EE 254

Electronic Instrumentation

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Content (Brief)

2. Op-Amp Applications

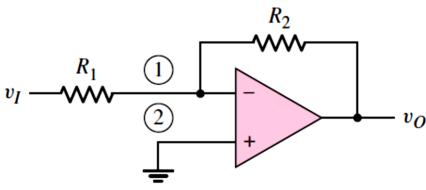
- ** Linear Applications
 - Inverting amplifiers
 - Noninverting amplifiers
 - Differential amplifiers
 - Summing amplifiers
 - Integrators
 - Differentiators
 - Low/ High pass filters
 - Instrumentational amplifiers

- ** Nonlinear Applications
 - Precision rectifiers
 - Peak detectors
 - Schmitt-trigger comparator
 - Logarithmic amplifiers

- **Short What is Inverting Amplifier?**
- \$\mathscr{C}\$ What is the circuit diagram?
- \$\text{\text{\text{W}}} What can you say about the gain of the amplifier?



What would be the output if the feedback resistor is removed?



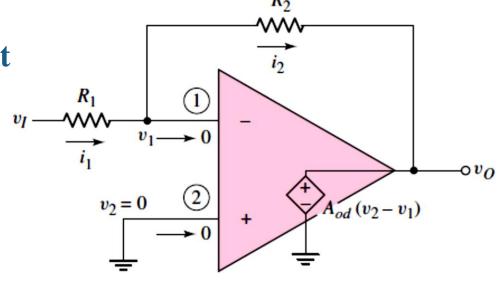
Inverting op-amp circuit

The voltage gain

$$A_{v} = \frac{v_{O}}{v_{I}}$$

The input current

$$i_1 = \frac{v_I - v_1}{R_1} = \frac{v_I}{R_1}$$



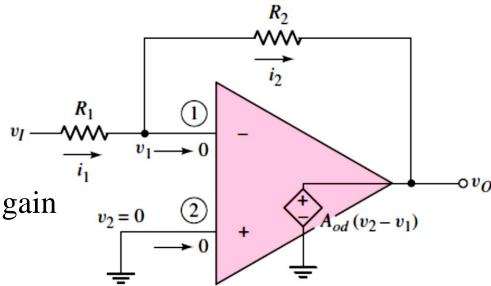
Inverting op-amp equivalent circuit

The output voltage

$$v_0 = v_1 - i_2 R_2 = 0 - \left(\frac{v_I}{R_1}\right) R_2$$

Therefore, the closed-loop voltage gain

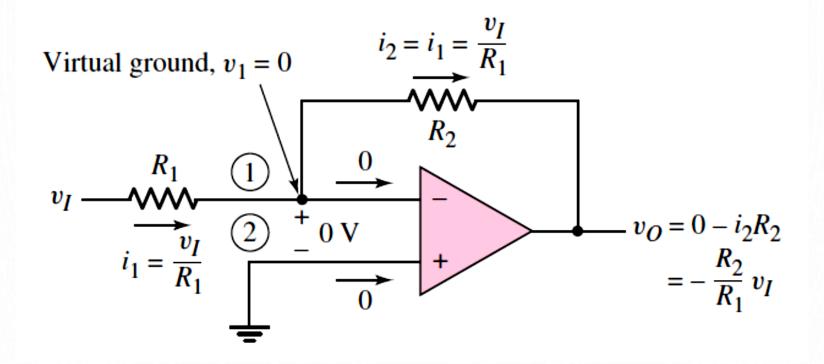
$$A_{v} = \frac{v_O}{v_I} = -\frac{R_2}{R_1}$$



Inverting op-amp equivalent circuit

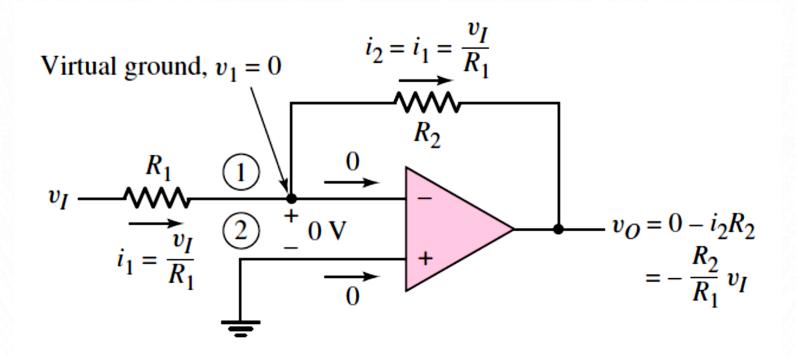
The input resistance

$$R_i = \frac{v_I}{i_1} = R_1$$

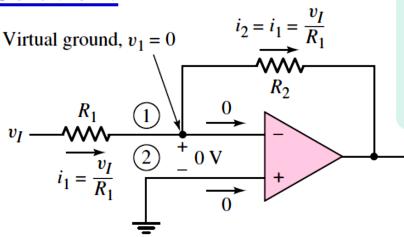


Currents and voltages in the inverting op-amp

The circuit configuration to be designed is shown in the figure. Design the circuit such that the voltage gain is $A_v = -5$. Assume the op-amp is driven by an ideal sinusoidal source, $v_s = 0.1 \sin \omega t$ (V), that can supply a maximum current of 5 μ A. Assume that frequency ω is low so that any frequency effects can be neglected.



Solution:



Given

$$A_v = -5$$

$$v_s = 0.1 \sin \omega t (V)$$

$$i_1(max) = 5 \mu A$$

The input current i_1

$$i_1 = \frac{v_I}{R_1} = \frac{v_S}{R_1}$$

If $i_1(max) = 5 \mu A$

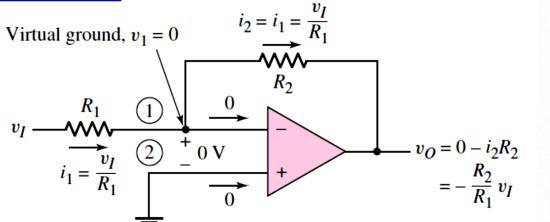
The closed-loop gain A_v is given by $A_v = -\frac{R_2}{R_1} = -5$

We then have $R_2 = 5R_1 = 5(20) = 100k\Omega$

<u>Trade-offs</u>: If the signal source has a **finite output resistance** and the desired output voltage is $v_0 = -0.5 \sin \omega t$, the circuit must be redesigned.

Assume the output resistance of the source is $R_S = 1 k\Omega$.

Solution:



Redesign Solution:

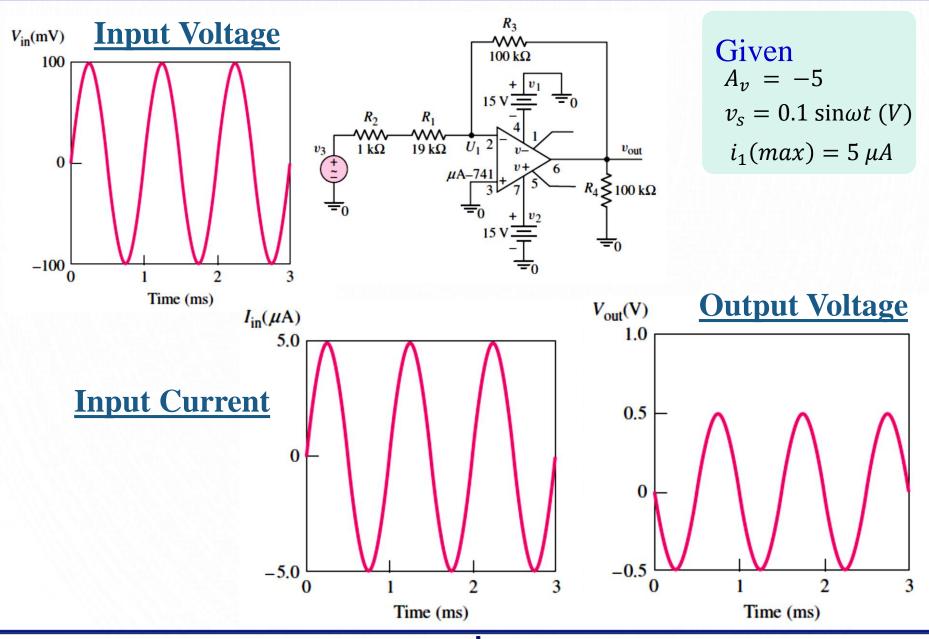
The output resistance of the signal source is now part of the input resistance to the op-amp.

Then,
$$R_1 + R_S = \frac{v_S(max)}{i_1(max)} = \frac{0.1}{5 \times 10^{-6}} \Rightarrow 20k\Omega$$

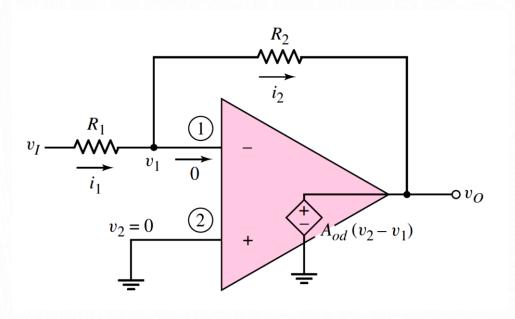
Since
$$R_S = 1 k\Omega$$
, We then have; $R_1 = 19 k\Omega$

The feedback resistor is then;

$$R_2 = 5(R_1 + R_s) = 5(19 + 1) = 100 k\Omega$$



Effect of Finite Gain



The current through R_1

$$i_1 = \frac{v_I - v_1}{R_1}$$

The current through R_2

$$i_2 = \frac{v_1 - v_0}{R_2}$$

The output voltage

$$v_0 = -A_{od}v_1$$

The terminal (1) voltage

$$v_1 = -\frac{v_O}{A_{od}}$$

Then,

$$i_1 = \frac{v_I + \frac{v_O}{A_{od}}}{R_1} = i_2 = \frac{-\frac{v_O}{A_{od}} - v_O}{R_2}$$

Effect of Finite Gain

$$\frac{v_I + \frac{v_O}{A_{od}}}{R_1} = \frac{-\frac{v_O}{A_{od}} - v_O}{R_2}$$

Solving for the closed-loop gain

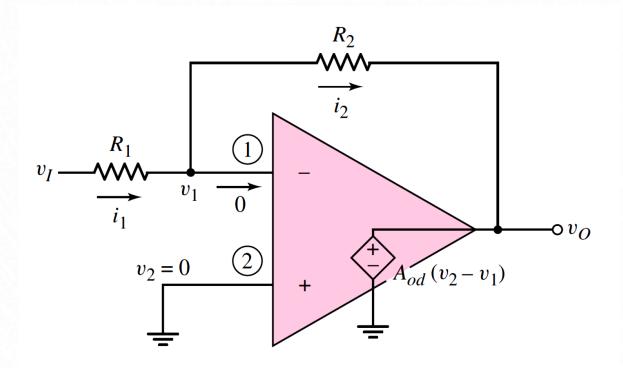
$$A_{v} = \frac{v_{O}}{v_{I}} = -\frac{R_{2}}{R_{1}} \frac{1}{\left[1 + \frac{1}{A_{od}} \left(1 + \frac{R_{2}}{R_{1}}\right)\right]}$$

If $A_{od} \rightarrow \infty$, the gain becomes same as the ideal closed-loop gain

$$A_{v} = \frac{v_{O}}{v_{I}} = \lim_{A_{od} \to \infty} \left(-\frac{R_{2}}{R_{1}} \frac{1}{\left[1 + \frac{1}{A_{od}} \left(1 + \frac{R_{2}}{R_{1}}\right)\right]} \right) = -\frac{R_{2}}{R_{1}}$$

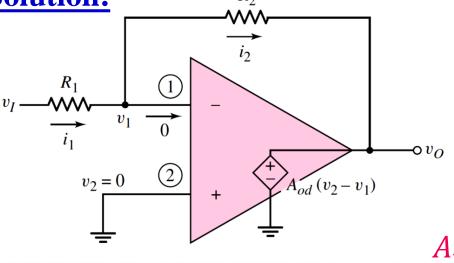
Ex. 02: Study the Effect of A_{od}

Determine the deviation from the ideal due to a finite differential gain. Consider an inverting op-amp with $R_1 = 10k\Omega$ and $R_2 = 100k\Omega$. Determine the closed-loop gain for: $A_{od} = 10^2$, 10^3 , 10^4 , 10^5 , and 10^6 . Calculate the percent deviation from the ideal gain.



Ex. 02: Study the Effect of A_{od}

Solution:



The ideal closed-loop gain;

$$A_v = -\frac{R_2}{R_1} = -\frac{100}{10} = -10$$

If
$$A_{od} = 10^2$$
,

$$A_{v} = \frac{v_{O}}{v_{I}} = -\frac{R_{2}}{R_{1}} \frac{1}{\left[1 + \frac{1}{A_{od}} \left(1 + \frac{R_{2}}{R_{1}}\right)\right]}$$

Then;

$$A_{v} = -\frac{100}{10} \cdot \frac{1}{\left[1 + \frac{1}{10^{2}} \left(1 + \frac{100}{10}\right)\right]} = -\frac{10}{(1 - 0.11)} = -9.01$$

Which is a 9.9% deviation from the ideal.

Ex. 02: Study the Effect of A_{od}

Solution:

For the other differential gain values, we have the following results:

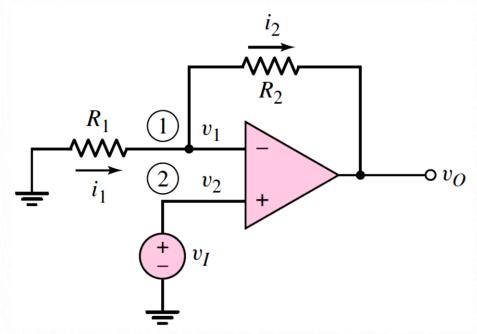
A_{od}	A_{v}	Deviation (%)
10^{2}	-9.01	9.9
10^{3}	-9.89	1.1
10^{4}	-9.989	0.11
10^{5}	-9.999	0.01
10^{6}	-9.9999	0.001

- In this case, the open loop gain must be at least 10^3 in order to be within 1% of the ideal gain.
- At low frequencies, most op-amp circuits have gains on the order of 10⁵, so achieving the required accuracy is not difficult.

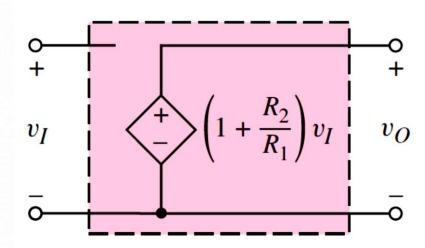
Non-Inverting Amplifier

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Non-inverting Amplifier



Non-inverting op-amp circuit



Equivalent circuit of ideal noninverting op-amp

We assume that no current enters the input terminals. Since $v_1 = v_2$, then $v_1 = v_I$, and current i_1 is given by

$$i_1 = -\frac{v_1}{R_1} = -\frac{v_I}{R_1}$$

Non-inverting Amplifier

Current i_2 is given by

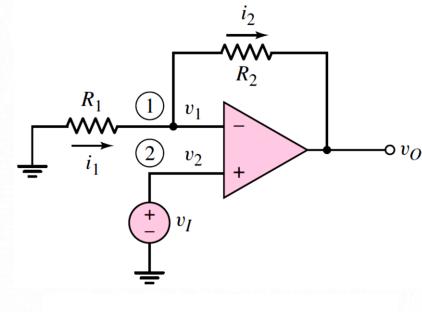
$$i_2 = \frac{v_1 - v_O}{R_2} = \frac{v_I - v_O}{R_2}$$

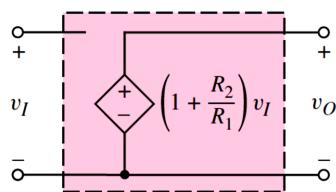
As before, $i_1 = i_2$, so that

$$-\frac{v_I}{R_1} = \frac{v_I - v_O}{R_2}$$

The closed-loop voltage gain A_v

$$A_{v} = \frac{v_{O}}{v_{I}} = 1 + \frac{R_{2}}{R_{1}}$$





Equivalent circuit of ideal noninverting op-amp