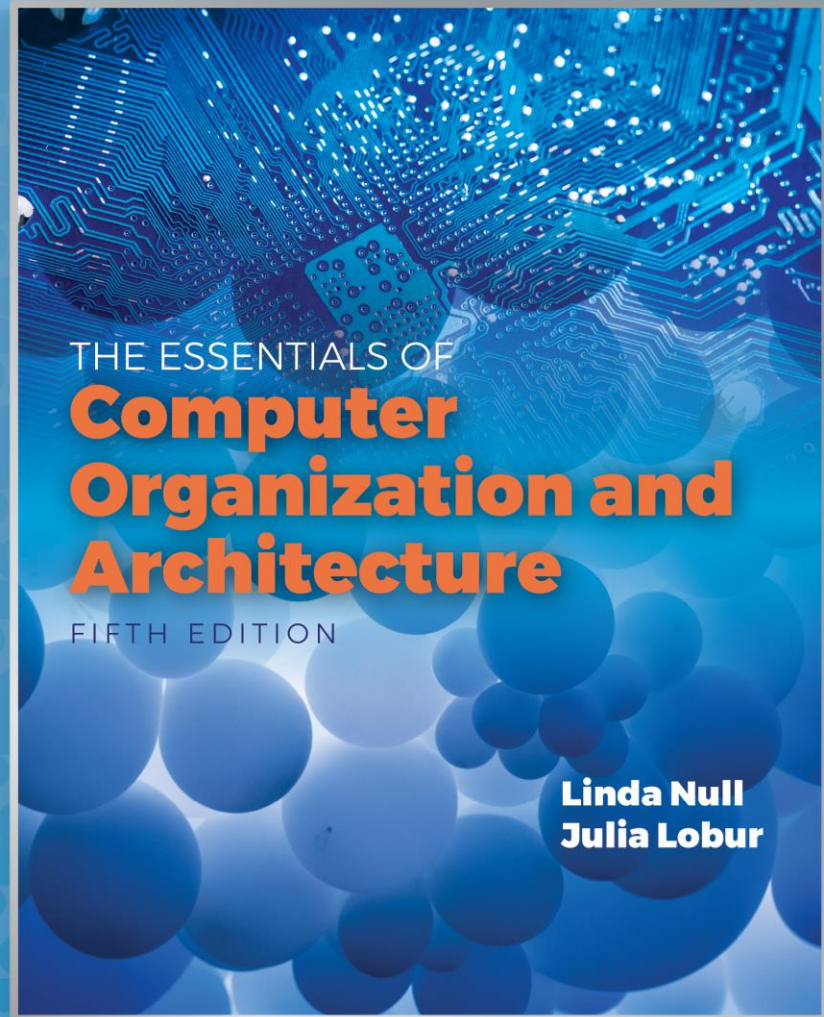


# Chapter 3

## Boolean Algebra and Digital Logic



# Objectives

- Understand the relationship between Boolean logic and digital computer circuits.
- Learn how to design simple logic circuits.
- Understand how digital circuits work together to form complex computer systems.

## 3.1 Introduction (1 of 2)

- In the latter part of the nineteenth century, George Boole incensed philosophers and mathematicians alike when he suggested that logical thought could be represented through mathematical equations.
  - How dare anyone suggest that human thought could be encapsulated and manipulated like an algebraic formula?
- Computers, as we know them today, are implementations of Boole's *Laws of Thought*.
  - John Atanasoff and Claude Shannon were among the first to see this connection.



## 3.1 Introduction (2 of 2)

- In the middle of the twentieth century, computers were commonly known as “thinking machines” and “electronic brains.”
  - Many people were fearful of them.
- Nowadays, we rarely ponder the relationship between electronic digital computers and human logic. Computers are accepted as part of our lives.
  - Many people, however, are still fearful of them.
- In this chapter, you will learn the simplicity that constitutes the essence of the machine.

## 3.2 Boolean Algebra (1 of 17)

- Boolean algebra is a mathematical system for the manipulation of variables that can have one of two values.
  - In formal logic, these values are “true” and “false.”
  - In digital systems, these values are “on” and “off,” 1 and 0, or “high” and “low.”
- Boolean expressions are created by performing operations on Boolean variables.
  - Common Boolean operators include AND, OR, and NOT.

## 3.2 Boolean Algebra (2 of 17)

- A Boolean operator can be completely described using a truth table.
- The truth table for the Boolean operators AND and OR are shown at the right.
- The AND operator is also known as a Boolean product. The OR operator is the Boolean sum.

X AND Y

| X | Y | XY |
|---|---|----|
| 0 | 0 | 0  |
| 0 | 1 | 0  |
| 1 | 0 | 0  |
| 1 | 1 | 1  |

X OR Y

| X | Y | X+Y |
|---|---|-----|
| 0 | 0 | 0   |
| 0 | 1 | 1   |
| 1 | 0 | 1   |
| 1 | 1 | 1   |

## 3.2 Boolean Algebra (3 of 17)

- The truth table for the Boolean NOT operator is shown at the right.
- The NOT operation is most often designated by a prime mark ( $\mathbf{x'}$ ). It is sometimes indicated by an overbar ( $\overline{\mathbf{x}}$ ) or an “elbow” ( $\neg\mathbf{x}$ ).

| NOT $\mathbf{x}$ |               |
|------------------|---------------|
| $\mathbf{x}$     | $\mathbf{x'}$ |
| 0                | 1             |
| 1                | 0             |

## 3.2 Boolean Algebra (4 of 17)

- A Boolean function has:
  - at least one Boolean variable,
  - at least one Boolean operator, and
  - at least one input from the set  $\{0,1\}$ .
- It produces an output that is also a member of the set  $\{0,1\}$ .

Now you know why the binary numbering system is so handy in digital systems.



## 3.2 Boolean Algebra (5 of 17)

- The truth table for the Boolean function:

$$F(x, y, z) = xz' + y$$

is shown at the right.

- To make evaluation of the Boolean function easier, the truth table contains extra (shaded) columns to hold evaluations of subparts of the function.

$$F(x, y, z) = xz' + y$$

| x | y | z | z' xz' |   | xz' + y |
|---|---|---|--------|---|---------|
| 0 | 0 | 0 | 1      | 0 | 0       |
| 0 | 0 | 1 | 0      | 0 | 0       |
| 0 | 1 | 0 | 1      | 0 | 1       |
| 0 | 1 | 1 | 0      | 0 | 1       |
| 1 | 0 | 0 | 1      | 1 | 1       |
| 1 | 0 | 1 | 0      | 0 | 0       |
| 1 | 1 | 0 | 1      | 1 | 1       |
| 1 | 1 | 1 | 0      | 0 | 1       |

## 3.2 Boolean Algebra (6 of 17)

- As with common arithmetic, Boolean operations have rules of precedence.
- The NOT operator has highest priority, followed by AND and then OR.
- This is how we chose the (shaded) function subparts in our table.

$$F(x, y, z) = xz' + y$$

| x | y | z | $z'$ | $xz'$ | $xz' + y$ |
|---|---|---|------|-------|-----------|
| 0 | 0 | 0 | 1    | 0     | 0         |
| 0 | 0 | 1 | 0    | 0     | 0         |
| 0 | 1 | 0 | 1    | 0     | 1         |
| 0 | 1 | 1 | 0    | 0     | 1         |
| 1 | 0 | 0 | 1    | 1     | 1         |
| 1 | 0 | 1 | 0    | 0     | 0         |
| 1 | 1 | 0 | 1    | 1     | 1         |
| 1 | 1 | 1 | 0    | 0     | 1         |

## 3.2 Boolean Algebra (7 of 17)

- Digital computers contain circuits that implement Boolean functions.
- The simpler that we can make a Boolean function, the smaller the circuit that will result.
  - Simpler circuits are cheaper to build, consume less power, and run faster than complex circuits.
- With this in mind, we always want to reduce our Boolean functions to their simplest form.
- There are a number of Boolean identities that help us to do this.

## 3.2 Boolean Algebra (8 of 17)

- Most Boolean identities have an AND (product) form as well as an OR (sum) form. We give our identities using both forms. Our first group is rather intuitive:

| Identity Name  | AND Form  | OR Form      |
|----------------|-----------|--------------|
| Identity Law   | $1x = x$  | $0 + x = x$  |
| Null Law       | $0x = 0$  | $1 + x = 1$  |
| Idempotent Law | $xx = x$  | $x + x = x$  |
| Inverse Law    | $xx' = 0$ | $x + x' = 1$ |



## 3.2 Boolean Algebra (9 of 17)

- Our second group of Boolean identities should be familiar to you from your study of algebra:

| Identity Name    | AND Form            | OR Form             |
|------------------|---------------------|---------------------|
| Commutative Law  | $xy = yx$           | $x+y = y+x$         |
| Associative Law  | $(xy)z = x(yz)$     | $(x+y)+z = x+(y+z)$ |
| Distributive Law | $x+yz = (x+y)(x+z)$ | $x(y+z) = xy+xz$    |

## 3.2 Boolean Algebra (10 of 17)

- Our last group of Boolean identities are perhaps the most useful.
- If you have studied set theory or formal logic, these laws are also familiar to you.

| Identity Name         | AND Form          | OR Form         |
|-----------------------|-------------------|-----------------|
| Absorption Law        | $x(x+y) = x$      | $x + xy = x$    |
| DeMorgan's Law        | $(xy)' = x' + y'$ | $(x+y)' = x'y'$ |
| Double Complement Law | $(x)'' = x$       |                 |

## 3.2 Boolean Algebra (11 of 17)

$$F(x, y, z) = xy + x'z + yz$$

- We can use Boolean identities to simplify:

$$\begin{aligned} F(x, y, z) &= xy + x'z + yz \\ &= xy + x'z + yz(1) && \text{(Identity)} \\ &= xy + x'z + yz(x + x') && \text{(Inverse)} \\ &= xy + x'z + (yz)x + (yz)x' && \text{(Distributive)} \\ &= xy + x'z + x(yz) + x'(zy) && \text{(Commutative)} \\ &= xy + x'z + (xy)z + (x'z)y && \text{(Associative twice)} \\ &= xy + (xy)z + x'z + (x'z)y && \text{(Commutative)} \\ &= xy(1 + z) + x'z(1 + y) && \text{(Distributive)} \\ &= xy(1) + x'z(1) && \text{(Null)} \\ &= xy + x'z && \text{(Identity)} \end{aligned}$$

## 3.2 Boolean Algebra (12 of 17)

- Sometimes it is more economical to build a circuit using the complement of a function (and complementing its result) than it is to implement the function directly.
- DeMorgan's law provides an easy way of finding the complement of a Boolean function.
- Recall DeMorgan's law states:  
$$(xy)' = x' + y' \quad \text{and} \quad (x + y)' = x' y'$$



## 3.2 Boolean Algebra (13 of 17)

- DeMorgan's law can be extended to any number of variables.
- Replace each variable by its complement and change all ANDs to ORs and all ORs to ANDs.
- Thus, we find the complement of:

$$F(x, y, z) = (xy) + (x'y) + (xz')$$

is:

$$\begin{aligned} F'(x, y, z) &= ((xy) + (x'y) + (xz'))' \\ &= (xy)' (x'y)' (xz')' \\ &= (x' + y') (x + y') (x' + z) \end{aligned}$$

## 3.2 Boolean Algebra (14 of 17)

- Through our exercises in simplifying Boolean expressions, we see that there are numerous ways of stating the same Boolean expression.
  - These “synonymous” forms are *logically equivalent*.
  - Logically equivalent expressions have identical truth tables.
- In order to eliminate as much confusion as possible, designers express Boolean functions in *standardized* or *canonical* form.

## 3.2 Boolean Algebra (15 of 17)

- There are two canonical forms for Boolean expressions: sum-of-products and product-of-sums.
  - Recall the Boolean product is the AND operation and the Boolean sum is the OR operation.
- In the sum-of-products form, ANDed variables are ORed together.

- For example:  $F(x, y, z) = xy + xz + yz$

- In the product-of-sums form, ORed variables are ANDed together.

- For example:  $F(x, y, z) = (x+y)(x+z)(y+z)$

## 3.2 Boolean Algebra (16 of 17)

- It is easy to convert a function to sum-of-products form using its truth table.
- We are interested in the values of the variables that make the function true (= 1).
- Using the truth table, we list the values of the variables that result in a true function value.
- Each group of variables is then ORed together.

$$F(x, y, z) = xz' + y$$

| x | y | z | $xz' + y$ |
|---|---|---|-----------|
| 0 | 0 | 0 | 0         |
| 0 | 0 | 1 | 0         |
| 0 | 1 | 0 | 1         |
| 0 | 1 | 1 | 1         |
| 1 | 0 | 0 | 1         |
| 1 | 0 | 1 | 0         |
| 1 | 1 | 0 | 1         |
| 1 | 1 | 1 | 1         |



## 3.2 Boolean Algebra (17 of 17)

- The sum-of-products form for our function is:

$$F(x, y, z) = (x'yz') + (x'yz) + (xy'z') + (xyz') + (xyz)$$

$$F(x, y, z) = xz' + y$$

| x | y | z | $xz' + y$ |
|---|---|---|-----------|
| 0 | 0 | 0 | 0         |
| 0 | 0 | 1 | 0         |
| 0 | 1 | 0 | 1         |
| 0 | 1 | 1 | 1         |
| 1 | 0 | 0 | 1         |
| 1 | 0 | 1 | 0         |
| 1 | 1 | 0 | 1         |
| 1 | 1 | 1 | 1         |

We note that this function is not in simplest terms. Our aim is only to rewrite our function in canonical sum-of-products form.

## 3.3 Logic Gates (1 of 6)

- We have looked at Boolean functions in abstract terms.
- In this section, we see that Boolean functions are implemented in digital computer circuits called gates.
- A gate is an electronic device that produces a result based on two or more input values.
  - In reality, gates consist of one to six transistors, but digital designers think of them as a single unit.
  - Integrated circuits contain collections of gates suited to a particular purpose.

## 3.3 Logic Gates (2 of 6)

- The three simplest gates are the AND, OR, and NOT gates.



X AND Y

| X | Y | XY |
|---|---|----|
| 0 | 0 | 0  |
| 0 | 1 | 0  |
| 1 | 0 | 0  |
| 1 | 1 | 1  |



X OR Y

| X | Y | X+Y |
|---|---|-----|
| 0 | 0 | 0   |
| 0 | 1 | 1   |
| 1 | 0 | 1   |
| 1 | 1 | 1   |



NOT X

| X | X' |
|---|----|
| 0 | 1  |
| 1 | 0  |

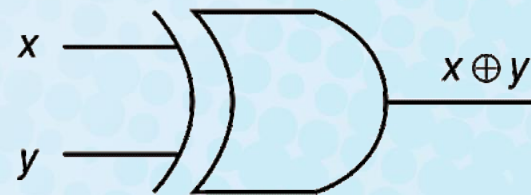
- They correspond directly to their respective Boolean operations, as you can see by their truth tables.

## 3.3 Logic Gates (3 of 6)

- Another very useful gate is the exclusive OR (XOR) gate.
- The output of the XOR operation is true only when the values of the inputs differ.

$x \text{ XOR } y$

| $x$ | $y$ | $x \oplus y$ |
|-----|-----|--------------|
| 0   | 0   | 0            |
| 0   | 1   | 1            |
| 1   | 0   | 1            |
| 1   | 1   | 0            |



**Note the special symbol  $\oplus$  for the XOR operation.**

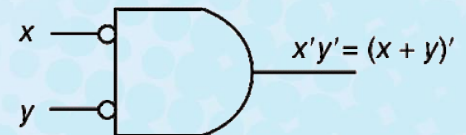
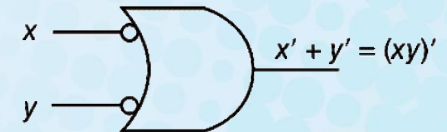
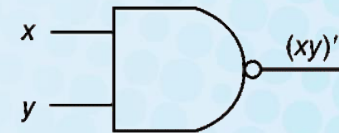


## 3.3 Logic Gates (4 of 6)

- NAND and NOR are two very important gates. Their symbols and truth tables are shown at the right.

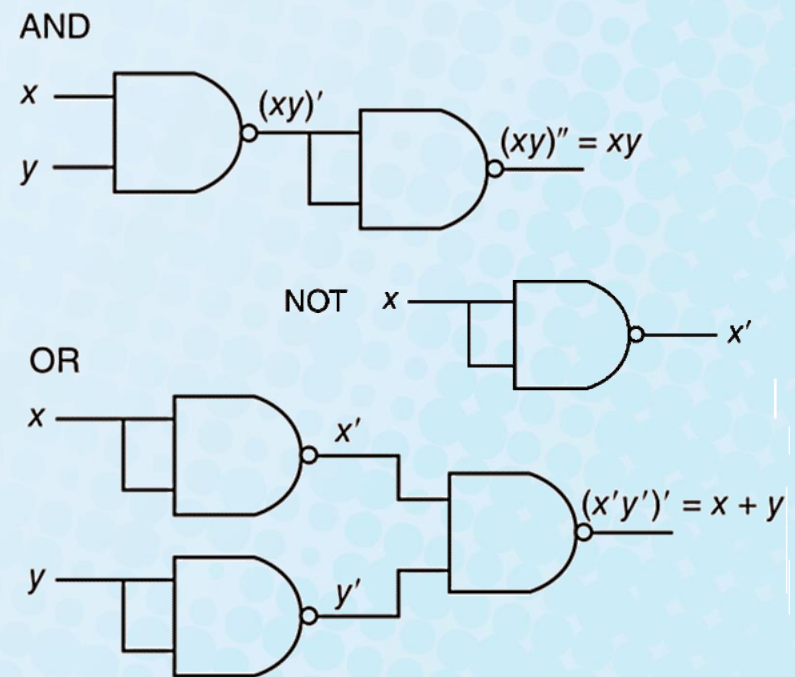
| X NAND Y |   |          |
|----------|---|----------|
| X        | Y | X NAND Y |
| 0        | 0 | 1        |
| 0        | 1 | 1        |
| 1        | 0 | 1        |
| 1        | 1 | 0        |

| X NOR Y |   |         |
|---------|---|---------|
| X       | Y | X NOR Y |
| 0       | 0 | 1       |
| 0       | 1 | 0       |
| 1       | 0 | 0       |
| 1       | 1 | 0       |



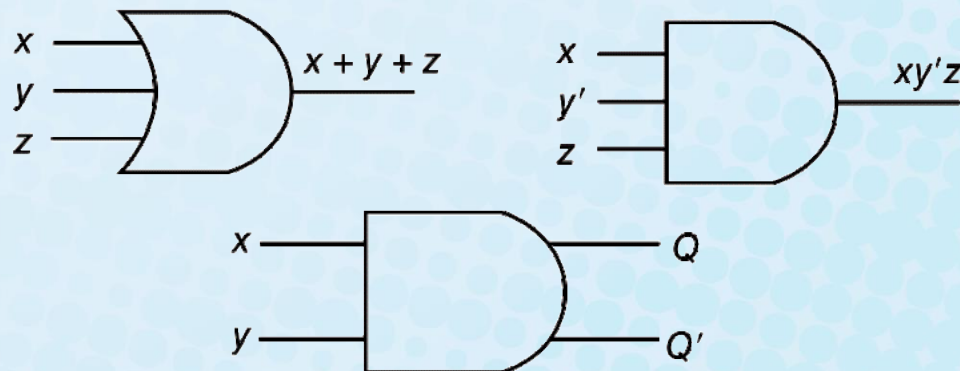
## 3.3 Logic Gates (5 of 6)

- NAND and NOR are known as *universal gates* because they are inexpensive to manufacture and any Boolean function can be constructed using only NAND or only NOR gates.



## 3.3 Logic Gates (6 of 6)

- Gates can have multiple inputs and more than one output.
  - A second output can be provided for the complement of the operation.
  - We'll see more of this later.



## 3.4 Karnaugh Maps

- Simplification of Boolean functions leads to simpler (and usually faster) digital circuits.
- Simplifying Boolean functions using identities is time-consuming and error-prone.
- This special section presents an easy, systematic method for reducing Boolean expressions.



## 3.4.1 Introduction

- In 1953, Maurice Karnaugh was a telecommunications engineer at Bell Labs.
- While exploring the new field of digital logic and its application to the design of telephone circuits, he invented a graphical way of visualizing and then simplifying Boolean expressions.
- This graphical representation, now known as a Karnaugh map, or Kmap, is named in his honor.

## 3.4.2 Description of Kmaps and Terminology (1 of 5)

- A Kmap is a matrix consisting of rows and columns that represent the output values of a Boolean function.
- The output values placed in each cell are derived from the minterms of a Boolean function.
- A *minterm* is a product term that contains all of the function's variables exactly once, either complemented or not complemented.

## 3.4.2 Description of Kmaps and Terminology (2 of 5)

- For example, the minterms for a function having the inputs  $x$  and  $y$  are  $\mathbf{x' y}$ ,  $\mathbf{x' y}$ ,  $\mathbf{xy'}$ , and  $\mathbf{xy}$ .
- Consider the Boolean function,  $\mathbf{F(x, y) = xy + xy'}$
- Its minterms are:

| Minterm | X | Y |
|---------|---|---|
| $X' Y'$ | 0 | 0 |
| $X' Y$  | 0 | 1 |
| $X Y'$  | 1 | 0 |
| $X Y$   | 1 | 1 |

## 3.4.2 Description of Kmaps and Terminology (3 of 5)

- Similarly, a function having three inputs, has the minterms that are shown in this diagram.

| Minterm  | X | Y | Z |
|----------|---|---|---|
| $X'Y'Z'$ | 0 | 0 | 0 |
| $X'Y'Z$  | 0 | 0 | 1 |
| $X'YZ'$  | 0 | 1 | 0 |
| $X'YZ$   | 0 | 1 | 1 |
| $XY'Z'$  | 1 | 0 | 0 |
| $XY'Z$   | 1 | 0 | 1 |
| $XYZ'$   | 1 | 1 | 0 |
| $XYZ$    | 1 | 1 | 1 |



## 3.4.2 Description of Kmaps and Terminology (4 of 5)

- A Kmap has a cell for each minterm.
- This means that it has a cell for each line for the truth table of a function.
- The truth table for the function  $F(x,y) = xy$  is shown at the right along with its corresponding Kmap.

$$F(X, Y) = XY$$

| X | Y | XY |
|---|---|----|
| 0 | 0 | 0  |
| 0 | 1 | 0  |
| 1 | 0 | 0  |
| 1 | 1 | 1  |

| X \ Y | 0 | 1 |
|-------|---|---|
|       | 0 | 1 |
| 0     | 0 | 0 |
| 1     | 0 | 1 |

## 3.4.2 Description of Kmaps and Terminology (5 of 5)

- As another example, we give the truth table and KMap for the function,  $F(x,y) = x + y$  at the right.
- This function is equivalent to the OR of all of the minterms that have a value of 1. Thus:

$$F(x, y) = x + y = x' y + x y' + x y$$

$$F(X, Y) = X + Y$$

| X | Y | X+Y |
|---|---|-----|
| 0 | 0 | 0   |
| 0 | 1 | 1   |
| 1 | 0 | 1   |
| 1 | 1 | 1   |

|   | Y | 0 | 1 |
|---|---|---|---|
| X | 0 | 0 | 1 |
|   | 1 | 1 | 1 |

## 3.4.3 Kmap Simplification for Two Variables (1 of 3)

- Of course, the minterm function that we derived from our Kmap was not in simplest terms.
  - That's what we started with in this example.
- We can, however, reduce our complicated expression to its simplest terms by finding adjacent 1s in the Kmap that can be collected into groups that are powers of two.
- In our example, we have two such groups.
  - Can you find them?

| x \ y | 0 | 1 |
|-------|---|---|
| 0     | 0 | 1 |
| 1     | 1 | 1 |

### 3.4.3 Kmap Simplification for Two Variables (2 of 3)

- The best way of selecting two groups of 1s form our simple Kmap is shown below.
- We see that both groups are powers of two and that the groups overlap.
- The next slide gives guidance for selecting Kmap groups.

| X \ Y | 0 | 1 |
|-------|---|---|
| 0     | 0 | 1 |
| 1     | 1 | 1 |



## 3.4.3 Kmap Simplification for Two Variables (3 of 3)

- The rules of Kmap simplification are:
  - Groupings can contain only 1s; no 0s.
  - Groups can be formed only at right angles; diagonal groups are not allowed.
  - The number of 1s in a group must be a power of 2 – even if it contains a single 1.
  - The groups must be made as large as possible.
  - Groups can overlap and wrap around the sides of the Kmap.

## 3.4.4 Kmap Simplification for Three Variables (1 of 7)

- A Kmap for three variables is constructed as shown in the diagram below.
- We have placed each minterm in the cell that will hold its value.
  - Notice that the values for the  $yz$  combination at the top of the matrix form a pattern that is not a normal binary sequence.

|     |   | $yz$     |         |         |          |
|-----|---|----------|---------|---------|----------|
|     |   | 00       | 01      | 11      | 10       |
| $x$ | 0 | $x'y'z'$ | $x'y'z$ | $x y z$ | $x'y z'$ |
|     | 1 | $x y'z'$ | $x y'z$ | $x y z$ | $x y z'$ |

### 3.4.4 Kmap Simplification for Three Variables (2 of 7)

- Thus, the first row of the Kmap contains all minterms where  $x$  has a value of zero.
- The first column contains all minterms where  $y$  and  $z$  both have a value of zero.

| $x \backslash yz$ | $yz$     |         |        |          |
|-------------------|----------|---------|--------|----------|
|                   | 00       | 01      | 11     | 10       |
| 0                 | $x'y'z'$ | $x'y'z$ | $xy z$ | $x'y z'$ |
| 1                 | $xy'z'$  | $xy'z$  | $xyz$  | $xyz'$   |

## 3.4.4 Kmap Simplification for Three Variables (3 of 7)

- Consider the function:

$$F(X, Y, Z) = X'Y'Z + X'YZ + XY'Z + XYZ$$

- Its Kmap is given below.
  - What is the largest group of 1s that is a power of 2?

| x \ yz | yz |    |    |    |
|--------|----|----|----|----|
|        | 00 | 01 | 11 | 10 |
| 0      | 0  | 1  | 1  | 0  |
| 1      | 0  | 1  | 1  | 0  |



## 3.4.4 Kmap Simplification for Three Variables (4 of 7)

- This grouping tells us that changes in the variables  $x$  and  $y$  have no influence upon the value of the function: They are irrelevant.

| x \ yz | yz |    |    |    |
|--------|----|----|----|----|
|        | 00 | 01 | 11 | 10 |
| 0      | 0  | 1  | 1  | 0  |
| 1      | 0  | 1  | 1  | 0  |

- This means that the function,

$$F(X, Y, Z) = X'Y'Z + X'YZ + XY'Z + XYZ$$

reduces to  $F(x) = z$ .

You could verify this reduction with identities or a truth table.

## 3.4.4 Kmap Simplification for Three Variables (5 of 7)

- Now for a more complicated Kmap. Consider the function:

$$F(X, Y, Z) = X'Y'Z' + X'Y'Z + X'YZ + X'YZ' + XY'Z' + XYZ'$$

- Its Kmap is shown below. There are (only) two groupings of 1s.
  - Can you find them?

| x \ yz | yz |    |    |    |
|--------|----|----|----|----|
|        | 00 | 01 | 11 | 10 |
| 0      | 1  | 1  | 1  | 1  |
| 1      | 1  | 0  | 0  | 1  |

## 3.4.4 Kmap Simplification for Three Variables (6 of 7)

- In this Kmap, we see an example of a group that wraps around the sides of a Kmap.
- This group tells us that the values of  $x$  and  $y$  are not relevant to the term of the function that is encompassed by the group.
  - What does this tell us about this term of the function?

What about the green group in the top row?

| x \ yz | yz |    |    |    |
|--------|----|----|----|----|
|        | 00 | 01 | 11 | 10 |
| 0      | 1  | 1  | 1  | 1  |
| 1      | 1  | 0  | 0  | 1  |

## 3.4.4 Kmap Simplification for Three Variables (7 of 7)

- The green group in the top row tells us that only the value of  $x$  is significant in that group.
- We see that it is complemented in that row, so the other term of the reduced function is  $\mathbf{X'}$
- Our reduced function is  $\mathbf{F(X, Y, Z) = X' + Z'}$

Recall that we had six minterms in our original function!

| x \ yz | yz |    |    |    |
|--------|----|----|----|----|
|        | 00 | 01 | 11 | 10 |
| 0      | 1  | 1  | 1  | 1  |
| 1      | 1  | 0  | 0  | 1  |



## 3.4.5 Kmap Simplification for Four Variables (1 of 4)

- Our model can be extended to accommodate the 16 minterms that are produced by a four-input function.
- This is the format for a 16-minterm Kmap:

|    |    | yz         |           |           |            |
|----|----|------------|-----------|-----------|------------|
|    |    | 00         | 01        | 11        | 10         |
| wx | 00 | $w'x'y'z'$ | $w'x'y'z$ | $w'xy z$  | $w'x'y z'$ |
|    | 01 | $w'x y'z'$ | $w'x y'z$ | $w'x y z$ | $w'x y z'$ |
|    | 11 | $w x'y'z'$ | $w x'y'z$ | $w x y z$ | $w x'y z'$ |
|    | 10 | $w x y'z'$ | $w x y'z$ | $w x y z$ | $w x y z'$ |

## 3.4.5 Kmap Simplification for Four Variables (2 of 4)

- We have populated the Kmap shown below with the nonzero minterms from the function:

$$F(W,X,Y,Z) = W'X'Y'Z' + W'X'Y'Z + W'X'YZ' + W'XYZ' + WX'Y'Z' + WX'Y'Z + WX'YZ'$$

- Can you identify (only) three groups in this Kmap?

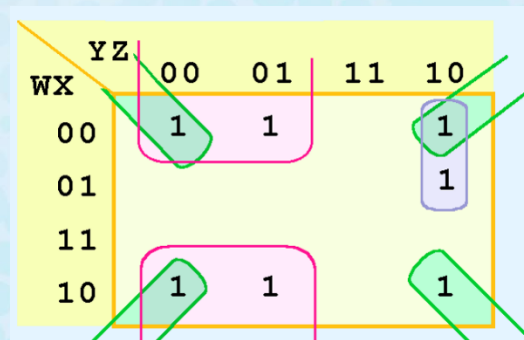
Recall that groups can overlap.

| WX \ YZ | YZ |    |    |    |
|---------|----|----|----|----|
|         | 00 | 01 | 11 | 10 |
| 00      | 1  | 1  |    | 1  |
| 01      |    |    |    | 1  |
| 11      |    |    |    |    |
| 10      | 1  | 1  |    | 1  |

## 3.4.5 Kmap Simplification for Four Variables (3 of 4)

- Our three groups consist of:
  - A purple group entirely within the Kmap at the right.
  - A pink group that wraps the top and bottom.
  - A green group that spans the corners.
- Thus we have three terms in our final function:

$$F(W, X, Y, Z) = X'Y' + X'Z' + W'YZ'$$



## 3.4.5 Kmap Simplification for Four Variables (4 of 4)

- It is possible to have a choice as to how to pick groups within a Kmap, while keeping the groups as large as possible.
- The (different) functions that result from the groupings below are logically equivalent.

|    |    | YZ |    |    |    |
|----|----|----|----|----|----|
|    |    | 00 | 01 | 11 | 10 |
| WX | 00 | 1  |    | 1  |    |
|    | 01 | 1  |    | 1  | 1  |
|    | 11 | 1  |    |    |    |
|    | 10 | 1  |    |    |    |
|    |    |    |    |    |    |

|    |    | YZ |    |    |    |
|----|----|----|----|----|----|
|    |    | 00 | 01 | 11 | 10 |
| WX | 00 | 1  |    | 1  |    |
|    | 01 | 1  |    | 1  | 1  |
|    | 11 | 1  |    |    |    |
|    | 10 | 1  |    |    |    |
|    |    |    |    |    |    |



## 3.4.6 Don't Care Conditions (1 of 5)

- Real circuits don't always need to have an output defined for every possible input.
  - For example, some calculator displays consist of 7-segment LEDs. These LEDs can display  $2^7 - 1$  patterns, but only ten of them are useful.
- If a circuit is designed so that a particular set of inputs can never happen, we call this set of inputs a don't care condition.
- They are very helpful to us in Kmap circuit simplification.

## 3.4.6 Don't Care Conditions (2 of 5)

- In a Kmap, a don't care condition is identified by an  $X$  in the cell of the minterm(s) for the don't care inputs, as shown here.
- In performing the simplification, we are free to include or ignore the  $X$ 's when creating our groups.

| WX \ YZ | YZ |    |    |    |
|---------|----|----|----|----|
|         | 00 | 01 | 11 | 10 |
| 00      | X  | 1  | 1  | X  |
| 01      |    | X  | 1  |    |
| 11      | X  |    | 1  |    |
| 10      |    |    | 1  |    |

## 3.4.6 Don't Care Conditions (3 of 5)

- In one grouping in the Kmap below, we have the function:

$$F(W, X, Y, Z) = W'X' + YZ$$

| WX \ YZ | YZ |    |    |    |
|---------|----|----|----|----|
|         | 00 | 01 | 11 | 10 |
| 00      | X  | 1  | 1  | X  |
| 01      |    | X  | 1  |    |
| 11      | X  |    | 1  |    |
| 10      |    |    | 1  |    |

## 3.4.6 Don't Care Conditions (4 of 5)

- A different grouping gives us the function:

$$F(W, X, Y, Z) = W'Z + YZ$$

| WX \ YZ | YZ |    |    |    |
|---------|----|----|----|----|
|         | 00 | 01 | 11 | 10 |
| 00      | X  | 1  | 1  | X  |
| 01      |    | X  | 1  |    |
| 11      | X  |    | 1  |    |
| 10      |    |    | 1  |    |



## 3.4.6 Don't Care Conditions (5 of 5)

- The truth table of:

$$F(W, X, Y, Z) = W' Y' + YZ$$

| WX \ YZ |    |    |    |    |
|---------|----|----|----|----|
|         | 00 | 01 | 11 | 10 |
| 00      | X  | 1  | 1  | X  |
| 01      |    | X  | 1  |    |
| 11      | X  |    | 1  |    |
| 10      |    |    | 1  |    |

- differs from the truth table of:

$$F(W, X, Y, Z) = W' Z + YZ$$

| WX \ YZ |    |    |    |    |
|---------|----|----|----|----|
|         | 00 | 01 | 11 | 10 |
| 00      | X  | 1  | 1  | X  |
| 01      |    | X  | 1  |    |
| 11      | X  |    | 1  |    |
| 10      |    |    | 1  |    |

- However, the values for which they differ, are the inputs for which we have don't care conditions.

## 3.4.7 Summary (1 of 2)

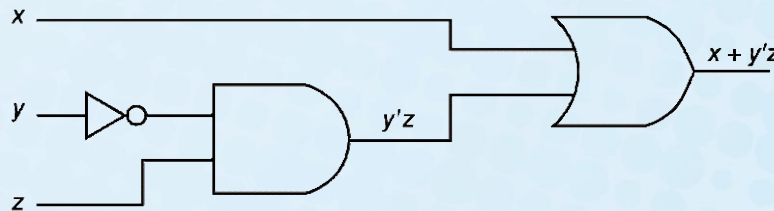
- Kmaps provide an easy graphical method of simplifying Boolean expressions.
- A Kmap is a matrix consisting of the outputs of the minterms of a Boolean function.
- In this section, we have discussed 2-, 3-, and 4-input Kmaps. This method can be extended to any number of inputs through the use of multiple tables.

## 3.4.7 Summary (2 of 2)

- Recapping the rules of Kmap simplification:
  - Groupings can contain only 1s; no 0s.
  - Groups can be formed only at right angles; diagonal groups are not allowed.
  - The number of 1s in a group must be a power of 2 – even if it contains a single 1.
  - The groups must be made as large as possible.
  - Groups can overlap and wrap around the sides of the Kmap.
  - Use don't care conditions when you can.

## 3.5 Digital Components (1 of 8)

- The main thing to remember is that combinations of gates implement Boolean functions.



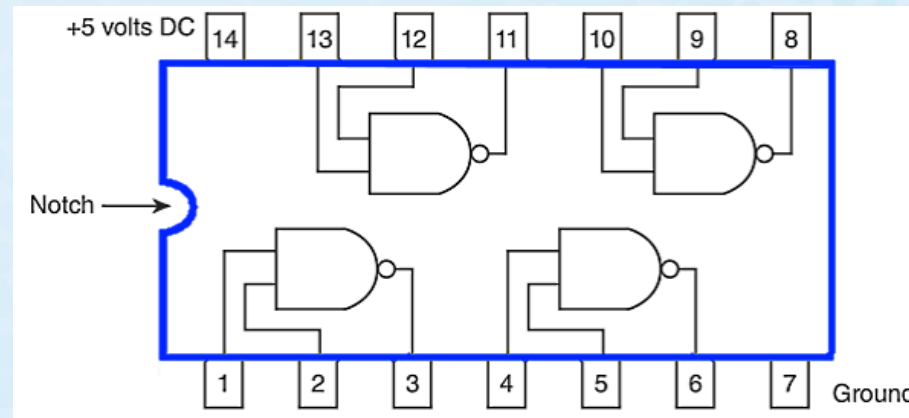
- The circuit below implements the Boolean function  $F(x, y, z) = x + y'z$ :

**We simplify our Boolean expressions so that we can create simpler circuits.**



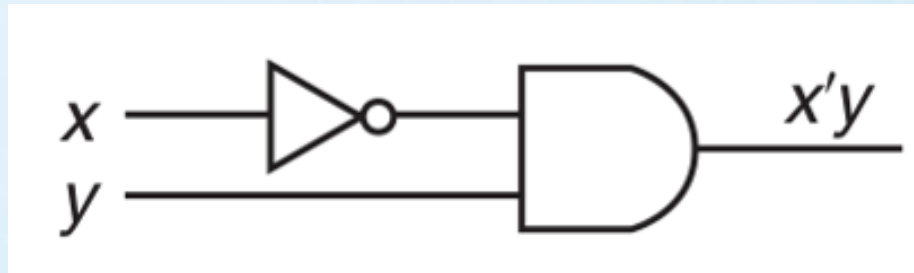
## 3.5 Digital Components (2 of 8)

- Standard digital components are combined into single integrated circuit packages.
- Boolean logic can be used to implement the desired functions.

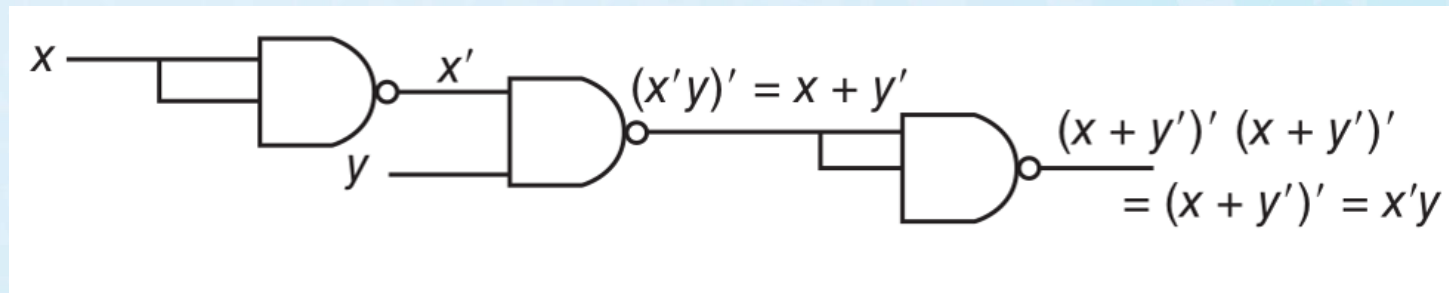


## 3.5 Digital Components (3 of 8)

- The Boolean circuit:

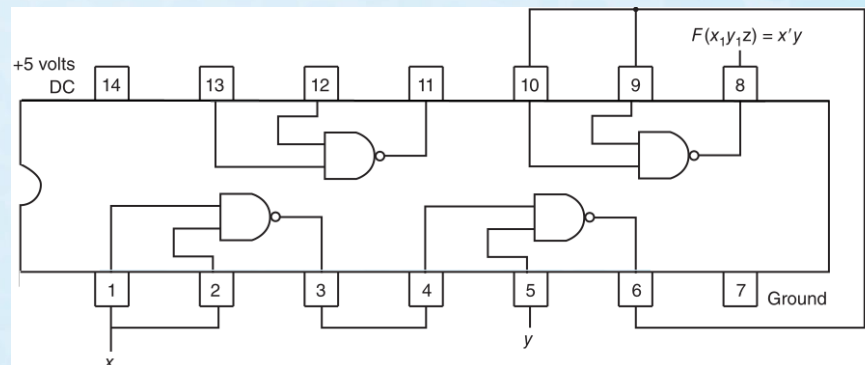
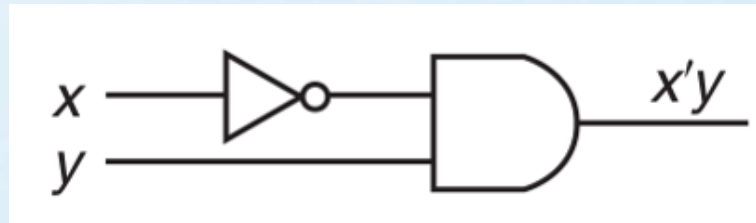


- Can be rendered using only NAND gates as:



## 3.5 Digital Components (4 of 8)

- So we can wire the pre-packaged circuit to implement our function:



## 3.5 Digital Components (5 of 8)

- Boolean logic is used to solve practical problems.
- Expressed in terms of Boolean logic practical problems can be expressed by truth tables.
- Truth tables can be readily rendered into Boolean logic circuits.



## 3.5 Digital Components (6 of 8)

- Suppose we are to design a logic circuit to determine the best time to plant a garden.
- We consider three factors (inputs):
  - (1) time, where 0 represents day and 1 represents evening;
  - (2) moon phase, where 0 represents not full and 1 represents full; and
  - (3) temperature, where 0 represents 45°F and below, and 1 represents over 45°F.
- We determine that the best time to plant a garden is during the evening with a full moon.

## 3.5 Digital Components (7 of 8)

- This results in the following truth table:

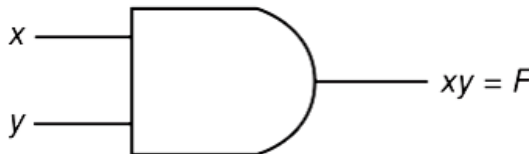
| Time (x) | Moon (y) | Temperature (z) | Plant? |
|----------|----------|-----------------|--------|
| 0        | 0        | 0               | 0      |
| 0        | 0        | 1               | 0      |
| 0        | 1        | 0               | 0      |
| 0        | 1        | 1               | 0      |
| 1        | 0        | 0               | 0      |
| 1        | 0        | 1               | 0      |
| 1        | 1        | 0               | 1      |
| 1        | 1        | 1               | 1      |

## 3.5 Digital Components (8 of 8)

- From the truth table, we derive the circuit:

| Time (x) | Moon (y) | Temperature (z) | Plant? |
|----------|----------|-----------------|--------|
| 0        | 0        | 0               | 0      |
| 0        | 0        | 1               | 0      |
| 0        | 1        | 0               | 0      |
| 0        | 1        | 1               | 0      |
| 1        | 0        | 0               | 0      |
| 1        | 0        | 1               | 0      |
| 1        | 1        | 0               | 1      |
| 1        | 1        | 1               | 1      |

$$F(x,y,z) = xyz' + xyz = xy$$



## 3.6 Combinational Circuits (1 of 12)

- We have designed a circuit that implements the Boolean function:

$$F(X, Y, Z) = X + Y'Z$$

- This circuit is an example of *a combinational logic circuit*.
- Combinational logic circuits produce a specified output (almost) at the instant when input values are applied.
  - In a later section, we will explore circuits where this is not the case.



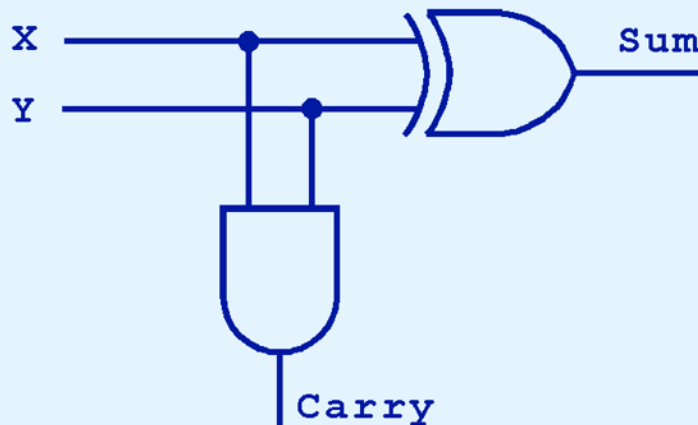
## 3.6 Combinational Circuits (2 of 12)

- Combinational logic circuits give us many useful devices.
- One of the simplest is the *half adder*, which finds the sum of two bits.
- We can gain some insight as to the construction of a half adder by looking at its truth table, shown at the right.

| Inputs |   | Outputs |       |
|--------|---|---------|-------|
| X      | Y | Sum     | Carry |
| 0      | 0 | 0       | 0     |
| 0      | 1 | 1       | 0     |
| 1      | 0 | 1       | 0     |
| 1      | 1 | 0       | 1     |

## 3.6 Combinational Circuits (3 of 12)

- As we see, the sum can be found using the XOR operation and the carry using the AND operation.



| Inputs |   | Outputs |       |
|--------|---|---------|-------|
| X      | Y | Sum     | Carry |
| 0      | 0 | 0       | 0     |
| 0      | 1 | 1       | 0     |
| 1      | 0 | 1       | 0     |
| 1      | 1 | 0       | 1     |

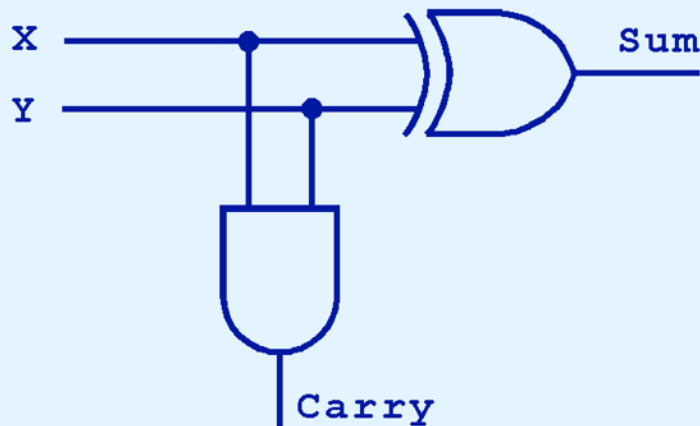
## 3.6 Combinational Circuits (4 of 12)

- We can change our half adder into to a full adder by including gates for processing the carry bit.
- The truth table for a full adder is shown at the right.

| Inputs |   |       | Outputs |     |
|--------|---|-------|---------|-----|
| X      | Y | Carry | Sum     | Out |
|        |   | In    |         |     |
| 0      | 0 | 0     | 0       | 0   |
| 0      | 0 | 1     | 1       | 0   |
| 0      | 1 | 0     | 1       | 0   |
| 0      | 1 | 1     | 0       | 1   |
| 1      | 0 | 0     | 1       | 0   |
| 1      | 0 | 1     | 0       | 1   |
| 1      | 1 | 0     | 0       | 1   |
| 1      | 1 | 1     | 1       | 1   |

## 3.6 Combinational Circuits (5 of 12)

- How can we change the half adder shown below to make it a full adder?

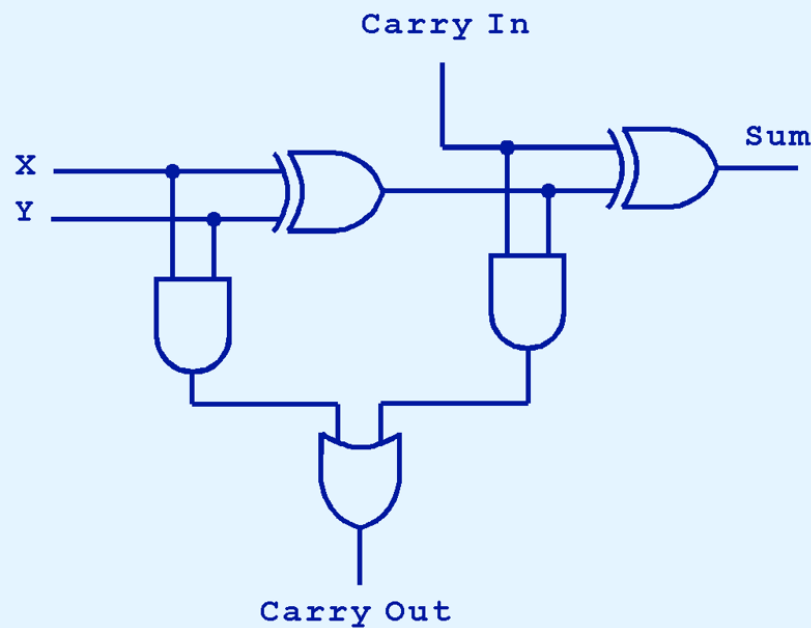


| Inputs |   |          | Outputs |           |
|--------|---|----------|---------|-----------|
| X      | Y | Carry In | Sum     | Carry Out |
| 0      | 0 | 0        | 0       | 0         |
| 0      | 0 | 1        | 1       | 0         |
| 0      | 1 | 0        | 1       | 0         |
| 0      | 1 | 1        | 0       | 1         |
| 1      | 0 | 0        | 1       | 0         |
| 1      | 0 | 1        | 0       | 1         |
| 1      | 1 | 0        | 0       | 1         |
| 1      | 1 | 1        | 1       | 1         |



## 3.6 Combinational Circuits (6 of 12)

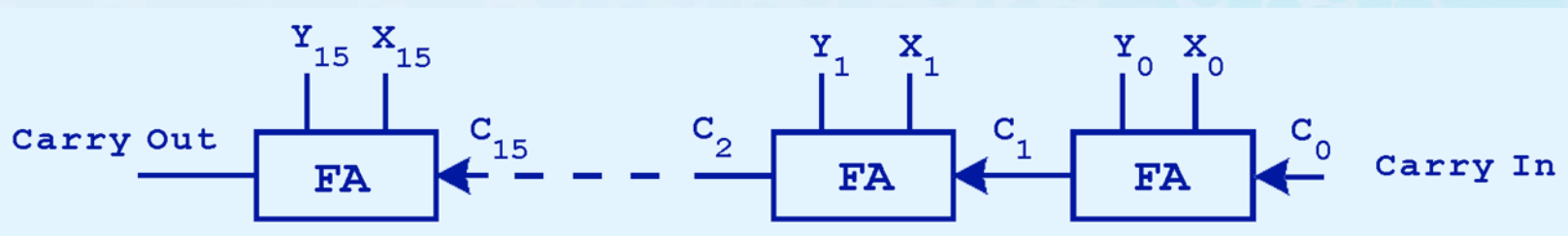
- Here's our completed full adder.



| Inputs |   |          | Outputs |           |
|--------|---|----------|---------|-----------|
| X      | Y | Carry In | Sum     | Carry Out |
| 0      | 0 | 0        | 0       | 0         |
| 0      | 0 | 1        | 1       | 0         |
| 0      | 1 | 0        | 1       | 0         |
| 0      | 1 | 1        | 0       | 1         |
| 1      | 0 | 0        | 1       | 0         |
| 1      | 0 | 1        | 0       | 1         |
| 1      | 1 | 0        | 0       | 1         |
| 1      | 1 | 1        | 1       | 1         |

## 3.6 Combinational Circuits (7 of 12)

- Just as we combined half adders to make a full adder, full adders can be connected in series.
- The carry bit “ripples” from one adder to the next; hence, this configuration is called a *ripple-carry adder*.



Today's systems employ more efficient adders.

## 3.6 Combinational Circuits (8 of 12)

- Decoders are another important type of combinational circuit.
- Among other things, they are useful in selecting a memory location according a binary value placed on the address lines of a memory bus.
- Address decoders with  $n$  inputs can select any of  $2^n$  locations.

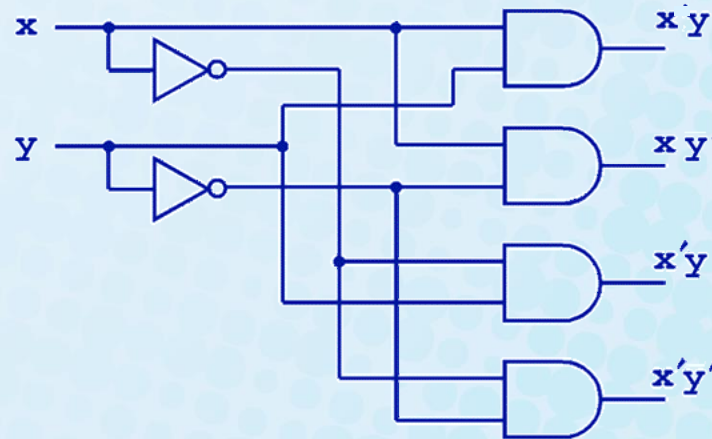
This is a block diagram for a decoder.



## 3.6 Combinational Circuits (9 of 12)

- This is what a 2-to-4 decoder looks like on the inside.

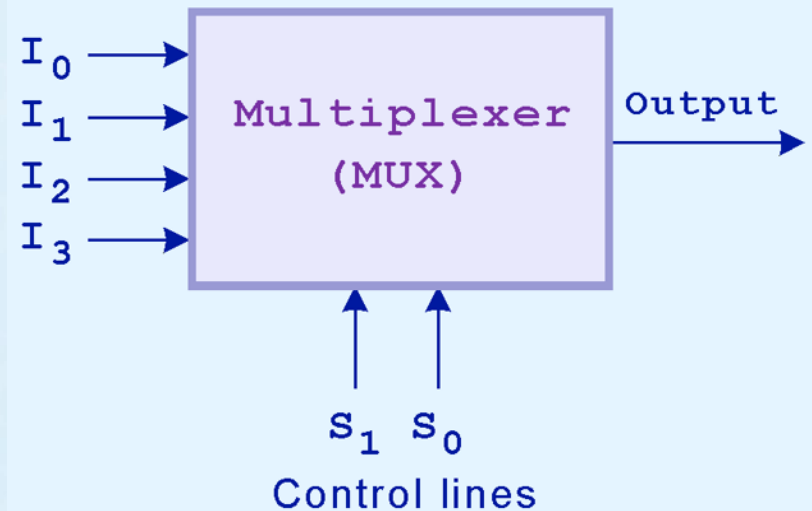
If  $x = 0$  and  $y = 1$ ,  
which output line  
is enabled?





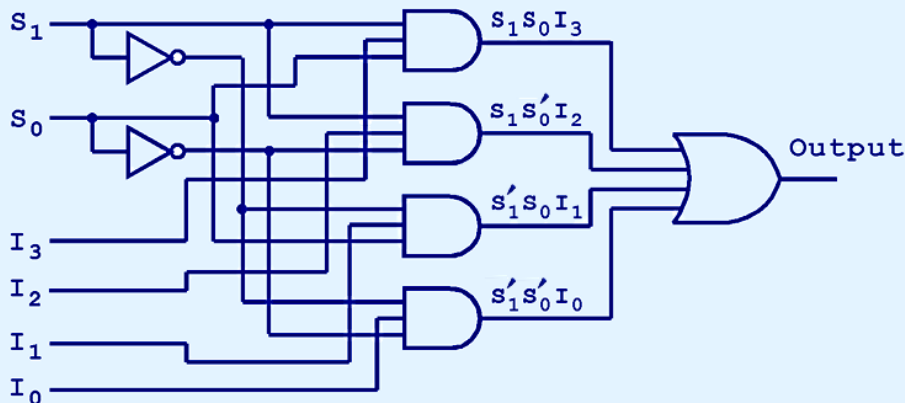
## 3.6 Combinational Circuits (10 of 12)

- A multiplexer does just the opposite of a decoder.
- It selects a single output from several inputs.
- The particular input chosen for output is determined by the value of the multiplexer's control lines.
- To be able to select among  $n$  inputs,  $\log_2 n$  control lines are needed.
- This is a block diagram for a multiplexer.



## 3.6 Combinational Circuits (11 of 12)

- This is what a 4-to-1 multiplexer looks like on the inside.

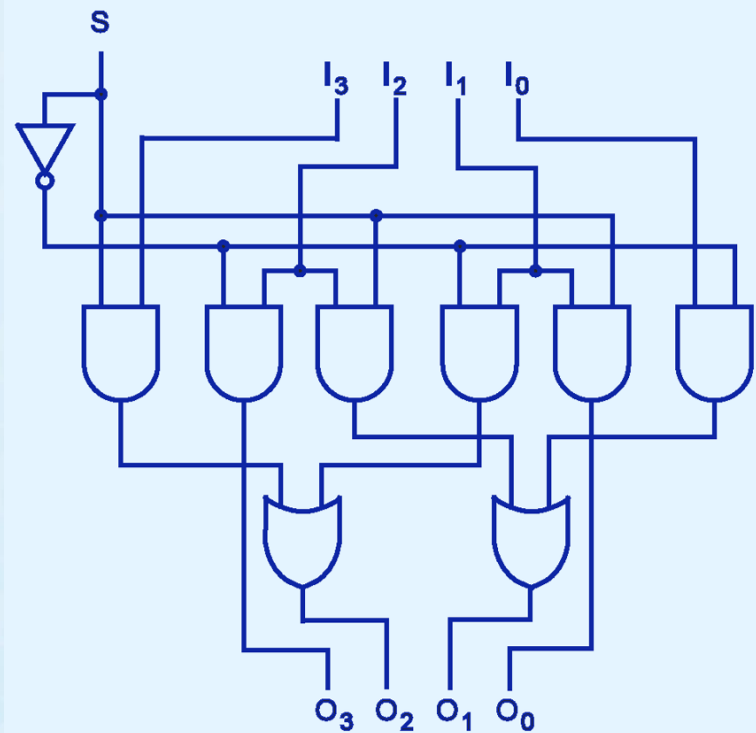


If  $S_0 = 1$  and  $S_1 = 0$ ,  
which input is  
transferred to the  
output?

## 3.6 Combinational Circuits (12 of 12)

- This shifter moves the bits of a nibble one position to the left or right.

If  $S = 0$ , in which direction do the input bits shift?



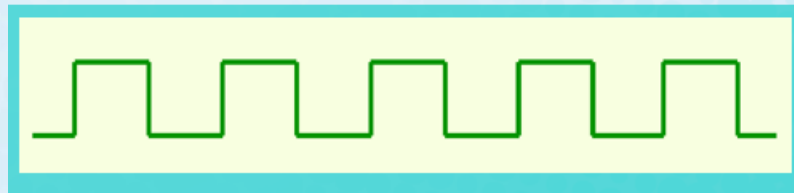
## 3.7 Sequential Circuits (1 of 12)

- Combinational logic circuits are perfect for situations when we require the immediate application of a Boolean function to a set of inputs.
- There are other times, however, when we need a circuit to change its value with consideration to its current state as well as its inputs.
  - These circuits have to “remember” their current state.
- *Sequential logic circuits* provide this functionality for us.



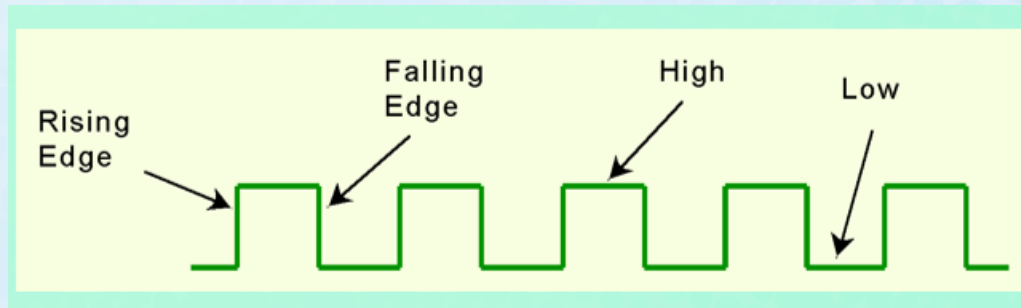
## 3.7 Sequential Circuits (2 of 12)

- As the name implies, sequential logic circuits require a means by which events can be sequenced.
- State changes are controlled by clocks.
  - A “clock” is a special circuit that sends electrical pulses through a circuit.
- Clocks produce electrical waveforms such as the one shown below.



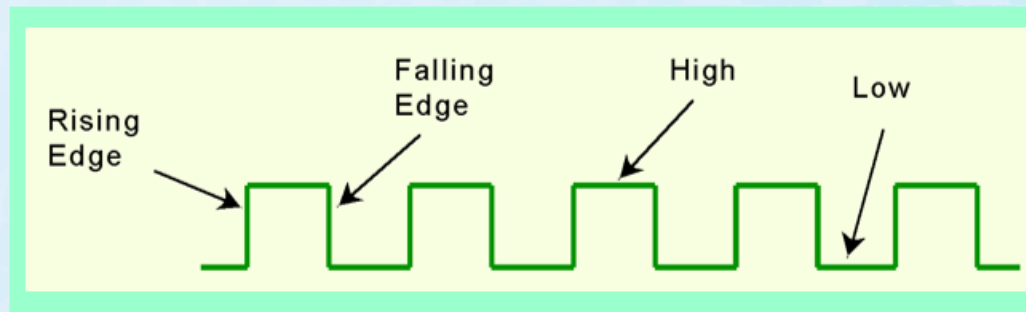
## 3.7 Sequential Circuits (3 of 12)

- State changes occur in sequential circuits only when the clock ticks.
- Circuits can change state on the rising edge, falling edge, or when the clock pulse reaches its highest voltage.



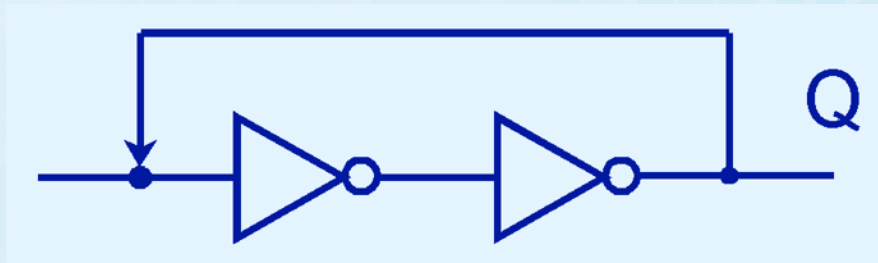
## 3.7 Sequential Circuits (4 of 12)

- Circuits that change state on the rising edge, or falling edge of the clock pulse are called *edge-triggered*.
- *Level-triggered circuits* change state when the clock voltage reaches its highest or lowest level.



## 3.7 Sequential Circuits (5 of 12)

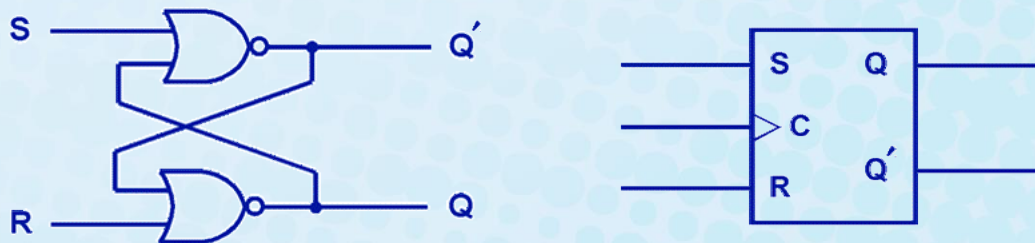
- To retain their state values, sequential circuits rely on *feedback*.
- Feedback in digital circuits occurs when an output is looped back to the input.
- A simple example of this concept is shown below.
  - If Q is 0 it will always be 0, if it is 1, it will always be 1. Why?





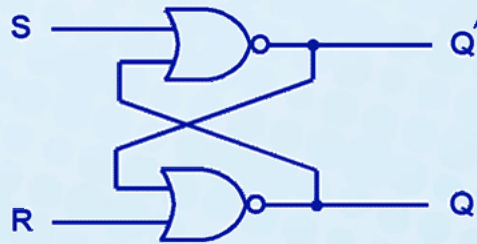
## 3.7 Sequential Circuits (6 of 12)

- You can see how feedback works by examining the most basic sequential logic components, the SR flip-flop.
  - The “SR” stands for set/reset.
- The internals of an SR flip-flop are shown below, along with its block diagram.



## 3.7 Sequential Circuits (7 of 12)

- The behavior of an SR flip-flop is described by a characteristic table.
- $Q(t)$  means the value of the output at time  $t$ .  $Q(t+1)$  is the value of  $Q$  after the next clock pulse.



| S | R | $Q(t+1)$           |
|---|---|--------------------|
| 0 | 0 | $Q(t)$ (no change) |
| 0 | 1 | 0 (reset to 0)     |
| 1 | 0 | 1 (set to 1)       |
| 1 | 1 | undefined          |

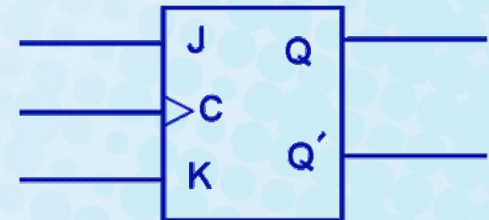
## 3.7 Sequential Circuits (8 of 12)

- The SR flip-flop actually has three inputs: S, R, and its current output, Q.
- Thus, we can construct a truth table for this circuit, as shown at the right.
- Notice the two undefined values. When both S and R are 1, the SR flip-flop is unstable.

| Present State |   |      | Next State |
|---------------|---|------|------------|
| S             | R | Q(t) | Q(t+1)     |
| 0             | 0 | 0    | 0          |
| 0             | 0 | 1    | 1          |
| 0             | 1 | 0    | 0          |
| 0             | 1 | 1    | 0          |
| 1             | 0 | 0    | 1          |
| 1             | 0 | 1    | 1          |
| 1             | 1 | 0    | undefined  |
| 1             | 1 | 1    | undefined  |

## 3.7 Sequential Circuits (9 of 12)

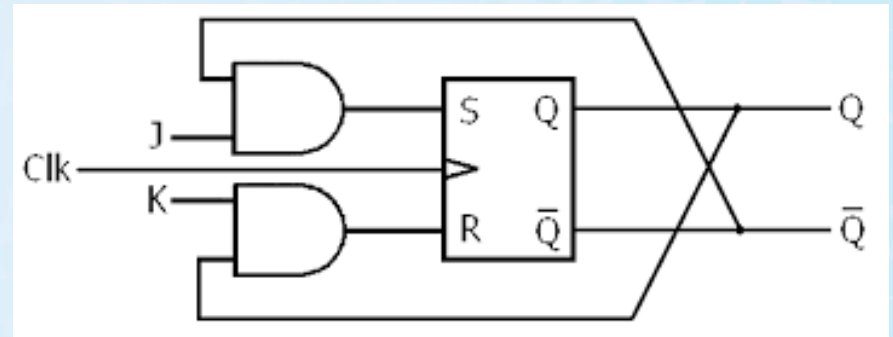
- If we can be sure that the inputs to an SR flip-flop will never both be 1, we will never have an unstable circuit. This may not always be the case.
- The SR flip-flop can be modified to provide a stable state when both inputs are 1.
- This modified flip-flop is called a JK flip-flop, shown at the right.





## 3.7 Sequential Circuits (10 of 12)

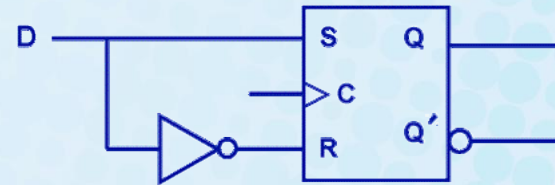
- At the right, we see how an SR flip-flop can be modified to create a JK flip-flop.
- The characteristic table indicates that the flip-flop is stable for all inputs.



| J | K | $Q(t+1)$           |
|---|---|--------------------|
| 0 | 0 | $Q(t)$ (no change) |
| 0 | 1 | 0 (reset to 0)     |
| 1 | 0 | 1 (set to 1)       |
| 1 | 1 | $Q'(t)$            |

## 3.7 Sequential Circuits (11 of 12)

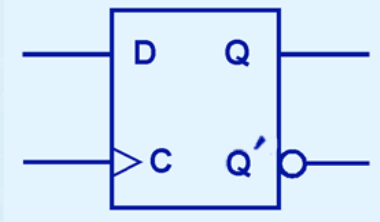
- Another modification of the SR flip-flop is the D flip-flop, shown below with its characteristic table.
- You will notice that the output of the flip-flop remains the same during subsequent clock pulses. The output changes only when the value of D changes.



| D | $Q(t+1)$ |
|---|----------|
| 0 | 0        |
| 1 | 1        |

## 3.7 Sequential Circuits (12 of 12)

- The D flip-flop is the fundamental circuit of computer memory.
  - D flip-flops are usually illustrated using the block diagram shown below.
- The characteristic table for the D flip-flop is shown at the right.



| D | $Q(t+1)$ |
|---|----------|
| 0 | 0        |
| 1 | 1        |

# Conclusion (1 of 3)

- Computers are implementations of Boolean logic.
- Boolean functions are completely described by truth tables.
- Logic gates are small circuits that implement Boolean operators.
- The basic gates are AND, OR, and NOT.
  - The XOR gate is very useful in parity checkers and adders.
- The “universal gates” are NOR, and NAND.



# Conclusion (2 of 3)

- Computer circuits consist of combinational logic circuits and sequential logic circuits.
- Combinational circuits produce outputs (almost) immediately when their inputs change.
- Sequential circuits require clocks to control their changes of state.
- The basic sequential circuit unit is the flip-flop: The behaviors of the SR, JK, and D flip-flops are the most important to know.

# Conclusion (3 of 3)

- The behavior of sequential circuits can be expressed using characteristic tables or through various finite state machines.