# K. J. SOMAIYA COLLEGE OF ENGINEERING DEPARTMENT OF ELECTRONICS ENGINEERING ELECTRONIC CIRCUITS

SINGLE STAGE FET AMPLIFIER

 $23^{rd}$  June, 2020 Numericals

1. For the circuit shown in Figure 1, find

- a)  $I_{DQ}$
- b)  $V_{SQ}$
- c)  $R_i$
- d)  $R_o$
- e)  $A_V$

Given:  $I_{DSS} = 10 \text{ mA}, V_P = -4 \text{ V}$ 

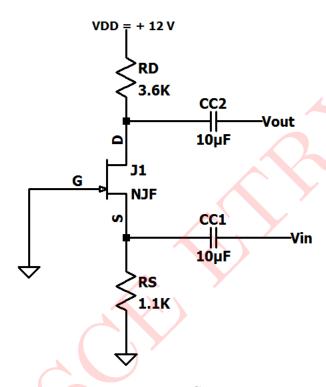


Figure 1: Circuit 1

### Solution:

The above circuit is a common gate amplifier.

## DC Analysis:

The capacitors act as open circuit.

$$f = 0, \quad \therefore X_C = \frac{1}{2\pi f C} = \infty$$

Applying KVL to input gate-source loop

$$-V_{GS} - I_D R_S = 0$$

$$\therefore V_{GS} = -I_D R_S = -1100 I_D$$

For JFET, 
$$I_D = I_{DSS} \left( 1 - \frac{V_{GS}}{V_P} \right)^2$$

$$\therefore V_{GS} = -1100 \times I_{DSS} \left( 1 - \frac{V_{GS}}{V_P} \right)^2$$

$$V_{GS} = -1100 \times 10 \times 10^{-3} \left( 1 + \frac{V_{GS}}{4} \right)^{2}$$

$$\therefore V_{GS} = -11\left(1 + \frac{V_{GS}}{2} + \frac{V_{GS}^2}{16}\right)$$

$$\therefore 0.6875V_{GS}^2 + 6.5V_{GS} + 11 = 0$$

$$\therefore V_{GS} = -2.2079 \text{ V or } V_{GS} = -7.2466 \text{ V}$$

Since  $V_{GS} > V_P$ ,  $V_{GS} = -2.2079 \text{ V}$ 

$$I_D = \frac{V_{GS}}{-1100} = 2.00718 \text{ mA}$$

## Small-signal parameters:

$$g_m = \frac{-2I_{DSS}}{V_P} \times \left(1 - \frac{V_{GS}}{V_P}\right)$$
$$\therefore g_m = \frac{-2 \times 10 \times 10^{-3}}{-4} \times \left(1 - \frac{2.2079}{4}\right) = 2.25 \text{ mA/V}$$

The small signal equivalent circuit is shown in Figure 2

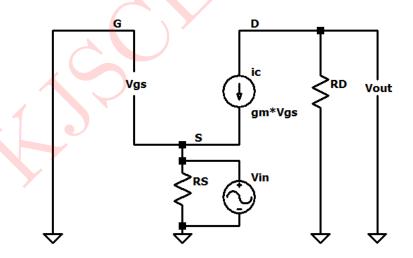


Figure 2: Small-signal equivalent circuit

Input resistance 
$$R_i = R_S || \frac{1}{g_m} = 316.5 \Omega$$

Output resistance  $R_o=R_D=3.6~{\rm k}\Omega$ 

Voltage gain  $A_V = g_m R_D = 2.25 \times 10^{-3} \times 3.6k$ 

$$A_V = 8.1$$

## SIMULATED RESULTS:

Above circuit is simulated using LTspice and the results are presented below:

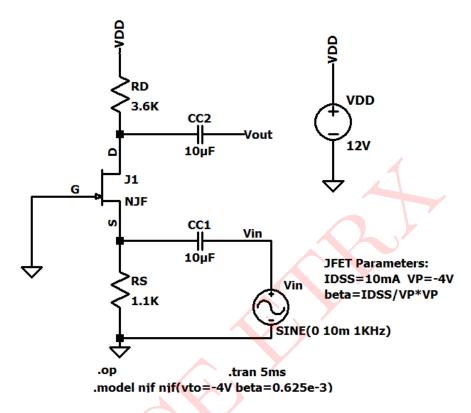


Figure 3: Circuit schematic

The waveforms for input and output voltage are shown in Figure 4.

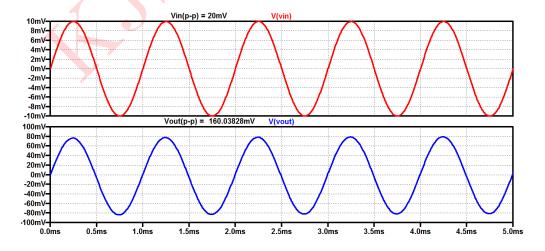


Figure 4: Input and output voltage waveforms

# Comparison of theoretical and simulated values:

Parameters	Theoretical	Simulated
$I_{DQ}$	$2.00718~\mathrm{mA}$	$2.00721~\mathrm{mA}$
$V_{GSQ}$	-2.2079  V	-2.20793  V
$A_V$	8.1	8.001914

Table 1: Numerical 1



- 2. For the circuit shown in Figure 5, calculate
  - a) Z
  - $\dot{b}$   $Z_o$
  - c)  $V_{GSQ}$
  - d)  $I_{DQ}$
  - e)  $A_V$

Given:  $k_n=1=0.2mA/V^2,\,V_{GS(th)}=3\mathrm{V},\,r_d=100~\mathrm{k}\Omega$ 

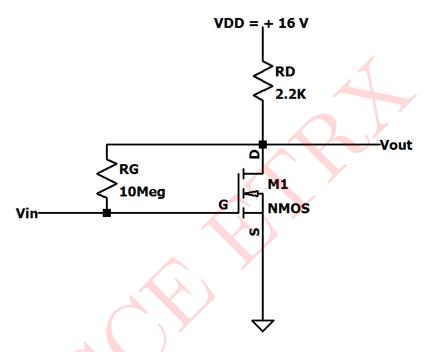


Figure 5: Circuit 2

Solution: The above circuit is a feedback amplifier circuit.

#### DC Analysis:

The capacitors act as open circuit.  $f = 0, :: X_C = \frac{1}{2\pi fC} = \infty$ 

$$k_n = \frac{I_D}{[V_{GS} - V_{GS(th)}]^2}$$

$$\therefore 0.2 \times 10^{-3} = \frac{I_D}{[V_{GS} - 3]^2}$$

$$\therefore I_D = 0.2 \times 10^{-3} \times [V_{GS} - 3]^2 \qquad \dots (1$$

$$V_{GS} = V_{DD} - I_D R_D$$

$$\therefore V_{GS} = 16 - (0.2 \times 10^{-3} \times [V_{GS} - 3]^2 \times 2.3k)$$

$$\therefore V_{GS} = 16 - 0.44[V_{GS} - 3]^2$$

$$\therefore V_{GS} = 16 - 0.44V_{GS}^2 + 2.64V_{GS} - 3.96$$

$$\therefore 0.44V_{GS}^2 - 1.64V_{GS} - 12.04 = 0$$

$$\therefore V_{GS} = 7.4167325 \text{ V}$$
 or  $V_{GS} = -3.68945 \text{ V}$ 

Since 
$$V_{GS} > V_{TN}$$
, :  $V_{GSQ} = 7.4167235 \text{ V}$ 

From equation (1), 
$$I_{DQ} = 0.2 \times 10^{-3} \times [7.4167235 - 3]^2$$

$$\therefore I_{DQ} = \mathbf{3.901489} \ \mathbf{mA}$$

## Small-signal parameters:

$$g_m = 2k_n(V_{GSQ} - V_{TN}) = 2 \times 0.2 \times 10^{-3} \times (7.4167235 - 3)$$

$$\therefore g_m = 1.766689 \frac{mA}{V}$$

$$r_d = \frac{1}{\lambda I_{DQ}}$$

$$\therefore \lambda = \frac{1}{r_d I_{DQ}} = \frac{1}{100 \times 10^3 \times 3.901489 \times 10^{-3}}$$

$$\therefore \lambda = 2.5631239 \times 10^{-3} V^{-1}$$

The mid-frequency AC equivalent circuit is shown in Figure 6

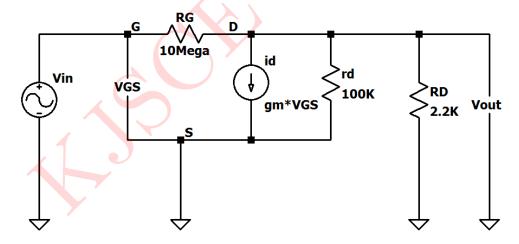


Figure 6: Mid-frequency equivalent circuit

Applying KCL at D, 
$$I_i = g_m V_{gs} + \frac{V_o}{r_d || R_D}$$

$$I_i = \frac{V_i - V_o}{R_G} \text{ and } V_i = V_{gs}$$

$$\therefore \frac{V_i - V_o}{R_G} = g_m V_i + \frac{V_o}{r_d || R_D}$$

$$V_i \left(\frac{1}{R_G} - g_m\right) = V_o \left(\frac{1}{R_G} + \frac{1}{r_d || R_D}\right)$$

$$\frac{1}{R_G} \text{ is very small, hence } \left(\frac{1}{R_G} - g_m\right) \approx -g_m$$

$$\therefore V_i \times -g_m = V_o \left(\frac{1}{R_G} + \frac{1}{r_d || R_D}\right)$$

$$\therefore A_V = \frac{V_o}{V_i} = \frac{-g_m}{\left(\frac{1}{R_G || r_d || R_D}\right)}$$

$$\therefore A_V = -g_m (R_G || r_d || R_D) = 1.766689 \times 10^{-3} \times (10M || 100k || 2.2k)$$

$$\therefore A_V = -3.802154$$

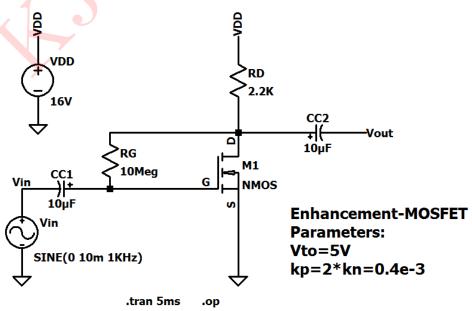
$$Z_i = R_G = 10 \text{ M}\Omega$$

### SIMULATED RESULTS:

 $\therefore Z_o = 2.152136 \text{ k}\Omega$ 

 $Z_o = R_G ||r_d|| R_D = 10M ||100k|| 2.2k$ 

Above circuit is simulated using LTspice and the results are presented below:



.model nmos nmos(vto=3V kp=0.4e-3 lambda=2.56e-3)

Figure 7: Circuit schematic

The input and output waveforms are shown in Figure 8

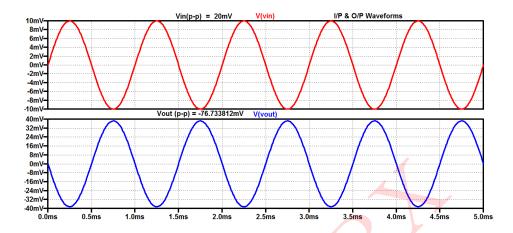


Figure 8: Input and output waveforms

## Comparison of theoretical and simulated values:

Parameters	Theoretical	Simulated
$I_{DQ}$	3.901489  mA	3.90149  mA
$V_{GSQ}$	7.4167235 V	7.41672 V
$A_V$	-3.802154	-3.84098

Table 2: Numerical 2