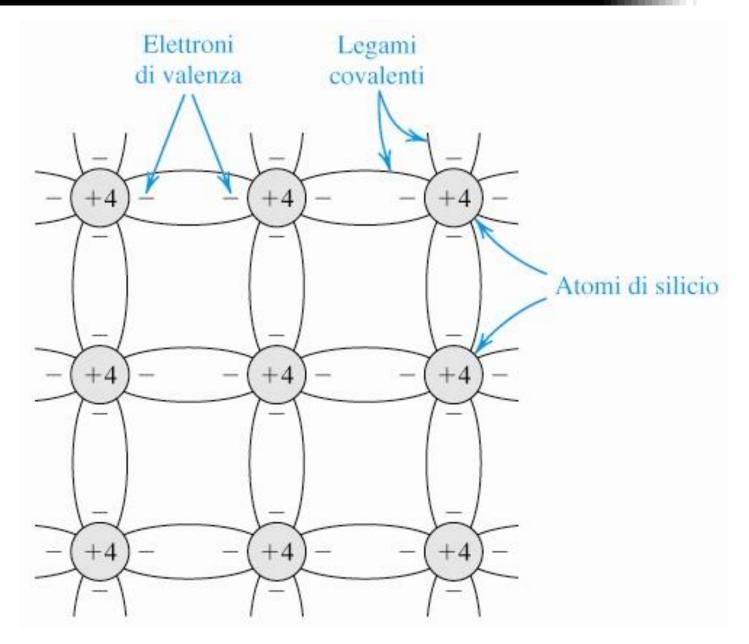
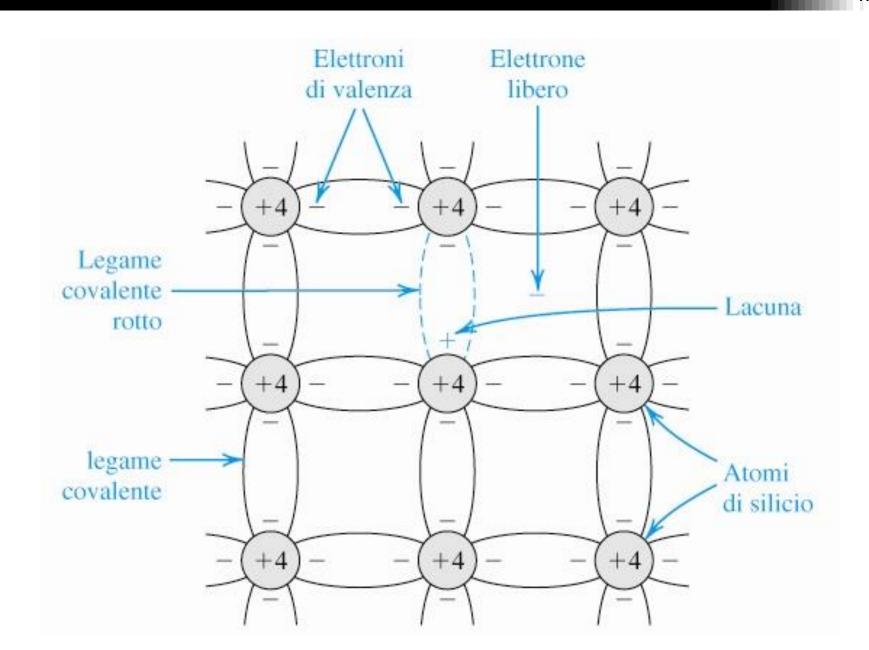
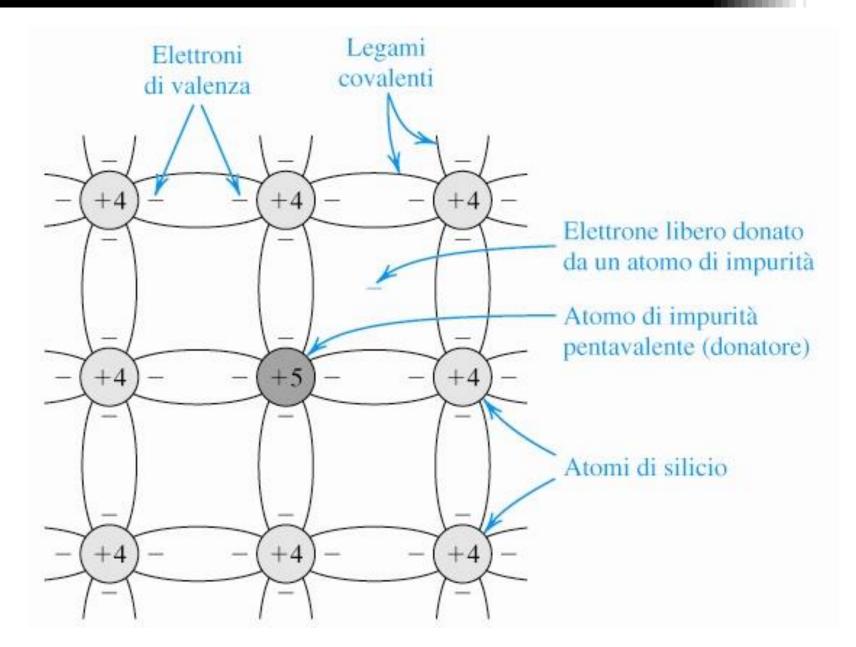
A.A. 2021-2022

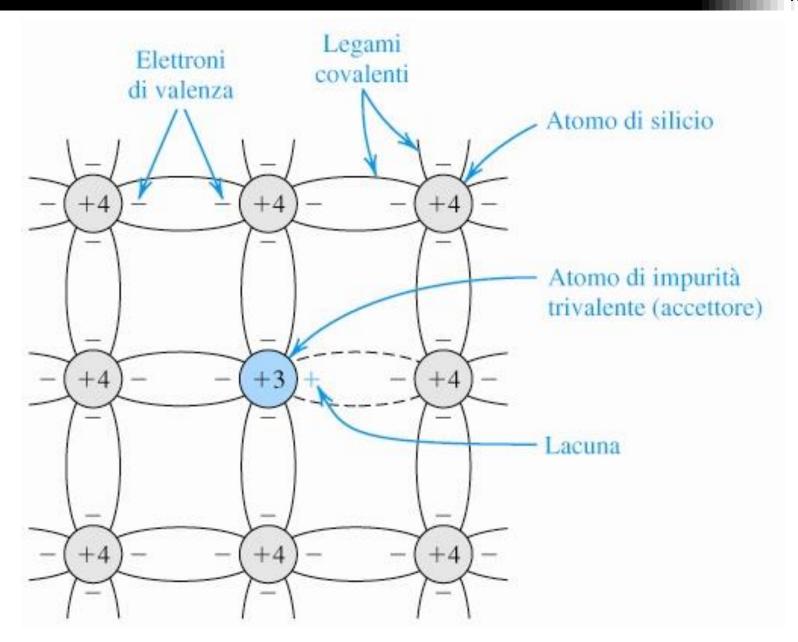
Elementi di Elettronica (INF) Prof. Paolo Crippa

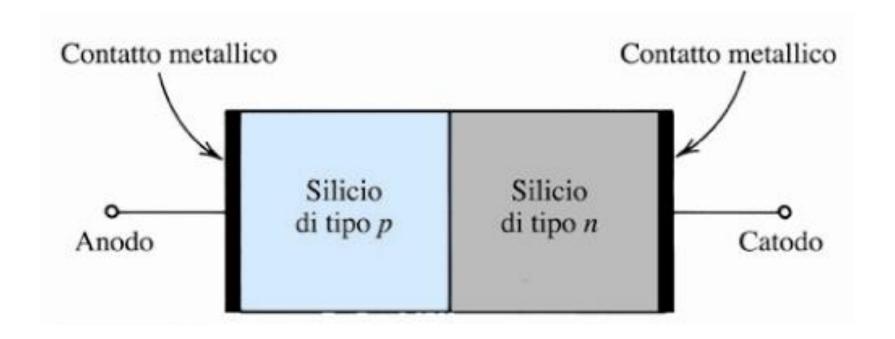
Diodo

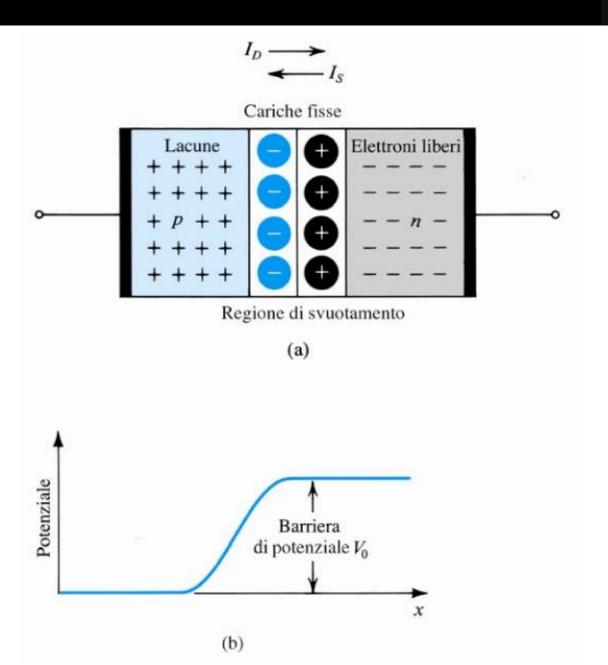


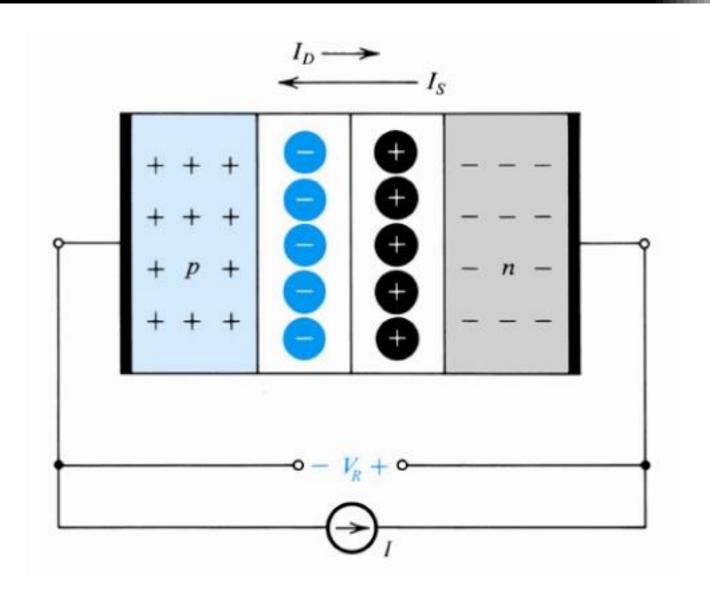


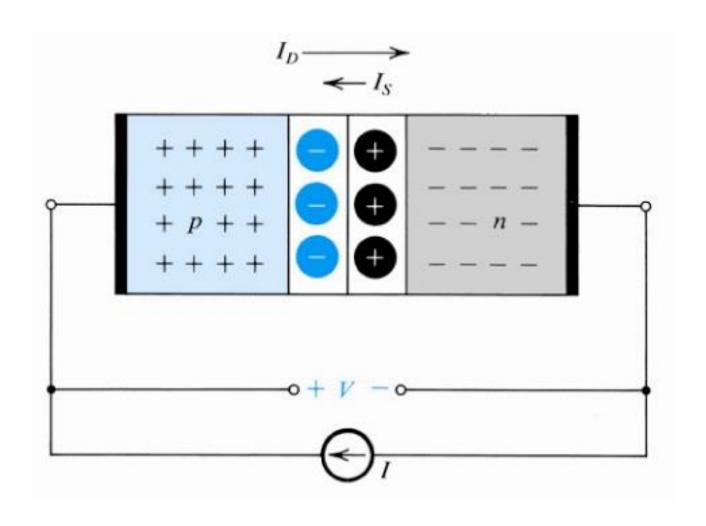




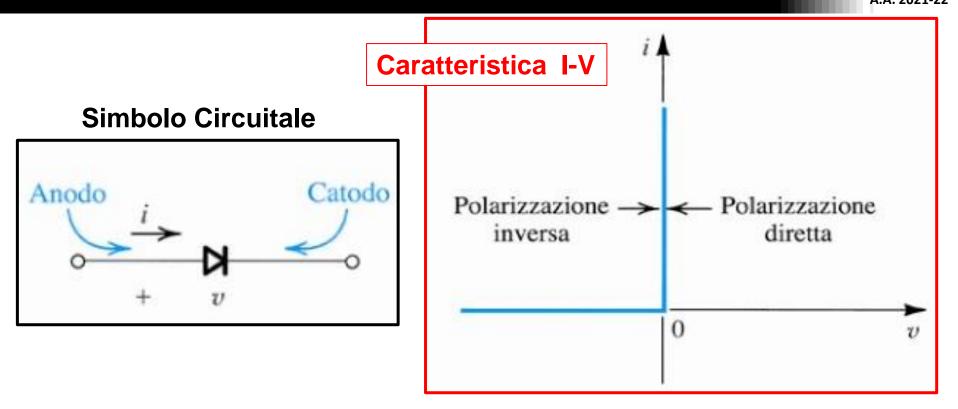




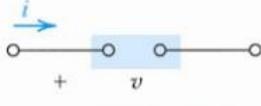




II Diodo Ideale

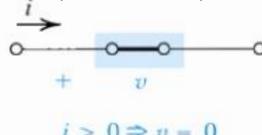


Circuito equivalente in polarizzazione inversa (circuito aperto)



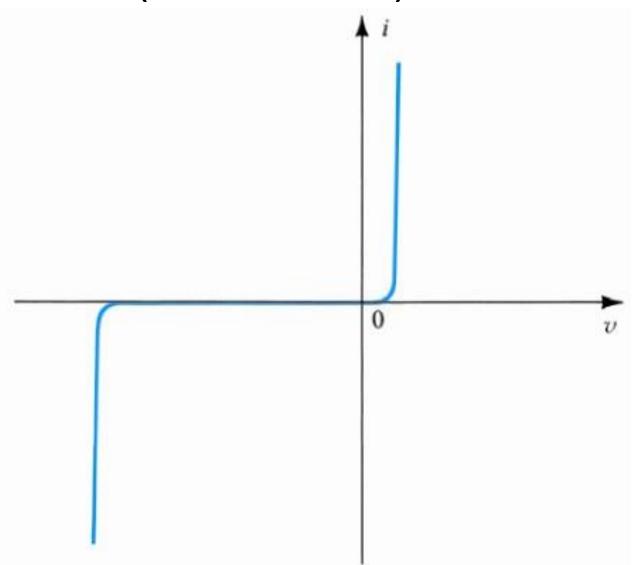
$$v < 0 \Rightarrow i = 0$$

Circuito equivalente in polarizzazione diretta (corto circuito)

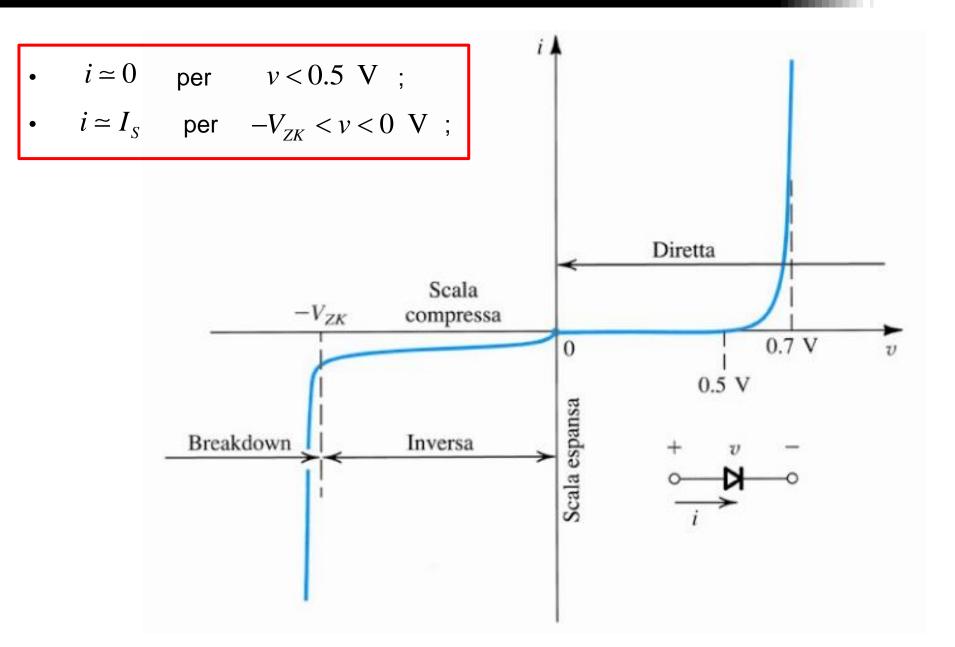


II Diodo Reale

Caratteristica I-V (corrente-tensione) del diodo al silicio reale



II Diodo Reale



Il Diodo: Equazione di Shockley

$$i = I_S \left(e^{v/nV_T} - 1 \right)$$

*I*_s corrente di saturazione

$$V_T = \frac{kT}{a}$$
 tensione termica

$$I_S = J_S A$$

 J_S = densità di corrente A = area del dispositivo

$$k = \text{costante di Boltzmann} = 1.38 \cdot 10^{-23} \text{ J/K}$$

 $T = \text{temperatura assoluta in K} = 273 + \text{temp. in } ^{\circ}\text{C}$

$$q$$
 = carica dell'elettrone = $1.602 \cdot 10^{-19}$ C

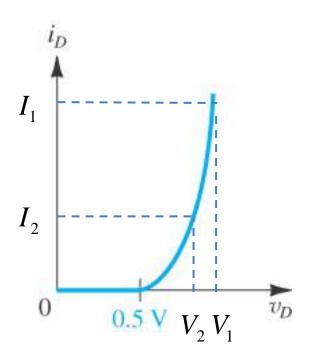
$$V_T(20^{\circ}\text{C}) = 25.2 \text{ mV} \approx 25 \text{ mV}$$

 $n = 1 \div 2$ parametro dipendente dalla tecnologia

$$v = V_T \ln \frac{i}{I_S}$$

II Diodo

Consideriamo due valori di corrente :



$$I_1 = I_S \left(e^{V_1/V_T} - 1 \right)$$

$$I_2 = I_S \left(e^{V_2/V_T} - 1 \right)$$

$$\frac{I_1}{I_2} = e^{(V_1 - V_2)/V_T}$$

Da cui:

$$V_1 - V_2 = V_T \ln \frac{I_1}{I_2}$$

$$V_1 - V_2 = 2.3 V_T \log \frac{I_1}{I_2}$$

se
$$\frac{I_1}{I_2} = 10$$
 \Rightarrow $V_1 - V_2 \cong 58 \text{ mV}$

se
$$\frac{I_1}{I_2} = 100$$
 \Rightarrow $V_1 - V_2 \cong 115 \text{ mV}$

II Diodo

Valori tipici di corrente in un circuito elettronico

$$v = V_T \ln \frac{i}{I_S} \quad , \qquad I_S = 10^{-15} \text{ A}$$

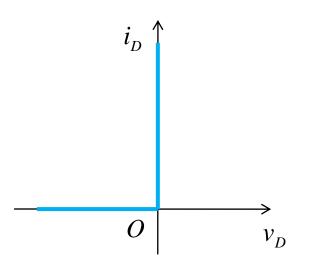
$$v = V_T \ln \frac{i_M}{I_S} = 25 \cdot 10^{-3} \ln \frac{10^{-3}}{10^{-15}} = 0.69 \text{ V}$$

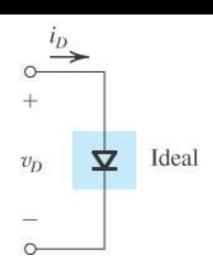
$$v = V_T \ln \frac{i_m}{I_S} = 25 \cdot 10^{-3} \ln \frac{10^{-6}}{10^{-15}} = 0.52 \text{ V}$$

$$v \sim 0.5 \div 0.7 \text{ V}$$

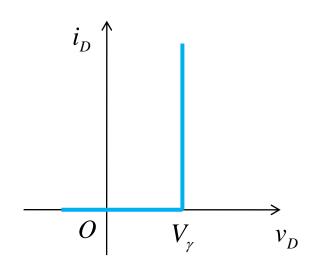
1	V
10 mA	0.748 V
1 mA	0.691 V
100 μΑ	0.633 V
10 μΑ	0.576 V
1 μΑ	0.518 V
100 nA	0.460 V
10 nA	0.403 V
1 nA	0.345 V

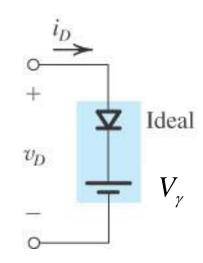
Il Diodo: Modelli Semplificati





$$\begin{cases} i_D = 0 & v_D \le 0 \\ v_D = 0 & i_D \ge 0 \end{cases}$$

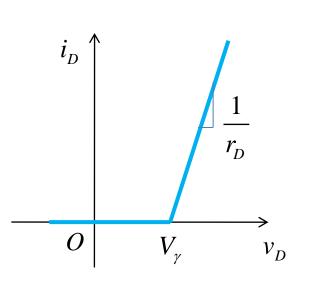


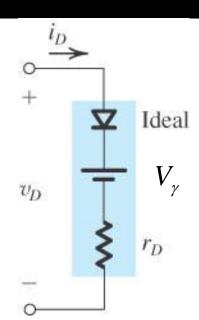


$$\begin{cases} i_D = 0 & v_D \le V_{\gamma} \\ v_D = V_{\gamma} & i_D \ge 0 \end{cases}$$

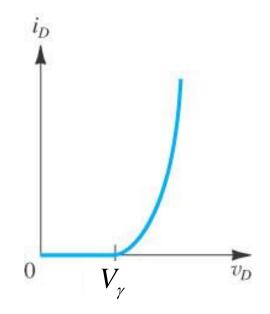
$$V_{_{\gamma}} \sim 0.5 \div 0.7$$

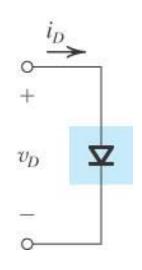
Il Diodo: Modelli Semplificati



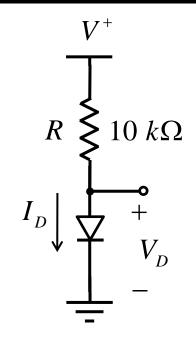


Ideal
$$V_{\gamma} = \begin{cases} i_D = 0 & v_D \leq V_{\gamma} \\ i_D = \frac{1}{r_D} (v_D - V_{\gamma}) & v_D \geq V_{\gamma} \end{cases}$$





$$i = I_S \left(e^{v/nV_T} - 1 \right)$$



Determinare I_D nei seguenti casi:

a)
$$V^+ = 10 V$$
;

b)
$$V^+ = 1 V$$
;

usando i diversi modelli del diodo.

Caso a) $V^{+} = 10 \text{ V}$

1)
$$V^+$$

$$R \nearrow V$$

$$V_D \text{ ideale}$$

$$V_D = 0 \text{ V}$$
, $I_D = \frac{V^+}{R} = \frac{10}{10 \cdot 10^3} = 1 \text{ mA}$

se
$$v \le 0$$
, $i = 0 \implies V_D = V^+ > 0$

2)
$$R \stackrel{V^+}{=}$$
 V ideale V V V

$$V_{\gamma} = 0.6 \text{ V}$$

$$I_D = \frac{V^+ - V_{\gamma}}{R} = \frac{10 - 0.6}{10 \cdot 10^3} = 0.94 \text{ mA}$$

 $V_{\gamma} = 0.6 \text{ V}$ $r_D = 200 \Omega$

3)
$$R = \begin{array}{c} V^{+} \\ \\ V \end{array}$$

$$V = \begin{array}{c} V^{+} \\ \\ V \end{array}$$

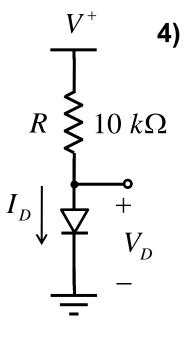
$$I_D = \frac{V^+ - V_{\gamma}}{R + r_D} = \frac{10 - 0.6}{(10 + 0.2) \cdot 10^3} = 0.92 \text{ mA}$$

se $v \le V_{\gamma}$, $i = 0 \implies v = V^+ > 0$

se
$$v \le V_{\gamma}$$
, $i = 0 \implies v = V^{+} > 0$ NO!

Elementi di Elettronica (INF) A.A. 2021-22

Circuiti con Diodi



Algoritmo di risoluzione dell'equazione esatta

$$I_D = I_S e^{V_D/V_T}$$

$$I_S = 10^{-15} \text{ A}$$

$$V^{+} = R I_{D} + V_{D} = R I_{D} + V_{T} \ln \frac{I_{D}}{I_{S}}$$

$$I_D = \frac{1}{R} \left(V^+ - V_T \ln \frac{I_D}{I_S} \right)$$

risolvendo iterativamente si ottiene il valore esatto:

$$I_D^{(k+1)} = 10^{-4} \left(10 - 25 \cdot 10^{-3} \ln \frac{I_D^k}{10^{-15}} \right)$$

$$I_D^{(0)} = 0.94 \text{ mA}$$

$$I_D^{(1)} = 0.931 \text{ mA}$$

$$I_D^{(2)} = 0.931 \text{ mA}$$

Soluzione:

$$I_D = 0.931 \text{ mA}$$

Caso b)
$$V^{+} = 1 V$$

1)
$$V_D = 0 \text{ V}$$
, $I_D = \frac{V^+}{R} = \frac{1}{10 \cdot 10^3} = 0.1 \text{ mA}$

2)
$$V_{\gamma} = 0.6 \text{ V}$$
 $I_D = \frac{V^+ - V_{\gamma}}{R} = \frac{1 - 0.6}{10 \cdot 10^3} = 0.04 \text{ mA}$

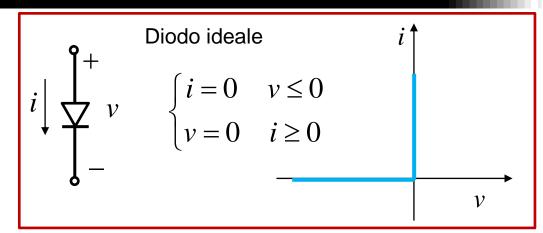
3)
$$V_{\gamma} = 0.6 \text{ V}$$
 $r_D = 200 \Omega$ $I_D = \frac{V^+ - V_{\gamma}}{R + r_D} = \frac{1 - 0.6}{(10 + 0.2) \cdot 10^3} = 0.039 \text{ mA}$

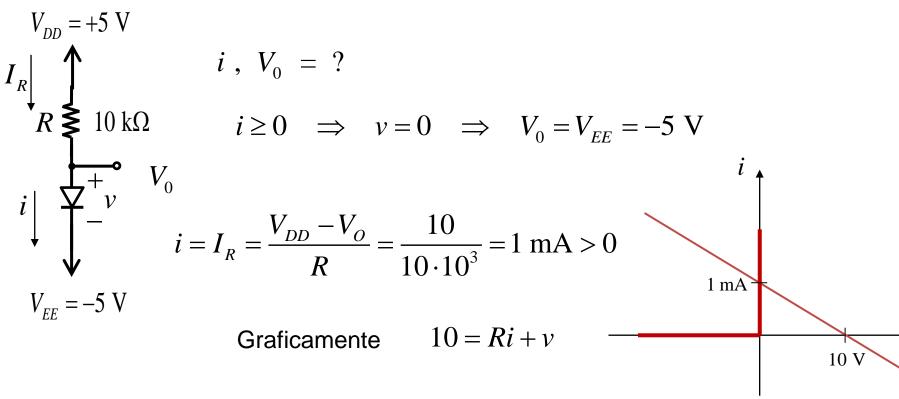
4)
$$I_D = 0.039 \text{ mA}$$

Il modello 1) dà risultati soddisfacenti solo quando la $\,V^{\scriptscriptstyle +}\,$ è elevata rispetto a $\,V_{\scriptscriptstyle
u}\,$

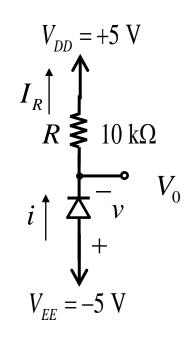
Elementi di Elettronica (INF) A.A. 2021-22

Es.1





Es.2



$$i, V_0 = ?$$

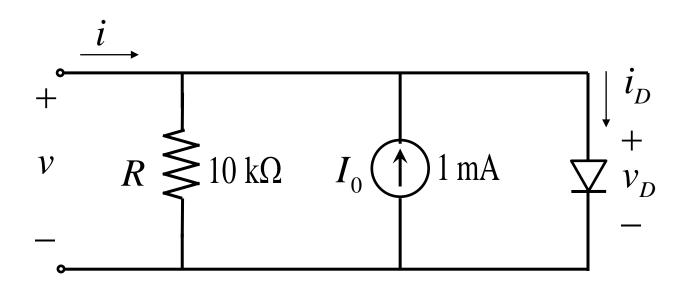
$$i \ge 0 \implies v = 0 \implies V_0 = V_{EE} = -5 \text{ V}$$

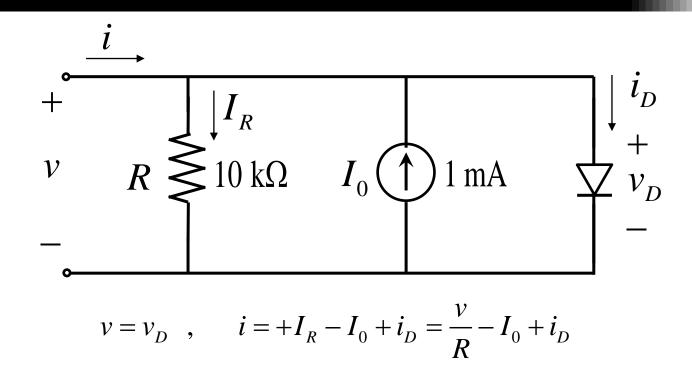
$$i = I_R = \frac{V_O - V_{DD}}{R} = \frac{-10}{10 \cdot 10^3} = -1 \text{ mA} < 0$$
 NO!

$$v < 0 \implies i = 0 \implies V_0 = 5 \text{ V} \implies v = -10 \text{ V} < 0 \text{ SI!}$$

Es.3

Determinare analiticamente e graficamente la caratteristica I-V del seguente bipolo:



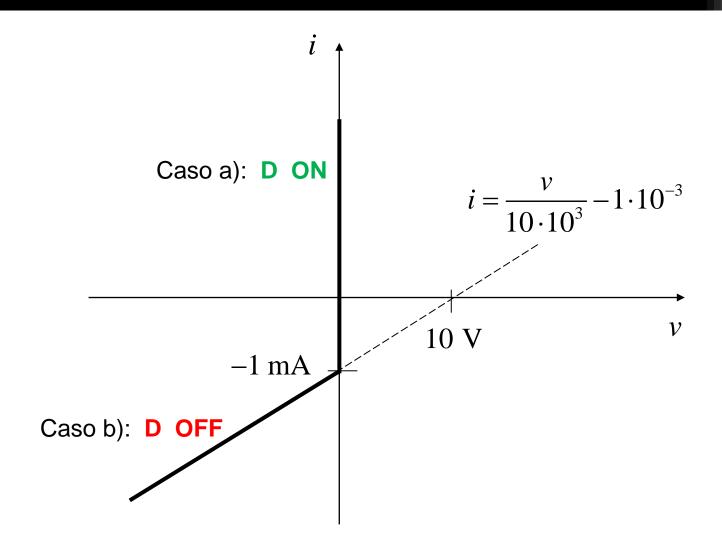


$$i_D \ge 0 \ , \ v_D = 0 \qquad \Longrightarrow \qquad v = 0 \ , \qquad i = -I_0 + i_D \qquad \qquad i_D = i + I_0 \ge 0 \qquad \qquad i \ge -I_0$$

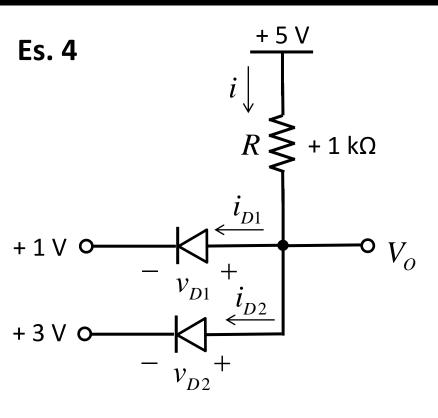
Caso b): D OFF

$$v_D \le 0$$
 , $i_D = 0$ \Rightarrow $i = \frac{v}{R} - I_0$

$$v \le 0$$



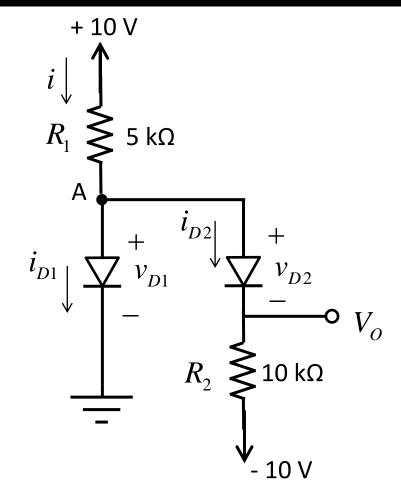
SI!



$$i_{D1} > 0$$
, $i_{D2} > 0$ \Rightarrow $v_{D1} = v_{D2} = 0$ \Rightarrow $V_0 = 1$, 3 V NO!

$$i_{D1} > 0$$
, $v_{D2} < 0$ \Rightarrow $v_{D1} = 0$, $i_{D2} = 0$ \Rightarrow $V_0 = 1 \text{ V}$
 \Rightarrow $v_{D2} = -2 \text{ V} < 0$, $i_{D1} = i = \frac{5-1}{1k} = 4 \text{ mA} > 0$



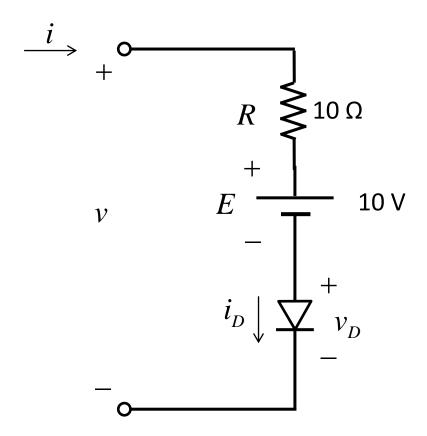


$$i_{D1} > 0 , i_{D2} > 0 \implies v_{D1} = v_{D2} = 0 \implies V_A = 0 \text{ V}, V_0 = 0 \text{ V}$$

$$\Rightarrow i_{D2} = \frac{10}{10k} = 1 \text{ mA} > 0$$
 $i = \frac{10}{5k} = 2 \text{ mA} \Rightarrow i_{D1} = 1 \text{ mA} > 0$

Es. 6

Determinare analiticamente e graficamente la caratteristica I-V del seguente bipolo:



Caso a): D ON

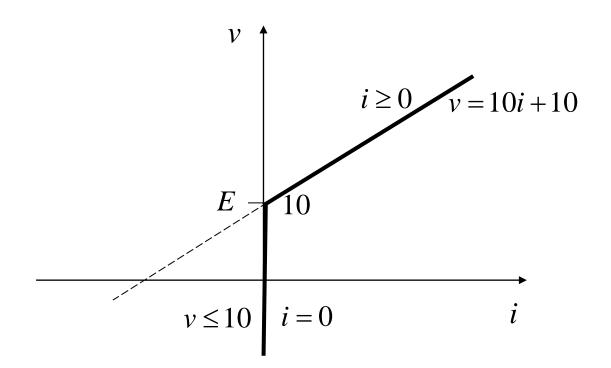
$$i_D \ge 0$$
, $v_D = 0$ \Rightarrow $v = Ri + E = 10i + 10$

$$i = i_D \implies i \ge 0$$

Caso b): D OFF

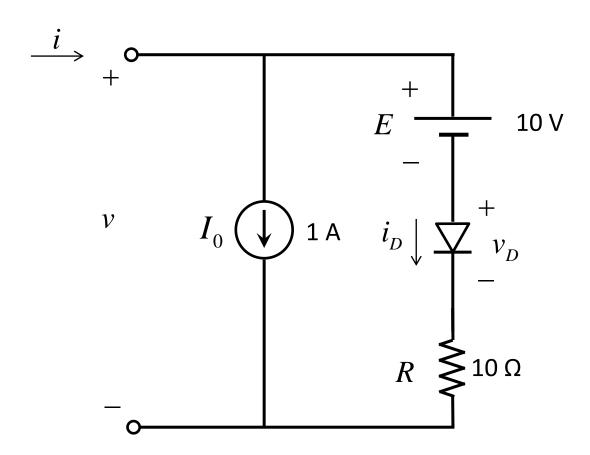
$$v_D \le 0$$
, $i_D = 0 \implies i = 0$

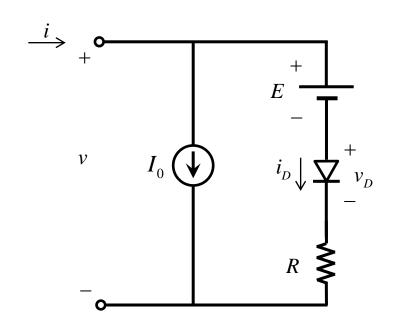
$$v_D = v - E \le 0 \implies v \le E = 10$$



Es. 7

Determinare analiticamente e graficamente la caratteristica I-V del seguente bipolo:



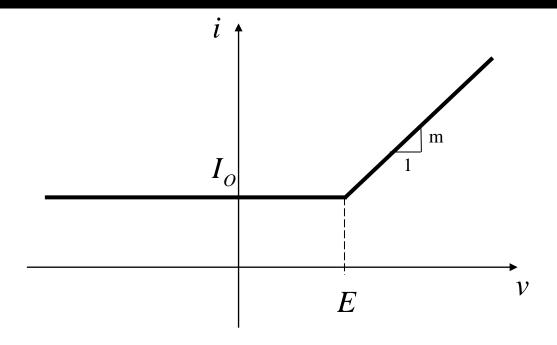


Caso a): D OFF

$$v_D \le 0$$
, $i_D = 0$ \Rightarrow $i = I_0$ $v = E + v_D$ \Rightarrow $v_D = v - E \le 0$ \Rightarrow $v \le E$

Caso b): D ON

$$i_D \ge 0$$
, $v_D = 0$ \Rightarrow $i = I_0 + \frac{v - E}{R}$ $m = \frac{1}{R}$ $i_D = \frac{v - E}{R} \ge 0$ \Rightarrow $v \ge E$



$$v \le E \qquad i = I_O$$

$$v \ge E \qquad i = m(v - E) + I_O$$

$$m = \frac{1}{R}$$

Logica a Diodi

Porta OR



$$V = 0 \text{ V}$$

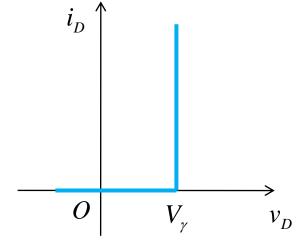
$$\Leftrightarrow$$

0 logico

$$V = 5 \text{ V}$$

$$\Leftrightarrow$$

1 logico



$$\begin{cases} i_D = 0 & v_D \le V_{\gamma} \\ v_D = V_{\gamma} & i_D \ge 0 \end{cases}$$

$$V_{_{\gamma}} = 0.7 \text{ V}$$

Logica a Diodi

$$V_A = V_B = 0 \text{ V}$$

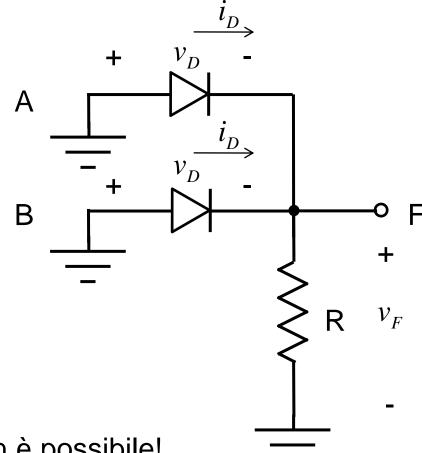
$$\text{se} \quad i_{\scriptscriptstyle D} \geq 0 \quad v_{\scriptscriptstyle D} = V_{\scriptscriptstyle \gamma}$$

$$\Rightarrow v_F = 2i_D R > 0 \text{ V}$$

$$ma v_F + v_D = 0 V$$

$$\Rightarrow v_D = V_v = -v_F < 0 \text{ V}$$

Allora: $v_D \leq V_{\gamma}$ $i_D = 0$



Non è possibile!

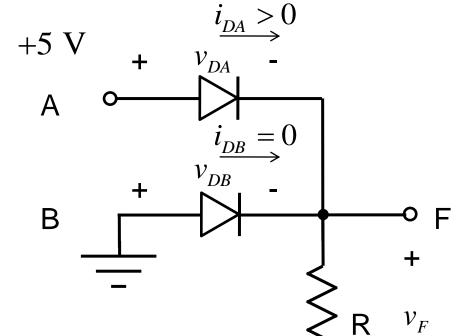
$$\Rightarrow$$
 $v_F = 0 \text{ V}$

Logica a Diodi

2)
$$V_A = 5 \text{ V}$$

 $V_B = 0 \text{ V}$

$$V_B = 0 \text{ V}$$



se

$$i_{DA} \ge 0$$
 $v_{DA} = V_{\gamma}$ $v_{DB} \le V_{\gamma}$ $i_{DB} = 0$

$$v_{DB} \le V_{\gamma} \quad i_{DB} = 0$$

$$\Rightarrow v_F = i_{DA}R > 0 \text{ V}$$

Allora

$$v_{DB} = -v_F \le V_{\gamma}$$

$$v_{DB} = -v_F \le V_{\gamma}$$
 \Rightarrow $v_F = 5 \text{ V} - V_{\gamma} = 4.3 \text{ V}$

3)
$$V_A = 0 \text{ V}$$

 $V_B = 5 \text{ V}$

$$\Rightarrow | v_F = 4.3 \text{ V}$$

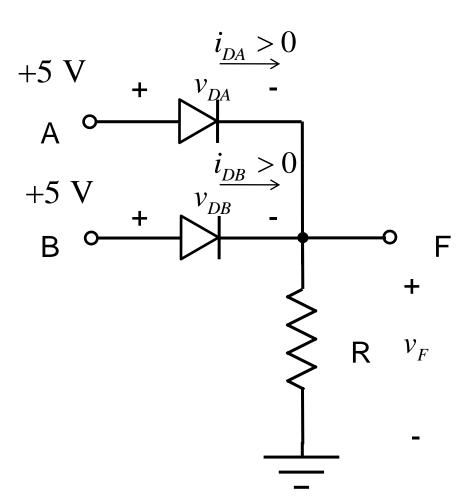
Analogo al caso 2)

$$V_A = V_B = 5 \, \mathrm{V}$$

$$i_{D\!A} \geq 0 \qquad v_{D\!A} = V_{\gamma}$$
 se
$$i_{D\!B} \geq 0 \qquad v_{D\!B} = V_{\gamma}$$

$$\Rightarrow |v_F = 5 \text{ V} - V_{\gamma} = 4.3 \text{ V} > 0$$

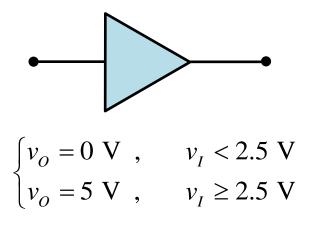
$$v_F = 2i_{DA}R > 0 \text{ V}$$

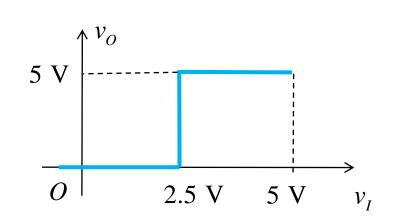


OK

$V_{_A}$	$V_{\scriptscriptstyle B}$	v_F	Caso ideale $V_{_{\gamma}}=0$
0 V	0 V	0 V	0 V
5 V	0 V	4.3 V	5 V
0 V	5 V	4.3 V	5 V
5 V	5 V	4.3 V	5 V

- Si ha una riduzione del livello in uscita a causa di $\,V_{\scriptscriptstyle \gamma}\,$
- Per ripristinare il livello di 5 V in uscita è necessario un circuito amplificatore





Porta OR



$$V = 0 \text{ V}$$

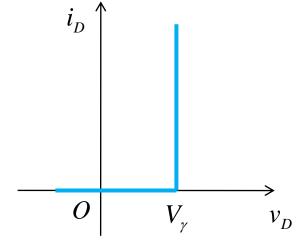
$$\Leftrightarrow$$

0 logico

$$V = 5 \text{ V}$$

$$\Leftrightarrow$$

1 logico



$$\begin{cases} i_D = 0 & v_D \le V_{\gamma} \\ v_D = V_{\gamma} & i_D \ge 0 \end{cases}$$

$$V_{_{\gamma}} = 0.7 \text{ V}$$

$$V_A = V_B = 0 \text{ V}$$

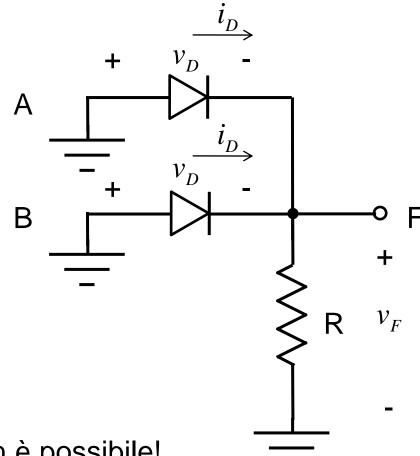
$$\text{se} \quad i_{\scriptscriptstyle D} \geq 0 \quad v_{\scriptscriptstyle D} = V_{\scriptscriptstyle \gamma}$$

$$\Rightarrow v_F = 2i_D R > 0 \text{ V}$$

$$ma v_F + v_D = 0 V$$

$$\Rightarrow v_D = V_{\gamma} = -v_F < 0 \text{ V}$$

Allora: $v_D \leq V_{\gamma}$ $i_D = 0$

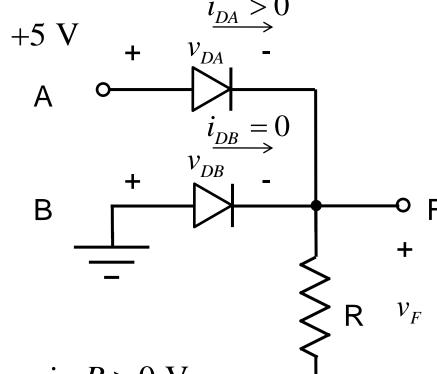


Non è possibile!

$$\Rightarrow$$
 $v_F = 0 \text{ V}$

2)
$$V_A = 5 \text{ V}$$

 $V_B = 0 \text{ V}$



$$i_{D\!A} \geq 0 \qquad v_{D\!A} = V_{\gamma}$$
 se
$$v_{D\!B} \leq V_{\gamma} \quad i_{D\!B} = 0$$

$$v_{DB} \leq V_{v}$$
 $i_{DB} = 0$

$$\Rightarrow v_F = i_{DA}R > 0 \text{ V}$$

$$v_{DB} = -v_F \le V_{\gamma}$$

$$v_{DB} = -v_F \le V_{\gamma}$$
 \Rightarrow $v_F = 5 \text{ V} - V_{\gamma} = 4.3 \text{ V}$

3)
$$V_A = 0 \text{ V}$$

 $V_B = 5 \text{ V}$

$$\Rightarrow | v_F = 4.3 \text{ V}$$

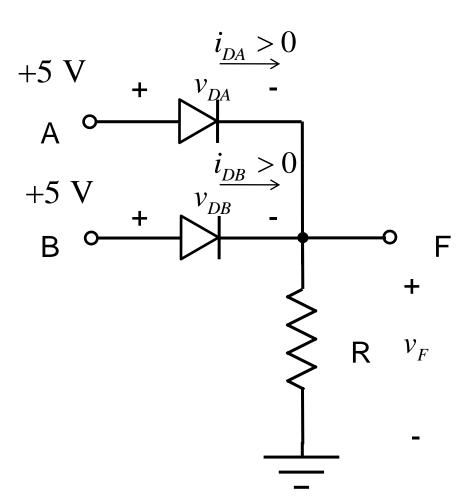
Analogo al caso 2)

$$V_A = V_B = 5 \, \mathrm{V}$$

$$i_{D\!A} \geq 0 \qquad v_{D\!A} = V_{\gamma}$$
 se
$$i_{D\!B} \geq 0 \qquad v_{D\!B} = V_{\gamma}$$

$$\Rightarrow |v_F = 5 \text{ V} - V_{\gamma} = 4.3 \text{ V} > 0$$

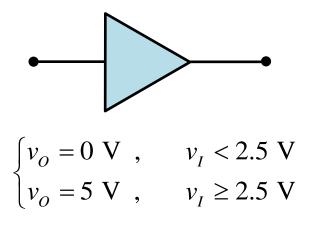
$$v_F = 2i_{DA}R > 0 \text{ V}$$

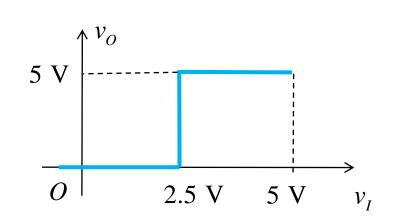


OK

$V_{_A}$	$V_{\scriptscriptstyle B}$	v_F	Caso ideale $V_{_{\gamma}}=0$
0 V	0 V	0 V	0 V
5 V	0 V	4.3 V	5 V
0 V	5 V	4.3 V	5 V
5 V	5 V	4.3 V	5 V

- Si ha una riduzione del livello in uscita a causa di $\,V_{\scriptscriptstyle \gamma}\,$
- Per ripristinare il livello di 5 V in uscita è necessario un circuito amplificatore

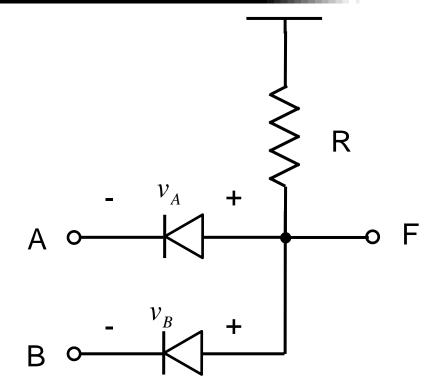


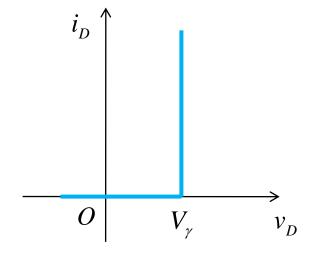


Porta AND

$$V = 0 \text{ V} \Leftrightarrow 0 \text{ logico}$$

$$V = 5 \text{ V} \Leftrightarrow 1 \text{ logico}$$





$$\begin{cases} i_D = 0 & v_D \le V_{\gamma} \\ v_D = V_{\gamma} & i_D \ge 0 \end{cases}$$

$$V_{\nu} = 0.7 \text{ V}$$

1)
$$V_{A} = V_{B} = 0 \text{ V}$$

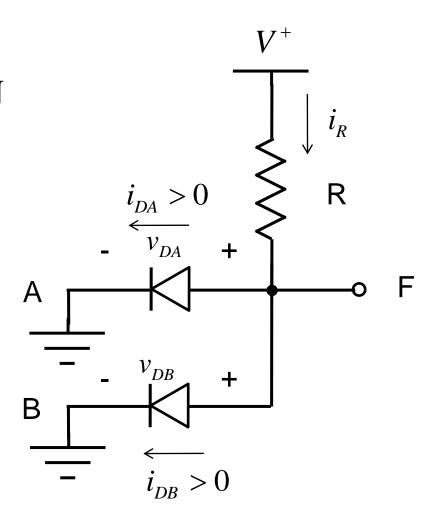
$$i_{DA}$$
, $i_{DB} \ge 0 \implies D_A$, $D_B \rightarrow ON$

$$i_{\mathrm{DA}}=i_{\mathrm{DB}}=i_{\mathrm{D}}$$

$$v_{DA} = v_{DB} = V_{\gamma} = v_F \qquad v_F = 0.7 \text{ V}$$

$$v_F = 0.7 \text{ V}$$

$$i_R = \frac{V^+ - v_F}{R} = \frac{V^+ - V_{\gamma}}{R} = 2i_D > 0$$



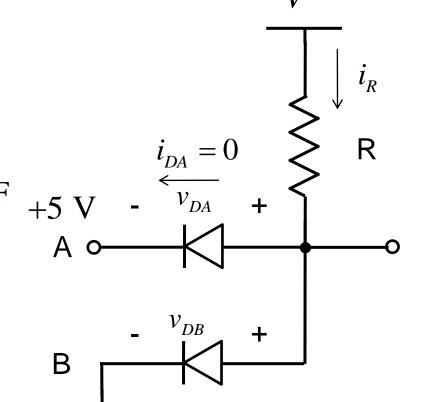
$$V_A = 5 \text{ V}$$
$$V_B = 0 \text{ V}$$

$$i_{DB} \ge 0$$
 $v_{DB} = V_{\gamma}$ $D_B \to ON$

$$v_{DA} = v_F - V^+ < 0 \implies D_A \rightarrow \text{OFF}$$

$$v_F = V_{\nu}$$

$$i_{R} = \frac{V^{+} - V_{\gamma}}{R} = i_{DB} > 0$$



$$v_F = 0.7 \text{ V}$$

$$V_A = 0$$

$$v_F = V_{\gamma} = 0.7 \text{ V}$$

$$V_A = 0 V$$

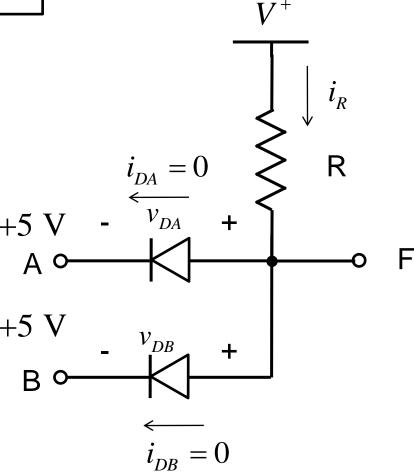
$$V_B = 5 V$$

$$V_A = V_B = 5 \, \mathrm{V}$$

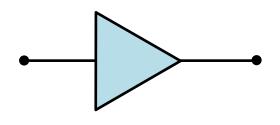
$$v_{DA}, \ v_{DB} < V_{\gamma}$$
 $\Rightarrow D_A, D_B \rightarrow \text{OFF}$

$$i_{DA} = i_{DB} = 0 = i_R$$

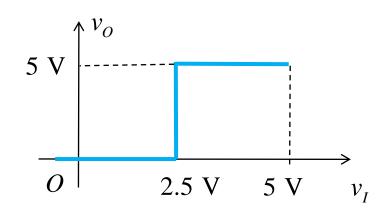
$$v_F = V^+ = 5 \text{ V}$$

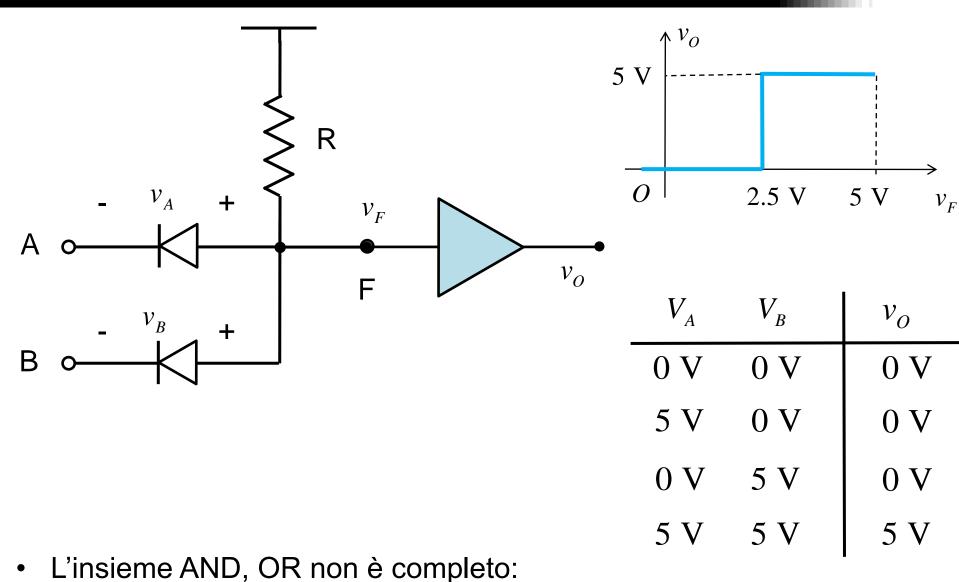


$V_{\scriptscriptstyle A}$	$V_{\scriptscriptstyle B}$	\mathcal{V}_F	Caso ideale $V_{_{\gamma}}=0$
0 V	0 V	0.7 V	0 V
5 V	0 V	0.7 V	0 V
0 V	5 V	0.7 V	0 V
5 V	5 V	5 V	5 V



$$\begin{cases} v_O = 0 \text{ V}, & v_I < 2.5 \text{ V} \\ v_O = 5 \text{ V}, & v_I \ge 2.5 \text{ V} \end{cases}$$





- L'Insieme AND, OR non e completo.
- Si deve realizzare il NOT (amplificatore invertente)

Il Diodo: Dipendenza della Temperatura

$$i_D = I_D$$

$$v_D = V_T \ln \left(\frac{I_D}{I_S}\right) = \frac{kT}{q} \ln \left(\frac{I_D}{I_S}\right)$$

$$\frac{dv_{D}}{dT} = \frac{v_{D}}{T} - V_{T} \frac{I_{S}}{I_{D}} \frac{I_{D}}{I_{S}^{2}} \frac{dI_{S}}{dT} = \frac{v_{D}}{T} - V_{T} \left(\frac{1}{I_{S}} \frac{dI_{S}}{dT} \right)$$

$$= \frac{v_{D}}{T} - \frac{3V_{T}}{T} - \frac{V_{G}}{T} = -\frac{V_{G} - v_{D}}{T} - \frac{3V_{T}}{T}$$

$$V_G = E_G/q$$

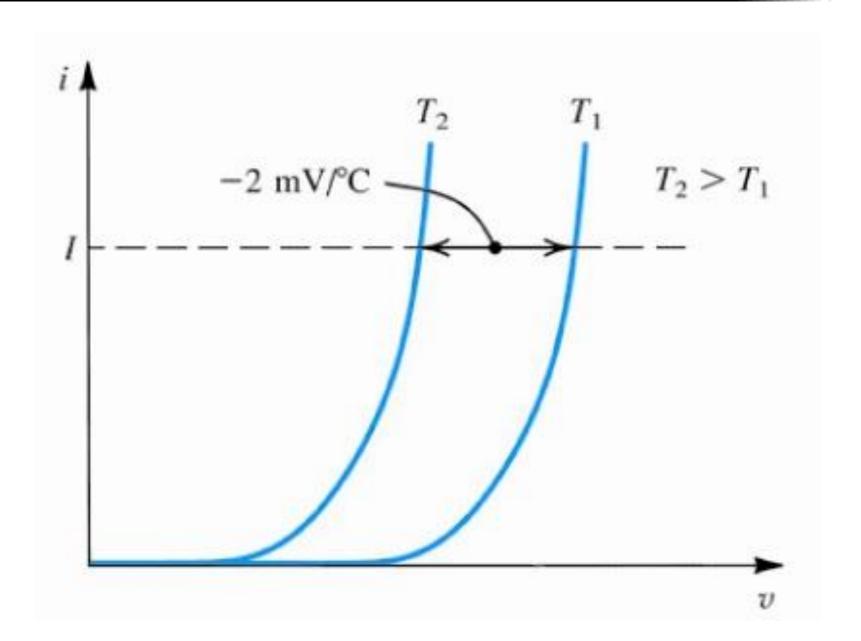
$$v_D \approx V_D = 0.6 \text{ V}$$

$$E_G = 1.12 \text{ eV}$$

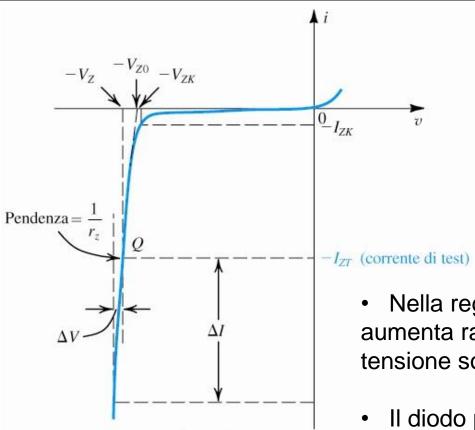
per
$$T = 300 \text{ K}$$

$$\frac{dv_D}{dT} = -2.3 \text{ mV/}^{\circ}\text{C}$$

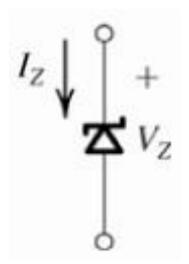
Il Diodo: Dipendenza della Temperatura



II Diodo Zener



 $\Delta V = \Delta I r_*$



- Nella regione di breakdown la corrente inversa aumenta rapidamente e i corrispondenti incrementi di tensione sono molto piccoli;
- Il diodo può lavorare nella regione di breakdown senza distruggersi, purché la potenza dissipata nel diodo venga opportunamente limitata da un circuito esterno ad un livello di guardia (il suo valore viene specificato nel data-sheet);
- I diodi che vengono costruiti appositamente per lavorare in questa regione vengono detti "diodi zener".

Il Diodo Zener come Regolatore di Tensione

$$V_{ZK} = 6.8 \text{ V}$$

$$I = \frac{10 - 6.8}{0.5 \cdot 10^3} = 6.4 \text{ mA}$$

• Inseriamo un carico R_L

$$V = +10 \text{ V}$$

$$I_{R} \downarrow \geqslant R = 0.5 \text{ k}\Omega$$

$$I' \downarrow \qquad \qquad \downarrow I_{L}$$

$$R_{L} = 2 \text{ k}\Omega$$

$$I_L = \frac{6.8}{2.10^3} = 3.4 \text{ mA}$$
 $V_Z \cong V_{ZK} = 6.8 \text{ V}$

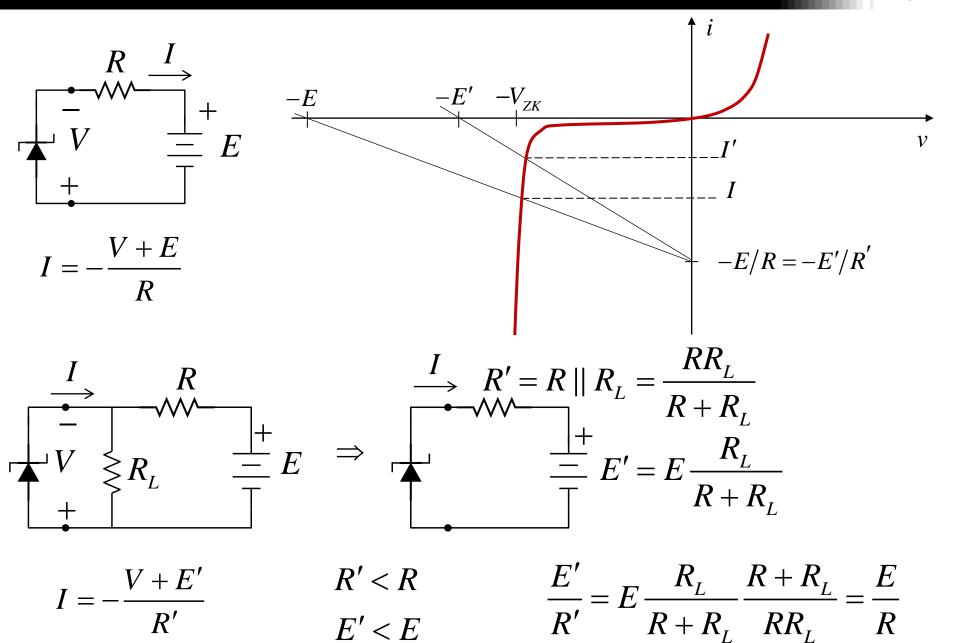
La corrente sul diodo cambia

$$I' = I_R - I_L = 6.4 - 3.4 = 3 \text{ mA}$$

I' < I

Il Diodo Zener come Regolatore di Tensione

Elementi di Elettronica (INF) A.A. 2021-22



Convenzioni sulle Notazioni

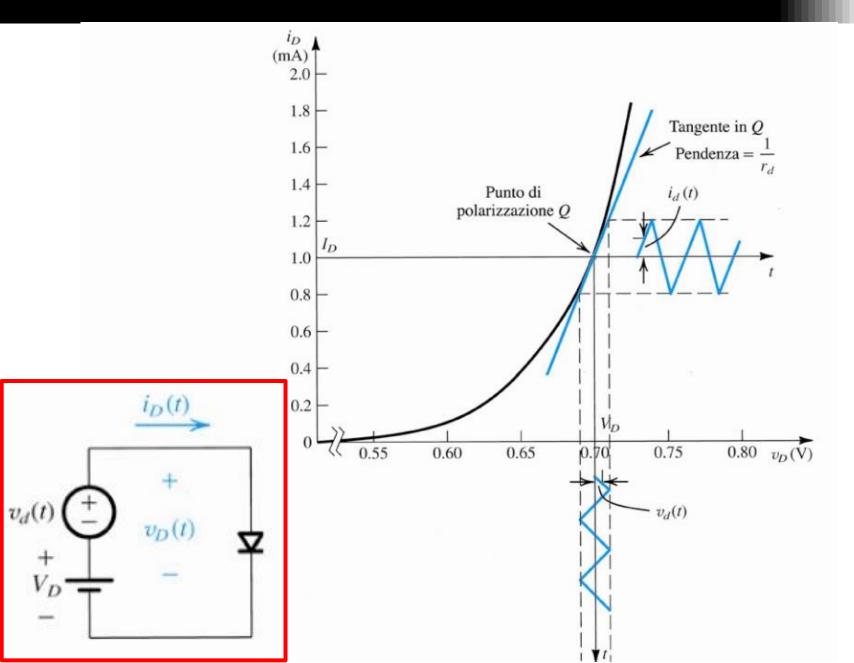
$$i_D = I_D + i_d$$

$$v_{AB} = V_{AB} + v_{ab}$$

$$i_{D}(t) = I_{D,0} + i_{d}(t)$$

$$v_{AB}(t) = V_{AB,0} + v_{ab}(t)$$

Il Diodo: Circuito alle Variazioni



 $V_T = \frac{kT}{}$

Il Diodo: Circuito alle Variazioni

Regione di polarizzazione diretta:

$$I_D = I_S e^{V_D/nV_T}$$

$$v_D(t) = V_D + v_d(t) \qquad i_D(t) = I_D + i_d(t)$$

$$i_D(t) = I_D + i_d(t)$$

$$i_D(t) = I_S e^{v_D(t)/nV_T}$$

$$i_{D}(t) = I_{S}e^{(V_{D}+v_{d}(t))/nV_{T}}$$

$$i_D(t) = I_S e^{V_D/nV_T} e^{v_d(t)/nV_T}$$

$$i_D(t) = I_D e^{v_d(t)/nV_T}$$

Approssimazione di piccolo segnale

$$\frac{v_d(t)}{nV_T} \ll 1 \qquad \Longrightarrow \qquad i_D(t) \simeq I_D\left(1 + \frac{v_d(t)}{nV_T}\right)$$

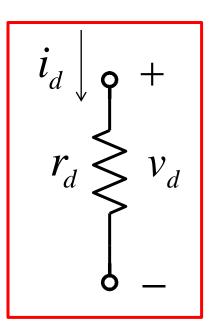
Il Diodo: Circuito alle Variazioni

$$i_D(t) = I_D + \frac{I_D}{nV_T} v_d(t)$$

$$i_{D}(t) = I_{D} + i_{d}(t)$$

$$i_d\left(t\right) = \frac{I_D}{nV_T} v_d\left(t\right)$$

MODELLO DEL DIODO PER PICCOLO SEGNALE



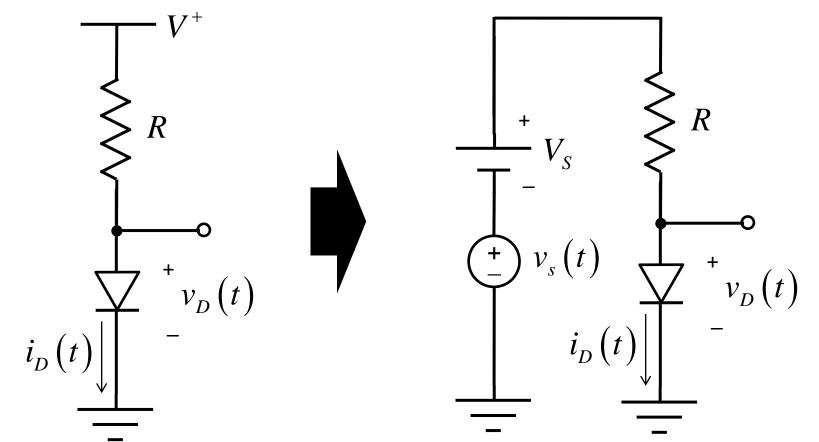
Resistenza del diodo per piccolo segnale o resistenza differenziale

$$r_d = \frac{nV_T}{I_D}$$

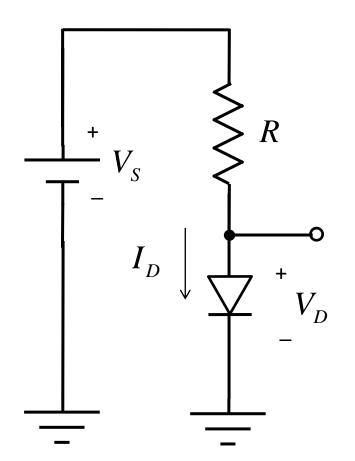
$$r_d = 1 / \left[\frac{\partial i_D}{\partial v_D} \right]_{i_D = I_D}$$

Il Diodo: Analisi Circuitale Completa

- Si consideri il circuito riportato qui sotto nel caso in cui sia $R = 10 \text{ k}\Omega$. L'alimentazione V⁺ fornisce una tensione continua di 10 V sulla quale è sovrapposta una sinusoide a 60 Hz con ampiezza di picco di 1 V (tale componente di segnale della tensione di alimentazione è un'imperfezione del progetto dell'alimentatore ed è conosciuta come *ripple* (ondeggiamento) dell'alimentazione).
- Calcolare sia la tensione in continua del diodo che il segnale sinusoidale che si presenta ai suoi capi. Si ipotizzi che il diodo abbia una caduta di tensione di $0.7 \, \text{V}$ in corrispondenza di una corrente di $1 \, \text{mA}$ e che sia n = 2.



Il Diodo: Analisi del Circuito in DC



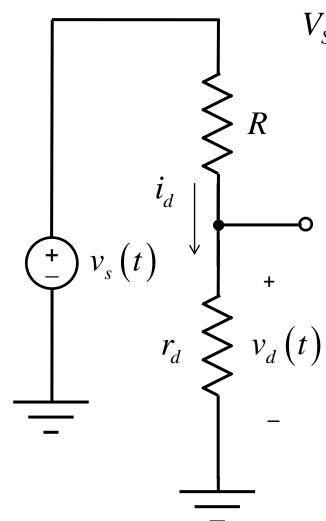
$$v_s(t) = 0$$
 $R = 10 \text{ k}\Omega$
 $V_s = V^+ = 10 \text{ V}$

$$I_D = \frac{V^+ - V_{\gamma}}{R} = \frac{10 - 0.7}{10 \cdot 10^{-3}} = 0.93 \text{ mA}$$

 $V_D = V_{\gamma} = 0.7 \text{ V}$

soluzione in DC (punto di riposo)

Il Diodo: Analisi del Circuito alle Variazioni



$$v_{s}(t) = \hat{V}_{s} \sin(2\pi f_{c} t) \qquad f_{c} = 60 \text{ Hz}$$

$$r_{d} = \frac{nV_{T}}{I_{D}} = \frac{2 \cdot 25 \cdot 10^{-3}}{0.93 \cdot 10^{-3}} = 53.8 \Omega$$

$$v_{d}(t) = \frac{r_{d}}{r_{d} + R} v_{s}(t) \qquad i_{d}(t) = \frac{v_{s}(t)}{r_{d} + R}$$

$$v_{d}(t) = \hat{v}_{d} \cdot \sin(2\pi f_{c} t)$$

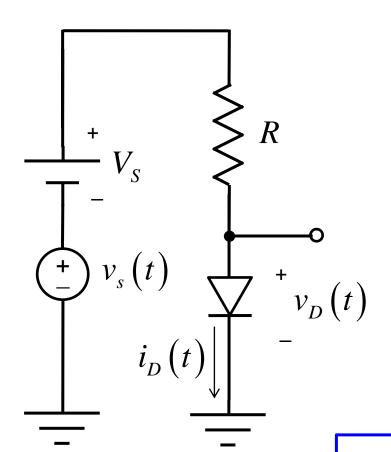
$$\hat{v}_{d}(picco) = \hat{V}_{s} \frac{r_{d}}{r_{d} + R} = 5.35 \text{ mV}$$

$$i_{d}(t) = \hat{i}_{d} \cdot \sin(2\pi f_{c} t)$$

$$\hat{i}_{d}(picco) = \frac{\hat{V}_{s}}{r_{d} + R} = 99.46 \mu\text{A}$$

$$V_S = 0 \text{ V}$$
 $I_{DQ} = I_D = 0.93 \text{ mA}$ $R = 10 \text{ k}\Omega$
$$v_s(t) = \hat{V_s} \sin(2\pi f_c t)$$
 $f_c = 60 \text{ Hz}$

Il Diodo: Analisi Completa



soluzione completa

$$v_D(t) = 0.7 + 5.35 \cdot 10^{-3} \sin(2\pi 60t)$$

$$i_D(t) = 0.93 \cdot 10^{-3} + 99.46 \cdot 10^{-6} \sin(2\pi 60t)$$