

# LABORATORY REPORT - CHAPTER 7

v7.1

Lastname, Firstname	KÖK, Orkun İbrahim
Student ID	22003656
Date	24.05.2023
Total Grade	/100

**Remarks:** Record all your measurements and write all your answers in the boxes provided.

## Preliminary Work

### 1. Envelope detector

- The envelope detector circuit of TRC-11 is given in Fig. 1. It serves to recover the AM modulation signal present on the RF signal. It does so by rectifying the positive side of the AM modulated signal. A low-pass circuit composed of C65 and R65 gets rid of the RF signal frequency.

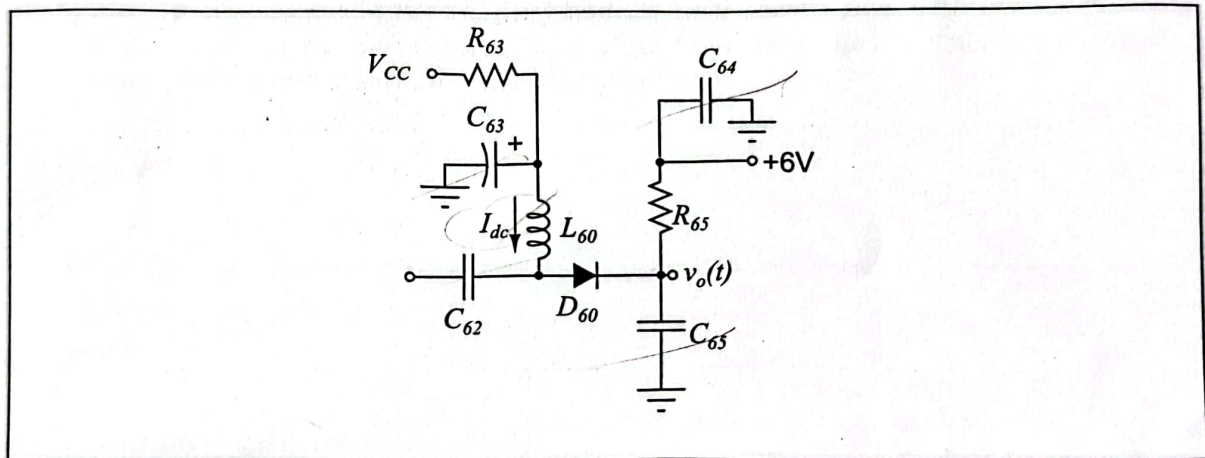
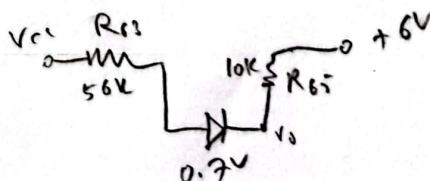


Figure 1: TRC-11 envelope detector circuit.

Calculate the dc bias current,  $I_{dc}$ , through  $R_{63}$ , the detector diode 1N4148, and  $R_{65}$ , from  $V_{CC}=12\text{ V}$  to  $+6\text{ V}$ , assuming  $C_{62}$ ,  $C_{63}$ ,  $C_{64}$ ,  $C_{65}$  are open-circuit and  $L_{60}$  is short-circuit. For this calculation, first assume that  $V_o$  of the diode is approximately  $0.7\text{ V}$ . Then, find the actual voltage  $V_o$  of 1N4148 for this bias current. You can do this by examining the datasheet of 1N4148 on page 407. The graph " $V_F$  -Forward Voltage" versus " $I_F$  -Forward Current" reveals this information. Record this voltage.



Designator	Comment	Description
C63	100 $\mu$ F	Electrolytic capacitor, polarized 16V
C64	100 nF	Capacitor, ceramic disc, 50V
C65	1 nF	Capacitor, ceramic disc, 50V
D60	1N4148	Silicon signal diode
L60	6.8 $\mu$ H	Inductor, axial leaded
R63	56 K	Resistor, carbon film, axial leaded, 1/4W
R65	10 K	Resistor, carbon film, axial leaded, 1/4W

Figure 2: Bill of materials for the envelope detector circuit

2. Calculate the DC voltage at the cathode of the diode ( $V_{ed}$ , envelope detector output voltage):

$$V_{ed} = 6 + I_{dc}R_{65}$$

3. From the same datasheet, find the capacitance,  $C_D$ , of 1N4148 when it is zero-biased.
4. L60 is a 6.8  $\mu$ H fixed inductor with a color code just like resistors, and the value is specified in  $\mu$ H. It resonates at 15 MHz with the parallel combination of C62 and capacitance,  $C_D$ , of the 1N4148 diode. The signal magnitude increases by the quality factor of the resonator. Find the value of C62 and pick the closest standard value.
5. The cutoff frequency of the low-pass network formed by C65 and R65 is set sufficiently high so that the envelope detector output can follow the variation at the high-frequency end of the modulation frequency. Find the corner frequency,  $f_c$ ,

$I_{dc} = 0.08 \text{ mA}$	$V_{dc} = 0.9$	$V_{ed} = 6.8 \text{ V}$
$C_D = 4.5 \text{ pF}$	$C_{62} = 12 \text{ pF}$	$f_c = 16 \text{ kHz}$

1.5. GRADE:

## 2. Automatic gain control (AGC)

1. A PIN diode acts like an RF resistance whose value is determined by the DC current flowing through it. It is inserted between the first and second IF amplifiers when the jumper JP70 is shorted. If the DC current through it increases, its resistance reduces and hence attenuates the signal going to the second IF amplifier as shown in Fig. 1. Refer to the datasheet of the PIN diode on page 412. Find the reverse breakdown voltage,  $V_R$ , and the series RF resistance,  $R_S$ , and the forward voltage,  $V_F$  of the PIN diode when 1 mA DC flows through it.



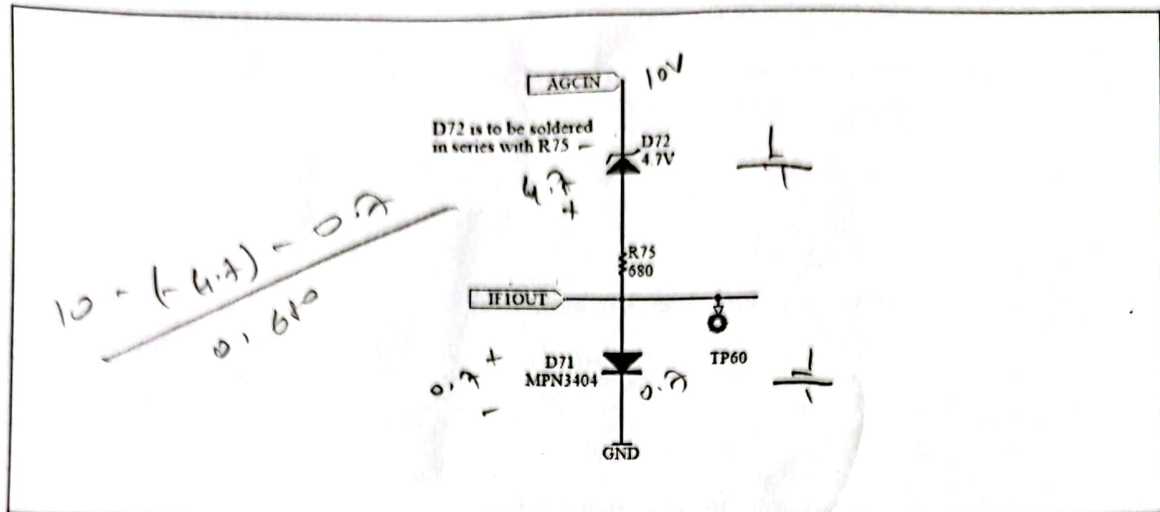


Figure 3: PIN diode inserted between first and second IF amplifiers.

$$V_R = 2.0V \quad R_S = 1.5 \Omega \quad V_F = 0.6V$$

## 2.1. GRADE:

2. The automatic gain control (AGC) circuit of TRC-11 is given in Fig. 4.

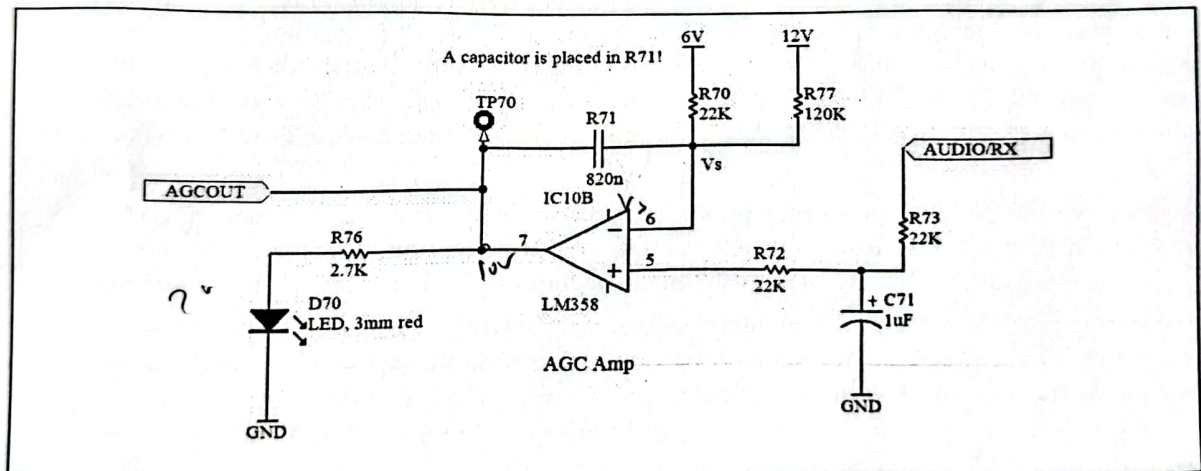


Figure 4: Automatic gain control circuit.

AGC circuit uses the OPAMP configured as an integrator in the feedback path. Integration gain is determined by 820 nF capacitor (soldered in R71 position) and the parallel combination of R70 and R77.

3. The set point,  $V_S$ , at the negative input of OPAMP is determined by R70, R77, and by the voltages they are connected to. Using nodal analysis or superposition theorem, find  $V_S$ .

$$V_S : \frac{V_S - 6}{22} + \frac{V_S - 12}{120} = 0$$

$$V_S \left( \frac{1}{22} + \frac{1}{120} \right) = \frac{6}{22} + \frac{12}{120}$$

Designator	Comment	Description
C71	1 $\mu$ F	Electrolytic capacitor, polarized 16V
D70	LED, 3mm red	Light-Emitting-Diode
D71	MPN3404	Pin diode (in series with R75)
D72	BZX55C4V7	Zener diode, 4.7 V
R70,R72,R73	22 K	Resistor, carbon film, axial leaded, 1/4W
R71	820 nF	Capacitor, ceramic disc, 50V
R75	680	Resistor, carbon film, axial leaded, 1/4W
R76	2.7K	Resistor, carbon film, axial leaded, 1/4W
R77	120K	Resistor, carbon film, axial leaded, 1/4W

**Figure 5:** Bill of Materials of the automatic gain control circuit.

$$V_S = 6.92V$$

### 2.3. GRADE:

The AGC circuit is intended to reduce the gain of the IF amplifier to prevent the saturation of the second stage. The AGC circuit observes the average value (by the integration operation) of the detector output and adjusts the IF gain by using a PIN diode placed in shunt in between the first and second IF stages. The average value of the detector output at the positive input of the OPAMP is compared with the set point,  $V_S$ , at the negative input of the OPAMP.

When there is no 15 MHz signal, the average envelope detector output voltage is  $V_{ed}$  found above. This value is compared with  $V_S$ . We should make sure that  $V_{ed}$  is less than  $V_S$ , not to reduce the gain of the IF amplifier in the absence of 15 MHz signal.

When the 15 MHz signal is present, the average value of envelope detector may become higher than the set point. In that case, the output of the OPAMP increases to a higher voltage. That increases the DC current through the PIN diode, reducing the RF resistance of the PIN diode and hence reducing the IF gain.

On the other hand, if the average detector output is less than the set point, the OPAMP output voltage decreases, decreasing the DC current through the PIN diode. This, in turn, increases the RF resistance of the PIN diode, causing an increase in the IF gain. Overall, the AGC circuit keeps the detector output at the same average level once the signal is sufficiently high. If the input signal is below the threshold, the OPAMP output is set to zero, making the DC current through the PIN diode also zero. Under that condition, the PIN diode is open-circuit for RF signals, maximizing the IF gain.

4. We have a visual indication of the AGC feedback circuit using a green LED, D70. It is connected to the OPAMP output through a resistor, R76. We would like the LED to turn



on as soon as the AGC circuit begins to work. Find the current,  $I_{LED}$ , flowing in the green LED when the AGC circuit works and OPAMP output reaches its maximum value of 10 V. For this purpose, you may assume an LED forward voltage drop of 2 V. Refer to the LED datasheet on page 415 to find the maximum allowable LED current,  $I_F$ .

$$I_{LED} = 2.76 \text{ mA} \quad I_F = 20 \text{ mA}$$

2.4. GRADE:

5. The OPAMP output that generates the DC current for the PIN diode can become as high as 10 V. Find the maximum DC current,  $I_{FMax}$ , through D72, R75, and D71 assuming the forward voltage drop of the PIN diode is 0.7 V. Find the RF resistance,  $R_{SMin}$ , of the PIN diode when this current is flowing through it using its datasheet.

$$I_{FMax} = 13.6 \text{ mA} \quad R_{SMin} = 0.7 \Omega$$

2.5. GRADE:

## Experimental Work

### 1. Envelope Detector

1. Install the components L60, R63, R65, D60, C63, C64, C65 (watch the polarity of C63) and solder them. For C62, use the value of the capacitor you calculated in the Preliminary Work.
2. Solder a loop of wire to TP62.
3. Apply the power. Using your multimeter, measure the voltage  $V_F$ , across the D60 diode. Does it agree with  $V_F$  that you have found from the datasheet of 1N4148?

$$\text{Measured } V_F = 0.5 \text{ V}$$

1.3. GRADE:

4. Set to signal generator to AM signal with 5 V<sub>pp</sub> amplitude and 50% modulation to obtain

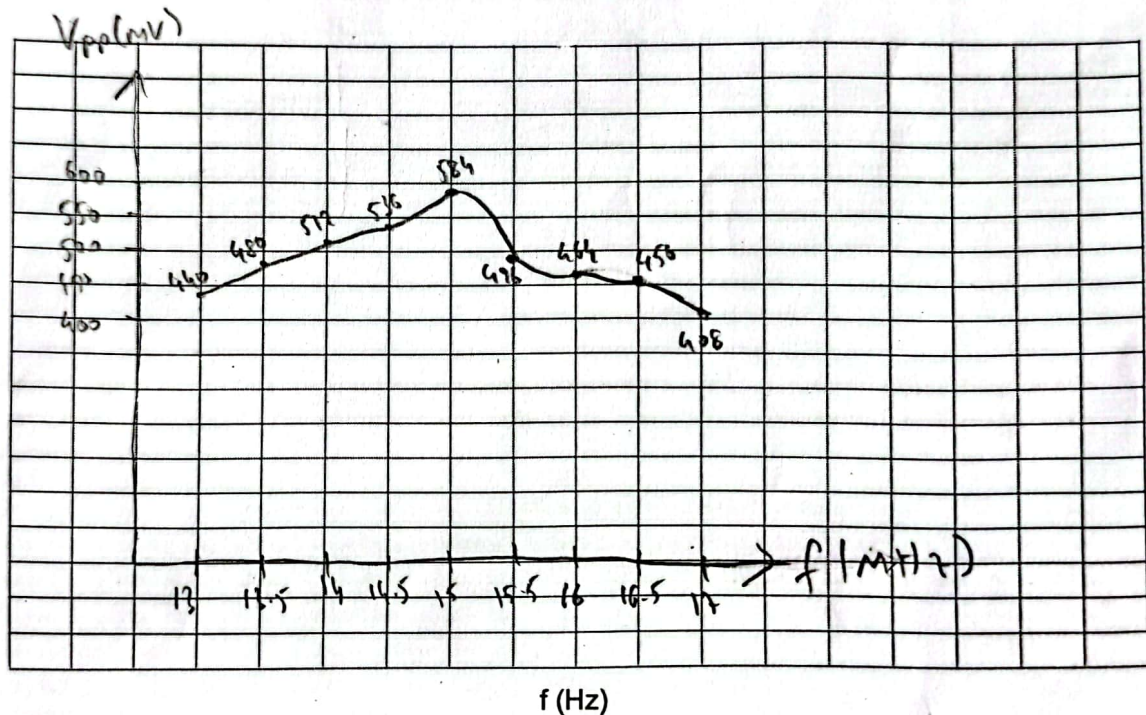
$$2.5[1 + 0.5 \cos(2\pi f_m t)] \cos(2\pi f_{IF} t)$$

where  $f_m$  is 1 KHz, and  $f_{IF}$  is 15 MHz. Press [RATE] to adjust the modulation frequency,  $f_m$ , to 1 KHz. Press [DEPTH] to adjust the modulation index to 50%. Connect the oscilloscope probe to the output of the generator. On the oscilloscope, set ACQUIRE to

PEAK DETECT. Set the time setting to 5 ms/div. Connect the "Modulation" output of the signal generator (on the back of the signal generator) to EXT trigger input of the oscilloscope using a coaxial BNC cable. Set the Trigger source of the oscilloscope to Ext, coupling AC. Make sure that the oscilloscope is triggered. Observe the AM waveform.

- Set the signal generator output voltage to 20 mV peak-to-peak. Connect the signal generator between TP60 and the ground. Connect the Ch 1 probe to the envelope detector output, TP62. On the oscilloscope, set ACQUIRE to PEAK DETECT. Apply the power to TRC-11. Record the peak-to-peak voltage,  $V_{pp}$ , as  $f_{IF}$  is varied between 13 to 17 MHz in 0.5 MHz steps. Since C62 value is adjusted to resonate with L60, you should see the maximum variation at 15 MHz. If the resonance occurs at frequencies greater than 17 MHz or lower than 13 MHz, you may change the value of C62 to the next higher or the next lower standard value.

$f_{IF}$ (MHz)	$V_{pp}$ (V)	$f_{IF}$ (MHz)	$V_{pp}$ (V)	$f_{IF}$ (MHz)	$V_{pp}$ (V)
13 MHz	440 mV	14.5 MHz	536 mV	16 MHz	464 mV
13.5 MHz	480 mV	15 MHz	584 mV	16.5 MHz	456 mV
14 MHz	512 mV	15.5 MHz	496 mV	17 MHz	408 mV

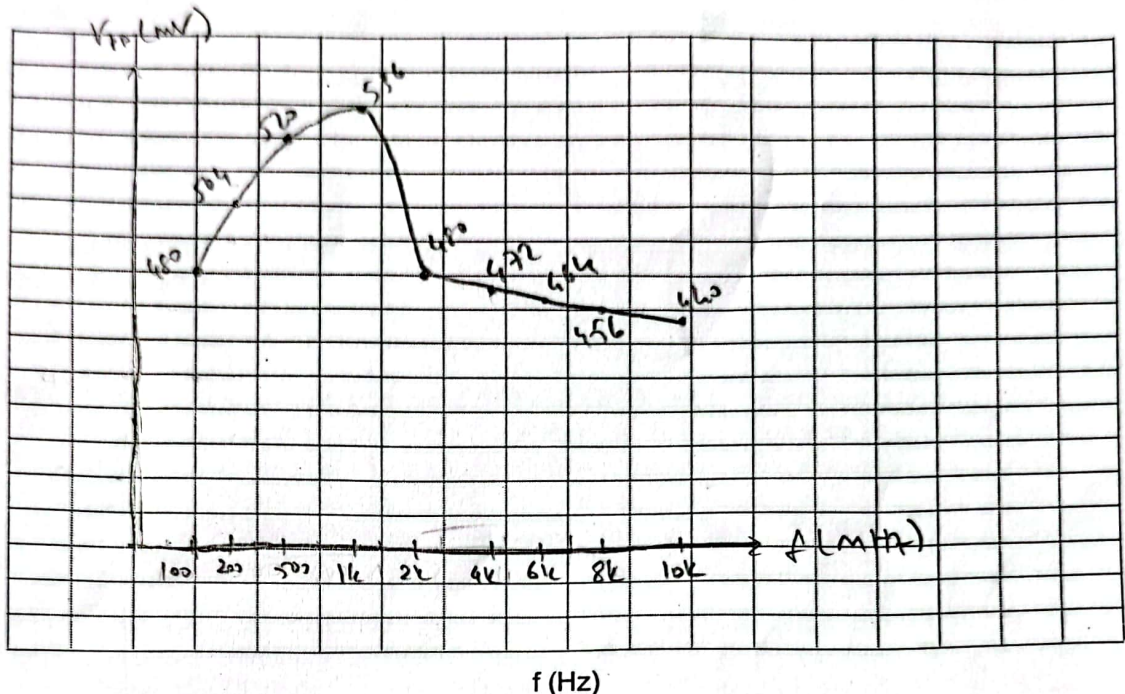


#### 1.5. GRADE:

- Set  $f_{IF}$  to 15 MHz and vary the modulation frequency  $f_m$  between 100 Hz and 10 kHz after pressing [RATE]. Modulation [DEPT] should be at 50%. Measure the envelope detector output at TP62. Choose the number of measurements and measurement frequencies adequately. Plot the peak-to-peak output voltage as a function of  $f_m$ .



$f_m$ (Hz)	$V_{pp}$ (V)	$f_m$ (Hz)	$V_{pp}$ (V)	$f_m$ (Hz)	$V_{pp}$ (V)
100 Hz	480 mV	1 kHz	520 mV	6 kHz	464 mV
200 Hz	504 mV	2 kHz	480 mV	8 kHz	456 mV
500 Hz	520 mV	4 kHz	472 mV	10 kHz	440 mV



#### 1.6. GRADE:

- Adjust the volume pot all the way counter-clockwise to reduce the volume. Plug in the earphones. Increase the volume until you hear a sound. This sound is the detected modulation signal.
- Connect the signal generator between TP40 and GND. Set the amplitude of the signal generator to the smallest amplitude (10mV). Set the frequency to the center frequency of your IF filter. This should be very close to 15.00 MHz. Set the AM modulation frequency  $f_m$  to 1 kHz. Set the modulation index to 50%. Apply the power. Now, you should hear the demodulated 1 kHz signal from the earphones. Observe 1 kHz signal at TP62 using the scope. Change the frequency of the carrier,  $f_{IF}$ , around 15 MHz in 1 KHz steps to find the maximum signal position.
- Increase the input signal amplitude. At some point, IF amplifiers saturate. The demodulated signal gets distorted. As the input amplitude increases further, the demodulated signal gets smaller. Record the input peak-to-peak signal level where the saturation begins and the signal observed on the scope begins to distort.

Input signal level for saturation= 0.03 V

### 1.9. GRADE:

## 2. Automatic gain control (AGC) circuit

1. Now, we are ready to incorporate the automatic gain control (AGC) circuitry to prevent saturation of the IF amplifiers. Refer to Fig. 4. Place R70, R72, R73, R77, and D71. Watch the direction of the PIN diode. Solder them. Cut the leads on the back side.
2. Solder one end of R75 that is connected to PIN diode, D71. Solder the cathode of D72 (the black striped end) to the other hole of R75. Solder the free ends of diode D72 and resistor R75 to each other. Note that there is no location for D72. This is circuit patch.
3. Solder C71.
4. Place the red LED, D70. Note that the longer lead is the positive terminal. Solder it. Place and solder R76.
5. Solder a loop of wire to TP70.
6. Connect the signal generator between TP40 and GND. Set the signal generator frequency to the center frequency of your IF filter (15 MHz). Set the amplitude to 10 mVpp. Set the AM modulation frequency to 1 kHz and modulation index to 50%. Using the oscilloscope, observe the demodulated signal at TP62. Increase the amplitude of the signal generator beyond the point where saturation took place. Observe that saturation is now avoided and a demodulation takes place without distortion at larger signal levels.
7. Observe that the LED turns on. You may need to increase the generator voltage or change the center frequency in small steps if the LED does not turn on. With a well-tuned receiver, the LED should turn on with about 10 mVpp at 15.000 MHz. Record the center frequency,  $f_0$ , and the smallest input amplitude which turns on the LED.
8. Debugging step: If the signal presence LED is always on, even without the input signal, measure the voltages at pins 5 and 6 of the IC10 using multimeter. These are + and - inputs of the OPAMP. Without the input signal, the voltage at + pin (pin 5) should be less than the voltage at - pin (pin 6). The expected values are 6.80 V at pin 5 and 6.84 V at pin 6. Because of resistance tolerances the voltages may be different than those. Change R77 to 100K (instead of 120K). In that case, the voltage at pin 6 should become higher (6.97 V) and the LED should turn off. With this change, the LED may turn on at a higher input signal level.
9. While at the smallest signal level, change the frequency in both directions with small steps to find out where the LED turns off.

$f_0 = 15 \text{ MHz}$  LED turns on with  $V_{inpp} = 500 \text{ mV}$

LED turns off at  $f_1 = 14.99998 \text{ MHz}$  and at  $f_2 = 15.000006 \text{ MHz}$

### 2.9. GRADE:

14.99998



10. Since the testing of IF amplifier is finished, remove the resistor R41. You may just cut its ungrounded lead.

**CHECK POINT:**