

# Optimization and Data Science

## Lecture 4: Fast Fourier Transformation and Applications

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## 1 Fast Fourier Transformation and Applications

- Recall: Discrete Fourier Transformation
- Fast Fourier Transformation (FFT)
- Motivating Example
- Derivation in the General Case
- FFT for real-valued data
- Interpretation and Application

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## Recall: DFT and inverse DFT as matrix-vector product

- The DFT mapping  $z \mapsto c$  is given by a matrix-vector multiplication

$$c = \frac{1}{m} M z \quad \text{with } M := \left( e^{-i \frac{2\pi k j}{m}} \right)_{k,j=0}^{m-1} \in \mathbb{C}^{m \times m}.$$

- The inverse DFT mapping  $c \mapsto z$  is performed with the inverse matrix:

$$z = m M^{-1} c.$$

Returns values  $z_j = f(j \frac{L}{m})$  of function  $f$  at equidistant points from its Fourier coefficients.

- The inverse matrix is given by

$$M^{-1} = \frac{1}{m} \left( e^{i \frac{2\pi k j}{m}} \right)_{k,j=0}^{m-1} \in \mathbb{C}^{m \times m}.$$

- **Note:** Some literature/algorithms define the DFT coefficients  $c_k$  without factor  $\frac{1}{m}$ . Then the factor  $m$  has to be omitted in the inverse DFT.

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# Fast Fourier Transformation (FFT)

- What is it?

Efficient implementation of the DFT

Only  $\mathcal{O}(m \log m)$  operations instead of  $\mathcal{O}(m^2)$  (standard matrix-vector product)

- Why are we studying this?

Named one of the Top Ten algorithms of the 20th century,  
in *Computing in Science & Engineering* 2000,  
American Institute of Physics, IEEE Computer Society,  
see <https://archive.siam.org/pdf/news/637.pdf>

- How does it work?

“Divide and conquer”

- What if we can use it?

Significant acceleration of Fourier analysis and synthesis

# Fast Fourier Transformation

- We exploit the periodicity of the function

$$t \mapsto e^{it} = \cos t + i \sin t.$$

- Defining  $\omega_m := e^{-i\frac{2\pi}{m}}$ , we get

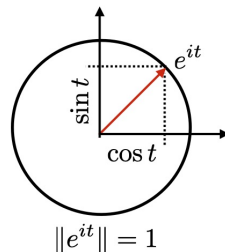
$$\omega_m^{kj} = \left( e^{-i\frac{2\pi}{m}} \right)^{kj} = e^{-i\frac{2\pi kj}{m}}.$$

- The Transformation matrix can be written as:

$$M_m = \left( e^{-i\frac{2\pi kj}{m}} \right)_{k,j=0}^{m-1} = \left( \omega_m^{kj} \right)_{k,j=0}^{m-1} \in \mathbb{C}^{m \times m}$$

- ... and the DFT as

$$c_k = \frac{1}{m} \sum_{j=0}^{m-1} z_j e^{-i\frac{2\pi kj}{m}} = \frac{1}{m} \sum_{j=0}^{m-1} z_j \omega_m^{kj}, \quad k = 0, \dots, m-1.$$



# Fast Fourier Transformation

- Transformation matrix

$$M_m = \left( e^{-i \frac{2\pi kj}{m}} \right)_{k,j=0}^{m-1} = \left( \omega_m^{kj} \right)_{k,j=0}^{m-1}$$

- We compute some special values:

$$\begin{aligned} kj = m : \quad \omega_m^m &= e^{-i \frac{2\pi m}{m}} = e^{-i 2\pi} = \cos(-2\pi) + i \sin(-2\pi) = 1 + i \cdot 0 = 1 \\ kj = m + \ell : \quad \omega_m^{m+\ell} &= \omega_m^m \omega_m^\ell = 1 \cdot \omega_m^\ell = \omega_m^\ell. \end{aligned} \quad (1)$$

- ... and if  $m = 2n, n \in \mathbb{N}$ :

$$\begin{aligned} kj = \frac{m}{2} = n : \quad \omega_m^n &= e^{-i \frac{2\pi n}{2m}} = e^{-i \pi} = \cos(-\pi) + i \sin(-\pi) = -1 + i \cdot 0 = -1 \\ kj = n + \ell : \quad \omega_m^{n+\ell} &= \omega_m^n \omega_m^\ell = -\omega_m^\ell \end{aligned} \quad (2)$$



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## Example ( $m = 4$ )

- Transformation matrix:

$$M_4 = \left( \omega_4^{kj} \right)_{k,j=0}^3 = \begin{pmatrix} \omega_4^0 & \omega_4^0 & \omega_4^0 & \omega_4^0 \\ \omega_4^0 & \omega_4^1 & \omega_4^2 & \omega_4^3 \\ \omega_4^0 & \omega_4^2 & \omega_4^4 & \omega_4^6 \\ \omega_4^0 & \omega_4^3 & \omega_4^6 & \omega_4^9 \end{pmatrix} \begin{array}{l} \leftarrow k=0 \\ \leftarrow k=1 \\ \leftarrow k=2 \\ \leftarrow k=3 \end{array}$$

$$j = \begin{array}{cccc} \uparrow & \uparrow & \uparrow & \uparrow \\ 0 & 1 & 2 & 3 \end{array}$$

- Periodicity: (1) on page 8  $\Rightarrow \omega_4^0 = \omega_4^4 = 1, \omega_4^{4+\ell} = \omega_4^\ell$  gives

$$M_4 = \begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & \omega_4 & \omega_4^2 & \omega_4^3 \\ 1 & \omega_4^2 & 1 & \omega_4^2 \\ 1 & \omega_4^3 & \omega_4^2 & \omega_4 \end{pmatrix}$$

## Example ( $m = 4$ )

- Transformation, using periodicity:

$$\begin{pmatrix} c_0 \\ c_1 \\ c_2 \\ c_3 \end{pmatrix} = \frac{1}{4} M_4 \begin{pmatrix} z_0 \\ z_1 \\ z_2 \\ z_3 \end{pmatrix} = \frac{1}{4} \begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & \omega_4 & \omega_4^2 & \omega_4^3 \\ 1 & \omega_4^2 & 1 & \omega_4^2 \\ 1 & \omega_4^3 & \omega_4^2 & \omega_4 \end{pmatrix} \begin{pmatrix} z_0 \\ z_1 \\ z_2 \\ z_3 \end{pmatrix}$$

- Exchange second and third row (means: split w.r.t. even and odd indices):

$$\begin{pmatrix} c_0 \\ c_2 \\ c_1 \\ c_3 \end{pmatrix} = \frac{1}{4} \begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & \omega_4^2 & 1 & \omega_4^2 \\ 1 & \omega_4 & \omega_4^2 & \omega_4^3 \\ 1 & \omega_4^3 & \omega_4^2 & \omega_4 \end{pmatrix} \begin{pmatrix} z_0 \\ z_1 \\ z_2 \\ z_3 \end{pmatrix}$$

## Example ( $m = 4$ )

- Transformation matrix with exchanged rows can be written as product of two matrices:

$$\begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & \omega_4^2 & 1 & \omega_4^2 \\ 1 & \omega_4 & \omega_4^2 & \omega_4^3 \\ 1 & \omega_4^3 & \omega_4^2 & \omega_4 \end{pmatrix} = \left( \begin{array}{cc|cc} 1 & 1 & 0 & 0 \\ 1 & \omega_4^2 & 0 & 0 \\ \hline 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & \omega_4^2 \end{array} \right) \left( \begin{array}{cc|cc} 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & 1 \\ \hline 1 & 0 & \omega_4^2 & 0 \\ 0 & \omega_4 & 0 & \omega_4^3 \end{array} \right)$$

where we used (see page 8):

$$\omega_4^4 = e^{-i\frac{2\pi 4}{4}} = e^{-i2\pi} = \cos(-2\pi) + i \sin(-2\pi) = 1 + i \cdot 0 = 1.$$

and

$$\omega_4^5 = \omega_4^4 \omega_4^1 = 1 \cdot \omega_4^1 = \omega_4.$$

## Example ( $m = 4$ )

- Transformation matrix with exchanged rows was written as product of two matrices:

$$\begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & \omega_4^2 & 1 & \omega_4^2 \\ 1 & \omega_4 & \omega_4^2 & \omega_4^3 \\ 1 & \omega_4^3 & \omega_4^2 & \omega_4 \end{pmatrix} = \left( \begin{array}{cc|cc} 1 & 1 & 0 & 0 \\ 1 & \omega_4^2 & 0 & 0 \\ \hline 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & \omega_4^2 \end{array} \right) \left( \begin{array}{cc|cc} 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & 1 \\ \hline 1 & 0 & \omega_4^2 & 0 \\ 0 & \omega_4 & 0 & \omega_4^3 \end{array} \right).$$

- The second block matrix can be simplified using (2) on page 8:  $\omega_4^2 = -1, \omega_4^3 = -\omega_4$ :

$$\left( \begin{array}{cc|cc} 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & 1 \\ \hline 1 & 0 & \omega_4^2 & 0 \\ 0 & \omega_4 & 0 & \omega_4^3 \end{array} \right) = \left( \begin{array}{cc|cc} 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & 1 \\ \hline 1 & 0 & -1 & 0 \\ 0 & \omega_4 & 0 & -\omega_4 \end{array} \right) = \begin{pmatrix} I_2 & I_2 \\ D_2 & -D_2 \end{pmatrix}$$

with the identity matrix  $I_2$  and the diagonal matrix  $D_2 = \text{diag}(1, \omega_4)$ .

## Example ( $m = 4$ ): Summary

- Transformation matrix with exchanged rows can be written as product of two matrices:

$$\begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & \omega_4^2 & 1 & \omega_4^2 \\ 1 & \omega_4 & \omega_4^2 & \omega_4^3 \\ 1 & \omega_4^3 & \omega_4^2 & \omega_4 \end{pmatrix} = \left( \begin{array}{cc|cc} 1 & 1 & 0 & 0 \\ 1 & \omega_4^2 & 0 & 0 \\ \hline 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & \omega_4^2 \end{array} \right) \left( \begin{array}{cc|cc} 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & 1 \\ \hline 1 & 0 & -1 & 0 \\ 0 & \omega_4 & 0 & -\omega_4 \end{array} \right)$$

$$= \begin{pmatrix} M_2 & 0_2 \\ 0_2 & M_2 \end{pmatrix} \begin{pmatrix} I_2 & I_2 \\ D_2 & -D_2 \end{pmatrix}$$

where all appearing block matrices have size  $(2 \times 2)$ :

$M_2$  : the transformation matrix of half size,  $0_2$  : zero matrix,

$I_2$  : identity matrix,  $D_2 = \text{diag}(1, \omega_4)$  : diagonal matrix.

⇒ We can solve the problem of size  $m = 4$  by solving two problems of size  $\frac{m}{2} = 2$ .

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## Generalization ( $m = 2n, n \in \mathbb{N}$ )

- We use again (1) on page 8:

$$\omega_m^{2\ell n} = \omega_m^{\ell m} = (\omega_m^m)^\ell = 1^\ell = 1$$

and

$$\omega_m^2 = e^{-i\frac{4\pi}{m}} = e^{-i\frac{4\pi}{2n}} = e^{-i\frac{2\pi}{n}} = \omega_n$$

- ... for the Fourier coefficients with even indices  $k = 2\ell$ :

$$\begin{aligned} c_{2\ell} &= \frac{1}{m} \sum_{j=0}^{2n-1} z_j \omega_m^{2\ell j} = \frac{1}{m} \left( \sum_{j=0}^{n-1} z_j \omega_m^{2\ell j} + \sum_{j=0}^{n-1} z_{n+j} \omega_m^{2\ell(n+j)} \right) \\ &= \frac{1}{m} \sum_{j=0}^{n-1} (z_j + z_{n+j}) \omega_m^{2\ell j} = \frac{1}{m} \sum_{j=0}^{n-1} (z_j + z_{n+j}) \omega_n^{\ell j}. \end{aligned}$$



# Generalization ( $m = 2n, n \in \mathbb{N}$ )

- Summary:

$$c_{2\ell} = \frac{1}{m} \sum_{j=0}^{n-1} (z_j + z_{n+j}) \omega_n^{\ell j}, \quad \ell = 0, \dots, n-1.$$

- Thus for the even indices:

$$c_{\text{even}} := \begin{pmatrix} c_0 \\ \vdots \\ c_{2(n-1)} \end{pmatrix} = \frac{1}{m} M_n \begin{pmatrix} z_0 + z_n \\ \vdots \\ z_{n-1} + z_{2(n-1)} \end{pmatrix} = \frac{1}{m} M_n \left( I_n \mid I_n \right) z$$

with  $I_n$  being the identity matrix of half size  $n = \frac{m}{2}$ .

## Generalization ( $m = 2n, n \in \mathbb{N}$ )

- We use, see (2) on page 8:

$$\omega_m^{(2\ell+1)(n+j)} = \omega_m^{2\ell n} \omega_m^{n+(2\ell+1)j} = \omega_m^{n+(2\ell+1)j} = -\omega_m^{(2\ell+1)j}$$

for the odd indices  $k = 2\ell + 1$ :

$$\begin{aligned} c_{2\ell+1} &= \frac{1}{m} \sum_{j=0}^{2n-1} z_j \omega_m^{(2\ell+1)j} = \frac{1}{m} \left( \sum_{j=0}^{n-1} z_j \omega_m^{(2\ell+1)j} + \sum_{j=0}^{n-1} z_{n+j} \omega_m^{(2\ell+1)(n+j)} \right) \\ &= \frac{1}{m} \sum_{j=0}^{n-1} z_j \omega_m^{(2\ell+1)j} - \sum_{j=0}^{n-1} z_{n+j} \omega_m^{(2\ell+1)j} \\ &= \frac{1}{m} \sum_{j=0}^{n-1} (z_j - z_{n+j}) \omega_m^j \omega_m^{2\ell j} = \frac{1}{m} \sum_{j=0}^{n-1} (z_j - z_{n+j}) \omega_m^j \omega_n^{\ell j}. \end{aligned}$$

## Generalization ( $m = 2n, n \in \mathbb{N}$ )

- Summary:

$$c_{2\ell+1} = \frac{1}{m} \sum_{j=0}^{n-1} (z_j - z_{n+j}) \omega_m^j \omega_n^{\ell j}, \quad \ell = 0, \dots, n-1.$$

- Thus, for the odd indices:

$$c_{\text{odd}} := \begin{pmatrix} c_1 \\ \vdots \\ c_{2n-1} \end{pmatrix} = \frac{1}{m} M_n \begin{pmatrix} (z_0 - z_n) \omega_m^0 \\ \vdots \\ (z_{n-1} - z_{2n-1}) \omega_m^{n-1} \end{pmatrix} = \frac{1}{m} M_n \left( D_n \mid -D_n \right) z$$

with the diagonal matrix

$$D_n = \text{diag} \left( \omega_{2n}^0, \omega_{2n}^1, \omega_{2n}^2, \dots, \omega_{2n}^{n-1} \right).$$

## Algorithmic realization in the general case ( $m = 2n, n \in \mathbb{N}$ )

- Transformation process  $z \mapsto c$  for  $c, z \in \mathbb{R}^{2n}$ , written as matrix-vector product

$$c = \frac{1}{2n} M_{2n} z, \quad M_{2n} \in \mathbb{R}^{2n \times 2n},$$

- ... can be written as:

$$\begin{pmatrix} c_{\text{even}} \\ c_{\text{odd}} \end{pmatrix} = \frac{1}{2n} \begin{pmatrix} M_n & 0_n \\ 0_n & M_n \end{pmatrix} \begin{pmatrix} I_n & I_n \\ D_n & -D_n \end{pmatrix} z$$

- Last matrix-vector product** is multiplication with diagonal matrix  $\rightsquigarrow$  effort  $\mathcal{O}(n)$ .
- $\rightsquigarrow$  Effort of problem of size  $2n \approx 2 \times$  effort of problem of size  $n$ .
- $\rightsquigarrow$  Recursive application (“divide and conquer”)
- $\rightsquigarrow$   $m = 2^k, k = \log_2 m$ : effort  $\mathcal{O}(m \log m)$  instead of  $\mathcal{O}(m^2)$  for **standard matrix-vector product**.

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## FFT for real-valued data

For **real-valued data**, complex coefficients are transformed into real ones using Euler's formula

$$e^{it} = \cos t + i \sin t.$$

### Theorem

Let  $m = 2n$ ,  $n \in \mathbb{N}$ ,  $z \in \mathbb{R}^m$  and  $c = (c_k)_{k=0}^{m-1} \in \mathbb{C}^m$  be the corresponding complex Fourier coefficients. Then

$$\frac{a_0}{2} + \sum_{k=1}^{n-1} \left( a_k \cos \frac{2\pi k t_j}{L} + b_k \sin \frac{2\pi k t_j}{L} \right) + \frac{a_n}{2} \cos \frac{2\pi n t_j}{L} = z_j, \quad t_j = j \frac{L}{m}, \quad j = 0, \dots, m-1, \quad (3)$$

where

$$a_k = 2 \operatorname{Re} c_k, \quad k = 0, \dots, n, \quad b_k = -2 \operatorname{Im} c_k, \quad k = 1, \dots, n-1.$$

$\operatorname{Re} z \in \mathbb{R}$  and  $\operatorname{Im} z \in \mathbb{R}$  denote the real and imaginary part of  $z \in \mathbb{C}$ , i.e.,  $z = \operatorname{Re} z + i \operatorname{Im} z$ .

# FFT for real-valued data

- The method above, i.e., treating real-valued data  $z \in \mathbb{R}^m$  as complex values  $z \in \mathbb{C}^m$  with zero imaginary part, wastes storage ( $m$  complex  $\hat{=}$   $2m$  real numbers for  $m$  real data).
- Alternative: given  $z = (z_j)_{j=0}^{m-1} \in \mathbb{R}^m$ ,  $m = 2n$ , create complex data  $\tilde{z} \in \mathbb{C}^n$  as

$$\tilde{z}_\ell := z_{2\ell} + iz_{2\ell+1}, \quad \ell = 0, \dots, n-1.$$

- Apply (complex) DFT to  $\tilde{z}$ , giving coefficients  $c = (c_k)_{k=0}^{n-1} \in \mathbb{C}^n$ .
- Then, the coefficients  $a_k, b_k$  used in (3) on page 22 can be computed as

$$a_k = \operatorname{Re} \left( \frac{1}{2}(c_k + \bar{c}_{n-k}) + \frac{1}{2i}(c_k - \bar{c}_{n-k})e^{-\frac{ik\pi}{n}} \right), \quad k = 0, 1, \dots, n,$$

$$b_k = -\operatorname{Im} \left( \frac{1}{2}(c_k + \bar{c}_{n-k}) + \frac{1}{2i}(c_k - \bar{c}_{n-k})e^{-\frac{ik\pi}{n}} \right), \quad k = 1, \dots, n,$$

with  $c_n = c_0$ .

- Here,  $\bar{w}$  denotes the **complex conjugate** of  $w = x + iy \in \mathbb{C}$ , i.e.,  $\bar{w} := x - iy \in \mathbb{C}$ .

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# Interpretation and application

- Assume we have given data  $z \in \mathbb{R}^m$
- ... and computed the real Fourier coefficients  $a, b$ .
- Interpretation of the formula of the Theorem on page 22:

$$\frac{a_0}{2} + \sum_{k=1}^{n-1} \left( a_k \cos \frac{2\pi kt_j}{L} + b_k \sin \frac{2\pi kt_j}{L} \right) + \frac{a_n}{2} \cos \frac{2\pi nt_j}{L} = z_j, \quad t_j = j \frac{L}{m}, j = 0, \dots, m-1.$$

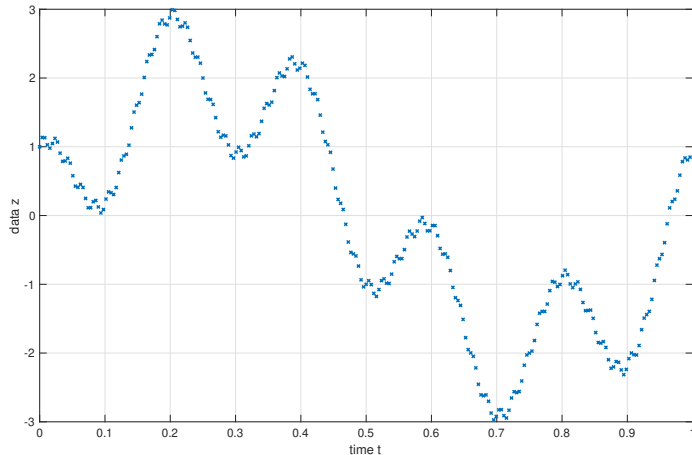
- We can deduce which frequencies  $k$  are dominant.
- We can omit frequencies with small Fourier coefficients  $\rightsquigarrow$  data compression.
- We can detect and omit high frequencies (which might be random perturbations).

## Example (last lecture)

- Given: dataset

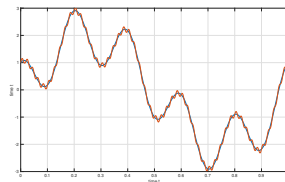
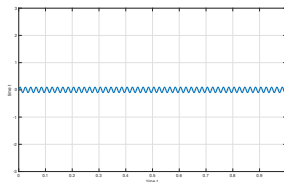
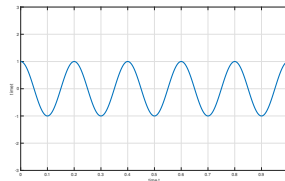
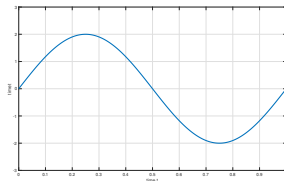
$$(t_j, z_j)_{j=0, \dots, m-1}, t_j, z_j \in \mathbb{R}.$$

- Example:  $t$  time,  
 $z$  measurements.
- We see “some”  
structure ...
- How to analyze this?



# Example

- Here we have the Fourier coefficients  $b_1 = 2$ ,  $a_5 = 1$ ,  $b_{50} = 0.1$ .
- The Fourier coefficients are the amplitudes of the periodic parts in the data corresponding to the frequency  $k$ .



# Multi-dimensional Fourier Transformation

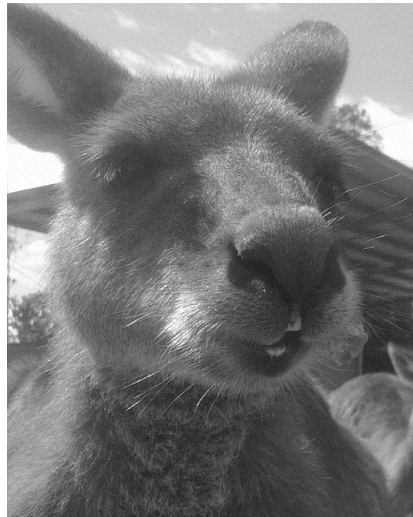
- ... is sequentially applied for each dimension.
- Example: color pictures are 3-D data
  - Image needs 1 Byte per color and pixel
  - Apply DFT for each color (e.g., 3 times for RGB pictures)
  - Apply for each row in a 2-D picture
  - Apply for the resulting columns
  - JPEG includes FFT as one step.
- Compression: delete the Fourier coefficients  $c_k \in \mathbb{C}$  with  $|c_k| < \epsilon$ , a given threshold.
- Noise elimination: delete high frequencies

## Example: Image compression with DFT

original size: 1060759



compressed, #fft coeff: 107264



## Example: Image compression with DFT

compressed, #fft coeff: 50089



compressed, #fft coeff: 3559



## Example: Denoising with DFT

**original**



**noisy**



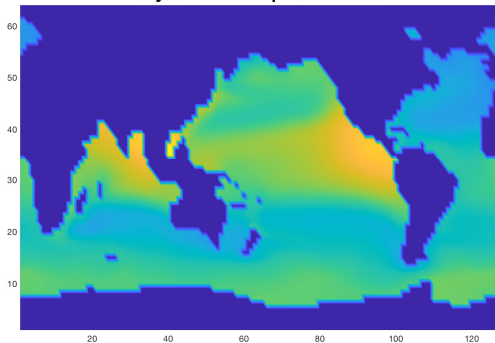
**filtered**



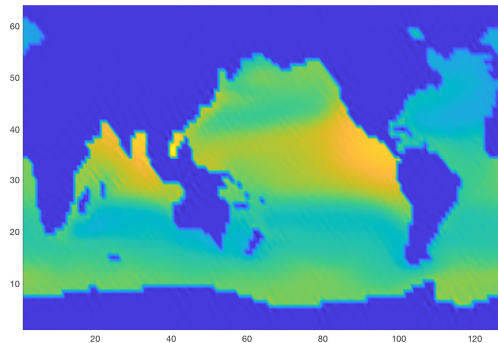
## Example: 3-dimensional data compression

Compression of 3-D data (simulation of nutrients in the ocean):

layer = 6 # data points: 52749



# modes = 1536 rel err = 0.013969





# Fourier Analysis: What is important

- Fourier analysis is an important and powerful tool to analyse, compress and denoise data in arbitrary dimensions.
- It performs a transformation of the data into the frequency domain.
- It is based on a transformation of the data considered as complex numbers.
- This transformation is a matrix-vector multiplication.
- Exploiting periodicity properties and applying the divide-and-conquer principle, a very efficient algorithm (the FFT) was developed.
- The FFT reduces the effort to  $\mathcal{O}(m \log m)$ , where  $m$  is the dimension of the data.
- Real-valued data are considered as complex-valued data, or considered as real and imaginary parts of complex numbers (to obtain even higher efficiency).
- Hardware-optimized implementations of the FFT are available in nearly all languages.
- It does not make sense to write another one, but to use a library function instead.