

LM4562 Dual High-Performance, High-Fidelity Audio Operational Amplifier

Check for Samples: [LM4562](#)

FEATURES

- Easily Drives 600Ω Loads
- Optimized for Superior Audio Signal Fidelity
- Output Short Circuit Protection
- PSRR and CMRR Exceed 120dB (Typ)
- SOIC, PDIP, and TO-99 Packages

APPLICATIONS

- Ultra High-Quality Audio Amplification
- High-Fidelity Preamplifiers
- High-Fidelity Multimedia
- State-of-the-Art Phono Pre Amps
- High-Performance Professional Audio
- High-Fidelity Equalization and Crossover Networks
- High-Performance Line Drivers
- High-Performance Line Receivers
- High-Fidelity Active Filters

KEY SPECIFICATIONS

- Power Supply Voltage Range: $\pm 2.5\text{V}$ to $\pm 17\text{V}$
- THD+N ($A_V = 1$, $V_{OUT} = 3V_{RMS}$, $f_{IN} = 1\text{kHz}$)
 - $R_L = 2\text{k}\Omega$: 0.00003% (typ)
 - $R_L = 600\Omega$: 0.00003% (typ)
- Input Noise Density: $2.7\text{nV}/\sqrt{\text{Hz}}$ (typ)
- Slew Rate: $\pm 20\text{V}/\mu\text{s}$ (typ)
- Gain Bandwidth Product: 55MHz (typ)
- Open Loop Gain ($R_L = 600\Omega$): 140dB (typ)
- Input Bias Current: 10nA (typ)
- Input Offset Voltage: 0.1mV (typ)
- DC Gain Linearity Error: 0.000009%

DESCRIPTION

The LM4562 is part of the ultra-low distortion, low-noise, high-slew-rate operational amplifier series optimized and fully specified for high-performance, high-fidelity applications. Combining advanced leading-edge process technology with state-of-the-art circuit design, the LM4562 audio operational amplifiers deliver superior audio signal amplification for outstanding audio performance. The LM4562 combines extremely low voltage noise density ($2.7\text{nV}/\sqrt{\text{Hz}}$) with vanishingly low THD+N (0.00003%) to easily satisfy the most demanding audio applications. To ensure that the most challenging loads are driven without compromise, the LM4562 has a high slew rate of $\pm 20\text{V}/\mu\text{s}$ and an output current capability of $\pm 26\text{mA}$. Further, dynamic range is maximized by an output stage that drives 2kΩ loads to within 1V of either power supply voltage and to within 1.4V when driving 600Ω loads.

The LM4562's outstanding CMRR (120dB), PSRR (120dB), and V_{OS} (0.1mV) give the amplifier excellent operational amplifier DC performance.

The LM4562 has a wide supply range of $\pm 2.5\text{V}$ to $\pm 17\text{V}$. Over this supply range the LM4562's input circuitry maintains excellent common-mode and power supply rejection, as well as maintaining its low input bias current. The LM4562 is unity gain stable. This Audio Operational Amplifier achieves outstanding AC performance while driving complex loads with values as high as 100pF.

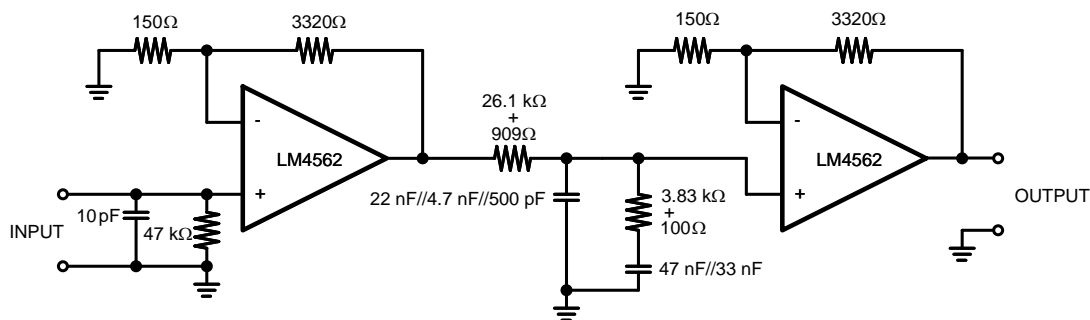
The LM4562 is available in an 8-lead narrow body SOIC, an 8-lead PDIP, and an 8-lead TO-99. Demonstration boards are available for each package.



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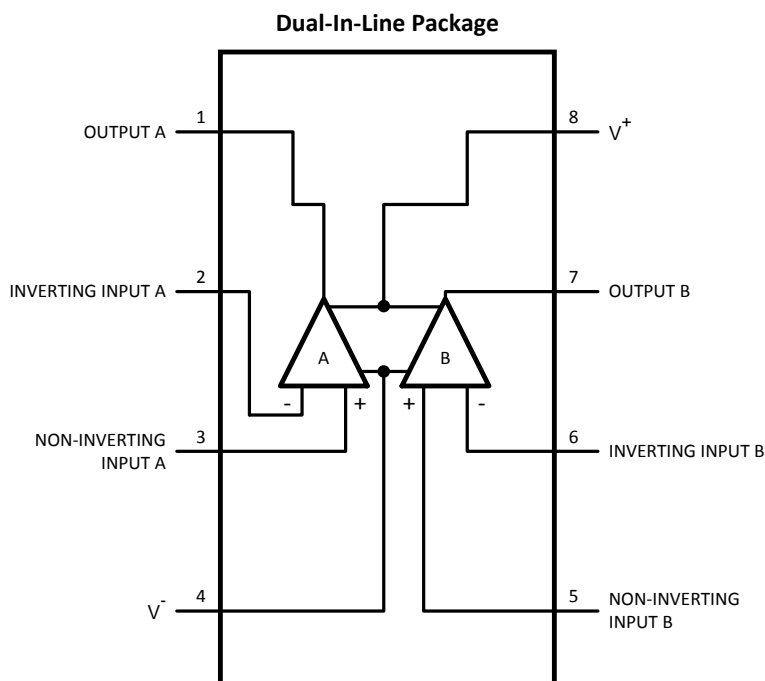
TYPICAL APPLICATION



A. 1% metal film resistors, 5% polypropylene capacitors

Figure 1. Passively Equalized RIAA Phono Preamplifier

CONNECTION DIAGRAMS



**Figure 2. 8-Lead SOIC (D Package)
8-Lead PDIP (P Package)**

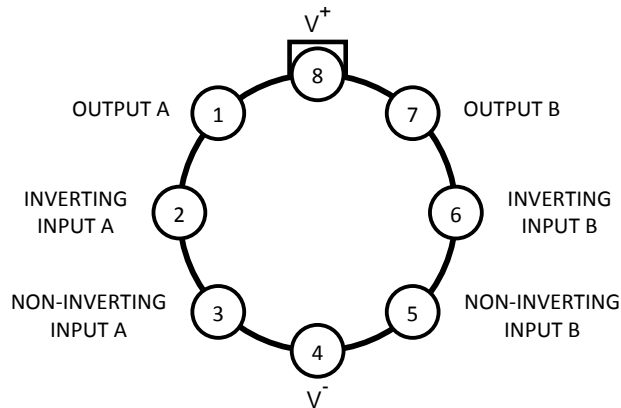


Figure 3. 8-Lead TO-99 (LMC Package)



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾⁽²⁾⁽³⁾

| | | |
|---|---------------------|----------------------------------|
| Power Supply Voltage ($V_S = V^+ - V^-$) | | 36V |
| Storage Temperature | | -65°C to 150°C |
| Input Voltage | | (V-) - 0.7V to (V+) + 0.7V |
| Output Short Circuit ⁽⁴⁾ | | Continuous |
| Power Dissipation | | Internally Limited |
| ESD Susceptibility ⁽⁵⁾ | | 2000V |
| ESD Susceptibility ⁽⁶⁾ | Pins 1, 4, 7 and 8 | 200V |
| | Pins 2, 3, 5 and 6 | 100V |
| Junction Temperature | | 150°C |
| Thermal Resistance | θ_{JA} (D) | 145°C/W |
| | θ_{JA} (P) | 102°C/W |
| | θ_{JA} (LMC) | 150°C/W |
| | θ_{JC} (LMC) | 35°C/W |
| Temperature Range ($T_{MIN} \leq T_A \leq T_{MAX}$) | | -40°C $\leq T_A \leq$ 85°C |
| Supply Voltage Range | | $\pm 2.5V \leq V_S \leq \pm 17V$ |

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.
- (2) Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see the Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (4) Amplifier output connected to GND, any number of amplifiers within a package.
- (5) Human body model, 100pF discharged through a 1.5kΩ resistor.
- (6) Machine Model ESD test is covered by specification EIAJ IC-121-1981. A 200pF cap is charged to the specified voltage and then discharged directly into the IC with no external series resistor (resistance of discharge path must be under 50Ω).

ELECTRICAL CHARACTERISTICS FOR THE LM4562⁽¹⁾⁽²⁾

The specifications apply for $V_S = \pm 15V$, $R_L = 2k\Omega$, $f_{IN} = 1kHz$, $T_A = 25^\circ C$, unless otherwise specified.

| Symbol | Parameter | Conditions | LM4562 | | Units (Limits) |
|-----------------------------|--|--|------------------------|--------------------------|-------------------------|
| | | | Typical ⁽³⁾ | Limit ⁽⁴⁾ | |
| THD+N | Total Harmonic Distortion + Noise | $A_V = 1$, $V_{OUT} = 3V_{rms}$ $R_L = 2k\Omega$ $R_L = 600\Omega$ | 0.00003 0.00003 | 0.00009 | % (max) |
| IMD | Intermodulation Distortion | $A_V = 1$, $V_{OUT} = 3V_{RMS}$ Two-tone, 60Hz & 7kHz 4:1 | 0.00005 | | % |
| GBWP | Gain Bandwidth Product | | 55 | 45 | MHz (min) |
| SR | Slew Rate | | ± 20 | ± 15 | V/ μs (min) |
| FPBW | Full Power Bandwidth | $V_{OUT} = 1V_{P-P}$, $-3dB$ referenced to output magnitude at $f = 1kHz$ | 10 | | MHz |
| t_s | Settling time | $A_V = -1$, 10V step, $C_L = 100pF$ 0.1% error range | 1.2 | | μs |
| e_n | Equivalent Input Noise Voltage | $f_{BW} = 20Hz$ to $20kHz$ | 0.34 | 0.65 | μV_{RMS} (max) |
| | Equivalent Input Noise Density | $f = 1kHz$ $f = 10Hz$ | 2.7 6.4 | 4.7 | nV/\sqrt{Hz} (max) |
| i_n | Current Noise Density | $f = 1kHz$ $f = 10Hz$ | 1.6 3.1 | | pA/\sqrt{Hz} |
| V_{OS} | Offset Voltage | | ± 0.1 | ± 0.7 | mV (max) |
| $\Delta V_{OS}/\Delta Temp$ | Average Input Offset Voltage Drift vs Temperature | $-40^\circ C \leq T_A \leq 85^\circ C$ | 0.2 | | $\mu V/^\circ C$ |
| PSRR | Average Input Offset Voltage Shift vs Power Supply Voltage | $\Delta V_S = 20V^{(5)}$ | 120 | 110 | dB (min) |
| ISO _{CH-CH} | Channel-to-Channel Isolation | $f_{IN} = 1kHz$ $f_{IN} = 20kHz$ | 118 112 | | dB |
| I_B | Input Bias Current | $V_{CM} = 0V$ | 10 | 72 | nA (max) |
| $\Delta I_{OS}/\Delta Temp$ | Input Bias Current Drift vs Temperature | $-40^\circ C \leq T_A \leq 85^\circ C$ | 0.1 | | nA/ $^\circ C$ |
| I_{OS} | Input Offset Current | $V_{CM} = 0V$ | 11 | 65 | nA (max) |
| V_{IN-CM} | Common-Mode Input Voltage Range | | +14.1 -13.9 | (V+) - 2.0 (V-) + 2.0 | V (min) |
| CMRR | Common-Mode Rejection | $-10V < V_{cm} < 10V$ | 120 | 110 | dB (min) |
| Z_{IN} | Differential Input Impedance | | 30 | | k Ω |
| | Common Mode Input Impedance | $-10V < V_{cm} < 10V$ | 1000 | | M Ω |
| A_{VOL} | Open Loop Voltage Gain | $-10V < V_{out} < 10V$, $R_L = 600\Omega$ | 140 | 125 | dB (min) |
| | | $-10V < V_{out} < 10V$, $R_L = 2k\Omega$ | 140 | | |
| | | $-10V < V_{out} < 10V$, $R_L = 10k\Omega$ | 140 | | |
| V_{OUTMAX} | Maximum Output Voltage Swing | $R_L = 600\Omega$ | ± 13.6 | ± 12.5 | V (min) |
| | | $R_L = 2k\Omega$ | ± 14.0 | | |
| | | $R_L = 10k\Omega$ | ± 14.1 | | |
| I_{OUT} | Output Current | $R_L = 600\Omega$, $V_S = \pm 17V$ | ± 26 | ± 23 | mA (min) |
| I_{OUT-CC} | Instantaneous Short Circuit Current | | +53 -42 | | mA |
| R_{OUT} | Output Impedance | $f_{IN} = 10kHz$ Closed-Loop Open-Loop | 0.01 13 | | Ω |

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.

(2) Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see the Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.

(3) Typical specifications are specified at $+25^\circ C$ and represent the most likely parametric norm.

(4) Tested limits are specified to AOQL (Average Outgoing Quality Level).

(5) PSRR is measured as follows: V_{OS} is measured at two supply voltages, $\pm 5V$ and $\pm 15V$. $PSRR = |20\log(\Delta V_{OS}/\Delta V_S)|$.

ELECTRICAL CHARACTERISTICS FOR THE LM4562⁽¹⁾⁽²⁾ (continued)

The specifications apply for $V_S = \pm 15V$, $R_L = 2k\Omega$, $f_{IN} = 1kHz$, $T_A = 25^\circ C$, unless otherwise specified.

| Symbol | Parameter | Conditions | LM4562 | | Units (Limits) |
|------------|---------------------------------|-----------------|------------------------|----------------------|-------------------|
| | | | Typical ⁽³⁾ | Limit ⁽⁴⁾ | |
| C_{LOAD} | Capacitive Load Drive Overshoot | 100pF | 16 | | % |
| I_S | Total Quiescent Current | $I_{OUT} = 0mA$ | 10 | 12 | mA (max) |

TYPICAL PERFORMANCE CHARACTERISTICS

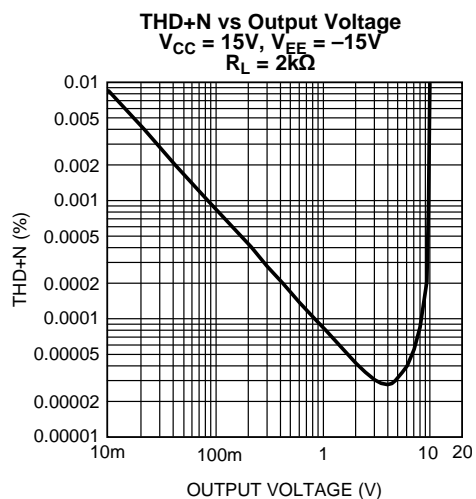


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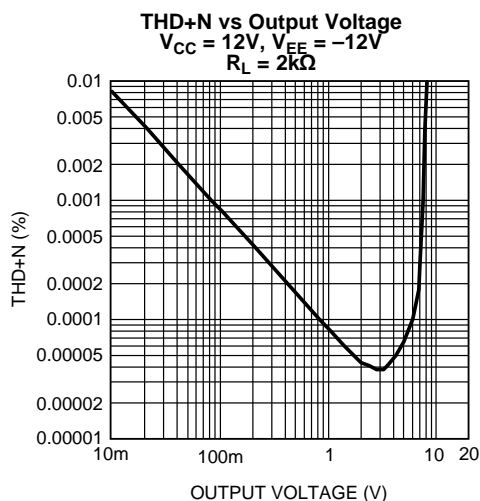


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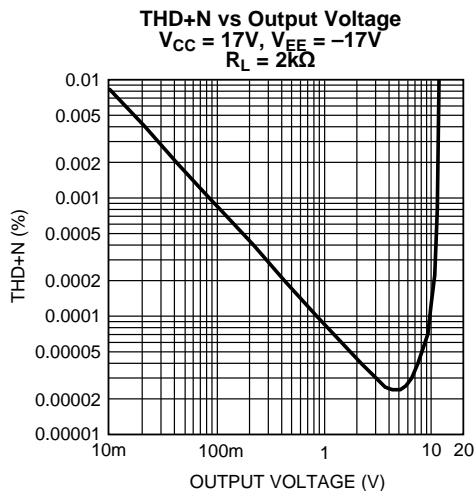


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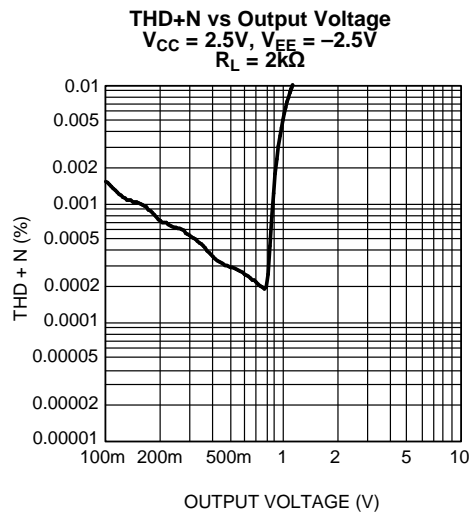


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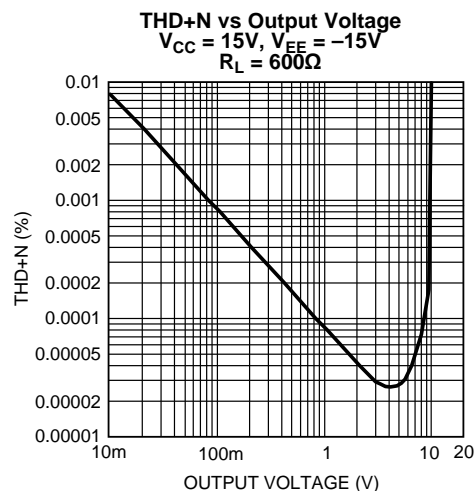


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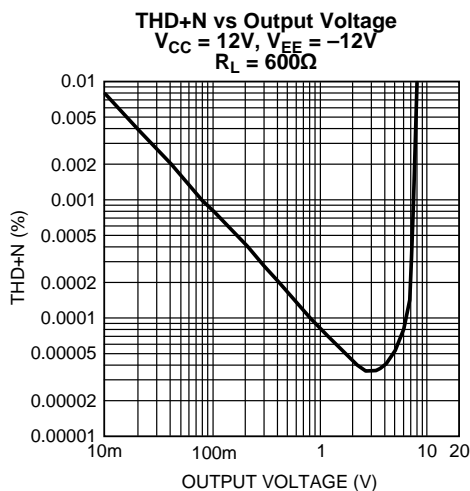


Figure 9.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

THD+N vs Output Voltage
 $V_{CC} = 17V$, $V_{EE} = -17V$
 $R_L = 600\Omega$

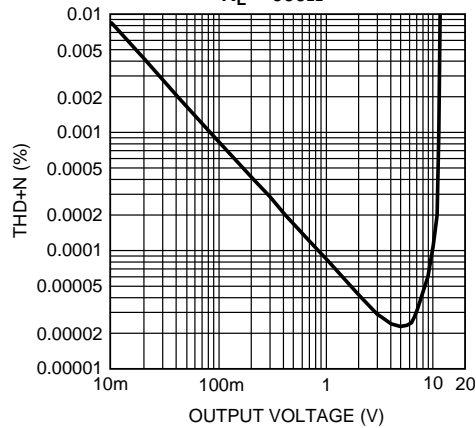


Figure 10.

THD+N vs Output Voltage
 $V_{CC} = 2.5V$, $V_{EE} = -2.5V$
 $R_L = 600\Omega$

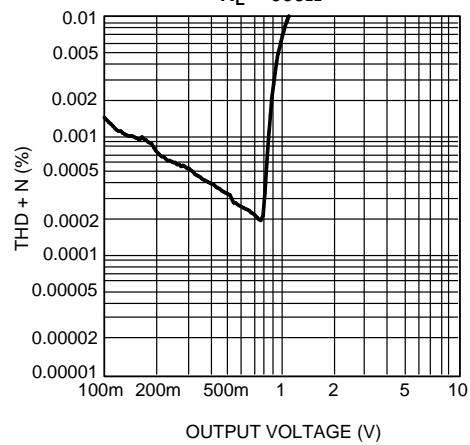


Figure 11.

THD+N vs Output Voltage
 $V_{CC} = 15V$, $V_{EE} = -15V$
 $R_L = 10k\Omega$

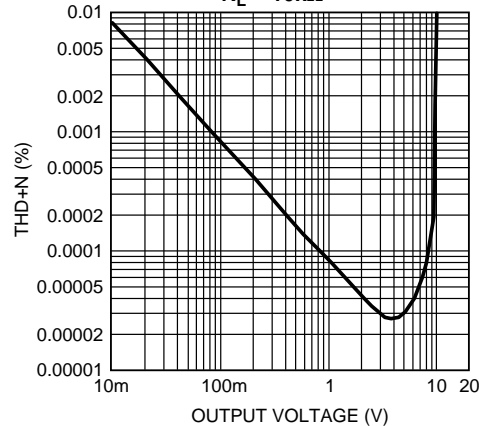


Figure 12.

THD+N vs Output Voltage
 $V_{CC} = 12V$, $V_{EE} = -12V$
 $R_L = 10k\Omega$

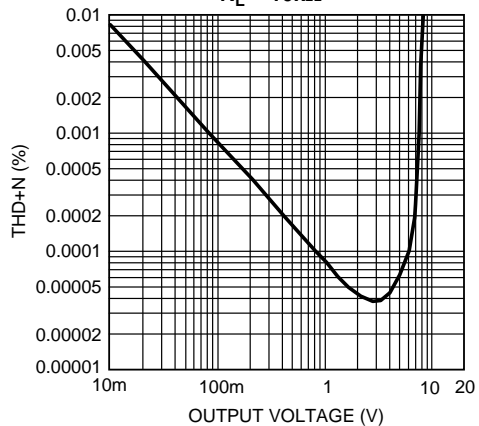


Figure 13.

THD+N vs Output Voltage
 $V_{CC} = 17V$, $V_{EE} = -17V$
 $R_L = 10k\Omega$

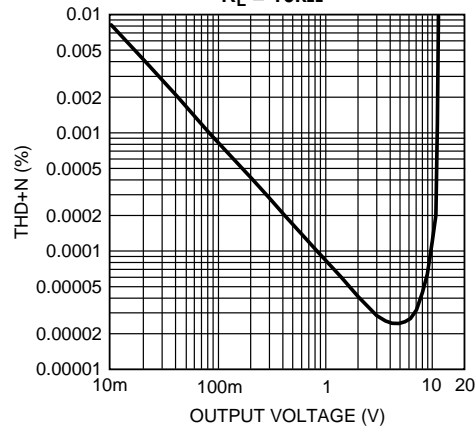


Figure 14.

THD+N vs Output Voltage
 $V_{CC} = 2.5V$, $V_{EE} = -2.5V$
 $R_L = 10k\Omega$

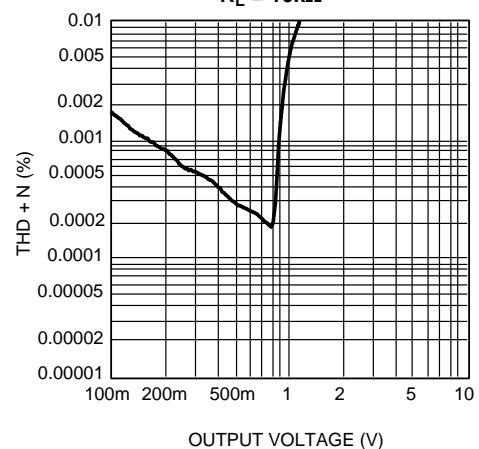


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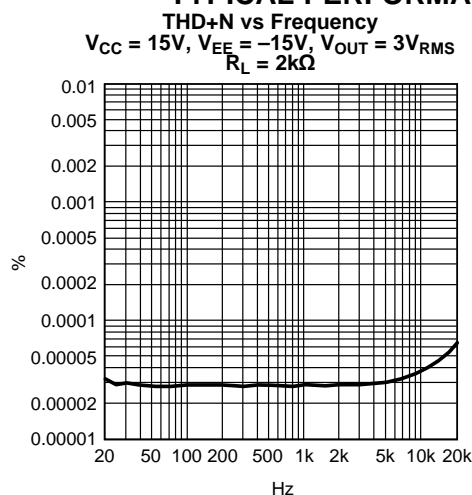
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

Figure 16.

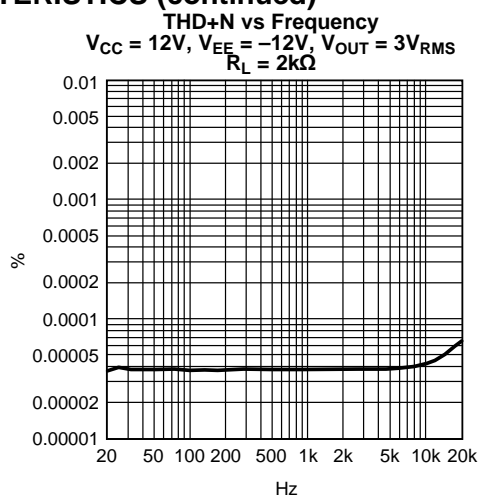


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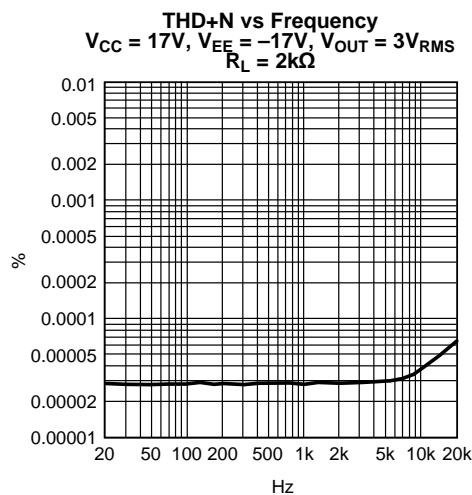


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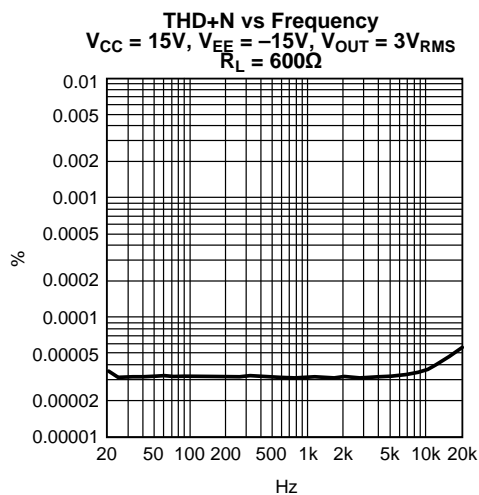


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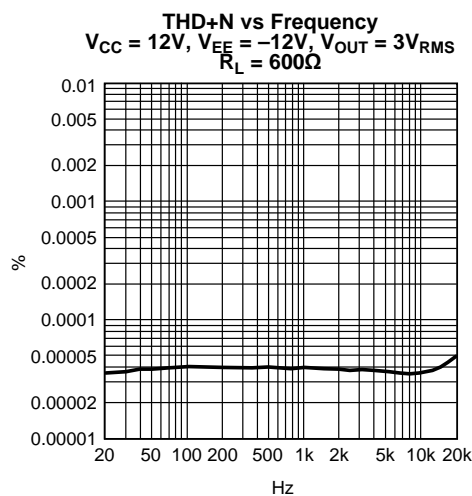


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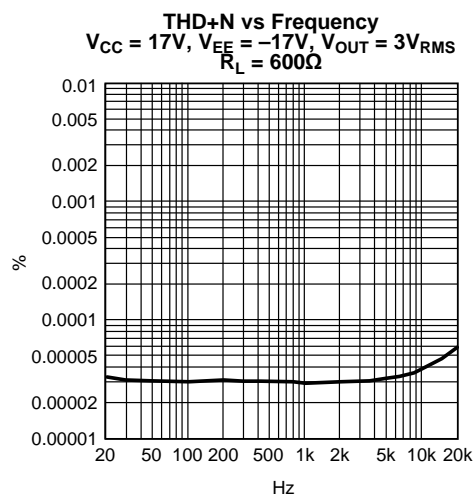


Figure 21.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

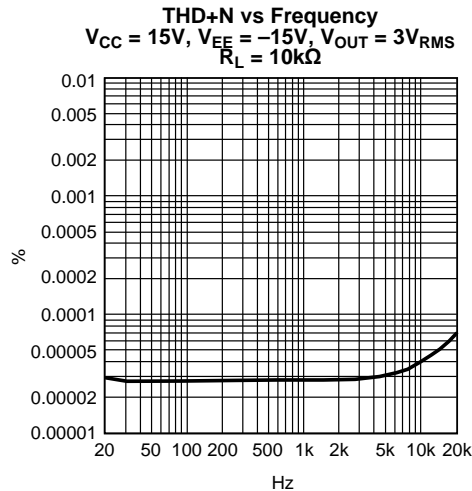


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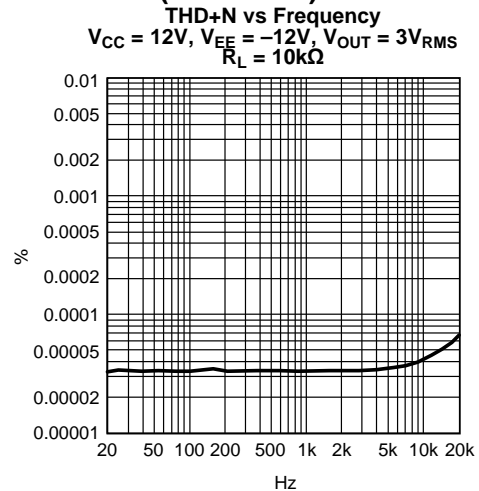


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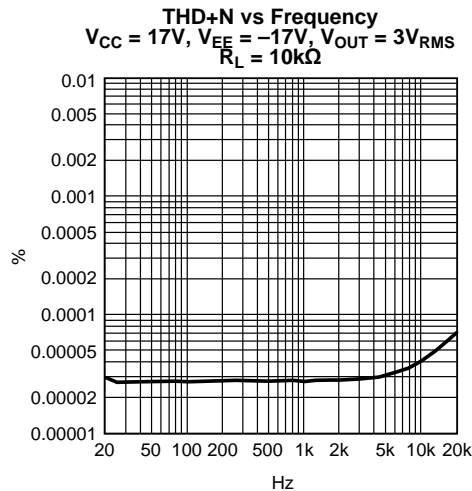


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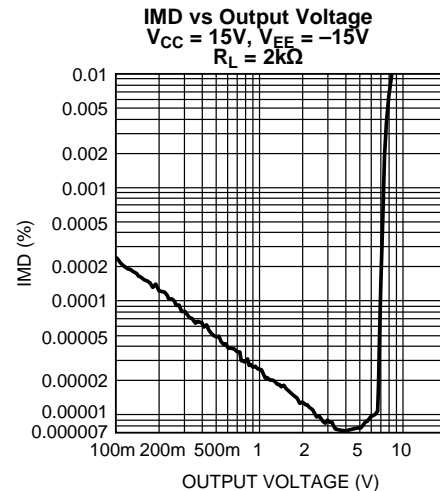


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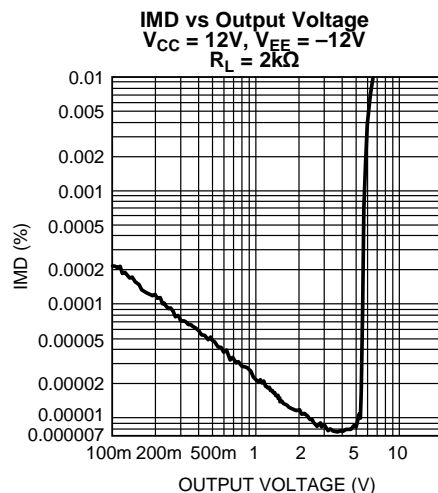


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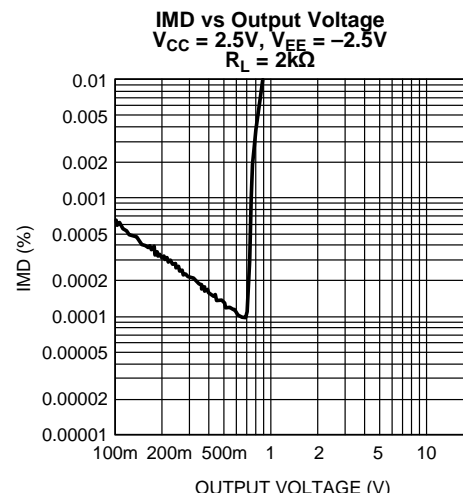
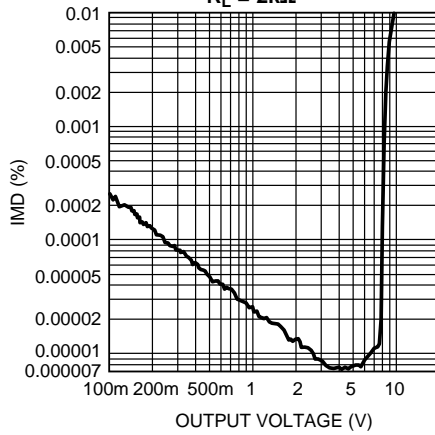


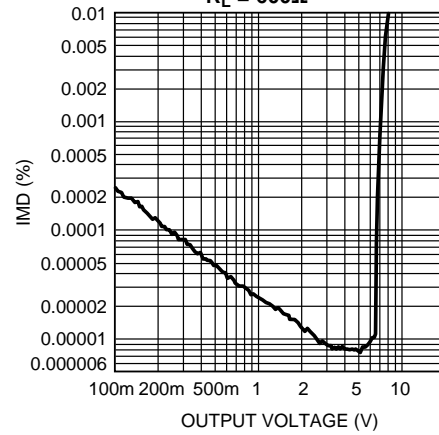
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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

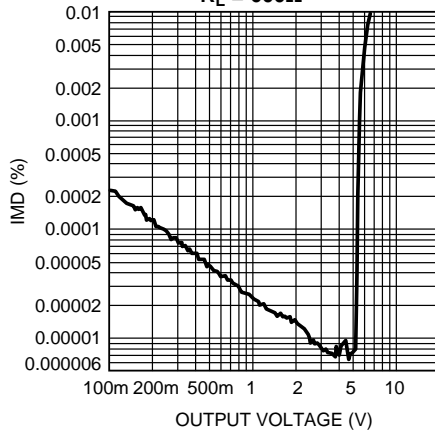
IMD vs Output Voltage
 $V_{CC} = 17V$, $V_{EE} = -17V$
 $R_L = 2k\Omega$

**Figure 28.**

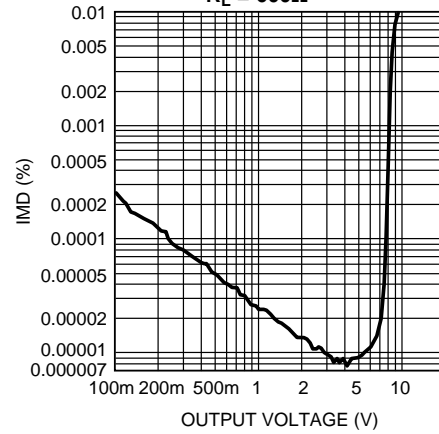
IMD vs Output Voltage
 $V_{CC} = 15V$, $V_{EE} = -15V$
 $R_L = 600\Omega$

**Figure 29.**

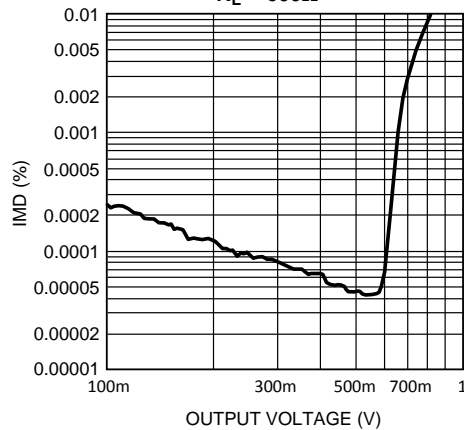
IMD vs Output Voltage
 $V_{CC} = 12V$, $V_{EE} = -12V$
 $R_L = 600\Omega$

**Figure 30.**

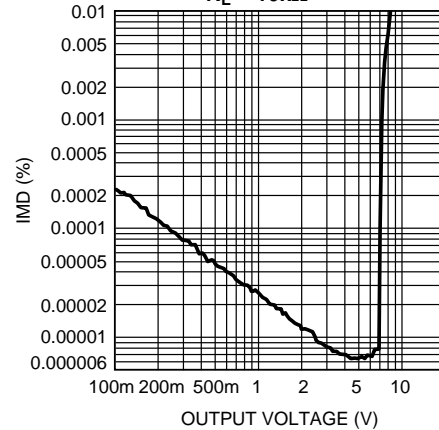
IMD vs Output Voltage
 $V_{CC} = 17V$, $V_{EE} = -17V$
 $R_L = 600\Omega$

**Figure 31.**

IMD vs Output Voltage
 $V_{CC} = 2.5V$, $V_{EE} = -2.5V$
 $R_L = 600\Omega$

**Figure 32.**

IMD vs Output Voltage
 $V_{CC} = 15V$, $V_{EE} = -15V$
 $R_L = 10k\Omega$

**Figure 33.**

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

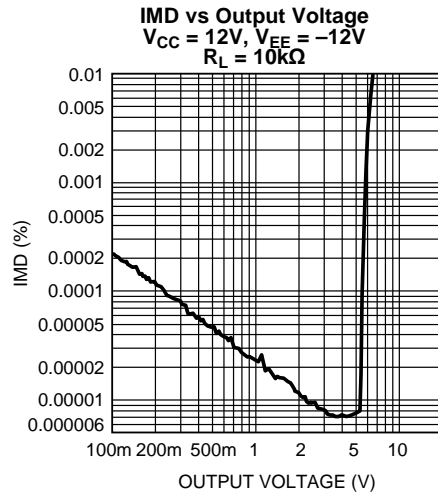


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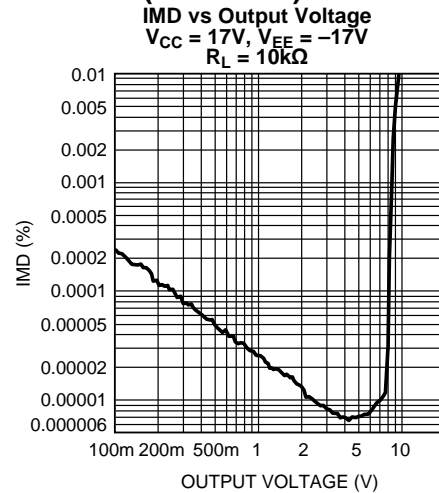


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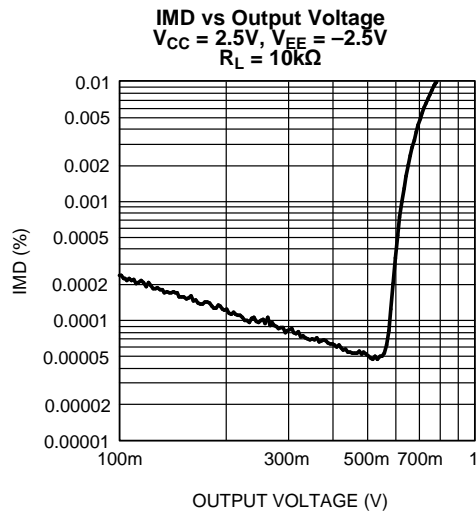


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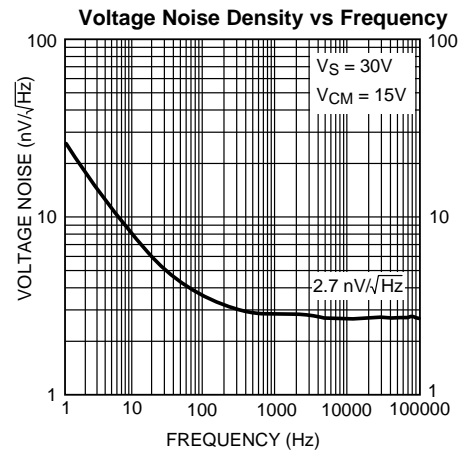


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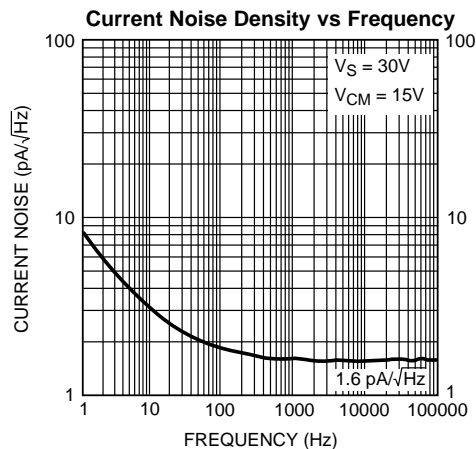


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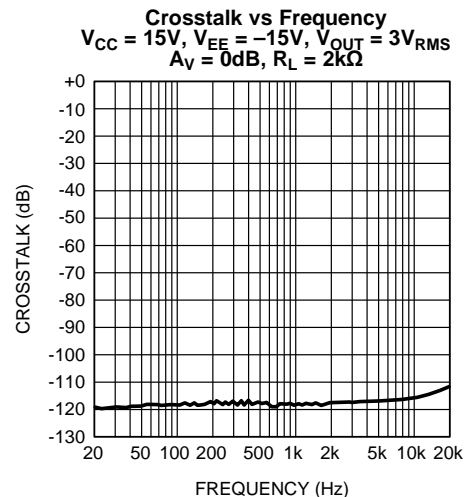
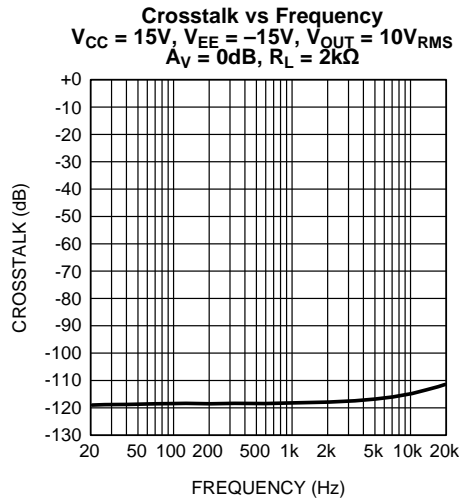
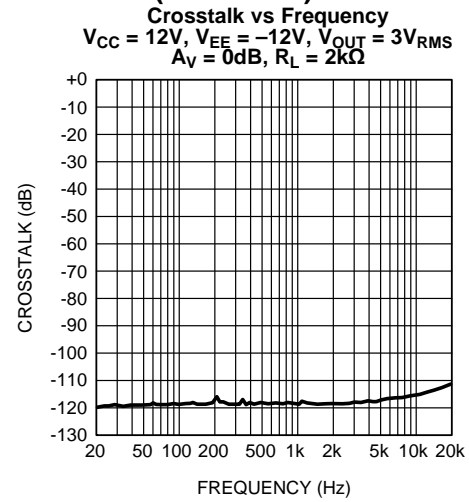
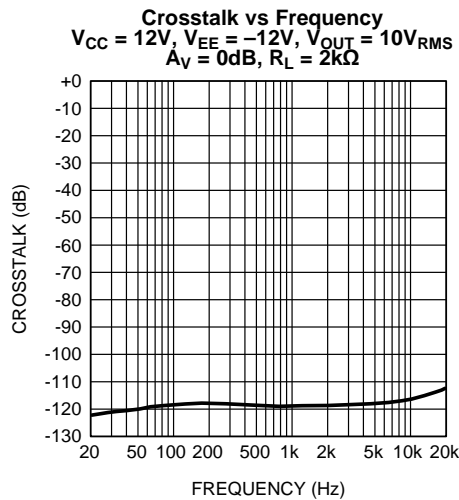
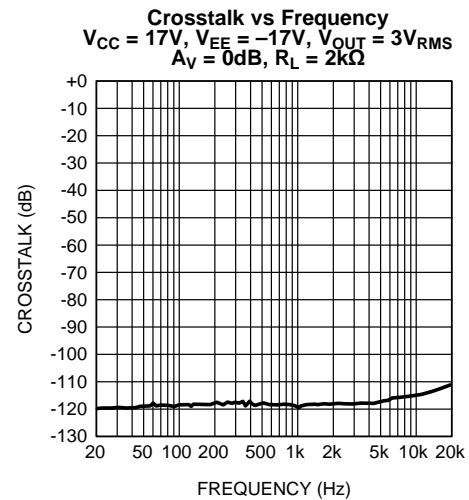
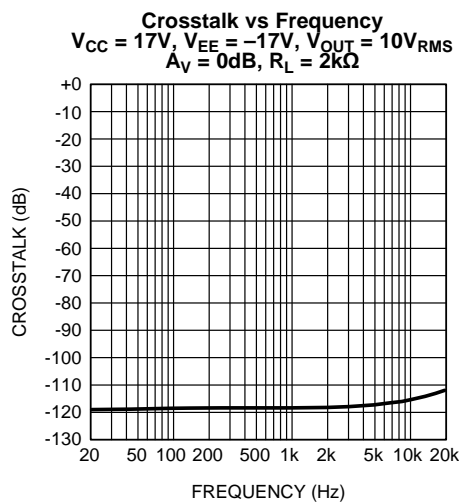
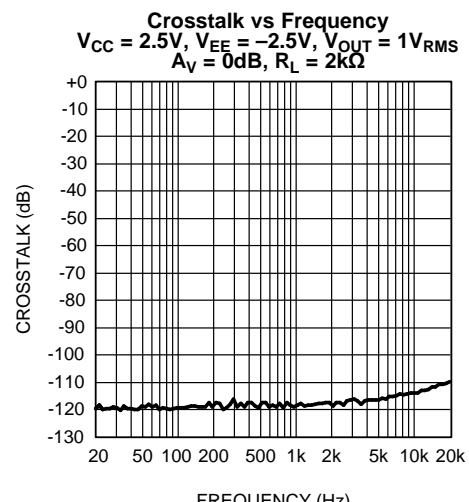


Figure 39.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)**Figure 40.****Figure 41.****Figure 42.****Figure 43.****Figure 44.****Figure 45.**

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

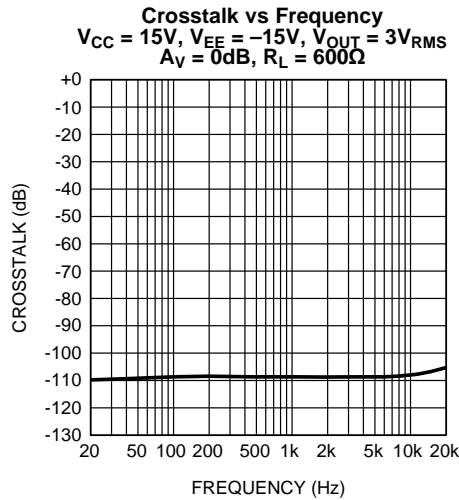


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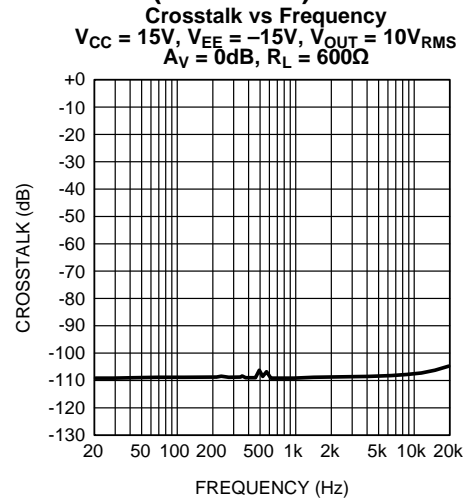


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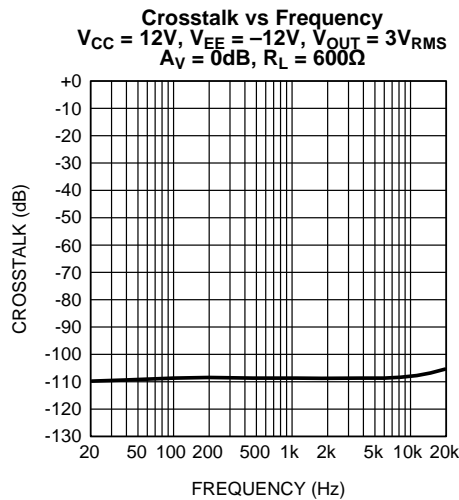


Figure 48.

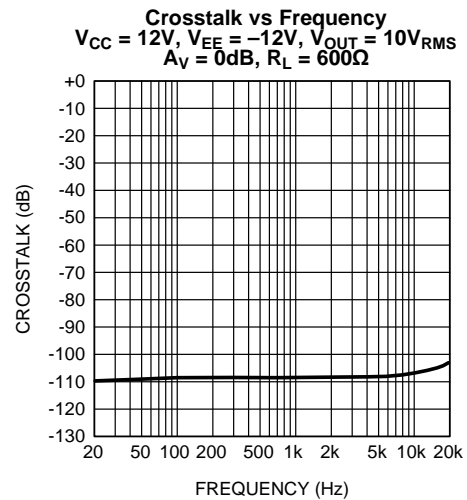


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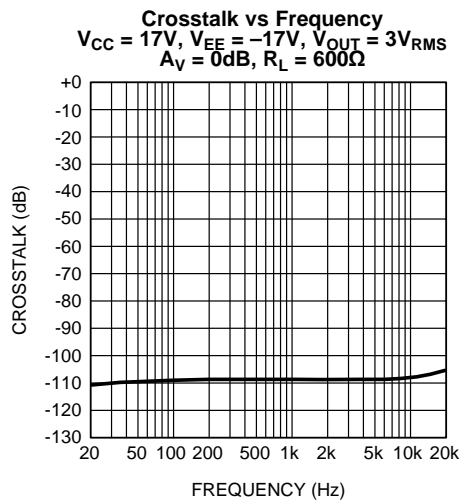


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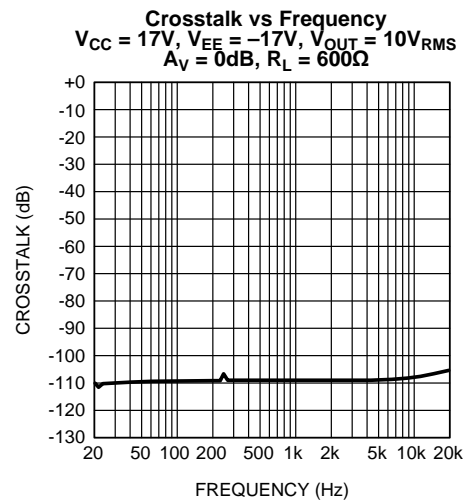
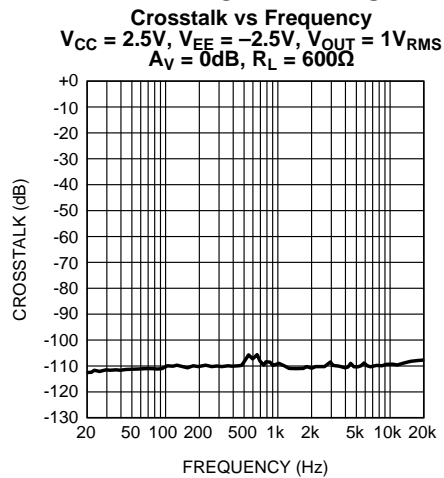
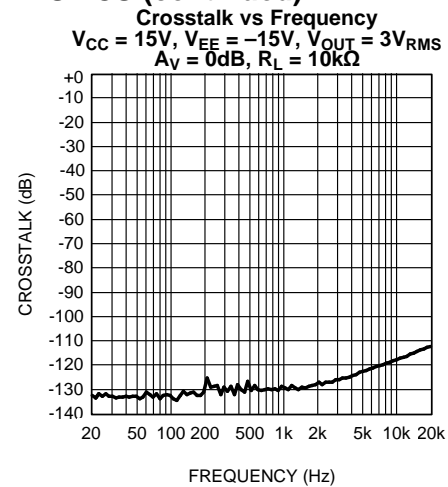
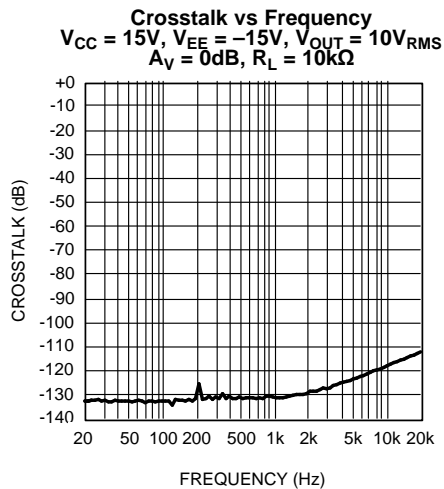
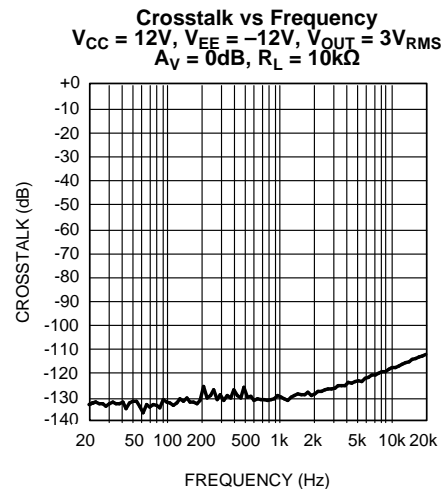
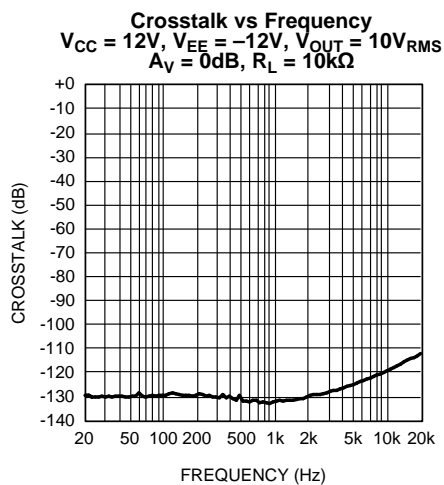
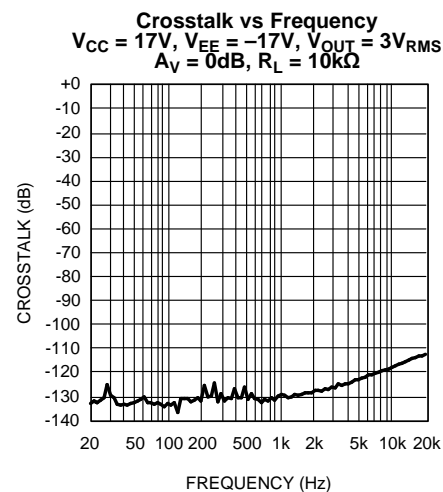


Figure 51.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)**Figure 52.****Figure 53.****Figure 54.****Figure 55.****Figure 56.****Figure 57.**

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

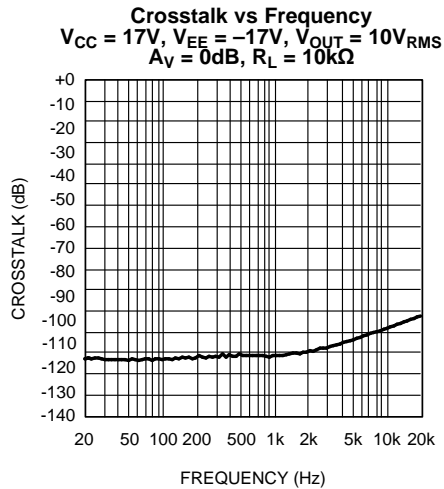


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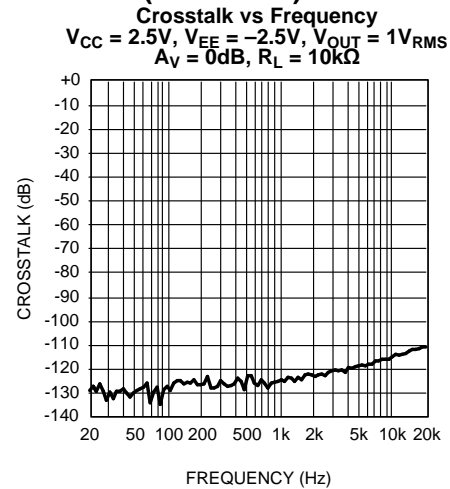


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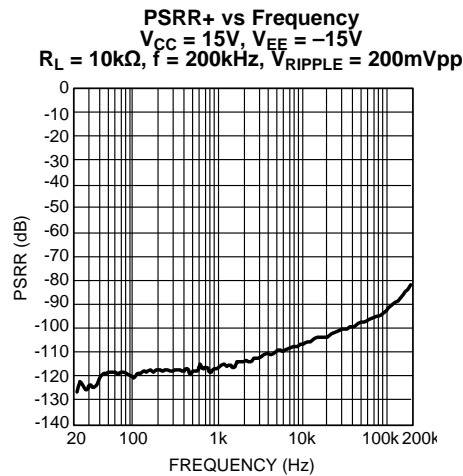


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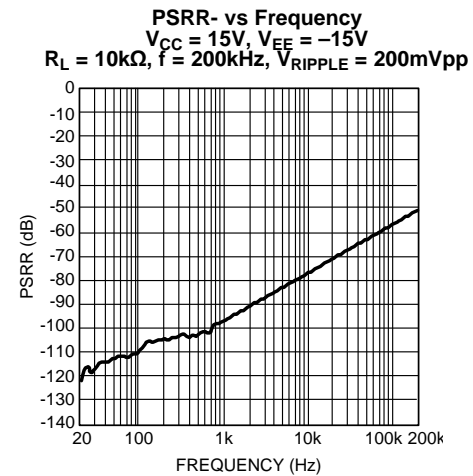


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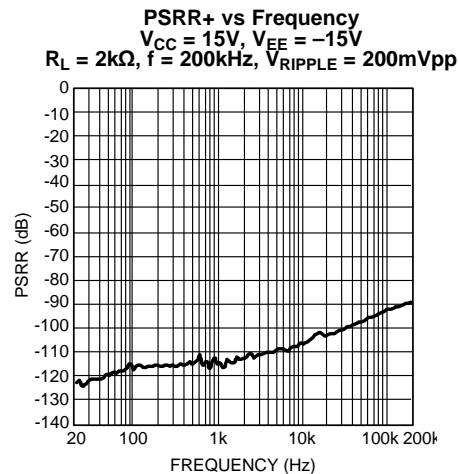


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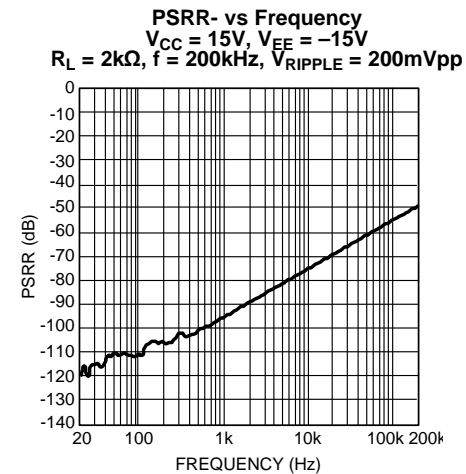


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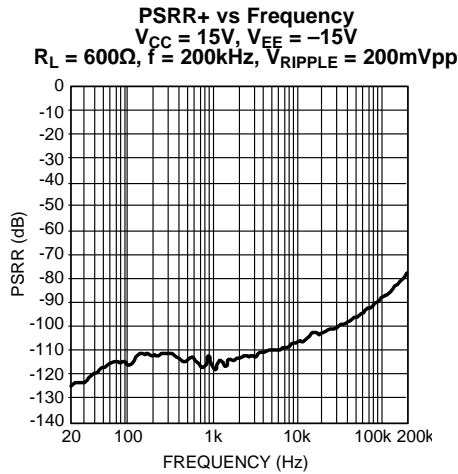
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

Figure 64.

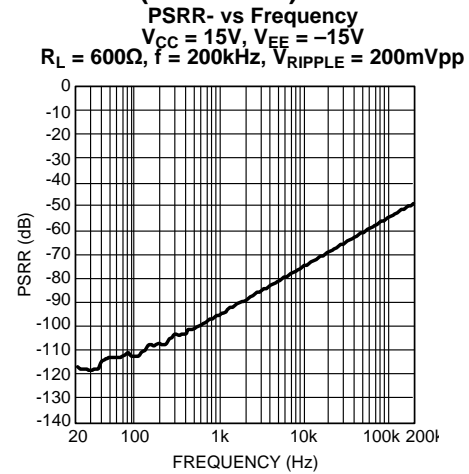


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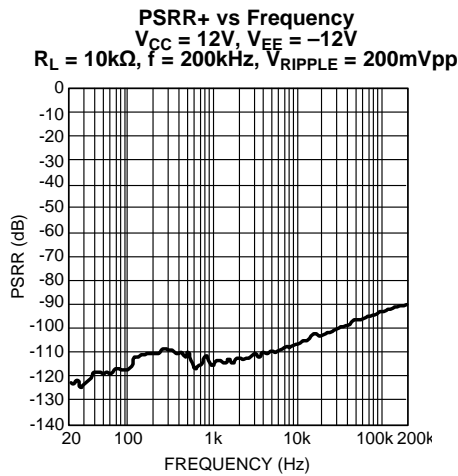


Figure 66.

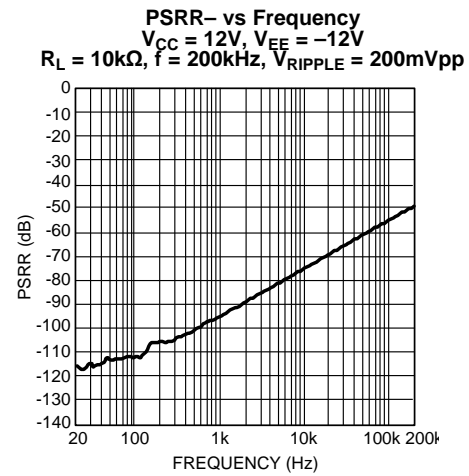


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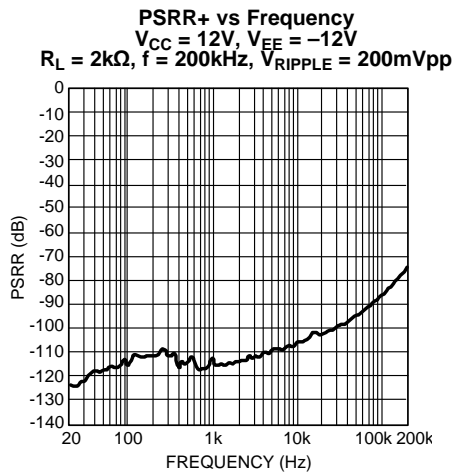


Figure 68.

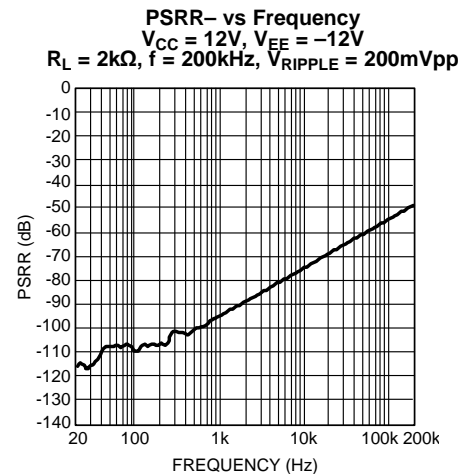


Figure 69.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

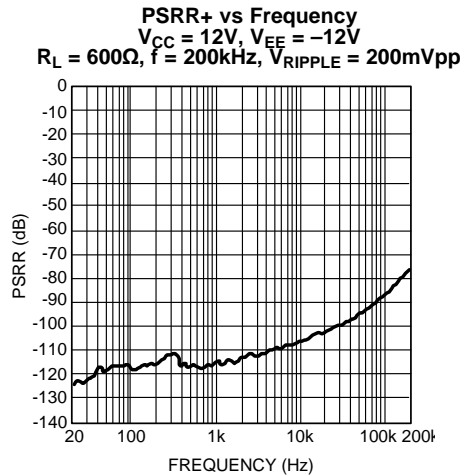


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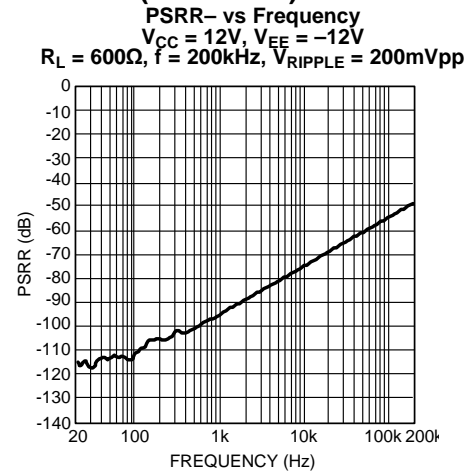


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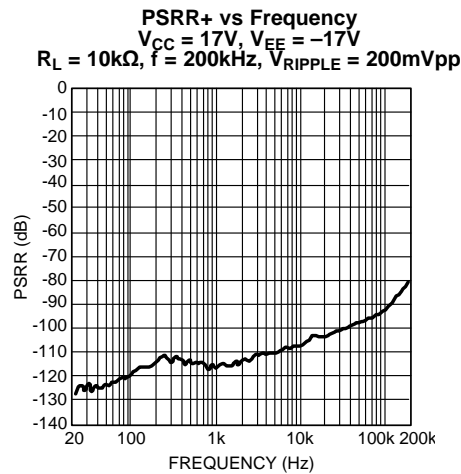


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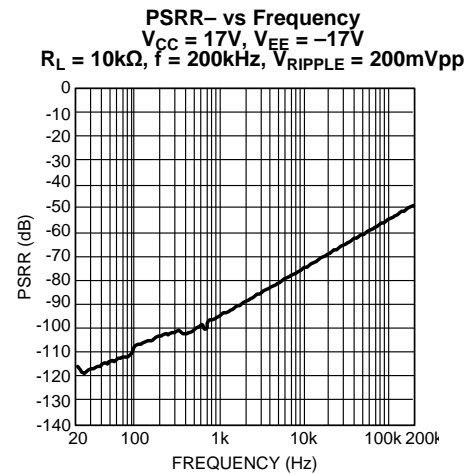


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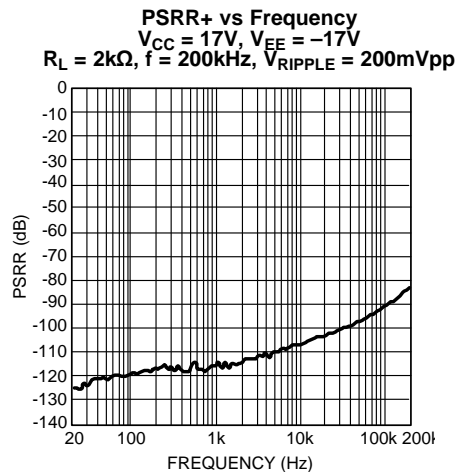


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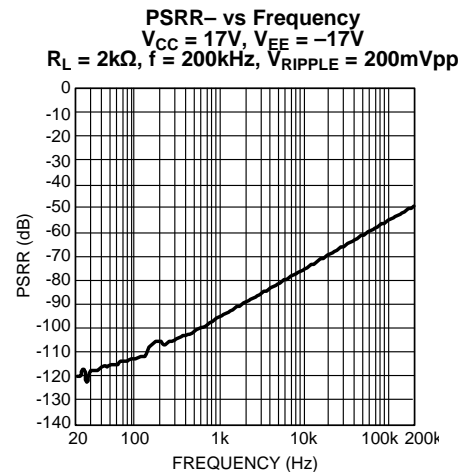


Figure 75.

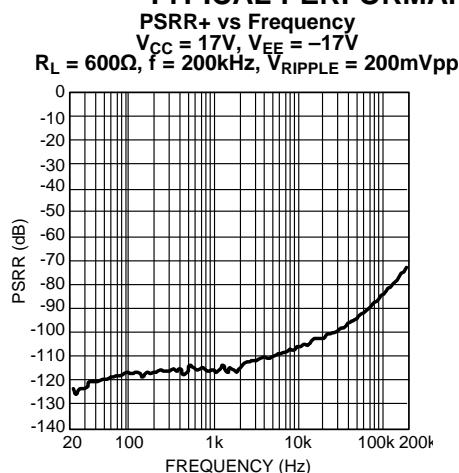
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

Figure 76.

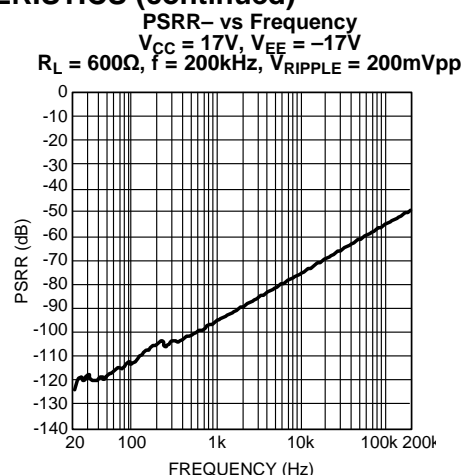


Figure 77.

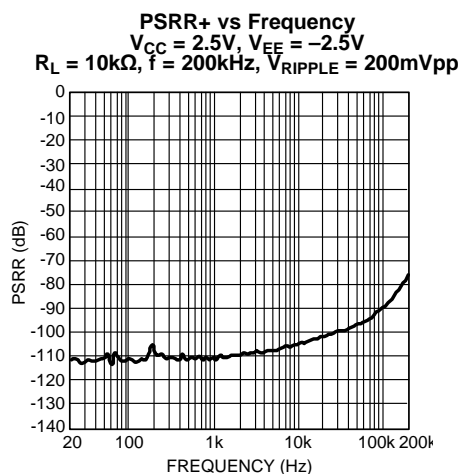


Figure 78.

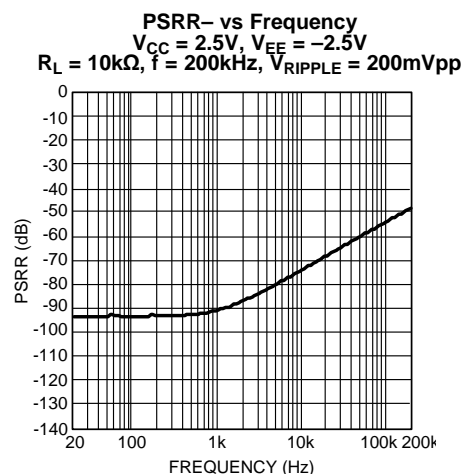


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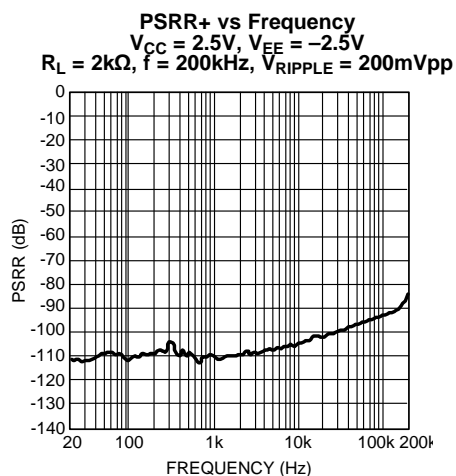


Figure 80.

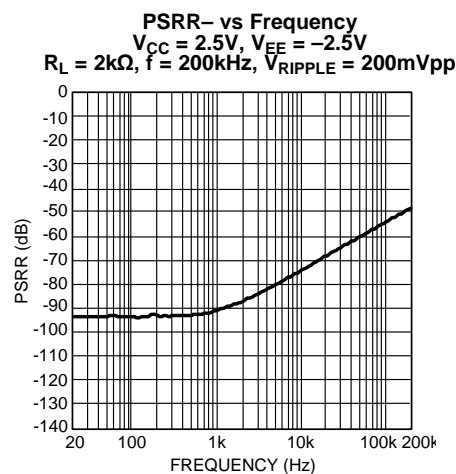


Figure 81.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

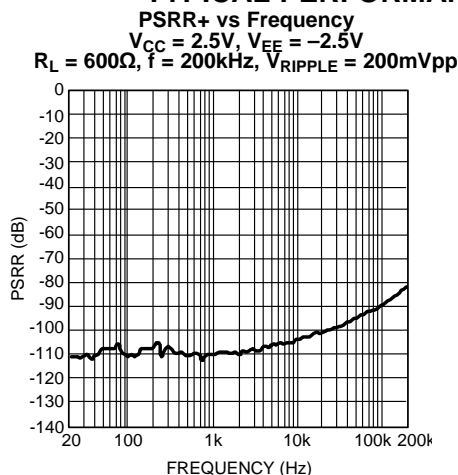


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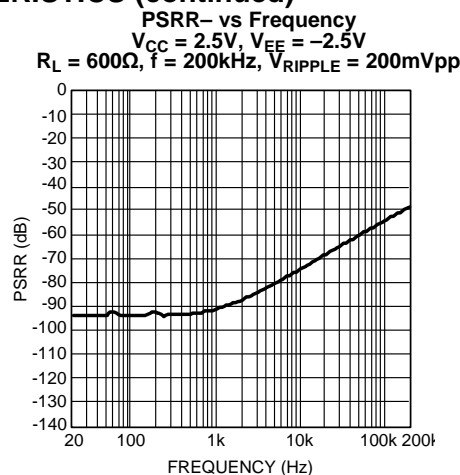


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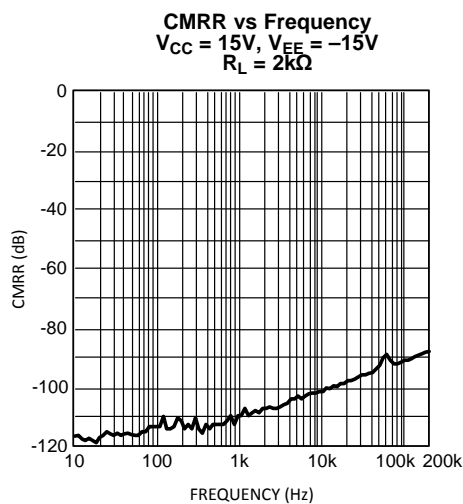


Figure 84.

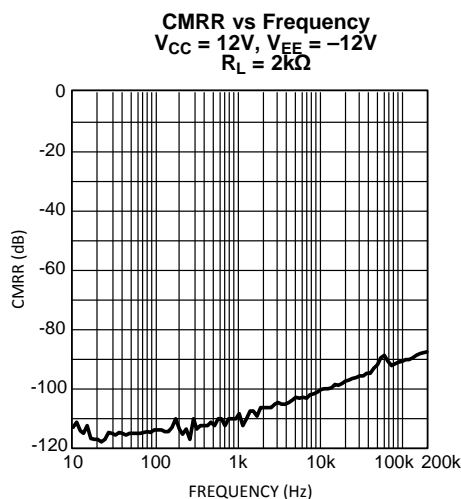


Figure 85.

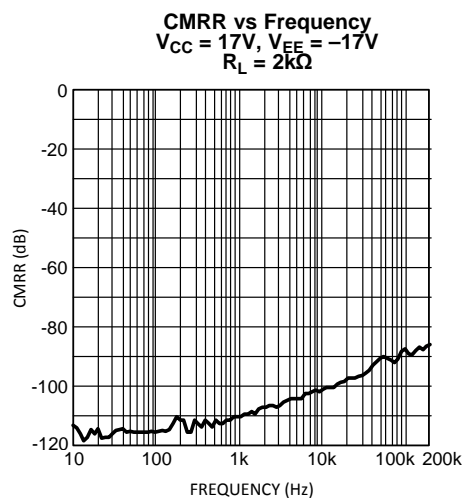


Figure 86.

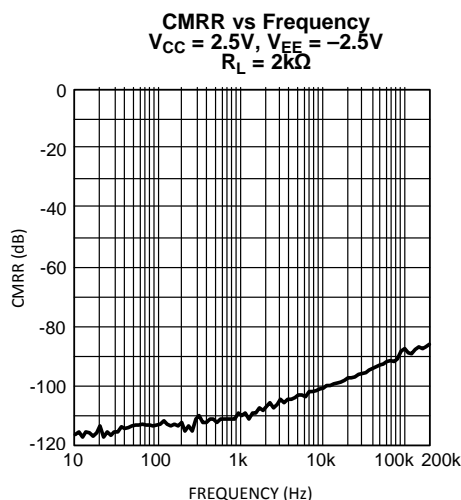


Figure 87.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

CMRR vs Frequency
 $V_{CC} = 15V$, $V_{EE} = -15V$
 $R_L = 600\Omega$

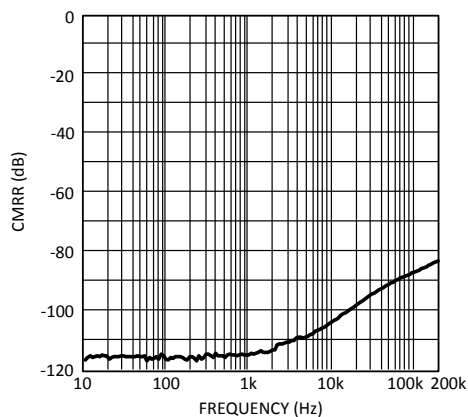


Figure 88.

CMRR vs Frequency
 $V_{CC} = 12V$, $V_{EE} = -12V$
 $R_L = 600\Omega$

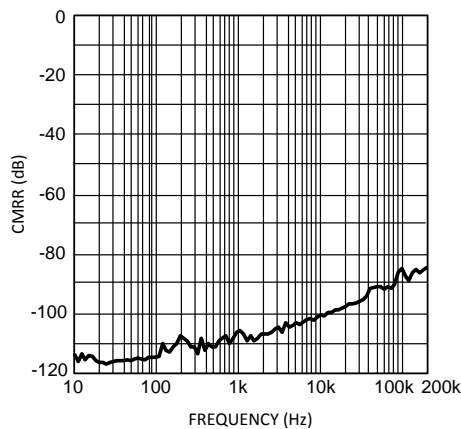


Figure 89.

CMRR vs Frequency
 $V_{CC} = 17V$, $V_{EE} = -17V$
 $R_L = 600\Omega$

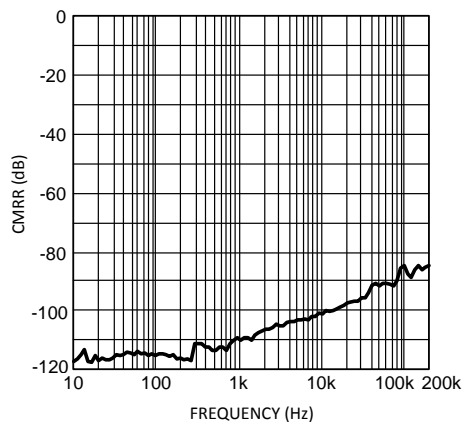


Figure 90.

CMRR vs Frequency
 $V_{CC} = 2.5V$, $V_{EE} = -2.5V$
 $R_L = 600\Omega$

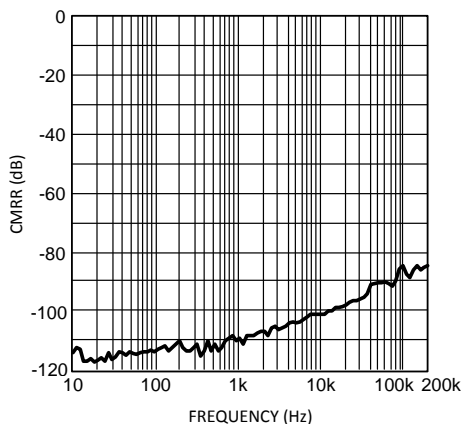


Figure 91.

CMRR vs Frequency
 $V_{CC} = 15V$, $V_{EE} = -15V$
 $R_L = 10k\Omega$

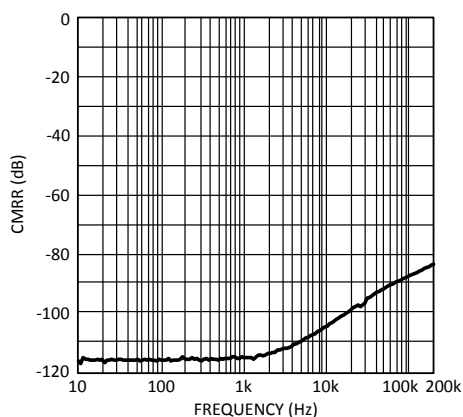


Figure 92.

CMRR vs Frequency
 $V_{CC} = 12V$, $V_{EE} = -12V$
 $R_L = 10k\Omega$

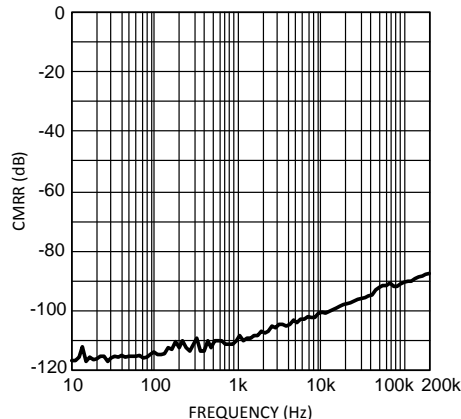


Figure 93.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

CMRR vs Frequency
 $V_{CC} = 17V$, $V_{EE} = -17V$
 $R_L = 10k\Omega$

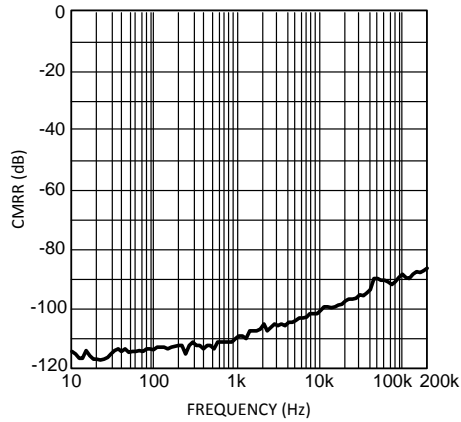


Figure 94.

CMRR vs Frequency
 $V_{CC} = 2.5V$, $V_{EE} = -2.5V$
 $R_L = 10k\Omega$

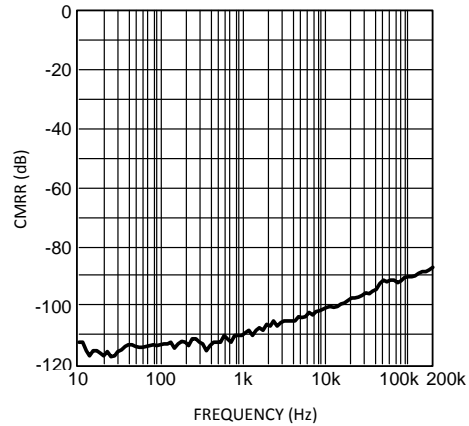


Figure 95.

Output Voltage vs Load Resistance
 $V_{DD} = 15V$, $V_{EE} = -15V$
 $THD+N = 1\%$

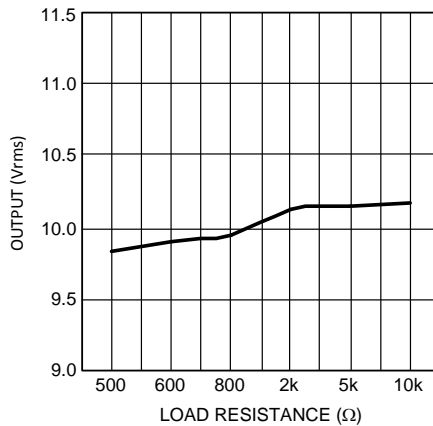


Figure 96.

Output Voltage vs Load Resistance
 $V_{DD} = 12V$, $V_{EE} = -12V$
 $THD+N = 1\%$

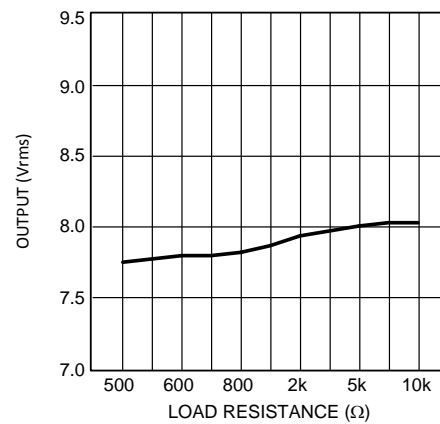


Figure 97.

Output Voltage vs Load Resistance
 $V_{DD} = 17V$, $V_{EE} = -17V$
 $THD+N = 1\%$

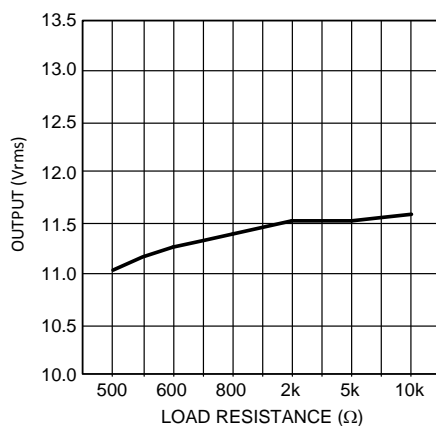


Figure 98.

Output Voltage vs Load Resistance
 $V_{DD} = 2.5V$, $V_{EE} = -2.5V$
 $THD+N = 1\%$

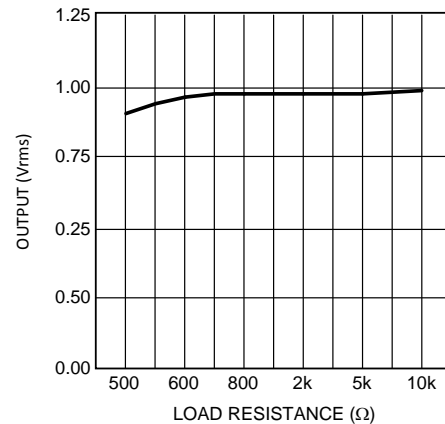
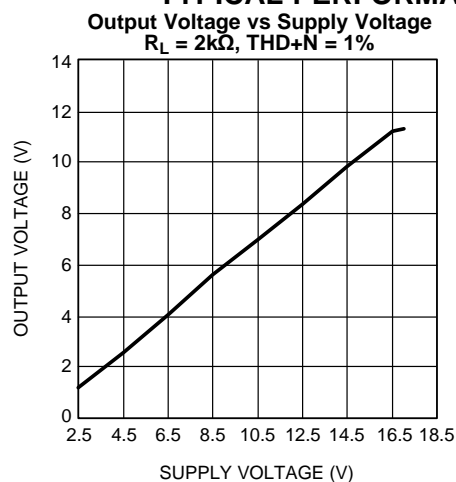
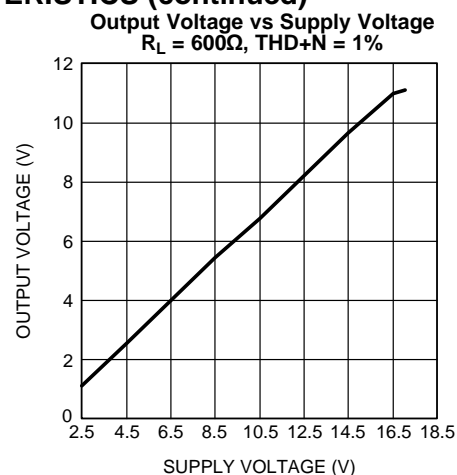
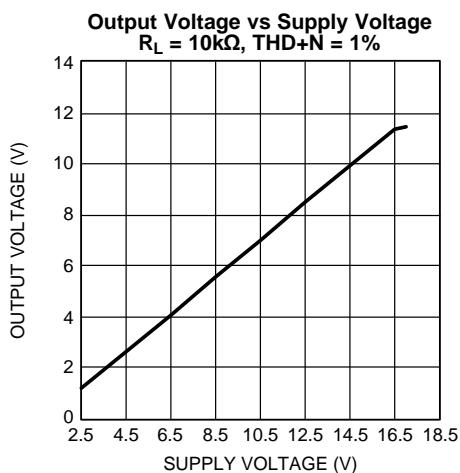
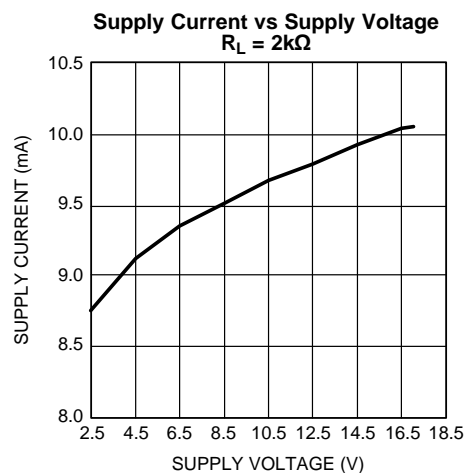
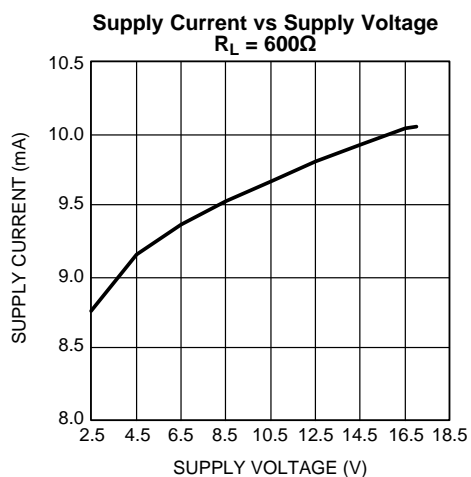
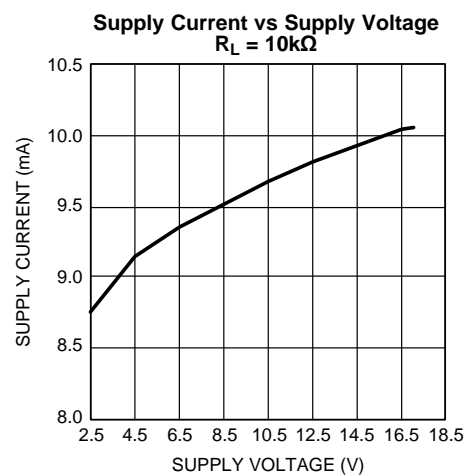


Figure 99.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)**Figure 100.****Figure 101.****Figure 102.****Figure 103.****Figure 104.****Figure 105.**

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

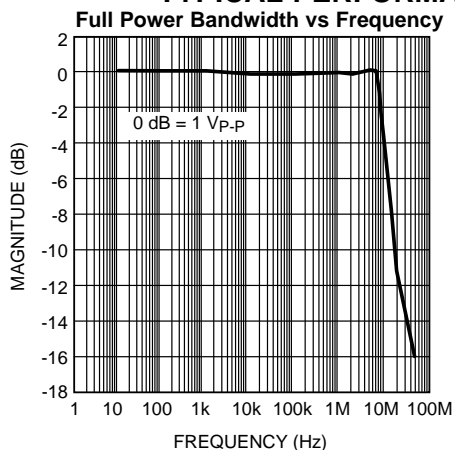


Figure 106.

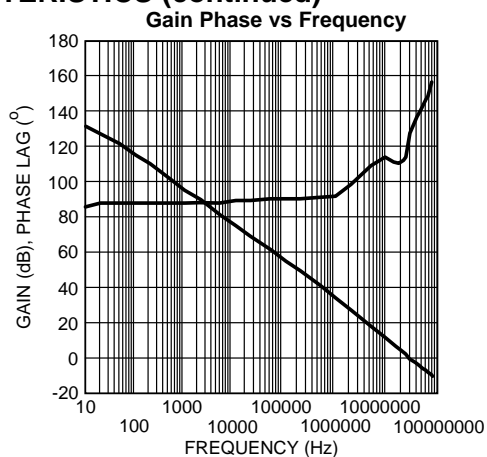


Figure 107.

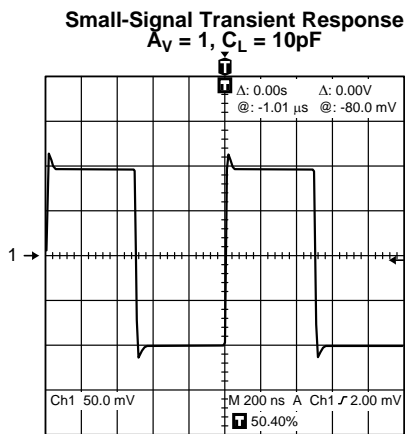


Figure 108.

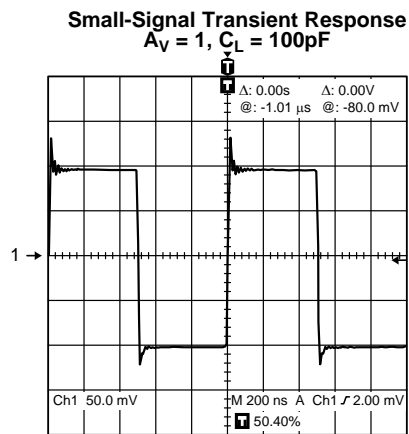


Figure 109.

APPLICATION INFORMATION

DISTORTION MEASUREMENTS

The vanishingly low residual distortion produced by LM4562 is below the capabilities of all commercially available equipment. This makes distortion measurements just slightly more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LM4562's low residual distortion is an input referred internal error. As shown in Figure 110, adding the 10Ω resistor connected between the amplifier's inverting and non-inverting inputs changes the amplifier's noise gain. The result is that the error signal (distortion) is amplified by a factor of 101. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced by 101, which means that measurement resolution increases by 101. To ensure minimum effects on distortion measurements, keep the value of R1 low as shown in Figure 110.

This technique is verified by duplicating the measurements with high closed loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities. This datasheet's THD+N and IMD values were generated using the above described circuit connected to an Audio Precision System Two Cascade.

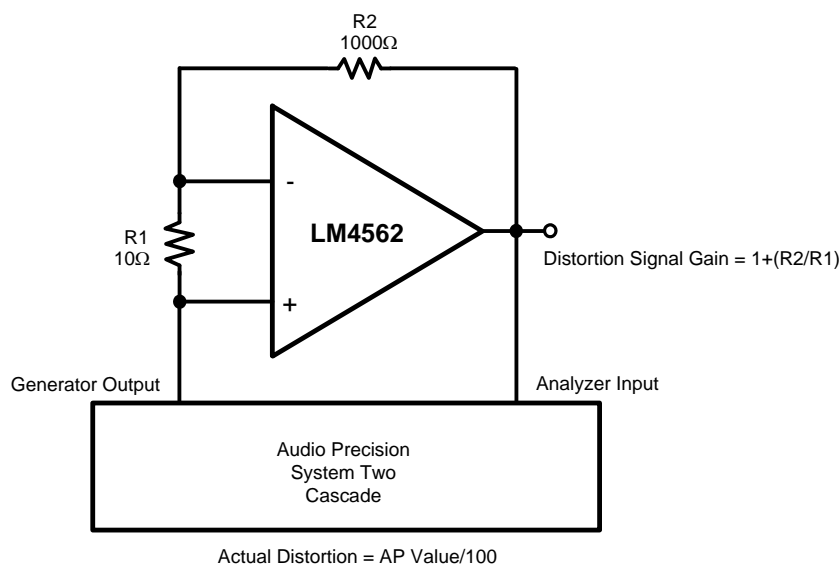
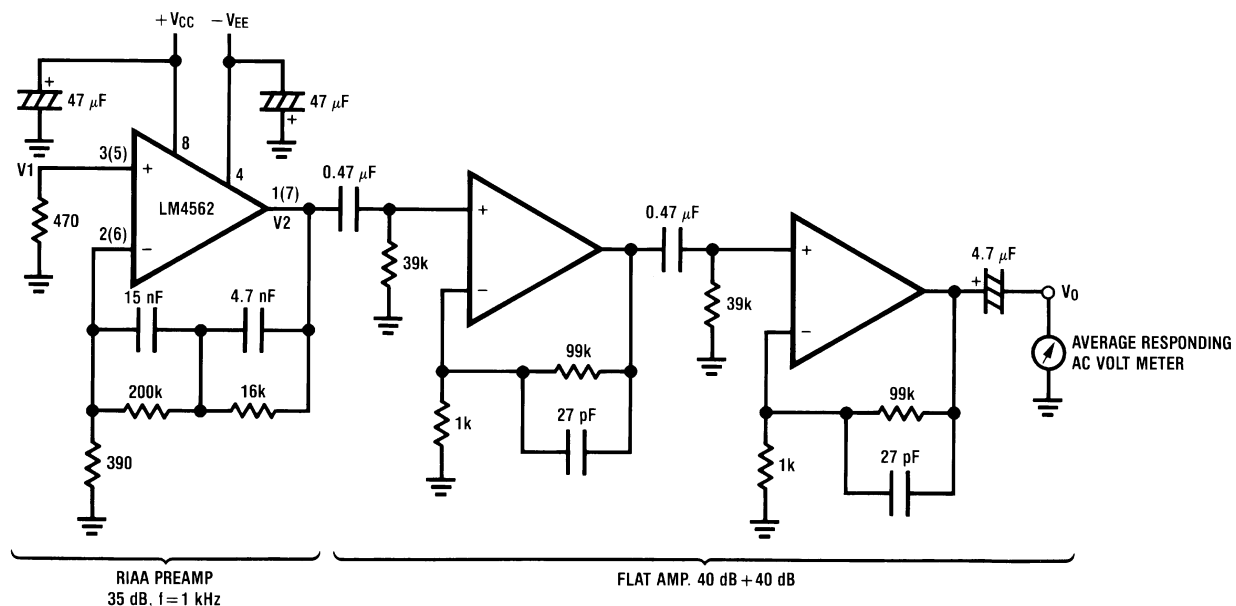


Figure 110. THD+N and IMD Distortion Test Circuit

The LM4562 is a high-speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.



- A. Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

Figure 111. Noise Measurement Circuit
Total Gain: 115 dB @ $f = 1$ kHz
Input Referred Noise Voltage: $e_n = V_0/560,000$ (V)

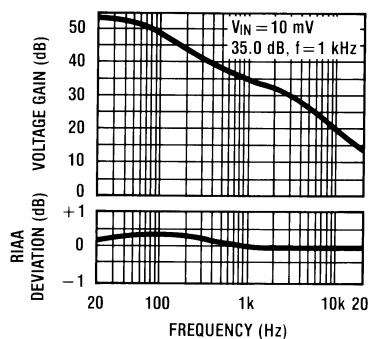


Figure 112. RIAA Preamp Voltage Gain, RIAA Deviation vs Frequency

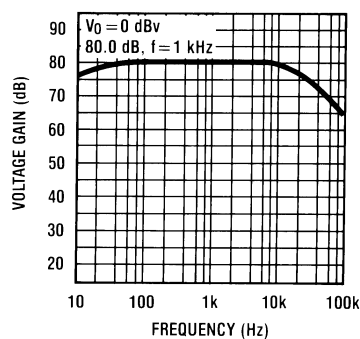
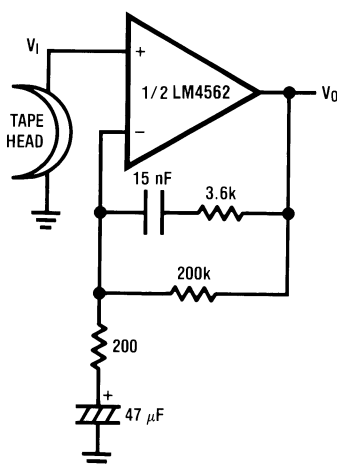


Figure 113. Flat Amp Voltage Gain vs Frequency

TYPICAL APPLICATIONS



$A_V = 34.5$
 $F = 1 \text{ kHz}$
 $E_n = 0.38 \mu\text{V}$
 A Weighted

Figure 114. NAB Preamp

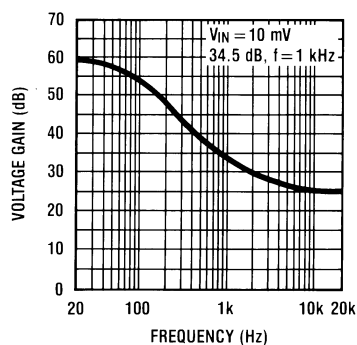
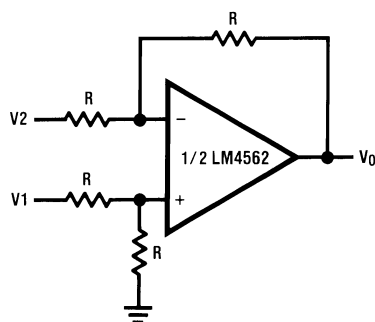
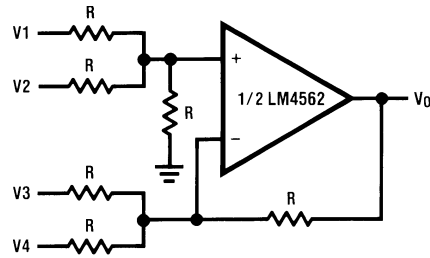


Figure 115. NAB Preamp Voltage Gain vs Frequency



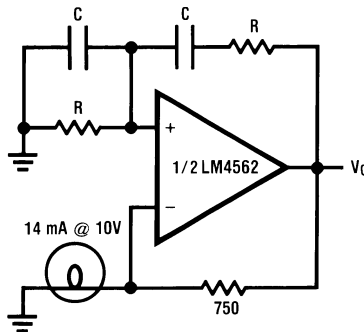
$$V_O = V_1 - V_2$$

Figure 116. Balanced to Single-Ended Converter



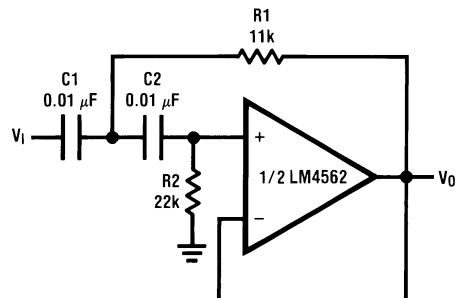
$$V_O = V_1 + V_2 - V_3 - V_4$$

Figure 117. Adder/Subtractor



$$f_o = \frac{1}{2\pi RC}$$

Figure 118. Sine Wave Oscillator



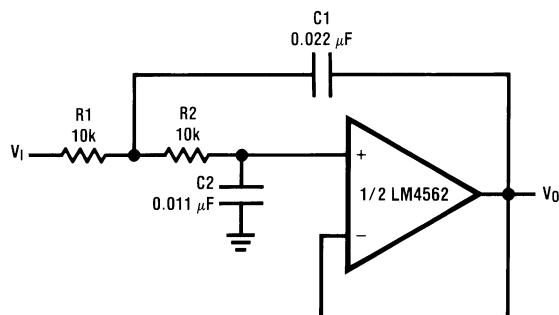
if $C_1 = C_2 = C$

$$R_1 = \frac{\sqrt{2}}{2\omega_o C}$$

$$R_2 = 2 \cdot R_1$$

Illustration is $f_o = 1 \text{ kHz}$

Figure 119. Second-Order High-Pass Filter (Butterworth)



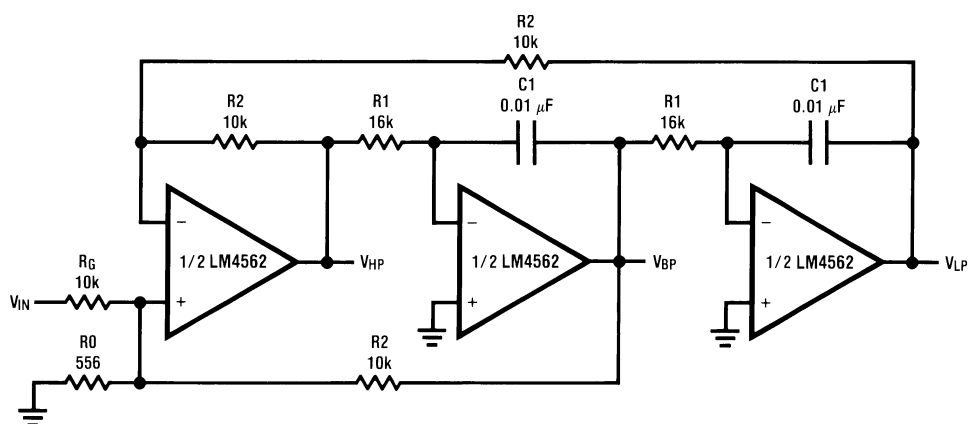
if $R1 = R2 = R$

$$C1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C2 = \frac{C1}{2}$$

Illustration is $f_0 = 1$ kHz

Figure 120. Second-Order Low-Pass Filter (Butterworth)



$$f_0 = \frac{1}{2\pi C1 R1}, Q = \frac{1}{2} \left(1 + \frac{R2}{R0} + \frac{R2}{RG} \right), A_{BP} = QA_{LP} = QA_{LH} = \frac{R2}{RG}$$

Illustration is $f_0 = 1$ kHz, $Q = 10$, $A_{BP} = 1$

Figure 121. State Variable Filter

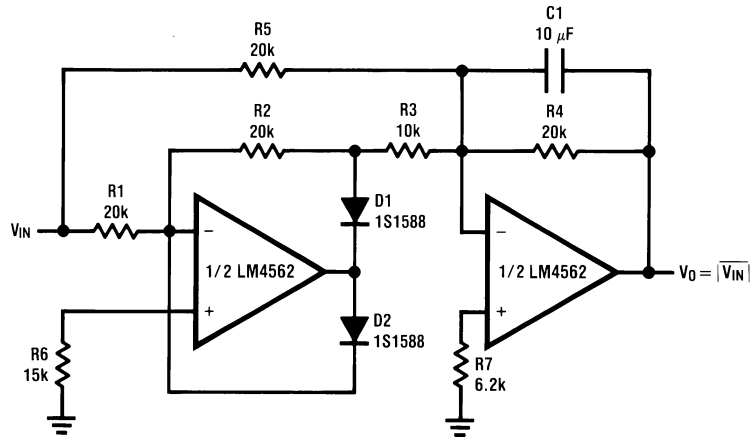


Figure 122. AC/DC Converter

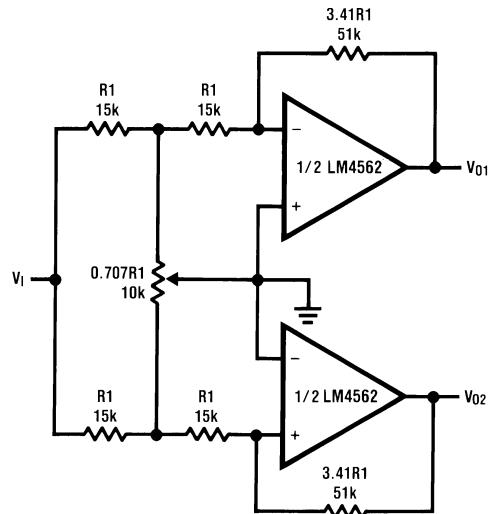


Figure 123. 2-Channel Panning Circuit (Pan Pot)

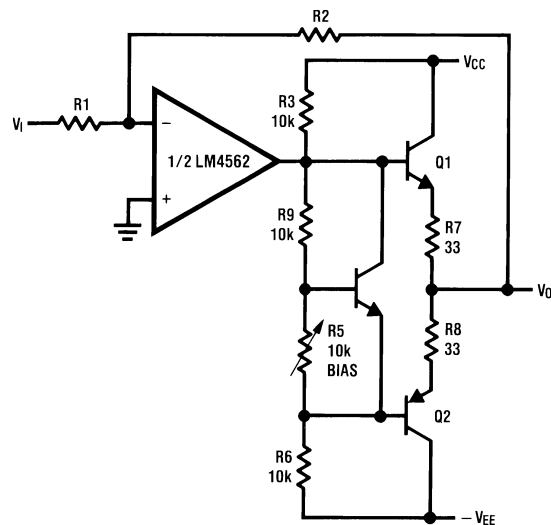
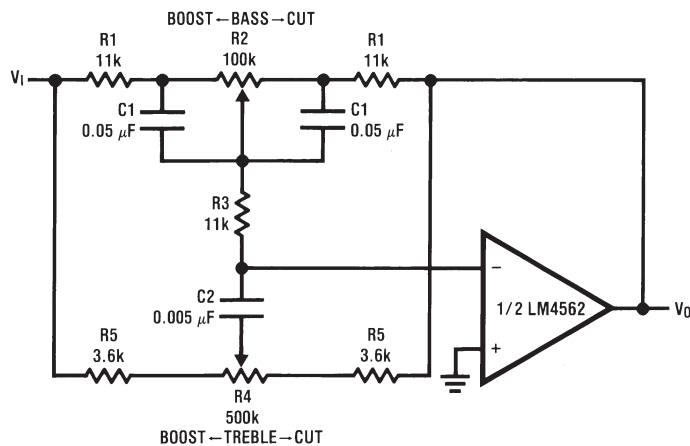


Figure 124. Line Driver



$$f_L \approx \frac{1}{2\pi R_2 C_1}, f_{LB} \approx \frac{1}{2\pi R_1 C_1}$$

$$f_H \approx \frac{1}{2\pi R_5 C_2}, f_{HB} \approx \frac{1}{2\pi (R_1 + R_5 + 2R_3) C_2}$$

The equations started above are simplifications, providing guidance of general -3dB point values, when the potentiometers are at their null position.

Illustration is:

$$f_L \approx 32 \text{ Hz}, f_{LB} \approx 320 \text{ Hz}$$

$$f_H \approx 11 \text{ kHz}, f_{HB} \approx 1.1 \text{ kHz}$$

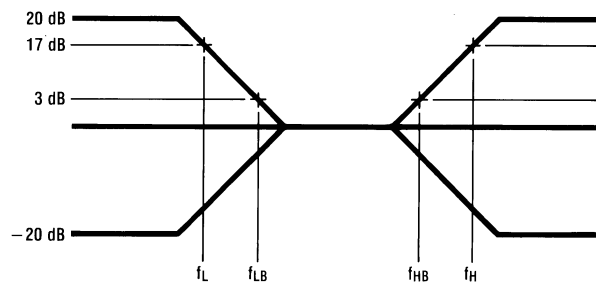
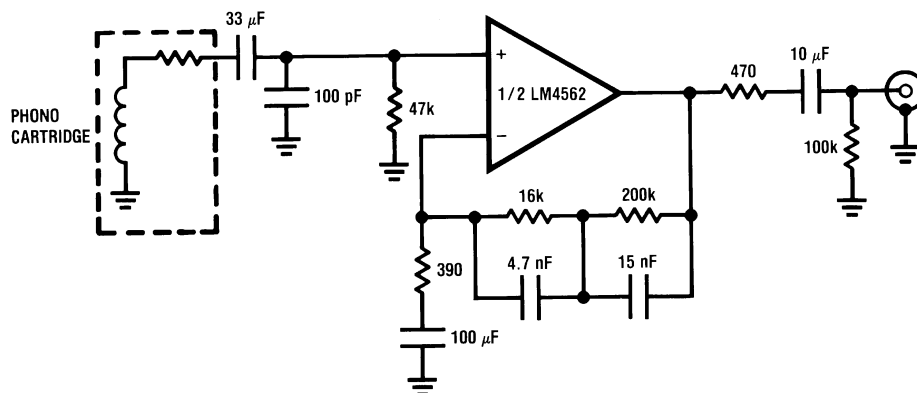
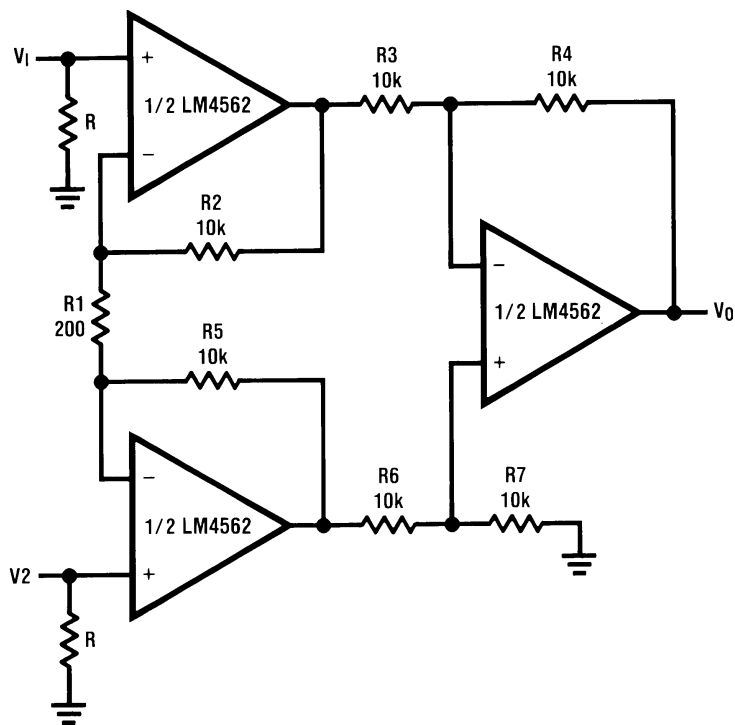


Figure 125. Tone Control



$A_v = 35 \text{ dB}$
 $E_n = 0.33 \text{ } \mu\text{V S/N} = 90 \text{ dB}$
 $f = 1 \text{ kHz}$
 A Weighted
 A Weighted, $V_{IN} = 10 \text{ mV}$
 @ $f = 1 \text{ kHz}$

Figure 126. RIAA Preamp



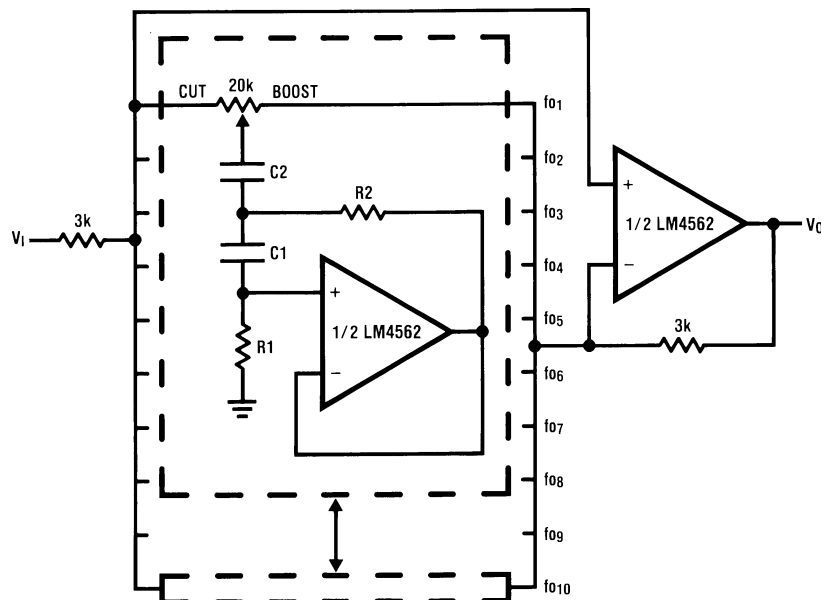
If $R_2 = R_5$, $R_3 = R_6$, $R_4 = R_7$

$$V_0 = \left(1 + \frac{2R_2}{R_1} \right) \frac{R_4}{R_3} (V_2 - V_1)$$

Illustration is:

$$V_0 = 101(V_2 - V_1)$$

Figure 127. Balanced Input Mic Amp



A. See [Table 1](#).

Figure 128. 10-Band Graphic Equalizer

Table 1. C_1 , C_2 , R_1 , and R_2 Values for [Figure 128](#)⁽¹⁾

| f_o (Hz) | C_1 | C_2 | R_1 | R_2 |
|------------|---------------|---------------|--------------|--------------|
| 32 | 0.12 μ F | 4.7 μ F | 75k Ω | 500 Ω |
| 64 | 0.056 μ F | 3.3 μ F | 68k Ω | 510 Ω |
| 125 | 0.033 μ F | 1.5 μ F | 62k Ω | 510 Ω |
| 250 | 0.015 μ F | 0.82 μ F | 68k Ω | 470 Ω |
| 500 | 8200pF | 0.39 μ F | 62k Ω | 470 Ω |
| 1k | 3900pF | 0.22 μ F | 68k Ω | 470 Ω |
| 2k | 2000pF | 0.1 μ F | 68k Ω | 470 Ω |
| 4k | 1100pF | 0.056 μ F | 62k Ω | 470 Ω |
| 8k | 510pF | 0.022 μ F | 68k Ω | 510 Ω |
| 16k | 330pF | 0.012 μ F | 51k Ω | 510 Ω |

(1) At volume of change = ± 12 dB $Q = 1.7$

REVISION HISTORY

| Rev | Date | Description |
|-----|----------|---|
| 1.0 | 08/16/06 | Initial release. |
| 1.1 | 08/22/06 | Updated the Instantaneous Short Circuit Current specification. |
| 1.2 | 09/12/06 | Updated the three $\pm 15\text{V}$ CMRR Typical Performance Curves. |
| 1.3 | 09/26/06 | Updated interstage filter capacitor values on page 1 Typical Application schematic. |
| 1.4 | 05/03/07 | Added the “general note” under the EC table. |
| 1.5 | 10/17/07 | Replaced all the PSRR curves. |
| 1.6 | 01/26/10 | Edited the equations on page 28 (under Tone Control). |
| J | 04/04/13 | Changed layout of National Data Sheet to TI format |

PACKAGING INFORMATION

| Orderable Device | Status (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan (2) | Lead/Ball Finish | MSL Peak Temp (3) | Op Temp (°C) | Device Marking (4/5) | Samples |
|------------------|---------------|--------------|--------------------|------|----------------|----------------------------|------------------|----------------------|--------------|-------------------------|-------------------------|
| LM4562HA/NOPB | ACTIVE | TO-99 | LMC | 8 | 20 | Green (RoHS & no Sb/Br) | POST-PLATE | Level-1-NA-UNLIM | -40 to 85 | LM4562HA | Samples |
| LM4562MA/NOPB | ACTIVE | SOIC | D | 8 | 95 | Green (RoHS & no Sb/Br) | CU SN | Level-1-260C-UNLIM | -40 to 85 | L4562 MA | Samples |
| LM4562MAX/NOPB | ACTIVE | SOIC | D | 8 | 2500 | Green (RoHS & no Sb/Br) | CU SN | Level-1-260C-UNLIM | -40 to 85 | L4562 MA | Samples |
| LM4562NA/NOPB | ACTIVE | PDIP | P | 8 | 40 | Green (RoHS & no Sb/Br) | CU SN | Level-1-NA-UNLIM | -40 to 85 | LM 4562NA | Samples |

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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TAPE AND REEL INFORMATION


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|----------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| LM4562MAX/NOPB | SOIC | D | 8 | 2500 | 330.0 | 12.4 | 6.5 | 5.4 | 2.0 | 8.0 | 12.0 | Q1 |

TAPE AND REEL BOX DIMENSIONS

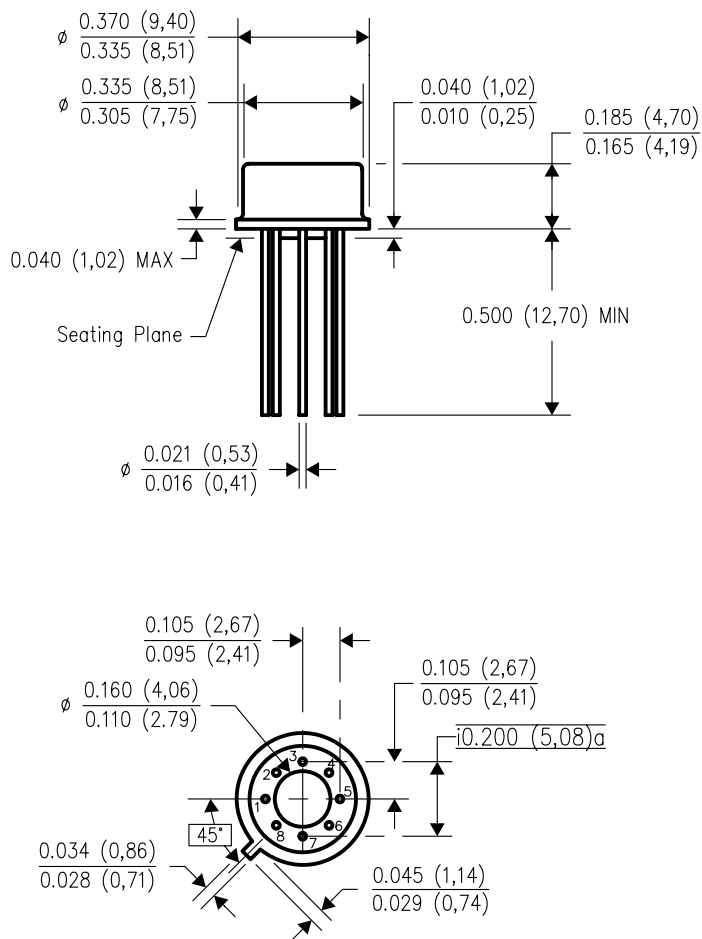


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|----------------|--------------|-----------------|------|------|-------------|------------|-------------|
| LM4562MAX/NOPB | SOIC | D | 8 | 2500 | 367.0 | 367.0 | 35.0 |

LMC (O-MBCY-W8)

METAL CYLINDRICAL PACKAGE



4202483/B 09/07

- NOTES:
- All linear dimensions are in inches (millimeters).
 - This drawing is subject to change without notice.
 - Leads in true position within 0.010 (0,25) R @ MMC at seating plane.
 - Pin numbers shown for reference only. Numbers may not be marked on package.
 - Falls within JEDEC MO-002/T0-99.

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 variation BA.

D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- $\triangle C$ Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- $\triangle D$ Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.

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