

**K. J. SOMAIYA COLLEGE OF ENGINEERING**  
**DEPARTMENT OF ELECTRONICS ENGINEERING**  
**ELECTRONIC CIRCUITS**  
**Design of single-stage Amplifier**

**Design 1:**

Design a single stage RC coupled BJT amplifier for the following specifications:

$$V_{orms} = 3V, f_L \leq 15Hz, S \leq 10, |A_V| \geq 120$$

Calculate  $A_V$ ,  $Z_i$  &  $Z_o$  of the amplifier you have designed.

**Solution:**

**1. Given Data:**

$$|A_V| \geq 120, S \leq 10, V_{orms} = 3V, f_L \leq 15Hz$$

**2. Circuit Diagram and Selection of Transistor:**

Transistor BC147B is selected with following specifications:

$$h_{FE(min)} = 200, h_{FE(typ)} = 290, h_{FE(max)} = 450$$

$$h_{ie} = 4.5k\Omega, V_{CE(sat)} = 0.25V$$

$$h_{fe(min)} = 240, h_{fe(typ)} = 330, h_{fe(max)} = 500$$

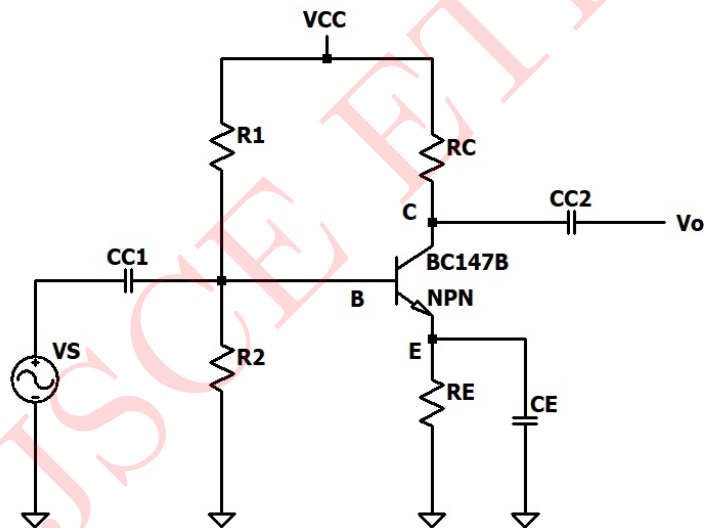


Figure 1: Circuit 1

**3. Selection of  $R_C$**

$$|A_V| = \frac{h_{fe(min)} \times R_C}{h_{ie}}$$

$$120 = \frac{240 \times R_C}{4.5k\Omega}$$

$$R_C = 2.25k\Omega$$

We select a higher standard value of  $R_C$  to increase the gain

$$\therefore \text{Select } R_C = 2.4k\Omega, \frac{1}{4}W$$

**4. Selection of Qpoint ( $V_{CE}$  &  $I_C$ )**

$$V_{o_{peak}} = V_{o_{rms}} \times \sqrt{2} = 3 \times \sqrt{2} = \mathbf{4.2426V}$$

$$V_{CE} = 1.5 [V_{o_{peak}} + V_{CE(sat)}]$$

The value is multiplied by 1.5 to take care of saturation voltages, variation in resistance, variation in supply voltage and device parameter variation.

$$V_{CE} = 1.5(4.2426 + 0.25) = \mathbf{6.74V}$$

$$I_{o_{peak}} = \frac{V_{o_{peak}}}{R_C} = \frac{4.2426V}{2.4k\Omega} = \mathbf{1.767mA}$$

$$I_C \geq I_{o_{peak}}$$

[For undistorted output signal]

Select  $I_C = \mathbf{2mA}$

$V_{CC} \geq 2V_{CE}$  [Selecting Qpoint at the center of DC load line for maximum output swing]

$$V_{CC} \geq 2 \times 4.2426V$$

$$V_{CC} \geq 8.4852V$$

Select  $V_{CC} = \mathbf{9V}$

## 5. Selection of $R_E$

For proper operation,

$$V_{RE} = 10\% \text{ of } V_{CC}$$

$$V_{RE} = 0.1V_{CC} = 0.1 \times 9 = \mathbf{0.9V}$$

Now,

$$V_{RE} = R_E \times I_E$$

$$R_E = \frac{V_{RE}}{I_E} \approx \frac{V_{RE}}{I_C}$$

$$R_E \approx \frac{0.9V}{2mA}$$

$$R_E \approx 450\Omega$$

Selecting a lower standard value

Select  $R_E = \mathbf{420\Omega, \frac{1}{4}W}$

## 6. Selection of $R_1$ & $R_2$

$$S = \frac{1 + h_{FE(max)}}{1 + h_{FE(max)} \left[ \frac{R_E}{R_E + R_B} \right]}$$

$$\beta = h_{FE(max)} = 450 \text{ \& } R_B = R_1 \parallel R_2$$

$$10 = \frac{1 + 450}{1 + 500 \left[ \frac{420}{420 + R_B} \right]}$$

$$R_{TH} = R_B = 3.8657k\Omega = \frac{R_1 R_2}{R_1 + R_2} \quad \text{.....(1)}$$

$$V_{TH} = V_B = \frac{R_2 V_{CC}}{R_1 + R_2} \quad \dots(2)$$

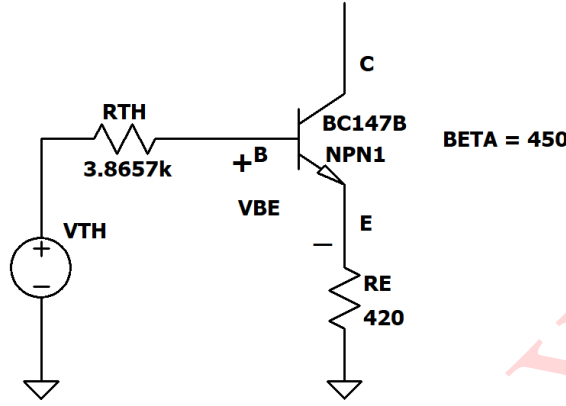


Figure 2: Thevenin's Equivalent Circuit

Applying KVL at B-E loop,

$$V_{TH} - I_B R_{TH} - V_{BE} - I_E R_E = 0$$

$$V_B - I_B R_B - V_{BE} - I_E R_E = 0$$

$$V_B = V_{BE} + \frac{I_C}{\beta} R_B + I_C R_E$$

$$= 0.7 + \frac{2mA}{500} \times 3.8657k\Omega + 2mA \times 420\Omega$$

$$= \mathbf{1.557V}$$

From equation (2),

$$V_B = 1.557 = \frac{R_2}{R_1 + R_2} \times 9V$$

$$\frac{R_2}{R_1 + R_2} = 0.173 \quad \dots(3)$$

Substituting equation (3) in equation (1),

$$R_1 \times (0.173) = 3.8657k\Omega$$

$$R_1 = 22.345k\Omega$$

Selecting a higher standard value so that circuit draws minimum current,

$$\text{Select } R_1 = \mathbf{24k\Omega}, \frac{1}{4}W$$

From equation (3),

$$\frac{R_2}{24k\Omega + R_2} = 0.173$$

$$R_2 = 5.02k\Omega$$

Selecting a lower standard value,

$$\text{Select } R_2 = \mathbf{4.7k\Omega}, \frac{1}{4}W$$

### 7. Selection of $C_E$

To ensure complete bypass of  $R_E$

$$X_{CE} < R_E$$

$$X_{CE} = \frac{R_E}{10} = 0.1R_E$$

$$\text{i.e } \frac{1}{2\pi f_L C_E} = 0.1R_E$$

$$\begin{aligned} C_E &= \frac{1}{2\pi f_L \times 0.1R_E} \\ &= \frac{1}{2\pi \times 15 \times 0.1 \times 420} \\ &= 252.62\mu F \end{aligned}$$

Selecting a higher standard value

Select  $C_E = 270\mu F/18V$

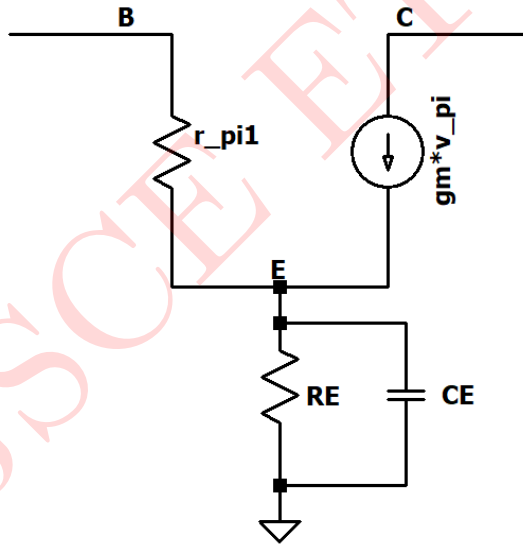


Figure 3: Small Signal Equivalent Circuit for CE

### 8. Selection of $C_{C_1}$

$$C_{C_1} = \frac{1}{2\pi R_{eq} f_L}$$

$$R_{eq} = R_1 \parallel R_2 \parallel h_{ie}$$

$$= 24k\Omega \parallel 4.7k\Omega \parallel 4.5k\Omega$$

$$= 3.93k\Omega \parallel 4.5k\Omega$$

$$= 2.097k\Omega$$

$$[h_{ie} = r_\pi]$$

$$C_{C_1} = \frac{1}{2\pi \times 2.097k\Omega \times 15Hz} = 5.059\mu F$$

Selecting a higher standard value

Select  $C_{C_1} = 5.6\mu\text{F}/18\text{V}$

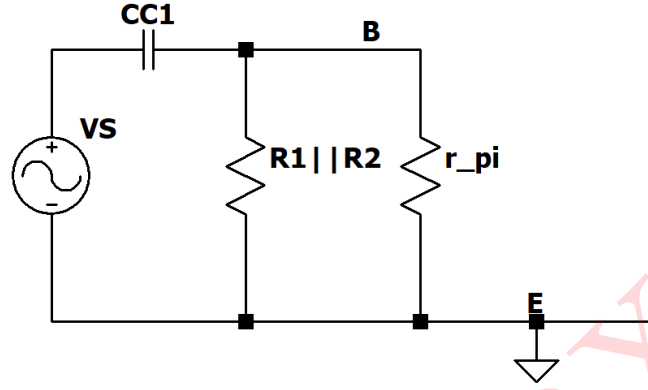


Figure 4: Small Signal Equivalent Circuit for  $C_{C_1}$

### 9. Selection of $C_{C_2}$

$$C_{C_2} = \frac{1}{2\pi R_{eq} f_L}$$

$$= \frac{1}{2\pi R_C f_L}$$

$$= \frac{1}{2\pi \times 2.4k\Omega \times 15Hz}$$

$$= 4.42\mu\text{F}$$

[Since,  $R_{eq} = R_C$ ]

Selecting a higher standard value,

Select  $C_{C_2} = 4.7\mu\text{F}/18\text{V}$

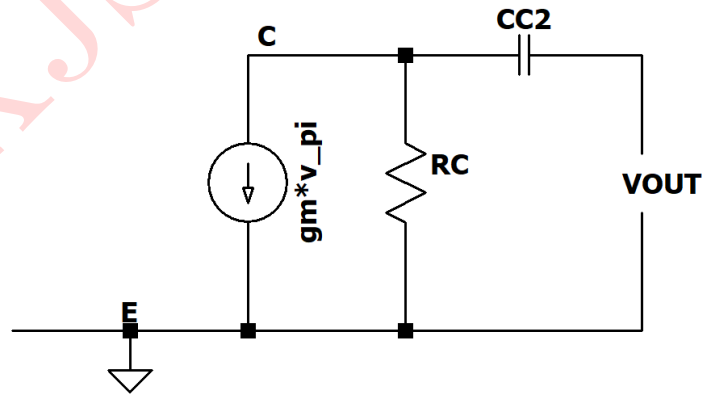


Figure 5: Small Signal Equivalent Circuit for  $C_{C_2}$

### 10. Complete Designed Circuit:

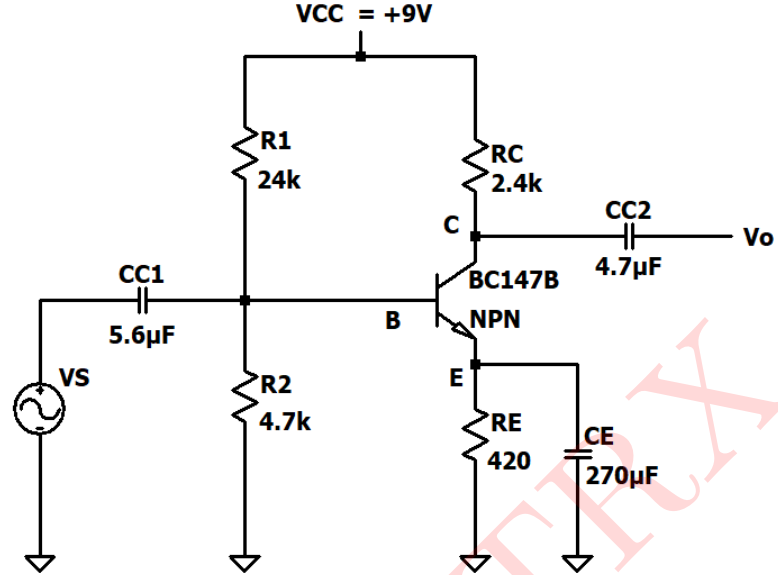


Figure 6: Designed Circuit

### 11. Calculation of $A_V$ , $Z_i$ & $Z_o$

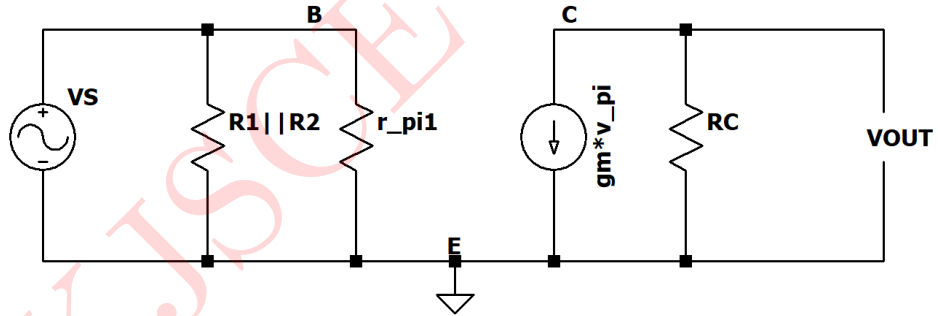


Figure 7: Small Signal Equivalent Circuit

Small Signal Voltage Gain ( $A_V$ ) =  $-g_m R_C$

$$g_m = \frac{I_{CQ}}{V_T} = \frac{2mA}{0.026V} = 76.923mA/V$$

$$\therefore A_V = -(76.923mA/V) \times (2.4k\Omega) = -184.615$$

Input Impedance ( $Z_i$ ):

$$Z_i = R_1 \parallel R_2 \parallel r_\pi$$

$$\begin{aligned} r_\pi &= \frac{\beta \times V_T}{I_{CQ}} \\ &= \frac{450 \times 0.026V}{2mA} \\ &= 5.85k\Omega \end{aligned}$$

$$Z_i = 24k\Omega \parallel 4.7k\Omega \parallel 5.85k\Omega = \mathbf{2.35k\Omega}$$

Output Impedance ( $Z_o$ ):

$$Z_o = R_C$$

$$\therefore Z_o = \mathbf{2.4k\Omega}$$

## SIMULATED RESULTS

The above circuit is simulated in LTspice and results are presented below:

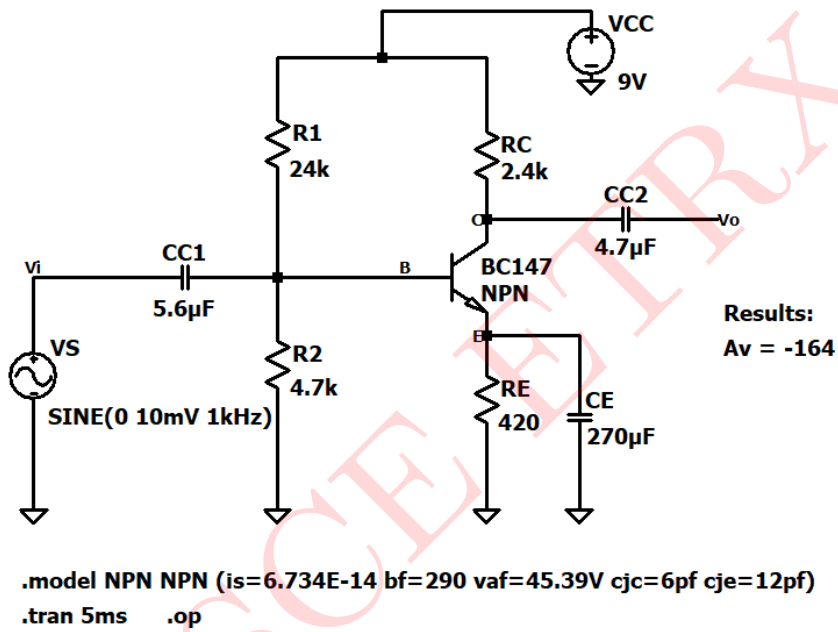
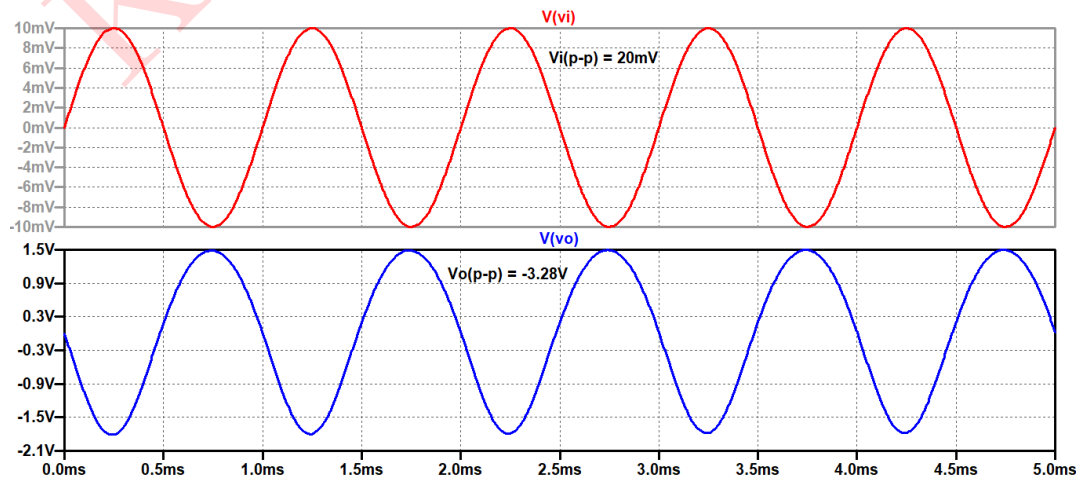


Figure 8: Circuit Schematic

The input and output waveform are shown in figure 8



**Comparison of Theoretical and Simulated results:**

Parameters	Theoretical	Simulated
$ A_V $	$> 120$	164
$I_{CQ}$	$2mA$	$1.96315mA$

Table 1: Design 1

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