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Oscillator circuits

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Design 1:

Design a RC phase shift oscillator to oscillate at 600Hz. Supply voltage is given as 10V.

Solution:

Step 1: Circuit diagram and selection of transistor.

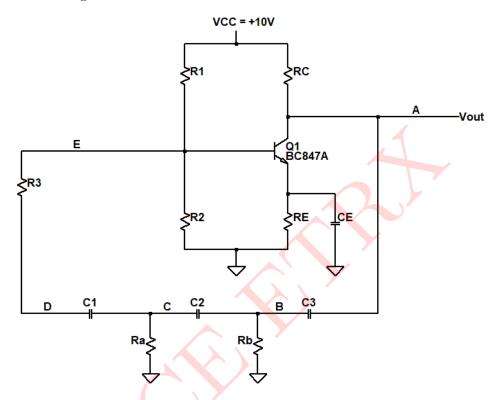


Figure 1: Circuit diagram

Here, $R_a = R_b$ and $R_3 > R_a$, R_b

$$C_1 = C_2 = C_3 = C$$
 and $\beta = 180$

Select transistor BC147A

$$h_{fe} = 220, h_{ie} = 2.7k\Omega, h_{FE} = 180, V_{CE(sat)} = 0.25$$

Step 2: Selection of R and C:

For RC phase shift oscillators,
$$f = \frac{1}{2\pi RC\sqrt{6+4k}}$$

where
$$k = \frac{R_C}{R}$$

Minimum value of k is 2.7

$$\therefore 600 = \frac{1}{2\pi RC\sqrt{6 + 4(2.7)}}$$
$$\therefore RC = \frac{1}{2\pi \times 600 \times \sqrt{6 + 4(2.7)}}$$

Assuming $C = 0.01 \mu F$,

$$\therefore R = \frac{1}{2\pi(600)(0.01\mu F)(\sqrt{6+4(2.7)})}$$

$$\therefore R = 6.4716k\Omega$$

Choosing higher standard value, $R = 6.8k\Omega$, 1/4W

Step 3: Selection of R_C :

We know,
$$k = \frac{R_C}{R}$$

$$\therefore R_C = kR$$

$$\therefore R_C = 2.7(6.8k\Omega)$$

$$\therefore R_C = 18.36k\Omega$$

Choosing higher standard value, $R_C = 22k\Omega$, 1/4W

Step 4: Selection of R_E :

 $V_E = 10\%$ of V_{CC} for good stability

$$V_E = 0.1(10)$$

$$\therefore V_E = 1$$

For maximum symmetrical output voltage swing, select Q-point at the centre of DC load line

$$V_{CEQ} = 0.5V_{CC} = 0.5(10)$$

$$\therefore V_{CEQ} = 5V$$

Applying KVL to C-E loop:

$$I_{CQ} = \frac{V_{CC} - V_{CEQ} - V_E}{R_C}$$

$$\therefore I_{CQ} = \frac{10 - 5 - 1}{22k\Omega}$$

$$\therefore I_{CQ} = 0.1818mA$$

$$\therefore I_{BQ} = \frac{I_{CQ}}{\beta} \qquad ...(\because I_{CQ} = \beta I_{BQ})$$

$$\therefore I_{BQ} = \frac{0.1818mA}{180} = 1.0101\mu A$$

Also,
$$I_{EQ} = (\beta + 1)I_{BQ}$$

$$I_{EQ} = 181(1.0101\mu A) = 0.1828mA$$

$$V_E = I_{EQ}R_E$$

$$\therefore R_E = \frac{V_E}{I_{EQ}} = \frac{1V}{0.1828mA}$$

$$\therefore R_E = 5.4696k\Omega$$

Choosing higher standard value,

$$R_E = 4.7k\Omega, 1/4W$$

Step 5: Selection of $R_1 \& R_2$:

Let
$$S = 8$$
,

$$\therefore S = \frac{\beta + 1}{1 + \beta \times \left[\frac{R_E}{R_B + R_E}\right]}$$

here,
$$R_B = R_1 \parallel R_2$$

$$\therefore 8 = \frac{180 + 1}{1 + \left\lceil \frac{180 \times 4.7k\Omega}{4.7k\Omega + R_B} \right\rceil}$$

On solving, $R_B = 34.4214k\Omega$

Also,
$$V_B = \frac{R_2 \times V_{CC}}{R_1 + R_2}$$

Applying KVL to B-E loop:

$$V_B = \left(\frac{I_{CQ}}{\beta}\right)R_B + V_{BE(ON)} + I_{EQ}R_E$$

$$\therefore V_B = \left(\frac{0.1818mA}{180}\right) (34.4214k\Omega) + 0.7 + (0.1828mA)(4.7k\Omega)$$

$$V_B = 1.5939V$$

From (1), (2) and (3):

$$R_1 = \frac{34.4214 \times 10}{1.5939} = 215.9571 \text{k}\Omega$$

Choosing higher standard value, $\therefore R_1 = 220k\Omega, 1/4W$

...(1)

From (2), we get:

$$\frac{10 \times R_2}{220k\Omega + R_2} = 1.5939$$

 $\therefore R_2 = 40.8059k\Omega$

Choosing lower standard value, $R_2 = 39k\Omega$, 1/4W

Step 6: Selection of C_E :

To ensure complete bypass, $X_{CE} \leq \frac{R_E}{10} \leq 0.1 R_E$

$$X_{CE} = \frac{1}{2\pi f_L C_E} = 0.1 R_E$$

$$\therefore C_E = \frac{1}{2\pi f_L(0.1)R_E} = \frac{1}{2\pi (600)(0.1)4.7k\Omega}$$

$$C_E = 5.6438 \times 10^{-7} F$$

Choosing higher standard value, $C_E = 1\mu F/25V$

Step 7: Selection of R_3 in R_C selection:

To avoid loading effect by input impedance of BJT towards R_3 ,

$$\therefore R_3 = R + R_i$$

here, $R_i = R_1 \parallel R_2 \parallel h_{ie}$

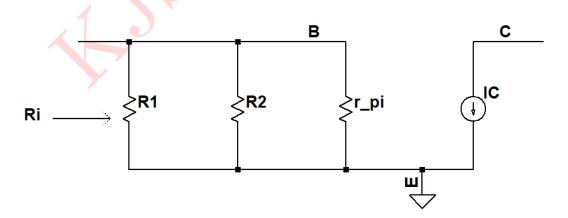


Figure 2: For calculation of R_i

$$\therefore R_i = 220k\Omega \parallel 39k\Omega \parallel 2.7k\Omega \qquad ...h_{ie} = r_{\pi}$$

$$\therefore R_i = 2.4965k\Omega$$

$$R_3 = 6.8k\Omega + 2.4965k\Omega$$

$$\therefore R_3 = 9.2965k\Omega$$

Choosing higher standard value, $R_3 = 10k\Omega$, 1/4W

Step 8: Designed completed circuit:

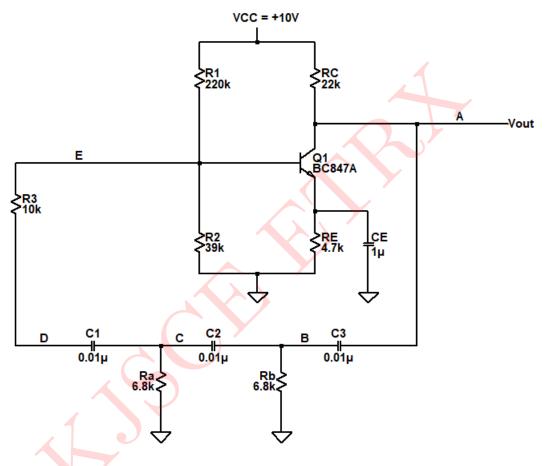


Figure 3: Designed circuit

Frequency of oscillations are given as,

$$f_L = \frac{1}{2\pi RC\sqrt{6+4k}}$$

$$f_L = \frac{1}{2\pi (6.8k\Omega)(0.01\mu F)(\sqrt{6+4(2.7)})}$$

$$\therefore f_L = 571.0269Hz$$

Time period of oscillations, $T = 1/f_L$

$$\therefore T = 1.7512 \text{ ms}$$

SIMULATED RESULTS:

Above circuit was simulated in LTspice and results obtained are as follows:

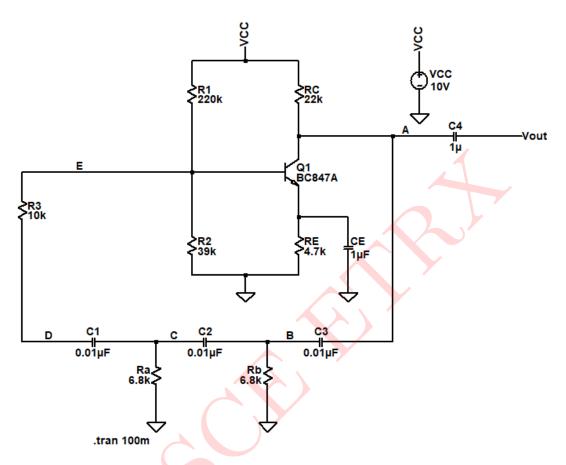


Figure 4: Circuit Schematic: Results

Waveforms observed are shown below:

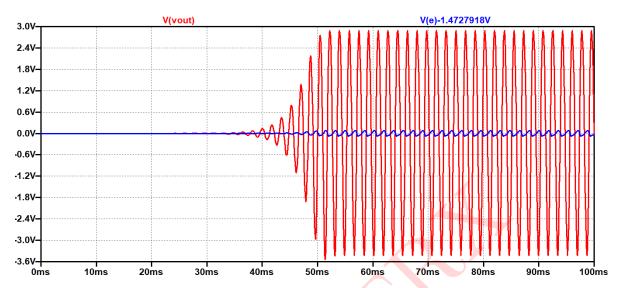


Figure 5: Output oscillations and feedback signal



Figure 6: Output oscillations

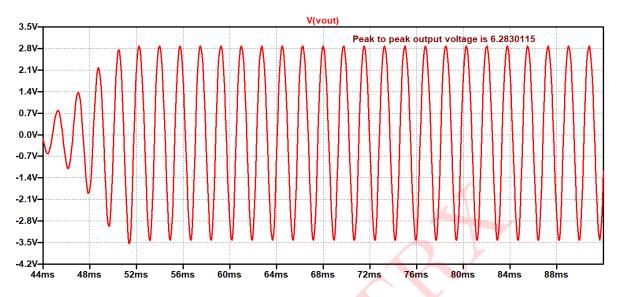


Figure 7: Output oscillations: Zoomed in

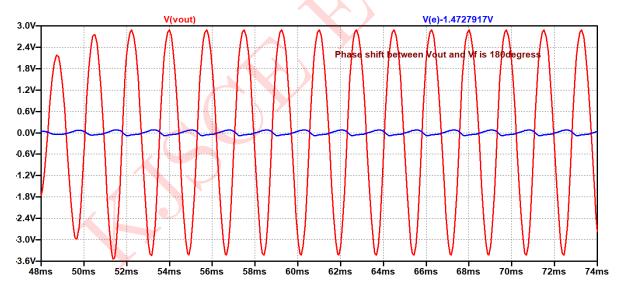


Figure 8: Phase shift between output and feedback signal

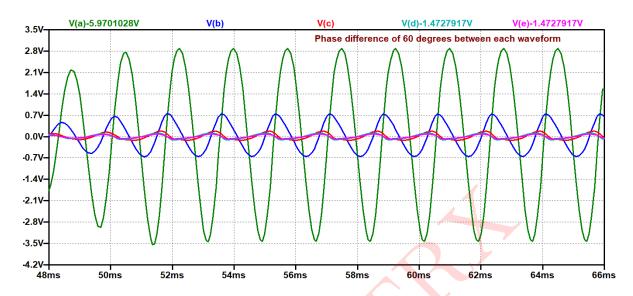


Figure 9: Phase shift between signals from each stage

Comparison between theoretical and simulated values:

Parameter	Theoretical value	Simulated value
Frequency of oscillations (f_L) :	571.0269Hz	570.9702Hz
Time period of oscillations:	$1.7512~\mathrm{ms}$	$1.7514~\mathrm{ms}$
Amplitude of oscillations (V_{p-p}) :	<u> </u>	6.2830V
Feedback signal V_f amplitude and new phase w.r.t V_{out}	_	$159.2456 \text{ mV}_{p-p}$ 180°
Feedback fraction: V_f/V_{out}	0.0253	0.0344
Phase shift offered by feedback network	180 °	180°

Table 1: Numerical 1