



Robotics 2

Trajectory Tracking Control

Prof. Alessandro De Luca

DIPARTIMENTO DI INGEGNERIA INFORMATICA
AUTOMATICA E GESTIONALE ANTONIO RUBERTI



SAPIENZA
UNIVERSITÀ DI ROMA



Inverse dynamics control

given the robot **dynamic model**

$$M(q)\ddot{q} + n(q, \dot{q}) = u$$

$c(q, \dot{q}) + g(q) + \text{friction model}$

and a twice-differentiable **desired trajectory** for $t \in [0, T]$

$$q_d(t) \rightarrow \dot{q}_d(t), \ddot{q}_d(t)$$

applying the **feedforward** torque in **nominal conditions**

$$u_d = M(q_d)\ddot{q}_d + n(q_d, \dot{q}_d)$$

yields exact reproduction of the desired motion, provided that $q(0) = q_d(0), \dot{q}(0) = \dot{q}_d(0)$ (initial **matched state**)



In practice ...

a number of differences from the **nominal condition**

- initial state is “**not matched**” to the desired trajectory $q_d(t)$
- **disturbances** on the actuators, truncation errors on data, ...
- **inaccurate knowledge** of robot dynamic parameters (link masses, inertias, center of mass positions)
- **unknown** value of the carried payload
- presence of **unmodeled** dynamics (complex friction phenomena, transmission elasticity, ...)



Introducing feedback

$\hat{u}_d = \hat{M}(q_d)\ddot{q}_d + \hat{n}(q_d, \dot{q}_d)$ with \hat{M} , \hat{n} **estimates** of terms (or coefficients) in the dynamic model

note: \hat{u}_d can be computed **off line** [e.g., by $\hat{N}E_\alpha(q_d, \dot{q}_d, \ddot{q}_d)$]

feedback is introduced to make the control scheme more robust

different possible implementations depending on amount of **computational load** share

- **OFF LINE** (\longleftrightarrow open loop)
- **ON LINE** (\longleftrightarrow closed loop)

two-step control design:

1. compensation (**feedforward**) or cancellation (**feedback**) of **nonlinearities**
2. synthesis of a **linear** control law stabilizing the trajectory error to zero



A series of trajectory controllers

(assuming the nominal case: $\hat{M} = M, \hat{n} = n$)

1. inverse dynamics compensation (FFW) + PD

$$u = \hat{u}_d + K_P(q_d - q) + K_D(\dot{q}_d - \dot{q})$$

local stabilization
of trajectory error
 $e(t) = q_d(t) - q(t)$

2. inverse dynamics compensation (FFW) + **variable** PD

$$u = \hat{u}_d + \hat{M}(q_d)[K_P(q_d - q) + K_D(\dot{q}_d - \dot{q})]$$

global if additional
conditions on
 K_P and K_D

3. **feedback linearization (FBL)** + [PD+FFW] = "COMPUTED TORQUE"

$$u = \hat{M}(q)[\ddot{q}_d + K_P(q_d - q) + K_D(\dot{q}_d - \dot{q})] + \hat{n}(q, \dot{q})$$

4. feedback linearization (FBL) + [PID+FFW]

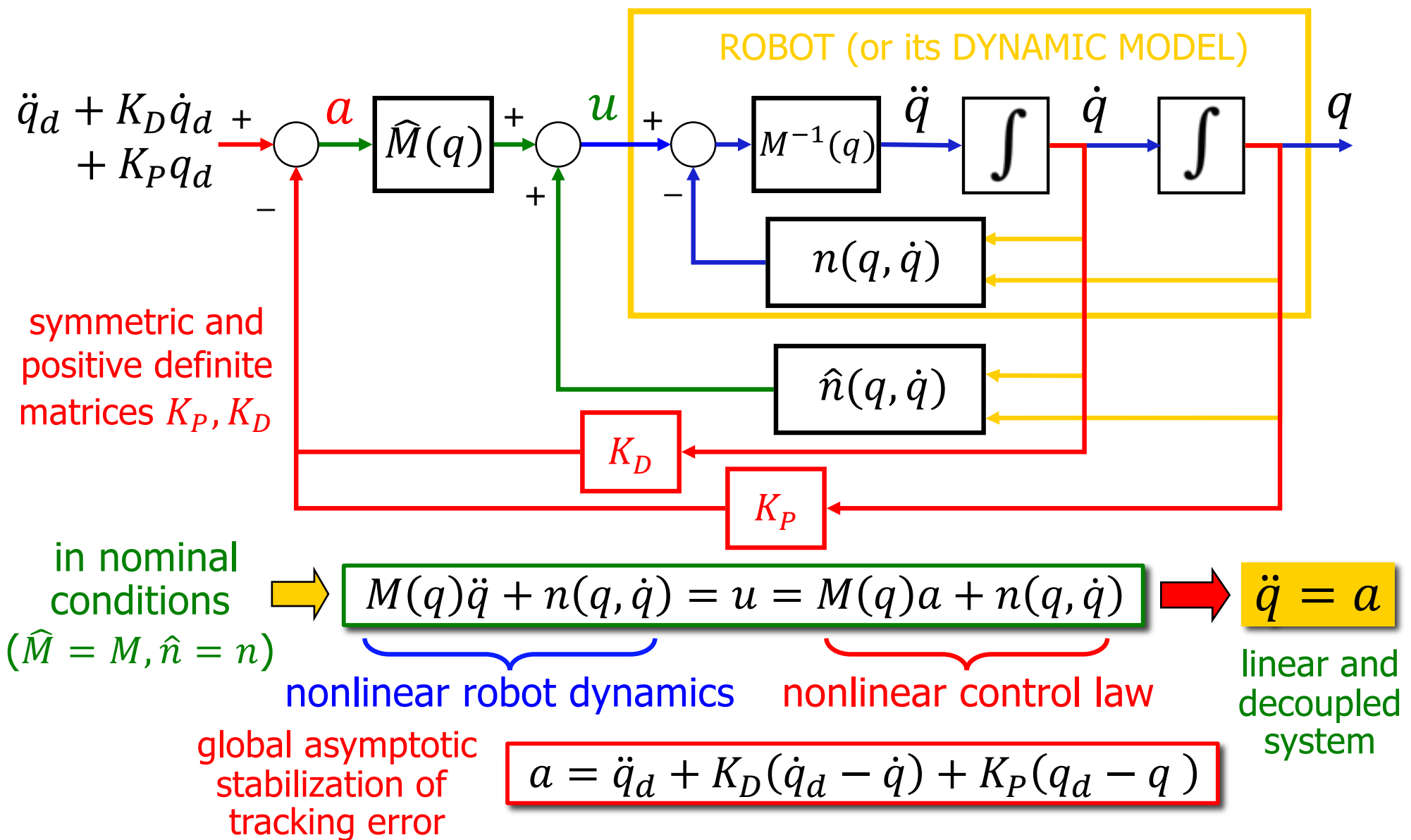
$$u = \hat{M}(q) \left[\ddot{q}_d + K_P(q_d - q) + K_D(\dot{q}_d - \dot{q}) + K_I \int (q_d - q) dt \right] + \hat{n}(q, \dot{q})$$

global stabilization for any $K_P > 0, K_D > 0$ (and not too large $K_I > 0$)

more robust to small uncertainties/disturbances, even if more complex to implement in real time

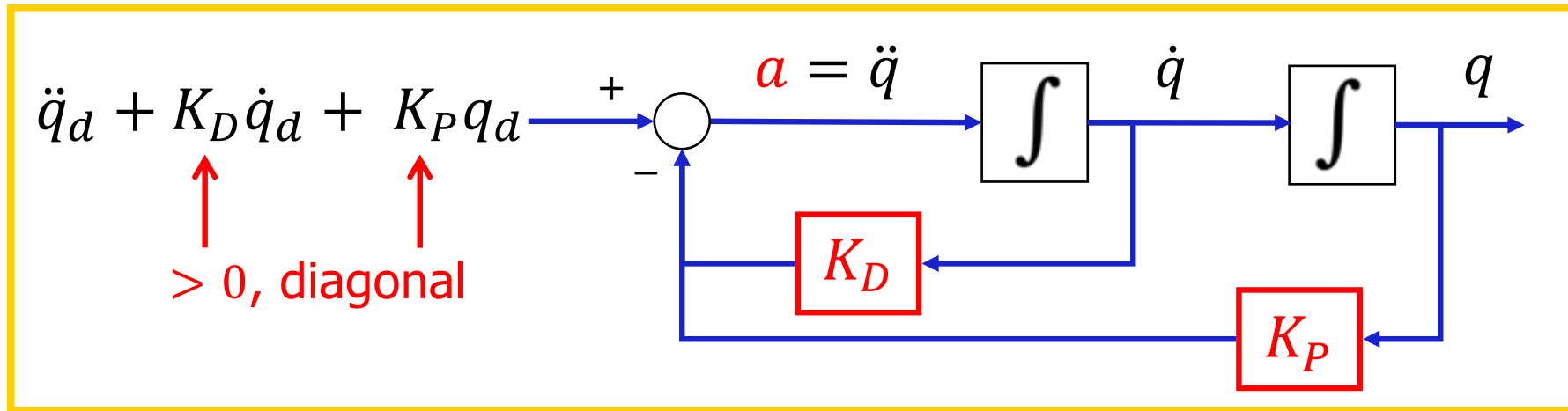


Feedback linearization control





Interpretation in the linear domain



under feedback linearization control, the robot has a dynamic behavior that is **invariant**, **linear** and **decoupled** in its whole state space ($\forall(q, \dot{q})$)

linearity

a unitary mass ($m = 1$) in the joint space !!

error transients $e_i = q_{di} - q_i \rightarrow 0$ **exponentially**, prescribed by K_{Pi}, K_{Di} choice

decoupling

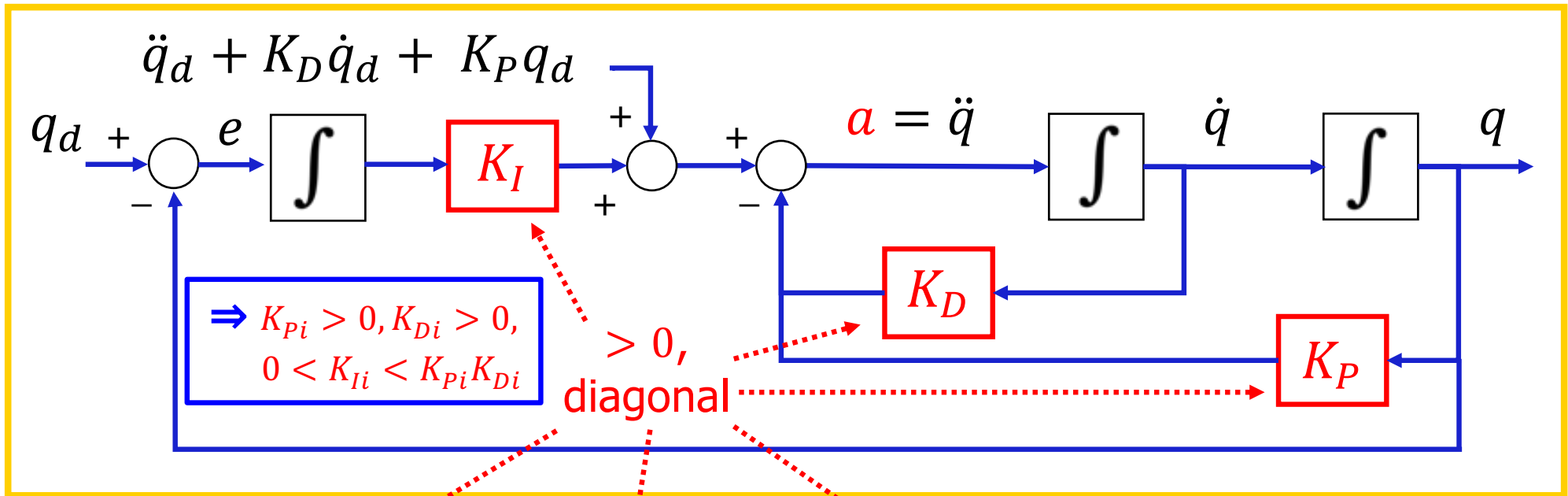
each joint coordinate q_i evolves **independently** from the others, forced by a_i

$$\ddot{e} + K_D \dot{e} + K_P e = 0 \iff \ddot{e}_i + K_{Di} \dot{e}_i + K_{Pi} e_i = 0$$



Addition of an integral term: PID

whiteboard...



$$\ddot{q} = a = \ddot{q}_d + K_D(\dot{q}_d - \dot{q}) + K_P(q_d - q) + K_I \int (q_d - q) d\tau \quad e = q_d - q$$

$$\Rightarrow (1) \quad e_i = q_{di} - q_i \quad (i = 1, \dots, N) \quad \Rightarrow (2) \quad \ddot{e}_i + K_{Di} \dot{e}_i + K_{Pi} e_i + K_{Pi} \int e_i d\tau = 0$$

$$\mathcal{L}[e_i(t)] \Rightarrow (3) \quad \left(s^2 + K_{Di} s + K_{Pi} + K_{Ii} \frac{1}{s} \right) e_i(s) = 0$$

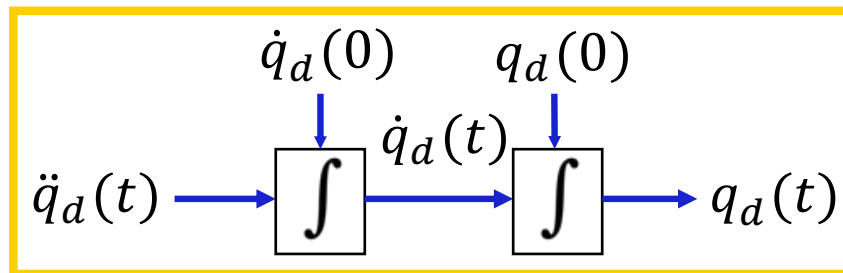
$$s \times \Rightarrow (4) \quad (s^3 + K_{Di} s^2 + K_{Pi} s + K_{Ii}) e_i(s) = 0 \quad \Rightarrow (5)$$

3	1	K_{Pi}
2	K_{Di}	K_{Ii}
1	$(K_{Di} K_{Pi} - K_{Ii}) / K_{Di}$	
0	K_{Ii}	

$\Rightarrow (6)$
exponential
stability
conditions by
Routh criterion

Remarks

- desired joint trajectory can be generated from Cartesian data



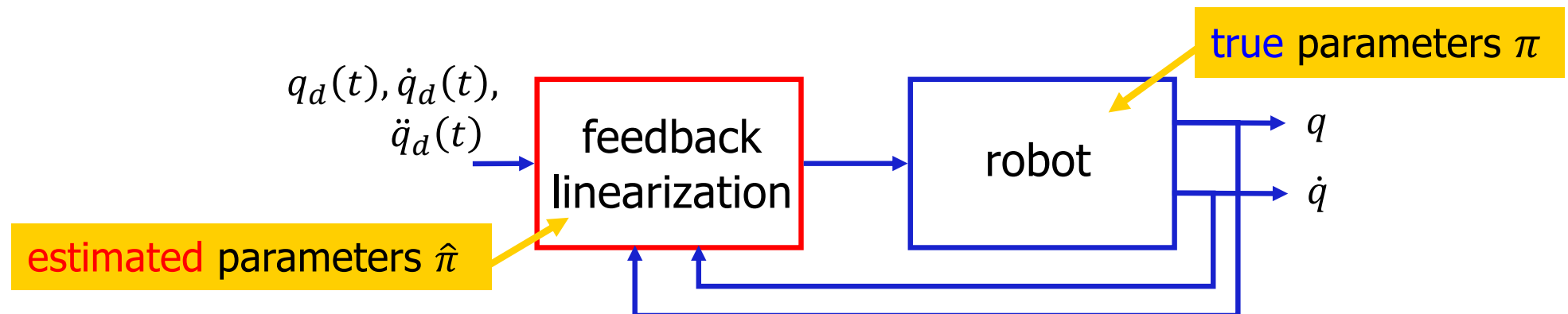
$$\ddot{p}_d(t), \dot{p}_d(0), p_d(0)$$

$$q_d(0) = f^{-1}(p_d(0))$$

$$\dot{q}_d(0) = J^{-1}(q_d(0))\dot{p}_d(0)$$

$$\ddot{q}_d(t) = J^{-1}(q_d)[\ddot{p}_d(t) - \dot{J}(q_d)\dot{q}_d]$$

- real-time computation by Newton-Euler algo: $u_{FBL} = \widehat{N}E_\alpha(q, \dot{q}, a)$
- simulation of feedback linearization control

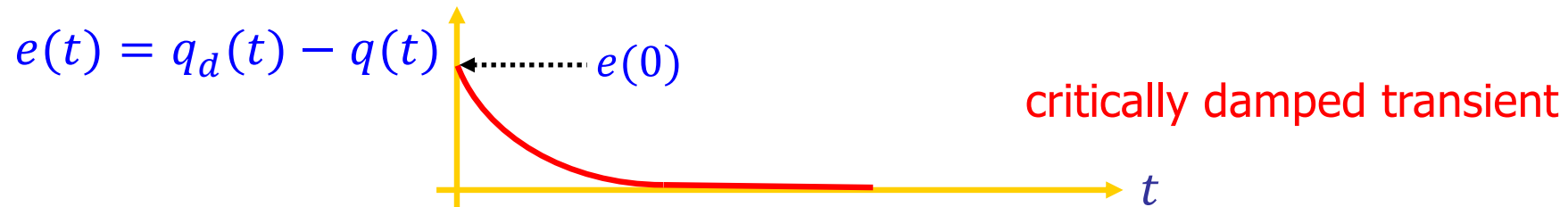


Hint: there is no use in simulating this control law in ideal case ($\hat{\pi} = \pi$); robot behavior will be identical to the linear and decoupled case of stabilized double integrators!!



Further comments

- **choice** of the diagonal elements of K_P , K_D (and K_I)
 - shaping the **error** transients, with an eye also to motor saturations...



- **parametric identification**
 - to be done in advance, using the property of **linearity** in the dynamic **coefficients** of the robot dynamic model
- choice of the **sampling time** of a digital implementation
 - compromise between **computational time** and **tracking accuracy**, typically $T_c = 0.5 \div 10$ ms
- exact linearization by (state) feedback is a general technique of **nonlinear control theory**
 - can be used for **robots with elastic joints, wheeled mobile robots, ...**
 - **non-robotics applications**: satellites, induction motors, helicopters, ...



Another example of feedback linearization design

- dynamic model of **robots with elastic joints**

- q = link position
 - θ = motor position (after reduction gears)
 - B_m = diagonal matrix (> 0) of inertia of the (balanced) motors
 - K = diagonal matrix (> 0) of (finite) stiffness of the joints
- } $2N$ generalized coordinates (q, θ)

$4N$ state variables
 $x = (q, \theta, \dot{q}, \dot{\theta})$

$$\begin{cases} M(q)\ddot{q} + c(q, \dot{q}) + g(q) + K(q - \theta) = 0 & (1) \\ B_m\ddot{\theta} + K(\theta - q) = u & (2) \end{cases}$$

- is there a control law that achieves **exact linearization via feedback**?

$$u = \alpha(q, \theta, \dot{q}, \dot{\theta}) + \beta(q, \theta, \dot{q}, \dot{\theta}) a$$

YES and it yields $\frac{d^4 q_i}{dt^4} = a_i, \quad i = 1, \dots, N$

linear and decoupled system:
 N chains of 4 integrators
(to be stabilized by linear control design)

Hint: differentiate (1) w.r.t. time until motor acceleration $\ddot{\theta}$ appears; substitute this from (2); choose u so as to cancel all nonlinearities ...



Alternative global trajectory controller

$$u = M(q)\ddot{q}_d + S(q, \dot{q})\dot{q}_d + g(q) + F_V\dot{q}_d + K_P e + K_D \dot{e}$$

↑
SPECIAL factorization such that
 $\dot{M} - 2S$ is skew-symmetric

↑ ↑
symmetric and
positive definite matrices

- global asymptotic stability of $(e, \dot{e}) = (0, 0)$ (trajectory tracking)
- proven by Lyapunov+Barbalat+LaSalle
- does not produce a complete cancellation of nonlinearities
 - the variables \dot{q} and \ddot{q} that appear linearly in the model are evaluated on the desired trajectory
- does not induce a linear and decoupled behavior of the trajectory error $e(t) = q_d(t) - q(t)$ in the closed-loop system
- however, it lends itself more easily to an adaptive version
- computation: by 4× standard or 1× modified NE algorithm

Analysis of asymptotic stability of the trajectory error - 1



$M(q)\ddot{q} + S(q, \dot{q})\dot{q} + g(q) + F_V\dot{q} = u$ robot dynamics (including friction)

control law $u = M(q)\ddot{q}_d + S(q, \dot{q})\dot{q}_d + g(q) + F_V\dot{q}_d + K_P e + K_D \dot{e}$

- Lyapunov candidate and its time derivative (with $e = q_d - q$)

$$V = \frac{1}{2} \dot{e}^T M(q) \dot{e} + \frac{1}{2} e^T K_P e \geq 0 \Rightarrow \dot{V} = \frac{1}{2} \dot{e}^T \dot{M}(q) \dot{e} + \underbrace{\dot{e}^T M(q) \ddot{e}} + e^T K_P \dot{e}$$

- the closed-loop system equations yield

$$M(q)\ddot{e} = -S(q, \dot{q})\dot{e} - (K_D + F_V)\dot{e} - K_P e$$

- substituting and using the skew-symmetric property of $\dot{M} - 2S$

$$\dot{V} = -\dot{e}^T (K_D + F_V) \dot{e} \leq 0 \quad \dot{V} = 0 \Leftrightarrow \dot{e} = 0$$

- since the system is time-varying (due to $q_d(t)$), direct application of LaSalle theorem is NOT allowed \Rightarrow use Barbalat lemma...

$$q = q_d(t) - e, \dot{q} = \dot{q}_d(t) - \dot{e} \Rightarrow V = V(\underbrace{e, \dot{e}}_{\text{error state } x}, t) = V(x, t)$$

\Rightarrow go to
slide 10 in
block 8

error state x



Analysis of asymptotic stability of the trajectory error - 2

- since i) V is lower bounded and ii) $\dot{V} \leq 0$, we have to check only condition iii) in order to apply **Barbalat lemma**

$$\ddot{V} = -2\dot{e}^T(K_D + F_V)\ddot{e} \quad \dots \text{is this bounded?}$$

- from i) + ii), V is bounded $\Rightarrow e$ and \dot{e} are bounded
 - assume that the desired trajectory has **bounded velocity** \dot{q}_d
- $\Rightarrow \dot{q}$ is bounded
- using the following **two properties** of dynamic model terms

$$0 < m \leq \|M^{-1}(q)\| \leq M < \infty \quad \|S(q, \dot{q})\| \leq \alpha_S \|\dot{q}\|$$

then also \ddot{e} will be **bounded** (in norm) since

$$\ddot{e} = -M^{-1}(q)[S(q, \dot{q})\dot{e} + K_P e + (K_D + F_V)\dot{e}]$$

$\begin{matrix} \uparrow & & \uparrow & \uparrow & \uparrow & & \uparrow \\ \text{bounded} & & \text{bounded} & \text{bounded} & \text{bounded} & & \text{bounded} \\ \text{in norm by } M & & \text{in norm by } \alpha_S \|\dot{q}\| & \leftarrow & \text{bounded} & & \end{matrix}$

$\Rightarrow \lim_{t \rightarrow \infty} \dot{V}(t) = 0$

Analysis of asymptotic stability of the trajectory error – end of proof



- we can conclude by proceeding as in **LaSalle theorem**

$$\dot{V} = 0 \iff \dot{e} = 0$$

- the closed-loop dynamics in this situation is

$$M(q)\ddot{e} = -K_p e$$

$$\Rightarrow \ddot{e} = 0 \iff e = 0 \quad \Rightarrow \quad (e, \dot{e}) = (0, 0)$$

is the largest
invariant set in $\dot{V} = 0$

\Rightarrow (global) asymptotic tracking
will be achieved





Regulation as a special case

- what happens to the control laws designed for trajectory tracking when q_d is **constant**? are there simplifications?
- **feedback linearization**

$$u = M(q)[K_P(q_d - q) - K_D\dot{q}] + c(q, \dot{q}) + g(q)$$

- no special simplifications
- however, this is a solution to the regulation problem with **exponential stability** (and decoupled transients at each joint!)

- **alternative global controller**

$$u = K_P(q_d - q) - K_D\dot{q} + g(q)$$

- we recover the PD + gravity cancellation control law!!



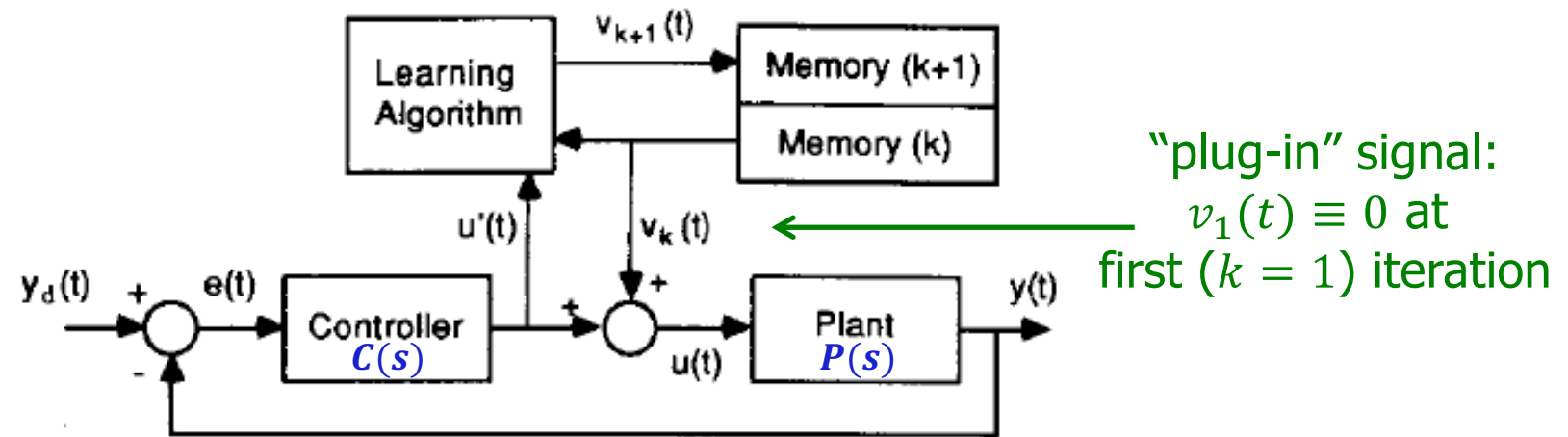
Trajectory execution without a model

- is it possible to accurately reproduce a desired smooth joint-space reference trajectory with **reduced or no information** on the robot dynamic model?
- this is feasible (and possibly simple) in case of **repetitive motion tasks** over a finite interval of time
 - trials are performed iteratively, storing the **trajectory error** information of the current execution [k -th iteration] and processing it off line before the next trial [$(k + 1)$ -iteration] starts
 - the robot should be **reinitialized** in the same initial state at the beginning of each trial (typically, with $\dot{q} = 0$)
 - the control law is made of a **non-model based part** (often, a decentralized PD law) + a **time-varying feedforward** which is updated before every trial
- this scheme is called **iterative trajectory learning**



Scheme of iterative trajectory learning

- control design can be illustrated on a **SISO linear system** in the Laplace domain



$$W(s) = \frac{y(s)}{y_d(s)} = \frac{P(s)C(s)}{1 + P(s)C(s)}$$

closed-loop system **without** learning
($C(s)$ is, e.g., a PD control law)

$$u_k(s) = u'_k(s) + v_k(s) = C(s)e_k(s) + v_k(s) \quad \text{control law at iteration } k$$

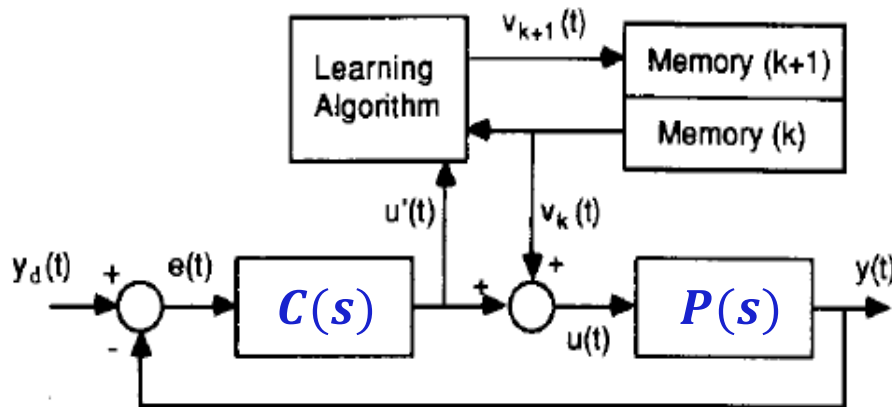
$$y_k(s) = W(s)y_d(s) + \frac{P(s)}{1 + P(s)C(s)} v_k(s) \quad \text{system output at iteration } k$$

Background math on feedback loops

whiteboard...



- **algebraic** manipulations on block diagram signals in the **Laplace** domain:
 $x(s) = \mathcal{L}[x(t)]$, $x = \{y_d, y, u', v, e\} \Rightarrow \{y_d, y_k, u'_k, v_k, e_k\}$, with transfer functions



$$\begin{aligned} y(s) &= P(s)u(s) = P(s)(v(s) + u'(s)) \\ &= P(s)v(s) + P(s)C(s)e(s) \\ &= P(s)v(s) + P(s)C(s)(y_d(s) - y(s)) \end{aligned}$$

$$\Rightarrow (1 + P(s)C(s)) y(s) = P(s)v(s) + P(s)C(s)y_d(s)$$

$$\Rightarrow y(s) = \frac{P(s)C(s)}{1 + P(s)C(s)} y_d(s) + \frac{P(s)}{1 + P(s)C(s)} v(s) = W(s)y_d(s) + W_v(s)v(s)$$

- **feedback control** law at iteration k

$$\begin{aligned} u'_k(s) &= C(s)(y_d(s) - y_k(s)) = C(s)y_d(s) - P(s)C(s)(v_k(s) + u'_k(s)) \\ \Rightarrow u'_k(s) &= \frac{C(s)}{1 + P(s)C(s)} y_d(s) - \frac{P(s)C(s)}{1 + P(s)C(s)} v_k(s) = W_c(s)y_d(s) - W(s)v_k(s) \end{aligned}$$

- **error** at iteration k

$$e_k(s) = y_d(s) - y_k(s) = y_d(s) - (W(s)y_d(s) + W_v(s)v_k(s)) = (1 - W(s))y_d(s) - W_v(s)v_k(s)$$

$$W_e(s) = 1/(1 + P(s)C(s))$$



Learning update law

- the **update** of the feedforward term is designed as

$$v_{k+1}(s) = \alpha(s)u'_k(s) + \beta(s)v_k(s) \quad \text{with } \alpha \text{ and } \beta \text{ suitable filters}$$

(also non-causal, of the FIR type)

recursive expression
of feedforward term

$$v_{k+1}(s) = \frac{\alpha(s)C(s)}{1 + P(s)C(s)}y_d(s) + (\beta(s) - \alpha(s)W(s))v_k(s)$$

recursive expression
of error $e = y_d - y$

$$e_{k+1}(s) = \frac{1 - \beta(s)}{1 + P(s)C(s)}y_d(s) + (\beta(s) - \alpha(s)W(s))e_k(s)$$

- if a **contraction condition** can be enforced

$$|\beta(s) - \alpha(s)W(s)| < 1 \quad \text{(for all } s = j\omega \text{ frequencies such that ...)}$$

then **convergence** is obtained for $k \rightarrow \infty$

$$v_\infty(s) = \frac{y_d(s)}{P(s)} \frac{\alpha(s)W(s)}{1 - \beta(s) + \alpha(s)W(s)} \quad e_\infty(s) = \frac{y_d(s)}{1 + P(s)C(s)} \frac{1 - \beta(s)}{1 - \beta(s) + \alpha(s)W(s)}$$



Proof of recursive updates

whiteboard...

- recursive expression for the **feedforward** v_k

$$\begin{aligned} v_{k+1}(s) &= \alpha(s)u'_k(s) + \beta(s)v_k(s) = \alpha(s)C(s)e_k(s) + \beta(s)v_k(s) \\ &= \alpha(s)C(s)[W_e(s)y_d(s) - W_v(s)v_k(s)] + \beta(s)v_k(s) \\ &= \frac{\alpha(s)C(s)}{1 + P(s)C(s)}y_d(s) + (\beta(s) - \alpha(s)W(s))v_k(s) \end{aligned}$$

- recursive expression for the **error** e_k

$$e_k(s) = y_d(s) - y_k(s) = y_d(s) - P(s)(v_k(s) + u'_k(s))$$

$$\Rightarrow v_k(s) = \frac{1}{P(s)}(y_d(s) - e_k(s)) - u'_k(s)$$



$$\begin{aligned} y_{k+1}(s) &= P(s)(v_{k+1}(s) + u'_{k+1}(s)) = P(s)(\alpha(s)u'_k(s) + \beta(s)v_k(s) + u'_{k+1}(s)) \\ &= P(s)\left(\alpha(s)C(s)e_k(s) + \beta(s)\frac{1}{P(s)}(y_d(s) - e_k(s)) - \beta(s)C(s)e_k(s) + C(s)e_{k+1}(s)\right) \end{aligned}$$

$$e_{k+1}(s) = y_d(s) - y_{k+1}(s)$$

$$= (1 - \beta(s))y_d(s) - [(\alpha(s) - \beta(s))P(s)C(s) - \beta(s)]e_k(s) - P(s)C(s)e_{k+1}(s)$$

$$\Rightarrow e_{k+1}(s) = \frac{1 - \beta(s)}{1 + P(s)C(s)}y_d(s) + (\beta(s) - \alpha(s)W(s))e_k(s)$$



Proof of convergence

whiteboard...

from recursive expressions

$$v_{k+1}(s) = \frac{\alpha(s)C(s)}{1 + P(s)C(s)} y_d(s) + (\beta(s) - \alpha(s)W(s)) v_k(s)$$

$$e_{k+1}(s) = \frac{1 - \beta(s)}{1 + P(s)C(s)} y_d(s) + (\beta(s) - \alpha(s)W(s)) e_k(s)$$

compute **variations** from k to $k + 1$ (repetitive term in trajectory y_d vanishes!)

$$\Delta v_{k+1}(s) = v_{k+1}(s) - v_k(s) = (\beta(s) - \alpha(s)W(s)) \Delta v_k(s)$$

$$\Delta e_{k+1}(s) = e_{k+1}(s) - e_k(s) = (\beta(s) - \alpha(s)W(s)) \Delta e_k(s)$$

by **contraction mapping** condition $|\beta(s) - \alpha(s)W(s)| < 1 \Rightarrow \{v_k\} \rightarrow v_\infty, \{e_k\} \rightarrow e_\infty$

$$v_\infty(s) = \frac{\alpha(s)C(s)}{1 + P(s)C(s)} y_d(s) + (\beta(s) - \alpha(s)W(s)) v_\infty(s)$$

$$e_\infty(s) = \frac{1 - \beta(s)}{1 + P(s)C(s)} y_d(s) + (\beta(s) - \alpha(s)W(s)) e_\infty(s)$$

$$\Rightarrow v_\infty(s) = \frac{y_d(s)}{P(s)} \frac{\alpha(s)W(s)}{1 - \beta(s) + \alpha(s)W(s)} \quad e_\infty(s) = \frac{y_d(s)}{1 + P(s)C(s)} \frac{1 - \beta(s)}{1 - \beta(s) + \alpha(s)W(s)}$$



Comments on convergence

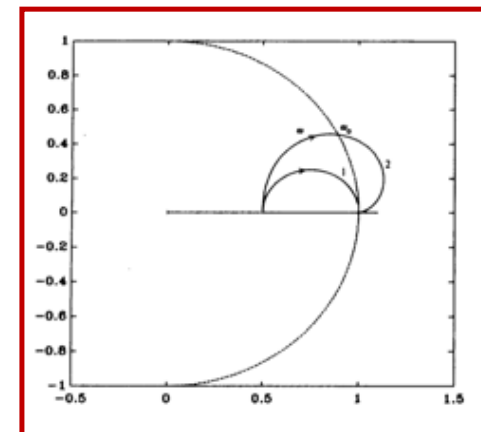
- if the choice $\beta = 1$ allows to satisfy the contraction condition, then convergence to **zero tracking error** is obtained

$$e_{\infty}(s) = 0$$

and the **inverse dynamics** command has been **learned**

$$v_{\infty}(s) = \frac{y_d(s)}{P(s)}$$

- in particular, for $\alpha(s) = 1/W(s)$ convergence would be in **1 iteration** only!
- the use of filter $\beta(s) \neq 1$ allows to obtain convergence (but with residual tracking error) even in presence of unmodeled high-frequency dynamics
- the **two filters** can be designed from very poor information on system dynamics, using classic tools (e.g., **Nyquist** plots)





Application to robots

- for N -dof robots modeled as

$$[B_m + M(q)]\ddot{q} + [F_V + S(q, \dot{q})]\dot{q} + g(q) = u$$

we choose as (initial = pre-learning) **control law**

$$u = u' = K_P(q_d - q) + K_D(\dot{q}_d - \dot{q}) + \hat{g}(q)$$

and design the **learning filters** (at each joint) using the **linear approximation**

$$W_i(s) = \frac{q_i(s)}{q_{di}(s)} = \frac{K_{Di}s + K_{Pi}}{\hat{B}_{mi}s^2 + (\hat{F}_{Vi} + K_{Di})s + K_{Pi}} \quad i = 1, \dots, N$$

- **initialization** of feedforward uses the best estimates

$$v_1 = [\hat{B}_m + \hat{M}(q_d)]\ddot{q}_d + [\hat{F}_V + \hat{S}(q_d, \dot{q}_d)]\dot{q}_d + \hat{g}(q_d)$$

or **simply** $v_1 = 0$ (in the worst case) at first trial $k = 1$

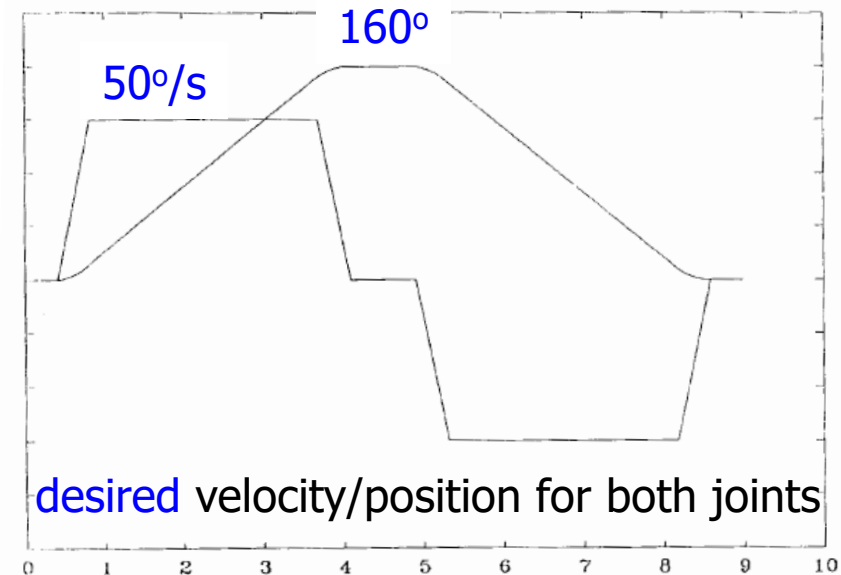
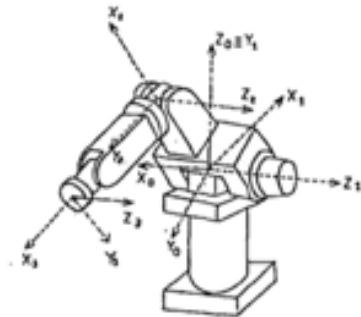
Experimental set-up

- joints 2 and 3 of 6R MIMO CRF robot prototype @DIS

≈ 90% gravity
balanced
through springs

high level of
dry friction

Harmonic Drives
transmissions
with ratio 160:1

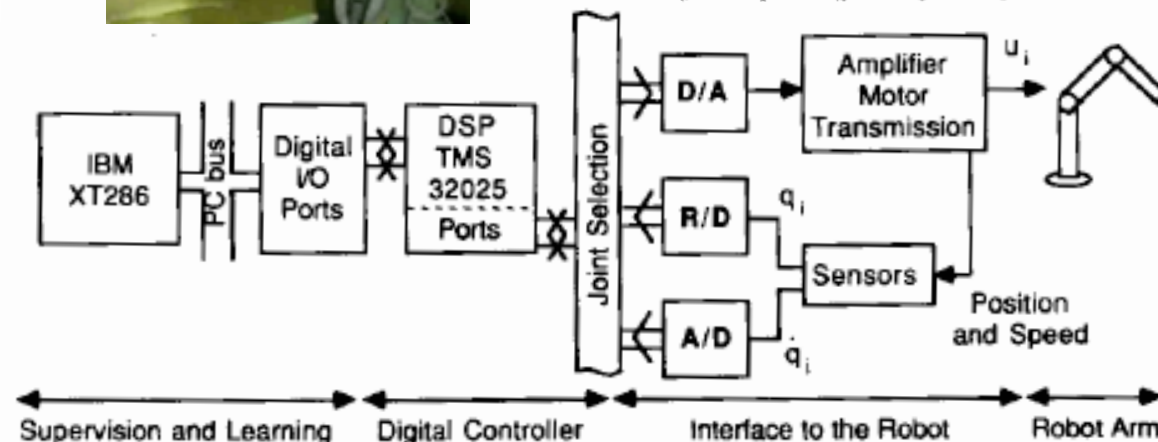


DSP $T_c = 400\mu s$

D/A = 12 bit

R/D = 16 bit/ 2π

A/D = 11 bit/(rad/s)



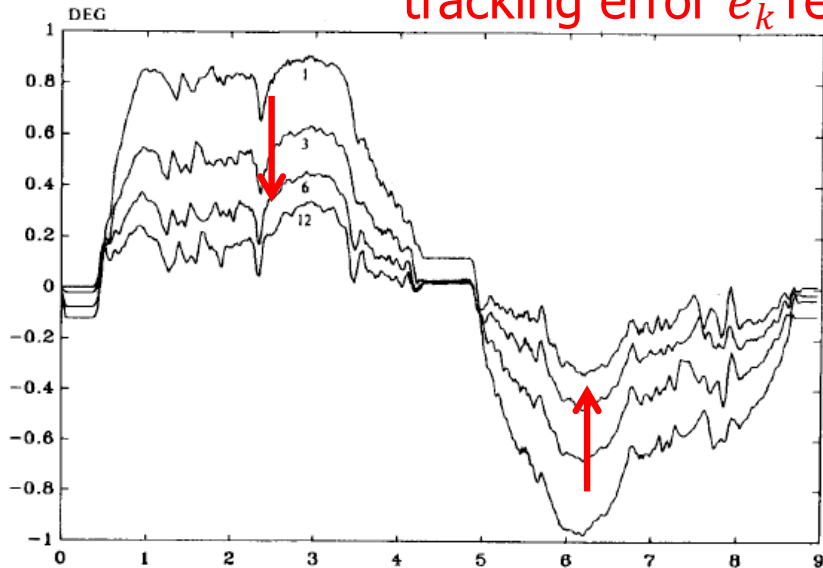
DC motors with
current amplifiers

resolvers and
tachometers

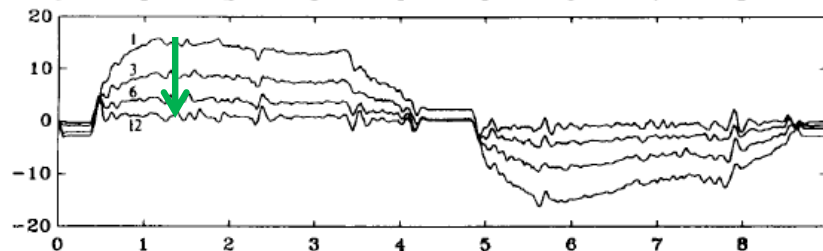
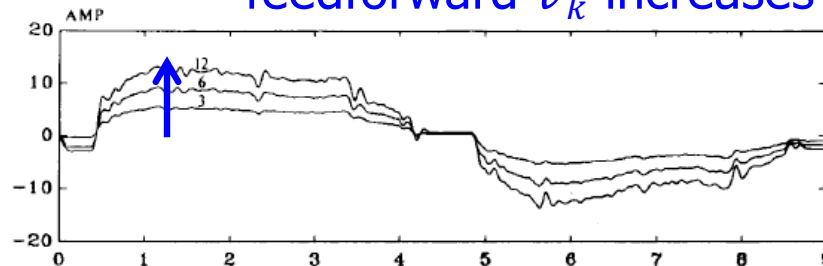
Experimental results

tracking error e_k reduces for $k = 1, 3, 6, 12$

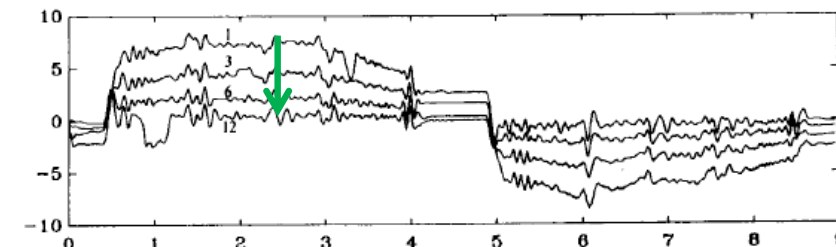
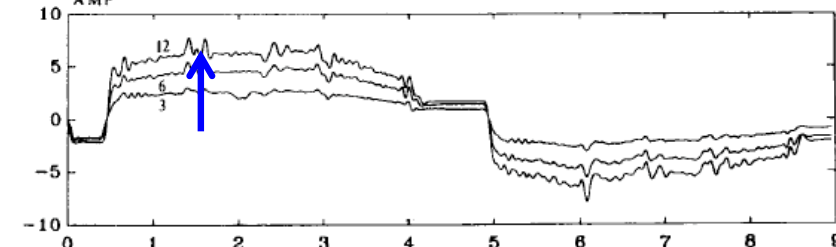
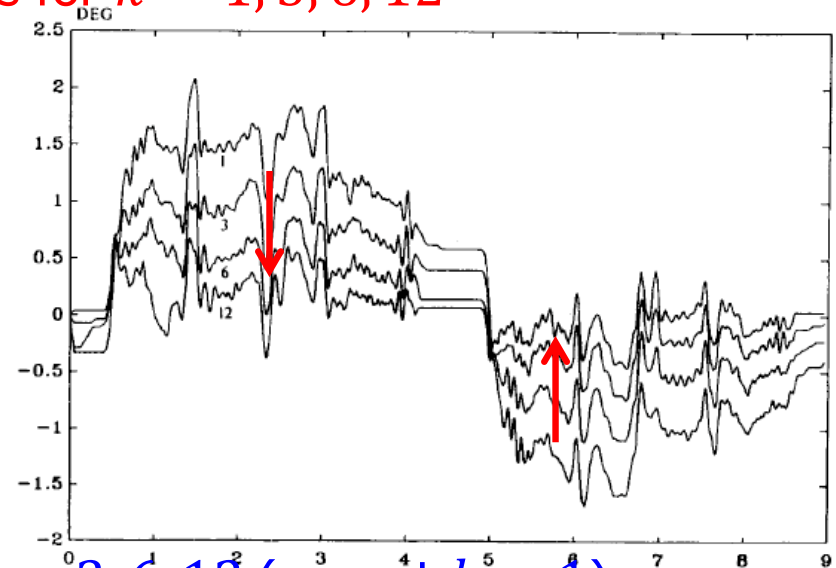
joint 2



feedforward v_k increases for $k = 3, 6, 12$ (zero at $k = 1$)



joint 3



feedback u'_k decreases for $k = 1, 3, 6, 12$



On-line learning control

- re-visitation of the learning idea so as to acquire the missing dynamic information in model-based trajectory control
- **on-line learning** approach
 - the robot improves tracking performance already while executing the task in feedback mode
- uses only position measurements from encoders
 - no need of joint torque sensors
- machine learning techniques used for
 - data collection and organization
 - regressor construction for estimating model perturbations
- **fast convergence**
 - starting with a reasonably good robot model
- extensions to underactuated robots or with flexible components



Control with approximate FBL

- dynamic model, its nominal part and (unstructured) uncertainty

$$M(q)\ddot{q} + n(q, \dot{q}) = \tau \quad M = \hat{M} + \Delta M \quad n = \hat{n} + \Delta n$$

- model-based (approximate) feedback linearization

$$\tau_{FBL} = \hat{M}(q)a + \hat{n}(q, \dot{q})$$

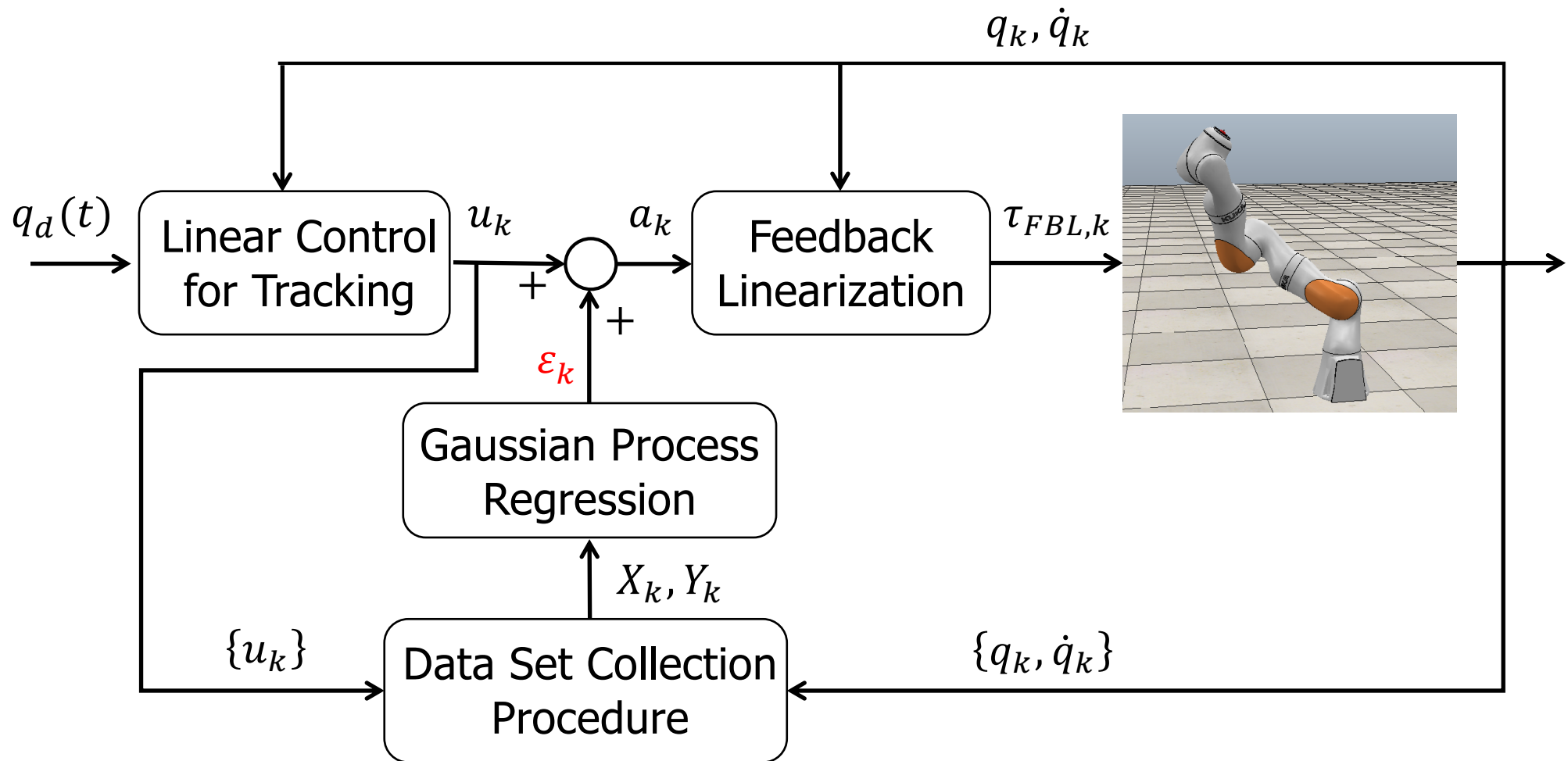
- resulting closed-loop dynamics with **perturbation**

$$\ddot{q} = a + \delta(q, \dot{q}, a) \quad \leftarrow \quad \delta = (M^{-1}\hat{M} - I)a + M^{-1}(\hat{n} - n)$$

- control law for tracking $q_d(t)$ is completed by using (at $t = t_k$) a linear design (PD with feedforward) and a learning **regressor** ε_k

$$\begin{aligned} a &= a_k = u_k + \varepsilon_k \\ &= \ddot{q}_{d,k} + K_P(q_{d,k} - q_k) + K_D(\dot{q}_{d,k} - \dot{q}_k) + \varepsilon_k \end{aligned}$$

On-line learning scheme





On-line regressor

- Gaussian Process (GP) regression to estimate the perturbation δ

- from input-output observations that are noisy, with $\omega \sim \mathcal{N}(0, \Sigma_\omega)$, the generated data points at the k -th control step are

$$X_k = (q_k, \dot{q}_k, u_k) \quad Y_k = \ddot{q}_k - u_k$$

- assuming the ensemble of n_d observations with a joint Gaussian distribution

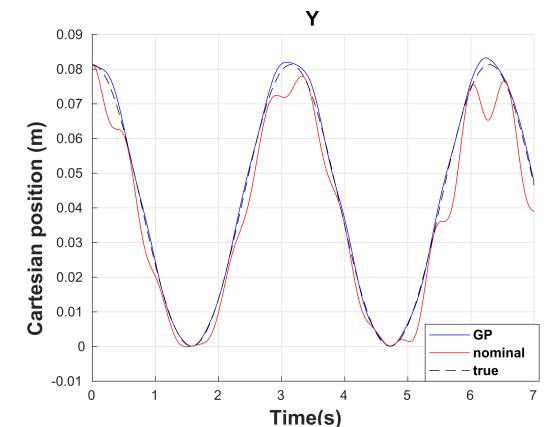
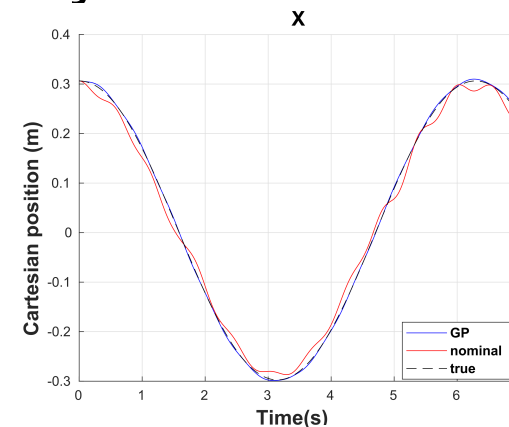
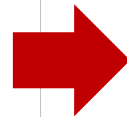
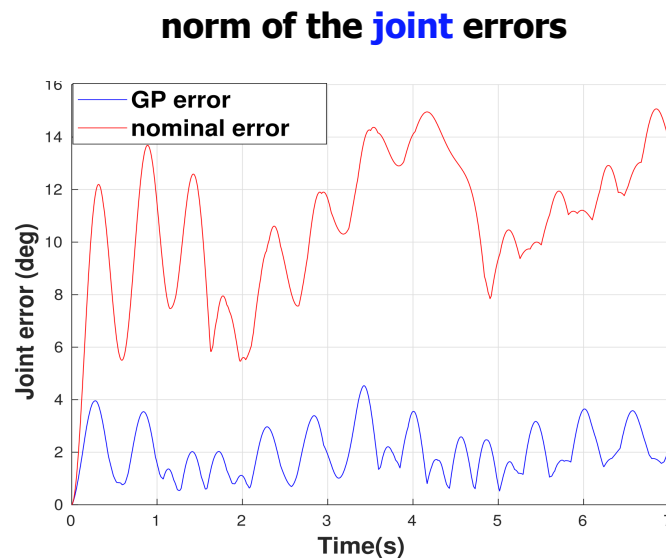
$$\begin{pmatrix} Y_{1:n_d-1} \\ Y_{n_d} \end{pmatrix} \sim \mathcal{N} \left(0, \begin{pmatrix} K & k \\ k^T & \kappa(X_{n_d}, X_{n_d}) \end{pmatrix} \right) \quad \begin{array}{l} \text{a Kernel} \\ \text{to be chosen} \end{array}$$

- the predictive distribution that approximates $\delta(\hat{X})$ for a generic query \hat{X} is

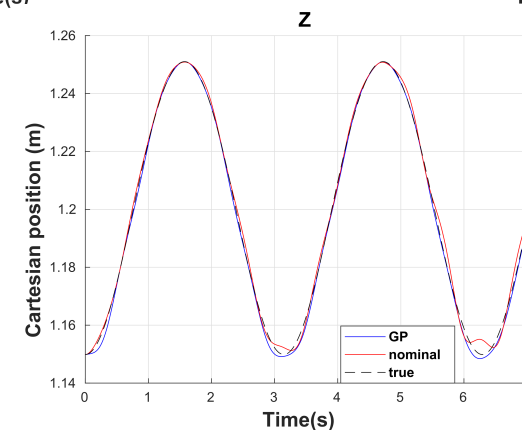
$$\left. \begin{array}{l} \varepsilon(\hat{X}) \sim \mathcal{N}(\mu(\hat{X}), \sigma^2(\hat{X})) \\ \text{with} \\ \mu(\hat{X}) = k^T(\hat{X})(K + \Sigma_\omega)^{-1}Y \\ \sigma^2(\hat{X}) = \kappa(\hat{X}, \hat{X}) - k^T(\hat{X})(K + \Sigma_\omega)^{-1}k(\hat{X}) \end{array} \right\} \Rightarrow \varepsilon_k = \varepsilon(X_k)$$

Simulation results

- Kuka LWR iiwa, 7-dof robot
- model perturbations: dynamic parameters with $\pm 20\%$ variation, uncompensated joint friction
- 7 separate GPs (one for each joint), each with 21 inputs at every $t = t_k$
- **sinusoidal** trajectories for each joint



**position
components
in the
Cartesian
space**



... at the **first and only** iteration!



Simulation results

video
(slowed
down)



An Online Learning Procedure for Feedback Linearization Control without Torque Measurements

M. Capotondi, G. Turrisi, C. Gaz, V. Modugno, G. Oriolo, A. De Luca

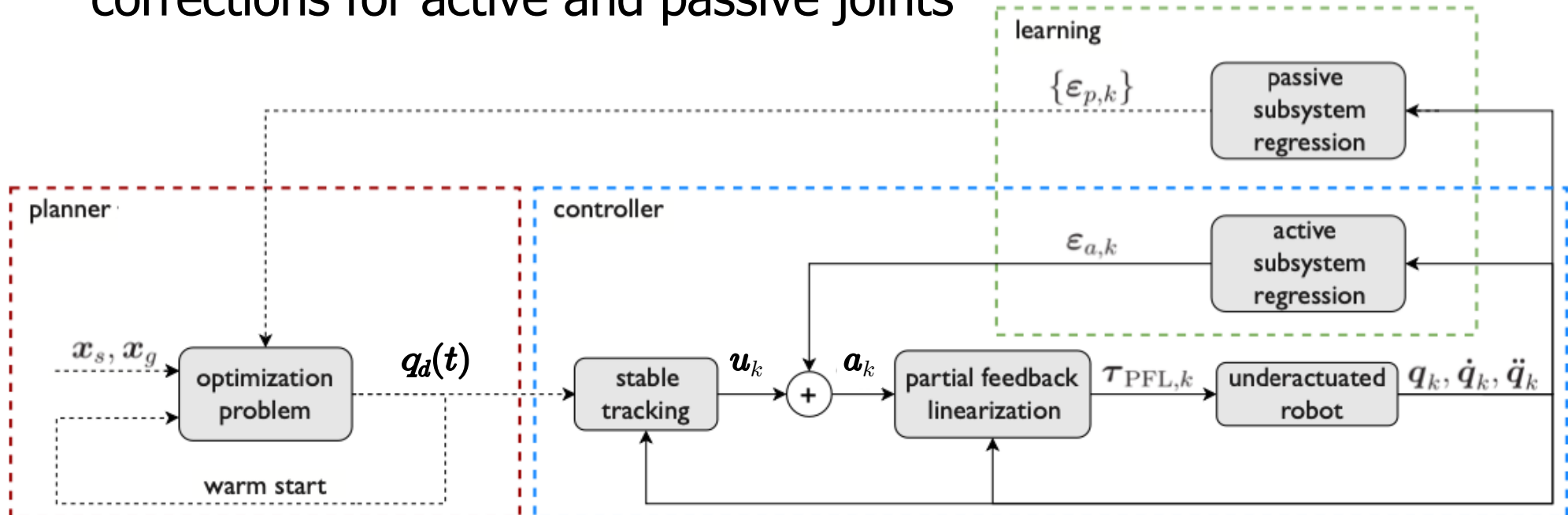
Robotics Lab, DIAG
Sapienza Università di Roma

October 2019

Extension to underactuated robots

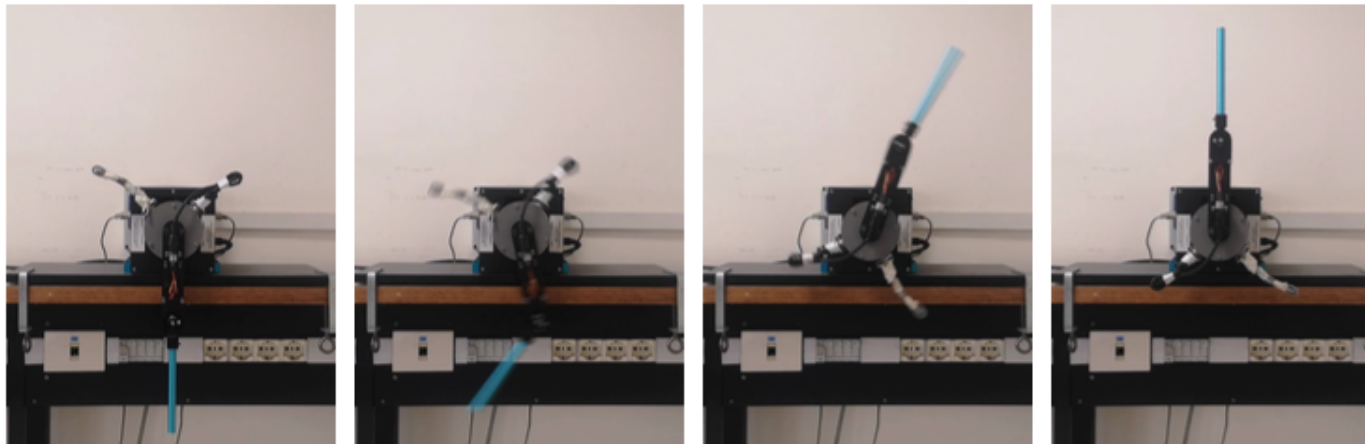
$$\begin{pmatrix} M_{aa}(q) & M_{ap}(q) \\ M_{ap}^T(q) & M_{pp}(q) \end{pmatrix} \begin{pmatrix} \ddot{q}_a \\ \ddot{q}_p \end{pmatrix} + \begin{pmatrix} n_a(q, \dot{q}) \\ n_p(q, \dot{q}) \end{pmatrix} = \begin{pmatrix} \tau \\ 0 \end{pmatrix}$$

- **planner** optimizes motion of **passive** joints (at every iteration)
- **controller** for **active** joints with **partial** feedback linearization
- **two regressors** (on/off-line) for learning the required acceleration corrections for active and passive joints

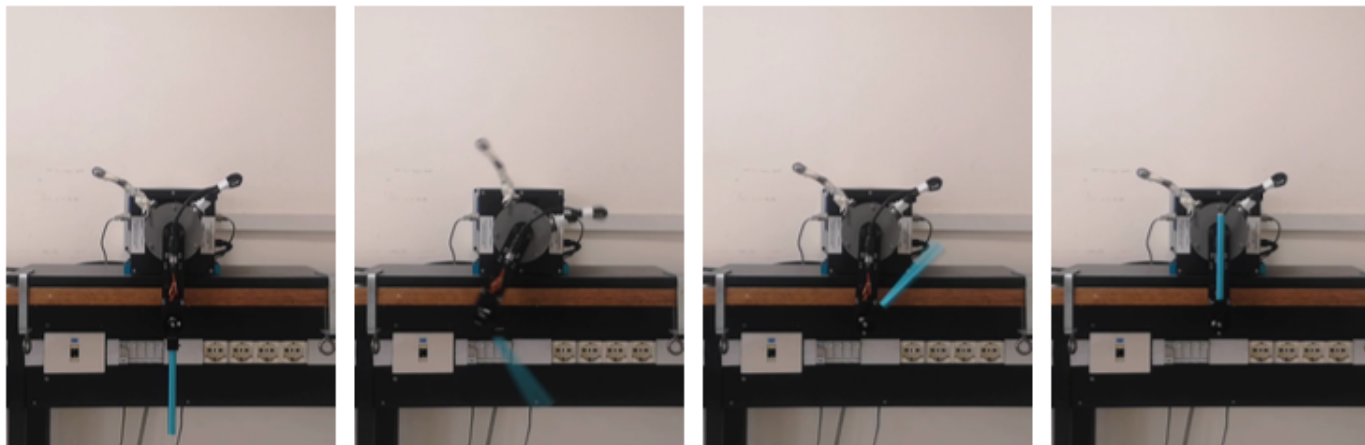


Experiments on the Pendubot

- Pendubot, 2-dof robot with passive second joint
- **swing-up** maneuvers from down-down to a new equilibrium state



⇒ up-up

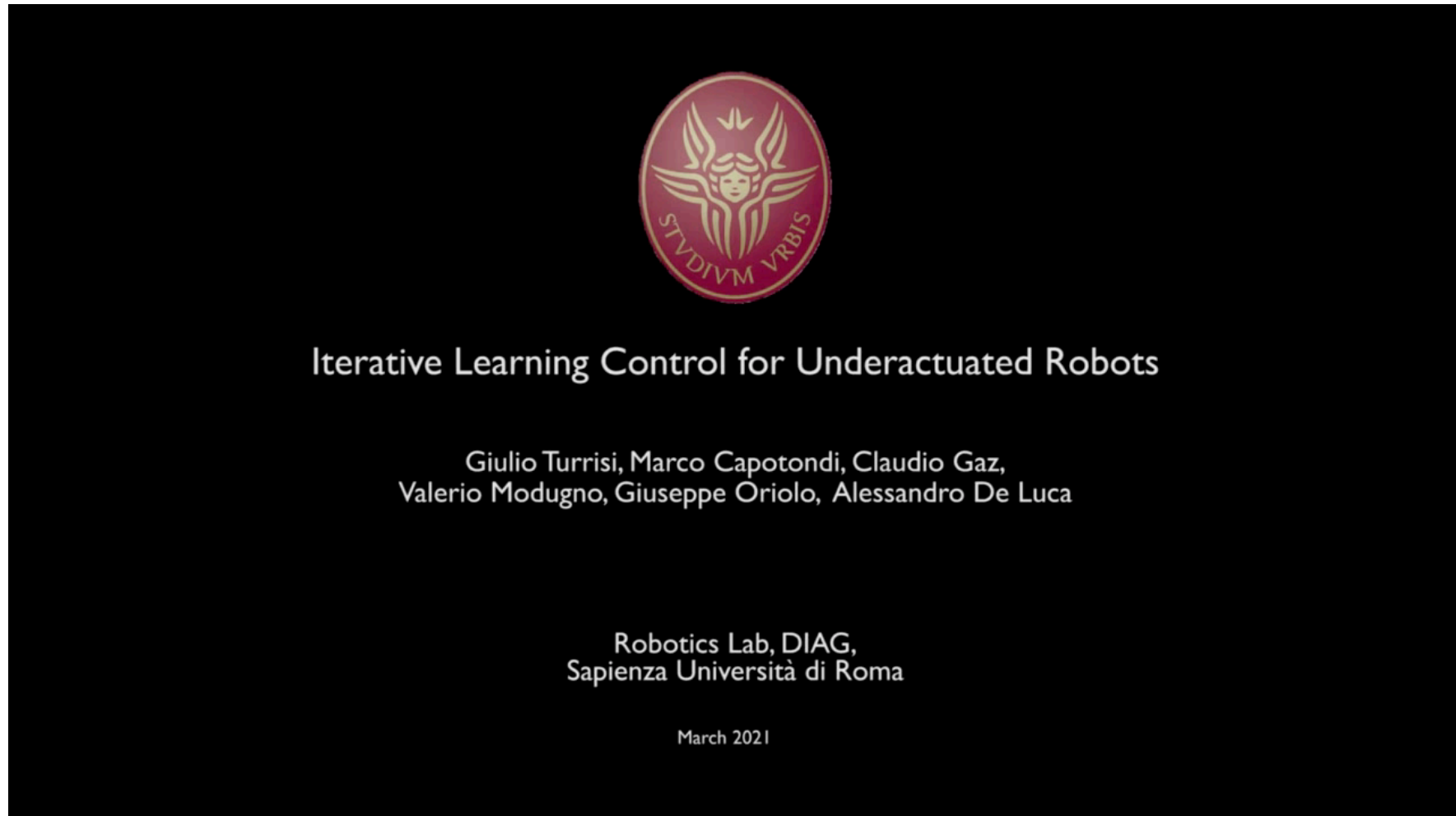


⇒ down-up



Experimental results

video



convergence in **2 iterations!**

latest video with more simulations & experiments
on YouTube https://youtu.be/1aKG_8gfvk