

**K. J. SOMAIYA COLLEGE OF ENGINEERING**  
**DEPARTMENT OF ELECTRONICS ENGINEERING**  
**ELECTRONIC CIRCUITS**  
**DIFFERENTIAL AMPLIFIER CIRCUITS**

16<sup>th</sup> July, 2020

Numerical

1. Determine the following for the circuit shown in Figure 1

Assume  $\beta_1 = \beta_2 = 100$

- Name the circuit
- Current flowing through resistors  $R_{S1}$ ,  $R_{S2}$  and  $R_{C2}$
- Find  $V_{C1}$ ,  $V_{C2}$ ,  $V_{CE1}$  and  $V_{CE2}$
- Differential voltage gain
- Common mode gain
- CMRR in dB

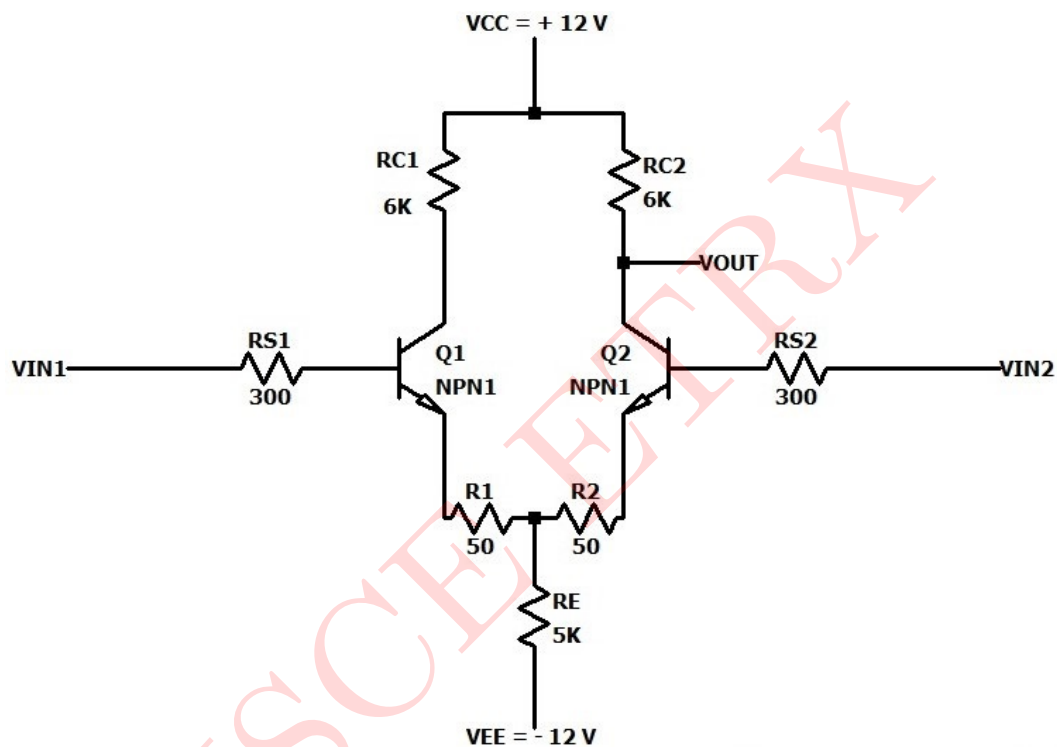


Figure 1: Circuit 1

**Solution:**

- a) The above circuit is a DIUO differential amplifier.

b) DC analysis:

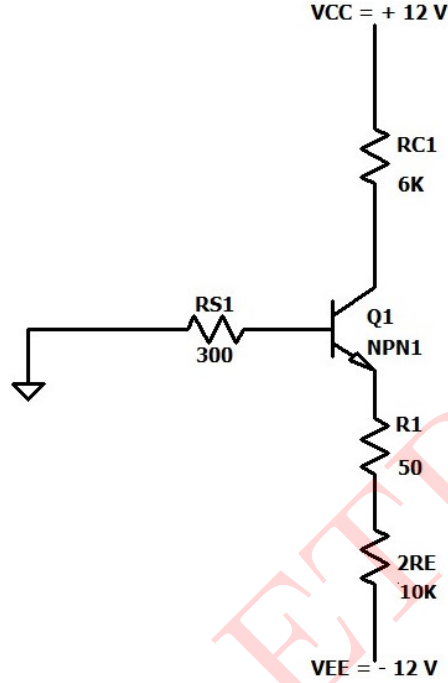


Figure 2: DC equivalent circuit

Applying KVL to the input base-emitter loop, we get

$$V_{EE} - I_{BQ}R_{S1} - V_{BE} - (1 + \beta)R_1I_{BQ} - 2(1 + \beta)R_EI_{BQ} = 0$$

$$I_{BQ} = \frac{V_{EE} - V_{BE}}{R_{S1} + (1 + \beta)(R_1 + 2R_E)}$$

$$\therefore I_{BQ} = 11.129 \mu\text{A}$$

$$I_{CQ1} = I_{CQ2} = I_{CQ} = \beta \times I_{BQ} = \mathbf{1.112916 \text{ mA}}$$

c) Applying KVL at output collector-emitter loop, we get

$$V_{CEQ} = V_{CC} + V_{EE} - I_{CQ}(R_C + 2R_E + R_1)$$

$$V_{CEQ} = 12 + 12 - (1.12916 \times 10^{-3}) \times (6k + 10k + 50)$$

$$\therefore V_{CEQ} = V_{CEQ1} = V_{CEQ2} = \mathbf{6.137955 \text{ V}}$$

$$\text{Q-point } (I_{CQ1}, V_{CEQ1}) = (1.112916 \text{ mA}, 6.137955 \text{ V})$$

$$V_C = V_{C1} = V_{C2} = V_{CC} - I_C R_C = 12 - (1.12916 \times 10^{-3} \times 6k) = \mathbf{5.3225 \text{ V}}$$

$$\text{d) } r_\pi = \frac{\beta V_T}{I_{CQ}} = \frac{100 \times 26 \times 10^{-3}}{1.112916 \times 10^{-3}} = 2.336205 \text{ k}\Omega$$

$$A_d = \frac{\beta R_C}{2(r_\pi + R_S + \beta R_1)} = \mathbf{39.2865}$$

$$e) A_{cm} = \left| \frac{R_C}{2R_E} \right| = \left| \frac{6k}{2 \times 5k} \right| = \mathbf{0.6}$$

$$f) \text{CMRR} = \left| \frac{A_d}{A_{cm}} \right| = 65.4775$$

$$\text{CMRR in dB} = 20 \log_{10} \left( \frac{A_d}{A_{cm}} \right) = 20 \log_{10}(65.4775) = \mathbf{36.32184178 \text{ dB}}$$

### **SIMULATED RESULTS:**

Above circuit is simulated using LTspice and the results are presented below:

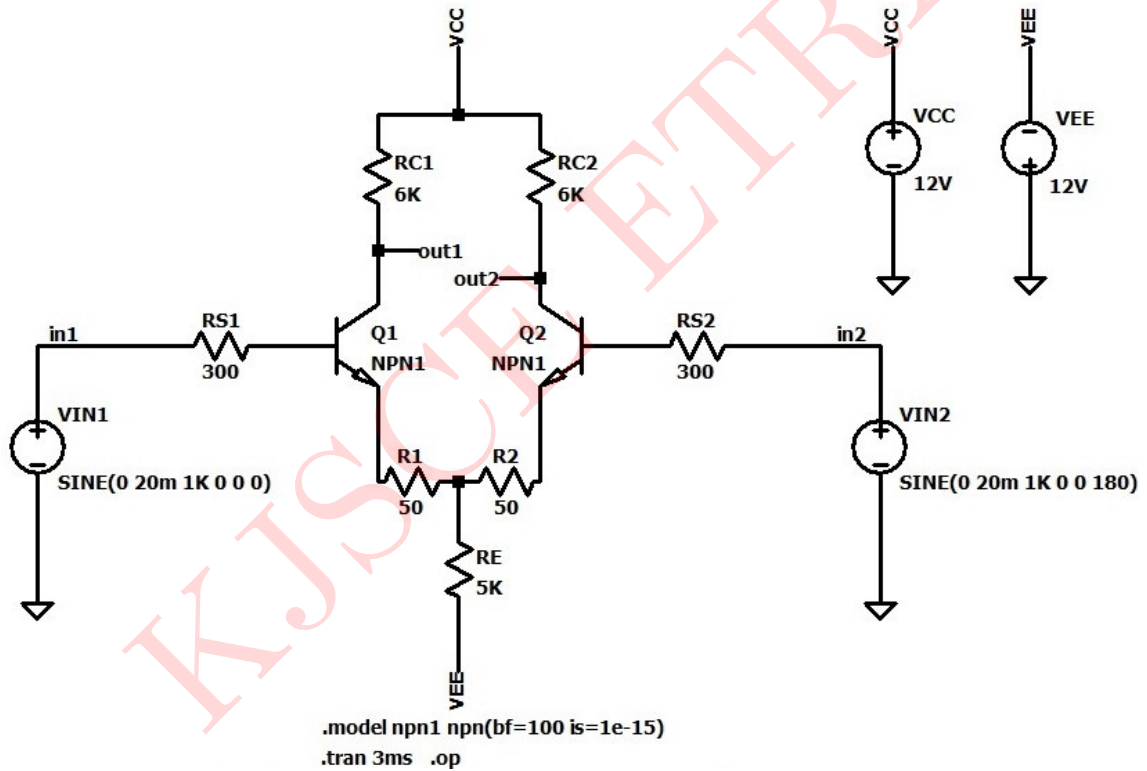


Figure 3: Circuit schematic

The waveforms for input and output voltage are shown in Figure 4

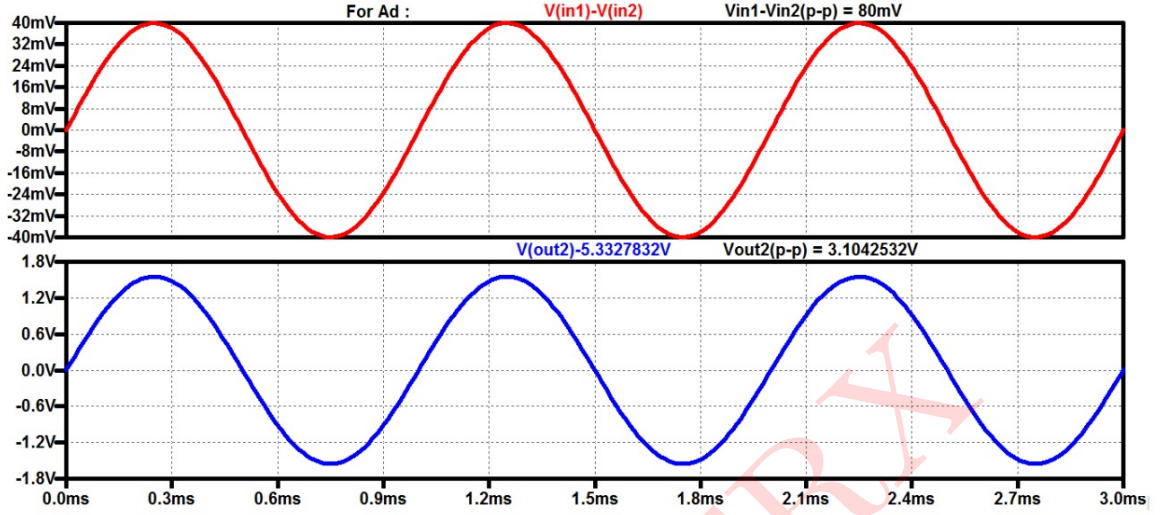


Figure 4: Input and output voltage waveforms

Comparison of theoretical and simulated values:

Parameters	Theoretical	Simulated
$I_{C1}, I_{C2}$	1.112916 mA	1.1112 mA
$V_{C1}, V_{C2}$	5.3225 V	5.33278 V
$V_{CE1}, V_{CE2}$	6.137955 V	6.053516 V
Differential voltage gain $A_d$	39.2865	38.803
Common mode voltage gain $A_{cm}$	0.6	—
CMRR in dB	36.3218 dB	—

Table 1: Numerical 1

2. Consider the differential amplifier in Figure 5.

The transistor parameters are:  $k_{n1} = k_{n2} = 40 \mu\text{A}/\text{V}^2$ ,  $V_{TN1} = V_{TN2} = 0.8 \text{ V}$ ,  $\lambda = 0.02\text{V}^{-1}$

- Determine  $I_S$ ,  $I_{D1}$ ,  $I_{D2}$ ,  $V_{D1}$ ,  $V_{D2}$ ,  $V_{DS1}$ ,  $V_{DS2}$
- Differential voltage gain  $A_d$
- Common mode voltage gain  $A_{cm}$
- CMRR in dB

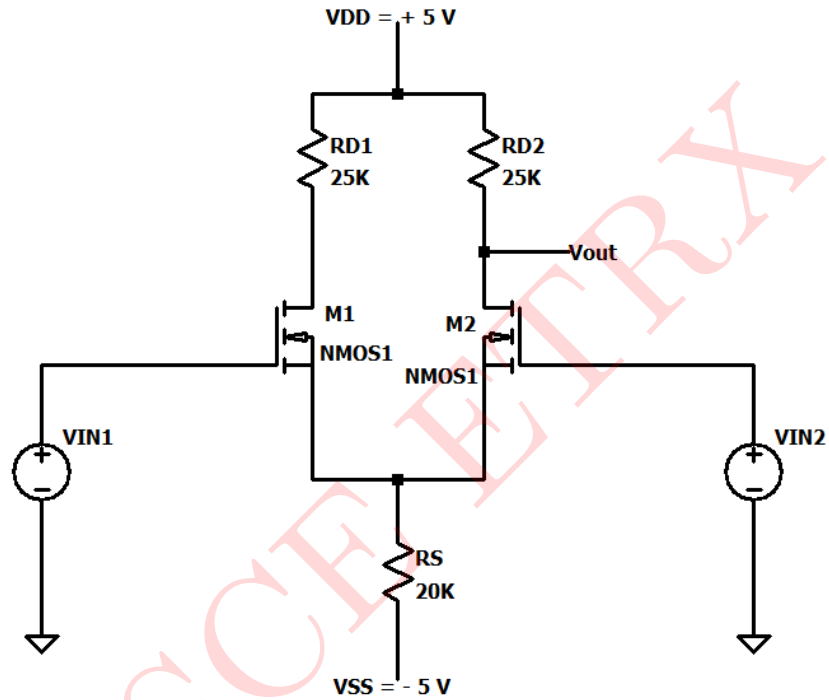


Figure 5: Circuit 2

**Solution:**

The above circuit is a DIUO mosfet differential amplifier.

For DC analysis, we consider only one transistor, as both transistors are identical.

b) DC analysis:

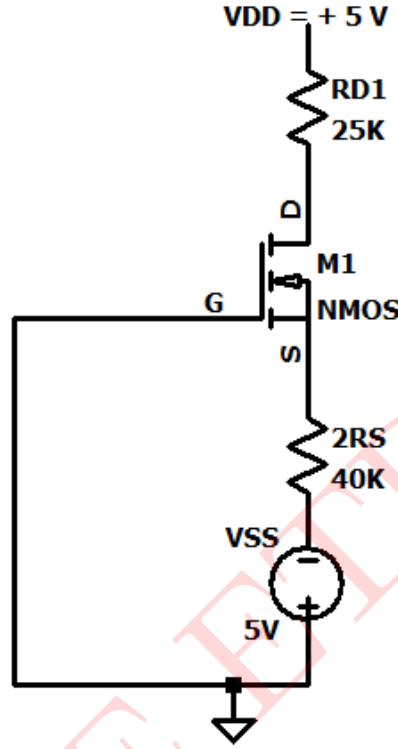


Figure 6: DC equivalent circuit

Applying KVL to the gate-source loop, we get

$$-V_{GS1} - 2I_{D1}R_S - V_{SS} = 0$$

$$\therefore V_{GS1} = -V_{SS} - 2I_{D1}R_S$$

$$V_{GS1} = 5 - 2I_{D1}(20k) \quad \dots\dots\dots(1)$$

Applying KVL at drain-source loop

$$V_{DD} - I_{D1}R_{D1} - V_{DS1} - 2I_{D1}R_S - V_{SS} = 0$$

$$V_{DS1} = 5 - I_{D1}(25k) - I_{D1}(40k) + 5$$

$$V_{DS1} = 10 - I_{D1}(65k) \quad \dots\dots\dots(2)$$

$$I_{D1} = k_n(V_{GS1} - V_{TN1})^2(1 + \lambda V_{DS1})$$

$$I_{D1} = 40 \times 10^{-6}(5 - I_{D1}(40k) - 0.8)^2(1 + 0.02(10 - I_{D1}(65k)))$$

$$I_{D1} = 40 \times 10^{-6}[21.168 - 403.2kI_{D1} + (1920 \times 10^{-6})I_{D1}^2 - I_{D1}(22.932k) + (436.8 \times 10^{-6})I_{D1}^2 - (2080 \times 10^9)I_{D1}^3]$$

$$(83.2 \times 10^{-6})I_{D1}^3 - 94.272kI_{D1}^2 + 18.04528I_{D1} - (8.467 \times 10^{-4}) = 0$$

$$\therefore I_{D_1} = 0.906106 \text{ mA or } 0.154076 \text{ mA or } 0.0728939 \text{ mA}$$

$$\text{For } I_{D_1} = 0.906106 \text{ mA, } V_{GS_1} = -31.2464 \text{ V}$$

$$\text{For } I_{D_1} = 0.154076 \text{ mA, } V_{GS_1} = -1.16304 \text{ V}$$

$$\text{For } I_{D_1} = 0.0728939 \text{ mA, } V_{GS_1} = 2.084244 \text{ V}$$

$V_{GS_1}$  cannot be negative, and should be greater than  $V_{TN_1}$

$$\therefore V_{GS_1} = 2.084244 \text{ V}$$

$$\therefore I_{D_1} = 0.0728939 \text{ mA} = I_{D_2}$$

$$V_{DS_1} = V_{DS_2} = 10 - 0.0728939 \times 10^{-3} \times 65 \times 10^3 = 5.261955 \text{ V}$$

$$V_{D_1} = V_{DD} - I_{D_1} R_{D_1} = V_{D_2} = 3.177675 \text{ V}$$

$$I_S = I_{D_1} = 0.0728939 \text{ mA}$$

**AC analysis:**

$$g_{m_1} = 2k_n(V_{GS_1} - V_{TN_1})(1 + \lambda V_{DS_1})$$

$$g_m = g_{m_1} = g_{m_2} = 2 \times 40 \times 10^{-6}(2.0842 - 0.8)(1 + 0.1052)$$

$$\therefore g_m = g_{m_1} = g_{m_2} = 0.113547 \text{ mA/V}$$

$$r_{d_2} = r_{d_1} = \frac{1}{\lambda I_{D_1}} = \frac{1}{0.02 \times 72.893 \times 10^{-6}} = 685.936 \text{ k}\Omega$$

$$|A_d| = \frac{g_m(r_d || R_D)}{2} = \frac{0.113547 \times 10^{-3} 685.936 k || 25k}{2} = 1.369422$$

$$|A_{cm}| = \frac{g_m(r_d || R_D)}{1 + 2g_m R_S} = 0.4942$$

$$\text{CMRR} = \left| \frac{A_d}{A_{cm}} \right| = \left| \frac{1.369}{0.4942} \right| = 2.770133$$

$$\text{CMRR in dB} = 20 \log_{10}(2.770133) = 8.85 \text{ dB}$$

### SIMULATED RESULTS:

Above circuit is simulated using LTSpice and the results are presented below:

$$\begin{aligned} \text{Ad} &= \text{Vout(p-p)} / \text{Vin(p-p)} \\ \text{Ad} &= 109.72048\text{mV} / 80\text{mV} \\ \text{Ad} &= 1.3715 \end{aligned}$$

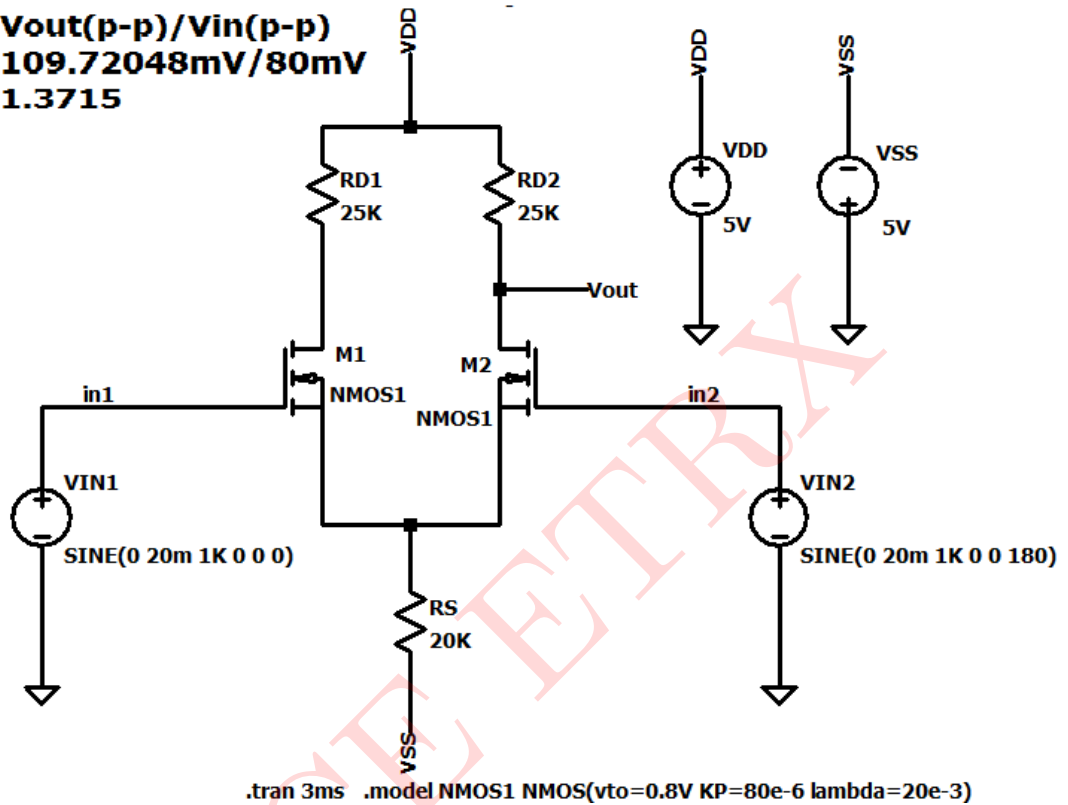


Figure 7: Circuit schematic



The waveforms for input and output voltage are shown in Figure 8

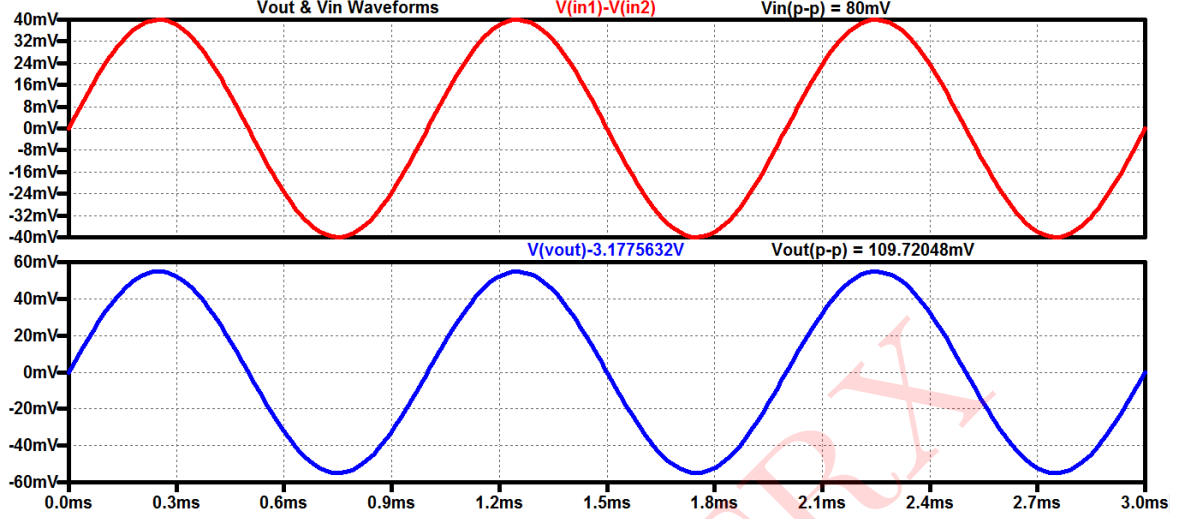


Figure 8: Input and output voltage waveforms

Comparison of theoretical and simulated values:

Parameters	Theoretical	Simulated
$I_S$	72.893 $\mu\text{A}$	72.8975 $\mu\text{A}$
$I_{D_1}$	72.893 $\mu\text{A}$	72.8975 $\mu\text{A}$
$I_{D_2}$	72.893 $\mu\text{A}$	72.8975 $\mu\text{A}$
$V_{D_1}$	3.177675 V	3.17756 V
$V_{D_2}$	3.177675 V	3.17756 V
$V_{DS_1}$	5.261955 V	5.26166 V
$V_{DS_2}$	5.261955 V	5.26166 V
Differential voltage gain $ A_d $	1.369422	1.37
Common mode voltage gain $A_{cm}$	0.4942	—
CMRR in dB	8.85 dB	—

Table 2: Numerical 2

3. For the circuit shown in Figure 9, consider  
 $R_{C1} = R_{C2} = 5 \text{ k}\Omega$ , current source = 2.5 mA and  $R_{in} = 50 \text{ }\Omega$   
 Calculate  
 a)  $I_{C1}, I_{C2}$   
 b)  $V_{C1}, V_{C2}$   
 c) Differential voltage gain  $A_d$

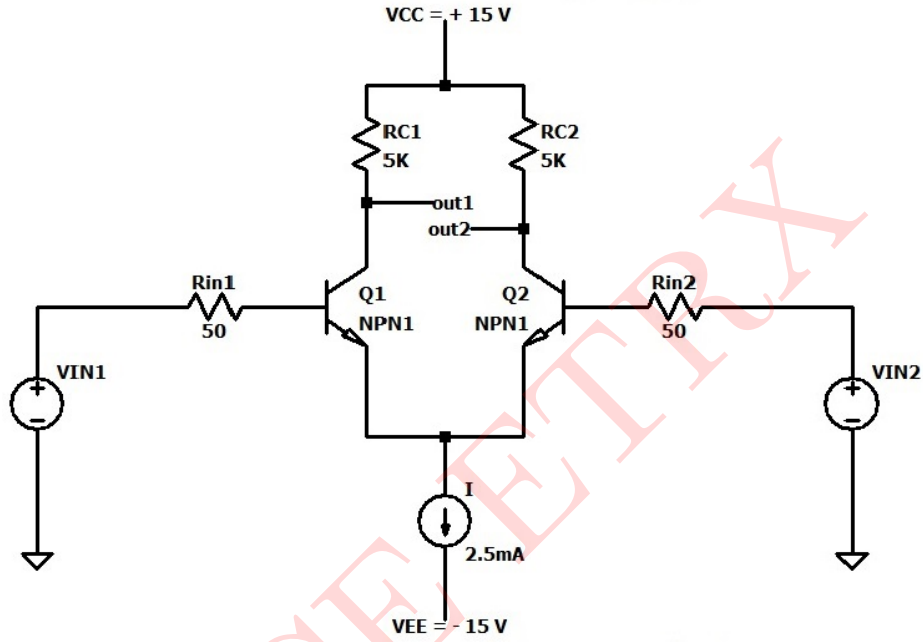


Figure 9: Circuit 3

**Solution:**

a) **DC Analysis:**

$$I_{E1} = I_{E2} = \frac{I}{2} = \frac{2.5 \times 10^{-3}}{2} = 1.25 \text{ mA}$$

$$I_E = I_C + I_B = I_C + \frac{I_C}{\beta} = I_C \left( 1 + \frac{1}{\beta} \right)$$

$$\therefore I_{C1} = \frac{I_{E1}}{\left( 1 + \frac{1}{\beta_1} \right)} = \frac{1.25 \times 10^{-3}}{1 + 0.01}$$

$$\therefore I_{C1} = I_{C2} = \mathbf{1.2376 \text{ mA}}$$

$$\text{b) DC value of } V_{o1} = V_{CC} - I_{C1} R_C = V_{C1} = 15 - (1.25 \times 10^{-3} \times 5k)$$

$$\therefore V_{C1} = \mathbf{8.75 \text{ V}}$$

$$V_{o2} = V_{CC} - I_{C2} R_{C2} = V_{C2} = 15 - (1.25 \times 10^{-3} \times 5k)$$

$$\therefore V_{C2} = \mathbf{8.75 \text{ V}}$$

c) Output is being taken from collector of transistor  $Q_1$

Assuming  $V_{o2} > V_{o1}$ ,

$$A_d = \frac{V_{o1}}{V_1 - V_2} = \frac{-\beta R_C}{2(r_\pi + R_{in})}$$

$$r_\pi = \frac{\beta V_T}{I_{CQ}} = \frac{100 \times 26 \times 10^{-3}}{1.2376 \times 10^{-3}} = 2.1008 \text{ k}\Omega$$

$$\therefore A_d = \frac{-100 \times 5k}{2(2.1008k + 50)} = -116.2358$$

### SIMULATED RESULTS:

Above circuit is simulated using LTspice and the results are presented below:

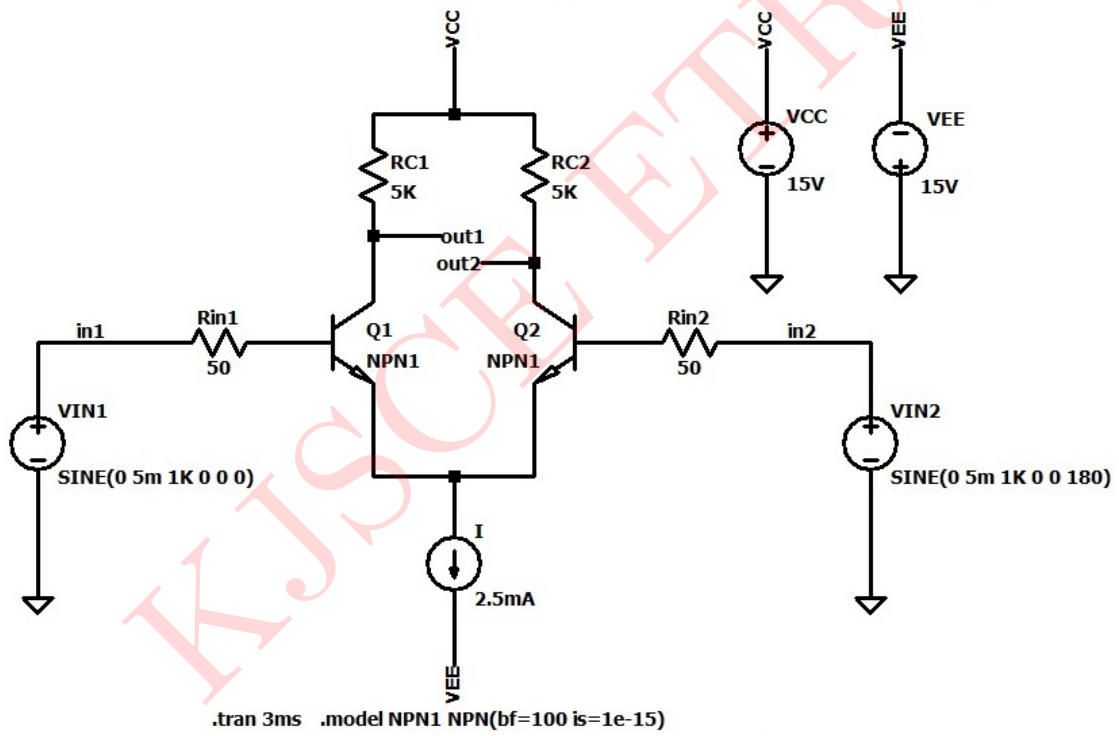


Figure 10: Circuit schematic

The waveforms for input and output voltage are shown in Figure 11

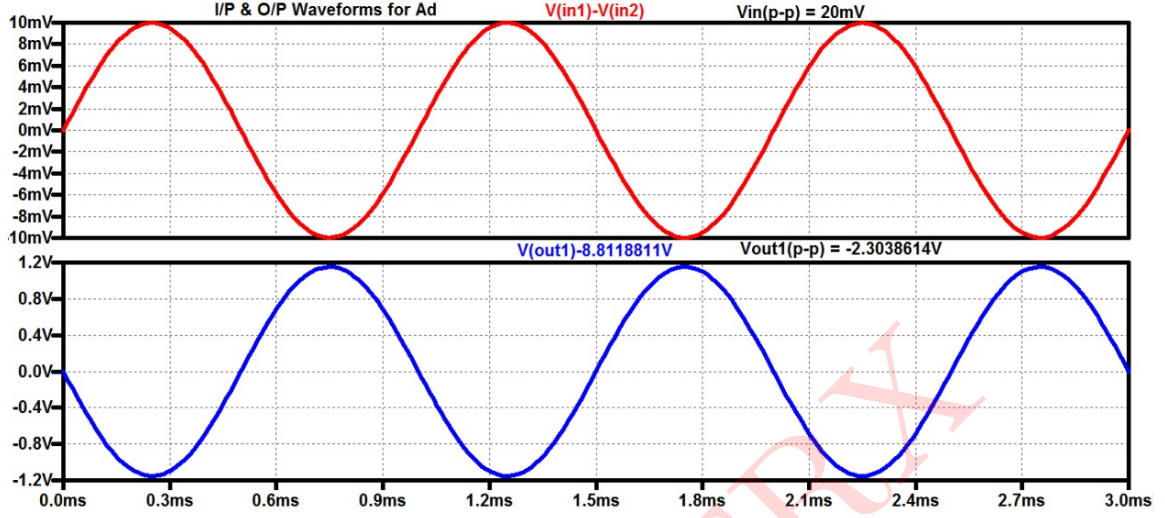


Figure 11: Input and output voltage waveforms

Comparison of theoretical and simulated values:

Parameters	Theoretical	Simulated
$I_{C_1}, I_{C_2}$	1.2376 mA	1.23762 mA
$V_{C_1}, V_{C_2}$	8.75 V	8.81188 V
Differential voltage gain $A_d$	-116.2358	-115.19307

Table 3: Numerical 3