



Instructions:

- Write down all your observations in your notebook.

Objectives:

- To design a basic Differential amplifier for given specifications
 - To understand the working and design of an Operational Differential Amplifier
 - Using Simulations to support analytical analysis.
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1. Differential Amplifier with Resistive Load

(a) Theory

Fig.[1] is a basic differential amplifier with a resistive load. M_1 and M_2 is input NMOS differential pair. R_2 and R_3 are resistive loads to this differential pair. M_3 realizes the tail current source. R_1 , M_3 and M_4 together realizes a current mirror. Take $V_{DD} = 10V$. A_1 , A_2 , and A_3 are ammeters used to measure respective branch currents. They don't play any role in differential amplifiers apart from measurement purposes.

Required specifications:

$$Gain > 12 \text{ dB}$$

$$V_{in,cm(min)} = 3.5 \text{ V}$$

$$5 \text{ V} < V_{out,cm} < 7 \text{ V}$$

Follow the below design steps

- i. Minimum input common mode voltage is defined as the minimum voltage at which all the MOSFETs are in desired operating regions. In our case, saturation region. Let current from M_3 be I_{tail} . Thus $V_{in,cm(min)}$ can be written as

$$V_{in,cm(min)} = V_{GS1} + V_{dsat3} \quad (1)$$

$$V_{in,cm(min)} = V_{TH1} + \sqrt{\frac{I_{tail}}{K_{n1}}} + \sqrt{\frac{2I_{tail}}{K_{n3}}} \quad (2)$$

Calculate the required I_{tail} to achieve the minimum input common mode specification. As a safety of margin, $V_{in,cm}$ will be greater than the calculated $V_{in,cm(min)}$ value.

- ii. Gain of differential amplifier, A_v can be written as

$$A_v = gm1 * R_2 \quad (3)$$

$$A_v = \sqrt{I_{tail}K_{n1}} * R_2 \quad (4)$$

Calculate R_2 required to meet gain specification.

- iii. Output common mode voltage ($V_{out,cm}$) can be written as

$$V_{out,cm} = V_{DD} - \frac{I_{tail}R_2}{2} \quad (5)$$

Evaluate the $V_{out,cm}$ value. If it does not meet the specifications then redesign certain parameters to meet the specifications. For example, if you have designed for a very high gain then you will probably get $V_{out,cm}$ lesser than the acceptable minimum value. This can risk M_1 and M_2 to be driven into the triode region. In this case, you can reduce the gain to increase $V_{out,cm}$. It is expected to get trade-offs between different specifications.

- iv. Let the current through M_4 be I_{ref} . Calculate the value of R_1 to achieve desired I_{ref} value.

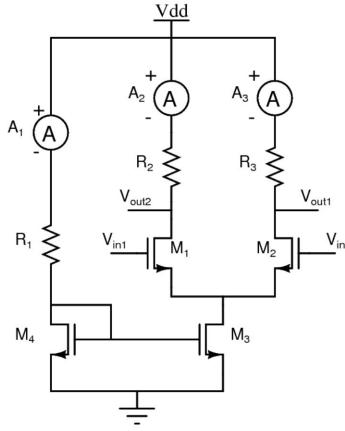


Figure 1: Basic Differential Amplifier

(b) **Simulation**

- i. Draw a schematic for Differential Amplifier with resistive load circuit [1] in LT-Spcie. Apply $V_{in1} = V_{in2} = 4.5V, V_{dd} = 10V$. Ensure all the MOSFETs are in the saturation region. Each MOSFET should satisfy $V_{DS} > V_{GS} - V_{TH}$ condition to be in the saturation region. If any of the MOSFETs fails the condition, redesign the circuit.
- ii. Tabulate all the node voltages, branch currents and operating region of MOSFETs. **[2 Marks]**
- iii. Characterize the large signal behavior of the designed differential amplifier. Fix $V_{in2} = 4.5V$ and sweep V_{in1} from 0 V to 10 V. Plot the $(V_{out1} - V_{out2})$ vs $(V_{in1} - V_{in2})$ transfer characteristics curve. How should ideally transfer characteristics curve look like? **[2 Marks]**
- iv. Apply 10 mV_{pp}, 1 KHz, sinusoidal differential input with 4.5 V common mode input voltage. Use 5 mV_{pp}, 1 KHz sinusoidal signal, with 4.5 V dc offset and apply it to V_{in1} and similarly apply 5 mV_{pp}, 1 KHz, 180 deg phase-shifted sinusoidal signal, with 4.5 V dc offset to V_{in2} .
- v. Plot the output V_{out1} and V_{out2} together. Observe the phase shift between two signals. Measure the differential gain. Compare with your hand calculations. **[2 Marks]**

(c) **Experiment**

Realize the Differential Amplifier on hardware for the designed values.

- i. Build circuit as shown in Fig.[1]. Apply $V_{in1} = V_{in2} = 4.5V$. Ensure all the MOSFETs are in the saturation region. Each MOSFET should satisfy $V_{DS} > V_{GS} - V_{TH}$ condition to be in the saturation region. If any MOSFET fails the condition then tweak the circuit accordingly.
- ii. Tabulate all the node voltages, branch current of M4 (I_{REF}) and operating region of MOSFET's. Ideally, V_{out1} and V_{out2} DC value should be equal since it is a balanced differential amplifier. If the measured value has a mismatch, then explain the reason. **[3 Marks]**
- iii. Perform transient analysis. Apply 40 mV_{pp}, 1 KHz, sinusoidal differential input with 4.5 V offset as common mode input voltage. Use two channels from AFG. Set channel 1 to 20 mV_{pp}, 1 KHz sinusoidal signal, with 4.5 V dc offset and apply it to V_{in1} and similarly set channel 2 to 20 mV_{pp}, 1 KHz, 180 deg phase-shifted sinusoidal signal, with 4.5 V dc offset. Remember to use the "Align Phase" feature on AFG to get the desired phase shift. Plot the output V_{out1} and V_{out2} together on oscilloscope. Observe the phase shift between these two signals. Plot V_{out1} and V_{in1} measure the differential gain. Compare your hand calculations with your own and justify your observations. **[3 Marks]**

Note: Grounds of AFG and DSO are internally shorted. Beware of this while taking measurements.

2. **Differential Amplifier with active load**

(a) **Theory**

Differential amplifier with current mirror load [2] is also called a Five Transistor OTA (5T-OTA). It is a very useful topology to build an Operational Amplifier (Op-amp). M_1, M_2 are NMOS Differential

pair. M_3, M_4 are the current mirror load. M_0 is a tail current source and M_5 mirrors the current to M_0 . Tail current (I_0) is decided by the value of R_D .

i. **Gain**

Gain of 5T-OTA is given as $-g_{m1}(r_{o2}||r_{o4})$. r_o is defined as $\frac{1}{\lambda * I_d}$. This gain is highly dependent on channel length modulation coefficient. You can get the values from the model file.

ii. **Output Common Mode Voltage ($V_{out_{dc}}$)**

Due to the negative feedback caused by the differential pair and current mirror, V_{out} , will be the same as V_{g3} in equilibrium condition (i.e. when $V_{in1} = V_{in2}$ and M_1, M_2 and M_3, M_4 are identical). V_{g3} can be expressed as $V_{DD} - V_{sg3}$. $V_{sg3} = \sqrt{\frac{I_0}{K_{n3}}} + V_{th3}$, where I_0 is the tail current source. Thus finally $V_{g3} = V_{out_{dc}} = V_{DD} - \sqrt{\frac{I_0}{K_{n3}}} - V_{th3}$

iii. **Input Common Mode Voltage**

Input common mode voltage (V_{incm}) should be such that all transistors should be in the saturation region. $V_{incm_{min}} = \sqrt{\frac{2I_0}{K_{n0}}} + \sqrt{\frac{I_0}{K_{n1}}} + V_{th1}$ and $V_{incm_{max}} = V_{out_{dc}} + V_{th1}$. Thus V_{incm} should be chosen in between these two limits.

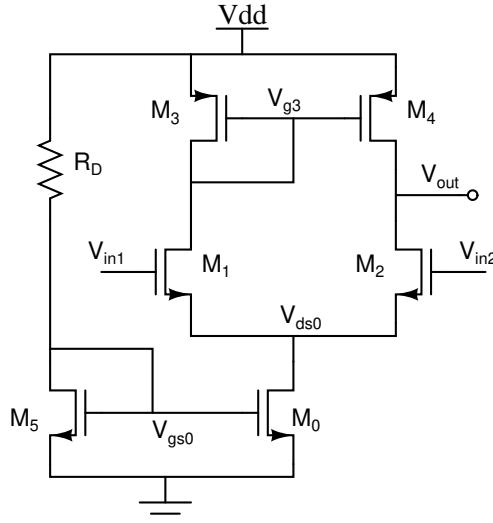


Figure 2: Five Transistor-OTA

(b) **Simulation**

- i. Design this amplifier for $V_{out_{dc}} = 6V$, $V_{dd} = 10V$.

(Hint: Refer to the equations in the theory part for $V_{out_{dc}}$. From this calculate the tail current source (I_0). Then calculate the limits of V_{incm} and choose V_{incm} accordingly (Don't choose very near to the extreme values). Since I_0 is known, calculate R_D). [3 Marks]

- ii. Apply $V_{in1} = V_{in2} = V_{incm}$ and run DC analysis. Tabulate all the node voltages, branch currents and operating region of MOSFETs. [2 Marks]

- iii. Apply V_{in1} as sine wave of $5 mV_{pp}$, 1 KHz and V_{in2} as sine wave of $5 mV_{pp}$, 1 KHz and 180 deg phase shifted (w.r.t to V_{in1}). Offset/DC value of these two waveforms should be V_{incm} . Plot the V_{out} and $V_{in1} - V_{in2}$ and report the gain. Compare with the hand calculated gain. [2 Marks]

- iv. Perform AC analysis. Determine the DC gain and compare it with the transient analysis gain. [2 Marks]

- v. Measure the common mode gain of the OTA. You have to short V_{in1} and V_{in2} and apply V_{in} as sine wave of $100 mV_{pp}$, 1 KHz and dc offset should be V_{incm} that you have calculated earlier. Plot the output waveform and mention the common mode gain of the OTA. [2 Marks]

(c) **Experiment**

- i. Build circuit as shown in Fig.[2]. Apply $V_{in1} = V_{in2} = V_{incm}$ (as calculated in simulation). Ensure all the MOSFETs are in the saturation region. Each MOSFET should satisfy $V_{DS} > V_{GS} - V_{TH}$ condition to be in the saturation region. If any one of the MOSFET fails the condition then tweak the circuit accordingly.
- ii. Tabulate all the node voltages, branch current of M_5 (I_{REF}) and operating region of MOSFET's. Ideally, V_{out} and V_{g3} DC value should be equal. [2 Marks]

- iii. Perform transient analysis. Apply 40 mV_{pp} , 1 KHz,sinusoidal differential input with V_{incm} offset as common mode input voltage. Use two channels from AFG. Set one channel to 20 mV_{pp} , 1 KHz sinusoidal signal, with V_{incm} as dc offset and apply it to V_{in1} and similarly set channel 2 to 20 mV_{pp} , 1 KHz, 180 deg phase shifted sinusoidal signal, with V_{incm} as dc offset. Remember to use "Align Phase" feature on AFG to get desired phase shift.
 Plot the output V_{out} and V_{in1} together on oscilloscope. Observe the phase shift between these two signals. Measure the differential gain. Compare with your hand calculations and justify your observations. [3 Marks]

Note: Grounds of AFG and DSO are internally shorted. Beware of this while taking measurements.

3. Some Application design around Five Transistor OTA

(a) Unity Gain Amplifier

- Built the unity gain buffer circuit as shown in the figure on the breadboard. [3].
- Perform transient analysis. Apply V_{in1} as 500 mV_{pp} , 1 KHz and V_{incm} (as calculated earlier) as DC offset. [2 Marks]

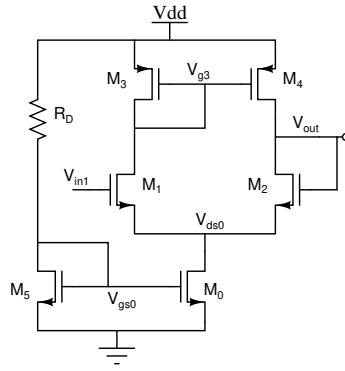


Figure 3: Unity amplifier

(b) Inverting Amplifier

- The Inverting amplifier circuit designed using five-transistor OTA is shown in Figure [4]. Apply V_{bias} as V_{incm} at the gate of M_1 that you have calculated when you were designing five transistor OTA. Take Resistor value $R_2 = 10R_1 = 10\text{M}\Omega$, $V_{dd} = 10\text{V}$. What is the need to have resistance values in $\text{M}\Omega$? [1 Marks]
- Apply $V_{bias} + vd$ at the one terminal of R_1 , vd is the sinusoidal voltage input with 50 mV_{pp} , 1KHz frequency. Plot V_{out} and $V_{bias} + vd$ and tabulate the theoretical and measured value of the amplifier gain and the phase shift. [2 Marks]

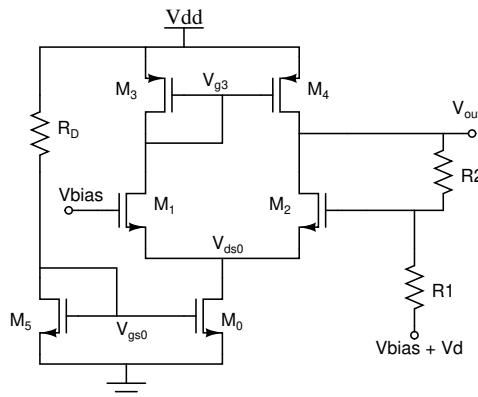


Figure 4: Inverting amplifier

(c) Differentiator Circuit (BONUS)

- i. By observing unity gain and inverting amplifier now you have to connect the five transistor OTA as Differentiator in negative feedback configuration. You can refer the previous experiment of Opamp based Differentiator. Apply Vin as triangular waveform of 100 mVpp, 1 KHz frequency and Plot input and output waveform. [2 Marks]