

微电子器件物理 双极型晶体管2

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本节课提纲

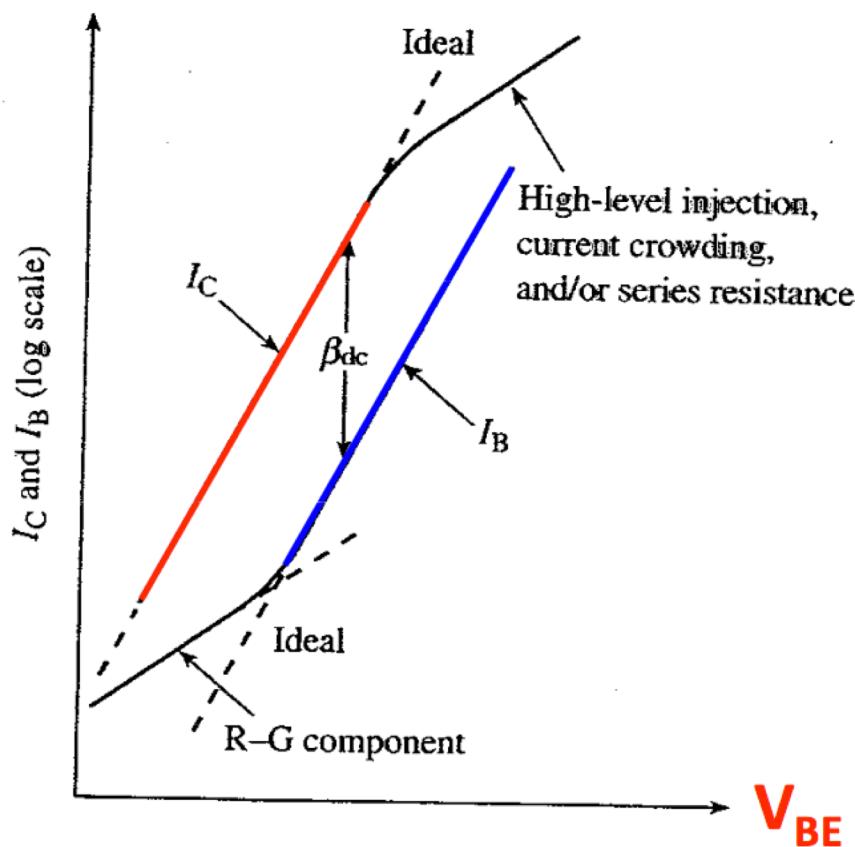
1. 三极管的电流增益
2. 基极的掺杂设计
3. 集电极的掺杂设计

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输出特性曲线

$$\frac{I_C}{A} \simeq -\frac{qD_n}{W_B} \frac{n_{i,B}^2}{N_B} (e^{qV_{BE}/kT} - 1) + \frac{qD_n}{W_B} \frac{n_{i,B}^2}{N_B} (e^{qV_{BC}/kT} - 1)$$



$$\frac{I_B}{A} = \frac{qD_p}{W_E} \frac{n_{i,E}^2}{N_E} (e^{qV_{BE}/kT} - 1)$$

$$\beta_{DC} = \frac{I_C}{I_B}$$

$\beta_{DC} \rightarrow$ 共射级增益

电流增益

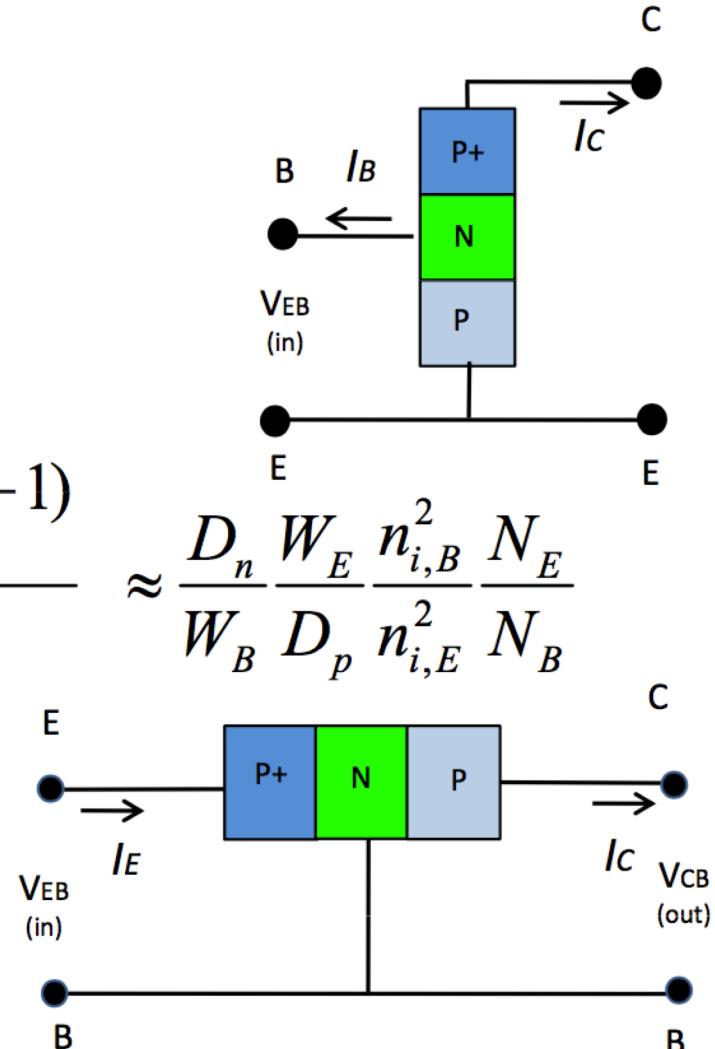
共射极电流增益

$$\beta_{DC} = \frac{I_C}{I_B}$$

$$= \frac{\frac{qD_n}{W_B} \frac{n_{i,B}^2}{N_B} (e^{qV_{BE}/kT} - 1) + \frac{qD_n}{W_B} \frac{n_{i,B}^2}{N_B} (e^{(qV_{BC}/kT)} - 1)}{\frac{qD_n}{W_E} \frac{n_{i,E}^2}{N_E} (e^{qV_{BE}/kT} - 1)} \approx \frac{D_n}{W_B} \frac{W_E}{D_p} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B}$$

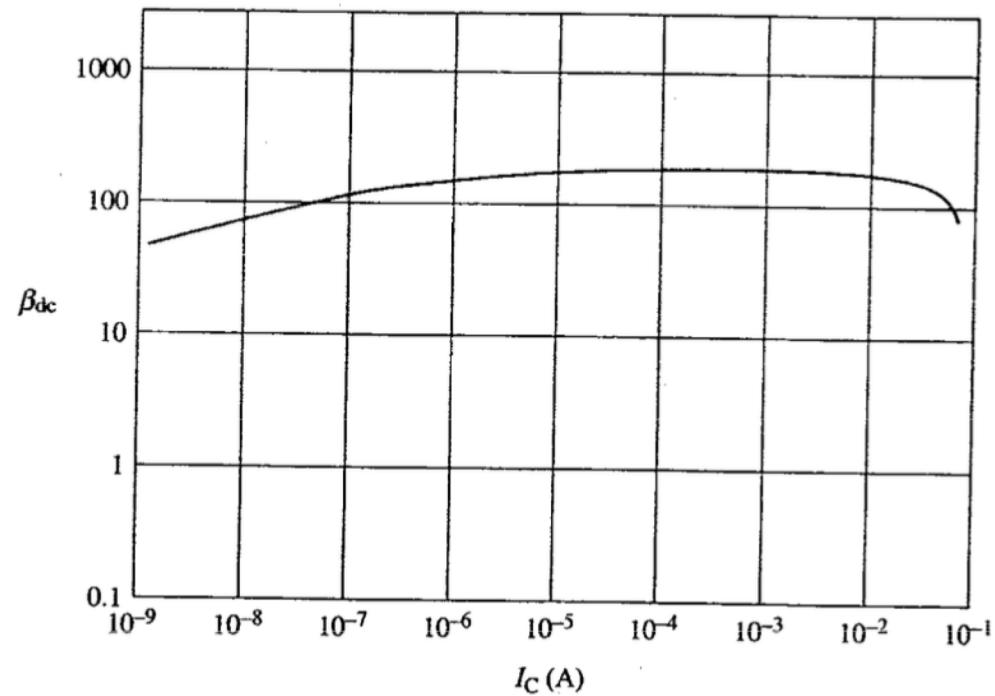
共基极电流增益

$$\alpha_{DC} = \frac{I_C}{I_E} \rightarrow \beta_{DC} = \frac{I_C}{I_B} = \frac{I_C}{I_E - I_C} = \frac{\alpha_{DC}}{1 - \alpha_{DC}}$$



电流增益

$$\beta_{DC} \approx \frac{D_n}{W_B} \frac{W_E}{D_p} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B}$$



如何设计一个好的三极管

For a given Emitter length

$$\beta_{DC} \approx \frac{D_n}{W_B} \frac{W_E}{D_p} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B}$$

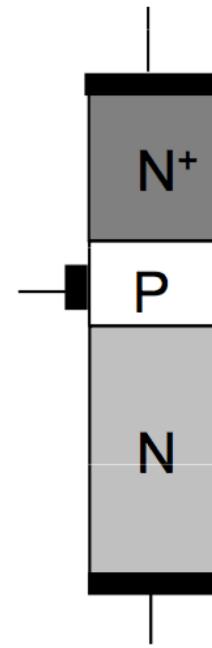
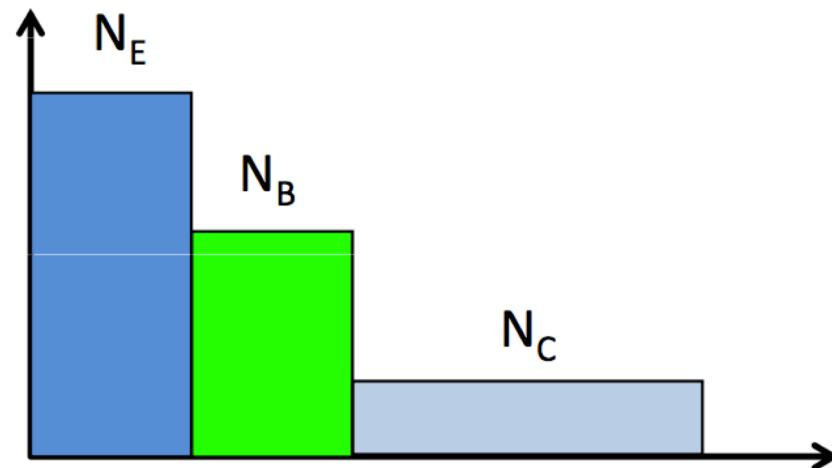
~1, same material

Make-Base short ...
(few mm in 1950s, 200 Å now)

Emitter doping higher
than Base doping

修改掺杂浓度

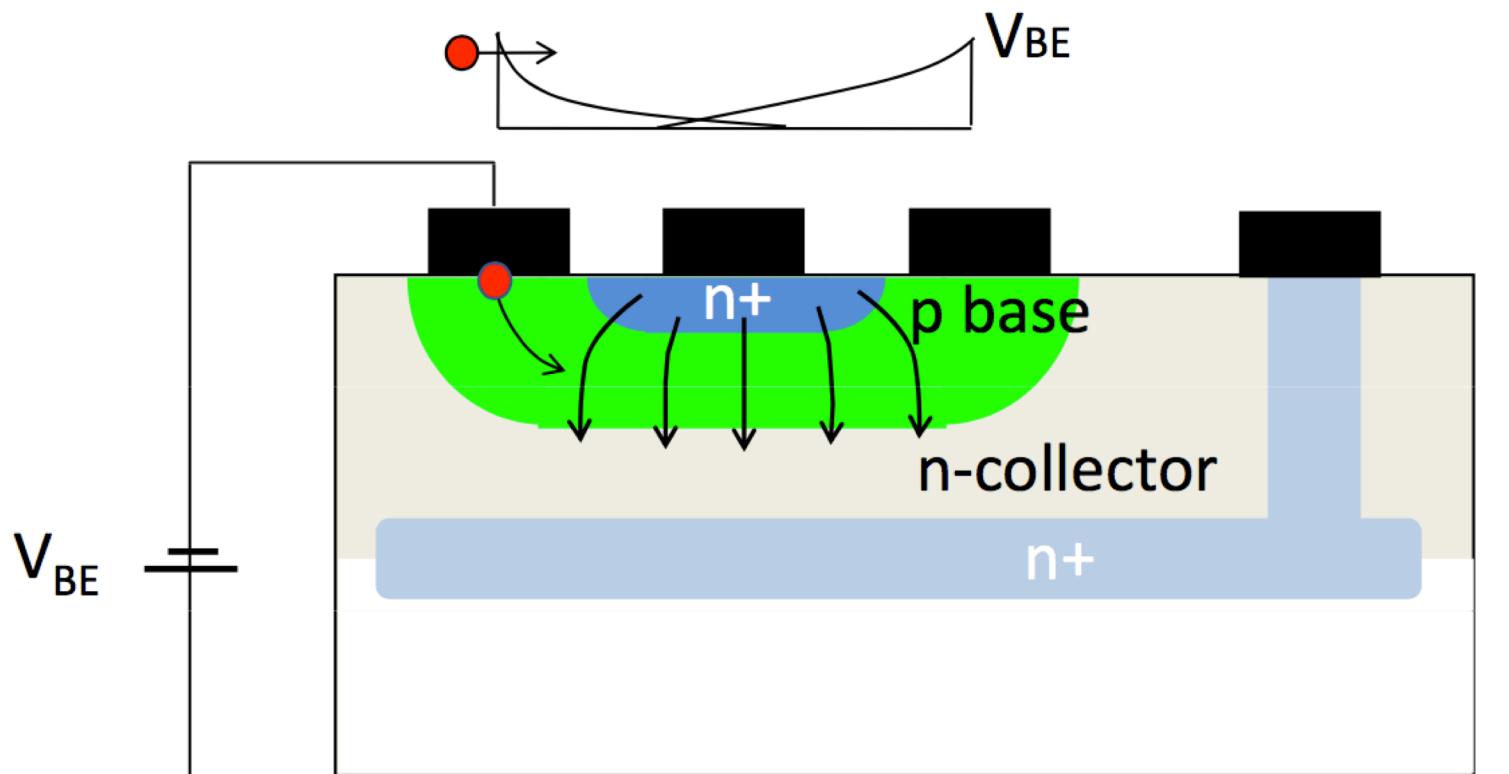
$$\beta_{DC} \approx \frac{D_n}{W_B} \frac{W_E}{D_p} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B}$$



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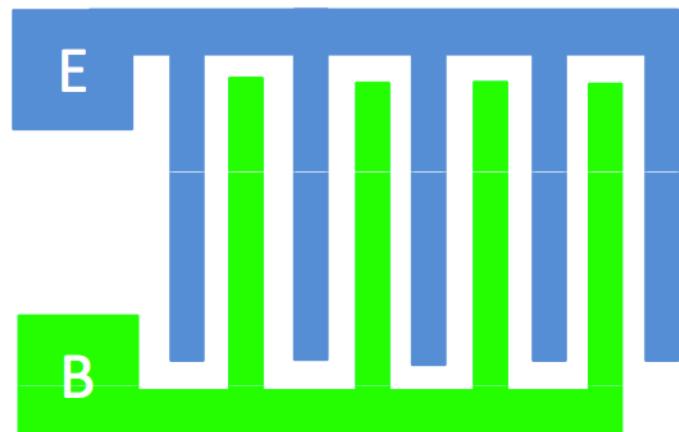
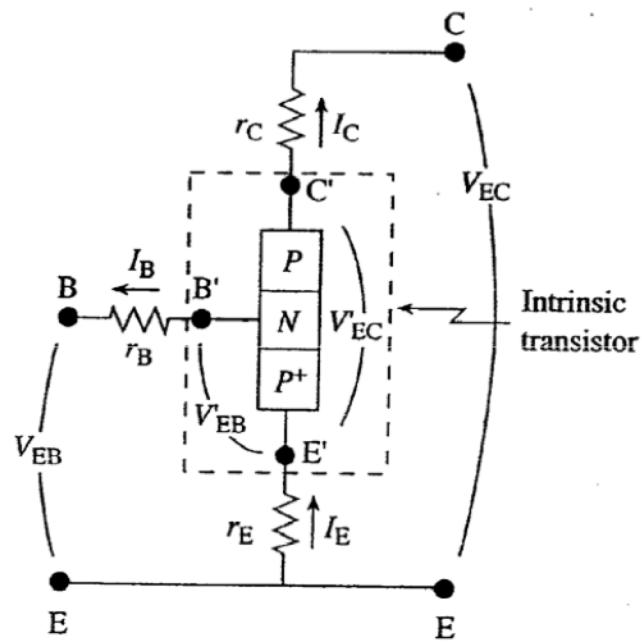
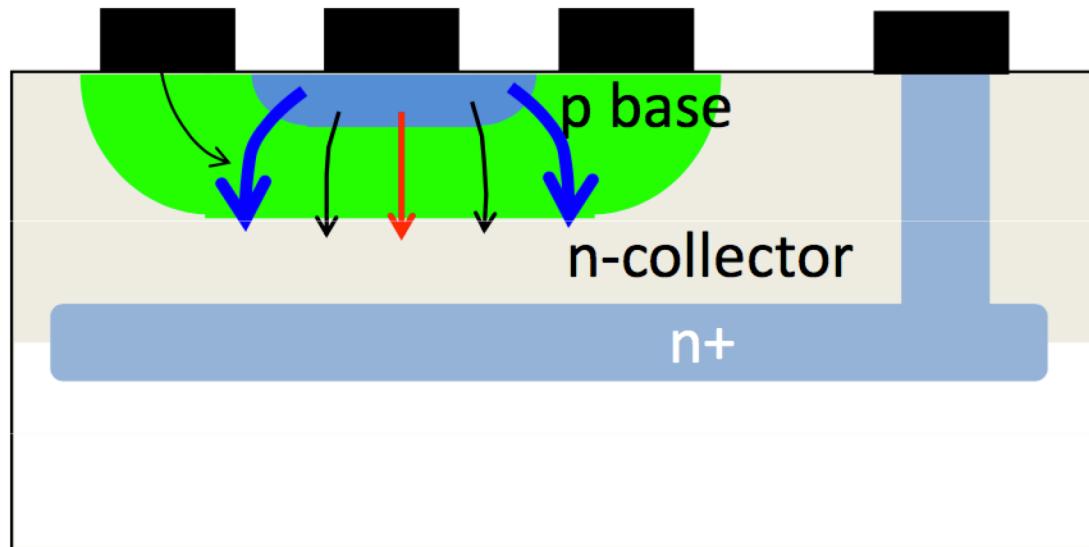
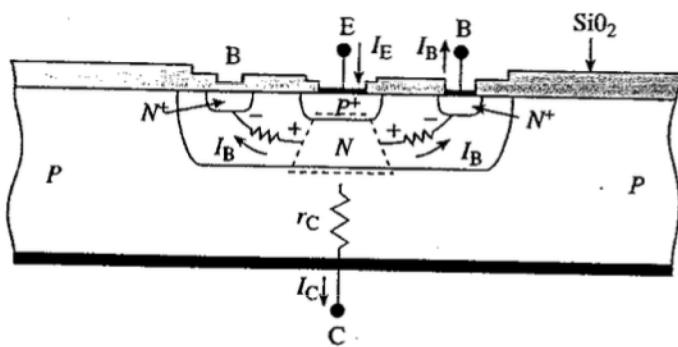
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基极低掺杂浓度引入的问题1

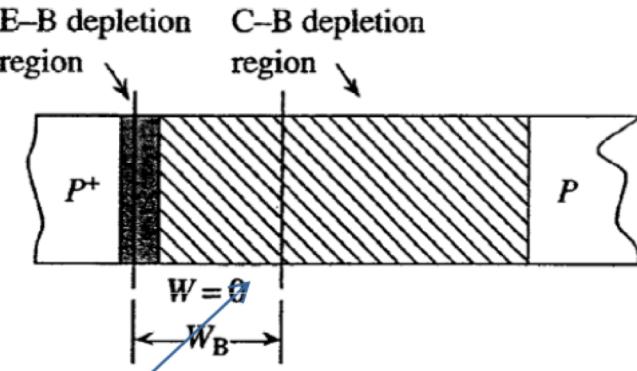
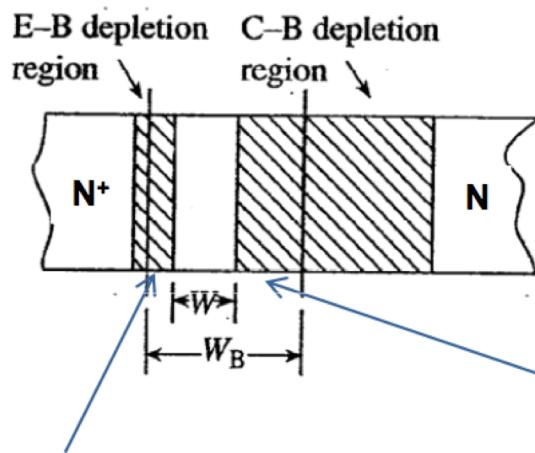
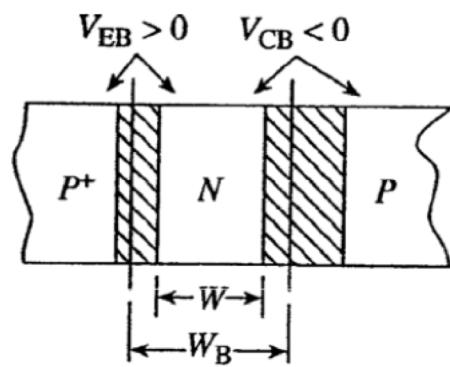
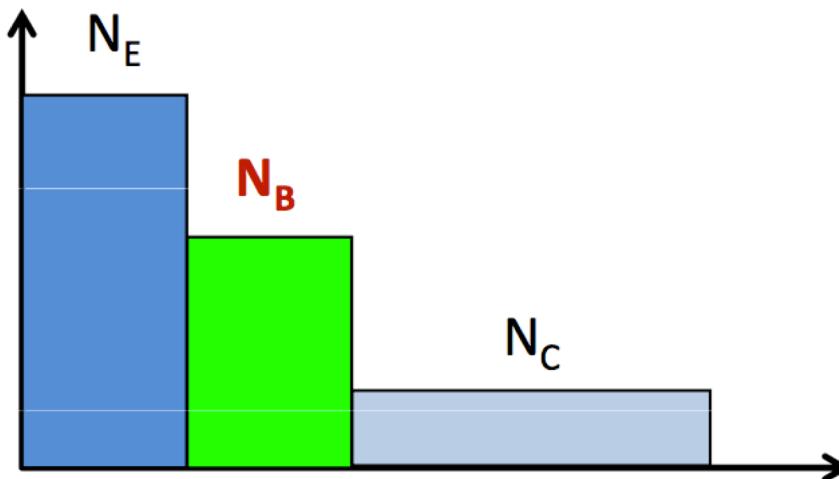


$$\beta = \frac{I_C}{I_B} = \frac{\int J_C(x)dx}{\int J_B(x)dx} = \frac{\int \frac{qD_n}{W_B} \frac{n_{i,B}^2}{N_B} \left(e^{qV'_B(x)\beta} - 1 \right) dx}{\int \frac{qD_p}{W_E} \frac{n_{i,E}^2}{N_E} \left(e^{qV'_B(x)\beta} - 1 \right) dx}$$

基极低掺杂浓度引入的问题2



基极低掺杂浓度引入的问题3

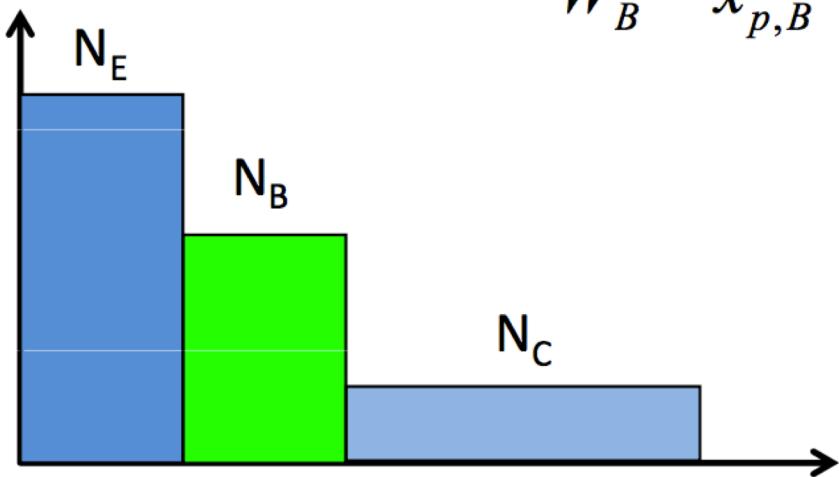


$$x_{p,BE} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_E}{N_B(N_E + N_B)} (V_{bi} - V_{BE})}$$

$$x_{p,BC} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_C}{N_B(N_C + N_B)} (V_{bi} - V_{BC})}$$

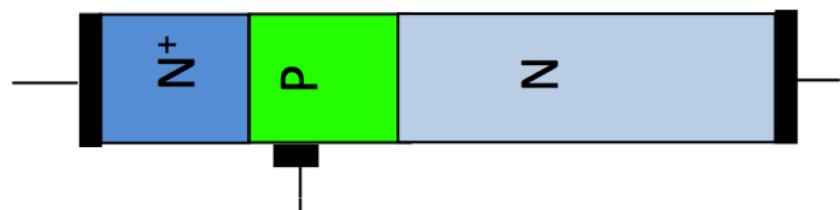
基极低掺杂浓度引入的问题4

$$\beta_{DC} \approx \frac{D_n}{W_B - x_{p,B} - x_{p,c}} \frac{W_E \cancel{n_{i,B}^2} N_E}{\cancel{D_p} \cancel{n_{i,E}^2} N_B}$$



$$x_{p,BE} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_E}{N_B(N_E + N_B)} (V_{bi} - V_{BE})}$$

$$x_{p,BC} = \sqrt{\frac{2k_s \epsilon_0}{q} \frac{N_C}{N_B(N_C + N_B)} (V_{bi} - V_{BC})}$$



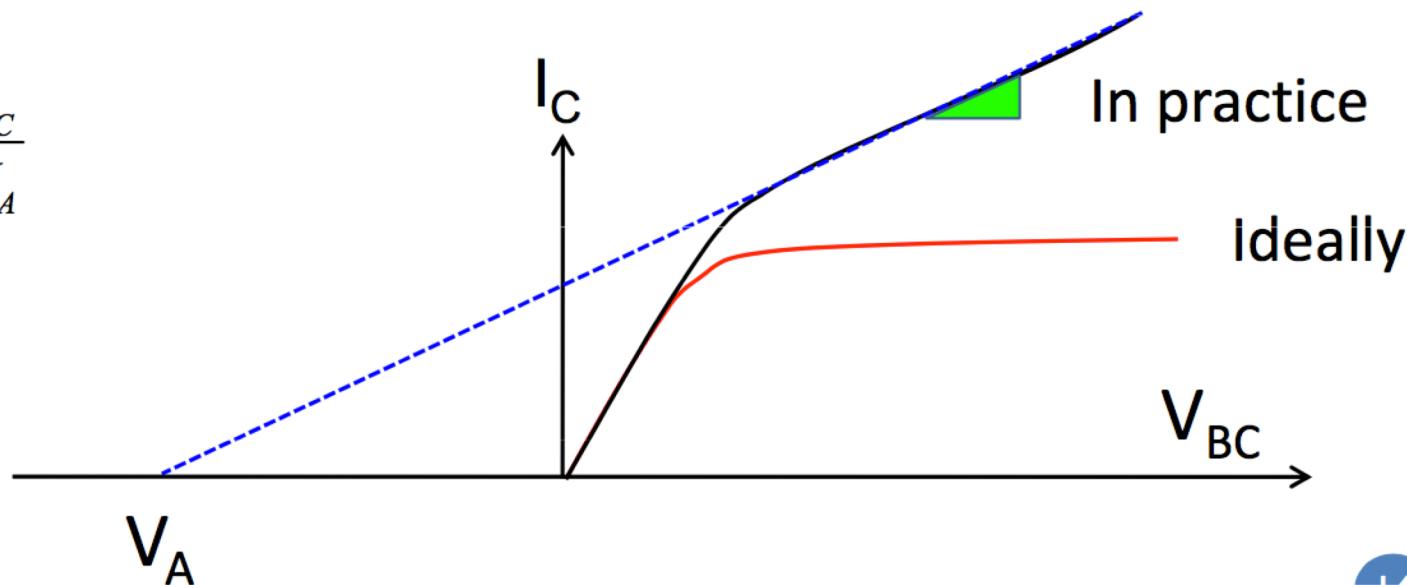
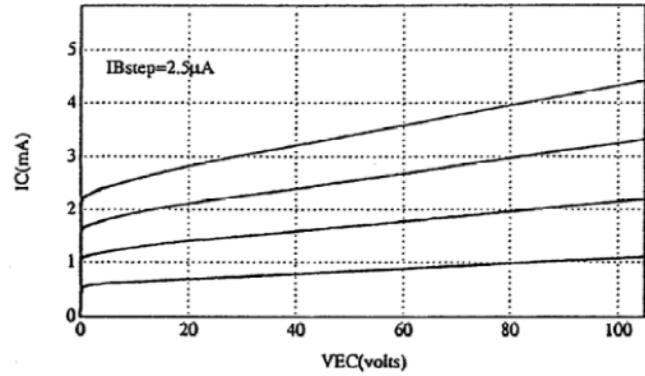
Gain depends on collector voltage (**bad**) ...

Early电压

$$\beta_{DC} \approx \frac{D_n}{W_B - x_{p.B} - x_{p.C}} \frac{W_E}{D_p} \cancel{\frac{n^2_{i,B}}{n^2_{i,E}}} \frac{N_E}{N_B}$$

$$I_{n,C} = -\frac{qD_n}{W_B} \frac{n^2_{i,B}}{N_B} (e^{(qV_{BE}/kT)} - 1) + \frac{qD_n}{W_B} \frac{n^2_{i,B}}{N_B} (e^{(qV_{BC}/kT)} - 1)$$

$$\frac{dI_C}{dV_{BC}} = \frac{I_C}{V_{BC} + V_A} \approx \frac{I_C}{V_A}$$

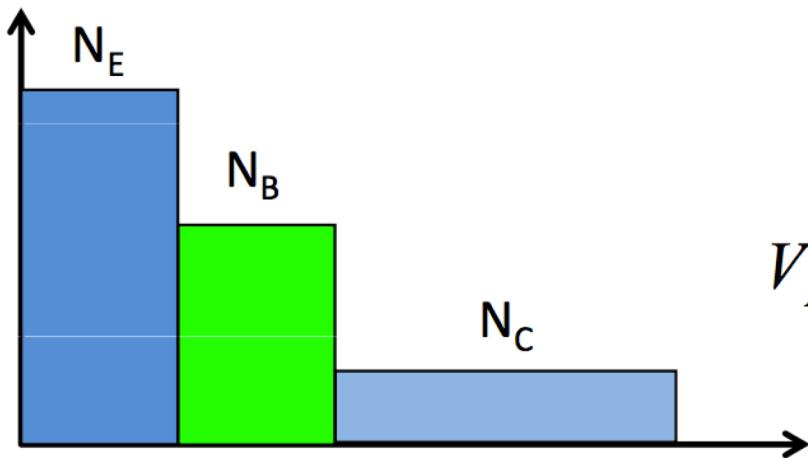


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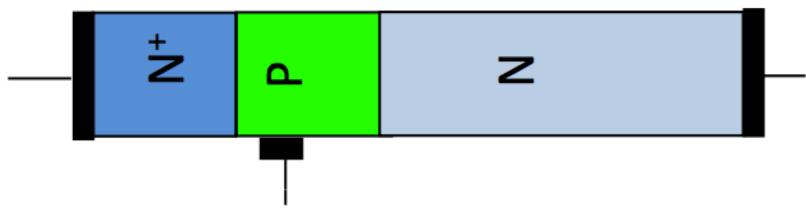
集电极掺杂

$$\beta \approx \frac{D_n}{W_B - x_{p,B} - x_{p,C}} \frac{W_E}{D_p} \frac{n_{i,B}^2 / N_E}{n_{i,E}^2 / N_B}$$



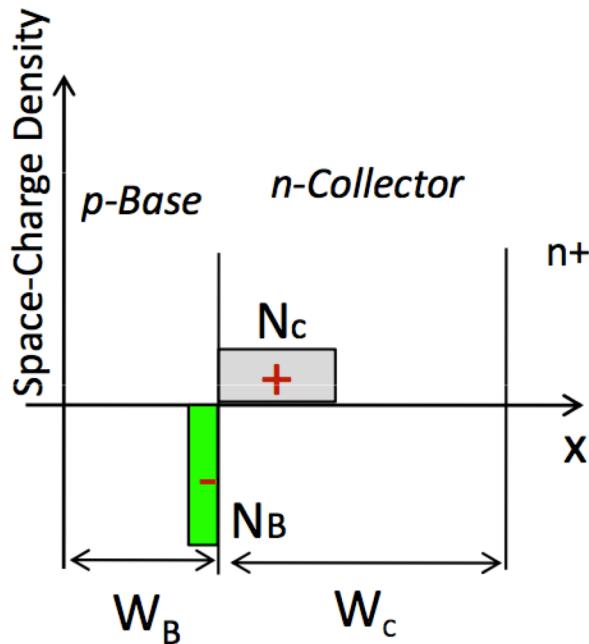
$$V_A = -\frac{qN_B W_B}{C_{CB}}$$

$$C_{CB} = \frac{\kappa_s \epsilon_0}{x_{n,C} + x_{p,B}}$$



If you want low base doping
then reduce collector doping
even more to increase
Collector depletion.....

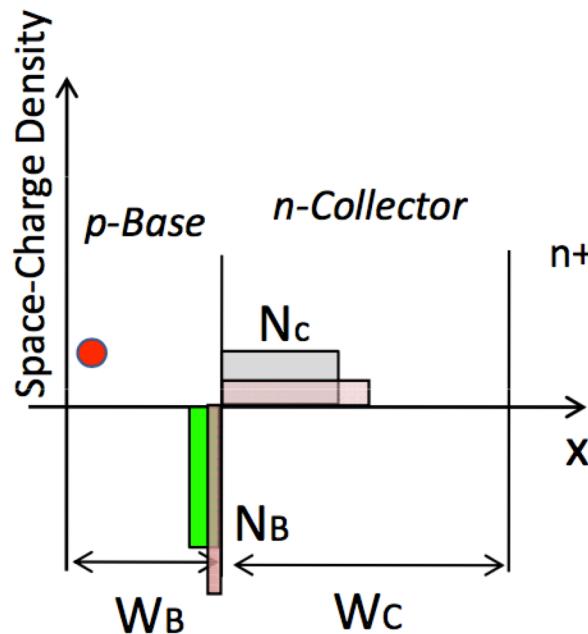
Kirk效应与基区扩展效应



$$N_B x_B = N_C x_C$$

$$V_{bi} - V_{BC} = \frac{q}{2\kappa_s \epsilon_0} [N_B x_B^2 + N_C x_C^2]$$

$$J_C = qv_{sat} n$$

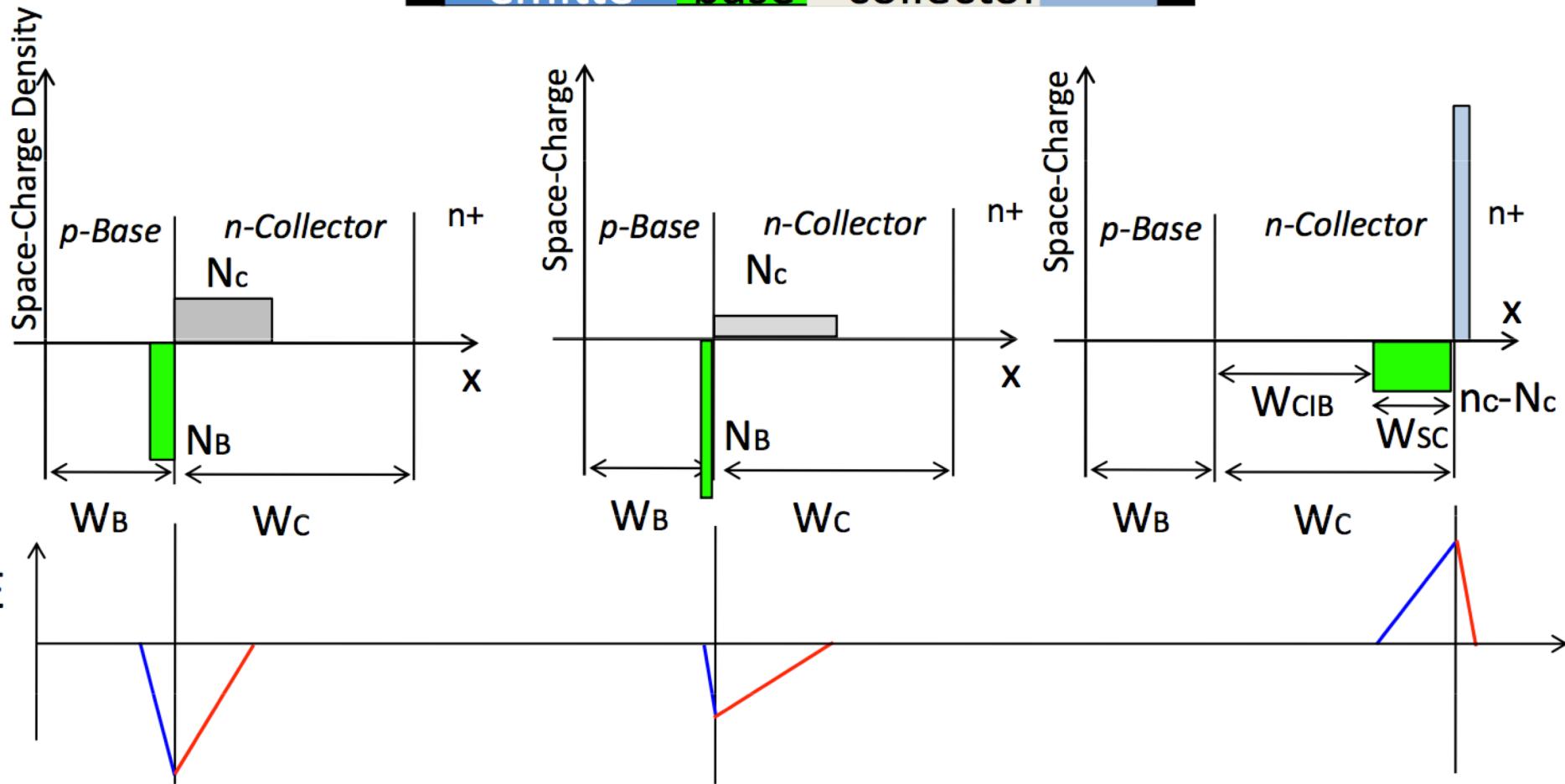
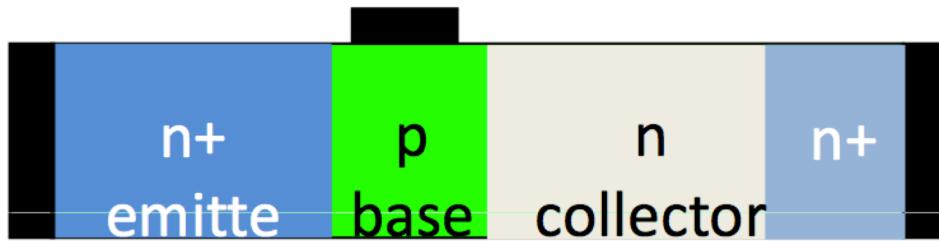


$$(N_B + n)x_B' = (N_C - n)x_C'$$

$$V_{bi} - V_{BC} = \frac{q}{2\kappa_s \epsilon_0} [(N_B + n)x_B'^2 + (N_C - n)x_C'^2]$$

$$x_C' = x_C \sqrt{\frac{1 + \frac{n}{N_B}}{1 - \frac{n}{N_C}}} = x_C \sqrt{\frac{1 + \frac{J_C}{qv_{sat}N_B}}{1 - \frac{J_C}{qv_{sat}N_C}}}$$

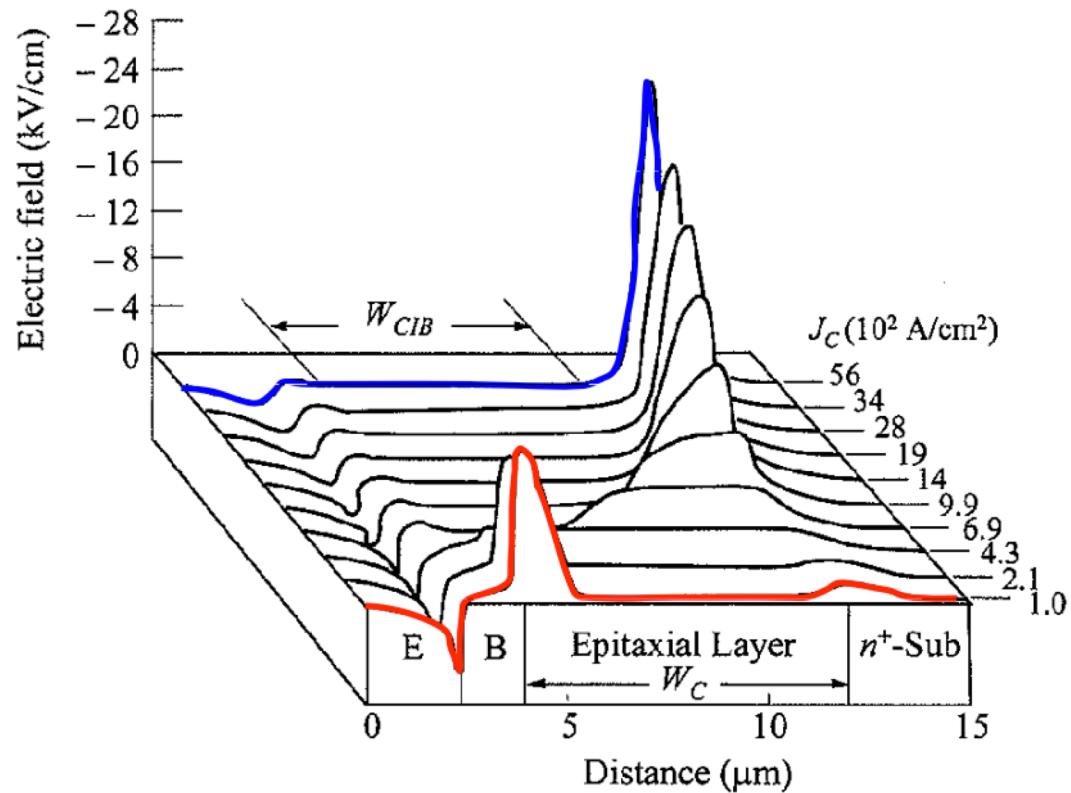
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Kirk效应与基区扩展效应

$$x_C' = x_C \sqrt{\frac{1 + \frac{J_C}{qv_{sat}N_B}}{1 - \frac{J_C}{qv_{sat}N_C}}}$$

$$J_{C,crit} = qv_{sat}N_C \equiv J_K$$



Can not reduce collector doping
arbitrarily without causing base pushout

增加发射区掺杂浓度？

禁带变窄导致电流增益急剧降低

$$\beta \approx \frac{D_n}{W_B} \frac{W_E}{D_p} \frac{n_{i,B}^2}{n_{i,E}^2} \frac{N_E}{N_B} = \frac{D_n}{W_B} \frac{W_E}{D_p} \frac{N_C N_V e^{-E_{g,B}/kT}}{N_C N_V e^{-E_{g,E}/kT}} \frac{N_E}{N_B} \approx e^{-\Delta E_g/kT} \frac{N_E}{N_B}$$

量子隧穿效应导致基极无法完全控制电流



Thanks!
Q&A