

## 1. Maxwell's Equations

- Gauss's Laws
- Faraday's Laws
- Ampere - Maxwell

## 2. Continuity

## 3. Wave

- Travelling wave
- EM wave
- Plane Polarised

## 4. Propagation

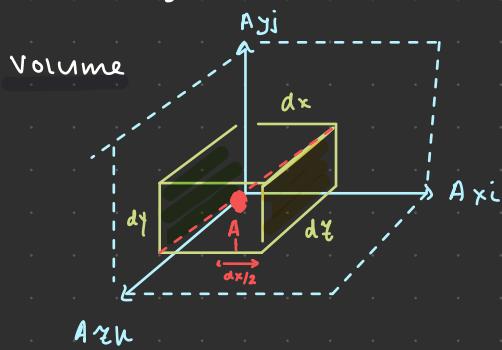
- Microscopic Ohm's law
- Conducting Medium

## Flux



- $d\phi$  of vector field (flux density) through infinitesimal surface  $d\vec{a}$  =  $d\phi = \underbrace{\vec{A} \cdot d\vec{a}}_{d\vec{a} \rightarrow \hat{n} d\vec{a}}$   $\xrightarrow{\hat{n}} \vec{A}$

$$\rightarrow \Phi = \int_S \vec{A} \cdot d\vec{a} \quad (\text{surface})$$



→ flux going in =  $-\hat{n}$   
out =  $\hat{n}$

$$d\phi_R = \left[ A_x + \frac{\partial A_x}{\partial x} \cdot \frac{dx}{2} \right] dx dy$$

$$d\phi_L = - \left[ A_x - \frac{\partial A_x}{\partial x} \cdot \frac{dx}{2} \right] dx dy$$

$$\left[ A_x + \frac{\partial A_x}{\partial x} \cdot \frac{dx}{2} \right] dx dy - \left[ A_x - \frac{\partial A_x}{\partial x} \cdot \frac{dx}{2} \right] dx dy$$

$$= d\phi = \frac{\partial A_x}{\partial x} \cdot dx dy dz$$

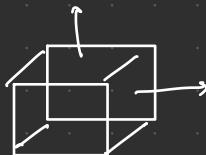
$$A_x = \frac{\partial A_x}{\partial x} \cdot dV \quad \left. \right\} \text{rate of change of field wrt } x$$

## Flux

$$A_y \Rightarrow \frac{\partial A_y}{\partial y} \cdot dV$$

Net Flux:  $d\Phi_{TOT} = \left( \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z} \right) dV$

$$\Phi = \int_V \left( \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z} \right) dV$$



Divergence :

Divergence  $\rightarrow$  how much flux generated at any given point in space?



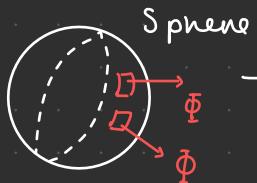
Positive  $\vec{\nabla}$   
↓  
source

$\searrow$   
Negative  $- \vec{\nabla}$   
↓  
sink

rate of change of field in direction of field

$$\vec{\nabla} \cdot \vec{A} = \left( \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z} \right)$$

## Gauss' Theorem / divergence

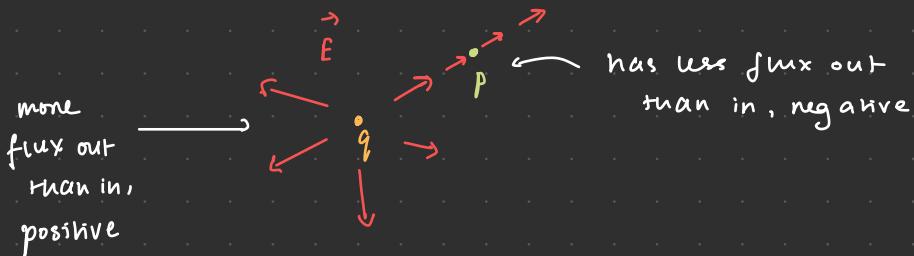


Total outward flux through closed surface area  $S$  = sum of normal components of  $\vec{A}$  through  $dxdydz$

$$\Phi = \oint_S \vec{A} \cdot d\vec{a} \longrightarrow \underbrace{\int_V \nabla \cdot \vec{A} dV}_{\text{Divergence of volume}}$$

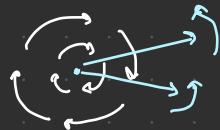
→ positive divergence  $\Rightarrow$  more field out than in

→ negative divergence  $\Rightarrow$  more field in than out

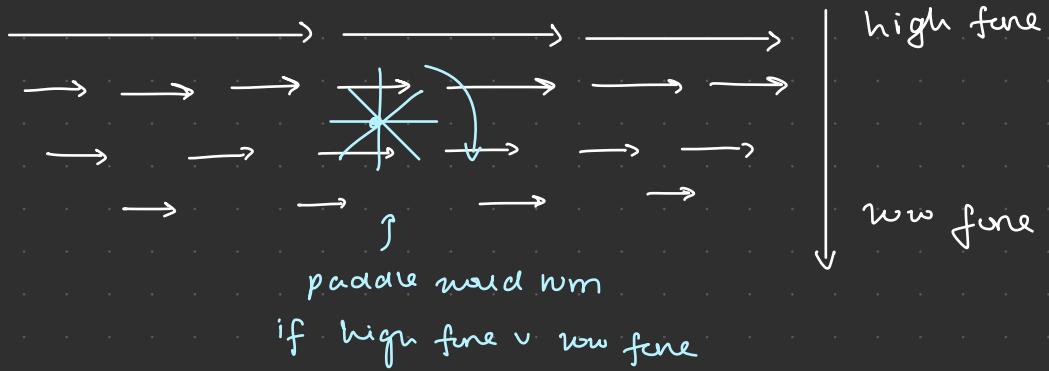


## Curl ( $\vec{\nabla} \times$ )

→ rate of change in vector field



→ circulating flow →



→ circulation is line integral of vector field along a path

$$\int_a^b \vec{A} \cdot d\vec{l}$$



→ if  $\oint \vec{A} \cdot d\vec{l}$ , closed path → conservative, line integral = 0

→ voltage fields are conservative

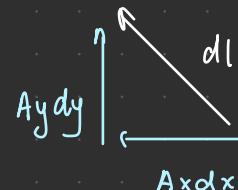
→ integrating along closed path →  $\left. \int_C \vec{A} \cdot d\vec{l} \right\}$  positive curl

$$\underline{\text{curl} (\vec{A} \times \vec{x})}$$



$$\oint \vec{A} \cdot d\vec{l}$$

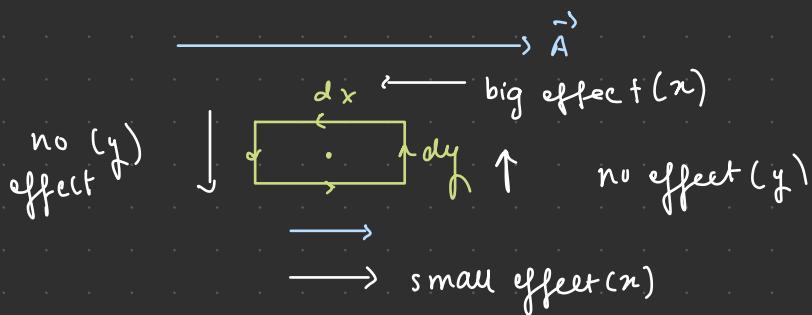
$\rightarrow d\vec{l}$  is infinitesimally small displacement along curve in space



$$\vec{A} \cdot d\vec{l} = A_x dx + A_y dy$$

$\rightarrow$  components of a closed loop integral:

$\hookrightarrow$  rate of change of vector field at  $90^\circ$  to axis



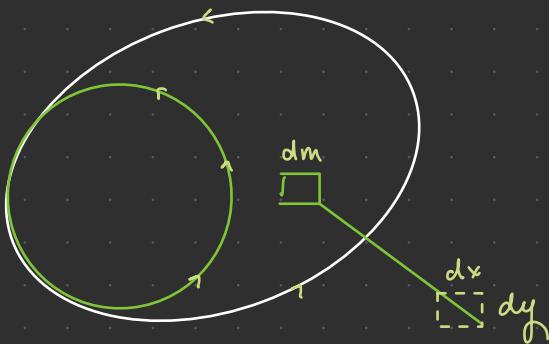
$$\oint \vec{A} \cdot d\vec{l} = \left[ \vec{i} \left( \frac{\partial A_z}{\partial y} - \frac{\partial A_y}{\partial z} \right) - \vec{j} \left( \frac{\partial A_z}{\partial x} - \frac{\partial A_x}{\partial z} \right) + \vec{k} \left( \frac{\partial A_y}{\partial x} - \frac{\partial A_x}{\partial y} \right) \right]$$



$$\oint \vec{A} \cdot d\vec{l} = (\vec{\nabla} \times \vec{A}) \cdot \vec{da}$$

## Stoke's Theorem

$$\oint \vec{A} \cdot d\vec{l} = (\vec{\nabla} \times \vec{A}) \cdot d\vec{a}$$



Laplacian  $\int \rightarrow$  divergence of gradient of scalar

$$\vec{\nabla} \cdot \vec{\nabla} \phi = \frac{\partial^2 \phi}{\partial x^2} + \frac{\partial^2 \phi}{\partial y^2} + \frac{\partial^2 \phi}{\partial z^2} = \left( \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2} \right) \phi$$



Laplacian of vector field  $\vec{A}$

$$\vec{\nabla} \cdot \vec{\nabla} \vec{A} = \vec{\nabla}^2 \vec{A} = \vec{\nabla}^2 A_x \vec{i} + \vec{\nabla}^2 A_y \vec{j} + \vec{\nabla}^2 A_z \vec{k}$$

Poisson's Equation:  $\vec{\nabla} \cdot \vec{\nabla} \cdot \vec{A} = \vec{\nabla}^2 \vec{A} = \vec{\nabla} (\vec{\nabla} \cdot \vec{A}) - \vec{\nabla}_x (\vec{\nabla} \times \vec{A})$

$$\vec{\nabla} \vec{A} = \begin{vmatrix} \frac{\partial A_x}{\partial x} & \frac{\partial A_x}{\partial y} & \frac{\partial A_x}{\partial z} \\ \frac{\partial A_y}{\partial x} & \frac{\partial A_y}{\partial y} & \frac{\partial A_y}{\partial z} \\ \frac{\partial A_z}{\partial x} & \frac{\partial A_z}{\partial y} & \frac{\partial A_z}{\partial z} \end{vmatrix}$$

Product Rules:

- vector that is gradient of a scalar  $\rightarrow$  no curl
- vector that is a curl of a scalar  $\rightarrow$  no divergence

# MAXWELL'S Equations

differential

↓  
in space

integral

↓  
surfaces, volumes

Parameter	Symbol	SI unit
Electric field	$\vec{E}(x, y, z)$	volt/m
Electric flux density	$\bar{D}(x, y, z)$	coulomb/m <sup>2</sup>
Displacement		Cm <sup>-2</sup>
Magnetic flux	$\Phi(x, y, z)$	weber = joule/A
Magnetic flux density	$\bar{B}(x, y, z)$	tesla = weber/m <sup>2</sup>
Magnetic inductance		T = NA <sup>-1</sup> m <sup>-1</sup>
Magnetic field	$\vec{H}(x, y, z)$	ampere/m = Am <sup>-1</sup>
Charge	$Q$ or $q$	coulombs
Current	$I$	ampere
Charge density	$\rho(x, y, z)$	coulomb/m <sup>3</sup>
Current density	$\bar{J}(x, y, z)$	ampere/m <sup>2</sup>
Electro motive force	Emf, V	volts
Vector Potential	$\vec{A}(x, y, z)$	definition
Scalar Potential	$\phi(x, y, z)$	volt

Parameter	Symbols	Free Space	SI Units
Permeability	$\mu = \mu, \mu_0$	$\mu_0 = 4\pi \times 10^{-7}$	henry/m N/A <sup>2</sup>
Permittivity	$\epsilon = \epsilon, \epsilon_0$	$\epsilon_0 = 10^{-9}/(36\pi)$	farad/m
Conductivity	$\gamma$ or $\sigma$	----	siemens/m (Ωm) <sup>-1</sup>
Light Speed	$c = 1/\sqrt{\mu_0 \epsilon_0}$	$3 \times 10^8$	m/s
Impedance	$\eta = \sqrt{\mu_0 / \epsilon_0}$	$120\pi = 377$	ohms, Ω

→ flux density = how field affects

the body it is acting on

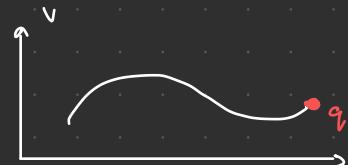
$$\star \vec{x} \cdot d\vec{a} \iff \vec{x} \cdot \hat{n} da$$

↳ unit vector (normal)

## Electric field

$$\vec{E}$$

$$\vec{F}(x, y, z) = Q \vec{E}(x, y, z)$$



→ charged particle moving around closed conservative loop

has no net force = 0

→ no work is done by field on the particle

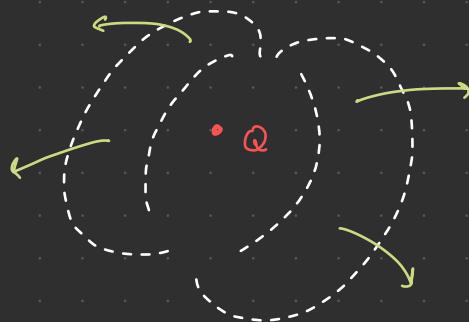
Recall Stoke's theorem:  $\oint \vec{E} \cdot d\vec{s} = \int (\vec{\nabla} \times \vec{E}) \cdot d\vec{a} = 0$

- closed loop, electrostatic field has no curl
- Electric field strength is measured at every point passing through the perpendiculars
- $\vec{E}$  field has no curl (closed loop)
- Stokes Theorem relates closed line integral to the surface



## Electric field

### Gauss' law for Static Electric field



→ charge in free space creates field lines

$$\epsilon \oint_S \vec{E} \cdot d\vec{a} = Q$$

$$\oint_D \vec{D} \cdot d\vec{a} = Q$$

Electric induction  $\iff \vec{D} = \epsilon \vec{E}$   $\leftarrow \vec{E}$  = electric field  
(unchanging)

$\vec{D}$  = amount of field induced on body

= flux per unit area

## Electric field $\rightarrow$ Gauss' law

↳ closed surface integral  $\Leftrightarrow$  volume integral

$$\oint_S \vec{A} \cdot d\vec{a} = \int_V \vec{\nabla} \cdot \vec{A} \cdot dV$$

$\rightarrow$  applying Gauss' Theorem:

\* integral form!  $\oint_S \vec{D} \cdot d\vec{a} = \int_V (\vec{\nabla} \cdot \vec{D}) \cdot dV = Q = \int_V p dV$



$\rightarrow p$  = charge density scalar function

$$p = \vec{\nabla} \cdot \vec{D}$$

$$\int_V \vec{\nabla} \cdot \vec{D} dV = \int_V p dV$$

## Electric field

- Divergence (how much field in/out) relates to how much charge

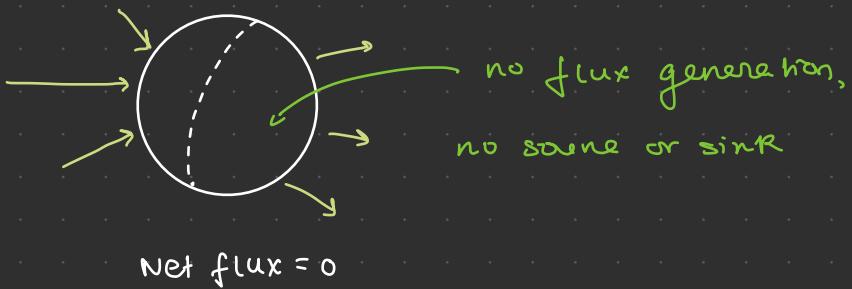
$$\vec{\nabla} \cdot \vec{E} = \frac{\rho}{\epsilon_0} \quad \curvearrowleft \text{charge density}$$

Point charge (charge $q$ C, distance $r$ m)	$\vec{E} = \frac{1}{4\pi\epsilon_0} \frac{q}{r^2} \hat{r}$
Conducting sphere (charge $Q$ C) <i>(perfect conductor)</i>	$\vec{E} = \frac{1}{4\pi\epsilon_0} \frac{Q}{r^2} \hat{r}$ (outside, $r$ m from centre) $\vec{E} = 0$ (inside)
Uniformly charged insulating sphere (radius $r_0$ m)	$\vec{E} = \frac{1}{4\pi\epsilon_0} \frac{Q}{r^2} \hat{r}$ (outside) $\vec{E} = \frac{1}{4\pi\epsilon_0} \frac{Qr}{r_0^3} \hat{r}$ (inside, $r$ m from centre)
Infinite line charge (linear charge density $\lambda$ C/m)	$\vec{E} = \frac{1}{2\pi\epsilon_0} \frac{\lambda}{r} \hat{r}$ ( $r$ m from line)
Infinite flat plane (surface charge density $\sigma$ C/m <sup>2</sup> )	$\vec{E} = \frac{\sigma}{2\epsilon_0} \hat{n}$

Magnetic field  $\vec{H}$

$\vec{H} \rightarrow$  field strength  $\rightarrow$  force at point  
 $\vec{B} \rightarrow$  flux density  $\rightarrow$  force on body due to  $\vec{H}$

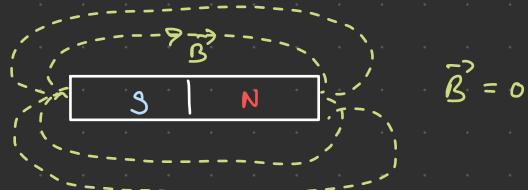
→ Total magnetic flux passing through closed surface = 0



→ Magnetic fields have no divergence

→ No point in space generates magnetic fields

Source  $(-\infty)$   $\longrightarrow$  Sink  $(\infty)$



→  $\mu_0 H_r = H$  (relative permittivity)

## Magnetic Fields

$$\oint_S \vec{B} \cdot \hat{n} da = 0$$

component of magnetic

field normal to surface

$\vec{B}$  = magnetic flux  
density

Differential form

$$\nabla \cdot \vec{B} = 0$$

divergence

## Faraday's Law

Electric current induced by time-varying magnetic field

→ by moving a magnet through closed loop

emf = force to move charge around loop

$$\text{EMF} = -\frac{d\Phi_m}{dt}$$
 ↪ induced electromotive force



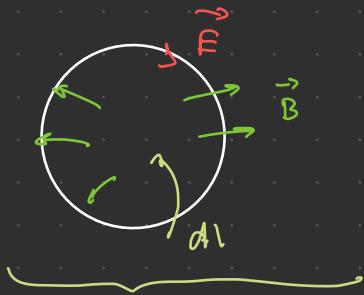
→ EMF would induce field +  $\nabla \psi$

work done in moving charge over a curve  $\oint_L \vec{E} \cdot d\vec{l}$

$$\oint \vec{E} \cdot d\vec{l} = -\frac{d\Phi_m}{dt}$$

→ Flux through surface =  $\int_S \vec{B} \cdot d\vec{s}$

## Magnetic field



$$\oint_C \vec{E} \cdot d\vec{l} = - \int \frac{d\Phi_M}{dt}$$

→ cause of change of flux

$$\frac{d \int_S \vec{B} \cdot d\vec{a}}{dt}$$

$$\oint_C \vec{E} \cdot d\vec{l} = - \frac{d}{dt} \int S \vec{B} \cdot d\vec{s} = - \int \frac{\partial \vec{B}}{\partial t} \cdot \hat{n} \cdot d\vec{a}$$

$$\boxed{\oint_C \vec{E} \cdot d\vec{l} = \int_S \vec{\nabla} \times \vec{E} \cdot d\vec{a}}$$

Lenz' law

$$= - \underbrace{\int_S \frac{\partial \vec{B}}{\partial t} \cdot d\vec{a}}_{\text{flux of M field, time}} \xrightarrow{\text{flux through S bounded by C}}$$

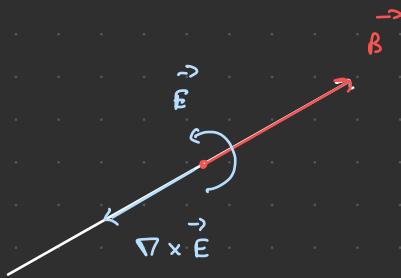
→ changing magnetic field creates circulating electric field

## Applying Stoule's Theorem

$$\oint \vec{E} \cdot d\vec{l} = \int_S \vec{\nabla} \times \vec{E} \cdot da = - \int_S \frac{d\vec{B}}{dt} \cdot da$$

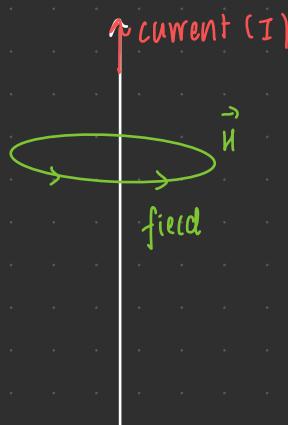
$$\rightarrow \vec{\nabla} \times \vec{E} = - \frac{\partial \vec{B}}{\partial t}$$

(because  $\vec{\nabla} \times \vec{E} = \mu \frac{\partial \vec{H}}{\partial t}$ )



## Ampere Maxwell's law

- changing electric field effect on magnetic field
- current inducing a magnetic field



$$\oint \vec{H} \cdot d\vec{l} = I$$

$$\vec{J}(x, y, z) = \text{current density } (\text{A/m}^2)$$

$$\int_S \vec{J}(x, y, z) \cdot da = I$$

$$\therefore \underbrace{\int \vec{J}(x, y, z) \cdot da}_{\text{current density}} = \underbrace{\oint \vec{H} \cdot dl}_{\text{magnetic field}} = I$$

## Ampere Maxwell's law

$$\oint_{l} \vec{H} \cdot d\vec{l} = \int_S \vec{J} \cdot d\vec{a}$$

Applying Stoke's Theorem  $\rightarrow$

$$\oint_C \vec{A} \cdot d\vec{l} = \int_S \nabla \times \vec{A} \cdot d\vec{s}$$

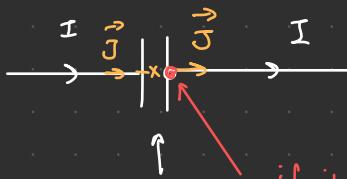
$$\int_S \nabla \times \vec{H} \cdot d\vec{a} = \int_S \vec{J} \cdot d\vec{a}$$

$$\nabla \times \vec{H} = \vec{J} \quad \rightarrow \text{curl of magnetic field} = \text{charge dens.}$$

$$\underbrace{\nabla \cdot (\nabla \times \vec{H})}_{\text{div of any curl}} = \nabla \cdot \vec{J} \quad \begin{aligned} &\rightarrow \text{take divergence of both sides} \\ &\rightarrow \text{div of any current } (\vec{J}) \neq 0 \end{aligned}$$

$\rightarrow$  i.e. capacitor

$\rightarrow$  there is a charge between 2 plates (called displacement current)

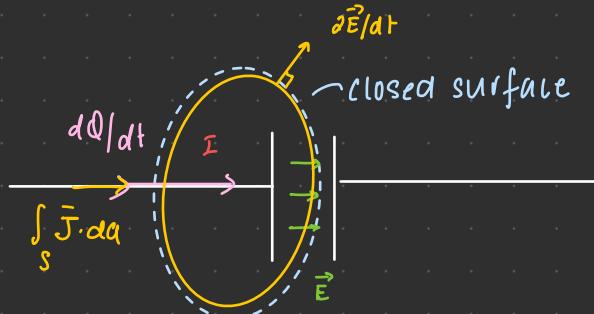


if it loops like  $\vec{J}$  is being generated, there is divergence

## Displacement Current

$$I = \frac{dQ}{dt}$$

→ current =  $\frac{\text{charge}}{\text{time}}$



- charge is moving in the volume
- when  $Q$  is increasing (recall - Gauss' law)

$$\Phi = \int_S \vec{D} \cdot d\vec{a} = Q$$

$$I = \frac{dQ}{dt} = \frac{d\Phi}{dt}$$

$$I = \frac{d \oint_S \vec{D} \cdot d\vec{a}}{dt} = \oint_S \left( \frac{d\vec{D}}{dt} \right) \cdot d\vec{a} = \int_S \vec{J} \cdot d\vec{a}$$

displacement  
current with current  
density

## Ampere - Maxwell

$$-\int \underbrace{\vec{J} \cdot d\vec{a}}_{\text{conduction (in)}} = \int \underbrace{\frac{d\vec{D}}{dt} \cdot d\vec{a}}_{\text{displacement (current out)}}$$

## Applying Gauss

$$\int_S \left( \vec{J} + \frac{d\vec{D}}{dt} \right) \cdot d\vec{a} = 0 \rightarrow \int_V \vec{\nabla} \cdot \left( \vec{J} + \frac{d\vec{D}}{dt} \right) \cdot d\vec{a} = 0$$



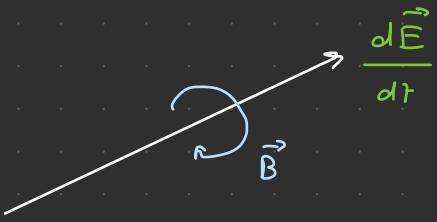
$$\vec{\nabla} \cdot \left( \vec{J} + \frac{d\vec{D}}{dt} \right) = 0$$

$$\text{Divergence} = 0$$

## Ampere - Maxwell

### Differential

$$\vec{\nabla} \times \vec{H} = \vec{J} + \frac{d\vec{D}}{dt}$$

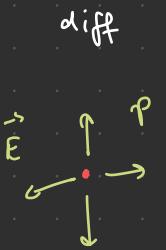


→ current → circling magnetic field

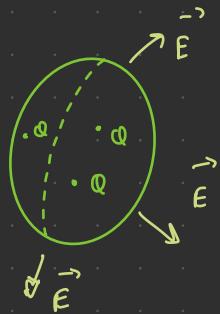
### Integral

$$\oint_L \vec{H} \cdot d\vec{l} = I_{\text{enc}} + \frac{d}{dt} \int_S \vec{D} \cdot d\vec{a}$$

Electric field ( $\vec{E}$ )



int



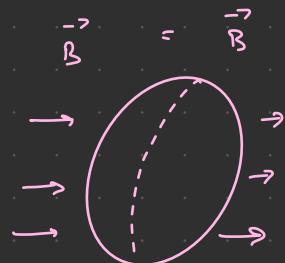
Magnetic field ( $\vec{B}$ )

diff

$$\vec{B} = \vec{B}$$

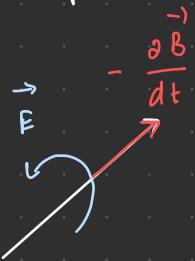
$$\nabla \cdot \vec{B} = 0$$

int

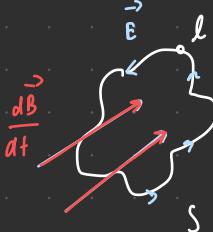


$$\vec{B} \longrightarrow \vec{E}$$

diff

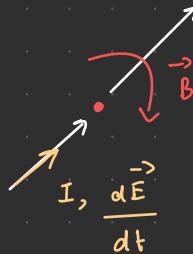


int

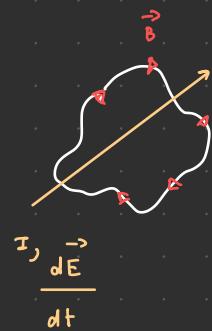


$$\vec{E}, I \longrightarrow \vec{B}$$

diff



int



→ magnetic field  
through a  
surface

## Continuity

→ conservation of charge

$$q(t) = \int_V p(t) \cdot dV$$



→ charge in =  $\int$  charge density

→ (current in - current out) through surface enclosing  
volume = net charge exiting body

$$\Rightarrow \oint_S \vec{J} \cdot d\vec{a}$$

→ rate of change of charge entering = net current entering

$$\frac{dq(t)}{dt} = - \oint_S \vec{J} \cdot d\vec{a}$$

Applying Gauss'  $\int_V \frac{dp(t)}{dt} \cdot dV = - \int_V \vec{\nabla} \cdot \vec{J} \cdot dV$  (volumes)

$$\frac{dp(t)}{dt} = - \vec{\nabla} \cdot \vec{J}$$

rate of  $\uparrow$  current  
charge density density

## Continuity Equation

$$\underbrace{\frac{dp(t)}{dt}}_{\text{rate of change of charge density}} + \vec{\nabla} \cdot \vec{J} = 0$$

rate of change  
of charge  
density

# Wave Equation

→ interacting with each other as waves

## Poisson's Equations

$$\underbrace{\vec{\nabla} \cdot \vec{\nabla} \vec{A}} = \vec{\nabla}^2 \vec{A} = \vec{\nabla}(\vec{\nabla} \cdot \vec{A}) - \vec{\nabla} \times (\vec{\nabla} \times \vec{A})$$

## Travelling wave

→ using function in space & time →  $s(x, t)$

→ describes function moving in time

$$\frac{\partial^2 s}{\partial x^2} = \frac{1}{v^2} \frac{\partial^2 s}{\partial t^2}$$

→ Given sin wave:  $\sin(x - wt)$

$$\frac{\partial^2 s}{\partial x^2} = -\sin(x - wt)$$

$$\frac{\partial^2 s}{\partial t^2} = -w^2 \sin(x - wt)$$

# Wave Equation

$$s(x, t) \longrightarrow F(x - vt) + G(x + vt)$$

+ ve                                    - ve

$F$  &  $G$  are arbitrarily differentiable waves

## Medium

- homogenous, isotropic
- insulator,  $\mu = \mu_0 \mu_r$ ,  $\epsilon = \epsilon_0 \epsilon_r$

## Maxwell's Eqns

2 eqns       $\left\{ \begin{array}{l} \vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} \\ \vec{\nabla} \times \vec{H} = \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t} \end{array} \right.$

2 unknowns      Faraday's Law      Ampere Maxwell

Take curl of Ampere's law:

$$\underbrace{\vec{\nabla} \times (\vec{\nabla} \times \vec{H})}_{\downarrow} = \epsilon \frac{\partial (\vec{\nabla} \times \vec{E})}{\partial t} \rightarrow \text{apply Poisson's law}$$

$$\frac{\partial \vec{J}}{\partial t} \Rightarrow 0$$

$$\vec{\nabla}(\vec{\nabla} \cdot \vec{H}) - \vec{\nabla}^2 H \quad * \text{ remember } \vec{\nabla} \cdot \vec{H} = 0 \text{ Always}$$

(Gauss's Law)

## Ampere's law

$$-\nabla^2 \vec{H} = \epsilon \frac{\partial (\vec{\nabla} \times \vec{E})}{\partial t}$$

from Faraday's Law  $\rightarrow \vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$

$$-\nabla^2 \vec{H} = -\epsilon \mu \frac{\partial \vec{H}}{\partial t} \cdot \frac{\partial}{\partial t}$$

$$\nabla^2 \vec{H} = \boxed{\epsilon \mu} \cdot \frac{\partial^2 \vec{H}}{\partial t^2}$$

Intrinsic Impedance  $\eta = \sqrt{\frac{\mu}{\epsilon}}$

In the wave equation  $\rightarrow \vec{\nabla}^2 \vec{E} = \boxed{\frac{1}{c^2}} \frac{\partial^2 \vec{E}}{\partial t^2}$

$$\therefore \epsilon \mu = \frac{1}{c^2}$$

$$\vec{\nabla}^2 \vec{H} = \boxed{\frac{1}{c^2}} \frac{\partial^2 \vec{H}}{\partial t^2}$$

$$\therefore c = \frac{1}{\sqrt{\epsilon \mu}}$$

- Steps:
1. Take curl of both sides
  2. Poisson's Law
  - 3.

## 3D Wave

$$\vec{\nabla}^2 \vec{E} = \frac{1}{c^2} \frac{\partial^2 \vec{E}}{\partial t^2}$$

$$\vec{\nabla}^2 \vec{H} = \frac{1}{c^2} \frac{\partial^2 \vec{H}}{\partial t^2}$$

Expand this using the Laplacian: (differentiate each direction)

$$H_E \frac{\partial^2 E_x}{\partial t^2} = \frac{\partial^2 E_x}{\partial x^2} + \frac{\partial^2 E_x}{\partial y^2} + \frac{\partial^2 E_x}{\partial z^2}$$

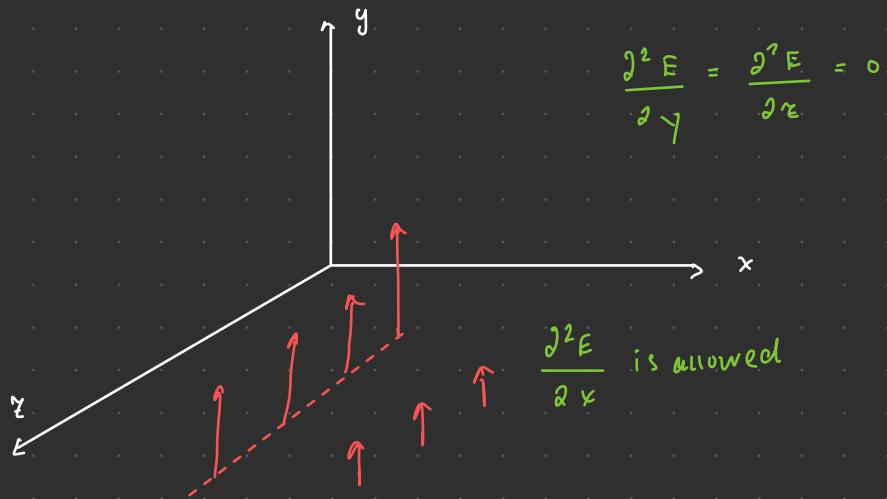
$$H_E \frac{\partial^2 E_y}{\partial t^2} = \frac{\partial^2 E_y}{\partial x^2} + \frac{\partial^2 E_y}{\partial y^2} + \frac{\partial^2 E_y}{\partial z^2}$$

$$H_E \frac{\partial^2 E_z}{\partial t^2} = \frac{\partial^2 E_z}{\partial x^2} + \frac{\partial^2 E_z}{\partial y^2} + \frac{\partial^2 E_z}{\partial z^2}$$

→ second derivatives in space → time

## Plane Polarised Wave

→ reduce 3D wave → 1D equivalent



→ only allow it to vary in one direction

$$\text{Polarisation} = E_x = E_z = 0$$

$$\text{Plane of wave} = \frac{\partial E_y}{\partial y} = \frac{\partial E_y}{\partial z} = 0$$

$$\frac{\partial^2 E_y}{\partial x^2} = \mu \epsilon \underbrace{\frac{\partial^2 E_y}{\partial t^2}}_{E_y} \rightarrow E_y \rightarrow F(x-vt) + G(x+vt)$$

$(F, G \neq \text{sinusoids}, \text{any } 2x \text{ differentiable functions})$

## Fields

Remember:

$$\vec{\nabla} \cdot \vec{D} = \rho$$

$$\vec{\nabla} \cdot \vec{B} = 0$$

$$\vec{\nabla} \times \vec{E} = - \frac{\partial \vec{B}}{\partial t}$$

$$\vec{\nabla} \times \vec{H} = \vec{J} + \frac{\vec{\partial D}}{\partial t}$$

(un)

$$\vec{\nabla} \times \vec{A} = \begin{vmatrix} \hat{i} & \hat{j} & \hat{k} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A_x & A_y & A_z \end{vmatrix} = \left( \frac{\partial A_z}{\partial x} - \frac{\partial A_y}{\partial z} \right) \hat{i} - \left( \frac{\partial A_z}{\partial x} - \frac{\partial A_x}{\partial z} \right) \hat{j} + \left( \frac{\partial A_y}{\partial x} - \frac{\partial A_x}{\partial y} \right) \hat{k}$$

## example

$$\vec{E} = 2b(x + y\varepsilon)t\hat{i} + 2zt\hat{j} + \alpha x^2\hat{k}$$

$$\vec{\nabla} \times \vec{E} = (0 - 2t)\hat{i} - (2ax - 2byt)\hat{j} + (0 - 2b\varepsilon t)\hat{k}$$

$$\vec{B} = t^2\hat{i} + (2axt - byt^2)\hat{j} + b\varepsilon t^2\hat{k}$$
$$-\frac{\partial \vec{B}}{\partial t} = -\left( (2t)\hat{i} + (2ax - 2byt)\hat{j} + 2b\varepsilon t\hat{k} \right)$$

Find values of  $\rho$   $\Rightarrow$  remember:  $\vec{\nabla} \cdot \vec{B} = 0$

$$\vec{\nabla} \cdot \vec{E} = 2bt\hat{i} + 0\hat{j} + 0\hat{k} = \frac{\rho}{\Sigma}$$

$$\therefore \rho = 2bt \cdot \varepsilon \quad (\text{holds for } \vec{\nabla} \cdot \vec{D} = \rho)$$

$\Rightarrow$  charge density =  $\rho$

Find value for current density  $\vec{J}$

$$\vec{J} = \nabla \times \vec{H} - \frac{\partial \vec{D}}{\partial t} \implies \frac{\nabla \times \vec{B}}{\mu} - \frac{\varepsilon \frac{\partial \vec{E}}{\partial t}}{\mu}$$

$$\nabla \times \vec{B} \quad (\text{curl}) = (0-0)\hat{i} - (0-0)\hat{j} + (2at - 0)\hat{k}$$

$$\left\{ \begin{array}{l} \vec{B} = t^2\hat{i} + (2axt - by\hat{t})\hat{j} + b\gamma t^2\hat{k} \\ \vec{E} = 2b(x+y\hat{z})t\hat{i} + 2\gamma t\hat{j} + ax^2\hat{k} \end{array} \right\}$$

$$\frac{\partial \vec{E}}{\partial t} = 2b(x+y\hat{z})\hat{i} + 2\gamma\hat{j} + 0\hat{k}$$

$$\vec{J} = \frac{2at\hat{k}}{\mu} - \varepsilon (2b(x+y\hat{z})\hat{i} + 2\gamma\hat{j})$$

$$\therefore -\varepsilon 2b(x+y\hat{z})\hat{i} - \varepsilon \cdot 2\gamma\hat{j} + \frac{2at\hat{k}}{\mu}$$

$$\text{Note that: } \vec{\nabla} \cdot \vec{j} + \frac{\partial p}{\partial t} = 0$$

$$p = 2bt + \varepsilon$$

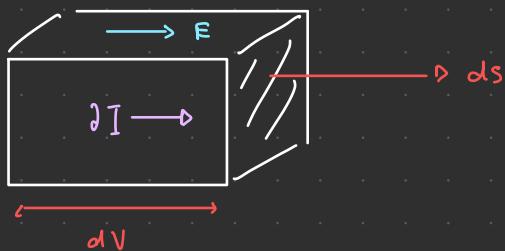
$$\vec{j} = -\varepsilon 2b(x + y\varepsilon) \hat{i} - \varepsilon \cdot 2\varepsilon \hat{j} + \frac{2at \hat{k}}{H}$$

$$\left. \begin{aligned} \vec{\nabla} \cdot \vec{j} &= -\varepsilon 2b \\ \frac{\partial p}{\partial t} &= 2b\varepsilon \end{aligned} \right\} \therefore -\varepsilon 2b + \varepsilon 2b = 0$$

## Propagation in a Conducting Medium

→ in free space  $\vec{H}$  and  $\vec{E}$  are in phase

### Microwave Ohm's Law



$$\text{conductance} = \frac{1}{R} = \sigma \left( \frac{\text{area}}{\text{length}} \right)$$

$$dI = \frac{1}{R} \cdot dV \quad \vec{J} \cdot \vec{ds} = \sigma \frac{|\vec{ds}|}{|\vec{dl}|} \vec{E} \cdot \vec{dl}$$

$$\vec{J} = \sigma \vec{E}$$

→ current density has to be in same direction as  $\vec{E}$  field

## Microwave Ohm's Law

$$\vec{J} = \sigma \vec{E}$$

- in steady state, lines of current don't terminate
- no accumulation of charge

→ i.e.  $\vec{\nabla} \cdot \vec{J} = 0$

$$\sigma \cdot \vec{\nabla} \cdot \vec{E} = 0$$



- single frequency, steady state
- e.g.  $E_y = E_m (\cos(\omega t - \beta x))$

# Analysis in Exponential Notation

---

$$E_y = E_m \cdot \cos(\omega t - \beta x)$$

→ write this as a real part of a complex function

$$\begin{aligned}
 E_y &= \operatorname{Real} \left( E_m e^{j(\omega t - \beta x)} \right) \\
 &= \operatorname{Real} \left( \underbrace{E_m e^{j\beta x}}_{=} \cdot \underbrace{e^{-j\omega t}}_{=} \right) \quad \text{at } E_0 = E_m e^{-j\beta x} \\
 &= \operatorname{Real} (E_0 e^{-j\omega t})
 \end{aligned}$$

## Complex form

$$\vec{E}(x, y, z, t) = \vec{E}_0(x, y, z) e^{j\omega t}$$

- Then:

$$\vec{E}(x, y, z, t) = (\alpha + j\beta)(\cos(\omega t) + j\sin(\omega t))$$

- The magnitude and phase of the signal is:

$$|\vec{E}(x, y, z, t)| = |\alpha + j\beta| |\cos(\omega t) + j\sin(\omega t)|$$

$$\angle \vec{E}(x, y, z, t) = \tan^{-1} \frac{\beta}{\alpha} + \tan^{-1} \frac{\sin \omega t}{\cos \omega t}$$

- The magnitude and phase of the signal is:

$$|\vec{E}(x, y, z, t)| = |\alpha + j\beta| |\cos(\omega t) + j\sin(\omega t)|$$

$$= \sqrt{\alpha^2 + \beta^2} \cdot 1$$

$$\angle \vec{E}(x, y, z, t) = \tan^{-1} \frac{\beta}{\alpha} + \tan^{-1} \frac{\sin \omega t}{\cos \omega t}$$

$$= \tan^{-1} \frac{\beta}{\alpha} + \omega t$$

# Analysis

$$\vec{E} = \vec{E}_0 e^{j\omega t} \longrightarrow \vec{H} = \vec{H}_0 e^{j\omega t}$$

$$\frac{\partial \vec{E}}{\partial t} = j\omega \vec{E}_0 e^{j\omega t} = j\omega \vec{E}$$

↳  $\frac{\partial}{\partial t} = j\omega$  rate of change over time  
of the field

- Analyse of how field effects occur in conducting medium  
i.e. current flow when charge is applied

$$\vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} \quad \left. \begin{array}{l} \\ \end{array} \right\} \quad \vec{\nabla} \times \vec{E}_0 = -j\omega \mu \vec{H}_0$$

$$\vec{\nabla} \times \vec{E}_0 e^{j\omega t} = -j\omega \mu \vec{H}_0 e^{-j\omega t} \quad \left. \begin{array}{l} \\ \end{array} \right\} \quad \text{divide}$$

$$\left. \begin{array}{l} \vec{\nabla} \times \vec{H} = \sigma \vec{E} + \epsilon \frac{\partial \vec{E}}{\partial t} \\ \vec{\nabla} \times \underbrace{\vec{H}_0 e^{j\omega t}}_{\text{complex}} = \sigma \vec{E}_0 e^{j\omega t} + j\omega \vec{E}_0 e^{j\omega t} \cdot \epsilon \end{array} \right\} \quad \vec{\nabla} \times \vec{H}_0 = (\sigma + j\omega \epsilon) \vec{E}_0$$

complex  
form of  
equations

## To Derive the wave Equation

→ conducting media has current density

$$\vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} \quad \longrightarrow \quad E \text{ field eqn} \rightarrow \mu \text{ (permittivity)}$$

$$\vec{\nabla} \times \vec{H} = \underbrace{G \vec{E}}_{\substack{\downarrow \\ J = G \vec{E}}} + \frac{\Sigma \partial \vec{E}}{\partial t} \quad \longrightarrow \quad H \text{ field has } \epsilon \text{ (permittivity)}$$

current density

## Space Equations

$$\vec{\nabla} \times \vec{E}_0 = -j\omega H \vec{H}_0$$

$$\vec{\nabla} \times \vec{H}_0 = (G + j\omega \epsilon) \vec{E}_0$$

Poisson's Equation :  $\underbrace{\nabla(\nabla \cdot \vec{E})}_{\substack{\parallel \\ 0}} - \nabla^2 \vec{E} = \vec{\nabla} \times \vec{\nabla} \times \vec{E}$

→ no divergence of an electric field

→ charge is always on the outside of a conducting medium

## Poisson's Equation

remember  $\rightarrow j\omega = \partial/dt$

- Solve these two simultaneous equations (Take the curl of both sides):  

$$\vec{\nabla} \times \vec{\nabla} \times \vec{E}_o = -j\omega\mu(\vec{\nabla} \times \vec{H}_o)$$
- Using poisons equation, and substitution for  $\vec{\nabla} \times \vec{H}_o$ :  

$$-\vec{\nabla}^2 \vec{E}_o = -j\omega\mu(\sigma + j\omega\epsilon)\vec{E}_o$$

- We are left with the equation representing the propagating E field in a conducting medium:

$$\vec{\nabla}^2 \vec{E}_o = j\omega\mu(\sigma + j\omega\epsilon)\vec{E}_o$$

- Let  $\sqrt{j\omega\mu(\sigma + j\omega\epsilon)} = \gamma$ , to get:

$$\vec{\nabla}^2 \vec{E}_o = \gamma^2 \vec{E}_o$$

- wave Equation

- $\gamma$  is complex and is called the **Propagation Constant**: ( $\gamma = \alpha + j\beta$ )  
 $\alpha$  = attenuation constant  
 $\beta$  = Phase constant

wave equation  $\rightarrow$

$$\underbrace{\vec{\nabla}^2 \vec{E}_o}_{\text{space derivative}} = \underbrace{\gamma^2 \vec{E}_o}_{\text{time derivative}}$$

$$\vec{E} \Leftrightarrow \vec{H}$$

$\rightarrow$  connecting the electric & magnetic field using the wave Egn.

## Plane Polarised Wave Analysis

free space

$$\left\{ \begin{array}{l} E_x = E_z = 0 \\ \frac{\partial E_y}{\partial y} = \frac{\partial E_y}{\partial z} = 0 \end{array} \right.$$

In a conducting medium, field propagation:

$$\frac{\partial^2 E_{0y}}{\partial x^2} = \gamma^2 E_{0y}$$

$\gamma$  = second order derivation

Plane polarised wave ( $E_{0y} = E_m e^{\pm jyx}$   $\rightarrow \epsilon_m$  is a constant)

$$\frac{\partial^2 E_m e^{\pm jyx}}{\partial x^2} = \gamma^2 E_m e^{\pm jyx}$$

use complex function  $\vec{E} = \vec{E}_0 j\omega t$

$$\vec{E} = E_0 j\omega t$$

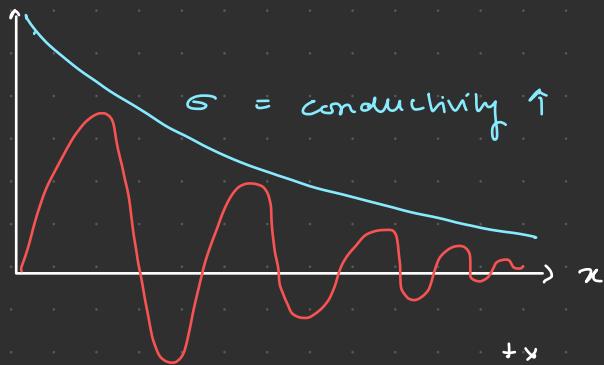
## Propagating field

$$E_y = E_m e^{j\omega t} e^{-\alpha x} e^{\pm j\beta x}$$

sinusoidal variation

attenuation

phase shift



→ Exponential attenuation  $\propto$  alpha ( $\alpha$ )

→  $\alpha \rightarrow \gamma = \mu \epsilon \sigma$

→ Gamma ( $\gamma$ ) = propagation constant

if  $\epsilon \uparrow, \alpha \uparrow, \beta \uparrow$

To find  $\vec{H}$  in a conducting medium

$$E_y = E_m e^{j\omega t} e^{-\alpha x} e^{-j\beta x}$$

$$E_{oy} = E_m e^{-\delta x} \quad \text{if } \gamma = \text{prop constant}$$

Maxwell's Equation:  $\vec{\nabla} \times \vec{E}_o = -j\omega \mu H_o$

for a 1-D plane polarised wave:  $\frac{\partial E_{oy}}{\partial x} \vec{k} = -j\omega \mu H_o$

↳ this means  $H_o$

only has components  
in the  $\hat{x}$ -direction

Plane  
Polarised  
Wave Eqns

$$E_z = E_x = 0$$

$$\frac{\partial E_y}{\partial y} = \frac{\partial E_y}{\partial z} = 0$$

$$\frac{\partial E_{oy}}{\partial x} \hat{n} = -j\omega \mu H_o = -\delta E_m e^{-\delta x} = -\delta E_{oy} \quad (\text{1d plane polarised})$$

$$\therefore \frac{H_o \hat{x}}{E_{oy}} = \frac{j\omega H}{\gamma} \quad \left\{ \begin{array}{l} \cdot \text{ connected by a complex term} \\ \cdot E \text{ and } H \text{ are not in phase} \end{array} \right.$$

$$\eta = \sqrt{\frac{E_o}{\mu_0}} \quad \text{in free space} \longrightarrow \eta = \frac{j\omega H}{\gamma} = \frac{j\omega H}{\sqrt{j\omega \mu (\epsilon + j\omega \epsilon)}}$$

$$\eta = \sqrt{\frac{j\omega H}{\epsilon + j\omega \epsilon}}$$

(in a medium)

## Propagation of $\vec{H}$ in conducting medium

( same as last page)  $\rightarrow$

$$\frac{\partial E_{oy}}{\partial x} = -j\omega \mu H_0 = -\underbrace{\sigma E_m c}_{\downarrow} \cancel{-j\sigma x} = -\gamma E_{oy}$$

depends on  $\sigma$  (prop. constant)

$\downarrow$   
depends on conductivity  
of the material

$$\rightarrow \gamma = \underbrace{\frac{j\omega H}{\gamma}}_{\text{gamma}} = \frac{j\omega H}{\sqrt{j\omega \mu (\epsilon + j\omega \epsilon)}}$$

## What is Propagation Constant $\gamma$

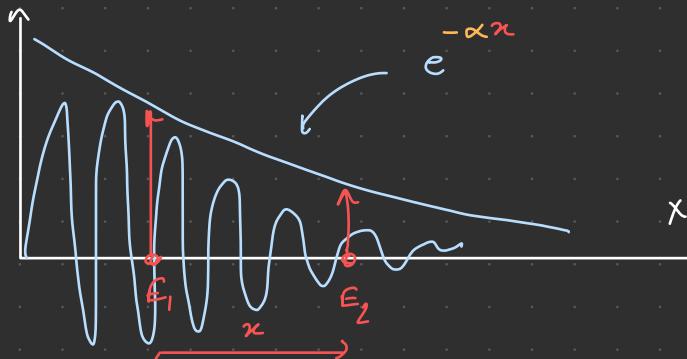
$$E_y = E_m e^{j\omega t} e^{-\alpha x} e^{-j\beta x}$$

$\alpha$  = attenuation constant  
(nepers/m)

$\gamma$  = phase constant (rad/s/m)

$$\gamma = \alpha + j\beta$$

if you need an  $\vec{E}$  field through  $x$  (medium),  
the size of wave attenuates



$$|E_2| = |E_1| e^{-\alpha x} = \alpha = \frac{-\ln \frac{|E_1|}{|E_2|}}{x} \text{ nepers/m}$$

usually assume  $x = 1$

## What is Propagation Constant $\gamma$

→ nepers are dimensionless

$$\Rightarrow 20 \log_{10} \left| \frac{E_2}{E_1} \right| \text{ dB}$$

$$\Rightarrow 1 \text{ neper} = 0.866 \text{ dB}$$

$$\frac{20 \ln \left| \frac{E_2}{E_1} \right|}{\ln 10} \Rightarrow \ln \frac{\left| E_2 \right|}{\left| E_1 \right|} \frac{20}{\ln 10} \Rightarrow -\alpha \times 8.686 \text{ dB}$$

1. Good Conductor  $\rightarrow$  conductivity is large

$$\sigma \gg w\epsilon$$

$$\text{if } \gamma = \sqrt{jw\mu(\sigma + jw\epsilon)}$$

$$= \sqrt{jw\mu \underbrace{\epsilon(1 + j\frac{w\epsilon}{\sigma})}_{}}$$

in a good conductor,  
this is negligible

In a good  
conductor

$$\gamma = \sqrt{jw\mu}$$

# 1. $\delta$ in a Good Conductor

$$\begin{aligned}\delta &= \sqrt{j\omega\mu\sigma} & \sqrt{j} &= \left( \frac{1}{\sqrt{2}} + j \frac{1}{\sqrt{2}} \right) \\ &= \sqrt{\omega\mu\sigma} \left( \frac{1}{\sqrt{2}} + j \frac{1}{\sqrt{2}} \right) \\ &= \sqrt{\omega\pi f\mu\sigma} \left( \frac{1}{\sqrt{2}} + j \frac{1}{\sqrt{2}} \right) && \text{convert } \omega = 2\pi f \\ &= \sqrt{\pi f\mu\sigma} \cdot (1 + j) && \text{to multiply across by } \sqrt{2} \\ && \downarrow & \downarrow \\ &\delta = \alpha + j\beta & \text{this means } \alpha \approx \beta & \delta = \alpha + j\beta\end{aligned}$$
$$\delta = \alpha + j\beta = \sqrt{\pi f\mu\sigma}$$

if  $\sigma$  is a conductor

- $\sigma$  is large
- $e^{-\alpha x}$  is large  $\rightarrow$  wave attenuates rapidly
- $v = \frac{\omega}{\beta} = \frac{\omega}{\sqrt{\pi f \mu \sigma}}$   $\rightarrow$  wave velocity is slow

remember:  $\omega$  = frequency (rad/s)

$\beta$  = phase constant (rad/m)

Look at different example  $\Rightarrow \delta$  in General (2.)

$$\text{Given: } \delta = \alpha + j\beta = \sqrt{j\omega\mu(\varsigma + j\omega\varepsilon)}$$

Step 1 : Take square

$$\begin{aligned}\delta^2 &= \alpha^2 - \beta^2 + 2\alpha\beta j = j\omega\mu(\varsigma + j\omega\varepsilon) \\ &= -\omega^2\mu\varepsilon + j\omega\mu\varsigma \quad (1)\end{aligned}$$

$\rightarrow$  Trying to find independent values for  $\alpha$  and  $\beta$

Step 2 : Take square of absolute value

$$\begin{aligned}|\delta|^2 &= \alpha^2 + \beta^2 = (\sqrt{|j\omega\mu||\varsigma + j\omega\varepsilon|})^2 \\ &= \sqrt{\sqrt{\omega^2\mu^2} \sqrt{\varsigma^2 + \omega^2\varepsilon^2}} \\ &= \omega\mu\sqrt{\varsigma^2 + \omega^2\varepsilon^2} \quad (2)\end{aligned}$$

# Rough Practice Test

---

$$\begin{aligned}
 1. \quad \delta &= \alpha + j\beta = \sqrt{jw\mu(\varsigma + jw\varepsilon)} \\
 &= \alpha^2 + j^2\beta^2 + 2j\alpha\beta \\
 &= \alpha^2 - \beta^2 + 2j\alpha\beta = jw\mu(\varsigma + jw\varepsilon) \\
 &= \boxed{jw\mu\varsigma - w^2\mu\varepsilon}
 \end{aligned}$$
  

$$\begin{aligned}
 2. \quad |\delta|^2 &= \alpha^2 + \beta^2 = |\sqrt{jw\mu(\varsigma + jw\varepsilon)}|^2 \\
 &= \sqrt{|jw\mu||\varsigma + jw\varepsilon|} \\
 &= \sqrt{\sqrt{j^2w^2\mu^2} \sqrt{\varsigma^2 + j^2w^2\varepsilon^2}} \\
 &= \sqrt{\sqrt{w^2\mu^2} \sqrt{\varsigma^2 + w^2\varepsilon^2}} \\
 &= w\mu \sqrt{\varsigma^2 + w^2\varepsilon^2}
 \end{aligned}$$


(Take absolute square value of everything)

## 2. $\delta$ in General

Step 3: Equate Real Parts

$$(1) \alpha^2 - \beta^2 = -\omega^2 \mu \varepsilon \quad \begin{matrix} \nearrow \text{only real parts!} \\ \downarrow \end{matrix}$$

$$(2) \alpha^2 + \beta^2 = \omega \mu \sqrt{\sigma^2 + \omega^2 \varepsilon^2} \quad \begin{matrix} \beta^2 \text{ cancel} \\ \text{out} \end{matrix}$$

$$(1) + (2) = 2\alpha^2 = \omega \mu \sqrt{\sigma^2 + \omega^2 \varepsilon^2} - \omega^2 \mu \varepsilon$$

$$\alpha = \sqrt{\frac{\omega \mu \sqrt{\sigma^2 + \omega^2 \varepsilon^2} - \omega^2 \mu \varepsilon}{2}} \quad \begin{matrix} \text{multiply} \\ \text{everything} \\ \text{out} \end{matrix}$$

$$= \omega \sqrt{\frac{\mu \varepsilon}{2} \left( \sqrt{1 + \sigma^2 / \varepsilon^2 \omega^2} - 1 \right)} \quad \begin{matrix} \downarrow \\ \text{divide} \\ \text{across by} \\ \omega^2 \varepsilon^2 \end{matrix}$$

following same steps for  $\beta$ :

$$\beta = \omega \sqrt{\frac{\mu \varepsilon}{2} \left( \sqrt{1 + \sigma^2 / \varepsilon^2 \omega^2} + 1 \right)}$$

Rough:

$$\sqrt{\frac{\frac{\omega \mu}{\omega} \sqrt{\frac{\sigma^2 + \omega^2 \varepsilon^2}{\omega^2}} - \frac{\omega^2 \varepsilon}{\omega}}{2}} = \sqrt{\frac{\mu \sqrt{\sigma^2 / \omega^2 + \varepsilon^2}}{2}} - \frac{\varepsilon \cdot \omega}{2}$$

$$\alpha = \sqrt{\frac{wH\sqrt{\epsilon^2 + w^2\Sigma^2} - w^2H\Sigma}{2}}$$

$$= \sqrt{\frac{H}{2} \left( w\sqrt{\epsilon^2 + w^2\Sigma^2} - w^2\Sigma \right)}$$

$$= \sqrt{\frac{\mu w^2}{2} \left( \frac{1}{w\Sigma} \sqrt{\frac{\epsilon^2}{w^2\Sigma^2} + 1} - 1 \right)}$$

$$= w \sqrt{\frac{\mu\Sigma}{2} \sqrt{\frac{\epsilon^2}{w^2\Sigma^2} + 1} - 1}$$

$$\beta = w \sqrt{\frac{\mu\Sigma}{2} \sqrt{\frac{\epsilon^2}{w^2\Sigma^2} + 1} + 1}$$

### 3. $\sigma$ in a Good Dielectric

(Insulator)

$$\sigma \ll \omega \epsilon$$

(opposite !! to good conductor)

$$\alpha = \omega \sqrt{\frac{\mu \epsilon}{2} \left( \sqrt{1 + \sigma^2 / \epsilon^2 \omega^2} - 1 \right)}$$

(attenuation const).

if  $\sigma \rightarrow$  use binomial expansion  $(\sqrt{1+x})$   
small

$$\text{of } \left( 1 + \frac{\sigma^2}{\epsilon^2 \omega^2} \right)^{1/2} \quad \underline{\text{Taylor Expansion}}$$

$\rightarrow$  by the time this reaches  $1 + \frac{x}{2} + \frac{x^2}{8} \dots$ ,

the value becomes negligibly small

$$\alpha = \omega \sqrt{\frac{\mu \epsilon}{2} \cdot \frac{\sigma^2}{\epsilon^2 \omega^2}} = \omega \underbrace{\sqrt{\frac{\mu \sigma^2}{2 \epsilon^2 \omega^2}}} = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}}$$

$$\left( 1 + \frac{\sigma^2}{\epsilon^2 \omega^2} \right)^{1/2} \rightarrow \underbrace{\frac{\sigma^2}{2 \epsilon^2 \omega^2}}_{= \frac{\omega \sigma}{\omega^2 \cdot \sqrt{\frac{\mu}{\epsilon}}}} = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}}$$

### 3. $\oint$ in a Good Dielectric

$$\beta = \omega \sqrt{\frac{\mu \epsilon}{2}} \left( \underbrace{\sqrt{1 + \frac{\epsilon^2}{\epsilon^2 w^2}}}_{\text{take binomial of this}} + 1 \right)^{-1/2}$$

$\downarrow \text{T.S.}$

$$\frac{\epsilon^2}{2 \cdot \epsilon^2 \cdot w^2} + 2$$

divide across by 2

$$= \omega \sqrt{\frac{\mu \epsilon}{2}} \left( 2 + \frac{\epsilon^2}{2 \epsilon^2 w^2} \right) \approx \omega \sqrt{\mu \epsilon} \left( 1 + \frac{\epsilon^2}{8 \epsilon^2 w^2} \right)$$

$$\checkmark (\text{velocity}) = \frac{\omega}{\beta} = \frac{1}{\omega \sqrt{\mu \epsilon} \left( 1 + \frac{\epsilon^2}{8 \epsilon^2 w^2} \right)}$$

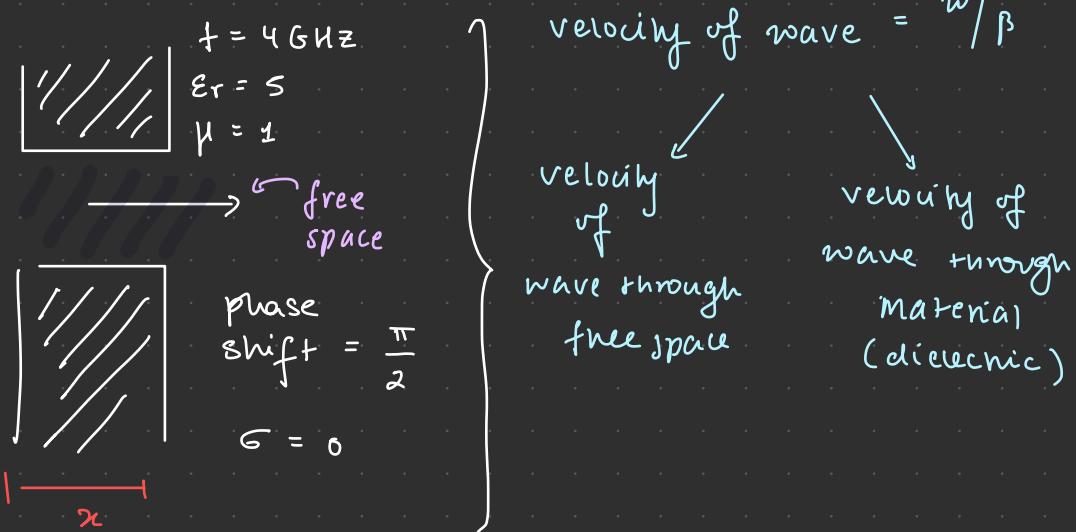
$$= \frac{1}{\mu \epsilon} \left( 1 - \frac{\epsilon^2}{8 \epsilon^2 w^2} \right)$$

$\rightarrow$  velocity is high, phase shift is small

# Summary

	velocity ( $v$ )	attenuation ( $\alpha$ )	phase shift ( $\beta$ )
Good conductor $\sigma \gg \omega \epsilon$	$\frac{\omega}{\sqrt{\mu f \mu \epsilon}}$	$\sqrt{\pi f \mu \epsilon}$	$\sqrt{\pi f \mu \epsilon}$
Good dielectric $\sigma \ll \omega \epsilon$	$\frac{1}{\mu \epsilon} \left( 1 - \frac{\sigma^2}{8\omega^2 \epsilon^2} \right)$	$\frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}}$	$\omega \sqrt{\mu \epsilon} \left( 1 + \frac{\sigma^2}{8\omega^2 \epsilon^2} \right)$
Insulator $\sigma = 0$	$\frac{1}{\sqrt{\mu \epsilon}}$	0	$\omega \sqrt{\mu \epsilon}$

## Example Q1 Q2.



$$v_1 = \frac{w}{\beta_1}, \quad v_2 = \frac{w}{\beta_2} \rightarrow \beta_1 = \frac{w}{v_1} \rightarrow \beta_2 = \frac{w}{v_2}$$

Phase difference between signals

$$\beta_2 x - \beta_1 x = \pi/2 = \frac{w x}{v_2} - \frac{w x}{v_1}$$

$$x = \frac{\pi/2}{w \left( \frac{1}{v_2} - \frac{1}{v_1} \right)}$$

$$v_1 = v \text{ through free space} = c = \frac{1}{\sqrt{\epsilon_0 \mu_0}}$$

$$v_2 = v \text{ through dielectric} = \frac{1}{\sqrt{\epsilon_0 \epsilon_r \mu_0}}$$

Q1 cont.

$$x = \frac{\pi / 2}{w(\gamma_{v_2} - \gamma_{v_1})}$$

$$= \frac{\pi}{2w(\sqrt{\epsilon_0 \epsilon_r \mu_0 \mu_r} - \sqrt{\epsilon_0 \mu_0})}$$

$$= 0.152 \text{ m} \quad (\text{given } \epsilon_0, \epsilon_r, \mu_0, \mu_r, w = 2\pi f)$$

Q2

$\sigma \gg w\epsilon = \text{conductor}$
$\sigma \ll w\epsilon = \text{dielectric}$

$$\epsilon_r = 35$$

$$\sigma = 0.2 \text{ S/m}$$

$$V = \frac{\omega}{\beta}, f = 100 \text{ MHz}$$

$\gamma = \alpha + j\beta \rightarrow$  use to find if good conductor / insulator / dielectric

$$\alpha = \omega \sqrt{\frac{\mu\epsilon}{2} \left( \sqrt{1 + \frac{\sigma^2}{\epsilon^2 \omega^2}} - 1 \right)}$$

nepers/m  $\rightarrow \frac{dB}{m}$

$$\beta = \omega \sqrt{\frac{\mu\epsilon}{2} \left( \sqrt{1 + \frac{\sigma^2}{\epsilon^2 \omega^2}} + 1 \right)}$$

rads/m

using  $\gamma = \alpha + j\beta = \sqrt{j\omega\mu(\epsilon + j\omega\epsilon)}$

→ direct definition

→ equate real + imaginary components

$$\sqrt{-\omega^2\mu\epsilon + j\omega\mu\sigma}$$

To polar:  $z = |r|(\cos\theta + j\sin\theta)$ ,  $r = |\gamma|$

Q2 cont...

$$\gamma = \sqrt{-\omega^2 \mu \varepsilon + j\omega \mu \sigma}$$

→ sub in values →

$-\omega^2 \mu \varepsilon = -153.8$

$j\omega \mu \sigma = 154.9$

→ get angle + abs value

$$|\gamma| = 220.4, \quad \angle \gamma = 184.2^\circ$$

$$\begin{aligned}\gamma &= \sqrt{220.4} \left( \cos(64.1) + j \sin(64.1) \right) \\ &= \sqrt{r} \left( \cos\left(\frac{\theta}{2}\right) + j \sin\left(\frac{\theta}{2}\right) \right)\end{aligned}$$

## Tutorial Q

### Question 2

A plane electromagnetic wave of frequency  $1\text{ GHz}$  is normally incident on a slab of material with a width of  $.005\text{ m}$ , which has a relative permittivity of  $15$ , a relative permeability of  $10$  and a conductivity of  $10\text{ }\Omega^{-1}\text{ m}^{-1}$ . There is a large hole in the slab and measurements made on the field that propagates through the slab and on the field that propagates through the hole. What is the phase difference and amplitude difference of the two parts of the wave (wave propagation through free space, wave propagating through the material) after the slab thickness. Do not consider boundary effects on the wave in the material.

$$f = 1\text{ GHz}$$

$$A_{in} = 5 \text{ Volts/m}$$

$$x = 0.005\text{ m}$$

$$A = A e^{-\alpha x}$$

$$\epsilon_r = 15 \text{ (permittivity)}$$

$$\mu_r = 10 \text{ (permeability)}$$

$$\sigma = 10 \text{ (conductivity)}$$



know that it's a good conductor because  $\sigma \gg \omega \epsilon$

$$\alpha = \frac{\sigma}{\omega \epsilon} = \frac{10}{2\pi f (15)} =$$

$$\beta = \alpha = \sqrt{\pi f \mu \sigma} \longrightarrow \beta_1 x - \beta_2 x$$

$$\Rightarrow (5.600(0.005) - 2\pi f \sqrt{\mu \epsilon})$$

$$\text{Amp diff} \rightarrow A = A e^{-\alpha x}$$

# Tutorial

$$\text{if } E_y = E_m \cos(\beta x - \omega t) \quad \int \rightarrow E_m \cos \beta x - \underbrace{\cancel{\omega t}}_{\cos(1) = \cos(-1)}$$

$\hookrightarrow$  define  $\vec{n}$  and  $\vec{E}$  field

$$E_m \cdot \cos(\beta x - \omega t) \quad \omega = v \beta$$

Solution: write as Real part of a complex expression

$$\therefore E_y = E_m \cos(\beta x - \omega t)$$

$$= E_m \cos(\omega t - \beta x)$$

$$= \text{Real} \left( E_m e^{j(\omega t - \beta x)} \right)$$

$$= \text{Real} \cdot E_m e^{j\omega t} e^{-j\beta x}$$

$$\text{let } E_0 = E_m e^{-j\beta x} \quad \therefore \quad \vec{E}_0 \cdot e^{j\omega t}$$

$\therefore$

This Plane Polarised Wave  $\Rightarrow$  only has

$$\frac{\partial B}{\partial t} = \frac{\partial}{\partial t} \frac{\partial E_y (\cos(\omega t - \beta x))}{\partial x} \quad \text{only diff by } z.$$

$$= E_y (\sin \omega t - \beta x) \cdot \beta$$

$$\int \frac{\partial B}{\partial t} = \vec{B} = E_y \cdot \cos(\omega t - \beta x) \cdot \frac{\beta}{\omega}$$

Maxwell Eqn

$$\frac{\partial \vec{B}}{\partial t} = -\vec{\nabla} \times \vec{E}$$

$$-\vec{\nabla} \times \vec{E} \text{ of plane p. wave}$$

$$= \frac{\partial E_y}{\partial x} \hat{k}$$

$\uparrow$  can decide  
this because  
 $E_y (\cos \beta x - \omega t)$   
only

$$\frac{\partial E_y}{\partial x}$$

has  
 $E_y(x)$

Propagation:  $\vec{J} = \vec{E}$

remember that  
 $\lambda = v/f$

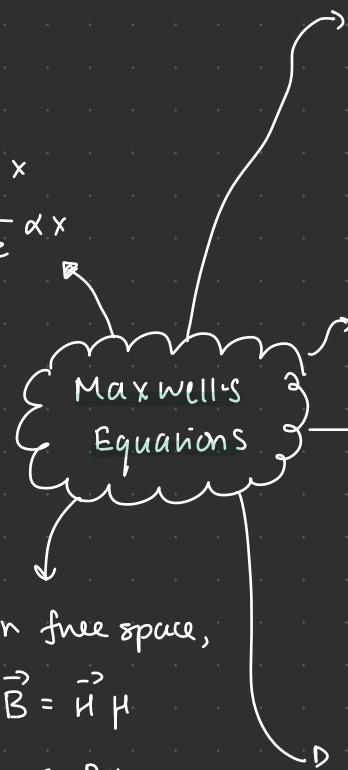
Phase Diff =  $\beta_1 x - \beta_2 x$

Amp Diff =  $A - A_{in} e^{-\alpha x}$

To convert to dB  $\rightarrow$

$\times 8.686$

$20 \log_{10} \left( \frac{A_{out}}{A_{in}} \right) = \text{dB gain}$



$$\frac{\partial \vec{B}}{\partial t} = -\vec{\nabla} \times \vec{E}$$

Plane Polarised wave means

$$\frac{\partial E_y}{\partial x} \rightarrow \text{only in } \hat{z} \cdot \hat{x}$$

in free space,

$$\vec{B} = \vec{H} \mu$$

$$w = \beta v.$$

$$v = \frac{1}{\sqrt{\epsilon \mu}}$$

### Relationships

$$\vec{B} = \frac{\beta}{w} E_y \cdot \hat{x}$$

$$\vec{B} = \frac{\beta}{w} \cdot E_y \hat{z}$$

$$\left\{ \begin{array}{l} \vec{H}_z \mu_0 \cdot v = \vec{E}_y \\ \vec{H}_z \mu_0 = \frac{\vec{E}_y}{\sqrt{\epsilon \mu_0}} \end{array} \right\}$$

$$\begin{aligned} \vec{B}_z &= \vec{H}_z \mu_0 = \frac{\beta}{w} \vec{E}_y \\ \frac{H_z \mu_0 \cdot w}{\beta} &= \vec{E}_y \end{aligned}$$

Relating  
 $H_z \longrightarrow E_y$

# Propagation Revision

- $\vec{\nabla} \times \vec{E} = \mu \frac{\partial \vec{H}}{\partial t}$  curl of  $E = \mu \times$  partial diff of  $H$  by  $t$
- $\vec{\nabla} \times \vec{H} = \vec{J} + \underbrace{\frac{\epsilon \cdot \partial \vec{E}}{\partial t}}$   
 $\vec{J} = \sigma \vec{E} = 0$
- Poisson's Equations:  $\vec{\nabla}(\vec{\nabla} \cdot \vec{E}) - \vec{\nabla}^2 \cdot \vec{E}$   
 $= \vec{\nabla} \times \vec{\nabla} \times \vec{E}$
- remember  $\frac{d}{dt} = j\omega$

E.g. derive propagation constant

given:  $E_y \Rightarrow E_m \cos \beta(x-vt)$

$$= E_m \cos(\beta x - \beta vt) \longrightarrow \beta \cdot v = \omega$$
$$= E_m \cos(\beta x - \omega t)$$
$$= E_m \cos(\omega t - \beta x) \longrightarrow \cos(-1) = \cos(1)$$

write as Real part of complex value in exponential form:

$$= E_m e^{j(\omega t - \beta x)}$$
$$= E_m e^{j\omega t} e^{-\beta jx} \longrightarrow \text{let } E_0 = E_m e^{-\beta jx} \text{ for } t=0$$
$$= E_0 e^{j\omega t}$$

if  $E = E_0 e^{j\omega t}$ ,  $\frac{\partial E}{\partial t} = j\omega E_0 e^{j\omega t}$

and let  $\frac{d}{dt} = j\omega$

Remember maxwell's law  $\vec{\nabla} \times \vec{E} = \vec{H} \frac{d\vec{H}}{dt}$  ( $\vec{H} = H_0 e^{j\omega t}$ )

- $\vec{\nabla} \times (E_0 e^{j\omega t}) = j\omega \mu (H_0 e^{-j\omega t})$  at  $t=0$ ,  $H_0 e^{j\omega t} = \vec{H}_0$
- $\vec{\nabla} \times \vec{E} = j\omega \mu \vec{H}_0$

Steps: start with  
 $\vec{\nabla} \times \vec{E}$  in H  
at  $t=0$

## Maxwell's Eqns Again!

$$\vec{\nabla} \times \vec{H} = \vec{J} + \frac{\varepsilon \frac{\partial \vec{E}}{\partial t}}{\mu} = \vec{G} \vec{E} + \varepsilon \cdot \frac{\partial \vec{E}}{\partial t} \quad (\text{H-field})$$

$$\vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} \quad (\text{E-field})$$

Poisson :  $\underbrace{\nabla(\nabla \cdot \vec{E})}_{\substack{\text{div of} \\ \text{div}}} - \underbrace{\nabla^2 \cdot \vec{E}}_{\nabla^2 \cdot \vec{E}} = \underbrace{\vec{\nabla} \times \vec{\nabla} \times \vec{E}}_{\substack{\text{curl of curl} \\ \text{of} \vec{E}}}$

(but  $\nabla \cdot \vec{E} = 0$ )

using Poisson's Equation :  $\vec{\nabla} \times \vec{\nabla} \times \vec{E}_0 = \vec{\nabla} \times \left( -\mu \frac{\partial \vec{H}}{\partial t} \right)$

$$\int \frac{\partial \vec{H}}{\partial t} \Rightarrow \int H_0 e^{-j\omega t} = j\omega H_0 e^{-j\omega t}$$

at  $t=0 \rightarrow j\omega H_0 \therefore \vec{\nabla} \times \vec{\nabla} \times \vec{E}_0 = \vec{\nabla} \times (\mu j\omega H_0)$

$\nabla \cdot \vec{E} = 0$ , no divergence of  $\vec{E}$  field  $\therefore$

$$-\nabla^2 \cdot \vec{E} = \vec{\nabla} \times \vec{\nabla} \times \vec{E} = \vec{\nabla} \times (\mu j\omega H_0)$$

$$-\nabla^2 \cdot \vec{E} = \mu \cdot j\omega (\vec{\nabla} \times H_0)$$

Take  $\vec{\nabla} \times H = \vec{J} + \frac{\varepsilon \frac{\partial \vec{E}}{\partial t}}{\mu}$

remember, if  $E_0 \rightarrow E_0 e^{j\omega t} e^{-j\omega X} \rightarrow E_0 e^{j\omega t} \rightarrow E_0 \cdot j\omega$

$$\text{if } -\nabla^2 \vec{E} = (\vec{\nabla} \times \vec{H}_0) \mu j \omega$$

$$\text{and } \vec{\nabla} \times \vec{H} = \vec{J} + \frac{\epsilon \partial \vec{E}}{\partial t}$$

$$= \sigma \vec{E} + \frac{\epsilon \partial \vec{E}}{\partial t} = \sigma \vec{E} + j\omega \vec{E}$$

because  $j\omega = \frac{d}{dt}$

$$\text{Remember: } -\nabla^2 E_0 = \vec{\nabla} \times \vec{H} \quad (\text{because } \vec{\nabla} \cdot \vec{E} = 0)$$

$$\vec{\nabla} \times \vec{H} = \vec{E} (\sigma + \epsilon \left( \frac{\partial}{\partial t} \right))$$

$$\begin{aligned} -\nabla^2 \cdot E_0 &= \vec{E} (\sigma + j\omega \epsilon) (\mu j \omega) \\ &= \mu j \omega (\sigma + j\omega \epsilon) \cdot \vec{E} \end{aligned}$$

$$-\nabla^2 E_{0y} = E_{0y} [\mu j \omega (\sigma + j\omega \epsilon)]$$

$$\gamma^2 = \mu j \omega (\sigma + j\omega \epsilon)$$

$$\gamma = \sqrt{\mu j \omega (\sigma + j\omega \epsilon)} \quad (\text{prop constant})$$

Q1

Past Nidkem to satisfy Eqns

$$\vec{E} = 2b(x^2 - z) \hat{i} + \hat{i} + 2yt\hat{j} + ay^2\hat{k} \quad E\text{-field}$$

$$\vec{B} = -2ay\hat{i} - b\hat{x}t^2\hat{j} + by\hat{k} \quad \text{magnetic flux density}$$

Maxwell's laws:  $\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$

$$\vec{\nabla} \cdot \vec{D} = P \quad \left. \right\} \vec{B} = \mu \vec{H}$$

$$\vec{\nabla} \times \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t}$$

Steps 1: curl of  $\vec{E}$ ,  
 2. diff of  $\vec{B}$  }  $\rightarrow$  easy!

$$\begin{vmatrix} i & j & k \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ 2b(x^2+z) & ay^2t & 2yt \end{vmatrix} = i(2ya - 0) - j(0 - 2b) + k(0)$$

$$= 2ay\hat{i} + 2b\hat{j} = \vec{\nabla} \times \vec{E}$$

$$\frac{\partial \vec{B}}{\partial t} = -2ay\hat{i} - 2bz\hat{j} + 0\hat{k}$$

should equal  
each other.

To determine charge  
and current  
densities

Remember

$$\vec{\nabla} \cdot \vec{D} = \rho / \epsilon$$

$$\vec{\nabla} \cdot \vec{D} = \rho \text{ (charge density)}$$

$$\vec{\nabla} \times \vec{H} = \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t}$$

$$\vec{\nabla} \cdot \vec{B} = 0$$

$$\vec{\nabla} \cdot \vec{D} = \rho$$

$$D = \epsilon \vec{E}$$

$\vec{E}$   
displacement  
electric field

$$\vec{B} = -2ay\hat{i} - b\hat{x}t^2\hat{j} + by\hat{k}$$

$$\vec{\nabla} \cdot \vec{E} = 4bx\hat{i} + 2t\hat{j} + 0\hat{k} = \rho / \epsilon = \frac{\text{charge density}}{\text{permittivity}}$$

$$\rho = \epsilon (4bx\hat{i} + 2t\hat{j})$$

$$\vec{\nabla} \times \vec{H} = \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t} = \frac{\vec{\nabla} \times \vec{B}}{\mu}$$

$$\vec{\nabla} \cdot \vec{J} + \frac{\partial \rho}{\partial t} = 0$$

## Charge + Charge density

$$\vec{\nabla} \times \vec{H} = \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t}$$

$$B = \mu H \therefore H = B/\mu$$

$$\frac{\vec{\nabla} \times \vec{B}}{\mu} = \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t} \quad \therefore \quad \vec{J} = \frac{\vec{\nabla} \times \vec{B}}{\mu} - \epsilon \frac{\partial \vec{E}}{\partial t}$$

$$\vec{\nabla} \cdot \vec{J} + \frac{\partial \rho}{\partial t} = 0 \quad (\text{divergence of charge} = -\text{diff. of density with } t)$$

when these are equal, can ensure they are =

# Maxwell's Eqns

## Summary

---

- $\vec{\nabla} \times \vec{E} = \mu \frac{\partial \vec{H}}{\partial t} = \frac{\partial \vec{B}}{\partial t}$
- $\vec{\nabla} \times \vec{H} = \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t}, \quad \vec{J} = -\sigma \vec{E}$
- $\vec{B} = \mu \vec{H}$
- $\vec{E} \cdot \vec{\epsilon} = \vec{D} \quad \longrightarrow \quad \vec{\nabla} \cdot \vec{E} = \rho / \epsilon$
- $\vec{\nabla} \cdot \vec{J} = -\frac{\partial \rho}{\partial t}$

Q2

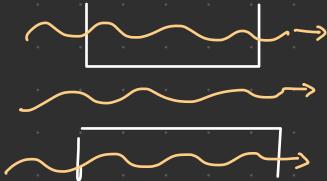
$$f = 4 \text{ GHz}$$

$$\epsilon_r = 5 \quad \text{permittivity}$$

$$\mu_r = 1 \quad \text{permeability}$$

$$G = 0$$

To find  $\alpha$



$$\text{phase diff} = 90^\circ$$

Remember:  $\omega = v \beta$ , frequency = velocity  $\times$  phase shift

$$\text{Phase shift} \Rightarrow \beta_1 x - \beta_2 x = \frac{\pi}{2}$$

$$\frac{\omega x}{v_1} - \frac{\omega x}{v_2} = \frac{\pi}{2}$$

$$\omega \left( \frac{1}{v_1} - \frac{1}{v_2} \right)$$

$\underbrace{\phantom{0}}$

$$v_1, v_2 \text{ through } \frac{1}{\sqrt{\epsilon_0 \epsilon_r \mu_0 \mu_r}}$$

$$\frac{1}{\sqrt{\epsilon_0 \epsilon_r \mu_0 \mu_r}}$$

## Poynting Vector

- vector that carried power of a wave
- direction of power transfer
- integrate space between coaxial cables  
(between conductors)

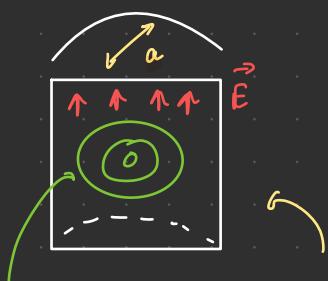
## Cylindrical current carrying conductor

Current  $\rightarrow$  proportional to  $\vec{E}$  field

$$\vec{j} = \sigma \vec{E}$$

$$H_r = \frac{\sigma I}{2\pi a^2}$$

$$E_r = \frac{J}{\sigma} = \frac{1}{\pi a^2 \sigma}$$

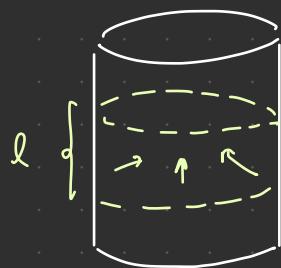


circulating  
magnetic  
field inside  
the wire  
(anti clockwise)

$\vec{P} = \vec{E} \times \vec{H} \Rightarrow$  points towards centre  
of conductor, gets weaker  
towards centre as  $\vec{H}$  gets  
weaker

## Poynting vector

$P$  at surface of conductor.  $P = E_r H_r$  where  $r = a$



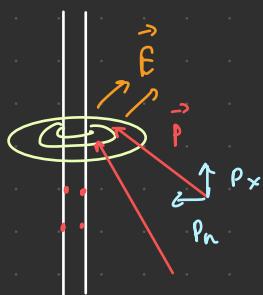
$$= \frac{a I^2}{2\pi^2 a^4 G} \quad \left. \begin{array}{l} \text{inside of wire} \\ \text{radially inward} \\ \text{from surface of wire} \end{array} \right\}$$

$$P_{\text{total}} = \frac{\sigma I^2}{2\pi^2 a^4 G} \cdot 2\pi a l$$

$$= \frac{I^2 l}{\pi a^2 G} = \frac{l}{\pi a^2 G} \cdot I^2 \quad \left. \begin{array}{l} \text{because resistance:} \\ R = \frac{1}{\pi a^2 G} = \frac{\text{length}}{\text{area}} \cdot \frac{1}{G} \end{array} \right\}$$

$$P_{\text{total}} = R I^2$$

wire carrying current:



Parallel  $\rightarrow$  transmitted

Normal  $\rightarrow$  lost due to resistance

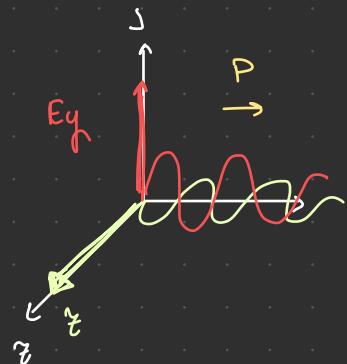
e.g. Plane Polarised wave

$$E_y = E_m \cos(\omega t - \beta x)$$

$$\vec{P} = \vec{E} \times \vec{H}$$

$$H_z = \eta E_y$$

$$P_x = \gamma E_m^2 \cos^2(\omega t - \beta x)$$



→ power transferred in same direction as wave

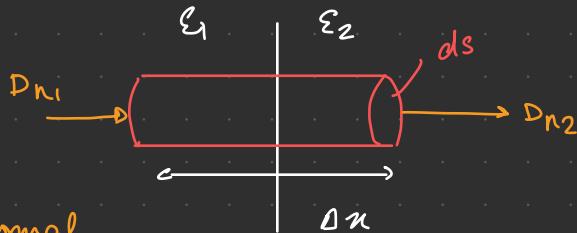
## Boundary Conditions

"Skin Effects"

for field distribution to exist:

1. Be a solution to Maxwell's Eqns
2. Boundary conditions

## Electric fields



$D_n$  = flux fields normal  
to boundary

Gauss law:  $\oint \vec{E} \cdot d\vec{s} = \frac{q_{enc}}{\epsilon}$ ,  $\vec{E} \cdot \vec{\epsilon} = 0 \Rightarrow (D_{n2} - D_{n1}) ds = ps ds$

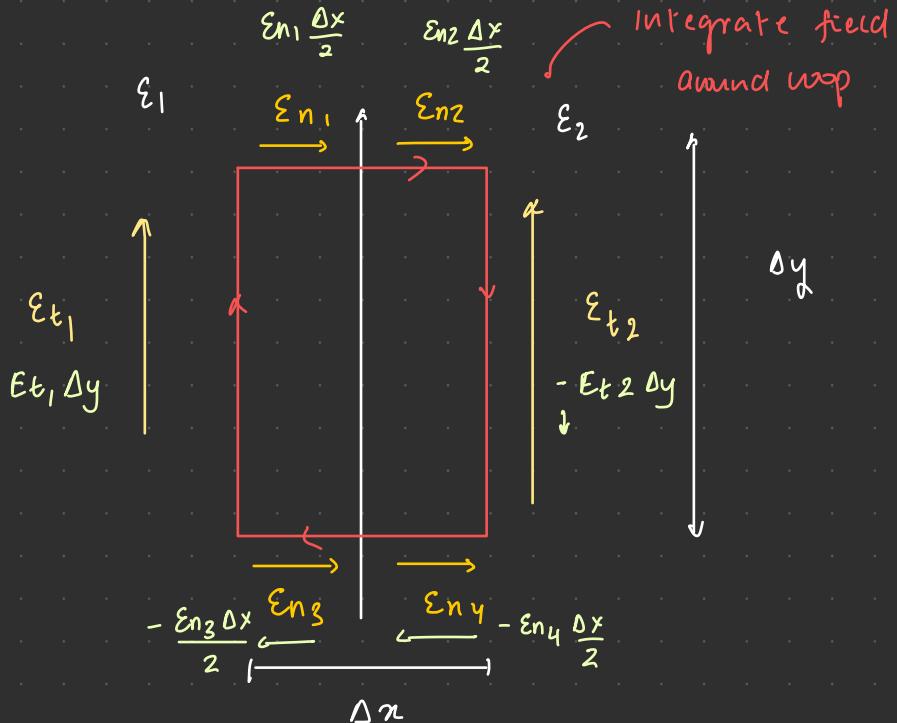
→ interaction of flux and field

$$\int \vec{D} \cdot d\vec{s} = q_{enc}$$

\* no surface charge so  $p_s = 0$ , flux normal =  $D_{n2} = D_{n1}$   
↳ means normal component is continuous

## Boundary conditions

$$\epsilon_1 = \epsilon_2$$



Combining using Faraday's law:  $\oint_C \vec{E} \cdot d\vec{l}$

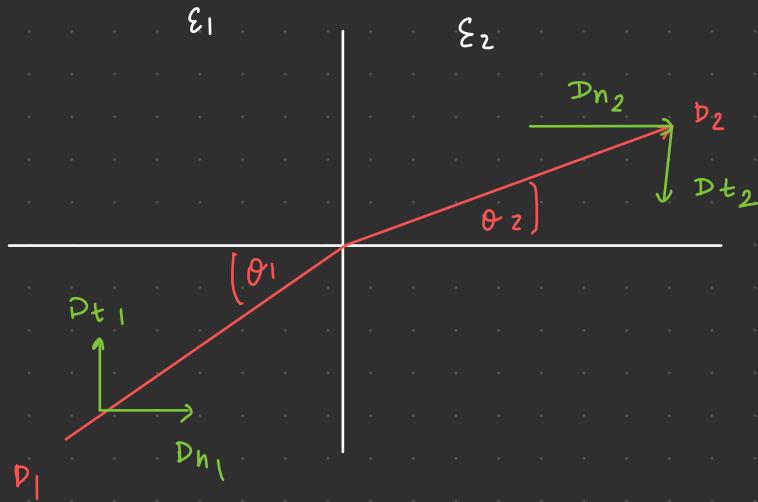
$$= - \frac{d}{dt} \int_S \underbrace{\vec{B} \cdot \hat{n} da}_{\phi \text{ flux}} = - \frac{d\phi}{dt} \text{ through surface}$$

Take limit from  $\Delta x \rightarrow 0$ :  $\oint_C \vec{E} \cdot d\vec{l} = E_{t1} \Delta y - E_{t2} \Delta y$

$$\therefore \boxed{E_{t2} = E_{t1}}$$

flux density normal = continuous, tangential  $\vec{E}$  field same.

## Electric Flux Refraction



$$\left. \begin{aligned} D_{n1} &= D_1 \cos \theta_1 \\ D_{n2} &= D_2 \cos \theta_2 \end{aligned} \right\} \quad \begin{aligned} E_{t1} &= \frac{D_1 \sin \theta_1}{\epsilon_1} \\ E_{t2} &= \frac{D_2 \sin \theta_2}{\epsilon_2} \end{aligned}$$

$\vec{E} = \frac{\vec{D}}{\epsilon}$

## Refraction

$$\sin \theta / \tan \theta = \frac{\sin \theta}{\cos \theta} \quad \left. \begin{array}{l} E_{t1} = E_{t2} \\ Dn_1 = Dn_2 \end{array} \right\}$$

$$\frac{E_{t1}}{Dn_1} = \frac{\tan \theta_1}{\epsilon_1} \quad \text{and} \quad \frac{E_{t2}}{Dn_2} = \frac{\tan \theta_2}{\epsilon_2}$$

$$\frac{\tan \theta_1}{\epsilon_1} = \frac{\tan \theta_2}{\epsilon_2}$$

$$\frac{\tan \theta_1}{\tan \theta_2} = \frac{\epsilon_1}{\epsilon_2}$$

# Magnetic Field at Boundary

↳ Same as electric

- Consider the magnetic field at a boundary between two dielectric materials with respective permeability  $\mu_1, \mu_2$ :



- From Gauss' Law for magnetic fields:  
 $(B_{n2} - B_{n1})ds = 0$ , since  $\nabla \cdot \vec{B} = 0$

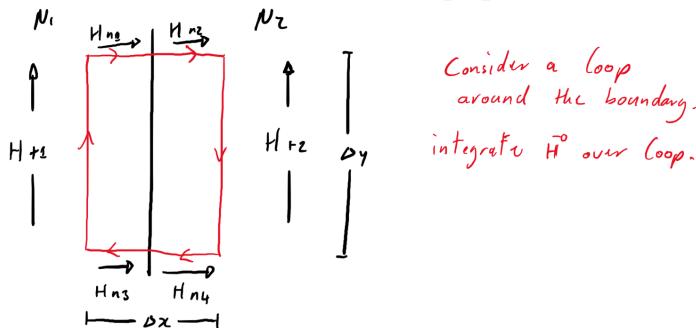
- Hence:

$$B_{n2} = B_{n1}$$

- The relationship dictates that the **normal component** of **Magnetic flux** at a dielectric boundary must be continuous.



- To find the relationship between the **tangential component** of the field at the boundary with respective permeability  $\mu_1, \mu_2$ :



$$\oint_C \vec{H} \cdot d\vec{l} = I_t = 0 \text{ when } \Delta n = 0$$

$$\oint_C \vec{H} \cdot d\vec{l} = -H_{t2}\Delta y + H_{t1}\Delta y = 0 \quad \therefore H_{t2} = H_{t1}$$

∴ tangential components of magnetic field are continuous

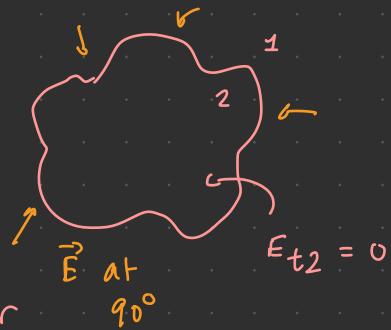
## Perfect Conductor

→ no field strength inside a conductor because charges cancel

if  $E_{t2} = 0$ ,  $E_{t1} = 0$ , no

tangential component

to  $\vec{E}$  field at surface of conductor

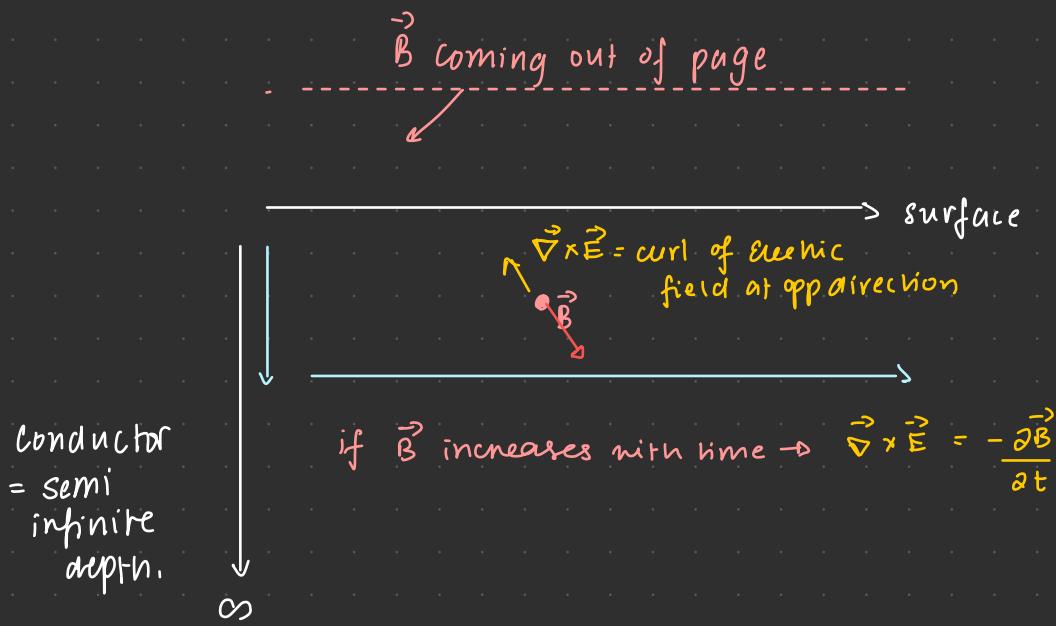


Depth of penetration

$\delta$  = skin depth.

$$\alpha = \frac{1}{2\pi f \epsilon}$$

## Imperfect Conductors

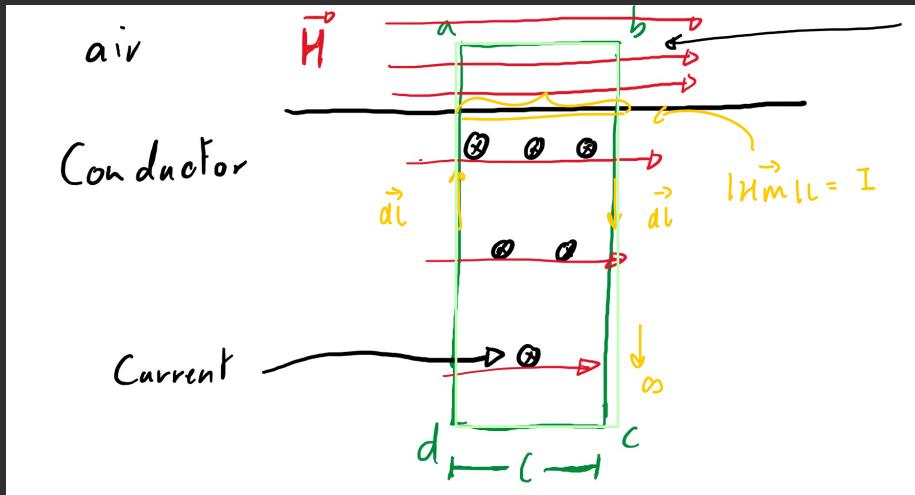


→ if there is an AC current, current will travel at "skin", very little inside

→  $\vec{B}$  decreases as moves down conductor

→  $\vec{J} = \sigma \vec{E}$ ,  $\vec{J}$  concentrated near surface

## Skin Effect



Total current enclosed:  $\oint_C \vec{H} \cdot d\vec{l} = I_+$

(right at surface from abcd)

$$ab + bc + cd + da$$

$$\text{over } l = ab$$

$$|H_m|l = ab = I$$

# Solution of wave in conducting medium :

- Solution of a wave in a conducting medium:

$$E_y = E_0 e^{j\omega t}$$

$$E_y = \underbrace{E_m}_{\text{surface}} e^{-\gamma x} e^{j\omega t}$$

- Also:  $\vec{J} = \sigma \vec{E}$

- So

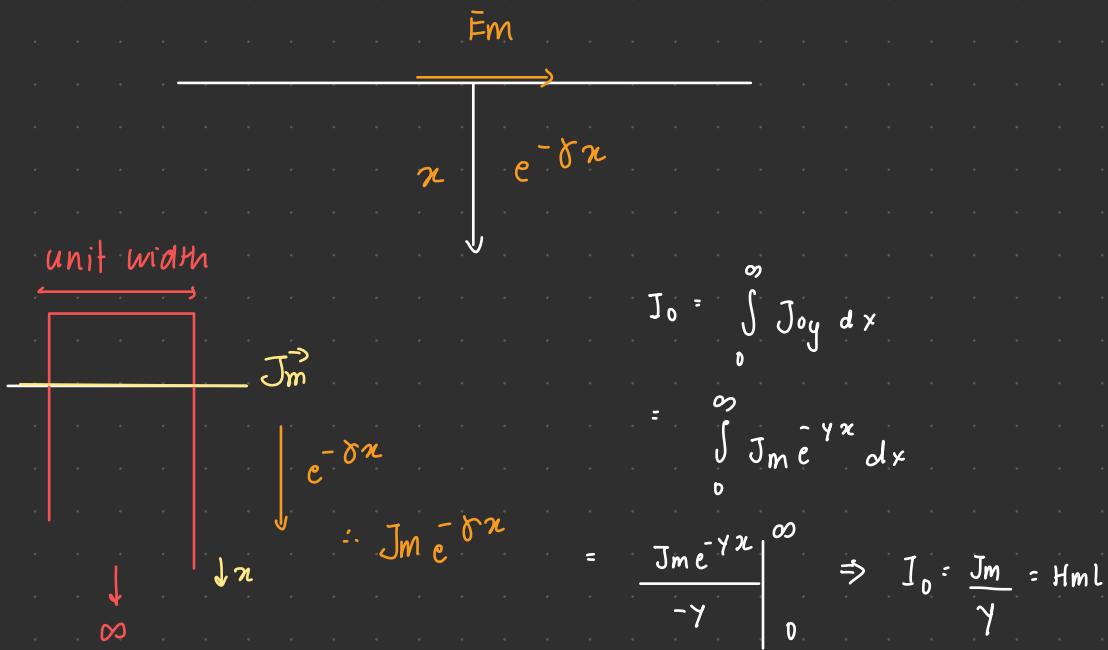
$$J_{0y} = J_m e^{-\gamma x}$$

$\delta \rightarrow$  frequency dependent

- For a conductor  $\alpha = \beta = \sqrt{\pi f \mu \sigma}$ :

$$\gamma = \alpha + j\beta$$

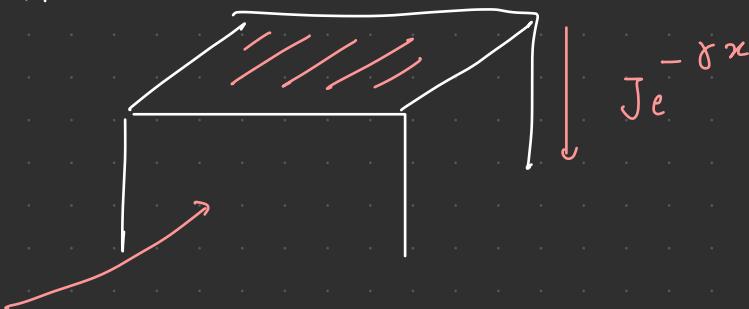
$$= \frac{1+j}{\delta} \quad , \quad \delta = \text{depth of penetration} = \frac{1}{\alpha} = \frac{1}{\beta}$$



$$J_m = \gamma H_m$$

## Skin Effect

Higher  $J_m$  at top



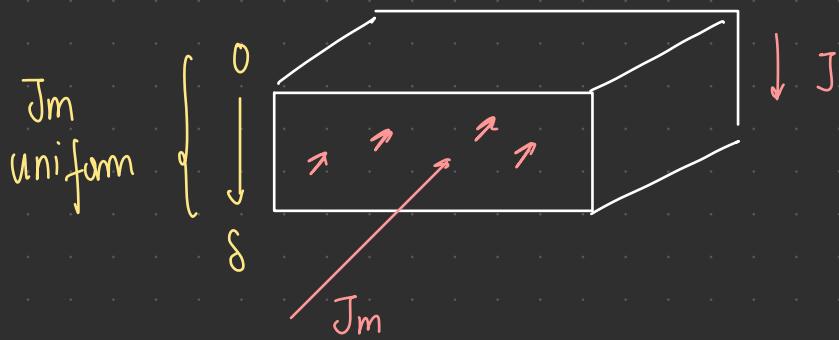
$$J_{oy} = J_e^{-\delta x}$$

Total power dissipated per Area:

$$\frac{J_m^2 \delta}{4 \sigma}$$

current  
density

## Model for Dissipation

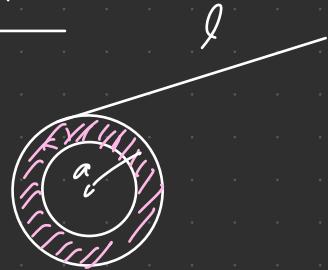


$$I_o = H_m \text{ (uniform for depth } \delta \text{)}$$

$$\text{Power} = \frac{J_m^2 \delta}{4 \sigma}$$

if dimensions  $> \delta$ , current flow uniformly  
to depth  $\delta$

## Cylindrical Conductor



$$\text{DC resistance of length } l, \quad R_{DC} = \frac{1}{\sigma} \cdot \frac{l}{\pi a^2}$$

$$\begin{aligned} \text{AC resistance: } R_{AC} &= \frac{1}{\sigma} \cdot \frac{l}{\pi a^2 - \pi (a - \delta)^2} \\ &= \frac{\frac{1}{\sigma} \cdot l}{\pi (2a\delta - \delta^2)} = \frac{l}{\sigma \cdot 2\pi a \delta} \end{aligned}$$

$$\text{if } \delta \ll a \text{ then } \frac{R_{AC}}{R_{DC}} \approx \frac{a}{2\delta}$$

$$\delta = \frac{1}{\alpha} = \frac{1}{\sqrt{c_f \pi N}}$$

If  $\downarrow \delta \uparrow R \uparrow P$

## Reflections

$$n = \sqrt{\frac{\mu}{\epsilon}}$$

$\mu$  = permeability

$\epsilon$  = permittivity

## Polarisation of Waves

horizontal  $\rightarrow$   $E$  field varies on horizontal plane wrt surface

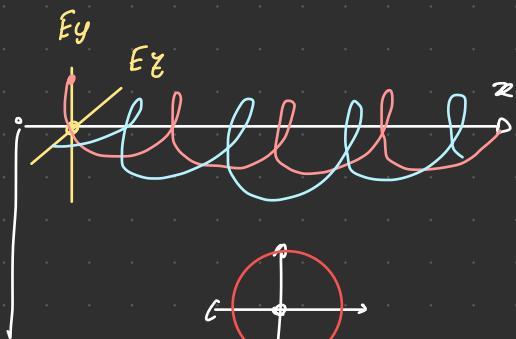
vertical:



## Elliptical Polarisation:

$$E_y = E_1 \sin(\omega t - \beta x)$$

$$E_z = E_2 \cos(\omega t - \beta x)$$

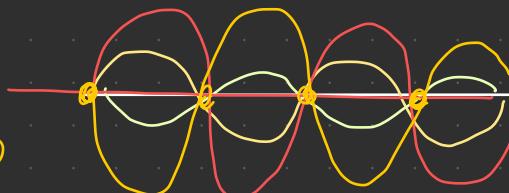


$\Rightarrow$  remember travelling:

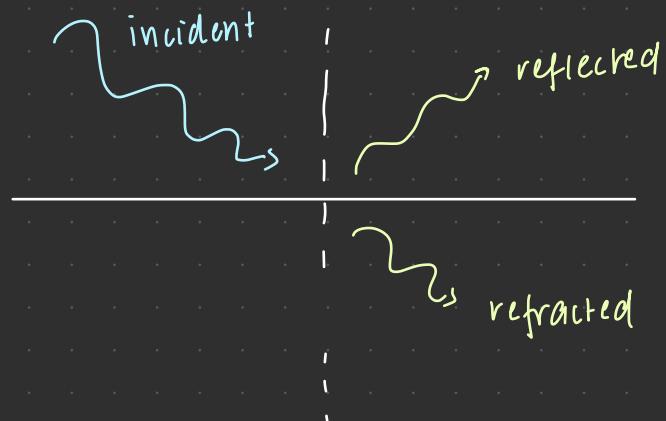


Standing wave:

nodes always at 0

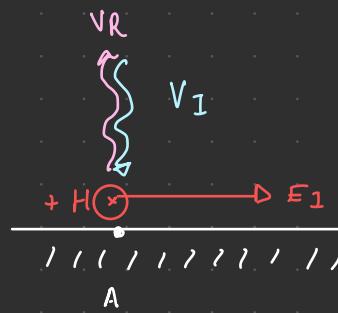


$\curvearrowleft$  reflects from wall  
 $\rightarrow$  stays in phase



Perfect conductor:

$$E_1 = E_m (\cos(\omega t - \beta x))$$



Wave incident on perfect conductor.

$\vec{E}$  field must be continuous

$$\begin{aligned} & \xrightarrow{\quad} E = 0 \\ & \xrightarrow{\quad} E \\ & \xrightarrow{\quad} E = 0 \end{aligned}$$

$$E_y = E_1 + E_R = 0 \quad (\text{in a perfect conductor})$$

at  $x = 0$

$$\left. \begin{aligned} E_y &= h_1 f(x - vt) + h_2 f(x + vt) \\ &= E_m \cos(\omega t - \beta x) + h_2 f(x + vt) \\ &= E_m \cos \beta(vt - x) + h_2 f(x + vt) \end{aligned} \right\} \begin{aligned} h_2 f(vt) &= -E_m \cos \beta(vt) \\ h_2 f(vt + x) &= -E_m \cos \beta(vt + x) \end{aligned}$$

## Reflection

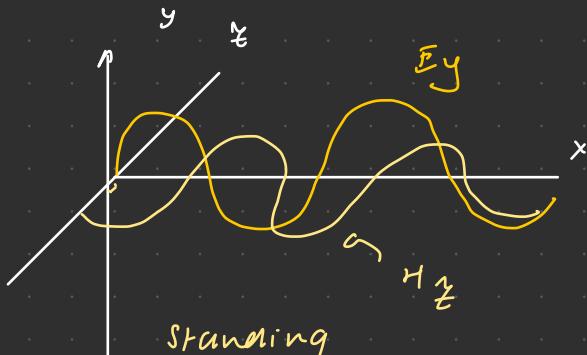
$$E_y = \underbrace{E_m \cos(\omega t - \beta x)}_{\text{incident}} - \underbrace{E_m \cos(\omega t + \beta x)}_{\text{reflection}}$$

$$E_y = 2E_m \sin(\omega t) \sin \beta x$$

Magnetic field equivalent:

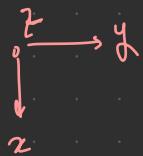
$$H_z = \frac{E_y}{\eta}$$

$$H_z = H_i + H_R = \frac{E_i}{\eta} - \frac{E_R}{\eta}$$

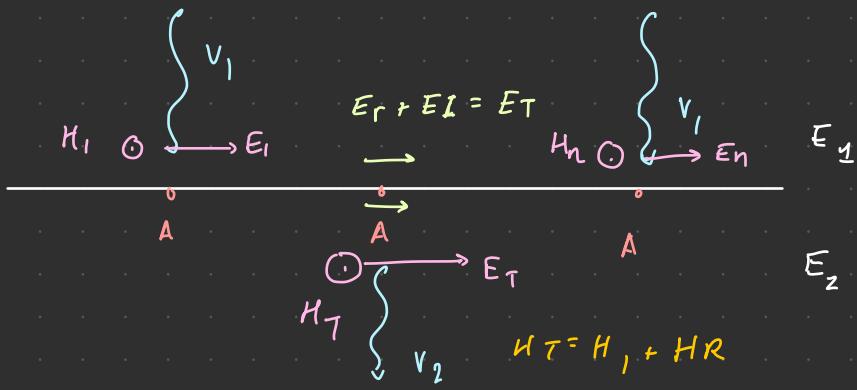


$$= \frac{E_m}{\eta} \left[ \cos(\omega t - \beta x) + \cos(\omega t + \beta x) \right]$$

both are standing waves



## Dielectric



To find:  $\frac{E_T}{E_i} = T$  = transmission coefficient

$\frac{E_r}{E_i} = P$  = reflection coefficient

$$\frac{E_1}{H_1} = \eta_1, \quad \frac{E_T}{H_T} = \eta_2, \quad \underbrace{\frac{E_R}{H_R}}_{=} = \eta \quad \text{where } \eta = \sqrt{\frac{\mu}{\epsilon}}$$

right hand rule.

for  $\mu_r = 1$  feris metanion!

= relative permittivity

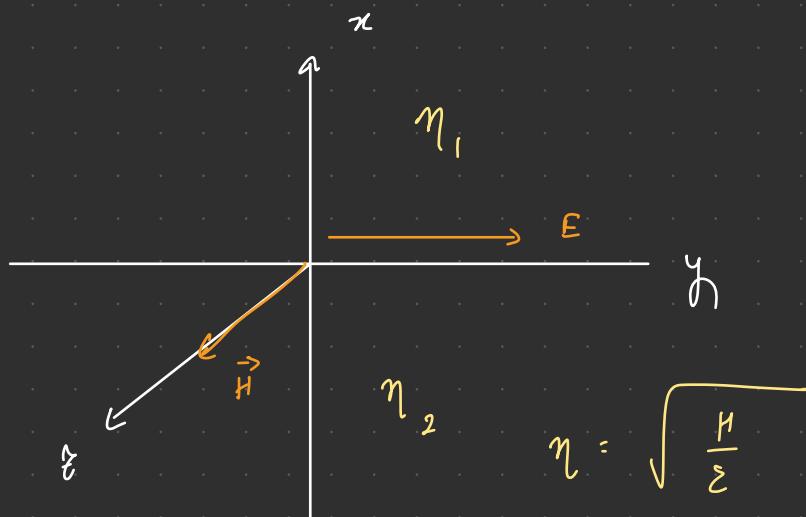
Rewrite eqns

$$\frac{E_T}{n_2} = \frac{E_I}{n_1} - \frac{E_R}{n_1}$$

$$\Rightarrow \frac{n_1}{n_2} E_I = E_I - E_R \quad \left( E_I = E_I + E_R \right)$$

$$= 2E_I = E_T \left( 1 + \frac{n_1}{n_2} \right)$$

$$= \frac{E_T}{E_I} = \frac{2n_2}{n_1 + n_2} = T = \text{transmission coefficient}$$



## Reflection

$$E_R = E_I - E_1$$

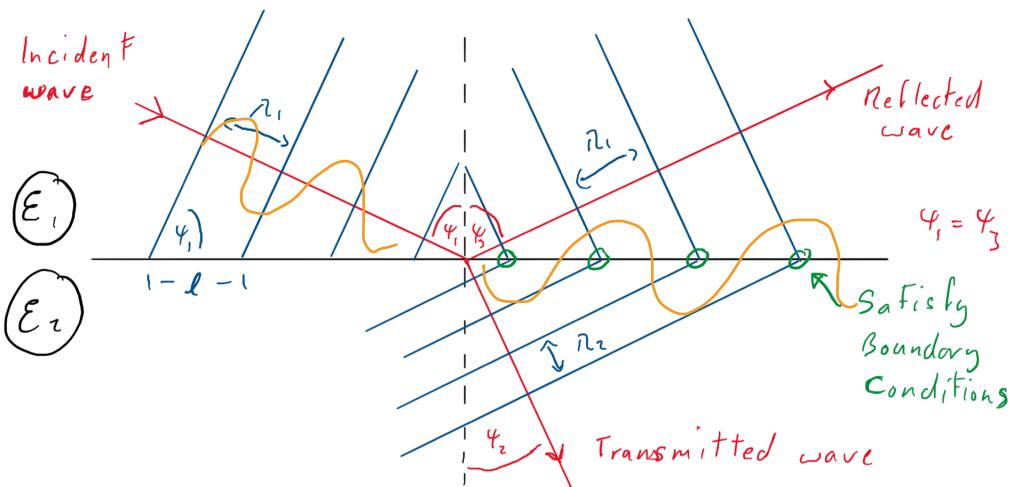
$$= E_1 \left( \frac{2\eta_2}{\eta_1 + \eta_2} - 1 \right)$$

$$= \frac{E_R}{E_1} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = p = \begin{matrix} \text{Reflection} \\ \text{Coefficient} \end{matrix}$$

## Oblique Incidence

at Boundary

---



$f_t = f_B$  To satisfy boundary condition

$$\sin \psi_1 = \frac{\lambda_1}{\ell} \quad \sin \psi_2 = \frac{\lambda_2}{\ell} \quad \therefore \quad \frac{\lambda_1}{\sin \psi_1} = \frac{\lambda_2}{\sin \psi_2}$$

$$\frac{v_1}{v_2} = \frac{\sin \psi_1}{\sin \psi_2}, \quad v = \sqrt{\frac{1}{\epsilon \mu_0}} \quad , \quad \mu_r = 1$$

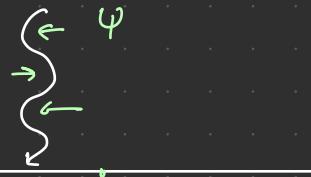
$$\sqrt{\frac{1}{\epsilon_1 \mu_0} \cdot \frac{\epsilon_2 \mu_0}{1}} = \sqrt{\frac{\epsilon_2}{\epsilon_1}} \quad \left. \right\} \text{Snell's Law}$$

## Obllique Incidence

Snell's Law

$$\vec{E} = E_m \cos(\omega t - \beta x) - E_m \cos(\omega t - \beta x)$$

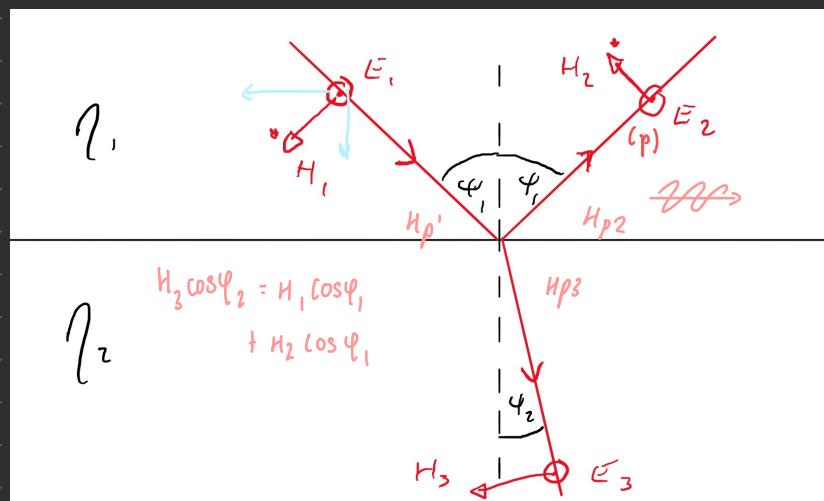
same frequency  $\{f_1, f_2\}$



$$f = \frac{v}{\lambda}$$

$$v = \frac{1}{\sqrt{\mu \epsilon}}$$

Horizontally polarised wave



## Fresnel's Eqns

T - Transmission coefficient

$$\frac{\varepsilon_1}{n_1} \cos \varphi_1 - \frac{\varepsilon_2}{n_2} \cos \varphi_2 = \frac{\varepsilon_3}{n_2} \cos \varphi_2$$

$$E_1 - E_2 = \varepsilon_3 \frac{n_1}{n_2} \frac{\cos \varphi_2}{\cos \varphi_1}$$

} relate  $\varepsilon_1$  and  $\varepsilon_3$

$$2E_1 = \varepsilon_3 \left( 1 + \frac{n_1}{n_2} \frac{\cos \varphi_2}{\cos \varphi_1} \right)$$

$$T = \frac{\varepsilon_3}{\varepsilon_1} = \frac{2n_2 \cos \varphi_1}{n_2 \cos \varphi_1 + n_1 \cos \varphi_2}$$

$$\text{Reflection coefficient} = \frac{\epsilon_2}{\epsilon_1}$$

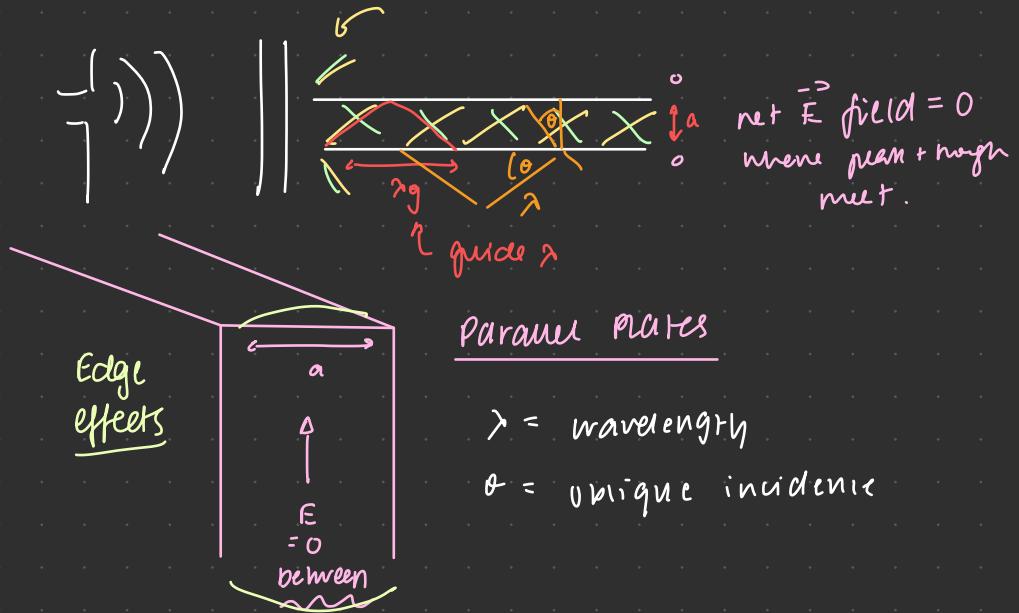
$$2\epsilon_2 = \epsilon_3 \left( 1 - \frac{n_1 \cos \varphi_2}{n_2 \cos \varphi_1} \right)$$

$$\boxed{\frac{\epsilon_2}{\epsilon_1} = \frac{n_2 \cos \varphi_1 - n \cos \varphi_2}{n_2 \cos \varphi_1 + n \cos \varphi_2} = p}$$

# Waveguides

Assumptions

same  $f$ , same  $\lambda$ , diff direction



Parallel Plates

$\lambda$  = wavelength

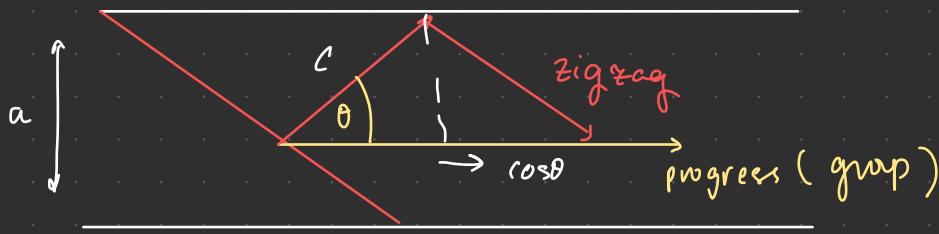
$\theta$  = oblique incidence

Cutoff = lower it'll be allowed to propagate

group velocity = speed of energy in waveguide

phase velocity = speed at phase change

## Progression in Guide



Cut off = max  $\lambda$  allowed :  $\cos \theta = \frac{\lambda}{\lambda g} \quad \left. \begin{array}{l} \\ \end{array} \right\} \text{get } \lambda_{\max}$

$$\tan \theta = \frac{\lambda g / 2}{a} = \frac{\lambda g}{2a}$$



$$\tan \theta = \frac{\sin \theta}{\cos \theta} = \frac{\sin \theta \cdot \lambda g}{\lambda} = \frac{\lambda g}{2a} \quad \therefore \lambda = 2a \sin \theta$$

$\lambda_{\max}$  when

$$\theta = 0,$$

$$\sin \theta = 1,$$

$$\boxed{\lambda_{\max} = 2a}$$

$$\lambda_c : 2a$$

$$f_c = \frac{c}{\lambda c} \quad \left. \begin{array}{l} \\ \end{array} \right\} c: \text{air}$$

$$\sin^2 \theta + \cos^2 \theta = 1$$

$$\sin \theta = \frac{\lambda}{2a} \quad \cos \theta = \frac{\lambda}{\lambda g}$$

$$\left(\frac{\lambda}{2a}\right)^2 + \left(\frac{\lambda}{\lambda g}\right)^2 = 1$$

$$\boxed{\frac{1}{4a^2} + \frac{1}{\lambda g^2} = \frac{1}{\lambda^2}} \quad 2a = \lambda c \\ \therefore$$

$$\frac{1}{\lambda c^2} + \frac{1}{\lambda g^2} = \frac{1}{\lambda^2}$$

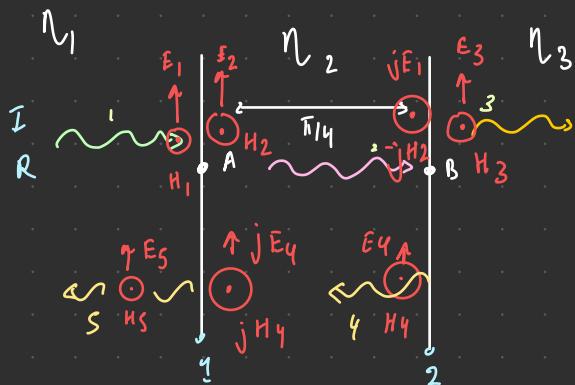
Group Velocity  $v_g = c \cdot \cos \theta$

Phase Velocity  $v_{ph} = f \cdot \lambda_g = \frac{c}{\lambda} \cdot \lambda_g$

$$v_{ph} > c, = \frac{c}{\sqrt{1 - \sin^2 \theta}}$$

$$= \frac{c}{\sqrt{1 - (\lambda / \lambda_c)^2}}$$

## Boundary conditions



single frequency  
situation

1. what value for  $n_2$  (in respect to  $n_1, n_3$ ) for no reflection of wave

→ all energy propagating through

$$n_3$$

→ when  $R=0$ , what is  $\boxed{I, n_1, n_2, n_3}$  ?

Incident wave has orthogonal E field, H field

Any wave  $-90^\circ \varphi$  lag  $\rightarrow e^{j\omega t} = e^{j(\omega t + \varphi)} = -je^{j\omega t}$   
 $e^{j\omega t} e^{j\varphi}$

$R=0$ , when  $E_5 = 0$

$$1 \quad \left\{ \begin{array}{l} E_1 + E_S = E_2 - jE_4 \quad 1 \\ H_1 + H_S = H_2 - jH_4 \quad 2 \end{array} \right. \quad \underline{\text{Maxwell's Eqns}}$$

$$E_1 = n_1 H_1$$

$$2 \quad \left\{ \begin{array}{l} E_4 - jE_2 = E_3 \quad 3 \\ H_4 - jH_2 = H_3 \quad 4 \end{array} \right. \quad \begin{array}{l} E_2 = n_2 H_2 \\ E_3 = n_2 H_3 \end{array}$$

Right hand rule  $\rightarrow$   $E_4 = n_2 (-H_4)$   
 $E_S = -n_1 H_S$

$$\text{Eqn. 2} \quad \frac{E_1}{n_1} - \frac{E_S}{n_1} = \frac{E_2}{n_2} - j \left( -\frac{E_4}{n_2} \right)$$

$$\frac{E_1}{n_1} - \frac{E_S}{n_1} = \frac{E_2}{n_2} + j \frac{E_4}{n_2}$$

$$E_1 - E_S = \frac{n_1}{n_2} (E_2 + jE_4) \quad 5.$$

$$-\frac{E_4}{n_2} - j \frac{E_3}{n_2} = \frac{E_3}{n_3} \quad . \quad E_3 = \frac{n_3}{n_2} (-E_4 - jE_2) \quad 6.$$

Take Eqn. 2 and 5

$$E_2 + jE_4 = jE_3$$
$$E_1 - E_5 = \frac{n_1}{n_2} (E_2 + jE_4)$$
$$E_1 - E_5 = \frac{n_1}{n_2} \cdot jE_3 \quad 7.$$

Take Eqn 1, 6

$$E_1 + E_5 = E_2 - jE_4$$

$$-j \frac{E_2 - E_4}{n_2} = \frac{E_3}{n_3} \quad \left. \right\} \text{ multiply with } j = \frac{E_2 - jE_4}{n_2} = \frac{jE_3}{n_3}$$

$$E_2 - jE_4 = j \frac{n_2}{n_3} \cdot E_3$$

$$E_1 + E_5 = j \frac{n_2}{n_3} \cdot E_3 \quad 8.$$

Add Eqn 7 + 8

$$2E_1 = \frac{n_1}{n_2} jE_3 + \frac{n_2}{n_3} jE_3$$

$$= jE_3 \left( \frac{n_1}{n_2} + \frac{n_2}{n_3} \right)$$

$$\therefore jE_3 = \frac{2E_1}{\left( \frac{n_1}{n_2} + \frac{n_2}{n_3} \right)} = \frac{2n_2 n_3}{n_1 n_3 + n_2^2} \quad 9.$$

Subtract Eqn 8 - 7

$$-2E_5 = jE_3 \left( \frac{n_1}{n_2} - \frac{n_2}{n_3} \right)$$

$$jE_3 = \frac{2n_2 n_3}{n_2^2 - n_1 n_3} \cdot E_5 \quad 10.$$

Equate Eqn 9 and 10

$$\frac{2 n_1 n_3}{n_2^2 - n_1 n_3} E_S = \frac{2 n_2 n_3}{n_1 n_3 + n_2^2} E_1$$

$$E_S = \frac{n_2^2 - n_1 n_3}{n_2^2 + n_1 n_3} E_1$$

this can't be 0

setting  $E_S = 0 \quad \therefore$

$$n_2^2 = n_1 n_3$$

$$n_2 = \sqrt{n_1 n_3}$$

# Exam 22/23

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## 1.3 Intrinsic Impedance

Maxwell Eqns:

$$\begin{aligned} \vec{\nabla} \times \vec{E} &= -\mu \frac{\partial \vec{H}}{\partial t} \\ \vec{\nabla} \times \vec{H} &= \vec{J}_+ + \epsilon \frac{\partial \vec{E}}{\partial t} \end{aligned} \quad \left. \begin{array}{l} \\ \end{array} \right\} \text{Take the curl:}$$

$$\vec{\nabla} \times \vec{\nabla} \times \vec{H} = \epsilon \frac{\partial (\vec{\nabla} \times \vec{E})}{\partial t}$$

Poisson's Equation:

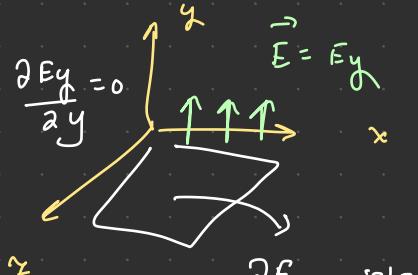
$$\vec{\nabla}(\vec{\nabla} \cdot \vec{H}) - \vec{\nabla}^2 \vec{H} = \frac{\epsilon \sigma (\vec{\nabla} \times \vec{E})}{\partial t} \quad \text{Remember } \vec{\nabla} \cdot \vec{H} = 0$$

$$-\vec{\nabla}^2 \vec{H} = \frac{\epsilon \sigma (-\mu \frac{\partial \vec{H}}{\partial t})}{\partial t} = -\mu \frac{\epsilon \sigma^2 \vec{H}}{\partial t^2}$$

\*  $-\vec{\nabla}^2 \vec{E} = -\epsilon \mu \frac{\partial^2 \vec{E}}{\partial t^2}$

Polarise wave:  $E_x = E_z = 0$

$$\frac{\partial^2 E_y}{\partial y^2} = \frac{\partial^2 E_y}{\partial z^2} = 0$$



Expand  $\underbrace{\vec{\nabla}^2 \vec{E}}_{\text{exists}} = \frac{\partial^2 \vec{E}}{\partial t^2}$  where  $c = \frac{1}{\sqrt{\mu \epsilon}}$

$$\frac{\partial^2 E_x}{\partial x^2} + \frac{\partial^2 E_x}{\partial y^2} + \frac{\partial^2 E_x}{\partial z^2} = \mu \epsilon \frac{\partial^2 E_x}{\partial t^2}$$

⋮      ⋮      ⋮      ⋮

$$\frac{\partial^2 E_y}{\partial x^2} + \frac{\partial^2 E_y}{\partial y^2} + \frac{\partial^2 E_y}{\partial z^2} = \mu \epsilon \frac{\partial^2 E_y}{\partial t^2}$$

$$\therefore \frac{\partial^2 E_y}{\partial x^2} = \mu \epsilon \frac{\partial^2 E_y}{\partial t^2}$$

$$E_y = E_m \cos \beta (x - vt) \rightarrow \beta v = \omega$$

$$\therefore E_m \cos (\omega t - \beta x)$$

Step: To find  $\vec{B}$  for system.

$$\frac{\partial \vec{B}}{\partial t} = -\vec{\nabla} \times \vec{E}$$

$$\begin{vmatrix} i & j & k \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A_x & A_y & A_z \end{vmatrix}$$

$$\frac{\partial B_y}{\partial t} = \underbrace{\frac{\partial E_z}{\partial y}}_{=0} - \frac{\partial E_y}{\partial z}$$

$$\frac{\partial B_z}{\partial t} = - \left( \frac{\partial E_x}{\partial z} - \frac{\partial E_z}{\partial x} \right) = 0$$

$$\frac{\partial B_x}{\partial t} = \cancel{\frac{\partial E_x}{\partial y}}^0 - \frac{\partial E_y}{\partial x}$$

$$\frac{\partial B_x}{\partial t} = - \frac{\partial E_y}{\partial x} \quad \text{where } E_y = \frac{1}{2} E_m \cos(\omega t - \beta x)$$

$$= - [ - \beta E_m (-\sin(\omega t - \beta x)) ]$$

$$\frac{\partial B_x}{\partial t} = -\beta E_m \sin(\omega t - \beta x)$$

$$\text{Integrate: } B_z = \int -B_E m \sin(\omega t - \beta x) dt$$

$$B_z = \frac{\beta E_m}{\omega} \cos(\omega t - \beta x)$$


---

$$\rightarrow E_y = E_m (\cos(\omega t - \beta x))$$

$$\rightarrow B_z = \underbrace{\frac{\beta}{\omega}}_{\text{radsim}} (E_y) \quad \therefore \frac{B_z}{\omega} = \frac{m}{s} = \frac{1}{v}$$

$\downarrow$   
 rads  
 $s$

$\frac{1}{v}$  = phase velocity

$$\rightarrow \mu H_z = \frac{1}{v} E_y \Rightarrow \mu \cdot v \cdot H_z = E_y$$

$$\text{know that } v = \frac{1}{\sqrt{\mu \epsilon}} \quad \therefore \quad \mu \cdot \frac{1}{\sqrt{\mu \epsilon}} \cdot H_z = E_y$$

$$\boxed{E_y = \sqrt{\frac{\mu}{\epsilon}} \cdot H_z}$$

Q2.

## Energy densities

To find: Total power emanating from closed surface:

Electric field:  $\frac{\vec{D} \cdot \vec{E}}{2} \text{ J/m}^3$

Magnetic field:  $\frac{\vec{B} \cdot \vec{H}}{2} \text{ J/m}^3$

Closed surface: Power out = - rate of change of stored energy.

$$\oint_S \vec{P} \cdot d\vec{s} = -\frac{\partial}{\partial t} \int_V \left( \underset{\text{density}}{\vec{E}} + \underset{\text{magnetic}}{\vec{H}} \right) \cdot dV$$

$$= \oint_S \vec{P} \cdot d\vec{s} = -\frac{\partial}{\partial t} \int_V \left( \frac{\vec{D} \cdot \vec{E}}{2} + \frac{\vec{B} \cdot \vec{H}}{2} \right) dV \quad \left\{ \begin{array}{l} \text{Remember:} \\ \vec{P} \vec{H} = \vec{B} \\ \sum \vec{E} = \vec{D} \end{array} \right.$$

$$\Rightarrow - \int_V \left( \frac{\partial \vec{H}}{\partial t} \cdot \vec{P} \vec{H} + \frac{\partial \vec{E}}{\partial t} \cdot \vec{P} \vec{E} \right) dV \quad \underbrace{\vec{H} \vec{H} \cdot \vec{H}}_{\vec{H} \vec{H}^2} = \vec{H} \vec{H}^2$$

$$= \int_V \left( \frac{\partial \vec{H}}{\partial t} \cdot \vec{H} + \frac{\partial \vec{E}}{\partial t} \cdot \vec{E} \right) dV \quad \text{chain rule.}$$

$$= \int_V \left( \vec{H} \cdot (\nabla \times \vec{E}) - \vec{E} \cdot (\nabla \times \vec{H}) \right) dV$$

$$\nabla \cdot (\vec{M} \times \vec{N}) = \vec{N} \cdot (\nabla \times \vec{M}) - \vec{M} \cdot (\nabla \times \vec{N}) \therefore \int_V \nabla \cdot (\vec{E} \times \vec{H}) dV$$

Gauss Theorem:  $\oint_S \vec{P} \cdot d\vec{s} = \oint_S (\vec{E} \times \vec{H}) \cdot d\vec{s}$

Q2.2Cylindrical Conductor

$$\sigma = 9 \times 10^3 \text{ S/m}$$

$$\sigma = 5.8 \times 10^7 \text{ S}^{-1} \text{ m}^{-1}$$

$$l = 1.5 \text{ m}$$

$$V = 12 \times 10^{-3} \text{ V}$$

Poynting Vector =  $\vec{E} \times \vec{H}$

$$\text{E field} = \frac{V}{l} = \frac{12 \times 10^{-3}}{1.5} = 8 \times 10^{-3} = \frac{8 \text{ mV}}{\text{m}}$$

Current Density:  $\vec{J} = \sigma \vec{E} \dots$

$$(5.8 \times 10^7) (8 \times 10^{-3}) = 464,000 \text{ A/m}^2$$

Trying to calculate:  $\vec{E} \times \vec{H}$  if  $R = \frac{P}{lA} \Rightarrow R = \frac{l}{2\pi a^2}$

$$\vec{H} = \frac{r \vec{J}}{2} = \frac{r I}{2\pi a^2} \quad \left. \right\} \text{ This varies} \rightarrow \sigma = \text{from centre}$$

$$\text{at } r = 4 \text{ mm} \rightarrow \frac{r J}{2} \Rightarrow 928$$

$$r = 2 \text{ mm} \rightarrow$$

$$r = 0 \text{ mm} \rightarrow$$

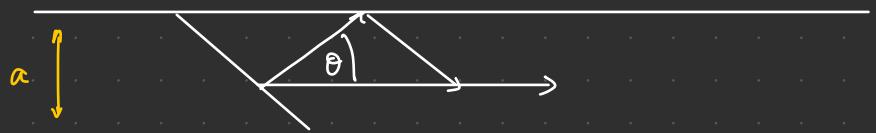
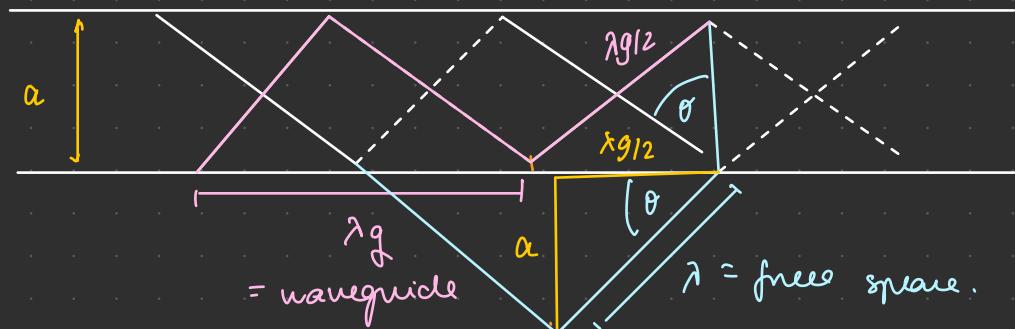
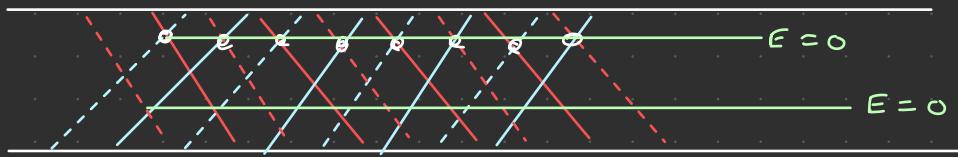
In a cylinder

$$H_r = \frac{\sigma I}{2\pi a^2}$$

$$E_r = \frac{J}{\sigma} = \frac{1}{\pi a^2 \sigma}$$

# Waveguides

4.1



$$\cos \theta = \frac{a}{\lambda} = \frac{\lambda g}{2}$$

$\lambda g$

$(\theta)$

$a$

$\Rightarrow \frac{\lambda}{\lambda g}$

$$\tan \theta = \frac{a}{\lambda g/2}$$

$a$

$\lambda g/2$

$= \frac{\lambda g/2}{a} \approx \frac{\lambda g}{2a}$

$$\cos \theta = \frac{\lambda}{\lambda g}, \quad \tan \theta = \frac{\lambda g}{2a}$$

$\sin \theta$  in terms of  $\cos \theta, \tan \theta$ :

$$\tan \theta = \frac{\sin \theta}{\cos \theta} = \sin \theta \cdot \frac{\lambda}{\lambda g} = \frac{\lambda g}{2a}$$

$$\therefore \frac{\lambda g}{2a} = \frac{\lambda g \cdot \sin \theta}{\lambda} \Rightarrow \frac{1}{2a} = \frac{\sin \theta}{\lambda} \Rightarrow \sin \theta = \frac{\lambda}{2a}$$

$$\Rightarrow \lambda = 2a \sin \theta$$

$$\lambda_c \text{ when } \sin \theta = 1, \quad \boxed{\lambda = 2a}$$

$$\text{if } a = 0.03m \therefore 2(0.03) = 0.06m$$

$$f_c = \frac{c}{\lambda_c} = \frac{3 \times 10^8}{0.06} = 50 \times 10^8 \text{ Hz}$$

4.2       $\lambda, \lambda_g, \lambda_c :$

$$\sin^2 \theta + \cos^2 \theta = 1$$

$$\downarrow \qquad \qquad \downarrow$$

$$\sin \theta = \frac{\lambda}{2a} \quad \cos \theta = \frac{\lambda g}{\lambda}$$

$$\left( \frac{\lambda}{2a} \right)^2 + \left( \frac{\lambda}{\lambda g} \right)^2 = 1 \quad * \text{ Divide across by } \lambda^2$$

$$\underbrace{\left( \frac{1}{2a} \right)^2}_{\text{this}} + \left( \frac{1}{\lambda g} \right)^2 = \frac{1}{\lambda^2}$$

Recall that

$$\text{this} = \lambda_c \quad \therefore \quad \left( \frac{1}{\lambda_c} \right)^2 + \left( \frac{1}{\lambda g} \right)^2 = \frac{1}{\lambda^2}$$

Q3  $\theta$  component zigzag:

given that  $f = 8 \times 10^{12} \text{ Hz}$

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8}{8 \times 10^{12}} = 3.75 \times 10^{-5} \text{ m}$$

Knowing  $\sin \theta = \frac{\lambda}{da} \therefore a = 0.03 \text{ m} \therefore$

$$= \frac{(3.75 \times 10^{-5})}{0.03}$$



$$\sin^{-1}(\text{ans}) = 36.86^\circ$$

4.4

$$V_g = c \cdot \cos \theta = (3 \times 10^8)(\cos(38.86)) \\ = 2.34 \times 10^8 \text{ m/s}$$

$$V_{ph} = \frac{c}{\cos \theta} = \frac{3 \times 10^8}{\cos(38.86)} = 3.84 \times 10^8 \text{ m/s}$$

Expectation:

$$\begin{cases} V_g < c \\ V_{ph} > c \end{cases} \quad \left. \begin{array}{l} V_g \times V_{ph} = c^2 \end{array} \right\}$$

# To verify Maxwell's Eqns

$$1. \nabla \cdot \vec{B} = 0$$

$$2. \nabla \cdot \vec{E} = \frac{\rho}{\epsilon_0}$$

$$3. \nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$

$$4. \nabla \times \vec{H} = \vec{J} + \frac{\partial \vec{B}}{\partial t}$$

} find  $\rho$  and  $\vec{J}$   
Eqn. 4 :