

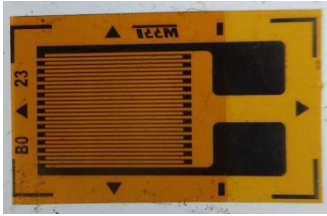
ANALOG CIRCUITS DESIGN

AE6: Operational Amplifiers V,
Differential circuits – instrumentation amplifiers

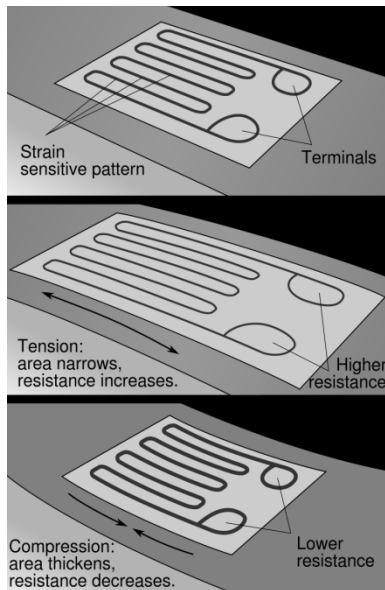
Course overview

1. Single ended versus differential
2. The difference amplifier
3. Common-mode analysis
4. Applications of differential circuits in the digital world

1. Single-ended versus differential: the challenge



The challenge is to measure a mechanical force using a strain gauge which is an insulating flexible backing supporting a metallic foil pattern.



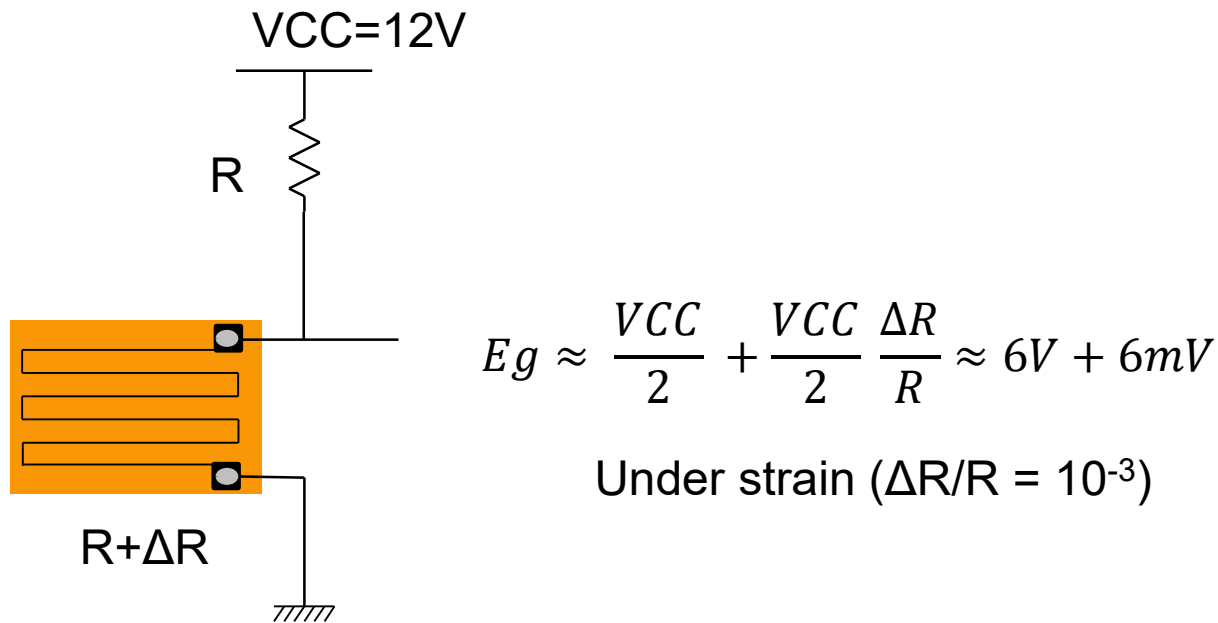
At rest, the metallic pattern exhibits a nominal resistance. The strain causes a deformation of the metallic foil and:

- When stretched, the metal becomes longer and narrower: its resistance increases,
- When compressed, the metal becomes shorter and wider: its resistance decreases



The relative variation of the resistance ($\Delta R/R$) is however in the $\pm 10^{-3}$ range, i.e. for a 100Ω nominal resistance, variation is only $\pm 0.1\Omega$

1. Single-ended versus differential: Single-ended measurement

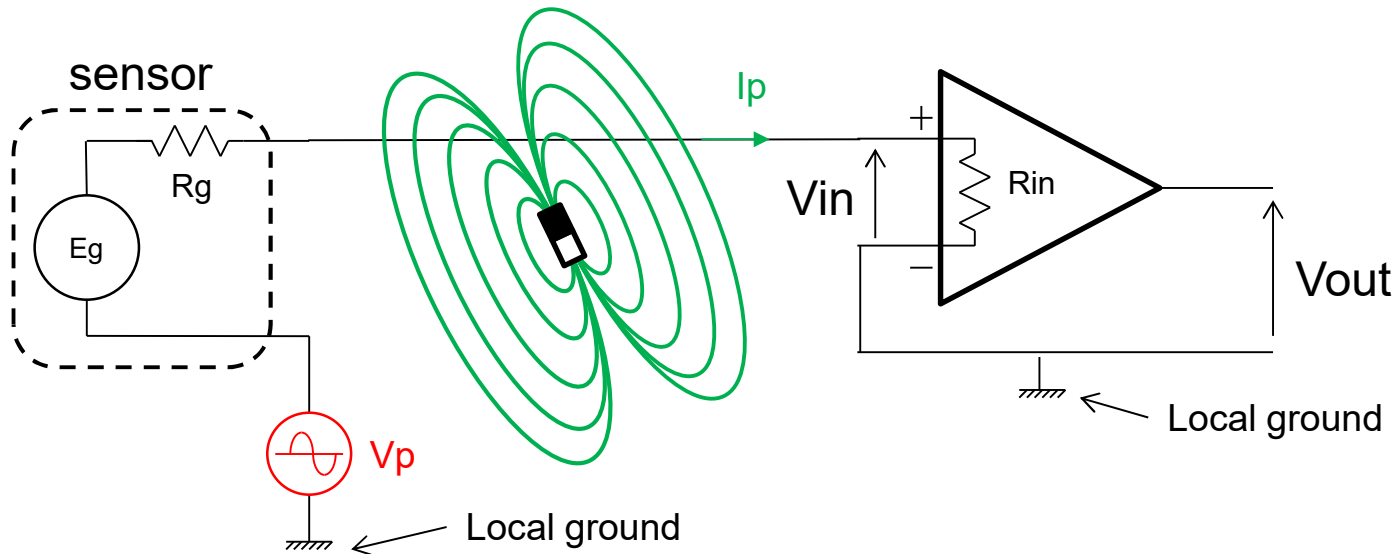


Single-ended arrangement: the signal to measure Eg is referred to the ground

Eg exhibits a large constant bias but very little variation in presence of strain

The large bias, unless it is cancelled, prohibits high gain for the input amplifier

1. Single-ended versus differential: single-ended measurement



Assuming that the bias is cancelled in E_g , only the small variation remains and:

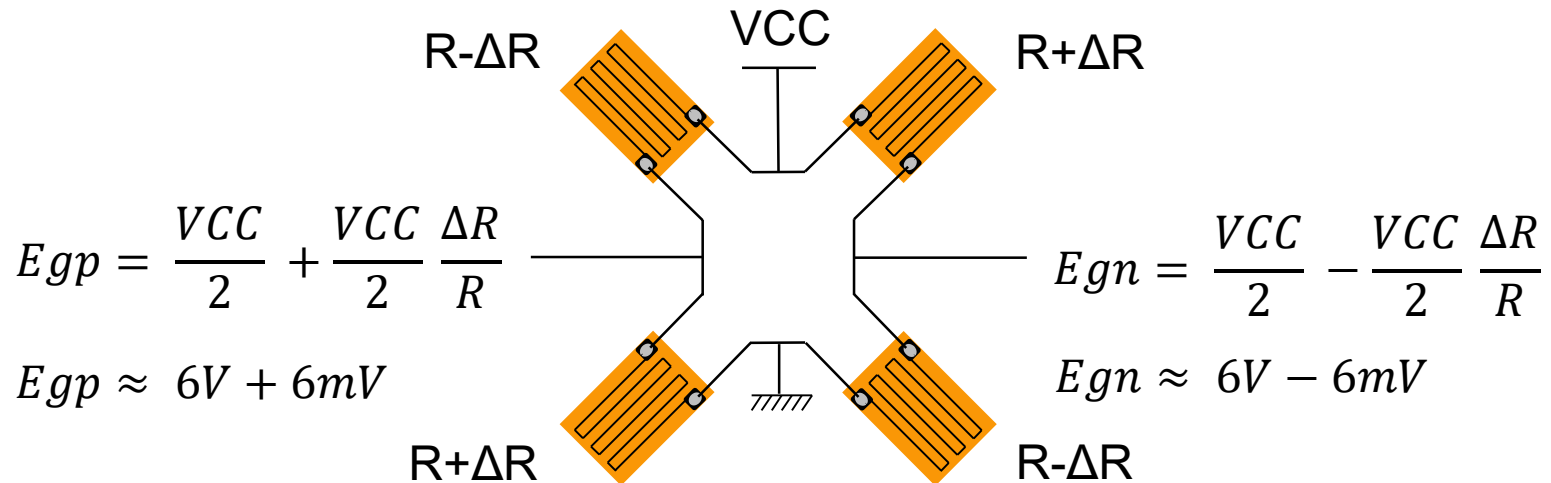
$$V_{out} = A V_{in} = A \left(E_g \frac{R_{in}}{R_{in} + R_g} + V_p \frac{R_{in}}{R_{in} + R_g} + I_p R_{in} \right)$$

V_p : parasitic voltage due to a large distance between the two local ground points

I_p : parasitic current created by an electromagnetic field

1. Single-ended versus differential: differential measurement

Under strain ($\Delta R/R = 10^{-3}$)

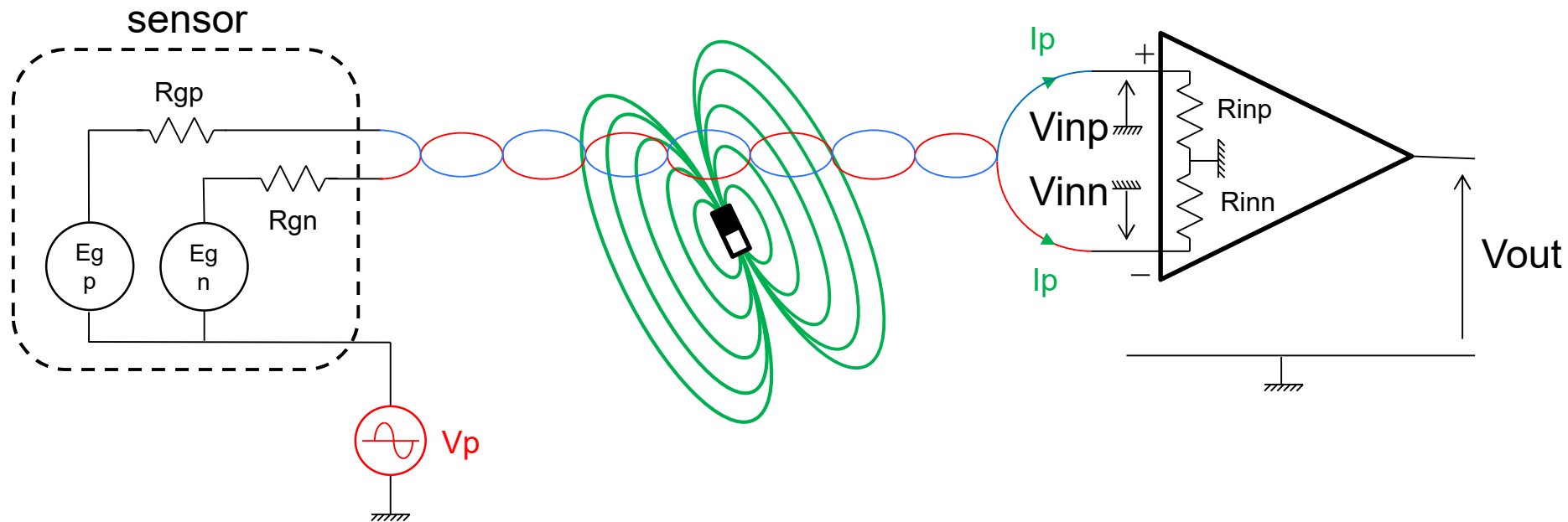


Differential arrangement: the signal to measure is now $E_{gp} - E_{gn}$

Ground still exists but is no longer the reference

The bias can be any, only the difference between E_{gp} and E_{gn} matters

1. Single-ended versus differential: differential measurement

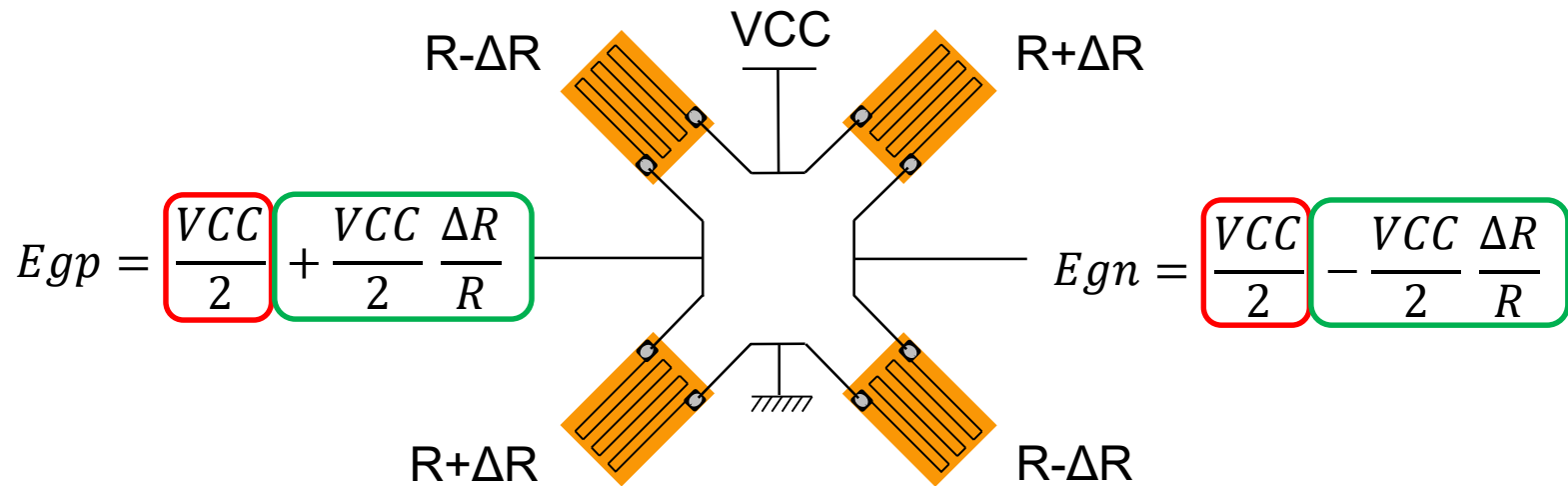


$$\begin{aligned}
 V_{out} &= A_p V_{inp} - A_n V_{inn} = A_p E_{gp} \frac{R_{inp}}{R_{inp} + R_{gp}} - A_n E_{gn} \frac{R_{inn}}{R_{inn} + R_{gn}} \\
 &\quad + A_p V_p \frac{R_{inp}}{R_{inp} + R_{gp}} - A_n V_p \frac{R_{inn}}{R_{inn} + R_{gn}} \\
 &\quad + A_p I_p R_{inp} - A_n I_p R_{inn}
 \end{aligned}$$

Parasitic terms cancel if symmetry is ensured :

- At amplifier's level $\rightarrow A_p = A_n, R_{inp} = R_{inn}$
- At sensor level $\rightarrow R_{gn} = R_{gp}$

1. Single-ended versus differential: common mode & differential mode



Differential operation involves two voltages (for example E_{gp} and E_{gn})

Common-mode voltage (V_{cm}): the amount which is the same on both E_{gp} and E_{gn} .

→ Here, $V_{cm} = V_{CC}/2$

Differential-mode voltage (V_d): the amount which is the difference of the terms having the same value but opposite polarity on both E_{gp} and E_{gn} .

→ Here, $V_d = V_{CC} (\Delta R/R)$

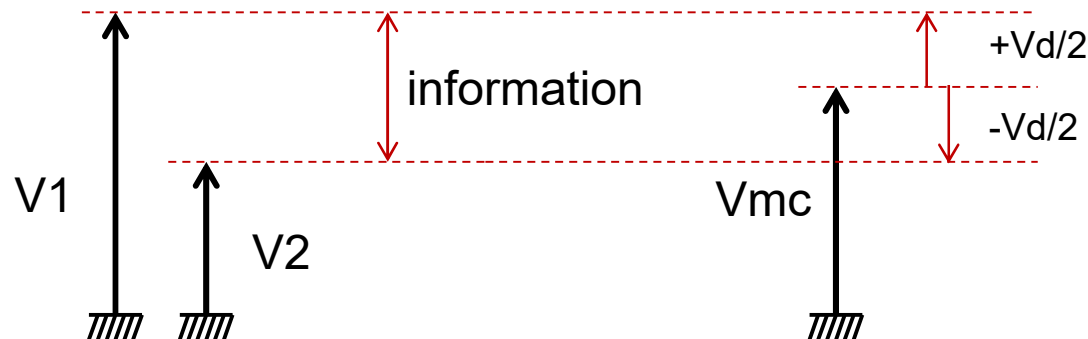
1. Single-ended versus differential: common mode & differential mode

In a general way, when information is conveyed by two voltages $V1$ and $V2$:

$V1$ and $V2$ can be expressed as $V1 = V_{cm} + \frac{V_d}{2}$ and $V2 = V_{cm} - \frac{V_d}{2}$

Information is carried by the differential-mode voltage $V_d = V1 - V2$

Perturbators are carried by the common-mode voltage $V_{cm} = \frac{V1+V2}{2}$



Goals for the differential amplifier:

Offering a stable and predictable gain for differential-mode signals

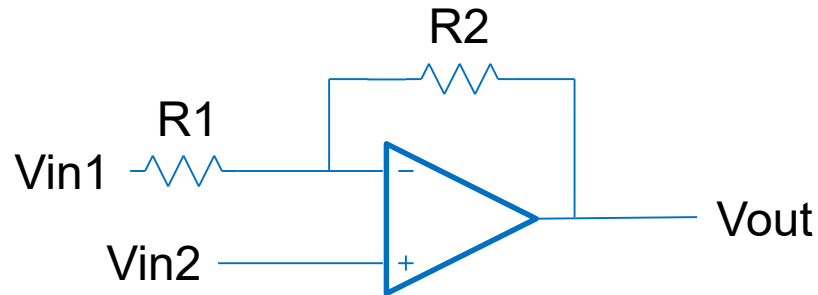
Reducing as much as possible the effects of common-mode perturbators (achieved through circuit symmetry)

2. The difference amplifier: basic circuit

Goals: the amplifier must exhibit at both inputs:



- The same gain with opposite sign
- The same (high) impedance

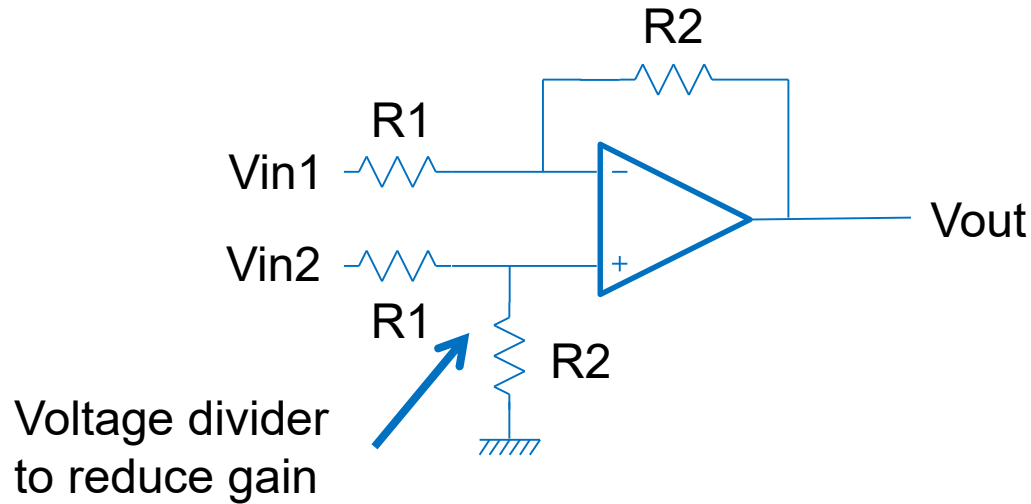



Gain:
$$V_{out} = \frac{-R2}{R1} V_{in1} + \left(1 + \frac{R2}{R1}\right) V_{in2}$$

Input impedance: $Z_{in1} = R1 \quad Z_{in2} \rightarrow \infty$

This circuit doesn't meet the expectations neither for gain nor input impedance

2. The difference amplifier: improved basic circuit



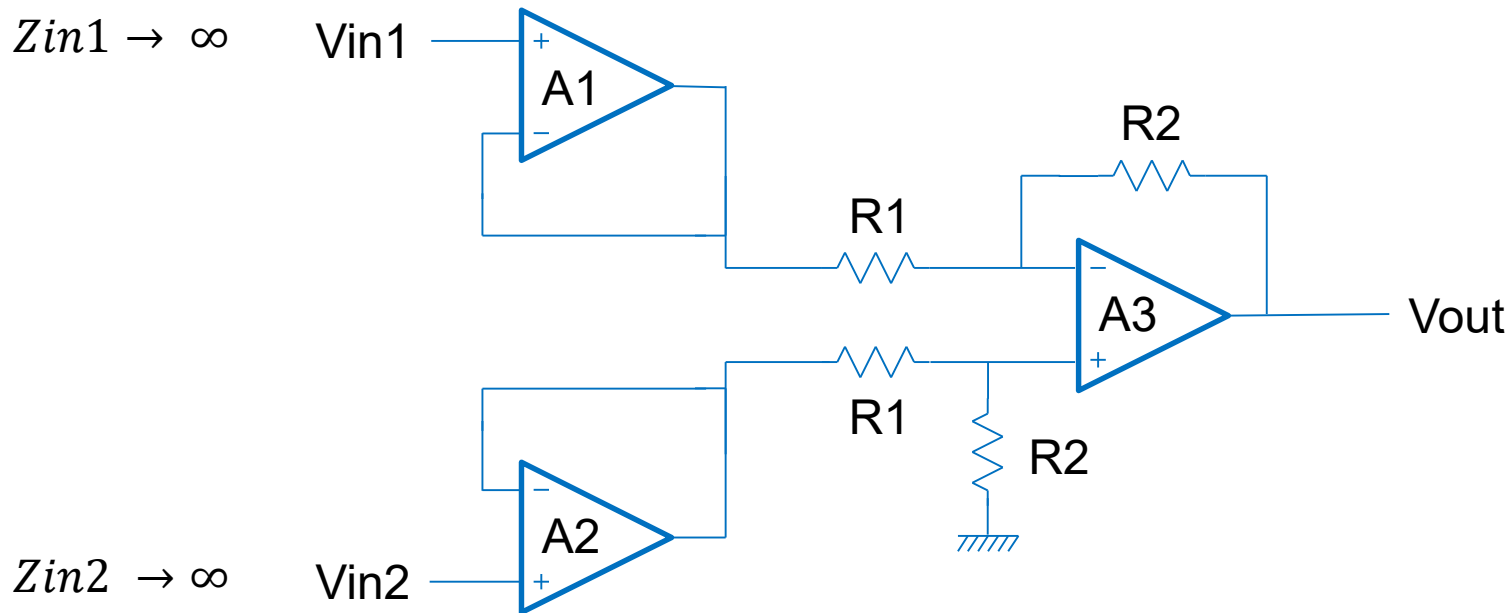
Gain: $V_{out} = \frac{R2}{R1} (V_{in2} - V_{in1})$  Importance of R2/R1 ratio matching

Input impedance: $Z_{in1} = R1$ $Z_{in2} = R1 + R2$

This circuit meets the expectations for gain but not for impedance

2. The difference amplifier: the instrumentation amplifier

Adding voltage followers increase and symmetrize input impedances



Problem: overall gain is only performed by last stage $\rightarrow R2/R1 > 1$

\rightarrow two external resistors needed

\rightarrow matching is difficult for external resistors

Gain of last stage is set to 1 ($R1 = R2$) \rightarrow matching greatly improved (internal res.)

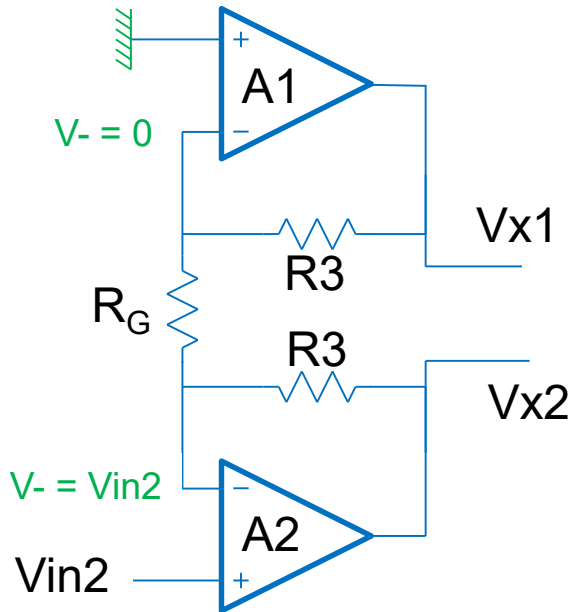
Gain is implemented on first stage (A1, A2) with only one external resistor R_G

2. The difference amplifier: the instrumentation amplifier

Which expression for the amplifier's gain?

1/ Amplifier's output voltage is:

$$V_{out} = \frac{R_2}{R_1} (V_{x2} - V_{x1})$$



2/ $V_{in2}=0$

3/ $V_{in1}=0$

$$V_{x1a} = \left(1 + \frac{R_3}{R_G}\right) V_{in1}$$

$$V_{x1b} = \frac{-R_3}{R_G} V_{in2}$$

$$V_{x2a} = \frac{-R_3}{R_G} V_{in1}$$

$$V_{x2b} = \left(1 + \frac{R_3}{R_G}\right) V_{in2}$$

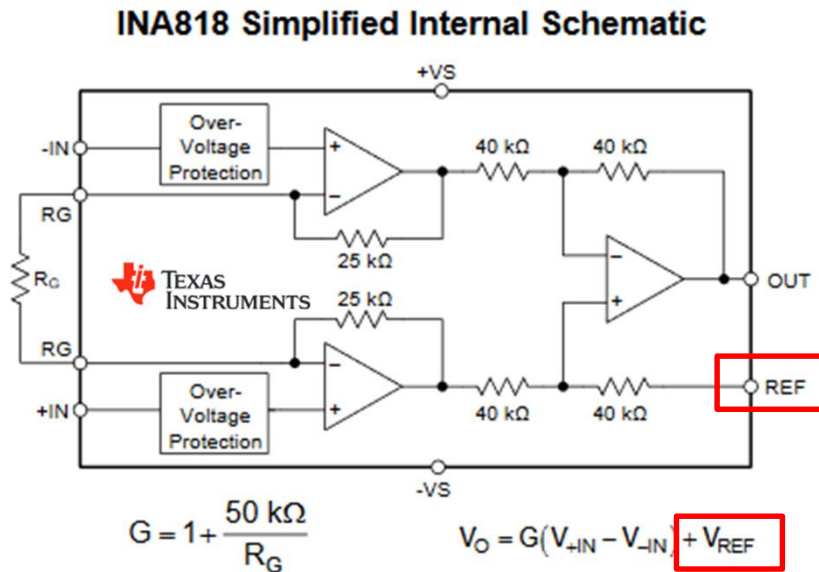
4/ Superposition: $V_{x1} = \left(1 + \frac{R_3}{R_G}\right) V_{in1} + \frac{-R_3}{R_G} V_{in2}$ $V_{x2} = \frac{-R_3}{R_G} V_{in1} + \left(1 + \frac{R_3}{R_G}\right) V_{in2}$

5/ Finally:

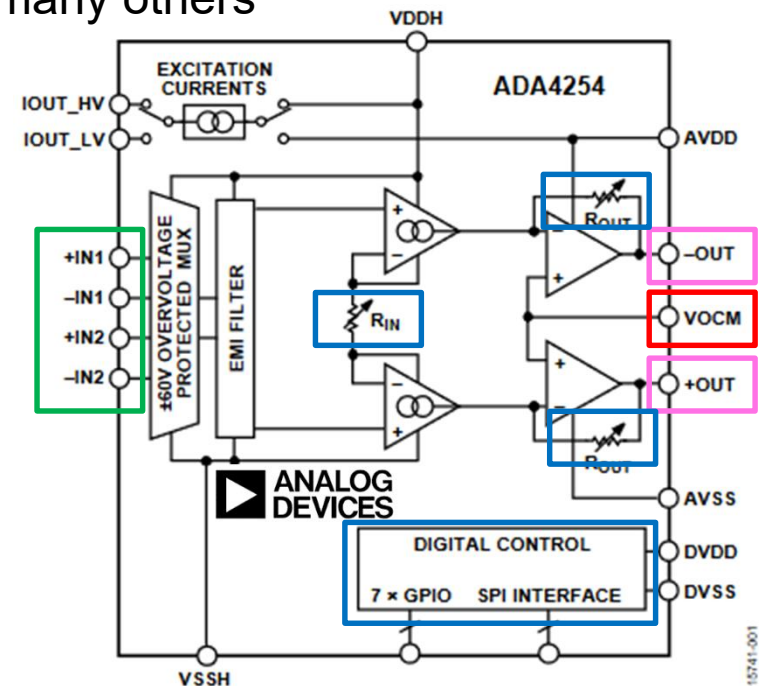
$$V_{out} = \frac{R_2}{R_1} \left(1 + \frac{2 R_3}{R_G}\right) (V_{in2} - V_{in1})$$

2. The difference amplifier: the instrumentation amplifier

Some commercially available circuits... among many others



- Second stage gain = 1 ($R_1=R_2=40\text{k}\Omega$)
- R_3 ($25\text{k}\Omega$) is laser-trimmed to guarantee an accurate absolute value
- Gain can be set between 1 and 1000
- Implements a V_{REF} input* to allow an offset at the output

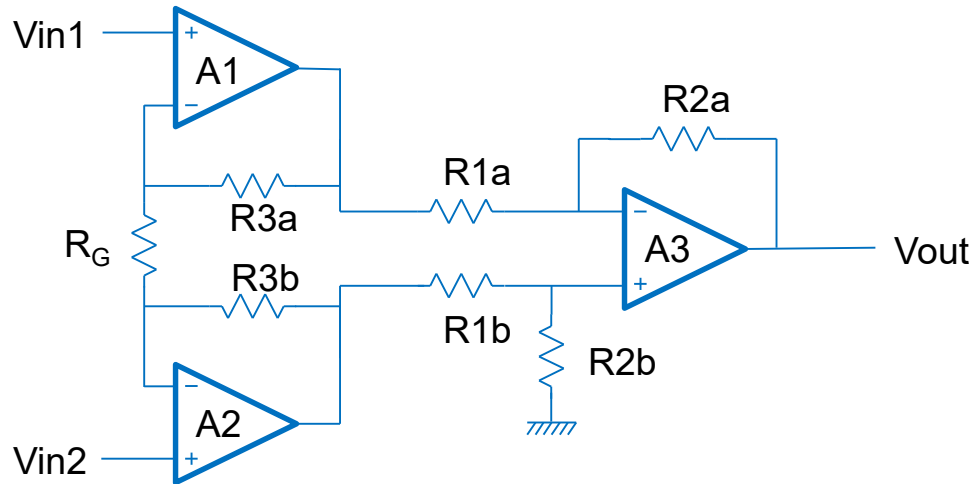


- Digitally programmable gain (36 different values)
- Calibration register for each gain value
- 2 multiplexed channels
- Differential output
- Implements a V_{OCM} input to allow an offset at the output

*The external voltage source connected to V_{REF} must exhibit a negligible output resistance with respect to $40\text{k}\Omega$

3. Common-mode analysis: when common-mode becomes differential..

Taking into account that resistors haven't exactly their nominal value:



The gain is slightly different for Vin1 and Vin2: $V_{out} = G_+ Vin2 - G_- Vin1$

Consequence if Vin1 and Vin2 have a non-zero common mode:

$$V_{out} = G_+ \left(V_{cm} + \frac{V_d}{2} \right) - G_- \left(V_{cm} - \frac{V_d}{2} \right) = \frac{V_d}{2} (G_+ + G_-) + V_{cm} (G_+ - G_-)$$

Slight gain error for $V_d/2$

V_{mc} is not fully canceled

A portion of the common-mode signal is viewed as a differential one

3. Common-mode analysis: Common-Mode Rejection Ratio

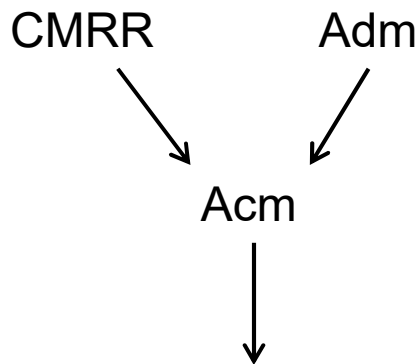
The Common-Mode Rejection Ratio (CMRR) is a measure of the ability of the amplifier to reject (cancel) the common-mode signals

$$CMRR = \frac{A_{dm}}{A_{cm}}$$

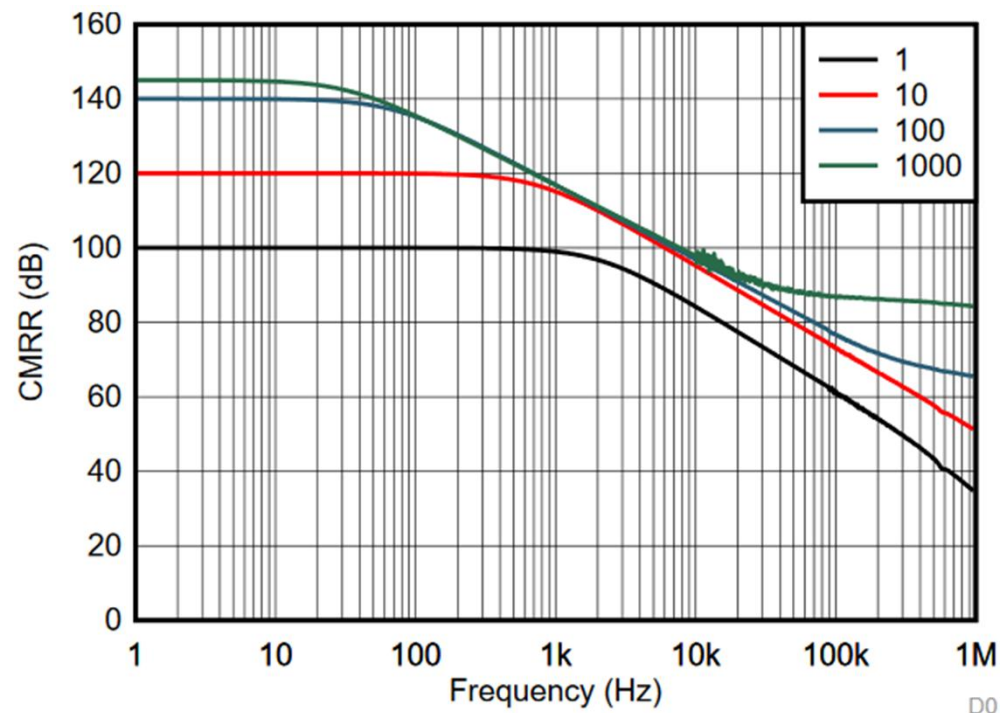
A_{dm} : differential-mode gain
 A_{cm} : common-mode gain

CMRR is frequency dependent

How to use the CMRR value?



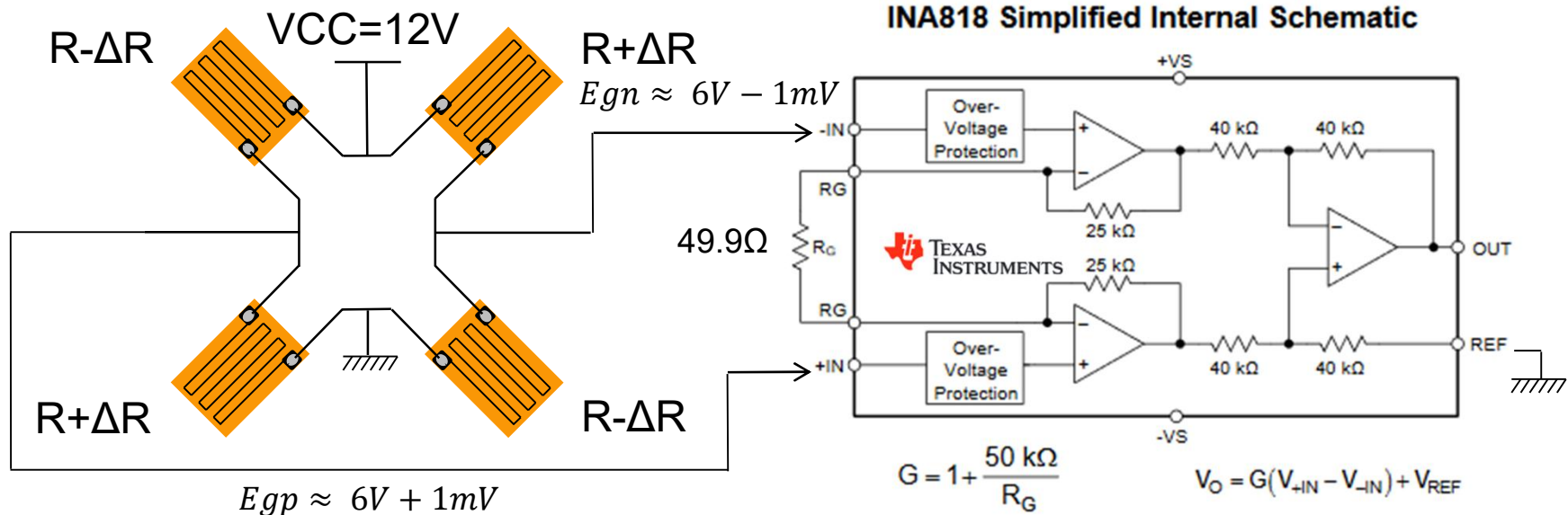
$$V_{out} = A_{dm} V_d + A_{cm} V_{cm}$$



Example: INA828

3. Common-mode analysis: Common-Mode Rejection Ratio

Common-mode error calculation example:



Gain = 1003 = A_{dm} $V_d = 2\text{mV}$ → Expected value: $V_{out} = 2.006\text{V}$

CMRR = 140dB (worst case @ $G=1000$) → $\text{CMRR} = 10^7$ → $A_{cm} \approx 10^{-4}$

Real value: $V_{out} = 1003 \times 2\text{mV} + 10^{-4} \times 6\text{V} = 2.0066\text{V}$ ($\approx 0.03\%$ error, $<0.5\text{LSB}$ @ 10 bits, $V_{ref} = 3.3\text{V}$)



V_{cm} range is limited, check datasheet

4. Applications of differential circuits in the digital world

Why differential signals for digital applications?

Because $i = C \frac{dV}{dt}$

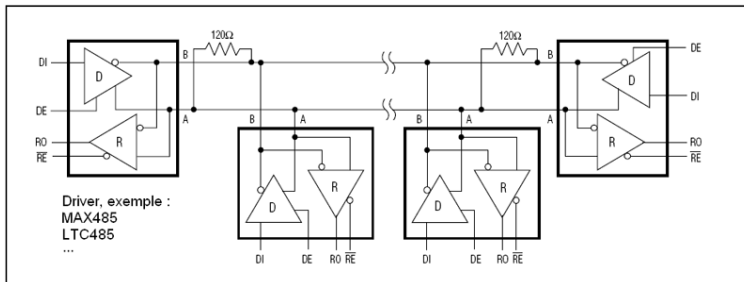
Digital signals → high dV
 + high throughput → low dt
 + Long distance → high capacitance

 = high current from supply

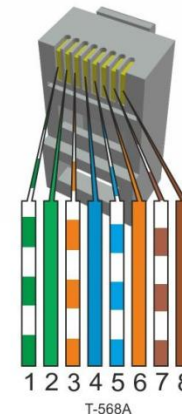
Reducing dV (lower supply voltage) helps reducing the supply current but also increases susceptibility to perturbators

Perturbators are common-mode signals, thus the interest in using differential signals

RS485 (computers, industry)
 CAN bus (automotive, industry)



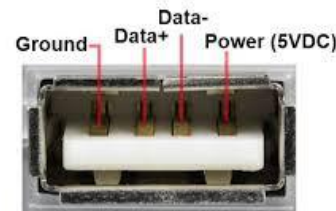
Ethernet



Pin	Description	10base-T	100Base-T	1000Base-T
1	Transmit Data+ or BiDirectional	TX+	TX+	BI_DA+
2	Transmit Data- or BiDirectional	TX-	TX-	BI_DA-
3	Receive Data+ or BiDirectional	RX+	RX+	BI_DB+
4	Not connected or BiDirectional	n/c	n/c	BI_DC+
5	Not connected or BiDirectional	n/c	n/c	BI_DC-
6	Receive Data- or BiDirectional	RX-	RX-	BI_DB-
7	Not connected or BiDirectional	n/c	n/c	BI_DD+
8	Not connected or BiDirectional	n/c	n/c	BI_DD-

PinoutsGuide.com

USB



LVDS

