

## Assignment 0

Q-1 Find out the z-transform and ROC of the following function

1)  $x(n) = \{1, 2, 1, 2, 3\}$

$$x(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

$$= 1 \cdot z^{-0} + 2z^{-1} + 1z^{-2} + 2z^{-3} + 3z^{-4}$$

$$= 1 + 2z^{-1} + z^{-2} + 2z^{-3} + 3z^{-4}$$

$$\text{ROC} = z - \{0, \infty\}$$

stability: system is stable because ROC includes unit circle

causal: system is causal because it is defined for  $n \geq 0$

2)  $x[n] = (n+1)^2 u(n)$

$$\begin{aligned} \rightarrow x(z) &= \sum_{n=-\infty}^{\infty} x(n)z^{-n} \\ &= \sum_{n=0}^{\infty} (n+1)^2 z^{-n} \end{aligned}$$

$$\text{ROC: } |z| > 1$$

stability: system is not stable because ROC does not include the unit circle.

consistency: The system is causal because it is defined only for  $n \geq 0$ .

$$3) x(n) = \log n$$

$$x(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} \log n z^{-n}$$

- The function  $\log(n)$  is not typically defined for  $n=0$  and is not causal function. It is also not well-defined for  $n < 1$ . Therefore z-transform isn't usually unstable for this function.

ROC: nA because standard z-transform is not exist.

stability: nA

causality: The system is not causal since  $\log(n)$  is not defined for  $n < 1$ .

$$4) x(n) = 2^n u(n) - 0.5^n u(-n-1)$$

$$\rightarrow x(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} 2^n u(n) z^{-n} - 0.5^n u(-n-1) z^{-n}$$

$$= \sum_{n=0}^{\infty} 2^n z^{-n} - \sum_{n=-\infty}^{\infty} 0.5^n z^{-n}$$

$$= \frac{1-2z^{-1}}{1-2z^{-1}} - \frac{1-0.5z}{1-0.5z}$$

$$= \frac{1}{z-2} - \frac{1}{1-0.5z}$$

ROC: ij for  $2^n u(n) \rightarrow |z| > 2$

ij for  $0.5^n u(n) \rightarrow |z| < 0.5$

The final ROC will be the intersection of both sequence which is empty so the z-transform does not exist in this combined form.

Stability: NA because no overlapping ROC.

Causality: The first part is causal but the second part is anti-causal.

$$g) x(n) = u(-n-6)$$

$$x(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$\sum_{n=-\infty}^{\infty} u(-n-6) z^{-n}$$

$$= \sum_{n=-\infty}^{-6} z^{-n} = z^6 \sum_{n=\infty}^{-6} (z^n - 1) = \frac{z^6}{1-z}$$

ROC: The ROC is  $|z| < 1$  where series is converges.

Stability: NA because ROC does not include the Unit circle.

Causality: The system is non causal because  $x[n]$  is defined  $n \leq 6$ .

6)  $x(n) = \delta(n+5)$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n} = \sum_{n=-\infty}^{\infty} \delta(n+5) z^{-n} = \sum_{n=-\infty}^{\infty} \frac{z^{-n}}{z^{-5}} = z^5$$

ROC:  $|z| > 1$

stability: The system is stable as the ROC includes the unit circle.

causality: The system is non-causal because the impulse is at  $n = -5$

Q-2 Find out the inverse z-transform of the following  $H(z)$  if the system is causal.

1)  $H(z) = 1 + z + z^{-1}$

$$= 1 + z + z^{-1} = (1 + z + z^{-1}) + (z + z^{-1})A + (z^{-1} + z^{-2})A^2$$

$z$  is for  $\delta[n]$

$z^{-1}$  is for  $\delta[n-1]$

$z^{-2}$  is for  $\delta[n-2]$

Here the  $\delta[n+1]$  is defined for  $n < 0$ , so the system is non causal.

2)  $H(z) = \log(1 + z^{-1})$

Power series of  $\log(1 + z^{-1})$  is

$$\sum_{k=1}^{\infty} (-1)^{k+1} \frac{z^{-k}}{k} = \frac{z^{-1}}{1-z}$$

so, the Inverse Z transform is

$$h[n] = \begin{cases} -1/k & (-1)^{k+1} \\ 0 & \text{otherwise} \end{cases}, \text{ for } n = k \quad (k \geq 1) \quad (s)$$

$$3) H(z) = \frac{z^2 + 2z - 3}{z^3 + z^2 - 4z + 3}$$

$$\text{- Here, } z^3 + z^2 - 4z + 3 = (z-1)^2(z+3) \quad (s)$$

$$\text{now, } H(z) = \frac{A}{z-1} + \frac{B}{(z-1)^2} + \frac{C}{z+3}$$

now multiply by denominator

$$A(z-1)(z+3) + B(z+3) + C(z-1)^2 \\ = z^2 + 2z - 3$$

$$\therefore A(z^2 + 2z - 3) + B(z+3) + C(z^2 - 2z + 1) = z^2 + 2z - 3$$

$$\therefore Az^2 + 2Az - 3A + Bz + 3B + Cz^2 - 2Cz + C = z^2 + 2z - 3$$

$$\therefore z^2(A+C) + z(2A+B-2C) + C(C+3B) = z^2 + 2z - 3$$

$$\text{so, } A+C=1 \quad (s) \quad (1+A)z^2 + (2A+B-2C)z + C(C+3B) = z^2 + 2z - 3$$

$$2A+B-2C=2$$

$$-3A+3B+C=-3$$

$$\text{so, } A=1, B=2, C=-1$$

$$\text{thus, } H(z) = \frac{1}{z-1} + \frac{2}{(z-1)^2} + \frac{1}{z+3} \quad (s)$$

so, the inverse Z transform are

$$h[n] = u[n] + 2(n+1)u[n] - (-3)^n u[n]$$

$$y) H(z) = \frac{(z^2+1)(2z-3)}{z^2-4z+3}$$

$$\text{Here, } (z^2+1)(2z-3) = 2z^3 - 3z^2 + 2z - 3$$

$$\text{so, } H(z) = \frac{2z^3 - 3z^2 + 2z - 3}{z^2 - 4z + 3}$$

$$\text{and, } z^2 - 4z + 3 = (z-1)(z-3)$$

$$\text{so, } H(z) = \frac{2z^3 - 3z^2 + 2z - 3}{(z-1)(z-3)}$$

$$\text{so, } H(z) = \frac{2z^3 - 3z^2 + 2z - 3}{(z-1)(z-3)}$$

$$\text{now, } H(z) = \frac{A}{z-1} + \frac{B}{z-3}$$

$$\text{so, } 2z^3 - 3z^2 + 2z - 3 = A(z-3) + B(z-1)$$

$$\text{now } z=3 \Rightarrow 8(3-1) = 2(2z) - 3(9) + 2(3) \Rightarrow 2z = 30 \Rightarrow z = 15$$

$$2B = 54 - 27 + 6 - 3$$

$$2B = 27 + 3 \Rightarrow B = 15$$

$$B = 15$$

$$\text{now } z=1 \Rightarrow 2(1) - 3(1) + 2(1) - 3 = -2A$$

$$(1-1)2-3+2-3 = -2A \Rightarrow A = 1$$

$$-2 = -2A$$

$$A = 1$$

$$\text{so, } H(z) = \frac{1}{z-1} + \frac{15}{z-3}$$

Now, Inverse Z-transform  $y[n] = H(z)$

$$H(z) = 4[z] + 15 \cdot 3^n [z]$$

$$= (1 + 15 \cdot 3^n) [z]$$

$$\text{Q} H(z) = \frac{s - 5}{(z-1)(z-3)^2}$$

$$\text{Here } H(z) = \frac{A}{z-1} + \frac{B}{z-3} + \frac{C}{(z-3)^2}$$

$$s = A(z-3)^2 + B(z-1)(z-3) + C(z-1)$$

$$s = A[2z^2 - 3z + 9] + B(z^2 - 4z + 3) + C(z-1)$$

$$s = Az^2 - 3Az + 9A + Bz^2 - 4Bz + 3B + Cz - C$$

$$s = z^2 [A+B] + z^1 [C-4B-3A] + z^0 [9A+3B-C]$$

$$\text{so, } A = \frac{s}{4}, C = \frac{s}{2}$$

$$\text{so, } H(z) = \frac{s/4}{z-1} + \frac{B}{z-3} + \frac{s/2}{(z-3)^2}$$

So, Inverse Transform  $y[n] = \frac{s}{4} [n] + B \cdot 3^n [n] + \frac{s}{2} n \cdot 3^n [n]$

Q-3) LCCD equation of LTI system is as following:

$$y(n) = y(n-1) + 5x(n) + 2x(n-1)$$

Find the impulse response frequency response magnitude response and phase response of system. If input is  $x[n] = 2^n u[n]$ . What will be the output.

- for impulse response,  $x(n) = \delta(n)$

$$\text{so, } h(n) = h(n-1) + 5\delta(n) + 2\delta(n-1) = (n+1)$$

$$\text{now } n=0 \geq h(0) = h(-1) + 5\delta(0) + 2\delta(-1)$$

assuming system is causal

$$(x(n)x(n))_d = (n)_d$$

$$h(0) = 0 + 5 + 2(0) = 5$$

$$\begin{aligned} \text{and } n=1 &> h(1) = h(0) + 5\delta(1) + 2\delta(0) \\ &= 5 + 5(0) + 2(1) \\ &= 7 \end{aligned}$$

$$\begin{aligned} \text{for } n=2 &> h(2) = h(2) + 5\delta(2) + 2\delta(1) \\ &= 7 + 2(0) + 5(0) \end{aligned}$$

$$\text{so, } h(n) = \begin{cases} 5, & n=0 \\ 7, & n>0 \end{cases}$$

- now, frequency response  $H(e^{j\omega}) = (s)H$

$$H(e^{j\omega}) = 5 + 7 \sum_{n=1}^{\infty} e^{-j\omega n}$$

$$\frac{8 - 5e^{-j\omega} - 5e^{-j\omega}}{8 + 5e^{-j\omega} - 5e^{-j\omega}}$$

$$H(e^{j\omega}) = 5 + \frac{1}{(s)e^{j\omega} - 1 - (s)e^{j\omega} + (s)e^{j\omega}}$$

$(s)e^{j\omega}$  - magnitude response  $sH = (s)e^{j\omega} + (s)e^{-j\omega}$

$(s)e^{j\omega}$  -

$$|H(e^{j\omega})| = \left| 5 + \frac{7}{e^{j\omega} - 1} \right|$$

Phase Response

$$H(e^{j\omega}) = \text{arg} \left( 5 + \frac{(-1)z^2}{e^{j\omega}-1} + (1-z)d \right) = (\omega)d, \text{ for } d < 0$$

$$(1-2d)z^2 + (1-d)z + (1-d)d = (0.2)z^2 + 0 = 0 \text{ for } d < 0$$

Output for input  $x(n) = 2^n u(n)$

$$y(n) = \sum_{k=0}^n h(k) x(n-k)$$

$$z = (0.5)z + z + 0 = (0.5)d$$

$$y(n) = 5(2^n) + 7 \sum_{k=1}^n 2^{n-k} = (1)d < 1 \text{ for } d < 0$$

$$= 5(2^n) + 7(2^{n-1})$$

$$= 12 \cdot 2^n - 7 \cdot 2^2 + (9)d = (9)d < 0 \text{ for } d < 0$$

$$(0.5)z + (0.5)z + F =$$

Q-4) If transfer function of LTI system is  $z^2 + 2z - 3$   
 $z^3 + z^2 - 4z + 3$ . Find output relationship between input  
and output. also find output unit step response  
of following system.

$$H(z) = \frac{z^2 + 2z - 3}{z^3 + z^2 - 4z + 3}$$

$$\sum_{i=1}^m f_i z = (w_i) H$$

$$\frac{y(z)}{x(z)} = \frac{z^2 + 2z - 3}{z^3 + z^2 - 4z + 3}$$

$$f_i + z = (w_i) H$$

$$(z^3 + z^2 - 4z + 3)y(z) \cdot (z^2 + 2z - 3)x(z)$$

$$z^2 y(z) + z^2 y(z) - 4z y(z) + 3y(z) = 2^2 x(z) + 2z x(z)$$

$$-3x(z)$$

$$\left| \begin{array}{l} f_i + z \\ -w_i \end{array} \right| = 1 (w_i) H$$

using difference equation

$$y(n+3) + y(n+2) - 4y(n+1) + 3y(n) = x(n+2) + 2x(n+1)$$

$x(n)$

- now, unit step response

$$x(z) = \frac{1}{1-z^{-1}}$$

The output  $y(z)$  is given by

$$y(z) = H(z) \cdot x(z)$$

$$= \frac{z^2 + 2z - 3}{z^3 + z^2 - 4z + 3} \cdot \frac{1}{1-z^{-1}}$$

Q-5] Find out convolution of  $2^n y(n)$  and  $5^n u(-n-1)$   
using Z transforms

$$\text{Let } x_1[n] = 2^n y(n) \quad S + (S-5)A = E_S$$

$$x_2[n] = 5^n u(-n-1) \quad S + AS - SA = E_S$$

$$\text{so, } x_1(z) = \sum_{n=0}^{\infty} 2^n z^{-n}$$

$$= \sum_{n=0}^{\infty} \left(\frac{2}{z}\right)^n$$

$$0 = AS + S^2 - S - A = S + A$$

$$= \frac{1}{1-\frac{2}{z}} = \frac{z}{z-2}$$

$$\text{so, } x_2(z) = \sum_{n=-\infty}^{\infty} 5^n z^{-n}$$

$$[ (n=-\infty) \rightarrow z - (n) \rightarrow n ] = 1$$

putting  $m = -n \rightarrow x_2(z) = \sum_{m=1}^{\infty} s^{-m} z^m$  with Pariu

$$(15) x_2(z) = \frac{(z/5)^2}{1-z/5}$$

$$= \frac{z^2}{25(1-z/5)}$$

90 92+2 104, with

1. = (s)x

1-5-1

$$\text{now, } x(z) = x_1(z) \cdot x_2(z)$$

$$= \frac{z^2}{z-2} \cdot \frac{z^2}{25(z-5)} \quad (s)r \cdot (s)H = (s)P$$

$$= \frac{z^3}{25(z-5)(z-2)}$$

now, Inverse transform is

$$\frac{z^3}{25(z-5)(z-2)} = \frac{A}{z-5} + \frac{B}{z-2}$$

$$z^3 = A(z-2) + B(z-5)$$

$$z^3 = Az - 2A + Bz - 5B$$

$$z^3 = z(A+B) - 2A + 5B$$

$$A+B=0 \quad 85B+2A=0$$

$$\text{Solving } A=0, B=0$$

$$\text{So, } y(n) = x_1(n) * x_2(n)$$

$$= \frac{1}{25} [2^n y(n) - 5^n y(n)]$$

Q-6) Find DTFT of the following sequence.

a)  $x(n) = (n+1) a^n u(n)$

$$\begin{aligned} x(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} (n+1) a^n u(n) e^{-jn\omega} \\ &= \sum_{n=0}^{\infty} (n+1) a^n e^{-jn\omega} \\ &= \frac{1}{(1-a e^{-j\omega})^2} \end{aligned}$$

b)  $x(n) = \sin(\omega_0 n)$

$$\begin{aligned} x(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} \sin(\omega_0 n) / \pi n \cdot e^{-jn\omega} \\ &= \begin{cases} 1, & |\omega - \omega_0| \leq \omega_0 \\ 0, & \text{otherwise} \end{cases} \end{aligned}$$

c)  $x(n) = \cos(\omega_0 n + \beta)$

$$x(e^{j\omega}) = \frac{1}{2} [e^{j\beta} s(\omega - \omega_0) + e^{j\omega_0} \delta(\omega + \omega_0)]$$

d)  $x(n) = e^{j\omega_0 n} u[n]$

$$\begin{aligned} x(e^{-j\omega}) &= \sum_{n=0}^{\infty} e^{j\omega_0 n} e^{-jn\omega} \\ &= \frac{1}{1 - e^{j(\omega_0 - \omega)}}, |\omega_0 - \omega| < 1 \end{aligned}$$

$$g(x(n)) = 1 \quad \text{for } -m < n < m$$

$$x(e^{j\omega}) = \sum_{n=-m}^m e^{-jn\omega}$$

$$= \frac{\sin(\omega(m+1)/2)}{\sin(\omega/2)} e^{-j\omega m/2} \quad (= (\omega/m) r)$$

Q-7 Determine IDTFT of the following sequence.

$$x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \sin(\pi/4) e^{jn\omega} d\omega$$

$$H_{BS}(e^{j\omega}) = \begin{cases} 1, & \text{if } |\omega| < \omega_0 \\ 0, & \text{if } |\omega| > \omega_0 \end{cases}$$

-The IDTFT of the above sequences

$$x(n) = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} H_{BS}(e^{jk\omega}) e^{jn\omega k} \quad (\omega \geq 0)$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} e^{jn\omega k} \quad (0 < \omega < \omega_0) = (n\pi) r$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} \frac{e^{jn\omega k}}{1 - e^{-j\omega k}} \quad [e^{jn\omega k} = 1 - e^{-j\omega k}] = (n\pi) r$$

$$= \frac{\sin(\pi/4)}{\pi n}$$

Q-8 Find out linear and circular convolution of the following sequences.

$$y(n) = x(n) * h(n)$$

$$= \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

$$= \{0, 0, 1, 2, 3, 4, 5, 6, 5, 4, 3, 2, 1, 0, 0\}$$

circular convolution.

$$y(n) = \sum_{k=0}^{n-1} x[k] h[(n-k) \bmod n]$$

$$= \{5, 4, 4, 4, 4, 4, 5, 6\}$$

- Q-9) consider an LTI system with  $|H(e^{j\omega})|$  and  $\arg |H(e^{j\omega})|$  is shown in figure. Find impulse response of the system

$$H(e^{j\omega}) = e^{j\alpha(\omega)}$$

$$h(t) = \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{j\omega t} e^{j\alpha(\omega)} d\omega$$

$$= \frac{1}{2\pi} (e^{j5\pi t/16} - e^{-j5\pi t/16})$$

$$+ \frac{1}{2\pi j} (e^{j\pi t/2} - e^{-j\pi t/2})$$

$$= \frac{2}{\pi} \sin \frac{5\pi t}{16}$$

$$\text{for } x(t) = \cos(\pi t) u(t)$$

$$y(t) = x(t) * h(t)$$

$$= \frac{2}{\pi} \int_{-\infty}^{\infty} \cos(\pi t) \sin \frac{5\pi t/16}{\pi} y(\tau) d\tau$$

$$= \frac{2}{\pi} \int_0^{\infty} \cos(\pi t) \sin \frac{5\pi t/16}{\pi} d\tau$$

$$= \frac{1}{\pi} \left[ \frac{\sin(5\pi t/16 - \pi t)}{5\pi/16} - \frac{\sin(5\pi t/16 + \pi t)}{5\pi/16} \right]$$

for unit step input  $x(t) = u(t)$

$$\begin{aligned} y(t) &= \frac{2}{\pi} \int_0^t \sin \frac{5\pi t}{6} dt \\ &= \frac{2}{\pi} \left[ \frac{-6}{5\pi} \cos \frac{5\pi t}{6} \right]_0^t \\ &= -\frac{12}{5\pi} \cos \frac{5\pi t}{6} + \frac{12}{5\pi} \end{aligned}$$

Q-10) Prove the following properties of z-transform

### i) Differentiation of $x(z)$

If  $x(z) = \{x[n]\}$ , then the z-transform of  $n \cdot x[n]$  is given by  $(z - \frac{1}{n})x(z)$

$$Z\{n \cdot x[n]\} = -z \frac{d}{dz} x(z)$$

$$(z - \frac{1}{n})x(z) = \frac{d}{dz} x(z)$$

Proof :

$$Z\{n \cdot x[n]\} = \sum_{n=-\infty}^{\infty} n \cdot x[n] z^{-n}$$

$$\text{so, } \frac{d}{dz} Z\{n \cdot x[n]\} = \frac{d}{dz} \left( \sum_{n=-\infty}^{\infty} x[n] z^{-n} \right)$$

$$\frac{d}{dz} Z\{n \cdot x[n]\} = \sum_{n=-\infty}^{\infty} x[n] \frac{d}{dz} z^{-n}$$

$$\left( \frac{1}{z} - \frac{1}{z^2} \right) Z\{n \cdot x[n]\} = \sum_{n=-\infty}^{\infty} x[n] \cdot (-n) z^{-n-1}$$

now multiplying  $-z$  both sides

$$-z \frac{dx(z)}{dz} = \sum_{n=-\infty}^{\infty} n x[n] z^{-n}$$

hence,  $\{x[n]\} = -z \frac{dx(z)}{dz}$

$$(w_i) x = (w_i) * x$$

### Time shifting

if  $x(z) = Z\{x[n]\}$ , then the  $Z$ -transform of  $x[n-k]$  is given by

$$Z\{x[n-k]\} = z^{-k} x(z)$$

Proof:

Consider the  $Z$ -transform of  $x[n-k]$

$$Z\{x[n-k]\} = \sum_{n=-\infty}^{\infty} x[n-k] z^{-n}$$

$$(w_i) x[n-k] = (w_i) x$$

now putting  $m=n-k$ , hence  $n=m+k$

$$Z\{x[n-k]\} = \sum_{m=-\infty}^{\infty} x[m] z^{-(m+k)}$$

$$= (z)^{-k} \sum_{m=-\infty}^{\infty} x[m] z^{-m}$$

this simplifies to:

$$Z\{x[n-k]\} = z^{-k} x(z)$$

(a-11) State and prove symmetry property of DTFT

Statement: The DTFT of a real valued sequence is Hermitian symmetric, meaning its magnitude is even and its phase is odd. Mathematically,  $X(-\omega) = \{X(\omega)\}^*$

$$x^*(-e^{j\omega}) = x(e^{j\omega})$$

where :

$x(-e^{j\omega})$  is the DFT of the real valued sequence  $x(n)$

$x(e^{-j\omega})$  is the complex conjugate of the DTFT evaluated at  $e^{j\omega}$

PROOF :

Let  $x(n)$  be a real valued sequence then its DTFT is

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n}$$

$$x^*(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x^*(n)e^{j\omega n}$$

Since  $x(n)$  is real,  $x^*(n) = x(n)$  therefore

$$x^*(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x(n)e^{j\omega n} = X(e^{j\omega})$$

Now let's evaluate  $x^*(e^{j\omega})$  at  $\omega = -\omega$

$$x(n) = \sum_{n=-\infty}^{\infty} x(n) e^{j\omega n}$$

Q-12)  $x(n) = \{1, 1, 2, 2\}, h(n) = \{1, 2, 3, 4\}$

$$\text{Convolution} = (a)$$

g)

1 2 3 4

1 1 2 3 4

2 2 8 6 8

9 1 0 1 2 2 1 0 2

1 0 0 0 0 0 2 2 1 9 4 / 1 0 2 6 2 8 2 1 0 2 0 0 0 0 7 1 , 0 2 9 4 .

lineas = {1, 3, 7, 13, 14, 8}

circulars = {15, 17, 15, 13}

$$\text{Convolution} = (a)$$

Q-13) Find 8 point DFT of  $x(n) = \cos(\frac{\pi n}{4})$ . Then find amplitude and phase spectrum.

Here  $x[n] = \cos^{\frac{\pi n}{4}}$  $x[n] = \{1, \sqrt{2}, 0, -\sqrt{2}, -1, -\sqrt{2}, 0, \sqrt{2}\}$ 

$$\text{now } X(k) = \sum_{n=0}^{7} x(n) e^{-j \frac{\pi}{4} kn}$$

$$x(0) = \sum_{n=0}^{7} x(n) = 0$$

$$x(1) = \sum_{n=0}^{7} x(n) e^{-j \frac{\pi}{4} kn}$$

$$= 1 + \sqrt{2} e^{-j \frac{\pi}{4}} + 0 - \sqrt{2} e^{-j \frac{3\pi}{4}} - 1 e^{-j \frac{5\pi}{4}} + \sqrt{2} + \sqrt{2} e^{-j \frac{7\pi}{4}}$$

Similarly for  $x(2), x(3), \dots, x(7)$  the non-zero values will occur at  $x(2)$  and  $x(6)$

for each of the following system, determine whether the system is i) stable ii) causal iii) linear iv) time invariant memoryless

$$y[n] = x[n - n_0]$$

- Here output is shifted version of input.  
So it is stable

- Here, if  $n_0 > 0$  it is causal else non causal

- Here, the system is linear shifted so it is linear

$$y[n] = x[-n]$$

the system is time reversal so it is stable

- the system is defined for  $n < 0$ , so it is non causal.

- The system depends on a different time index so, it is not memoryless.

$$y[n] = \log x(n)$$

If  $x > 0$  then it is stable

- the system is defined for  $n > 0$  only so it is causal

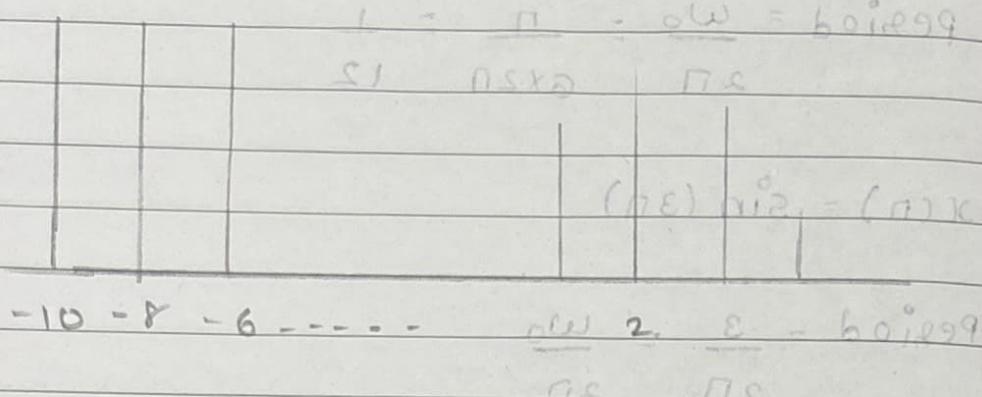
- the system is non-linear

(a)  $y[n] = (x[n])^2$  is non linear

Q-15) Draw following signal  $x(n) = 4(n) + 3\delta(n-1) + 2\delta(n-2) + 4\delta(n-3)$

$x(n_1)$

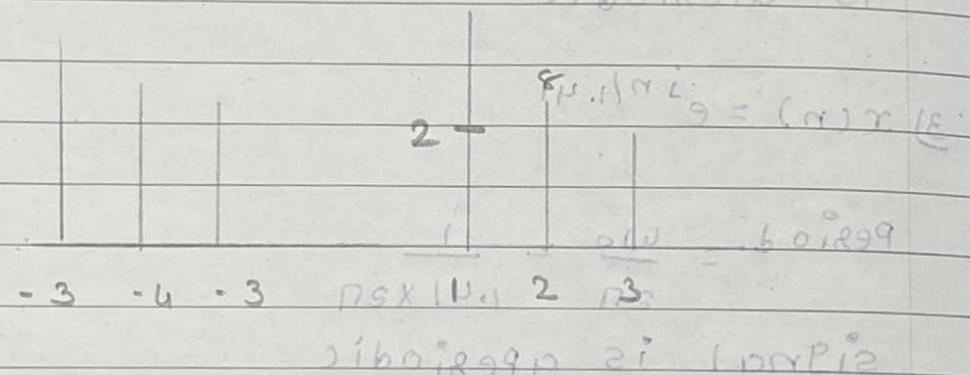
$$e^{j\omega n} = (\alpha)^n$$



$2x(n)$

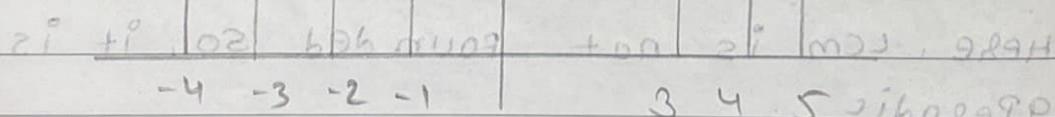
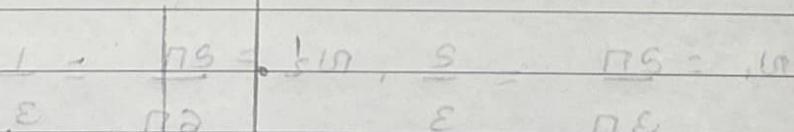
1.  $\alpha \neq 1$   $\Rightarrow$  Additive constant term  $\Rightarrow$   $x(n) = 4(n) + 3$

$$zibnibnibni$$



$x(-n)$

$$(x(n))_{n \geq 0} + (x(n))_{n < 0} = (n)x(n)$$



find out whether the signal is Periodical or aperiodic. Find the period of following signals.

1)  $x(n) = e^{j\pi n/16}$

$$\text{period} = \frac{\omega_0}{2\pi} = \frac{\pi}{6 \times 2\pi} = \frac{1}{12}$$

2)  $x(n) = \sin(3n)$

$$\text{period} = \frac{3}{2\pi} = \frac{\omega_0}{2\pi}$$

$\frac{3}{2\pi}$  is not rational number so the signal is aperiodic

3)  $x(n) = e^{jn/1.41}$

$$\text{period} = \frac{\omega_0}{2\pi} = \frac{1}{1.41 \times 2\pi}$$

signal is aperiodic

4)  $x(n) = \sin(3\pi n) + \cos(6\pi n)$

$$\omega_1 = \frac{2\pi}{3\pi} = \frac{2}{3}, \omega_2 = \frac{2\pi}{6\pi} = \frac{1}{3}$$

Here, LCM is not founded so, it is aperiodic

$$\text{Q} \quad x(n) = \frac{\sin n}{\pi n}$$

this is a sinc function which is aperiodic.