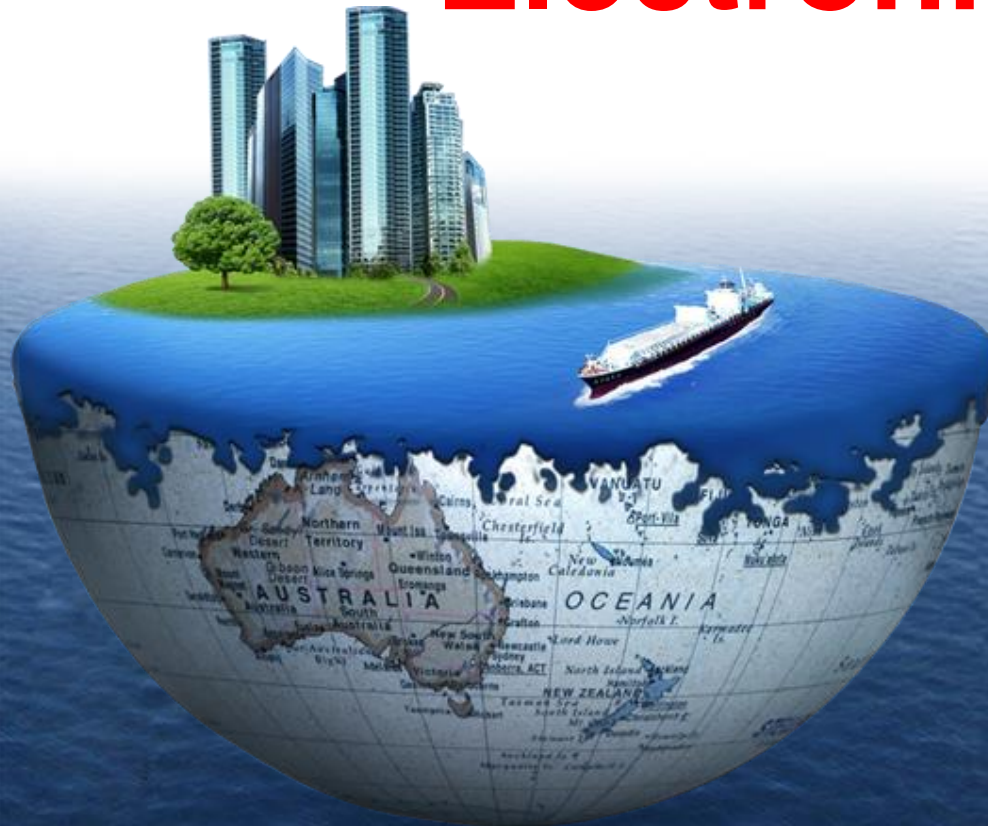




Electronic Engineering COM 121



**Assist. Prof.
Basma M. Yousef**

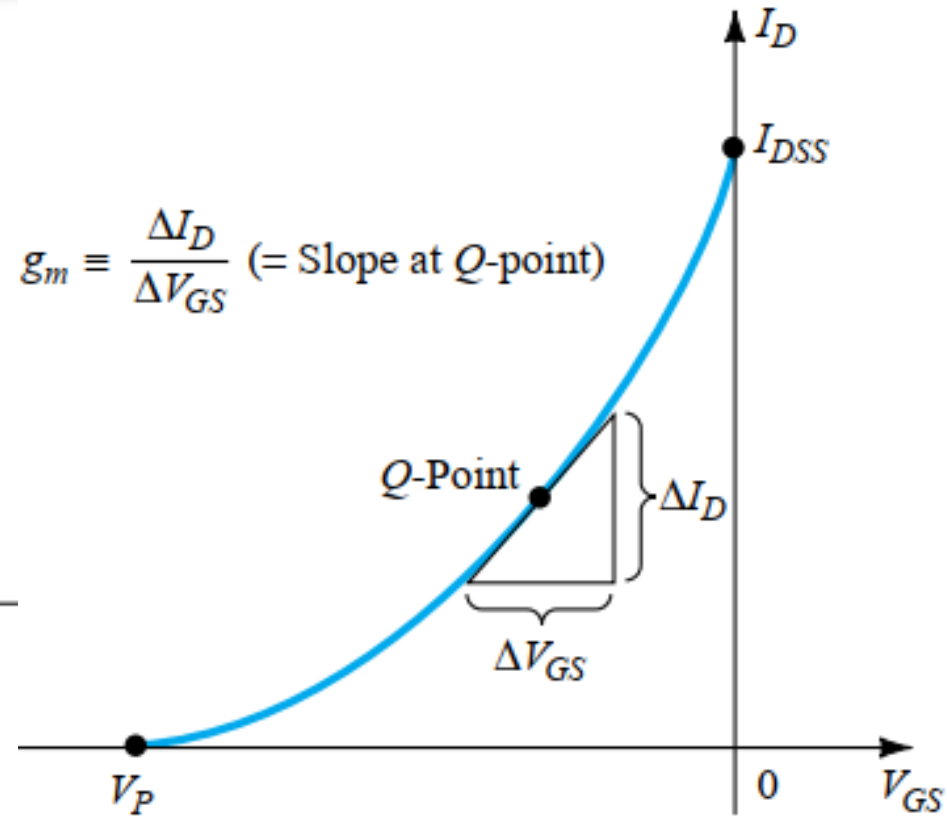
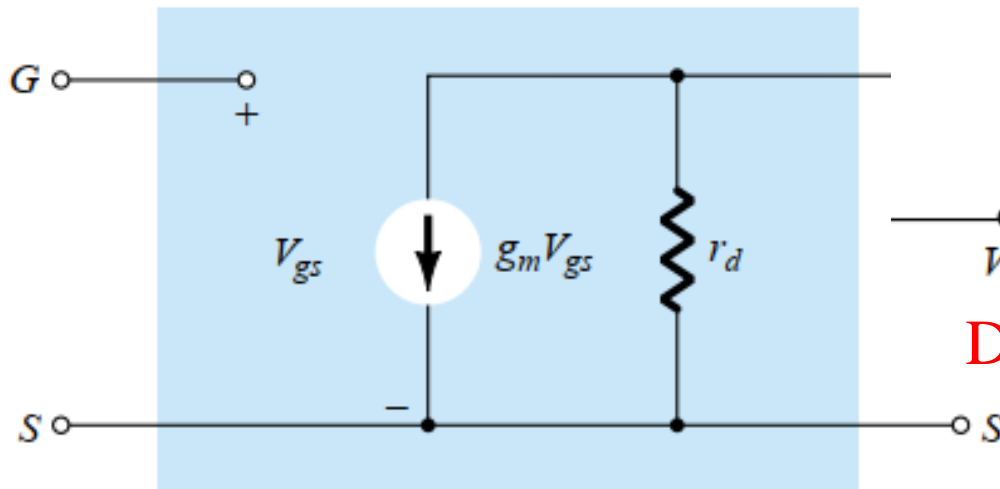
Introduction

- The **BJT** transistor is a **bipolar** device, the prefix (bi) revealing that the conduction level is a function of two charge carriers, electrons and holes.
- The **FET** is a **unipolar** device depending solely on either electron (n-channel) or hole (p-channel) conduction.
- There are **two types** of FETs transistors. **JFET** and **MOSFET**.
- .MOSFET type can be broken down into **depletion MOSFET** and **Enhancement MOSFET**.
- The gate-to-source voltage controls the drain-to-source (channel) current of an FET. $I_D = I_{DSS} (1 - V_{GS}/V_P)^2$.

Introduction

$$g_m = m = \frac{\Delta y}{\Delta x} = \frac{\Delta I_D}{\Delta V_{GS}}$$

$$g_m = \frac{2I_{DSS}}{|V_P|} \left[1 - \frac{V_{GS}}{V_P} \right]$$

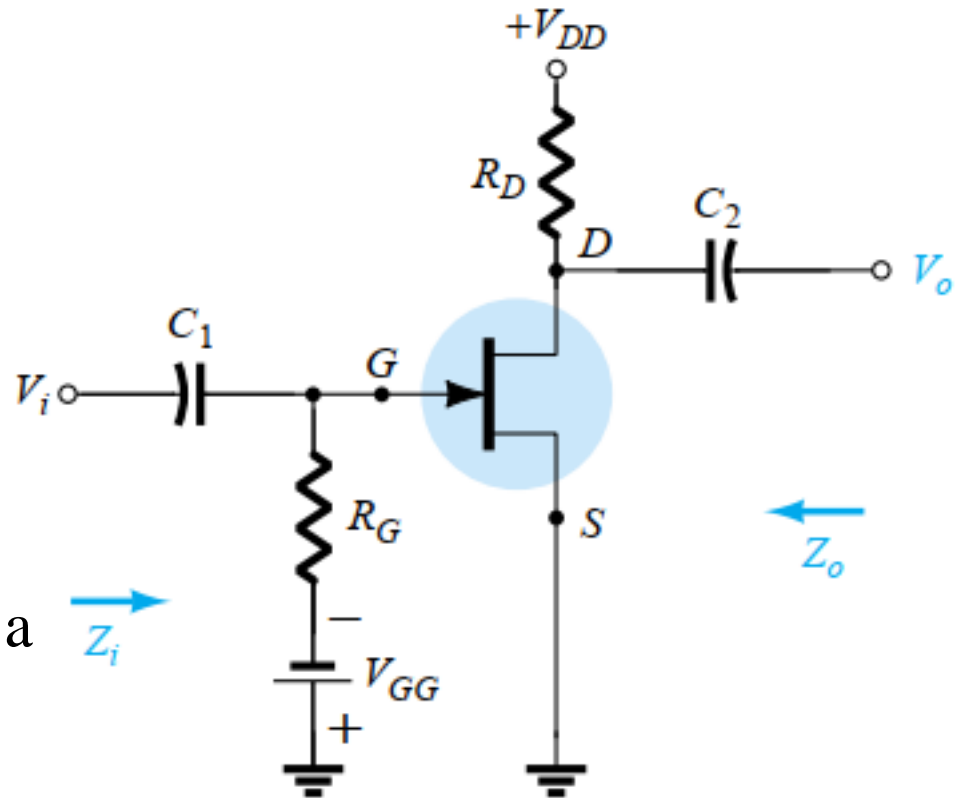


Definition of g_m using transfer characteristic.

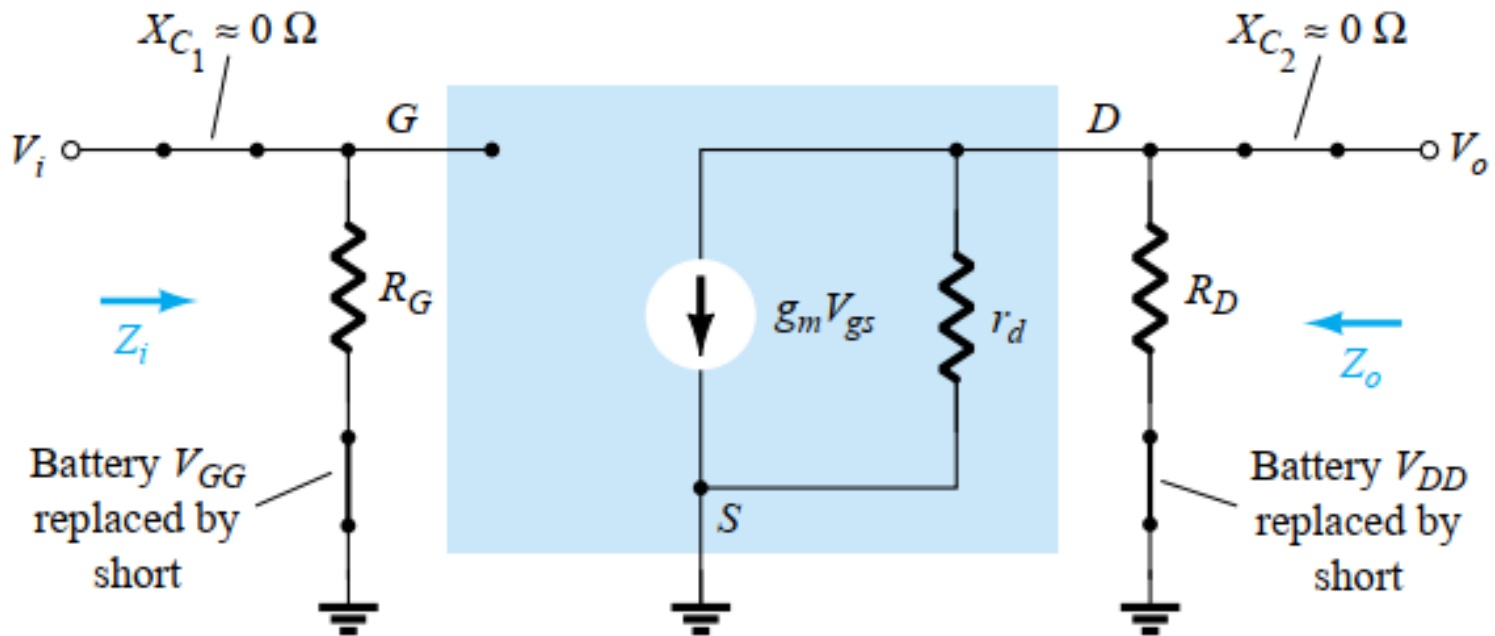
FET ac equivalent circuit

JFET Fixed-bias Configuration

- In AC analysis both capacitors have the short-circuit equivalent
- $X_c = \frac{1}{2\pi fC}$
- The dc batteries V_{GG} and V_{DD} are set to zero volts by a short-circuit equivalent.



JFET Fixed-bias Configuration

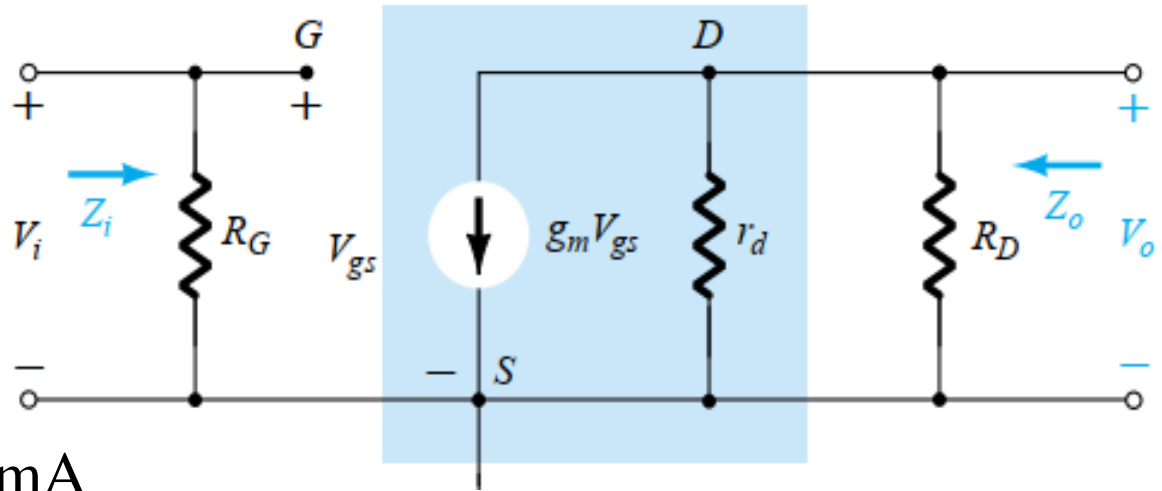


JFET AC équivalent fixed-bias configuration circuit

JFET Fixed-bias Configuration

$$Z_i = R_G$$

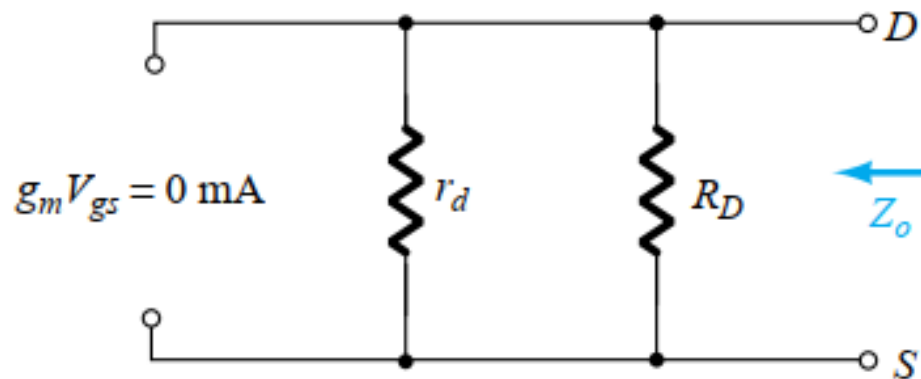
- To get Z_o set $V_i = 0\text{ v}$
- so V_{gs} as 0 v also.
- The result is $g_m V_{gs} = 0\text{ mA}$



$$Z_o = R_D \parallel r_d$$

$$Z_o \cong R_D$$

$$r_d \geq 10R_D$$



JFET Fixed-bias Configuration

$$V_o = -g_m V_{gs} (r_d \parallel R_D)$$

$$V_{gs} = V_i$$

$$V_o = -g_m V_i (r_d \parallel R_D) \quad \rightarrow$$

$$A_v = \frac{V_o}{V_i} = -g_m (r_d \parallel R_D)$$

$$A_v = \frac{V_o}{V_i} = -g_m R_D$$

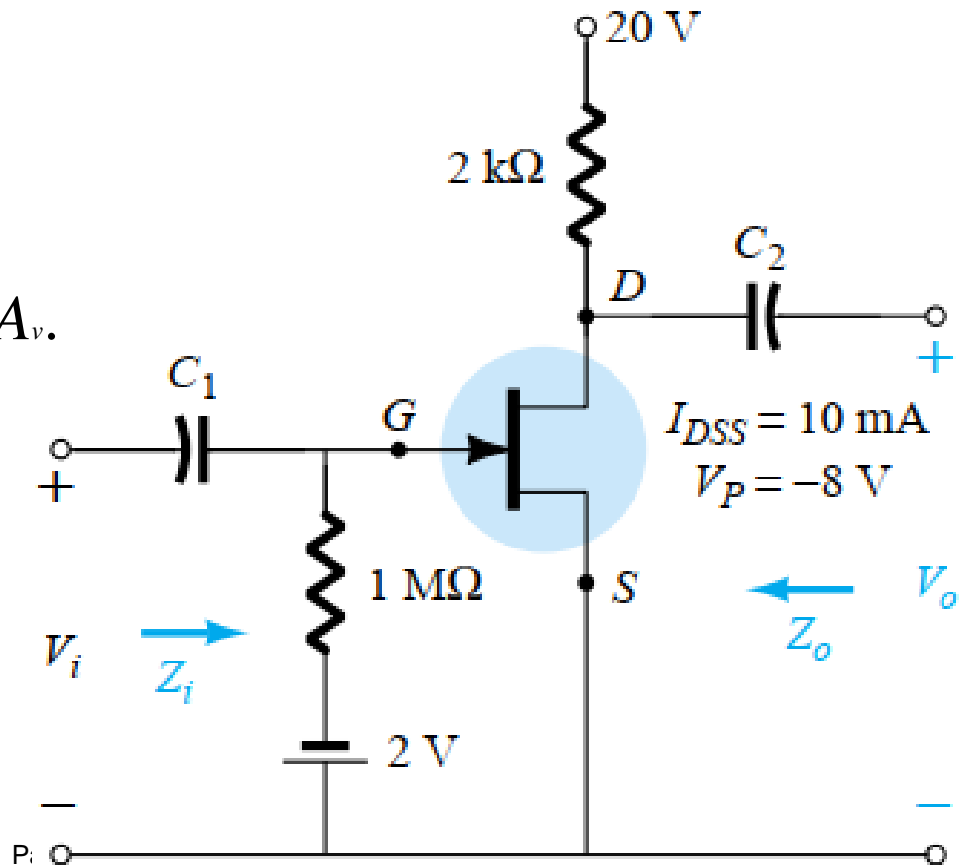
$$r_d \geq 10 R_D$$

Phase Relationship: The negative sign in the resulting equation for A_v clearly reveals a phase shift of 180° between input and output voltages.

Example

For fixed-bias configuration in the figure had an operating point defined by $V_{GSQ} = 2\text{ V}$ and $I_{DQ} = 5.625\text{ mA}$, with $I_{DSS} = 10\text{ mA}$ and $V_P = 8\text{ V}$. The value of y_{os} is provided as 40 S .

- Determine g_m .
- Find r_d .
- Determine Z_i .
- Calculate Z_o .
- Determine the voltage gain A_v .
- Determine A_v ignoring the effects of r_d .



Example

$$(a) \ g_{m0} = \frac{2I_{DSS}}{|V_P|} = \frac{2(10 \text{ mA})}{8 \text{ V}} = 2.5 \text{ mS}$$

$$g_m = g_{m0} \left(1 - \frac{V_{GS_Q}}{V_P} \right) = 2.5 \text{ mS} \left(1 - \frac{(-2 \text{ V})}{(-8 \text{ V})} \right) = \mathbf{1.88 \text{ mS}}$$

$$(b) \ r_d = \frac{1}{y_{os}} = \frac{1}{40 \ \mu\text{S}} = \mathbf{25 \text{ k}\Omega}$$

$$(c) \ Z_i = R_G = \mathbf{1 \text{ M}\Omega}$$

$$(d) \ Z_o = R_D \parallel r_d = 2 \text{ k}\Omega \parallel 25 \text{ k}\Omega = \mathbf{1.85 \text{ k}\Omega}$$

$$(e) \ A_v = -g_m(R_D \parallel r_d) = -(1.88 \text{ mS})(1.85 \text{ k}\Omega) \\ = \mathbf{-3.48}$$

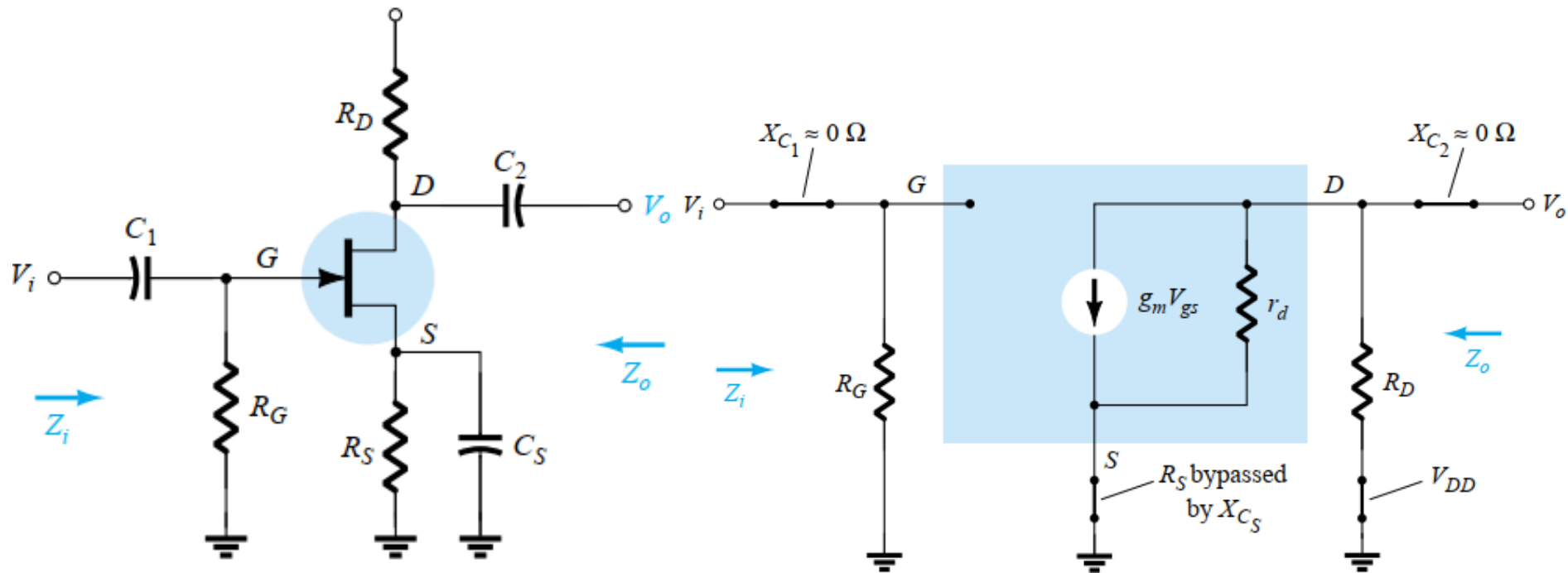
$$(f) \ A_v = -g_m R_D = -(1.88 \text{ mS})(2 \text{ k}\Omega) = \mathbf{-3.76}$$

As demonstrated in part (f), a ratio of $25 \text{ k}\Omega : 2 \text{ k}\Omega = 12.5 : 1$ between r_d and R_D resulted in a difference of 8% in solution.

JFET Self-bias Configuration

Bypassed R_S

- In AC analysis both capacitors have the short-circuit equivalent



JFET Self-bias Configuration

$$Z_i = R_G$$

$$A_v = -g_m(r_d || R_D)$$

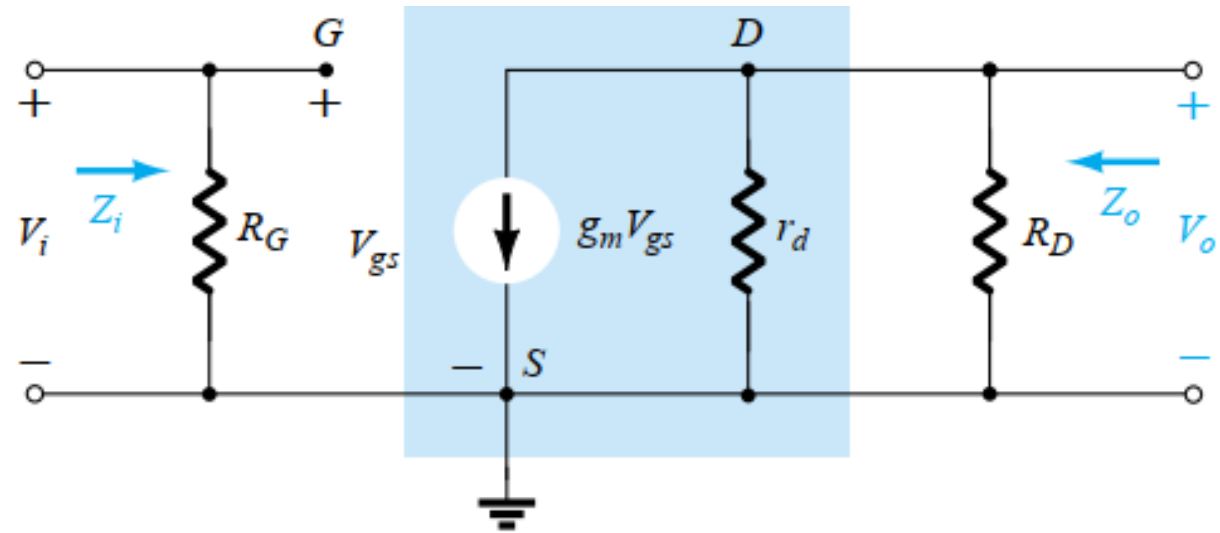
$$Z_o = r_d || R_D$$

$$A_v = -g_m R_D$$

$$r_d \geq 10 R_D$$

$$Z_o \cong R_D$$

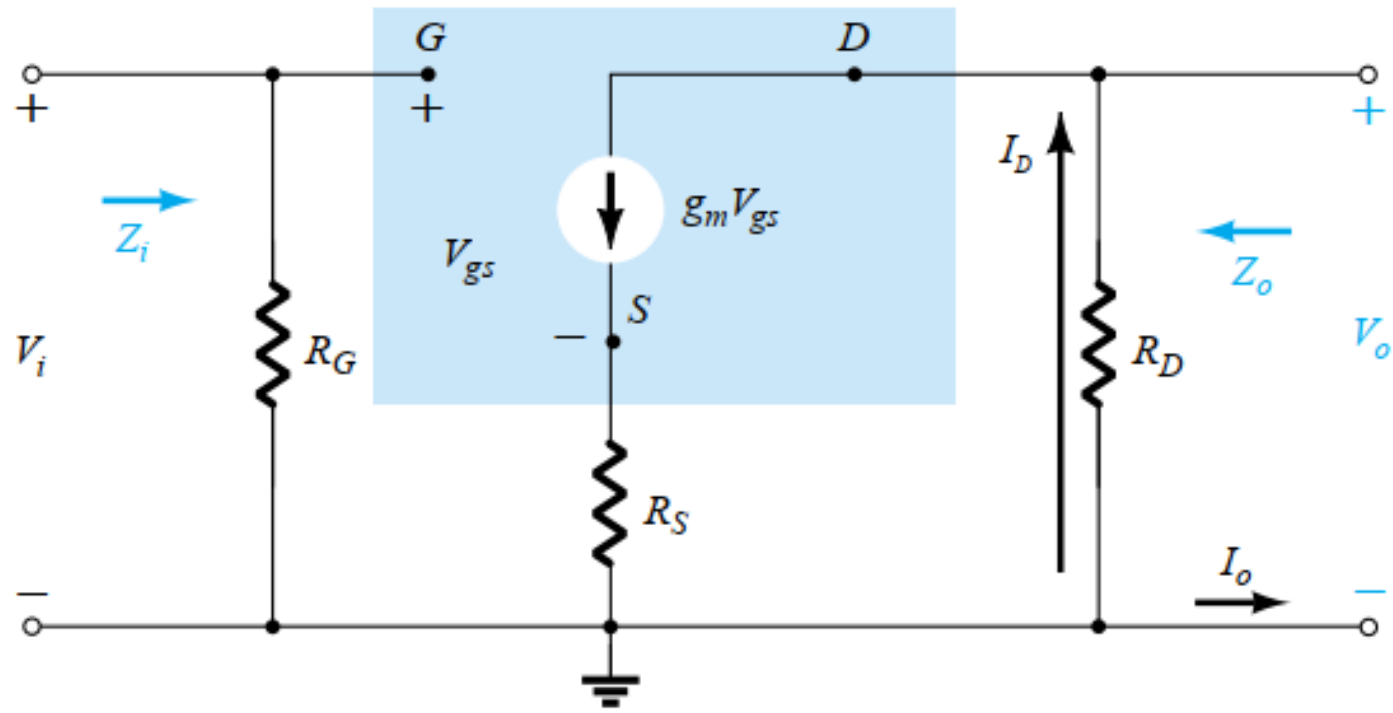
$$r_d \geq 10 R_D$$



JFET Self-bias Configuration

Un bypassed R_S

$$Z_i = R_G$$



JFET Self-bias Configuration

$$Z_o = \left. \frac{V_o}{I_o} \right|_{V_i = 0}$$

$$I_o + I_D = g_m V_{gs}$$

$$V_{gs} = -(I_o + I_D)R_S$$

$$I_o + I_D = -g_m (I_o + I_D)R_S = -g_m I_o R_S - g_m I_D R_S$$

$$I_o[1 + g_m R_S] = -I_D[1 + g_m R_S]$$

$$I_o = -I_D \quad V_o = -I_D R_D$$

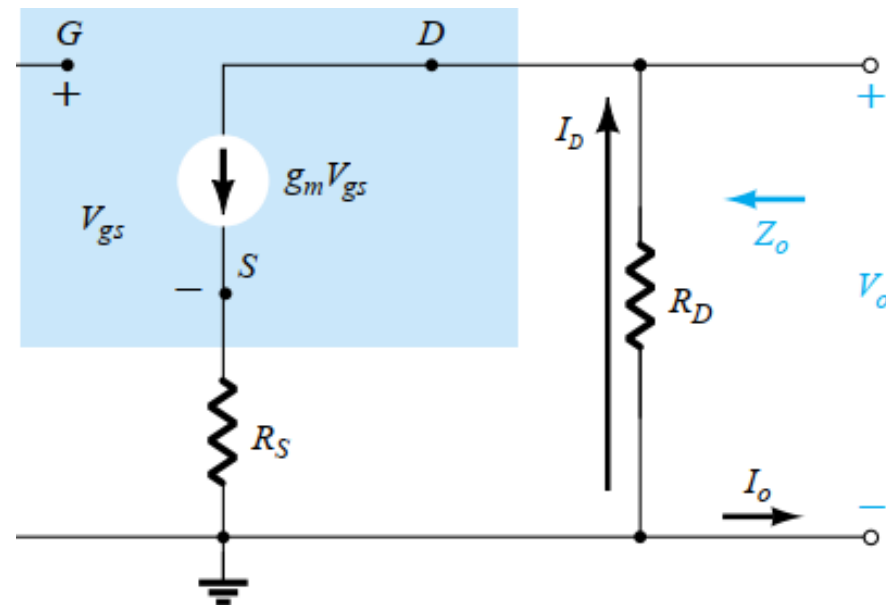
$$V_o = -(-I_o)R_D = I_o R_D$$

$$Z_o = \frac{V_o}{I_o} = R_D$$

$$r_d = \infty \Omega$$

$$A_v = \frac{V_o}{V_i} = -\frac{g_m R_D}{1 + g_m R_S}$$

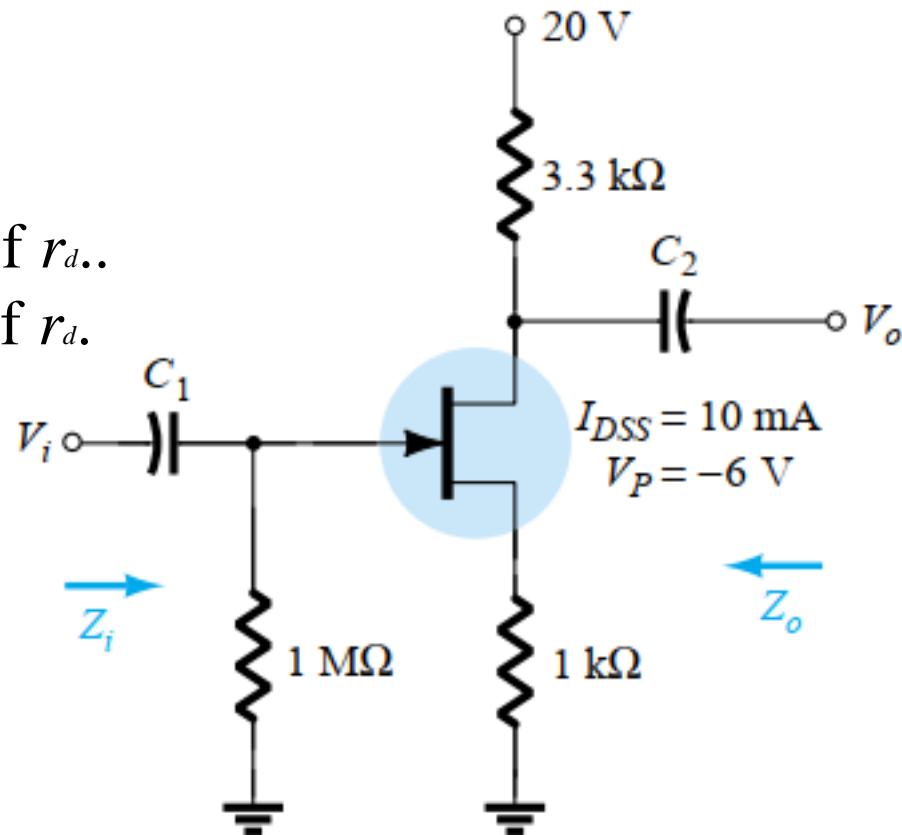
$$r_d \geq 10(R_D + R_S)$$



JFET Self-bias Configuration

The self-bias configuration in the figure has an operating point defined by $V_{GSQ} = 2.6 \text{ V}$ and $I_{DQ} = 2.6 \text{ mA}$, with $I_{DSS} = 8 \text{ mA}$ and $V_P = 6 \text{ V}$. The value of y_{os} is given as 20 S .

- Determine g_m .
- Find r_d .
- Find Z_i .
- Calculate Z_o without the effects of r_d .
- Calculate A_v without the effects of r_d .



JFET Self-bias Configuration

$$(a) \ g_{m0} = \frac{2I_{DSS}}{|V_P|} = \frac{2(8 \text{ mA})}{6 \text{ V}} = 2.67 \text{ mS}$$

$$g_m = g_{m0} \left(1 - \frac{V_{GS_Q}}{V_P} \right) = 2.67 \text{ mS} \left(1 - \frac{(-2.6 \text{ V})}{(-6 \text{ V})} \right) = \mathbf{1.51 \text{ mS}}$$

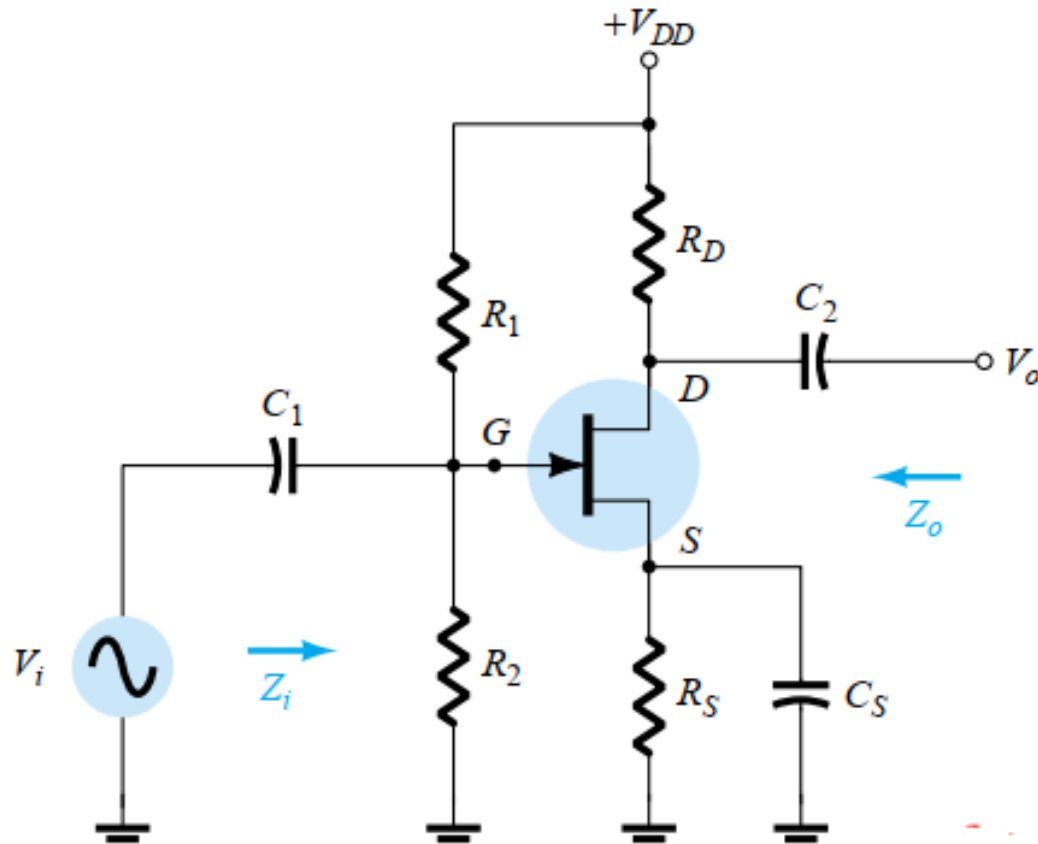
$$(b) \ r_d = \frac{1}{y_{os}} = \frac{1}{20 \ \mu\text{S}} = \mathbf{50 \text{ k}\Omega}$$

$$(c) \ Z_i = R_G = \mathbf{1 \text{ M}\Omega}$$

$$Z_o = R_D = \mathbf{3.3 \text{ k}\Omega}$$

$$A_v = \frac{-g_m R_D}{1 + g_m R_S} = \frac{-(1.51 \text{ mS})(3.3 \text{ k}\Omega)}{1 + (1.51 \text{ mS})(1 \text{ k}\Omega)} = \mathbf{-1.98}$$

JFET Voltage-divider Configuration



JFET Voltage-divider Configuration

$$Z_i = R_1 \parallel R_2$$

$$Z_o = r_d \parallel R_D$$

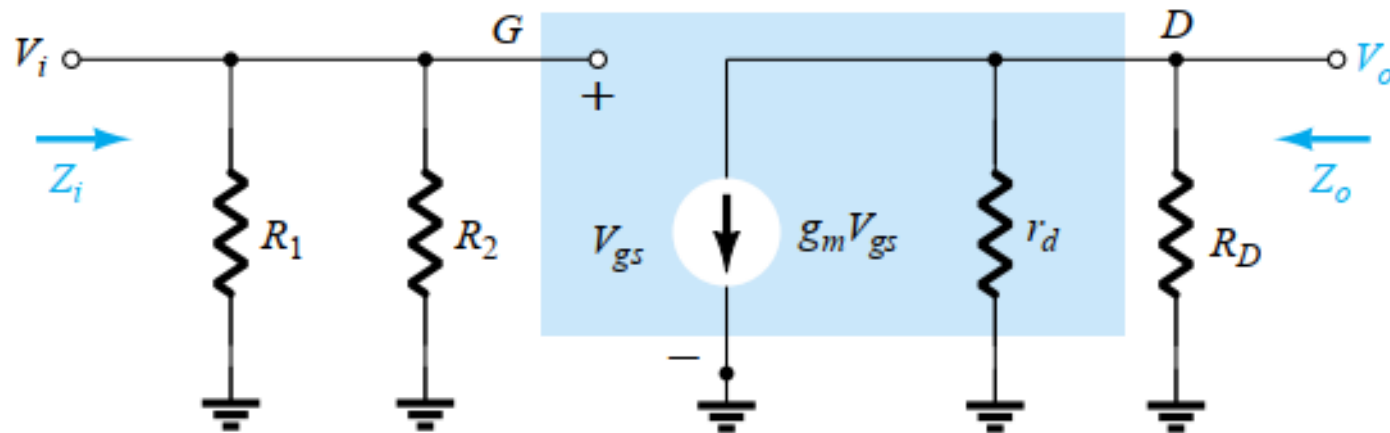
$$Z_o \cong R_D$$

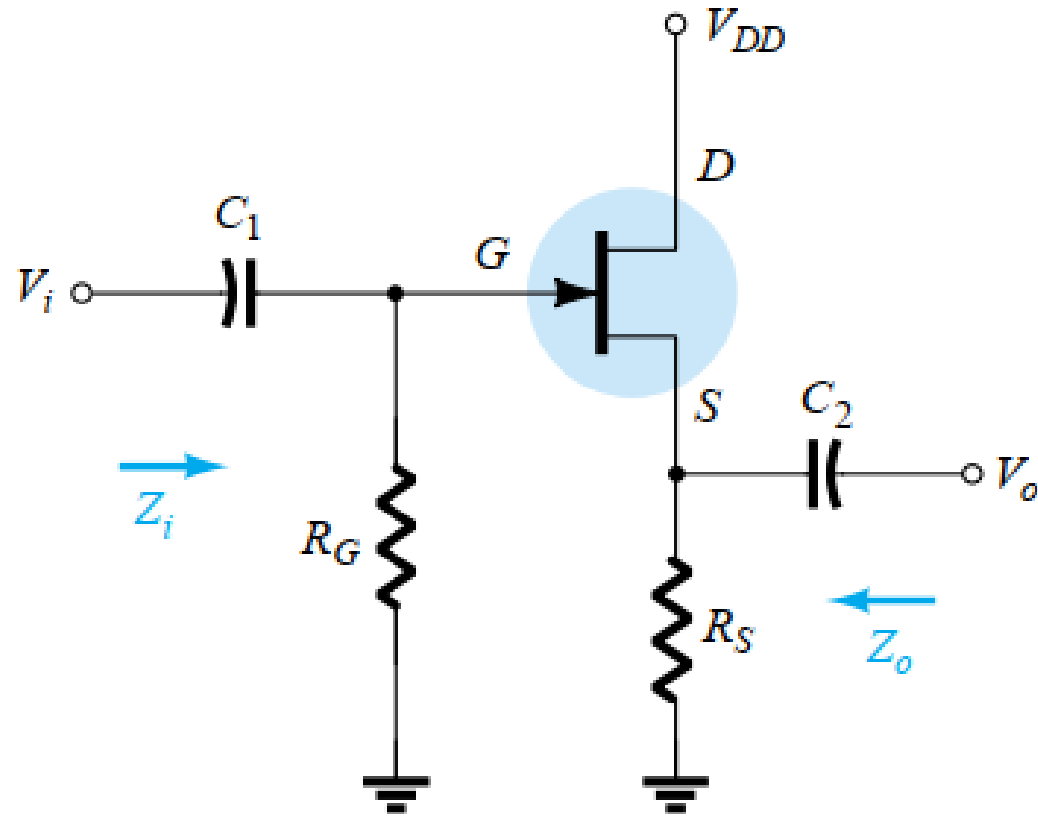
$$r_d \geq 10R_D$$

$$V_{gs} = V_i \quad V_o = -g_m V_{gs} (r_d \parallel R_D) \Rightarrow A_v = \frac{V_o}{V_i} = \frac{-g_m V_{gs} (r_d \parallel R_D)}{V_{gs}}$$

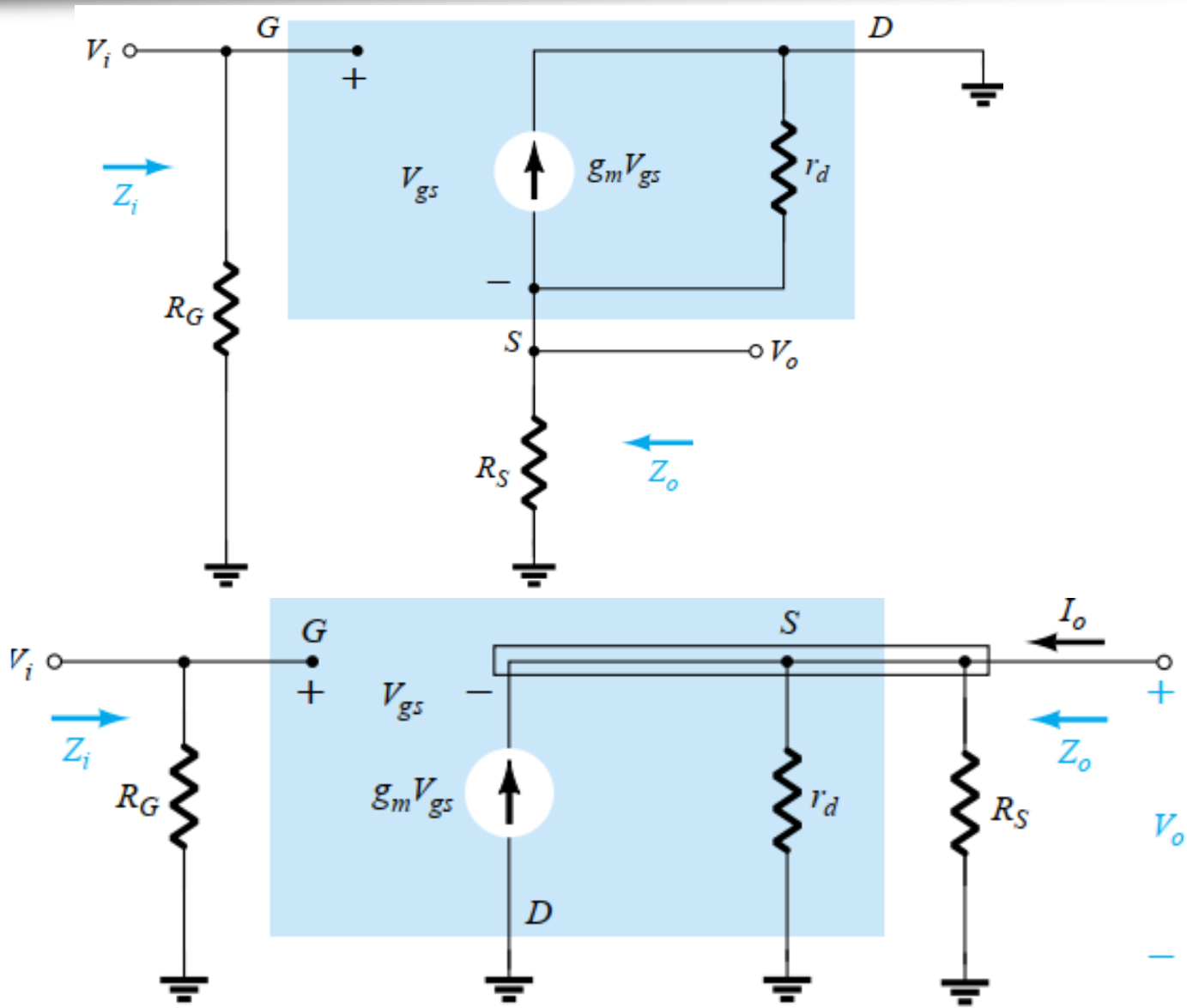
$$A_v = \frac{V_o}{V_i} \cong -g_m R_D$$

$$r_d \geq 10R_D$$





JFET Source-follower Configuration

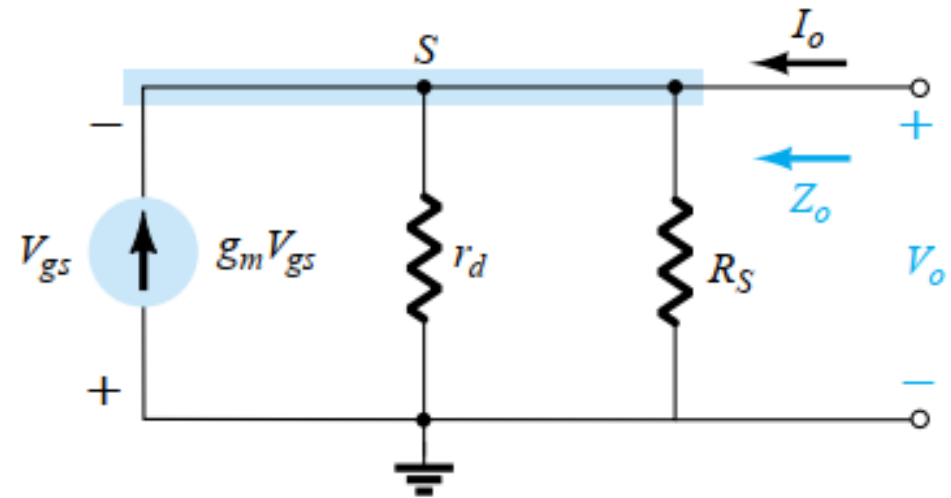


$$Z_i = R_G$$

Applying Kirchhoff's current law at node s

$$\begin{aligned} I_o + g_m V_{gs} &= I_{r_d} + I_{R_S} \\ &= \frac{V_o}{r_d} + \frac{V_o}{R_S} \end{aligned}$$

$$\begin{aligned} I_o &= V_o \left[\frac{1}{r_d} + \frac{1}{R_S} \right] - g_m V_{gs} \\ &= V_o \left[\frac{1}{r_d} + \frac{1}{R_S} \right] - g_m [-V_o] \\ &= V_o \left[\frac{1}{r_d} + \frac{1}{R_S} + g_m \right] \end{aligned}$$



JFET Source-follower Configuration

$$Z_o = \frac{V_o}{I_o} = \frac{V_o}{V_o \left[\frac{1}{r_d} + \frac{1}{R_S} + g_m \right]} = \frac{1}{\frac{1}{r_d} + \frac{1}{R_S} + g_m} = \frac{1}{\frac{1}{r_d} + \frac{1}{R_S} + \frac{1}{1/g_m}}$$

$$Z_o = r_d \parallel R_S \parallel 1/g_m$$

$$Z_o \cong R_S \parallel 1/g_m \quad r_d \geq 10R_S$$

$$V_o = g_m V_{gs} (r_d \parallel R_S) \quad V_i = V_{gs} + V_o$$

$$V_o = g_m (V_i - V_o) (r_d \parallel R_S) \quad \longrightarrow \quad V_o = g_m V_i (r_d \parallel R_S) - g_m V_o (r_d \parallel R_S)$$

$$V_o [1 + g_m (r_d \parallel R_S)] = g_m V_i (r_d \parallel R_S)$$

$$A_v = \frac{V_o}{V_i} = \frac{g_m (r_d \parallel R_S)}{1 + g_m (r_d \parallel R_S)}$$



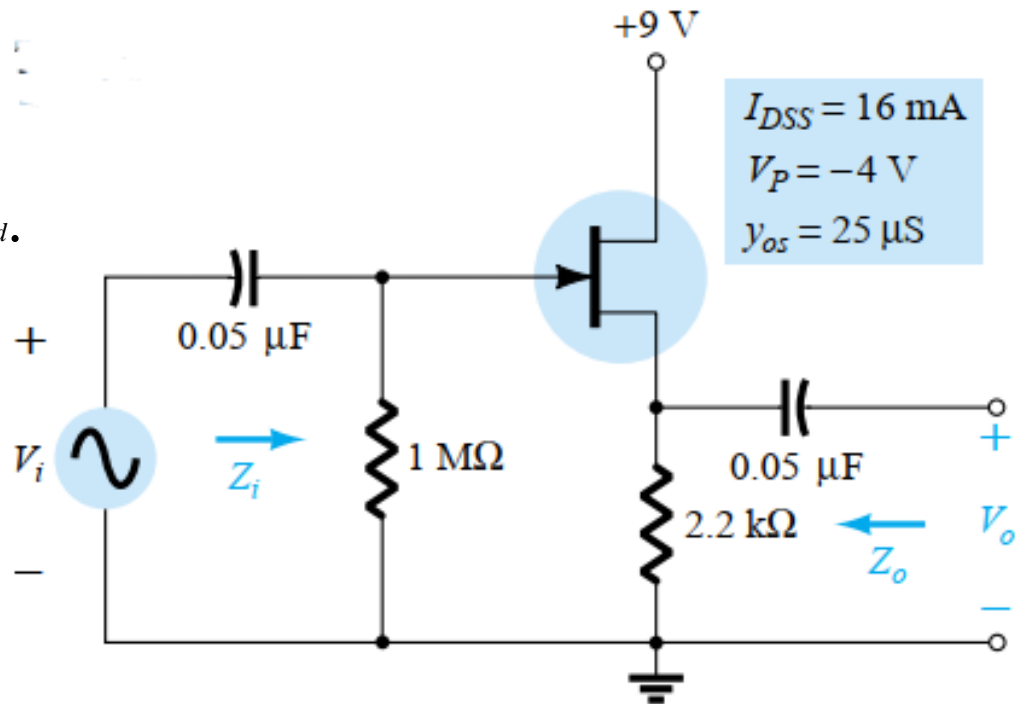
$$A_v = \frac{V_o}{V_i} \cong \frac{g_m R_S}{1 + g_m R_S}$$

$$r_d \geq 10R_S$$

Example

A dc analysis of the source-follower network of Figure will result in $V_{GSQ} = 2.86 \text{ V}$ and $I_{DQ} = 4.56 \text{ mA}$.

- Determining g_m .
- Find r_d .
- Determine Z_i .
- Calculate Z_o without r_d .
- Determine A_v without r_d .



Example

$$(a) \quad g_{m0} = \frac{2I_{DSS}}{|V_P|} = \frac{2(16 \text{ mA})}{4 \text{ V}} = 8 \text{ mS}$$

$$g_m = g_{m0} \left(1 - \frac{V_{GS_Q}}{V_P} \right) = 8 \text{ mS} \left(1 - \frac{(-2.86 \text{ V})}{(-4 \text{ V})} \right) = \mathbf{2.28 \text{ mS}}$$

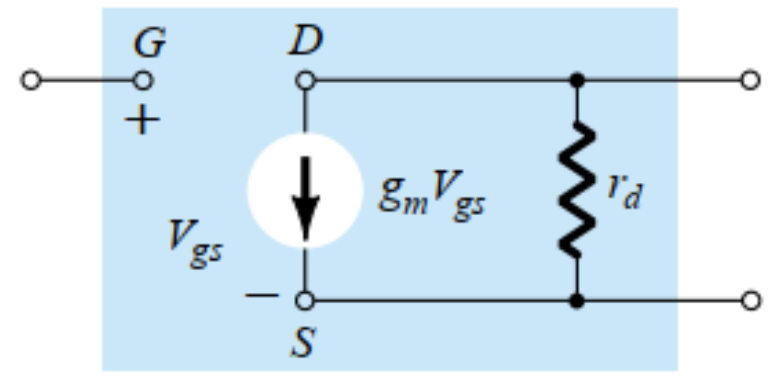
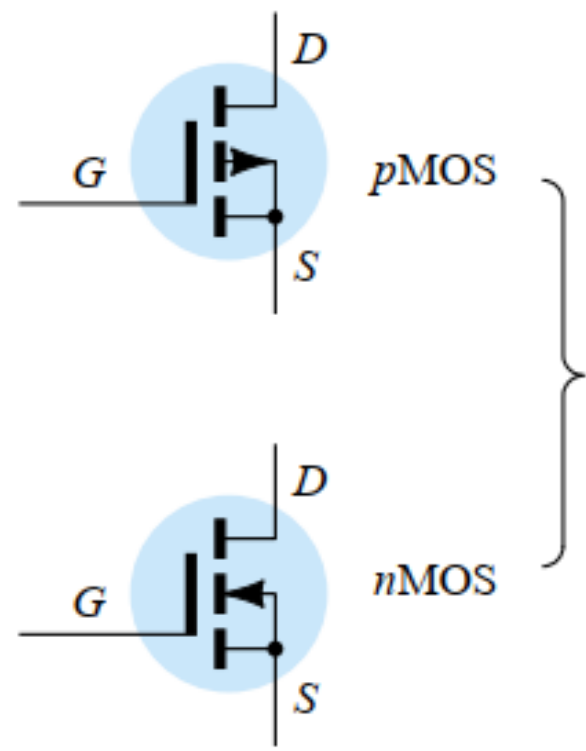
$$(b) \quad r_d = \frac{1}{y_{os}} = \frac{1}{25 \text{ } \mu\text{S}} = \mathbf{40 \text{ k}\Omega}$$

$$(c) \quad Z_i = R_G = \mathbf{1 \text{ M}\Omega}$$

$$Z_o = R_S \parallel 1/g_m = 2.2 \text{ k}\Omega \parallel 438.6 \text{ } \Omega = \mathbf{365.69 \text{ } \Omega}$$

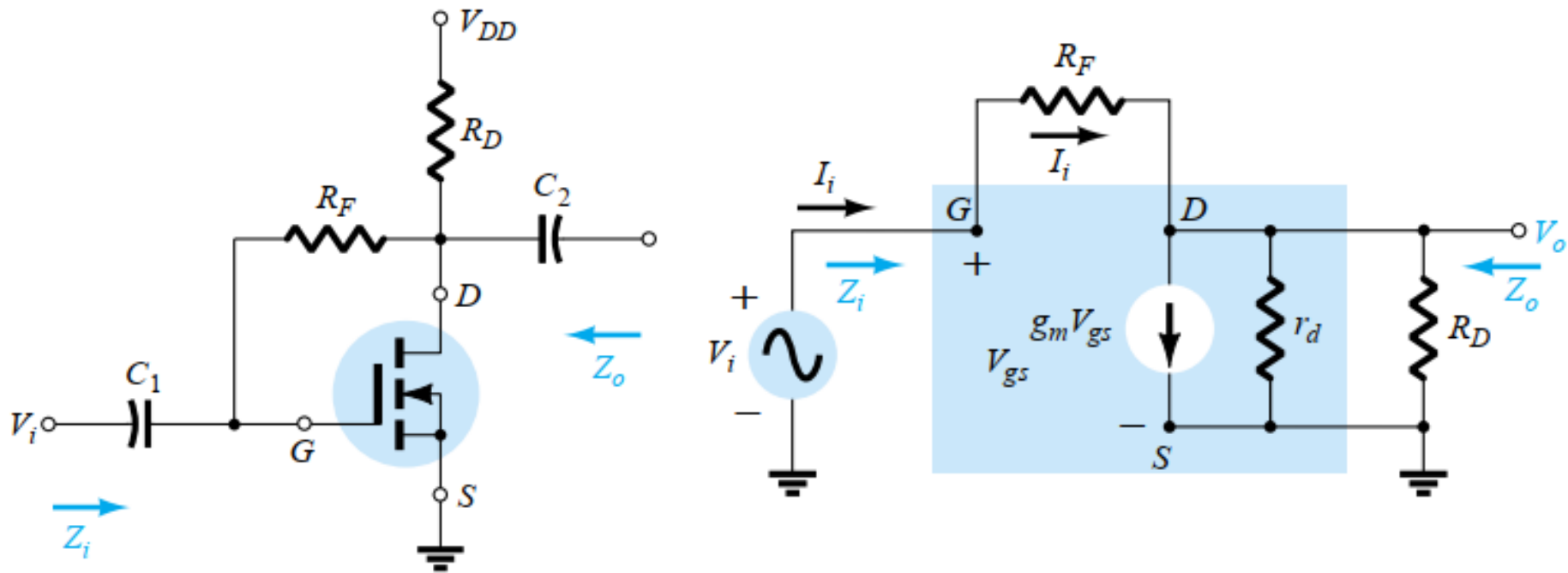
$$\begin{aligned} A_v &= \frac{g_m R_S}{1 + g_m R_S} = \frac{(2.28 \text{ mS})(2.2 \text{ k}\Omega)}{1 + (2.28 \text{ mS})(2.2 \text{ k}\Omega)} \\ &= \frac{5.02}{1 + 5.02} = \mathbf{0.83} \end{aligned}$$

Enhancement-Type MOSFETs



$$g_m = |y_{fs}|, \quad r_d = \frac{1}{|y_{os}|}$$

Recall from dc calculations that R_G could be replaced by a short-circuit equivalent since $I_G = 0$ A and therefore $V_{RG} = 0$ V.



Enhancement-Type MOSFETS

$$I_i = g_m V_{gs} + \frac{V_o}{r_d \parallel R_D}$$

$$V_{gs} = V_i$$

$$I_i = g_m V_i + \frac{V_o}{r_d \parallel R_D}$$

$$I_i - g_m V_i = \frac{V_o}{r_d \parallel R_D}$$

$$V_o = (r_d \parallel R_D)(I_i - g_m V_i)$$

$$I_i = \frac{V_i - V_o}{R_F} = \frac{V_i - (r_d \parallel R_D)(I_i - g_m V_i)}{R_F}$$

Enhancement-Type MOSFETs

$$I_i R_F = V_i - (r_d \parallel R_D) I_i + (r_d \parallel R_D) g_m V_i$$

$$V_i [1 + g_m (r_d \parallel R_D)] = I_i [R_F + r_d \parallel R_D]$$

$$Z_i = \frac{V_i}{I_i} = \frac{R_F + r_d \parallel R_D}{1 + g_m (r_d \parallel R_D)}$$

Typically, $R_F \gg r_d \parallel R_D$, so that

$$Z_i \cong \frac{R_F}{1 + g_m (r_d \parallel R_D)}$$

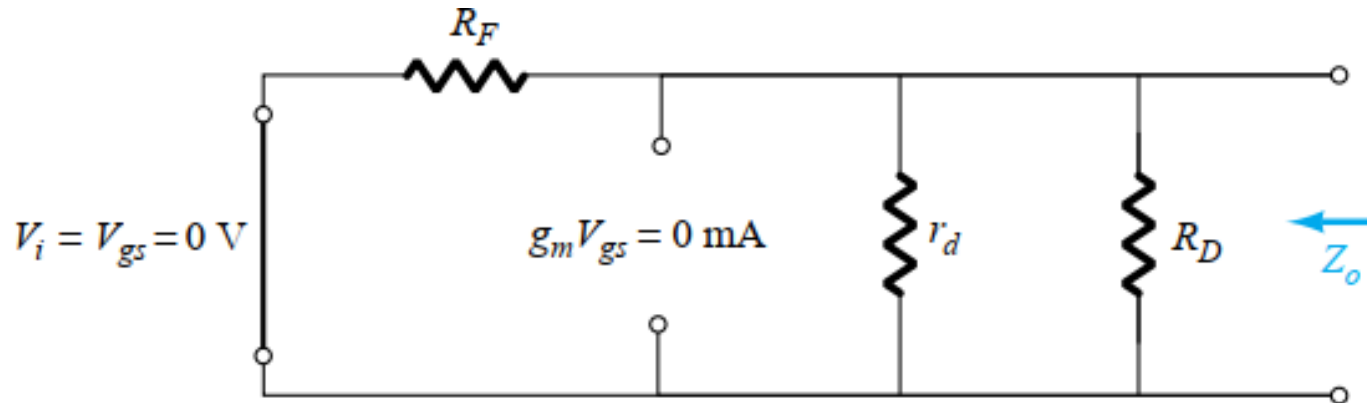
For $r_d \geq 10R_D$,

$$Z_i \cong \frac{R_F}{1 + g_m R_D}$$

$R_F \gg r_d \parallel R_D, r_d \geq 10R_D$

Enhancement-Type MOSFETs

$$Z_o = R_F \parallel r_d \parallel R_D$$



Normally, R_F is so much larger than $r_d \parallel R_D$ that

$$Z_o \cong r_d \parallel R_D$$

and with $r_d \geq 10R_D$,

$$Z_o \cong R_D$$

$$R_F \gg r_d \parallel R_D, r_d \geq 10R_D$$

Enhancement-Type MOSFETS

A_v : Applying Kirchhoff's current law at node D

$$I_i = g_m V_{gs} + \frac{V_o}{r_d \parallel R_D}$$

$$V_{gs} = V_i \text{ and } I_i = \frac{V_i - V_o}{R_F}$$

$$\frac{V_i - V_o}{R_F} = g_m V_i + \frac{V_o}{r_d \parallel R_D}$$

$$\frac{V_i}{R_F} - \frac{V_o}{R_F} = g_m V_i + \frac{V_o}{r_d \parallel R_D}$$

$$V_o \left[\frac{1}{r_d \parallel R_D} + \frac{1}{R_F} \right] = V_i \left[\frac{1}{R_F} - g_m \right]$$

Enhancement-Type MOSFETS

$$A_v = \frac{V_o}{V_i} = \frac{\left[\frac{1}{R_F} - g_m \right]}{\left[\frac{1}{r_d \parallel R_D} + \frac{1}{R_F} \right]}$$

$$\frac{1}{r_d \parallel R_D} + \frac{1}{R_F} = \frac{1}{R_F \parallel r_d \parallel R_D} \quad g_m \gg \frac{1}{R_F}$$

so that

$$A_v = -g_m(R_F \parallel r_d \parallel R_D)$$

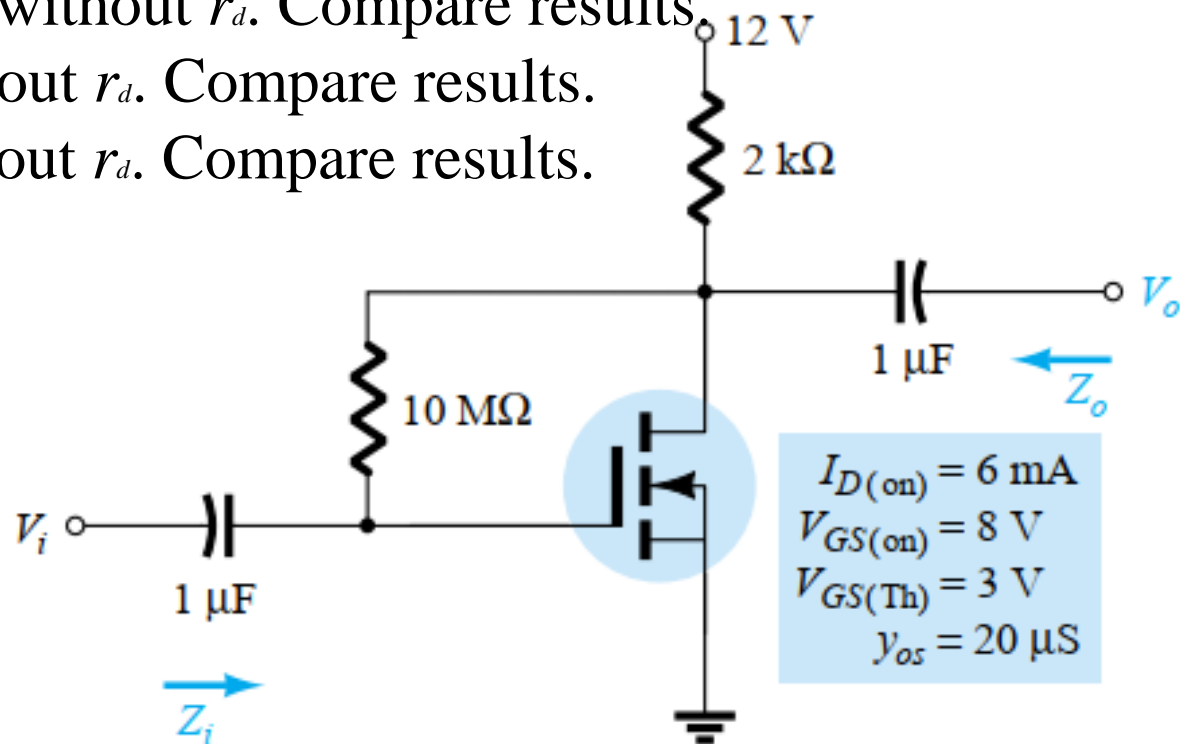
Since R_F is usually $\gg r_d \parallel R_D$ and if $r_d \geq 10R_D$,

$$A_v \cong -g_m R_D \quad R_F \gg r_d \parallel R_D, r_d \geq 10R_D$$

Example

E-MOSFET of Figure $k = 0.24 \times 10^{-3} \text{ A V}^2$, $V_{GQ} = 6.4 \text{ V}$,
and $I_{DQ} = 2.75 \text{ mA}$.

- Determine g_m .
- Find r_d .
- Calculate Z_i with and without r_d . Compare results.
- Find Z_o with and without r_d . Compare results.
- Find A_v with and without r_d . Compare results.



Example

(d) With r_d :

$$\begin{aligned}Z_o &= R_F \parallel r_d \parallel R_D = 10 \text{ M}\Omega \parallel 50 \text{ k}\Omega \parallel 2 \text{ k}\Omega = 49.75 \text{ k}\Omega \parallel 2 \text{ k}\Omega \\&= \mathbf{1.92 \text{ k}\Omega}\end{aligned}$$

Without r_d :

$$Z_o \cong R_D = \mathbf{2 \text{ k}\Omega}$$

(e) With r_d :

$$\begin{aligned}A_v &= -g_m(R_F \parallel r_d \parallel R_D) \\&= -(1.63 \text{ mS})(10 \text{ M}\Omega \parallel 50 \text{ k}\Omega \parallel 2 \text{ k}\Omega) \\&= -(1.63 \text{ mS})(1.92 \text{ k}\Omega) \\&= \mathbf{-3.21}\end{aligned}$$

Without r_d :

$$\begin{aligned}A_v &= -g_m R_D = -(1.63 \text{ mS})(2 \text{ k}\Omega) \\&= \mathbf{-3.26}\end{aligned}$$

THANK
YOU

