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1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32F101xC, STM32F101xD and STM32F101xE high-densityaccess line microcontrollers. For more details on the whole STMicroelectronics STM32F101xx family, please refer to Section 2.2: Full compatibility throughout the family.

The high-density STM32F101xx datasheet should be read in conjunction with the STM32F10xxx reference manual.

For information on programming, erasing and protection of the internal Flash memory please refer to the *STM32F10xxx Flash programming manual*.

The reference and Flash programming manuals are both available from the STMicroelectronics website *www.st.com*.

For information on the Cortex[®]-M3 core please refer to the Cortex[®]-M3 Technical Reference Manual, available from the www.arm.com website.





2 Description

The STM32F101xC, STM32F101xD and STM32F101xE access line family incorporates the high-performance ARM® Cortex®-M3 32-bit RISC core operating at a 36 MHz frequency, high-speed embedded memories (Flash memory up to 512 Kbytes and SRAM up to 48 Kbytes), and an extensive range of enhanced I/Os and peripherals connected to two APB buses. All devices offer one 12-bit ADC, four general-purpose 16-bit timers, as well as standard and advanced communication interfaces: up to two I²Cs, three SPIs and five USARTs.

The STM32F101xx high-density access line family operates in the –40 to +85 °C temperature range, from a 2.0 to 3.6 V power supply. A comprehensive set of power-saving mode allows the design of low-power applications.

These features make the STM32F101xx high-density access line microcontroller family suitable for a wide range of applications such as medical and handheld equipment, PC peripherals and gaming, GPS platforms, industrial applications, PLC, printers, scanners alarm systems and video intercom.



2.1 Device overview

The STM32F101xx high-density access line family offers devices in 3 different package types: from 64 pins to 144 pins. Depending on the device chosen, different sets of peripherals are included, the description below gives an overview of the complete range of peripherals proposed in this family.

Figure 1 shows the general block diagram of the device family.

Table 2. STM32F101xC, STM32F101xD and STM32F101xE features and peripheral counts

Perip	STM32F101Rx			STM32F101Vx			STM32F101Zx				
Flash memo	256 384 512		256	384 512		256	384	512			
SRAM in K	bytes	32	4	8	32	4	8	32	4	8	
FSMC			No			Yes ⁽¹⁾			Yes		
Timers	General- purpose		4								
	Basic					2					
	SPI		3								
Comm	I ² C	2									
	USART	5									
GPIOs	•	51			80			112			
12-bit ADC		Yes			Yes			Yes			
Number of	channels		16		16			16			
12-bit DAC		1									
Number of	channels	2									
CPU freque	ency	36 MHz									
Operating v	2.0 to 3.6 V										
Operating to	Ambient temperature: -40 to +85 °C (see <i>Table 10</i>) Junction temperature: -40 to +105 °C (see <i>Table 10</i>)										
Package		LQFP64			LQFP100			LQFP144			

For the LQFP100 package, only FSMC Bank1 and Bank2 are available. Bank1 can only support a
multiplexed NOR/PSRAM memory using the NE1 Chip Select. Bank2 can only support a 16- or 8-bit
NAND Flash memory using the NCE2 Chip Select. The interrupt line cannot be used since Port G is not
available in this package.



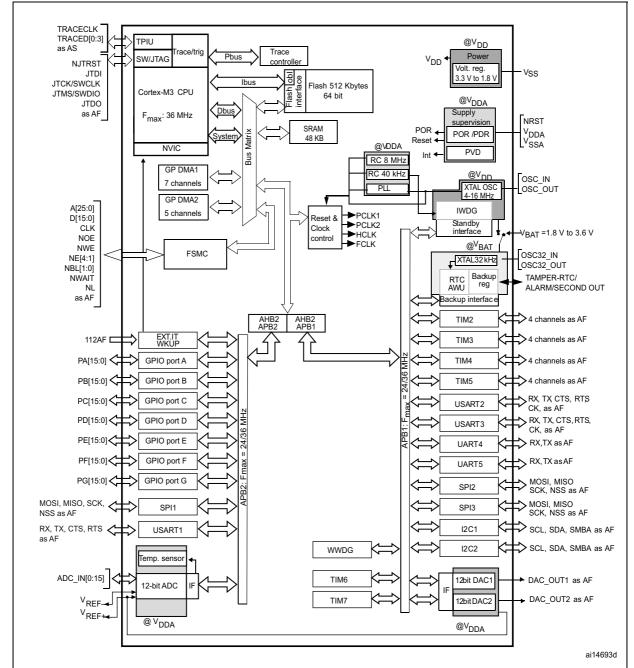


Figure 1. STM32F101xC, STM32F101xD and STM32F101xE access line block diagram

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^{1.} $T_A = -40$ °C to +85 °C (junction temperature up to 105 °C).

^{2.} AF = alternate function on I/O port pin.

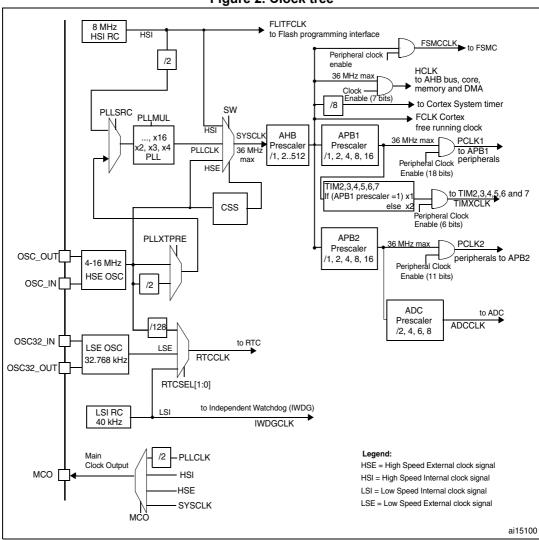


Figure 2. Clock tree

- When the HSI is used as a PLL clock input, the maximum system clock frequency that can be achieved is 36 MHz.
- 2. To have an ADC conversion time of 1 μ s, APB2 must be at 14 MHz or 28 MHz.

2.2 Full compatibility throughout the family

The STM32F101xx is a complete family whose members are fully pin-to-pin, software and feature compatible. In the reference manual, the STM32F101x4 and STM32F101x6 are identified as low-density devices, the STM32F101x8 and STM32F101xB are referred to as medium-density devices, and the STM32F101xC, STM32F101xD and STM32F101xE are referred to as high-density devices .

Low- and high-density devices are an extension of the STM32F101x8/B medium-density devices, they are specified in the STM32F101x4/6 and STM32F101xC/D/E datasheets, respectively.

Low-density devices feature lower Flash memory and RAM capacities, less timers and peripherals. High-density devices have higher Flash memory and RAM densities, and additional peripherals like FSMC and DAC, while remaining fully compatible with the other members of the family.

The STM32F101x4, STM32F101x6, STM32F101xC, STM32F101xD and STM32F101xE are a drop-in replacement for the STM32F101x8/B devices, allowing the user to try different memory densities and providing a greater degree of freedom during the development cycle.

Moreover, the STM32F101xx access line family is fully compatible with all existing STM32F103xx performance line and STM32F102xx USB access line devices.

	Memory size											
	Low-densi	ity devices	Medium-der	sity devices	High-density devices							
Pinout	16 KB 32 KB Flash Flash ⁽¹⁾		64 KB Flash	128 KB Flash	256 KB Flash	384 KB Flash	512 KB Flash					
	4 KB RAM	6 KB RAM	10 KB RAM	16 KB RAM	32 KB RAM	48 KB RAM	48 KB RAM					
144					5 × USARTs 4 × 16-bit timers. 2 × basic timers 3 × SPIs, 2 × I ² Cs, 1 × ADC. 2 ×							
100			3 × USARTs									
64	2 × USARTs	_	3 × 16-bit tim 2 × SPIs, 2 ×		DACs FSMC (100 and 144 pins)							
48	1 × SPI, 1 ×	· I ² C	1 × ADC									
36	1 × ADC				-							

Table 3. STM32F101xx family

2.3 Overview

2.3.1 ARM® Cortex®-M3 core with embedded Flash and SRAM

The ARM® Cortex®-M3 processor is the latest generation of ARM® processors for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while



For orderable part numbers that do not show the A internal code after the temperature range code (6), the
reference datasheet for electrical characteristics is that of the STM32F101x8/B medium-density devices.

delivering outstanding computational performance and an advanced system response to interrupts.

The ARM[®] Cortex[®]-M3 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an ARM[®] core in the memory size usually associated with 8- and 16-bit devices.

The STM32F101xC, STM32F101xD and STM32F101xE access line family having an embedded ARM® core, is therefore compatible with all ARM® tools and software.

Figure 1 shows the general block diagram of the device family.

2.3.2 Embedded Flash memory

256 to 512 Kbytes of embedded Flash are available for storing programs and data.

2.3.3 CRC (cyclic redundancy check) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a fixed generator polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

2.3.4 Embedded SRAM

Up to 48 Kbytes of embedded SRAM accessed (read/write) at CPU clock speed with 0 wait states.

2.3.5 FSMC (flexible static memory controller)

The FSMC is embedded in the STM32F101xC, STM32F101xD and STM32F101xE access line family. It has four Chip Select outputs supporting the following modes: PC Card/Compact Flash, SRAM, PSRAM, NOR and NAND.

Functionality overview:

- The three FSMC interrupt lines are ORed in order to be connected to the NVIC
- Write FIFC
- Code execution from external memory except for NAND Flash and PC Card
- The targeted frequency is HCLK/2, so external access is at 18 MHz when HCLK is at 36 MHz

2.3.6 LCD parallel interface

The FSMC can be configured to interface seamlessly with most graphic LCD controllers. It supports the Intel 8080 and Motorola 6800 modes, and is flexible enough to adapt to specific LCD interfaces. This LCD parallel interface capability makes it easy to build cost-effective graphic applications using LCD modules with embedded controllers or high-performance solutions using external controllers with dedicated acceleration.



2.3.7 Nested vectored interrupt controller (NVIC)

The STM32F101xC, STM32F101xD and STM32F101xE access line embeds a nested vectored interrupt controller able to handle up to 60 maskable interrupt channels (not including the 16 interrupt lines of Cortex®-M3) and 16 priority levels.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface
- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail-chaining
- · Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimal interrupt latency.

2.3.8 External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 19 edge detector lines used to generate interrupt/event requests. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 112 GPIOs can be connected to the 16 external interrupt lines.

2.3.9 Clocks and startup

System clock selection is performed on startup, however the internal RC 8 MHz oscillator is selected as default CPU clock on reset. An external 4-16 MHz clock can be selected, in which case it is monitored for failure. If failure is detected, the system automatically switches back to the internal RC oscillator. A software interrupt is generated if enabled. Similarly, full interrupt management of the PLL clock is available when necessary (for example with failure of an indirectly used external oscillator).

Several prescalers are used to configure the AHB frequency, the high-speed APB (APB2) domain and the low-speed APB (APB1) domain. The maximum frequency of the AHB and APB domains is 36 MHz. See *Figure 2* for details on the clock tree.

2.3.10 Boot modes

At startup, boot pins are used to select one of three boot options:

- Boot from user Flash: you have an option to boot from any of two memory banks. By default, boot from Flash memory bank 1 is selected. You can choose to boot from Flash memory bank 2 by setting a bit in the option bytes.
- Boot from system memory
- Boot from embedded SRAM

The bootloader is located in system memory. It is used to reprogram the Flash memory by using USART1.



2.3.11 Power supply schemes

- V_{DD} = 2.0 to 3.6 V: external power supply for I/Os and the internal regulator. Provided externally through V_{DD} pins.
- V_{SSA}, V_{DDA} = 2.0 to 3.6 V: external analog power supplies for ADC, DAC, Reset blocks, RCs and PLL (minimum voltage to be applied to V_{DDA} is 2.4 V when the ADC or DAC is used). V_{DDA} and V_{SSA} must be connected to V_{DD} and V_{SS}, respectively.
- V_{BAT} = 1.8 to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.

For more details on how to connect power pins, refer to Figure 9: Power supply scheme.

2.3.12 Power supply supervisor

The device has an integrated power-on reset (POR)/power-down reset (PDR) circuitry. It is always active, and ensures proper operation starting from/down to 2 V. The device remains in reset mode when V_{DD} is below a specified threshold, $V_{POR/PDR}$, without the need for an external reset circuit.

The device features an embedded programmable voltage detector (PVD) that monitors the V_{DD}/V_{DDA} power supply and compares it to the V_{PVD} threshold. An interrupt can be generated when V_{DD}/V_{DDA} drops below the V_{PVD} threshold and/or when V_{DD}/V_{DDA} is higher than the V_{PVD} threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software. Refer to Table 12: Embedded reset and power control block characteristics for the values of $V_{POR/PDR}$ and V_{PVD} .

2.3.13 Voltage regulator

The regulator has three operation modes: main (MR), low power (LPR) and power down.

- MR is used in the nominal regulation mode (Run)
- LPR is used in the Stop modes.
- Power down is used in Standby mode: the regulator output is in high impedance: the kernel circuitry is powered down, inducing zero consumption (but the contents of the registers and SRAM are lost)

This regulator is always enabled after reset. It is disabled in Standby mode.

2.3.14 Low-power modes

The STM32F101xC, STM32F101xD and STM32F101xE access line supports three low-power modes to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

Stop mode

Stop mode achieves the lowest power consumption while retaining the content of SRAM and registers. All clocks in the 1.8 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low-power mode.



The device can be woken up from Stop mode by any of the EXTI line. The EXTI line source can be one of the 16 external lines, the PVD output or the RTC alarm.

Standby mode

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.8 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, SRAM and register contents are lost except for registers in the Backup domain and Standby circuitry.

The device exits Standby mode when an external reset (NRST pin), a IWDG reset, a rising edge on the WKUP pin, or an RTC alarm occurs.

Note:

The RTC, the IWDG, and the corresponding clock sources are not stopped by entering Stop or Standby mode.

2.3.15 DMA

The flexible 12-channel general-purpose DMAs (7 channels for DMA1 and 5 channels for DMA2) are able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers.

The two DMA controllers support circular buffer management, removing the need for user code intervention when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with support for software trigger on each channel. Configuration is made by software and transfer sizes between source and destination are independent.

DMA can be used with the main peripherals: SPI, I²C, USART, general-purpose and basic timers TIMx, DAC and ADC.

2.3.16 RTC (real-time clock) and backup registers

The RTC and the backup registers are supplied through a switch that takes power either on V_{DD} supply when present or through the V_{BAT} pin. The backup registers are forty-two 16-bit registers used to store 84 bytes of user application data when V_{DD} power is not present. They are not reset by a system or power reset, and they are not reset when the device wakes up from the Standby mode.

The real-time clock provides a set of continuously running counters which can be used with suitable software to provide a clock calendar function, and provides an alarm interrupt and a periodic interrupt. It is clocked by a 32.768 kHz external crystal, resonator or oscillator, the internal low-power RC oscillator or the high speed external clock divided by 128. The internal low-speed RC has a typical frequency of 40 kHz. The RTC can be calibrated using an external 512 Hz output to compensate for any natural quartz deviation. The RTC features a 32-bit programmable counter for long term measurement using the Compare register to generate an alarm. A 20-bit prescaler is used for the time base clock and is by default configured to generate a time base of 1 second from a clock at 32.768 kHz.

2.3.17 Timers and watchdogs

The high-density STM32F101xx access line devices include up to four general-purpose timers, two basic timers, two watchdog timers and a SysTick timer.

Table 4 compares the features of the general-purpose and basic timers.

57/

Counter Counter Prescaler **DMA** request Capture/compare Complementary **Timer** resolution factor generation channels outputs type TIM2. Any integer Up, TIM3, 16-bit between 1 4 No down, Yes TIM4, and 65536 up/down TIM5 Any integer TIM6. 16-bit 0 Up between 1 Yes Nο TIM7 and 65536

Table 4. Timer feature comparison

General-purpose timers (TIMx)

There are up to 4 synchronizable general-purpose timers (TIM2, TIM3, TIM4 and TIM5) embedded in the STM32F101xC, STM32F101xD and STM32F101xE access line devices. These timers are based on a 16-bit auto-reload up/down counter, a 16-bit prescaler and feature 4 independent channels each for input capture/output compare, PWM or one-pulse mode output. This gives up to 16 input captures / output compares / PWMs on the largest packages.

The general-purpose timers can work together with the advanced-control timer via the Timer Link feature for synchronization or event chaining. Their counter can be frozen in debug mode. Any of the general-purpose timers can be used to generate PWM outputs. They all have independent DMA request generation.

These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

Basic timers TIM6 and TIM7

These timers are mainly used for DAC trigger generation. They can also be used as a generic 16-bit time base.

Independent watchdog

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 40 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

Window watchdog

The window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.



SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0.
- Programmable clock source

2.3.18 I2C bus

Up to two I²C bus interfaces can operate in multi-master and slave modes. They support standard and fast modes.

They support 7/10-bit addressing mode and 7-bit dual addressing mode (as slave). A hardware CRC generation/verification is embedded.

They can be served by DMA and they support SMBus 2.0/PMBus.

2.3.19 Universal synchronous/asynchronous receiver transmitters (USARTs)

The STM32F101xC, STM32F101xD and STM32F101xE access line embeds three universal synchronous/asynchronous receiver transmitters (USART1, USART2 and USART3) and two universal asynchronous receiver transmitters (UART4 and UART5).

These five interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN Master/Slave capability. The five interfaces are able to communicate at speeds of up to 2.25 Mbit/s.

USART1, USART2 and USART3 also provide hardware management of the CTS and RTS signals, Smart Card mode (ISO 7816 compliant) and SPI-like communication capability. All interfaces can be served by the DMA controller except for UART5.

2.3.20 Serial peripheral interface (SPI)

Up to three SPIs are able to communicate up to 18 Mbits/s in slave and master modes in full-duplex and simplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes.

All SPIs can be served by the DMA controller.

2.3.21 GPIOs (general-purpose inputs/outputs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high current-capable.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

2.3.22 ADC (analog to digital converter)

A 12-bit analog-to-digital converter is embedded into STM32F101xC, STM32F101xD and STM32F101xE access line devices. It has up to 16 external channels, performing conversions in single-shot or scan modes. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC can be served by the DMA controller.

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start trigger and injection trigger, respectively, to allow the application to synchronize A/D conversion and timers.

2.3.23 DAC (digital-to-analog converter)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in inverting configuration.

This dual digital Interface supports the following features:

- two DAC converters: one for each output channel
- 8-bit or 12-bit monotonic output
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- external triggers for conversion
- input voltage reference V_{RFF+}

Seven DAC trigger inputs are used in the STM32F101xC, STM32F101xD and STM32F101xE access line family. The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

2.3.24 Temperature sensor

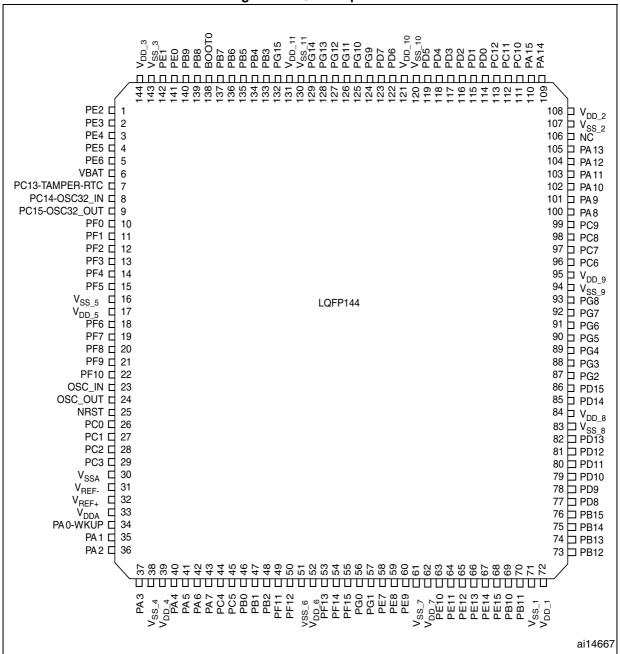
The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 2 V < V_{DDA}

2.3.26 Embedded Trace Macrocell™

The ARM® Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32F10xxx through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. The TPA is connected to a host computer using Ethernet, or any other high-speed channel. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer running debugger software. TPA hardware is commercially available from common development tool vendors. It operates with third party debugger software tools.

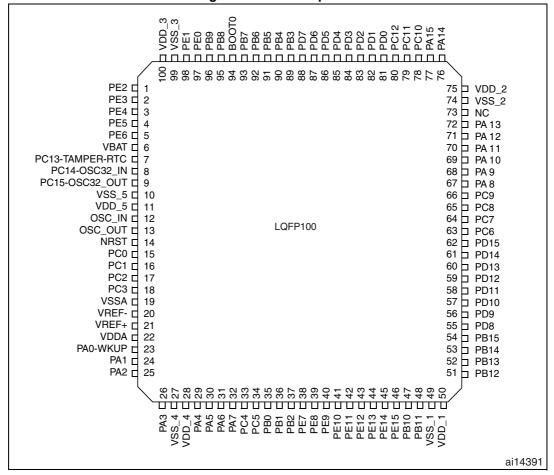
3 Pinouts and pin descriptions

Figure 3. LQFP144 pinout



1. The above figure shows the package top view.

Figure 4. LQFP100 pinout



1. The above figure shows the package top view.



VBAT ☐ 1 PC13-TAMPER-RTC 45 PA 12 44 PA 11 43 PA 10 42 PA 9 PD1-OS C_OUT NRST 41 PA 8 PC0 🗖 8 40 PC9 39 PC8 LQFP64 38 PC7 37 PC6 12 13 Vssa 36 PB 15 VDDA 35 PB 14 34 PB 13 PA 0-WKUP 14 PA 1 15 PA 2 16 17 18 19 20 21 22 23 24 25 26 27 28 29 30 31 32 33 PB 12 PA 3 VSS_4 VDD_4 PA 4 PA 5 PA 6 PA 7 PC 7 PC 9 PB 1 ai14392

Figure 5. LQFP64 pinout

1. The above figure shows the package top view.

Table 5. STM32F101xC/STM32F101xD/STM32F101xE pin definitions

Pins							Alternate functi	ons ⁽⁴⁾
LQFP144	LQFP64	LQFP100	Pin name	Type ⁽¹⁾	I / O Level ⁽²⁾	Main function ⁽³⁾ (after reset)	Default	Remap
1	-	1	PE2	I/O	FT	PE2	TRACECLK/ FSMC_A23	-
2	ı	2	PE3	I/O	FT	PE3	TRACED0/FSMC_A19	-
3	-	3	PE4	I/O	FT	PE4	TRACED1/FSMC_A20	-
4	-	4	PE5	I/O	FT	PE5	TRACED2/FSMC_A21	-
5	-	5	PE6	I/O	FT	PE6	TRACED3/FSMC_A22	-
6	1	6	V_{BAT}	S	-	V_{BAT}	-	-
7	2	7	PC13-TAMPER-RTC ⁽⁵⁾	I/O	-	PC13 ⁽⁶⁾	TAMPER-RTC	-
8	3	8	PC14-OSC32_IN ⁽⁵⁾	I/O	-	PC14 ⁽⁶⁾	OSC32_IN	-
9	4	9	PC15-OSC32_OUT ⁽⁵⁾	I/O	ı	PC15 ⁽⁶⁾	OSC32_OUT	
10	-	1	PF0	I/O	FT	PF0	FSMC_A0	-
11	-	-	PF1	I/O	FT	PF1	FSMC_A1	-



Table 5. STM32F101xC/STM32F101xD/STM32F101xE pin definitions (continued)

Pins							Alternate functions ⁽⁴⁾		
LQFP144	LQFP64	LQFP100	Pin name	Type ⁽¹⁾	I / O Level ⁽²⁾	Main function ⁽³⁾ (after reset)	Default	Remap	
12	-	-	PF2	I/O	FT	PF2	FSMC_A2	-	
13	-	-	PF3	I/O	FT	PF3	FSMC_A3	-	
14	-	-	PF4	I/O	FT	PF4	FSMC_A4	-	
15	-	-	PF5	I/O	FT	PF5	FSMC_A5	-	
16	-	10	V _{SS_5}	S	-	V _{SS_5}	-	-	
17	-	11	V _{DD_5}	S	-	V _{DD_5}	-	-	
18	-	-	PF6	I/O	-	PF6	FSMC_NIORD	-	
19	-	-	PF7	I/O	-	PF7	FSMC_NREG	-	
20	-	-	PF8	I/O	-	PF8	FSMC_NIOWR	-	
21	-	-	PF9	I/O	-	PF9	FSMC_CD	-	
22	-	-	PF10	I/O	-	PF10	FSMC_INTR	-	
23	5	12	OSC_IN	I	-	OSC_IN	-	PD0 ⁽⁷⁾	
24	6	13	OSC_OUT	0	-	OSC_OUT	-	PD1 ⁽⁷⁾	
25	7	14	NRST	I/O	-	NRST	-	-	
26	8	15	PC0	I/O	-	PC0	ADC_IN10	-	
27	9	16	PC1	I/O	-	PC1	ADC_IN11	-	
28	10	17	PC2	I/O	-	PC2	ADC_IN12	-	
29	11	18	PC3	I/O	-	PC3	ADC_IN13	-	
30	12	19	V _{SSA}	S	-	V_{SSA}	-	-	
31	-	20	V _{REF-}	S	-	V _{REF-}	-	-	
32	-	21	V _{REF+}	S	-	V _{REF+}	-	-	
33	13	22	V_{DDA}	S	-	V_{DDA}	-	-	
34	14	23	PA0-WKUP	I/O	-	PA0	WKUP/ USART2_CTS ⁽⁸⁾ / ADC_IN0/TIM5_CH1/ TIM2_CH1_ETR ⁽⁸⁾	-	
35	15	24	PA1	I/O	-	PA1	USART2_RTS ⁽⁸⁾ / ADC_IN1/TIM5_CH2 TIM2_CH2 ⁽⁸⁾	-	
36	16	25	PA2	I/O	-	PA2	USART2_TX ⁽⁸⁾ / TIM5_CH3/ADC_IN2/ TIM2_CH3 ⁽⁸⁾	-	

Table 5. STM32F101xC/STM32F101xD/STM32F101xE pin definitions (continued)

Pins							Alternate functions ⁽⁴⁾		
LQFP144	LQFP64	LQFP100	Pin name	Type ⁽¹⁾	I / O Level ⁽²⁾	Main function ⁽³⁾ (after reset) Default		Remap	
37	17	26	PA3	I/O	-	PA3	USART2_RX ⁽⁸⁾ / TIM5_CH4 / ADC_IN3/ TIM2_CH4 ⁽⁸⁾	-	
38	18	27	V_{SS_4}	S	ı	V_{SS_4}	-	-	
39	19	28	V_{DD_4}	S	ı	V_{DD_4}	-	-	
40	20	29	PA4	I/O	ı	PA4	SPI1_NSS/ DAC_OUT1 ADC_IN4 / USART2_CK ⁽⁸⁾	-	
41	21	30	PA5	I/O	ı	PA5	SPI1_SCK/ DAC_OUT2/ADC_IN5	-	
42	22	31	PA6	I/O	-	PA6	SPI1_MISO / ADC_IN6 / TIM3_CH1 ⁽⁸⁾	-	
43	23	32	PA7	I/O	-	PA7	SPI1_MOSI / ADC_IN7/ TIM3_CH2 ⁽⁸⁾	-	
44	24	33	PC4	I/O	-	PC4	ADC_IN14 -		
45	25	34	PC5	I/O	-	PC5	ADC_IN15 -		
46	26	35	PB0	I/O	-	PB0	ADC_IN8 / TIM3_CH3 ⁽⁸⁾	-	
47	27	36	PB1	I/O	ı	PB1	ADC_IN9/TIM3_CH4 ⁽⁸⁾	-	
48	28	37	PB2	I/O	FT	PB2/BOOT1	-	-	
49	-	1	PF11	I/O	FT	PF11	FSMC_NIOS16	-	
50	-	1	PF12	I/O	FT	PF12	FSMC_A6	-	
51	-	-	V_{SS_6}	S	-	V_{SS_6}	-	-	
52	-	-	V_{DD_6}	S	-	V_{DD_6}	-	-	
53	-	-	PF13	I/O	FT	PF13	FSMC_A7	-	
54	-	-	PF14	I/O	FT	PF14	FSMC_A8	-	
55	-	-	PF15	I/O	FT	PF15	FSMC_A9	-	
56	-	-	PG0	I/O	FT	PG0	FSMC_A10	-	
57	-	-	PG1	I/O	FT	PG1	FSMC_A11	-	
58	-	38	PE7	I/O	FT	PE7	FSMC_D4	-	
59	-	39	PE8	I/O	FT	PE8	FSMC_D5	-	
60	-	40	PE9	I/O	FT	PE9	FSMC_D6	_	
61	-	-	V _{SS_7}	S	-	V _{SS_7}	<u>-</u>		



Table 5. STM32F101xC/STM32F101xD/STM32F101xE pin definitions (continued)

Pins							Alternate functions ⁽⁴⁾		
LQFP144	LQFP64	LQFP100	Pin name	Type ⁽¹⁾	Main function ⁽³⁾ (after reset)		Default	Remap	
62	-	-	V _{DD_7}	S	-	V _{DD_7}	-	-	
63	-	41	PE10	I/O	FT	PE10	FSMC_D7	-	
64	-	42	PE11	I/O	FT	PE11	FSMC_D8	-	
65	-	43	PE12	I/O	FT	PE12	FSMC_D9	-	
66	-	44	PE13	I/O	FT	PE13	FSMC_D10	-	
67	-	45	PE14	I/O	FT	PE14	FSMC_D11	-	
68	-	46	PE15	I/O	FT	PE15	FSMC_D12	-	
69	29	47	PB10	I/O	FT	PB10	I2C2_SCL/ USART3_TX ⁽⁸⁾	TIM2_CH3	
70	30	48	PB11	I/O	FT	PB11	I2C2_SDA/ USART3_RX ⁽⁸⁾	TIM2_CH4	
71	31	49	V _{SS_1}	S	-	V _{SS_1}	-	-	
72	32	50	V _{DD_1}	S	-	V _{DD_1}	-	-	
73	33	51	PB12	I/O	FT	PB12	SPI2_NSS ⁽⁸⁾ / I2C2_SMBA USART3_CK ⁽⁸⁾	-	
74	34	52	PB13	I/O	FT	PB13	SPI2_SCK ⁽⁸⁾ / USART3_CTS ⁽⁸⁾	-	
75	35	53	PB14	I/O	FT	PB14	SPI2_MISO ⁽⁸⁾ / USART3_RTS ⁽⁸⁾	-	
76	36	54	PB15	I/O	FT	PB15	SPI2_MOSI ⁽⁸⁾	-	
77	-	55	PD8	I/O	FT	PD8	FSMC_D13	USART3_TX	
78	-	56	PD9	I/O	FT	PD9	FSMC_D14	USART3_RX	
79	-	57	PD10	I/O	FT	PD10	FSMC_D15	USART3_CK	
80	-	58	PD11	I/O	FT	PD11	FSMC_A16	USART3_CTS	
81	-	59	PD12	I/O	FT	PD12	FSMC_A17	TIM4_CH1 / USART3_RTS	
82	-	60	PD13	I/O	FT	PD13	FSMC_A18 TIM4_CH2		
83	-	-	V _{SS_8}	S	-	V _{SS_8}	-	-	
84	-	-	V _{DD_8}	S	-	V _{DD_8}			
85	-	61	PD14	I/O	FT	PD14	FSMC_D0	TIM4_CH3	
86	-	62	PD15	I/O	FT	PD15	FSMC_D1	TIM4_CH4	
87	-	ı	PG2	I/O	FT	PG2	FSMC_A12	-	

Table 5. STM32F101xC/STM32F101xD/STM32F101xE pin definitions (continued)

Pins							Alternate functions ⁽⁴⁾		
LQFP144	LQFP64	LQFP100	Pin name	Type ⁽¹⁾	I / O Level ⁽²⁾	Main function ⁽³⁾ (after reset)	Default	Remap	
88	-	-	PG3	I/O	FT	PG3	FSMC_A13 -		
89	-	-	PG4	I/O	FT	PG4	FSMC_A14	-	
90	-	-	PG5	I/O	FT	PG5	FSMC_A15	-	
91	-	-	PG6	I/O	FT	PG6	FSMC_INT2	-	
92	-	1	PG7	I/O	FT	PG7	FSMC_INT3	-	
93	-	-	PG8	I/O	FT	PG8	-	-	
94	-	1	V _{SS_9}	S	-	V _{SS_9}	-	-	
95	-	1	V _{DD_9}	S	-	V _{DD_9}	-	-	
96	37	63	PC6	I/O	FT	PC6	-	TIM3_CH1	
97	38	64	PC7	I/O	FT	PC7	-	TIM3_CH2	
98	39	65	PC8	I/O	FT	PC8	-	TIM3_CH3	
99	40	66	PC9	I/O	FT	PC9	-	TIM3_CH4	
100	41	67	PA8	I/O	FT	PA8	USART1_CK/ MCO	-	
101	42	68	PA9	I/O	FT	PA9	USART1_TX ⁽⁸⁾	-	
102	43	69	PA10	I/O	FT	PA10	USART1_RX ⁽⁸⁾	-	
103	44	70	PA11	I/O	FT	PA11	USART1_CTS	-	
104	45	71	PA12	I/O	FT	PA12	USART1_RTS	-	
105	46	72	PA13	I/O	FT	JTMS-SWDIO	-	PA13	
106	-	73				Not conne	ected		
107	47	74	V _{SS_2}	S	-	V _{SS_2}	-	-	
108	48	75	V _{DD_2}	S	-	V _{DD_2}	-	-	
109	49	76	PA14	I/O	FT	JTCK-SWCLK	-	PA14	
110	50	77	PA15	I/O	FT	JTDI	SPI3_NSS	TIM2_CH1_ETR/ PA15 /SPI1_NSS	
111	51	78	PC10	I/O	FT	PC10	UART4_TX	USART3_TX	
112	52	79	PC11	I/O	FT	PC11	UART4_RX	USART3_RX	
113	53	80	PC12	I/O	FT	PC12	UART5_TX	USART3_CK	
114	-	81	PD0	I/O	FT	OSC_IN ⁽⁸⁾	FSMC_D2 ⁽⁹⁾	-	
115	-	82	PD1	I/O	FT	OSC_OUT ⁽⁸⁾	FSMC_D3 ⁽⁹⁾	_	



Table 5. STM32F101xC/STM32F101xD/STM32F101xE pin definitions (continued)

Pins							Alternate functions ⁽⁴⁾		
LQFP144	LQFP64	LQFP100	Pin name	Type ⁽¹⁾	I / O Level ⁽²⁾	Main function ⁽³⁾ (after reset)	Default	Remap	
116	54	83	PD2	I/O	FT	PD2	TIM3_ETR/UART5_RX	-	
117	-	84	PD3	I/O	FT	PD3	FSMC_CLK	USART2_CTS	
118	-	85	PD4	I/O	FT	PD4	FSMC_NOE	USART2_RTS	
119	-	86	PD5	I/O	FT	PD5	FSMC_NWE	USART2_TX	
120	-	1	V _{SS_10}	S	1	V _{SS_10}	-	-	
121	-	-	V _{DD_10}	S	1	V _{DD_10}	-	-	
122	-	87	PD6	I/O	FT	PD6	FSMC_NWAIT	USART2_RX	
123		88	PD7	I/O	FT	PD7	FSMC_NE1/ FSMC_NCE2	USART2_CK	
124	-	ı	PG9	I/O	FT	PG9	FSMC_NE2/ FSMC_NCE3	-	
125	-	-	PG10	I/O	FT	PG10	FSMC_NE3/ FSMC_NCE4_1	-	
126	-	-	PG11	I/O	FT	PG11	FSMC_NCE4_2	-	
127	-	-	PG12	I/O	FT	PG12	FSMC_NE4	-	
128	-	-	PG13	I/O	FT	PG13	FSMC_A24	-	
129	-	-	PG14	I/O	FT	PG14	FSMC_A25	-	
130	-	-	V _{SS_11}	S		V _{SS_11}	-	-	
131	1	1	V _{DD_11}	S	•	V _{DD_11}	-	-	
132	-	-	PG15	I/O	FT	PG15	-	-	
133	55	89	PB3	I/O	FT	JTDO	SPI3_SCK	TIM2_CH2 /PB3 TRACESWO SPI1_SCK	
134	56	90	PB4	I/O	FT	NJTRST	SPI3_MISO	PB4 / TIM3_CH1 SPI1_MISO	
135	57	91	PB5	I/O		PB5	I2C1_SMBA/ SPI3_MOSI	TIM3_CH2 / SPI1_MOSI	
136	58	92	PB6	I/O	FT	PB6	I2C1_SCL/ TIM4_CH1 ⁽⁸⁾ USART1_		
137	59	93	PB7	I/O	FT	PB7	I2C1_SDA/FSMC_NADV TIM4_CH2 ⁽⁸⁾	USART1_RX	
138	60	94	воото	Ι	i	BOOT0	-	-	
139	61	95	PB8	I/O	FT	PB8	TIM4_CH3 ⁽⁸⁾	I2C1_SCL	

Alternate functions(4) **Pins** O Level⁽²⁾ Main -QFP144 function⁽³⁾ LQFP100 LQFP64 Pin name Default (after reset) Remap 96 TIM4 CH4 (8) 140 62 PB9 I/O FT PB9 I2C1 SDA TIM4 ETR(8)/ PE0 141 97 I/C FT PE0 FSMC NBL0 FSMC_NBL1 142 98 PE1 I/O FT PE1 _ 143 63 99 S _ V_{SS 3} V_{SS 3} 144 100 S 64 $V_{DD 3}$ _ V_{DD_3} _ _

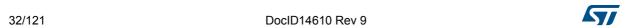
Table 5. STM32F101xC/STM32F101xD/STM32F101xE pin definitions (continued)

- 1. I = input, O = output, S = supply.
- 2 FT = 5 V tolerant
- 3. Function availability depends on the chosen device.
- 4. If several peripherals share the same I/O pin, to avoid conflict between these alternate functions only one peripheral should be enabled at a time through the peripheral clock enable bit (in the corresponding RCC peripheral clock enable register).
- PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 in output mode is limited: the speed should not exceed 2 MHz with a maximum load of 30 pF and these IOs must not be used as a current source (e.g. to drive an LED).
- 6. Main function after the first backup domain power-up. Later on, it depends on the contents of the Backup registers even after reset (because these registers are not reset by the main reset). For details on how to manage these IOs, refer to the Battery backup domain and BKP register description sections in the STM32F10xxx reference manual, available from the STMicroelectronics website: www.st.com.
- 7. For the LQFP64 package, the pins number 5 and 6 are configured as OSC_IN/OSC_OUT after reset, however the functionality of PD0 and PD1 can be remapped by software on these pins. For the LQFP100 and LQFP144 packages, PD0 and PD1 are available by default, so there is no need for remapping. For more details, refer to Alternate function I/O and debug configuration section in the STM32F10xxx reference manual
- This alternate function can be remapped by software to some other port pins (if available on the used package). For more
 details, refer to the Alternate function I/O and debug configuration section in the STM32F10xxx reference manual, available
 from the STMicroelectronics website: www.st.com.
- 9. For devices delivered in LQFP64 packages, the FSMC function is not available.



Table 6. FSMC pin definition

	FSMC											
Pins	CF	CF/IDE	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND 16 bit	LQFP100 ⁽¹⁾						
PE2	-	-	A23	A23	-	Yes						
PE3	-	-	A19	A19	-	Yes						
PE4	-	-	A20	A20	_	Yes						
PE5	-	-	A21	A21	-	Yes						
PE6	-	-	A22	A22	_	Yes						
PF0	A0	A0	A0	-	-	-						
PF1	A1	A1	A1	-	-	-						
PF2	A2	A2	A2	-	-	-						
PF3	A3	-	А3	-	-	-						
PF4	A4	-	A4	-	-	-						
PF5	A5	-	A5	-	-	-						
PF6	NIORD	NIORD	-	-	-	-						
PF7	NREG	NREG	-	-	-	-						
PF8	NIOWR	NIOWR	-	-	-	-						
PF9	CD	CD	-	-	-	-						
PF10	INTR	INTR	-	-	-	-						
PF11	NIOS16	NIOS16	-	-	-	-						
PF12	A6	-	A6	-	-	-						
PF13	A7	-	A7	-	-	-						
PF14	A8	-	A8	-	-	-						
PF15	A9	-	A9	-	-	-						
PG0	A10	-	A10	-	-	-						
PG1	-	-	A11	-	-	-						
PE7	D4	D4	D4	DA4	D4	Yes						
PE8	D5	D5	D5	DA5	D5	Yes						
PE9	D6	D6	D6	DA6	D6	Yes						
PE10	D7	D7	D7	DA7	D7	Yes						
PE11	D8	D8	D8	DA8	D8	Yes						
PE12	D9	D9	D9	DA9	D9	Yes						
PE13	D10	D10	D10	DA10	D10	Yes						
PE14	D11	D11	D11	DA11	D11	Yes						
PE15	D12	D12	D12	DA12	D12	Yes						
PD8	D13	D13	D13	DA13	D13	Yes						





4 Memory mapping

The memory map is shown in *Figure 6*.

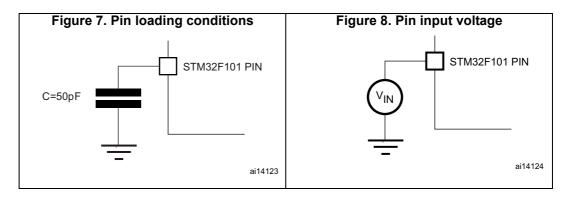


0xA000 1000 - 0xBFFF FFFF Reserved FSMC register 0xA000 0000 - 0xA000 0FFF FSMC bank 4 PCCARD 0x9000 0000 - 0x9FFF FFFF FSMC bank 3 NAND (NAND2) 0x8000 0000 - 0x8FFF FFFF FSMC bank 2 NAND (NAND1) 0x7000 0000 - 0x7FFF FFFF FSMC bank 1 NOR/PSRAM 4 0x6C00 0000 - 0x6FFF FFFF FSMC bank 1 NOR/PSRAM 3 0x6800 0000 - 0x6BFF FFFF FSMC bank 1 NOR/PSRAM 2 0x6400 0000 - 0x67FF FFFF FSMC bank 1 NOR/PSRAM 1 0x6000 0000 - 0x63FF FFFF Reserved 0x4002 3400 - 0x5FFF FFFF CRC 0x4002 3000 - 0x4002 33FF 0x4002 2400 - 0x4002 2FFF Flash interface 0x4002 2000 - 0x4002 23FF Reserved 0x4002 1400 - 0x4002 1FFF RCC 0x4002 1000 - 0x4002 13FF Reserved 0x4002 0400 - 0x4002 0FFF DMA2 0x4002 0400 - 0x4002 07FF DMA1 0x4002 0000 - 0x4002 03FF 0x4001 8400 - 0x4001 FFFF 0x4001 8000 - 0x4001 83FF Reserved Reserved 0xFFFF FFFF 0x4001 3C00 - 0x4001 7FFF 512-Mbyte Reserved 0x4001 3800 - 0x4001 3BFF block 7 USART1 0x4001 3400 - 0x4001 37FF Cortex-M3's 0x4001 3000 - 0x4001 33FF SPI1 internal 0x4001 2800 - 0x4001 2FFF 0xE000 0000 Reserved peripherals ADC1 0x4001 2400 - 0x4001 27FF 0xDFFF FFFF 0x4001 2000 - 0x4001 23FF Port G 512-Mbyte 0x4001 1C00 - 0x4001 1FFF block 6 0x4001 1800 - 0x4001 1BFF Port E Not used 0x4001 1400 - 0x4001 17FF Port D 0xC000 0000 0xBFFF FFFF 0x4001 1000 - 0x4001 13FF Port C 0x4001 0C00 - 0x4001 0FFF Port B 512-Mbyte 0x4001 0800 - 0x4001 0BFF 0x4001 0400 - 0x4001 07FF block 5 EXTI 0x4001 0000 - 0x4001 03FF FSMC register AFIO 0x4000 7800 - 0x4000 FFFF Reserved 0xA000 0000 0x9FFF FFFF 0x4000 7400 - 0x4000 77FF DAC PWR 0x4000 7000 - 0x4000 73FF 512-Mbyte BKP 0x4000 6C00 - 0x4000 6FFF block 4 0x4000 5C00 - 0x4000 6BFF FSMC bank3 0x4000 5800 - 0x4000 5BFF & bank4 I2C2 0x8000 0000 0x7FFF FFFF I2C1 0x4000 5400 - 0x4000 57FF UART5 0x4000 5000 - 0x4000 53FF 512-Mbyte block 3 0x4000 4C00 - 0x4000 4FFF UART4 FSMC bank1 0x4000 4800 - 0x4000 4BFF USART3 & bank2 USART2 0x4000 4400 - 0x4000 47FF 0x6000 0000 0x5FFF FFFF Reserved 0x4000 4000 - 0x4000 43FF 512-Mbyte 0x4000 3C00 - 0x4000 3FFF SPI2 0x4000 3800 - 0x4000 3BFF block 2 Peripherals 0x4000 3400 - 0x4000 37FF IWDG 0x4000 3000 - 0x4000 33FF 0x4000 0000 0x3FFF FFFF WWDG 0x4000 2C00 - 0x4000 2FFF RTC 0x4000 2800 - 0x4000 2BFF 512-Mbyte 0x4000 1800 - 0x4000 27FF block 1 Reserved SRAM TIMZ 0x4000 1400 - 0x4000 17FF TIM6 0x4000 1000 - 0x4000 13FF 0x2000 0000 0x1FFF FFFF 0x4000 0C00 - 0x4000 0FFF TIM5 512-Mbyte 0x4000 0800 - 0x4000 0BFF TIM4 block 0 0x4000 0400 - 0x4000 07FF TIM3 Code 0x4000 0000 - 0x4000 03FF TIM2 0x0000 0000 0x3FFF FFI Reserved 0x2000 C000 0x2000 BFFF SRAM (48 KB aliased 0x2000 0000 0x1FFF F800 - 0x1FFF F80F Option Bytes 0x1FFF F000- 0x1FFF F7FF stem memory 0x1FFF EFFF Reserved 0x0808 0000 0x0807 FFFF Flash 0x0800 0000 0x07FF FFFF Reserved 0x0008 0000 0x0007 FFFF Aliased to Flash or system memory depending on 0x0000 0000 **BOOT** pins ai14811c

Figure 6. Memory map

5.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 8.



5.1.6 Power supply scheme

Figure 9. Power supply scheme $V_{B\underline{A}\underline{T}}$ Backup circuitry Power switch 1.8-3.6V (OSC32K,RTC, Wake-up logic Backup registers) shifter Ю GP I/Os Logic Kernel logic (CPU, Digital & Memories) VDD1/2/.../11 Regulator 11 × 100 nF $+ 1 \times 4.7 \mu F$ Vss1/2/.../11 V_{DDA} V_{REF+} ADC/ Analog: 10 nF V_{REF}-DAC RCs, PLI + 1 µF V_{SSA} ai15401

Caution: In *Figure 9*, the 4.7 μ F capacitor must be connected to V_{DD3} .

5.1.7 Current consumption measurement

IDD_VBAT VBAT VDD VDD VDDA

Figure 10. Current consumption measurement scheme

5.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 7: Voltage characteristics*, *Table 8: Current characteristics*, and *Table 9: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Symbol	Ratings	Min	Max	Unit
V _{DD} – V _{SS}	External main supply voltage (including V_{DDA} and V_{DD}) ⁽¹⁾	-0.3	4.0	.,
V _{IN} ⁽²⁾	Input voltage on five volt tolerant pin	V _{SS} – 0.3	V _{DD} + 4.0	V
V _{IN} (-)	Input voltage on any other pin	V _{SS} - 0.3	4.0	
$ \Delta V_{DDx} $	Variations between different V _{DD} power pins	-	50	
V _{SSX} - V _{SS}	Variations between all the different ground pins	-	50	mV
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)		3.12: Absolute ings (electrical itivity)	

Table 7. Voltage characteristics

All main power (V_{DD}, V_{DDA}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.

^{2.} V_{IN} maximum must always be respected. Refer to *Table 8: Current characteristics* for the maximum allowed injected current values.

Table 8. Current characteristics

Symbol	Ratings	Max.	Unit
ΣI_{VDD}	Total current into V _{DD} /V _{DDA} power lines (source) ⁽¹⁾	150	
Σl _{VSS}	Total current out of V _{SS} ground lines (sink) ⁽¹⁾	150	
	Output current sunk by any I/O and control pin	25	
I _{IO}	Output current source by any I/Os and control pin	- 25	mA
, (2)	Injected current on five volt tolerant pins ⁽³⁾	-5/+0	
I _{INJ(PIN)} ⁽²⁾	Injected current on any other pin ⁽⁴⁾	± 5	
Σl _{INJ(PIN)}	Total injected current (sum of all I/O and control pins) ⁽⁵⁾	± 25	

- All main power (V_{DD}, V_{DDA}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
- 2. Negative injection disturbs the analog performance of the device. See note 3 below Table 58 on page 98.
- Positive injection is not possible on these I/Os. A negative injection is induced by V_{IN}<V_{SS}. I_{INJ(PIN)} must never be exceeded. Refer to *Table 7: Voltage characteristics* for the maximum allowed input voltage values.
- 4. A positive injection is induced by V_{IN}>V_{DD} while a negative injection is induced by V_{IN}<V_{SS}. I_{INJ(PIN)} must never be exceeded. Refer to *Table 7: Voltage characteristics* for the maximum allowed input voltage values.
- When several inputs are submitted to a current injection, the maximum ΣI_{INJ(PIN)} is the absolute sum of the
 positive and negative injected currents (instantaneous values).

Table 9. Thermal characteristics

Symbol	Ratings	Value	Unit
T _{STG}	Storage temperature range	-65 to +150	°C
T _J	Maximum junction temperature	150	°C



5.3 Operating conditions

5.3.1 General operating conditions

Table 10. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f _{HCLK}	Internal AHB clock frequency	-	0	36	
f _{PCLK1}	Internal APB1 clock frequency	-	0	36	MHz
f _{PCLK2}	Internal APB2 clock frequency	-	0	36	
V _{DD}	Standard operating voltage	-	2	3.6	V
V (1)	Analog operating voltage (ADC not used)	Must be the same potential	2	3.6	V
V _{DDA} ⁽¹⁾	Analog operating voltage (ADC used)	as V _{DD} ⁽²⁾	2.4	3.6	v
V_{BAT}	Backup operating voltage	-	1.8	3.6	V
		LQFP144	-	666	
P_{D}	Power dissipation at T _A = 85 °C ⁽³⁾	LQFP100	- 434 m\		mW
		LQFP64	-	444	
TA	Ambient temperature	Maximum power dissipation	-40	85	°C
IA	Ambient temperature	low-power dissipation ⁽⁴⁾	-4 0	105	°C
TJ	Junction temperature range	-	-40	105	°C

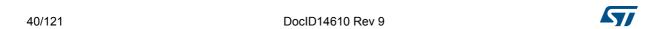
^{1.} When the ADC is used, refer to Table 55: ADC characteristics.

5.3.2 Operating conditions at power-up / power-down

The parameters given in *Table 11* are derived from tests performed under the ambient temperature condition summarized in *Table 10*.

Table 11. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit
+	V _{DD} rise time rate		0	∞	us/V
t∨DD	V _{DD} fall time rate	-	20	∞	μ5/ ν



^{2.} It is recommended to power V_{DD} and V_{DDA} from the same source. A maximum difference of 300 mV between V_{DD} and V_{DDA} can be tolerated during power-up and operation.

If T_A is lower, higher P_D values are allowed as long as T_J does not exceed T_Jmax (see Section 6.4: Thermal characteristics).

^{4.} In low-power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_Jmax (see *Section 6.4: Thermal characteristics*).

5.3.3 Embedded reset and power control block characteristics

The parameters given in *Table 12* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

Table 12. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		PLS[2:0]=000 (rising edge)	2.1	2.18	2.26	V
		PLS[2:0]=000 (falling edge)	2	2.08	2.16	V
		PLS[2:0]=001 (rising edge)	2.19	2.28	2.37	V
		PLS[2:0]=001 (falling edge)	2.09	2.18	2.27	V
		PLS[2:0]=010 (rising edge)	2.28	2.38	2.48	V
		PLS[2:0]=010 (falling edge)	2.18	2.28	2.38	V
		PLS[2:0]=011 (rising edge)	2.38	2.48	2.58	V
V	Programmable voltage detector level selection	PLS[2:0]=011 (falling edge)	2.28	2.38	2.48	V
V_{PVD}		PLS[2:0]=100 (rising edge)	2.47	2.58	2.69	V
		PLS[2:0]=100 (falling edge)	2.37	2.48	2.59	V
		PLS[2:0]=101 (rising edge)	2.57	2.68	2.79	V
		PLS[2:0]=101 (falling edge)	2.47	2.58	2.69	V
		PLS[2:0]=110 (rising edge)	2.66	2.78	2.9	V
		PLS[2:0]=110 (falling edge)	2.56	2.68	2.8	V
		PLS[2:0]=111 (rising edge)	2.76	2.88	3	V
		PLS[2:0]=111 (falling edge)	2.66	2.78	2.9	V
V _{PVDhyst} ⁽²⁾	PVD hysteresis	-	-	100	-	mV
	Power on/power down	Falling edge	1.8 ⁽¹⁾	1.88	1.96	V
V _{POR/PDR}	reset threshold	Rising edge	1.84	1.92	2.0	V
V _{PDRhyst} ⁽²⁾	PDR hysteresis	-	-	40	-	mV
t _{RSTTEMPO} ⁽²⁾	Reset temporization	-	1.5	2.5	3.5	ms

^{1.} The product behavior is guaranteed by design down to the minimum $V_{\mbox{POR}/\mbox{PDR}}$ value.

^{2.} Guaranteed by design, not tested in production.

5.3.4 Embedded reference voltage

The parameters given in *Table 13* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{REFINT}	Internal reference voltage	-40 °C < T _A < +85 °C	1.16	1.20	1.24	V
T _{S_vrefint} (1)	ADC sampling time when reading the internal reference voltage	-	-	5.1	17.1 ⁽²⁾	μs
V _{RERINT} ⁽²⁾	Internal reference voltage spread over the temperature range	V _{DD} = 3 V ±10 mV	-	-	10	mV
T _{Coeff} ⁽²⁾	Temperature coefficient	-	-	-	100	ppm/ °C

Table 13. Embedded internal reference voltage

5.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in *Figure 10: Current consumption measurement scheme*.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to Dhrystone 2.1 code.

Maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in input mode with a static value at V_{DD} or V_{SS} (no load)
- All peripherals are disabled except if it is explicitly mentioned
- The Flash access time is adjusted to f_{HCLK} frequency (0 wait state from 0 to 24 MHz, 1 wait state from 24 to 36 MHz)
- Prefetch in on (reminder: this bit must be set before clock setting and bus prescaling)
- When the peripherals are enabled f_{PCLK1} = f_{HCLK/2}, f_{PCLK2} = f_{HCLK}

The parameters given in *Table 14* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

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^{1.} Shortest sampling time can be determined in the application by multiple iterations.

^{2.} Guaranteed by design, not tested in production.

Table 14. Maximum current consumption in Run mode, code with data processing running from Flash

Symbol	Parameter	Conditions		Max ⁽¹⁾	Unit
Symbol	Parameter		fhcLK	T _A = 85 °C	Uiiii
			36 MHz	39	
	Supply current in Run mode	External clock ⁽²⁾ , all peripherals enabled	24 MHz	27	
			16 MHz	20	
			8 MHz	11	mA
IDD		External clock ⁽²⁾ , all peripherals disabled	36 MHz	22	IIIA
			24 MHz	16.5	
			16 MHz	12.5	
			8 MHz	8	

- 1. Guaranteed by characterization results, not tested in production.
- 2. External clock is 8 MHz and PLL is on when f_{HCLK} > 8 MHz.

Table 15. Maximum current consumption in Run mode, code with data processing running from RAM

Symbol	Parameter	er Conditions	£ .	Max ⁽¹⁾	Unit
Symbol	raiametei		f _{HCLK}	T _A = 85 °C	Offic
			36 MHz	34	
		External clock (2), all	24 MHz	24	
		peripherals enabled	16 MHz	17	
	Supply current in		8 MHz	10	mA
I _{DD}	Run mode		36 MHz	18	IIIA
		External clock ⁽²⁾ all	24 MHz	13	
		peripherals disabled	16 MHz	10	
			8 MHz	6	

- 1. Guaranteed by characterization results, tested in production at V_{DD} max, f_{HCLK} max.
- 2. External clock is 8 MHz and PLL is on when f_{HCLK} > 8 MHz.

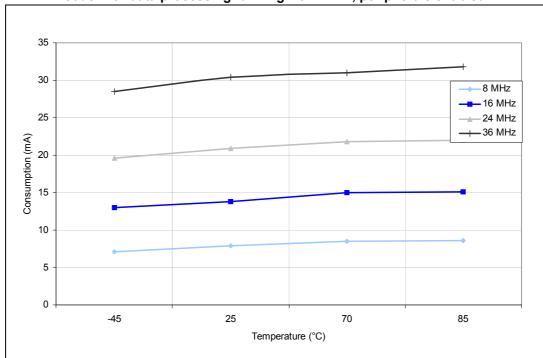
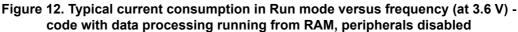


Figure 11. Typical current consumption in Run mode versus frequency (at 3.6 V) - code with data processing running from RAM, peripherals enabled



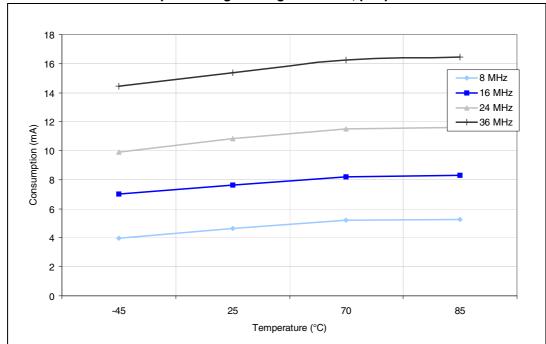


Table 16. Maximum current consumption in Sleep mode, code running from Flash or RAM

Symbol	Parameter	Conditions	_	Max ⁽¹⁾	Unit
Symbol	Parameter	Conditions	fHCLK	T _A = 85 °C	Ullit
			36 MHz	24	
	Supply current in Sleep mode	External clock ⁽²⁾ all peripherals enabled	24 MHz	17	
			16 MHz	12.5	
,			8 MHz	8	mA
I _{DD}		External clock ⁽²⁾ , all peripherals disabled	36 MHz	6	IIIA
			24 MHz	5	
			16 MHz	4.5	
			8 MHz	4	

Guaranteed by characterization results, tested in production at V_{DD} max, f_{HCLK} max with peripherals enabled.

Table 17. Typical and maximum current consumptions in Stop and Standby modes

		Parameter Conditions		Typ ⁽¹⁾			
Symbol	Parameter			V _{DD} / V _{BAT} = 2.4 V	V_{DD}/V_{BA} = 3.3 V	T _A = 85 °C	Unit
	Supply current	Regulator in Run mode, Low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	-	34.5	35	379	
I _{DD}	in Stop mode	Regulator in Low-power mode, Low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	-	24.5	25	365	
-00		Low-speed internal RC oscillator and independent watchdog ON	-	3	3.8	ı	μA
	Supply current in Standby	Low-speed internal RC oscillator ON, independent watchdog OFF	-	2.8	3.6	-	
	mode	Low-speed internal RC oscillator and independent watchdog OFF, low-speed oscillator and RTC OFF	1	1.9	2.1	5 ⁽²⁾	
I _{DD_VBAT}	Backup domain supply current	Low-speed oscillator and RTC ON	1.05	1.1	1.4	2 ⁽²⁾	

^{1.} Typical values are measured at T_A = 25 °C.



^{2.} External clock is 8 MHz and PLL is on when $f_{\mbox{\scriptsize HCLK}}$ > 8 MHz.

^{2.} Guaranteed by characterization results, not tested in production.

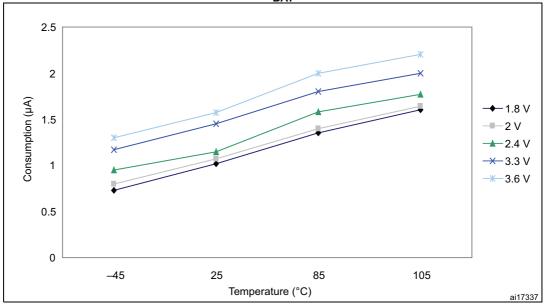
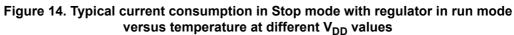
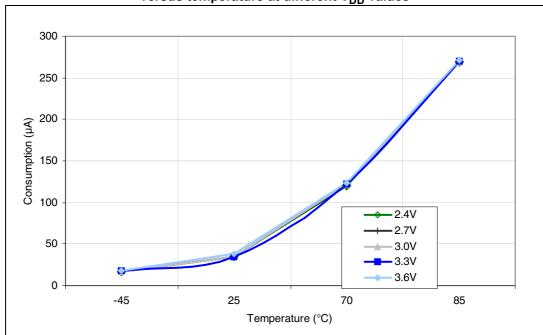


Figure 13. Typical current consumption on V_{BAT} with RTC on vs. temperature at different V_{BAT} values





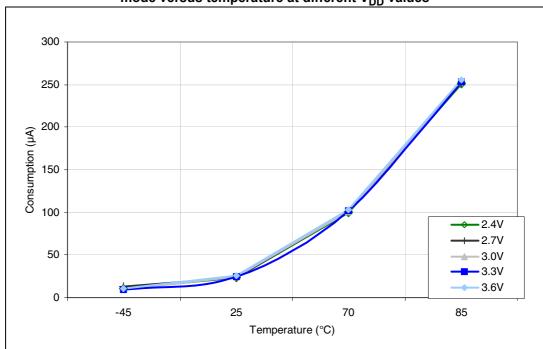
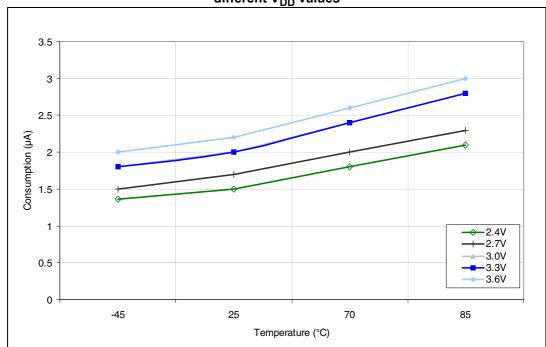


Figure 15. Typical current consumption in Stop mode with regulator in low-power mode versus temperature at different V_{DD} values

Figure 16. Typical current consumption in Standby mode versus temperature at different $V_{\rm DD}$ values



Typical current consumption

The MCU is placed under the following conditions:

- All I/O pins are in input mode with a static value at V_{DD} or V_{SS} (no load)
- All peripherals are disabled except if it is explicitly mentioned
- The Flash access time is adjusted to f_{HCLK} frequency (0 wait state from 0 to 24 MHz, 1 wait state from 24 to 36 MHz)
- Prefetch is on (reminder: this bit must be set before clock setting and bus prescaling)
- When the peripherals are enabled f_{PCLK1} = f_{HCLK/4}, f_{PCLK2} = f_{HCLK/2}, f_{ADCCLK} = f_{PCLK2}/4
- When the peripherals are enabled f_{PCLK1} = f_{HCLK}, f_{PCLK2} = f_{HCLK}, f_{ADCCLK} = f_{PCLK2}/2

The parameters given in *Table 18* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

Table 18. Typical current consumption in Run mode, code with data processing running from Flash

				Typ ⁽¹⁾	Typ ⁽¹⁾	
Symbol	Parameter	Conditions	f _{HCLK}	All peripherals enabled ⁽²⁾	All peripherals disabled	Unit
			36 MHz	26.6	16.2	
			24 MHz	18.5	11.4	
			16 MHz	12.8	8.2	
			8 MHz	7.2	5	
		External clock ⁽³⁾	4 MHz	4.2	3.1	
	Supply current in		2 MHz	2.7	2.1	
			1 MHz	2	1.7	
			500 kHz	1.6	1.4	
			125 kHz	1.3	1.2	mA
I _{DD}	Run mode		36 MHz	26	15.6	IIIA
			24 MHz	17.9	10.8	
		Running on high speed	16 MHz	12.2	7.6	
		internal RC	8 MHz	6.6	4.4	
		(HSI), AHB prescaler	4 MHz	3.6	2.5	
		used to	2 MHz	2.1	1.5	
		reduce the frequency	1 MHz	1.4	1.1	
			500 kHz	1	0.8	
			125 kHz	0.7	0.6	

^{1.} Typical values are measures at T_A = 25 °C, V_{DD} = 3.3 V.

^{2.} Add an additional power consumption of 0.8 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is on (ADON bit is set in the ADC_CR2 register).

^{3.} External clock is 8 MHz and PLL is on when f_{HCLK} > 8 MHz.

Table 19. Typical current consumption in Sleep mode, code running from Flash or RAM

				Typ ⁽¹⁾	Typ ⁽¹⁾	
Symbol	Parameter	Conditions	f _{HCLK}	All peripherals enabled ⁽²⁾	All peripherals disabled	Unit
			36 MHz	15.1	3.6	

External clock(3)

Supply current in Sleep mode

- 1. Typical values are measures at T_A = 25 °C, V_{DD} = 3.3 V.
- 2. Add an additional power consumption of 0.8 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is on (ADON bit is set in the ADC_CR2 register).
- 3. External clock is 8 MHz and PLL is on when $f_{HCLK} > 8$ MHz.

On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 20*. The MCU is placed under the following conditions:

- $\bullet \hspace{0.5cm}$ all I/O pins are in input mode with a static value at V_{DD} or V_{SS} (no load)
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
 - with all peripherals clocked off
 - with only one peripheral clocked on
- ambient operating temperature and V_{DD} supply voltage conditions summarized in Table 7.



Table 20. Peripheral current consumption⁽¹⁾

Peri	pherals	μ A /MHz
	DMA1	20.42
	DMA2	19.03
AHB (up to36 MHz)	FSMC	52.36
	CRC	2.36
	BusMatrix ⁽²⁾	9.72
	APB1-Bridge	7.78
	TIM2	33.06
	TIM3	31.94
	TIM4	31.67
	TIM5	31.94
	TIM6	8.06
	TIM7	8.06
	SPI2/I2S2 ⁽³⁾	8.33
	SPI3/I2S3 ⁽³⁾	8.33
	USART2	12.22
APB1 (up to 18 MHz)	USART3	12.22
	UART4	12.22
	UART5	12.22
	I2C1	10.28
	I2C2	10.00
	USB	18.06
	DAC ⁽⁴⁾	8.06
	WWDG	3.89
	PWR	1.11
	BKP	1.11
	IWDG	5.28

μA/MHz **Peripherals** APB2-Bridge 4.17 **GPIOA** 8.47 **GPIOB** 8.47 **GPIOC** 6.53 **GPIOD** 8.47 **GPIOE** 6.53 APB2 (up to 36 MHz) **GPIOF** 6.53 **GPIOG** 6.11 SPI1 4.72 USART1 12.50 TIM1 22.92 TIM8 22.92 ADC1⁽⁵⁾⁽⁶⁾ 17.32

Table 20. Peripheral current consumption⁽¹⁾ (continued)

- 1. f_{HCLK} = 36 MHz, f_{APB1} = $f_{HCLK/2}$, f_{APB2} = f_{HCLK} , default prescaler value for each peripheral.
- 2. The BusMatrix is automatically active when at least one master peripheral is ON.
- 3. When the I2S is enabled, a current consumption of 0.02 mA must be added.
- 4. When DAC_OUT1 or DAC_OUT2 is enabled, a current consumption of 0.36 mA must be added.
- 5. Specific conditions for ADC: f_{HCLK} = 28 MHz, f_{APB1} = $f_{HCLK/2}$, f_{APB2} = f_{HCLK} , f_{ADCCLK} = f_{APB2} /2. When ADON bit in the ADC_CR2 register is set to 1, the current consumption is equal to 0.54 mA.
- 6. When the ADC is enabled, a current consumption of 0.08 mA must be added.

5.3.6 External clock source characteristics

High-speed external user clock generated from an external source

The characteristics given in *Table 21* result from tests performed using an high-speed external clock source, and under ambient temperature and supply voltage conditions summarized in *Table 10*.

Table 21. High-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSE_ext}	User external clock source frequency ⁽¹⁾		1	8	25	MHz
V _{HSEH}	OSC_IN input pin high level voltage		0.7V _{DD}	-	V _{DD}	V
V _{HSEL}	OSC_IN input pin low level voltage	-	V _{SS}	ı	0.3V _{DD}	V
$\begin{matrix} t_{w(\text{HSE})} \\ t_{w(\text{HSE})} \end{matrix}$	OSC_IN high or low time ⁽¹⁾		5	ı	-	ns
t _{r(HSE)}	OSC_IN rise or fall time ⁽¹⁾		-	-	20	113



Table 21. High-speed external user clock characteristics

Parameter Conditions Min Typ Max

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
C _{in(HSE)}	OSC_IN input capacitance ⁽¹⁾	-	-	5	-	pF
DuCy _(HSE)	Duty cycle	-	45	-	55	%
ΙL	OSC_IN Input leakage current	$V_{SS} \le V_{IN} \le V_{DD}$	-	-	±1	μΑ

^{1.} Guaranteed by design, not tested in production

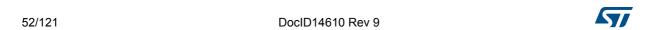
Low-speed external user clock generated from an external source

The characteristics given in *Table 22* result from tests performed using an low-speed external clock source, and under ambient temperature and supply voltage conditions summarized in *Table 10*.

Table 22. Low-speed user external clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{LSE_ext}	User external clock source frequency ⁽¹⁾	-	-	32.768	1000	kHz
V _{LSEH}	OSC32_IN input pin high level voltage	-	0.7V _{DD}	-	V_{DD}	V
V _{LSEL}	OSC32_IN input pin low level voltage	-	V _{SS}	-	0.3V _{DD}	V
t _{w(LSE)}	OSC32_IN high or low time ⁽¹⁾	-	450	-	-	ns
$\begin{matrix} t_{r(LSE)} \\ t_{f(LSE)} \end{matrix}$	OSC32_IN rise or fall time ⁽¹⁾	-	-	1	50	113
C _{in(LSE)}	OSC32_IN input capacitance ⁽¹⁾	-	-	5	-	pF
DuCy _(LSE)	Duty cycle	-	30	-	70	%
IL	OSC32_IN Input leakage current	$V_{SS} \le V_{IN} \le V_{DD}$	-	-	±1	μΑ

^{1.} Guaranteed by design, not tested in production.



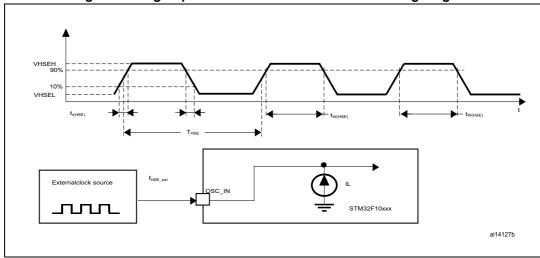
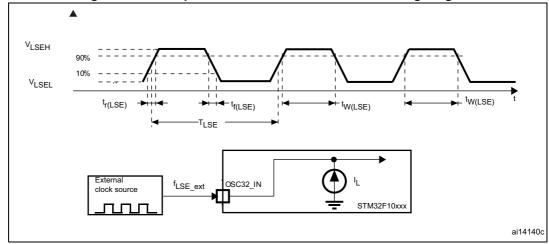


Figure 17. High-speed external clock source AC timing diagram







High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 16 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 23*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{OSC_IN}	Oscillator frequency	-	4	8	16	MHz
R _F	Feedback resistor	-	-	200	-	kΩ
С	Recommended load capacitance versus equivalent serial resistance of the crystal (R _S) ⁽³⁾	R _S = 30 Ω	-	30	-	pF
i ₂	HSE driving current	V_{DD} = 3.3 V V_{IN} = V_{SS} with 30 pF load	1	ı	1	mA
9 _m	Oscillator transconductance	Startup	25	-	-	mA/V
t _{SU(HSE)} ⁽⁴⁾	Startup time	V _{DD} is stabilized	ı	2	-	ms

Table 23. HSE 4-16 MHz oscillator characteristics⁽¹⁾⁽²⁾

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 19*). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} . Refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website *www.st.com*.

^{1.} Resonator characteristics given by the crystal/ceramic resonator manufacturer.

^{2.} Guaranteed by characterization results, not tested in production.

^{3.} The relatively low value of the RF resistor offers a good protection against issues resulting from use in a humid environment, due to the induced leakage and the bias condition change. However, it is recommended to take this point into account if the MCU is used in tough humidity conditions.

^{4.} t_{SU(HSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

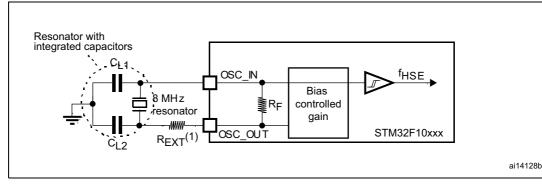


Figure 19. Typical application with an 8 MHz crystal

1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in Table 24. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

	Table 24. LSE oscillator characteristics (f _{LSE} = 32.768 kHz) ^{(1) (2)}							
Symbol	Parameter	Co	Min	Тур	Max	Unit		
R _F	Feedback resistor	-	-	-	5	-	МΩ	
С	Recommended load capacitance versus equivalent serial resistance of the crystal (R _S)	R _S = 30 KΩ	-	-	-	15	pF	

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R _F	Feedback resistor	-	-	-	5	-	МΩ
С	Recommended load capacitance versus equivalent serial resistance of the crystal (R _S)	R _S = 30 KΩ	-	-	-	15	pF
l ₂	LSE driving current	$V_{DD} = 3.3 \text{ V}$ $V_{IN} = V_{SS}$	-	-	-	1.4	μA
9 _m	Oscillator transconductance	-	-	5	-	-	μA/V
			T _A = 50 °C	-	1.5	-	
			T _A = 25 °C	-	2.5	-	
			T _A = 10 °C	-	4	-	
4 (3)	Otantum tima	V _{DD} is	T _A = 0 °C	-	6	-	
t _{SU(LSE)} ⁽³⁾	Startup time	stabilized	T _A = -10 °C	-	10	-	S
			T _A = -20 °C	-	17	-	
			T _A = -30 °C	-	32	-	
			T _A = -40 °C	-	60	-	

^{1.} Guaranteed by characterization results, not tested in production.

 $t_{\text{SU(LSE)}}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer



Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers"

Note:

For C_{L1} and C_{L2} , it is recommended to use high-quality ceramic capacitors in the 5 pF to 15 pF range selected to match the requirements of the crystal or resonator. C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} .

Load capacitance C_L has the following formula: $C_L = C_{L1} \times C_{L2} / (C_{L1} + C_{L2}) + C_{stray}$ where C_{stray} is the pin capacitance and board or trace PCB-related capacitance. Typically, it is between 2 pF and 7 pF.

Caution:

To avoid exceeding the maximum value of C_{L1} and C_{L2} (15 pF) it is strongly recommended to use a resonator with a load capacitance $C_L \le 7$ pF. Never use a resonator with a load capacitance of 12.5 pF.

Example: if you choose a resonator with a load capacitance of C_L = 6 pF, and C_{stray} = 2 pF, then $C_{L,1}$ = $C_{L,2}$ = 8 pF.

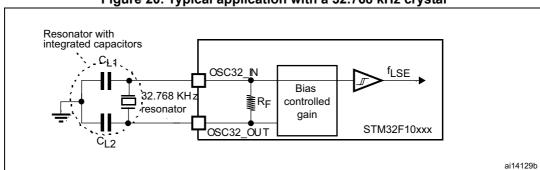


Figure 20. Typical application with a 32.768 kHz crystal

5.3.7 Internal clock source characteristics

The parameters given in $Table\ 25$ are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in $Table\ 10$.

High-speed internal (HSI) RC oscillator

Table 25. HSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions		Min	Тур	Max	Unit													
f _{HSI}	Frequency	-		-	8	-	MHz													
DuCy _(HSI)	Duty cycle	-		45	-	55	%													
		User-trimmed register ⁽²⁾	d with the RCC_CR	-	-	1 ⁽³⁾	%													
	Accuracy of the HSI		$T_A = -40 \text{ to } 105 ^{\circ}\text{C}$	-2	-	2.5	%													
ACC _{HSI}	oscillator	Factory-	T _A = -10 to 85 °C	-1.5	-	2.2	%													
		calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	calibrated ⁽⁴⁾	T _A = 0 to 70 °C	-1.3	-	2	%
				T _A = 25 °C	-1.1	-	1.8	%												
t _{su(HSI)} ⁽⁴⁾	HSI oscillator startup time	-		1	-	2	μs													
I _{DD(HSI)} ⁽⁴⁾	HSI oscillator power consumption	-		-	80	100	μА													

^{1.} V_{DD} = 3.3 V, T_A = -40 to 85 °C unless otherwise specified.



- Refer to application note AN2868 "STM32F10xxx internal RC oscillator (HSI) calibration" available from the ST website www.st.com
- 3. Guaranteed by design, not tested in production.
- 4. Guaranteed by characterization results, not tested in production.

Low-speed internal (LSI) RC oscillator

Table 26. LSI oscillator characteristics (1)

Symbol	Parameter	Min	Тур	Max	Unit
f _{LSI} ⁽²⁾	Frequency	30	40	60	kHz
t _{su(LSI)} (3)	LSI oscillator startup time	-	-	85	μs
I _{DD(LSI)} ⁽³⁾	LSI oscillator power consumption	-	0.65	1.2	μA

- 1. V_{DD} = 3 V, T_A = -40 to 85 °C unless otherwise specified.
- 2. Guaranteed by characterization results, not tested in production.
- 3. Guaranteed by design, not tested in production.

Wakeup time from low-power mode

The wakeup times given in *Table 27* are measured on a wakeup phase with an 8-MHz HSI RC oscillator. The clock source used to wake up the device depends from the current operating mode:

- Stop or Standby mode: the clock source is the RC oscillator
- Sleep mode: the clock source is the clock that was set before entering Sleep mode.

All timings are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

Table 27. Low-power mode wakeup timings

Symbol	Parameter	Тур	Unit
t _{WUSLEEP} (1)	Wakeup from Sleep mode	1.8	μs
₊ (1)	Wakeup from Stop mode (regulator in run mode)	3.6	116
t _{WUSTOP} ⁽¹⁾	Wakeup from Stop mode (regulator in low-power mode)	5.4	μs
t _{WUSTDBY} (1)	Wakeup from Standby mode	50	μs

 The wakeup times are measured from the wakeup event to the point at which the user application code reads the first instruction.



5.3.8 PLL characteristics

The parameters given in *Table 28* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

Value **Symbol** Unit **Parameter** Max⁽¹⁾ Min⁽¹⁾ Тур PLL input clock⁽²⁾ 1 8.0 25 MHz f_{PLL_IN} PLL input clock duty cycle 40 60 % PLL multiplier output clock 16 36 MHz f_{PLL} OUT PLL lock time 200 μs **t**LOCK Jitter 300 Cycle-to-cycle jitter ps

Table 28. PLL characteristics

5.3.9 Memory characteristics

Flash memory

The characteristics are given at $T_A = -40$ to 85 °C unless otherwise specified.

Max⁽¹⁾ **Symbol Conditions** Min Unit **Parameter** Тур $T_A = -40$ to +85 °C 52.5 70 16-bit programming time 40 t_{prog} μs Page (2 KB) erase time $T_A = -40 \text{ to } +85 \text{ }^{\circ}\text{C}$ 20 40 ms t_{ERASE} $T_A = -40$ to +85 °C t_{ME} Mass erase time 20 40 ms Read mode f_{HCLK} = 36 MHz with 1 28 mA wait state, $V_{DD} = 3.3 \text{ V}$ Write mode $f_{HCLK} = 36 \text{ MHz}, V_{DD} =$ 7 mΑ 3.3 V Supply current I_{DD} Erase mode f_{HCLK} = 36 MHz, V_{DD} = 5 mΑ Power-down mode / Halt, μΑ 50 $V_{DD} = 3.0 \text{ to } 3.6 \text{ V}$ 2 ٧ Programming voltage 3.6 V_{proq}

Table 29. Flash memory characteristics



^{1.} Guaranteed by characterization results, not tested in production.

Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by f_{PLL_OUT}.

^{1.} Guaranteed by design, not tested in production.

Value **Symbol Parameter Conditions** Unit Min⁽¹⁾ $T_A = -40 \, ^{\circ}\text{C}$ to 85 $^{\circ}\text{C}$ Endurance 10 kcycles N_{END} $T_A = 85 \, ^{\circ}C$, 1 kcycle⁽²⁾ 30 Data retention Years t_{RET} $T_A = 55 \, ^{\circ}C$, 10 kcycle⁽²⁾ 20

Table 30. Flash memory endurance and data retention

5.3.10 FSMC characteristics

Asynchronous waveforms and timings

Figure 21 through Figure 24 represent asynchronous waveforms and Table 31 through Table 34 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- AddressSetupTime = 0
- AddressHoldTime = 1
- DataSetupTime = 1

^{1.} Guaranteed by characterization results, not tested in production.

^{2.} Cycling performed over the whole temperature range.

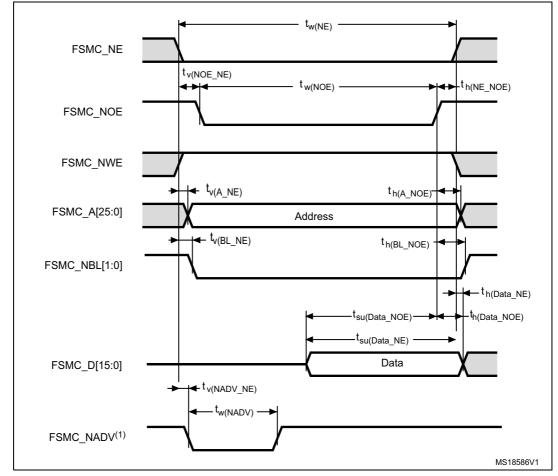


Figure 21. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms

1. Mode 2/B, C and D only. In Mode 1, FSMC_NADV is not used.

Note: $FSMC_BusTurnAroundDuration = 0$.

Table 31. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings⁽¹⁾ (2)

Symbol	Parameter	Min	Max	Unit
$t_{W(NE)}$	FSMC_NE low time	5t _{HCLK} – 1.5	5t _{HCLK} + 2	ns
t _{v(NOE_NE)}	FSMC_NEx low to FSMC_NOE low	0.5	1.5	ns
t _{w(NOE)}	FSMC_NOE low time	5t _{HCLK} - 1.5	5t _{HCLK} + 1.5	ns
t _{h(NE_NOE)}	FSMC_NOE high to FSMC_NE high hold time	-1.5	-	ns
t _{v(A_NE)}	FSMC_NEx low to FSMC_A valid	-	7	ns
t _{h(A_NOE)}	Address hold time after FSMC_NOE high	0.1	-	ns
t _{v(BL_NE)}	FSMC_NEx low to FSMC_BL valid	-	0	ns
t _{h(BL_NOE)}	FSMC_BL hold time after FSMC_NOE high	0	-	ns
t _{su(Data_NE)}	Data to FSMC_NEx high setup time	2t _{HCLK} + 25	-	ns
t _{su(Data_NOE)}	Data to FSMC_NOEx high setup time	2t _{HCLK} + 25	-	ns
t _{h(Data_NOE)}	Data hold time after FSMC_NOE high	0	-	ns

Figure 22. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms

1. Mode 2/B, C and D c0 13Mode10 13 FSeC_NAD i(s nt used.s)]TJ17.98 0 017.98 1 .

Table 32. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{h(Data_NWE)}	Data hold time after FSMC_NWE high	t _{HCLK}	-	ns
t _{v(NADV_NE)}	FSMC_NEx low to FSMC_NADV low	-	5.5	ns
t _{w(NADV)}	FSMC_NADV low time	-	t _{HCLK} + 1.5	ns

^{1.} C_L = 15 pF.

Figure 23. Asynchronous multiplexed NOR/PSRAM read waveforms

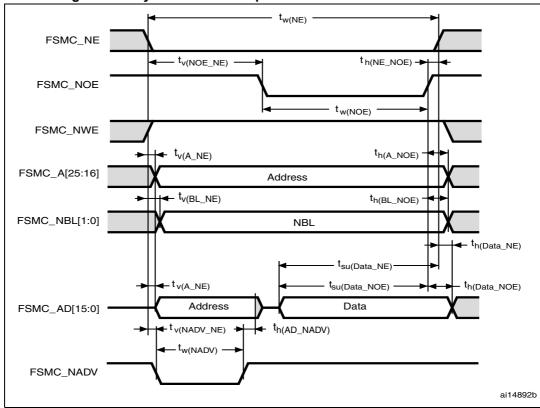


Table 33. Asynchronous multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter Min		Max	Unit
t _{w(NE)}	FSMC_NE low time	7t _{HCLK} – 2	7t _{HCLK} + 2	ns
t _{v(NOE_NE)}	FSMC_NEx low to FSMC_NOE low	3t _{HCLK} - 0.5	3t _{HCLK} + 1.5	ns
$t_{w(NOE)}$	FSMC_NOE low time	4t _{HCLK} – 1	4t _{HCLK} + 2	ns
t _{h(NE_NOE)}	FSMC_NOE high to FSMC_NE high hold time	-1	-	ns
t _{v(A_NE)}	FSMC_NEx low to FSMC_A valid	-	0	ns
t _{v(NADV_NE)}	FSMC_NEx low to FSMC_NADV low	3	5	ns
t _{w(NADV)}	FSMC_NADV low time	t _{HCLK} -1.5	t _{HCLK} + 1.5	ns
t _{h(AD_NADV)}	FSMC_AD (address) valid hold time after FSMC_NADV high	t _{HCLK}	-	ns

^{2.} Guaranteed by characterization results, not tested in production.

Table 33. Asynchronous multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾ (continued)

Symbol	Parameter	Min	Max	Unit
t _{h(A_NOE)}	Address hold time after FSMC_NOE high	t _{HCLK}	-	ns
t _{h(BL_NOE)}	FSMC_BL hold time after FSMC_NOE high	0	-	ns
t _{v(BL_NE)}	FSMC_NEx low to FSMC_BL valid	-	0	ns
t _{su(Data_NE)}	Data to FSMC_NEx high setup time	2t _{HCLK} + 24	-	ns
t _{su(Data_NOE)}	Data to FSMC_NOE high setup time	2t _{HCLK} + 25	-	ns
t _{h(Data_NE)}	Data hold time after FSMC_NEx high	0	-	ns
t _{h(Data_NOE)}	Data hold time after FSMC_NOE high	0	-	ns

^{1.} $C_L = 15 pF$.



^{2.} Guaranteed by characterization results, not tested in production.

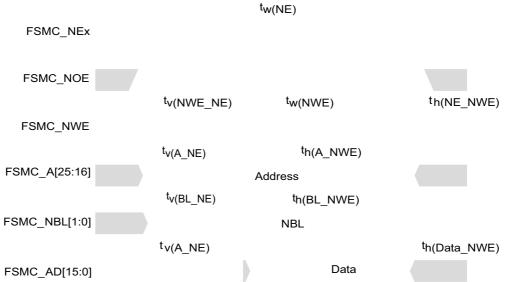


Figure 24. Asynchronous multiplexed NOR/PSRAM write waveforms

Table 34. Asynchronous multiplexed NOR/PSRAM write timings⁽¹⁾⁽²⁾

Symbol	bol Parameter Min		Max	Unit
t _{w(NE)}	FSMC_NE low time	5t _{HCLK} – 1	5t _{HCLK} + 2	ns
t _{v(NWE_NE)}	FSMC_NEx low to FSMC_NWE low	1t _{HCLK}	1t _{HCLK} + 1	ns
t _{w(NWE)}	FSMC_NWE low time	3t _{HCLK} – 1	2	ns
t _{h(NE_NWE)}	FSMC_NWE high to FSMC_NE high hold time	t _{HCLK} – 1	-	ns
t _{v(A_NE)}	FSMC_NEx low to FSMC_A valid	-	7	ns
t _{v(NADV_NE)}	FSMC_NEx low to FSMC_NADV low	3	5	ns
t _{w(NADV)}	FSMC_NADV low time	t _{HCLK} – 1	t _{HCLK} + 1	ns
t _{h(AD_NADV)}	FSMC_AD (address) valid hold time after FSMC_NADV high	t _{HCLK} – 3	-	ns
t _{h(A_NWE)}	Address hold time after FSMC_NWE high	1t _{HCLK}	-	ns
t _{v(BL_NE)}	FSMC_NEx low to FSMC_BL valid	-	1.6	ns
t _{h(BL_NWE)}	FSMC_BL hold time after FSMC_NWE high	t _{HCLK} – 1.5	-	ns
t _{v(Data_NADV)}	FSMC_NADV high to Data valid	-	t _{HCLK} + 1.5	ns
t _{h(Data_NWE)}	Data hold time after FSMC_NWE high	t _{HCLK} – 5	-	ns

^{1.} C_L = 15 pF.

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^{2.} Guaranteed by characterization results, not tested in production..

Synchronous waveforms and timings

Figure 25 through Figure 28 represent synchronous waveforms and Table 36 through Table 38 provide the corresponding timings. The results shown in these tables are obtained with the following FSMC configuration:

- BurstAccessMode = FSMC_BurstAccessMode_Enable;
- MemoryType = FSMC_MemoryType_CRAM;
- WriteBurst = FSMC WriteBurst Enable;
- CLKDivision = 1; (0 is not supported, see the STM32F10xxx reference manual)
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

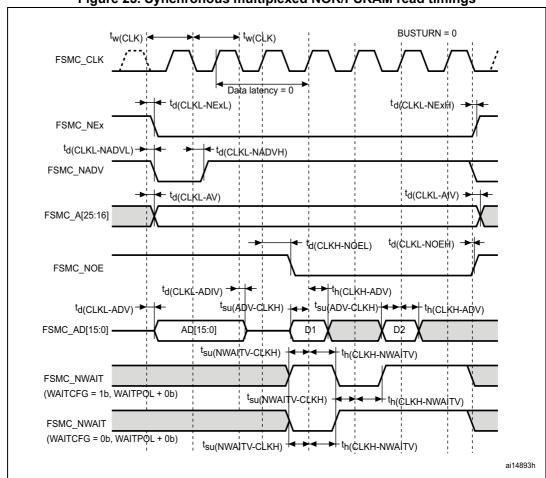


Figure 25. Synchronous multiplexed NOR/PSRAM read timings

Table 35. Synchronous multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FSMC_CLK period	55.5	-	ns
t _{d(CLKL-NExL)}	FSMC_CLK low to FSMC_NEx low (x = 02)	-	1.5	ns
t _{d(CLKL-NExH)}	FSMC_CLK low to FSMC_NEx high (x = 02)	2	-	ns
t _{d(CLKL-NADVL)}	FSMC_CLK low to FSMC_NADV low	-	4	ns
t _{d(CLKL-NADVH)}	FSMC_CLK low to FSMC_NADV high	5	-	ns
t _{d(CLKL-AV)}	FSMC_CLK low to FSMC_Ax valid (x = 1625)	-	0	ns
t _{d(CLKL-AIV)}	FSMC_CLK low to FSMC_Ax invalid (x = 1625)	2	-	ns
t _{d(CLKH-NOEL)}	FSMC_CLK high to FSMC_NOE low	-	1	ns
t _{d(CLKL-NOEH)}	FSMC_CLK low to FSMC_NOE high	0.5	-	ns
t _{d(CLKL-ADV)}	FSMC_CLK low to FSMC_AD[15:0] valid	-	12	ns
t _{d(CLKL-ADIV)}	FSMC_CLK low to FSMC_AD[15:0] invalid	0	-	ns
t _{su(ADV-CLKH)}	FSMC_A/D[15:0] valid data before FSMC_CLK high	6	-	ns
t _{h(CLKH-ADV)}	FSMC_A/D[15:0] valid data after FSMC_CLK high	0	-	ns
t _{su(NWAITV-CLKH)}	FSMC_NWAIT valid before FSMC_CLK high	8	-	ns
t _{h(CLKH-NWAITV)}	FSMC_NWAIT valid after FSMC_CLK high	2	-	ns

^{1.} C_L = 15 pF.



^{2.} Guaranteed by characterization results, not tested in production..

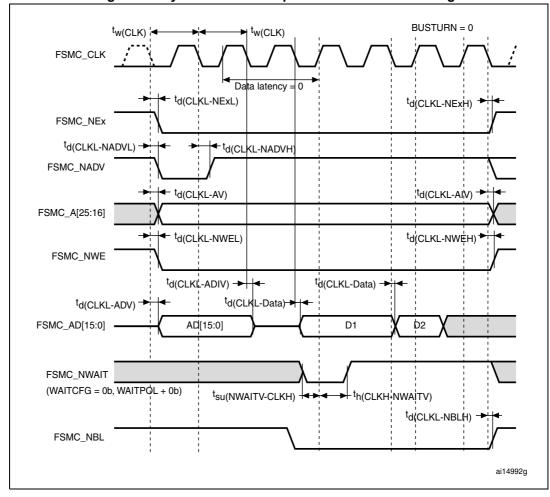


Figure 26. Synchronous multiplexed PSRAM write timings



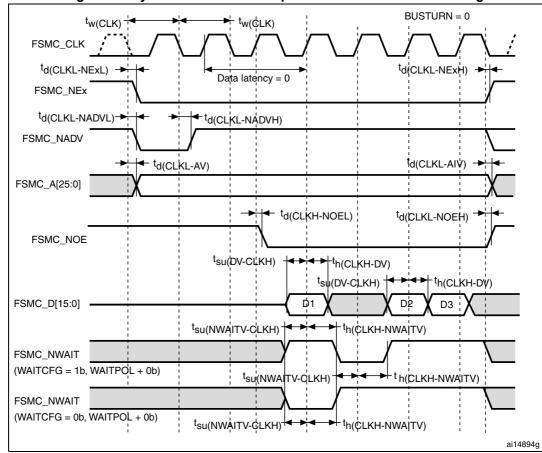


Figure 27. Synchronous non-multiplexed NOR/PSRAM read timings

Table 37. Synchronous non-multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FSMC_CLK period	55.5	-	ns
t _{d(CLKL-NExL)}	FSMC_CLK low to FSMC_NEx low (x = 02)	-	1.5	ns
t _{d(CLKL-NExH)}	FSMC_CLK low to FSMC_NEx high (x = 02)	2	-	ns
t _{d(CLKL-NADVL)}	FSMC_CLK low to FSMC_NADV low	-	4	ns
t _{d(CLKL-NADVH)}	FSMC_CLK low to FSMC_NADV high	5	-	ns
t _{d(CLKL-AV)}	FSMC_CLK low to FSMC_Ax valid (x = 025)	-	0	ns
t _{d(CLKL-AIV)}	FSMC_CLK low to FSMC_Ax invalid (x = 025)	4	-	ns
t _{d(CLKH-NOEL)}	FSMC_CLK high to FSMC_NOE low	-	1.5	ns
t _{d(CLKL-NOEH)}	FSMC_CLK low to FSMC_NOE high	1.5	-	ns
t _{su(DV-CLKH)}	FSMC_D[15:0] valid data before FSMC_CLK high	6.5	-	ns
t _{h(CLKH-DV)}	FSMC_D[15:0] valid data after FSMC_CLK high	7	-	ns
t _{su(NWAITV-CLKH)}	FSMC_NWAIT valid before FSMC_SMCLK high	7	-	ns
t _{h(CLKH-NWAITV)}	FSMC_NWAIT valid after FSMC_CLK high	2	-	ns

^{1.} $C_L = 15 pF$.



2. Guaranteed by characterization results, not tested in production.

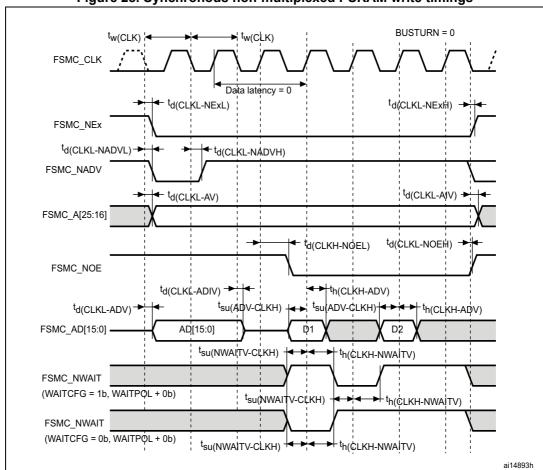


Figure 28. Synchronous non-multiplexed PSRAM write timings

Table 38. Synchronous non-multiplexed PSRAM write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FSMC_CLK period	55.5	-	ns
t _{d(CLKL-NExL)}	FSMC_CLK low to FSMC_NEx low (x = 02)	-	2	ns
t _{d(CLKL-NExH)}	FSMC_CLK low to FSMC_NEx high (x = 02)	2	-	ns
t _{d(CLKL-NADVL)}	FSMC_CLK low to FSMC_NADV low	-	4	ns
t _{d(CLKL-NADVH)}	FSMC_CLK low to FSMC_NADV high	5	-	ns
t _{d(CLKL-AV)}	FSMC_CLK low to FSMC_Ax valid (x = 1625)	-	0	ns
t _{d(CLKL-AIV)}	FSMC_CLK low to FSMC_Ax invalid (x = 1625)	2	-	ns
t _{d(CLKL-NWEL)}	FSMC_CLK low to FSMC_NWE low	-	1	ns
t _{d(CLKL-NWEH)}	FSMC_CLK low to FSMC_NWE high	1	-	ns
t _{d(CLKL-Data)}	FSMC_D[15:0] valid data after FSMC_CLK low	-	6	ns
t _{su(NWAITV-CLKH)}	FSMC_NWAIT valid before FSMC_CLK high	7	-	ns

Table 38. Synchronous non-multiplexed PSRAM write timings⁽¹⁾⁽²⁾ (continued)

Symbol	Parameter	Min	Max	Unit
t _{h(CLKH-NWAITV)}	FSMC_NWAIT valid after FSMC_CLK high	2	-	ns
t _{d(CLKL-NBLH)}	FSMC_CLK low to FSMC_NBL high	1	-	ns

^{1.} $C_L = 15 pF$.

PC Card/CompactFlash controller waveforms and timings

Figure 29 through *Figure 34* represent synchronous waveforms and *Table 39* provides the corresponding timings. The results shown in this table are obtained with the following FSMC configuration:

- COM.FSMC_SetupTime = 0x04;
- COM.FSMC WaitSetupTime = 0x07;
- COM.FSMC_HoldSetupTime = 0x04;
- COM.FSMC_HiZSetupTime = 0x00;
- ATT.FSMC SetupTime = 0x04;
- ATT.FSMC_WaitSetupTime = 0x07;
- ATT.FSMC_HoldSetupTime = 0x04;
- ATT.FSMC HiZSetupTime = 0x00;
- IO.FSMC_SetupTime = 0x04;
- IO.FSMC WaitSetupTime = 0x07;
- IO.FSMC HoldSetupTime = 0x04;
- IO.FSMC_HiZSetupTime = 0x00;
- TCLRSetupTime = 0;
- TARSetupTime = 0;



^{2.} Guaranteed by characterization results, not tested in production.

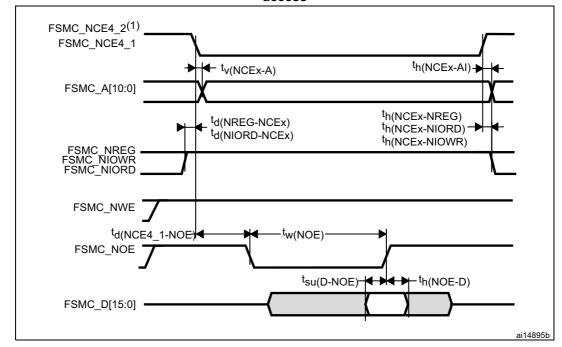
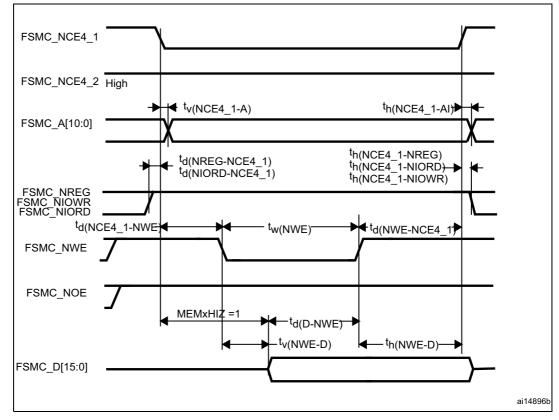


Figure 29. PC Card/CompactFlash controller waveforms for common memory read access

1. FSMC_NCE4_2 remains high (inactive during 8-bit access.

Figure 30. PC Card/CompactFlash controller waveforms for common memory write access



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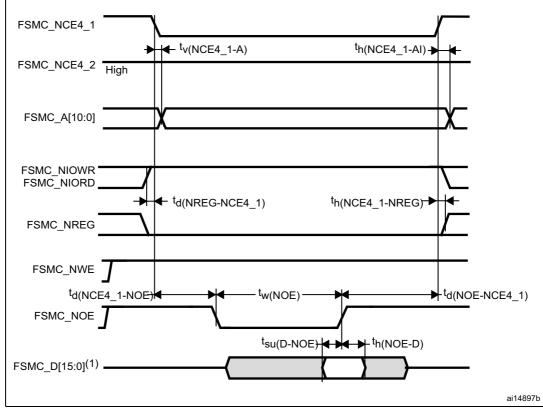


Figure 31. PC Card/CompactFlash controller waveforms for attribute memory read access

1. Only data bits 0...7 are read (bits 8...15 are disregarded).



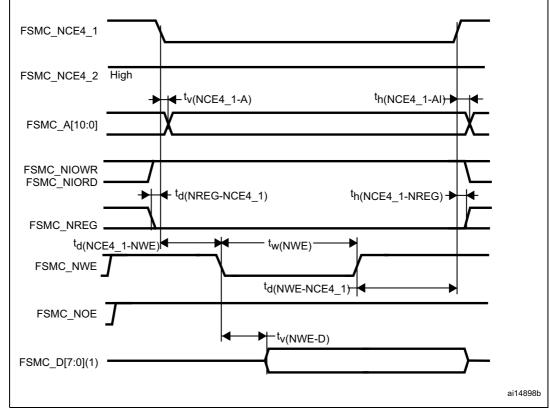


Figure 32. PC Card/CompactFlash controller waveforms for attribute memory write access

1. Only data bits 0...7 are driven (bits 8...15 remains HiZ).

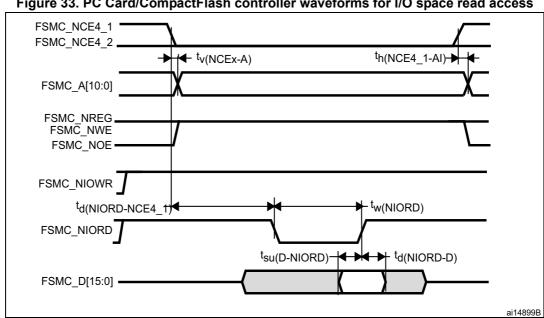


Figure 33. PC Card/CompactFlash controller waveforms for I/O space read access

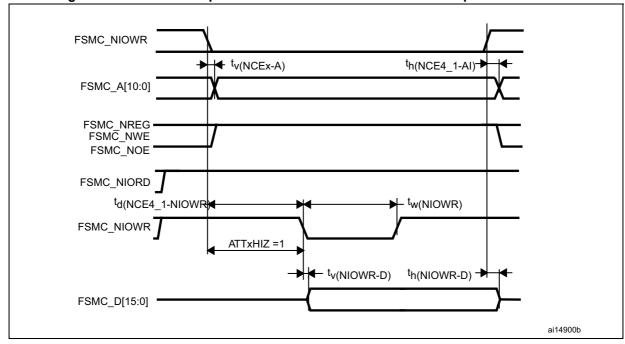


Figure 34. PC Card/CompactFlash controller waveforms for I/O space write access

Table 39. Switching characteristics for PC Card/CF read and write cycles⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _V (NCEx-A) t _V (NCE4_1-A)	FSMC_NCEx low (x = 4_1/4_2) to FSMC_Ay valid (y = 010) FSMC_NCE4_1 low (x = 4_1/4_2) to FSMC_Ay valid (y = 010)	-	0	ns
th(NCEx-AI) th(NCE4_1-AI)	FSMC_NCEx high (x = 4_1/4_2) to FSMC_Ax invalid (x = 010) FSMC_NCE4_1 high (x = 4_1/4_2) to FSMC_Ax invalid (x = 010)	2.5	-	ns
$t_{d(NREG-NCEx)} \\ t_{d(NREG-NCE4_1)}$	FSMC_NCEx low to FSMC_NREG valid FSMC_NCE4_1 low to FSMC_NREG valid	-	5	ns
t _{h(NCEx-NREG)} t _{h(NCE4_1-NREG)}	FSMC_NCEx high to FSMC_NREG invalid FSMC_NCE4_1 high to FSMC_NREG invalid	t _{HCLK} + 3	-	ns
t _{d(NCE4_1-NOE)}	FSMC_NCE4_1 low to FSMC_NOE low	-	5t _{HCLK} + 2	ns
t _{w(NOE)}	FSMC_NOE low width	8t _{HCLK} -1.5	8t _{HCLK} + 1	ns
t _{d(NOE-NCE4_1}	FSMC_NOE high to FSMC_NCE4_1 high	5t _{HCLK} + 2	-	ns
t _{su(D-NOE)}	FSMC_D[15:0] valid data before FSMC_NOE high	25	-	ns
t _{h(NOE-D)}	FSMC_D[15:0] valid data after FSMC_NOE high	15	-	ns
t _{w(NWE)}	FSMC_NWE low width	8t _{HCLK} – 1	8t _{HCLK} + 2	ns
t _{d(NWE-NCE4_1)}	FSMC_NWE high to FSMC_NCE4_1 high	5t _{HCLK} + 2	-	ns
t _{d(NCE4_1-NWE)}	FSMC_NCE4_1 low to FSMC_NWE low	-	5t _{HCLK} + 1.5	ns
t _{v(NWE-D)}	FSMC_NWE low to FSMC_D[15:0] valid	-	0	ns
t _{h(NWE-D)}	FSMC_NWE high to FSMC_D[15:0] invalid	11t _{HCLK}	-	ns



5t_{HCLK} - 5

5t_{HCLK} - 5

4.5

9

8t_{HCLK} + 2

 $5t_{HCLK} + 3ns$

 $5t_{HCLK} + 2.5$

ns

ns

ns

ns

ns

ns

ns

Symbol	Parameter	Min	Max	Unit
t _{d(D-NWE)}	FSMC_D[15:0] valid before FSMC_NWE high	13t _{HCLK}	-	ns
t _{w(NIOWR)}	FSMC_NIOWR low width	8t _{HCLK} + 3	-	ns
t _{v(NIOWR-D)}	FSMC_NIOWR low to FSMC_D[15:0] valid	-	5t _{HCLK} +1	ns
t _{h(NIOWR-D)}	FSMC_NIOWR high to FSMC_D[15:0] invalid	11t _{HCLK}	-	ns

Table 39. Switching characteristics for PC Card/CF read and write cycles⁽¹⁾⁽²⁾ (continued)

t_{su(D-NIORD)}

t_{d(NIORD-D)}

t_{d(NCE4_1-NIOWR)}

th(NCE4_1-NIOWR)

t_{d(NIORD-NCE4_1)}

th(NCE4_1-NIORD)

th(NCEx-NIOWR)

t_{d(NIORD-NCEx)}

t_{h(NCEx-NIORD)}

FSMC NIORD low width

NAND controller waveforms and timings

FSMC_NCE4_1 low to FSMC_NIOWR valid

FSMC NCEx high to FSMC NIOWR invalid

FSMC_NCEx low to FSMC_NIORD valid

FSMC_NCE4_1 low to FSMC_NIORD valid

FSMC NCEx high to FSMC NIORD invalid

FSMC_NCE4_1 high to FSMC_NIORD invalid

FSMC_D[15:0] valid after FSMC_NIORD high

FSMC_D[15:0] valid before FSMC_NIORD high

FSMC_NCE4_1 high to FSMC_NIOWR invalid

Figure 35 through *Figure 38* represent synchronous waveforms and *Table 40* provides the corresponding timings. The results shown in this table are obtained with the following FSMC configuration:

- COM.FSMC SetupTime = 0x01;
- COM.FSMC_WaitSetupTime = 0x03;
- COM.FSMC_HoldSetupTime = 0x02;
- COM.FSMC_HiZSetupTime = 0x01;
- ATT.FSMC_SetupTime = 0x01;
- ATT.FSMC_WaitSetupTime = 0x03;
- ATT.FSMC HoldSetupTime = 0x02;
- ATT.FSMC_HiZSetupTime = 0x01;
- Bank = FSMC_Bank_NAND;
- MemoryDataWidth = FSMC MemoryDataWidth 16b;
- ECC = FSMC_ECC_Enable;
- ECCPageSize = FSMC ECCPageSize 512Bytes;
- TCLRSetupTime = 0;
- TARSetupTime = 0;

 $t_{\text{w(NIORD)}}$ 1. $C_{\text{I}} = 15 \text{ pF}.$

^{2.} Guaranteed by characterization results, not tested in production.

FSMC_NCEx Low

ALE (FSMC_A17)
CLE (FSMC_A16)

FSMC_NWE

FSMC_NOE (NRE)

td(ALE-NOE)

tsu(D-NOE)

th(NOE-ALE)

ai14901b

Figure 35. NAND controller waveforms for read access

Figure 36. NAND controller waveforms for write access

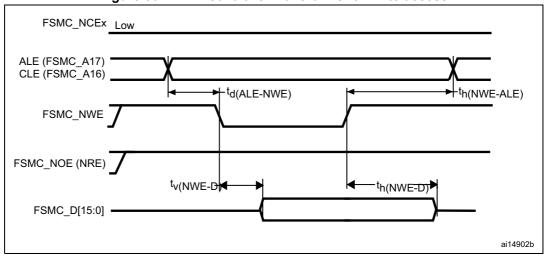
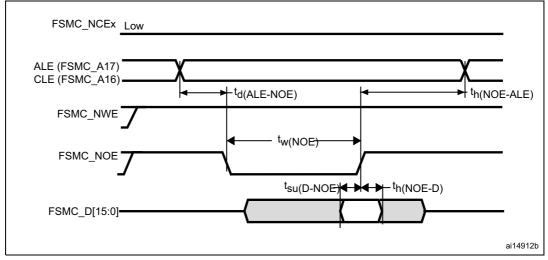


Figure 37. NAND controller waveforms for common memory read access



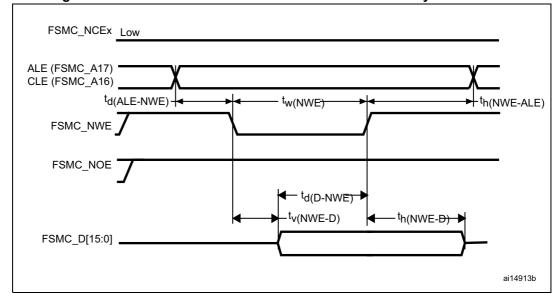


Figure 38. NAND controller waveforms for common memory write access

Table 40. Switching characteristics for NAND Flash read and write cycles⁽¹⁾

Symbol	Parameter	Min	Max	Unit
t _{d(D-NWE)} ⁽²⁾	FSMC_D[15:0] valid before FSMC_NWE high	5t _{HCLK} + 12	-	ns
t _{w(NOE)} ⁽²⁾	FSMC_NOE low width	4t _{HCLK} – 1.5	4t _{HCLK} + 1.5	ns
t _{su(D-NOE)} ⁽²⁾	FSMC_D[15:0] valid data before FSMC_NOE high	25	-	ns
t _{h(NOE-D)} (2)	FSMC_D[15:0] valid data after FSMC_NOE high	7	-	ns
t _{w(NWE)} ⁽²⁾	FSMC_NWE low width	4t _{HCLK} – 1	4t _{HCLK} + 2.5	ns
t _{v(NWE-D)} ⁽²⁾	FSMC_NWE low to FSMC_D[15:0] valid	-	0	ns
t _{h(NWE-D)} ⁽²⁾	FSMC_NWE high to FSMC_D[15:0] invalid	2t _{HCLK} + 4ns	-	ns
t _{d(ALE-NWE)} (3)	FSMC_ALE valid before FSMC_NWE low	-	3t _{HCLK} + 1.5	ns
t _{h(NWE-ALE)} (3)	FSMC_NWE high to FSMC_ALE invalid	3t _{HCLK} + 4.5	-	ns
t _{d(ALE-NOE)} (3)	FSMC_ALE valid before FSMC_NOE low	-	3t _{HCLK} + 2	ns
t _{h(NOE-ALE)} (3)	FSMC_NWE high to FSMC_ALE invalid	3t _{HCLK} + 4.5	-	ns

^{1.} C_L = 15 pF.

5.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

^{2.} Guaranteed by characterization results, not tested in production.

^{3.} Guaranteed by design, not tested in production.

Functional EMS (Electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports), the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 41*. They are based on the EMS levels and classes defined in application note AN1709.

Symbol	Parameter	Conditions	Level/Class
V _{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD} = 3.3$ V, LQFP144, $T_A = +25$ °C, $f_{HCLK} = 36$ MHz conforms to IEC 61000-4-2	2B
V _{EFTB}	Fast transient voltage burst limits to be applied through 100 pF on V _{DD} and V _{SS} pins to induce a functional disturbance	$V_{DD} = 3.3$ V, LQFP144, $T_A = +25$ °C, $f_{HCLK} = 36$ MHz conforms to IEC 61000-4-4	4A

Table 41. EMS characteristics

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and pre qualification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second. To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).



Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device is monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Max vs. [f_{HSE}/f_{HCLK}] Monitored Conditions **Parameter** Unit Symbol frequency band 8/36 MHz 0.1 MHz to 30 MHz 8 $V_{DD} = 3.3 \text{ V}, T_A = 25 ^{\circ}\text{C},$ 30 MHz to 130 MHz 27 dBµV LQFP144 package Peak level S_{EMI} compliant with 130 MHz to 1 GHz 26 IEC 61967-2 SAE EMI Level 4

Table 42. EMI characteristics

5.3.12 Absolute maximum ratings (electrical sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the JESD22-A114/JESD22-C101 standard.

	<u> </u>				
Symbol	Ratings	Conditions	Class	Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}		T _A = +25 °C, conforming to JESD22-A114	2	2000	V
V _{ESD(CDM)}	Electrostatic discharge voltage (charge device model)	T _A = +25 °C, conforming to JESD22-C101	III	500	V

Table 43. ESD absolute maximum ratings

Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78 IC latch-up standard.

Table 44. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T _A = +85 °C conforming to JESD78A	II level A



^{1.} Guaranteed by characterization results, not tested in production.

5.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DD} (for standard, 3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

Functional susceptibilty to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (>5 LSB TUE), out of spec current injection on adjacent pins or other functional failure (for example reset, oscillator frequency deviation).

The test results are given in Table 45

Table 45. I/O current injection susceptibility

		Functional s		
Symbol	Description	Negative injection	Positive injection	Unit
I _{INJ}	Injected current on OSC_IN32, OSC_OUT32, PA4, PA5, PC13	-0	+0	
	Injected current on all FT pins	-5	+0	mA
	Injected current on any other pin	-5	+5	

5.3.14 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in *Table 46* are derived from tests performed under the conditions summarized in *Table 10*. All I/Os are CMOS and TTL compliant.

Table 46. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V	Standard IO input low level voltage		-0.3		0.28*(V _{DD} -2 V)+0.8 V	٧
V _{IL}	IO FT ⁽¹⁾ input low level voltage	-	-0.3		0.32*(V _{DD} -2V)+0.75 V	٧
	Standard IO input high level voltage		0.41*(V _{DD} -2 V)+1.3 V		V _{DD} +0.3	٧
V_{IH}	IO FT ⁽¹⁾ input high level	V _{DD} > 2 V	0.42*(\/ 2.\/)+1.\/		5.5	V
	voltage	$V_{DD} \le 2 V$	- 0.42*(V _{DD} -2 V)+1 V		5.2	ľ
V_{hys}	Standard IO Schmitt trigger voltage hysteresis ⁽²⁾	-	200		-	mV
,0	IO FT Schmitt trigger voltage hysteresis ⁽²⁾		5% V _{DD} ⁽³⁾		-	mV
_	Input leakage current (4)	V _{SS} ≤ V _{IN} ≤ V _{DD} Standard I/Os	-		±1	
l _{lkg}	input leakage current V	V _{IN} = 5 V I/O FT	-		3	μA
R _{PU}	Weak pull-up equivalent resistor ⁽⁵⁾	$V_{IN} = V_{SS}$	30	40	50	kΩ
R _{PD}	Weak pull-down equivalent resistor ⁽⁵⁾	$V_{IN} = V_{DD}$	30	40	50	kΩ
C _{IO}	I/O pin capacitance	-	-	5	-	pF

FT = Five-volt tolerant. In order to sustain a voltage higher than V_{DD}+0.3 the internal pull-up/pull-down resistors must be disabled.

All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in *Figure 39* and *Figure 40* for standard I/Os, and in *Figure 41* and *Figure 42* for 5 V tolerant I/Os.

^{2.} Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization results, not tested in production.

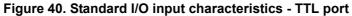
^{3.} With a minimum of 100 mV.

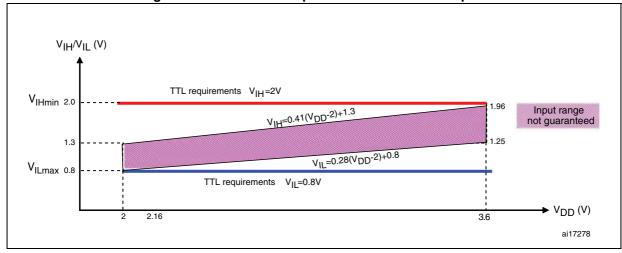
^{4.} Leakage could be higher than maximum value if negative current is injected on adjacent pins.

^{5.} Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimum (~10% order).

 $V_{IH}/V_{IL}(V)$ V_{IH}=0.41(V_{DD}-2)+1.3 CMOS standard requirement V_{IH}=0.65V_{DD} Input range V_{IHmin 1.3} 1.25 not guaranteed V_{ILmax 0.8} $V_{IL} = 0.28 (V_{DD}^{2}) + 0.8$ CMOS standard requirement V_{IL}=0.35V_{DD} **►** V_{DD} (V) 2 2.7 3.3 3.6 ai17277b

Figure 39. Standard I/O input characteristics - CMOS port





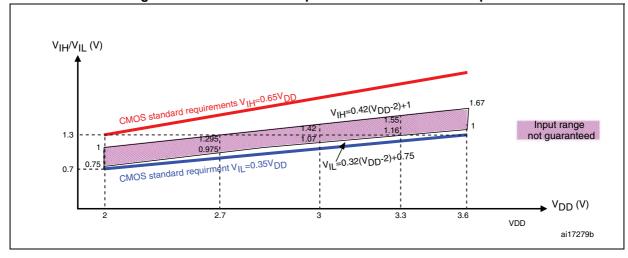
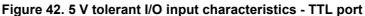
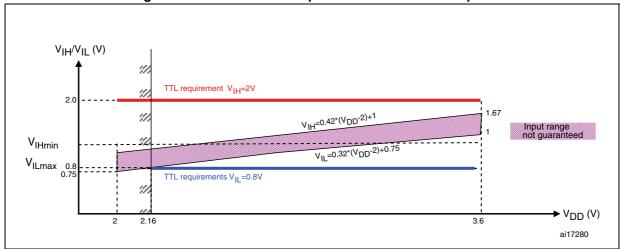


Figure 41. 5 V tolerant I/O input characteristics - CMOS port





Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to ± 8 mA, and sink or source up to ± 20 mA (with a relaxedV_{OL}/V_{OH}) except PC13, PC14 and PC15 which can sink or source up to ± 3 mA. When using the GPIOs PC13 to PC15 in output mode, the speed should not exceed 2 MHz with a maximum load of 30 pF.

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in *Section 5.2*:

- The sum of the currents sourced by all the I/Os on V_{DD}, plus the maximum Run consumption of the MCU sourced on V_{DD}, cannot exceed the absolute maximum rating I_{VDD} (see *Table 8*).
- The sum of the currents sunk by all the I/Os on V_{SS} plus the maximum Run consumption of the MCU sunk on V_{SS} cannot exceed the absolute maximum rating I_{VSS} (see *Table 8*).

Output voltage levels

Unless otherwise specified, the parameters given in *Table 47* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*. All I/Os are CMOS and TTL compliant.

Table 47. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
V _{OL} ⁽¹⁾	Output Low level voltage for an I/O pin when 8 pins are sunk at the same time	CMOS port ⁽²⁾ , I _{IO} = +8 mA,	-	0.4	V
V _{OH} ⁽³⁾	Output High level voltage for an I/O pin when 8 pins are sourced at the same time	$1_{\text{IO}} - +6 \text{ m/A},$ 2.7 V < V _{DD} < 3.6 V	V _{DD} -0.4	-	V
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin when 8 pins are sunk at the same time	TTL port ⁽²⁾ I _{IO} = +8 mA	-	0.4	V
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin when 8 pins are sourced at the same time	$1_{10} - 40 \text{ IIA}$ 2.7 V < V _{DD} < 3.6 V	2.4	-	v
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin when 8 pins are sunk at the same time	I _{IO} = +20 mA ⁽⁴⁾	-	1.3	V
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin when 8 pins are sourced at the same time	2.7 V < V _{DD} < 3.6 V	V _{DD} -1.3	-	V
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin when 8 pins are sunk at the same time	I _{IO} = +6 mA ⁽⁴⁾	-	0.4	V
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin when 8 pins are sourced at the same time	2 V < V _{DD} < 2.7 V	V _{DD} -0.4	-	V

^{1.} The $I_{|O}$ current sunk by the device must always respect the absolute maximum rating specified in *Table 8* and the sum of $I_{|O}$ (I/O ports and control pins) must not exceed I_{VSS} .



^{2.} TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

^{3.} The I_{IO} current sourced by the device must always respect the absolute maximum rating specified in *Table 8* and the sum of I_{IO} (I/O ports and control pins) must not exceed I_{VDD} .

^{4.} Guaranteed by characterization results, not tested in production.

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 43* and *Table 48*, respectively.

Unless otherwise specified, the parameters given in *Table 48* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*

Table 48. I/O AC characteristics⁽¹⁾

MODEx [1:0] bit value ⁽¹⁾	Symbol	Parameter	Conditions	Max	Unit
	f _{max(IO)out}	Maximum frequency ⁽²⁾	C _L = 50 pF, V _{DD} = 2 V to 3.6 V	2	MHz
10	t _{f(IO)out}	Output high to low level fall time	C = 50 pF V = 2 V to 2 6 V	125 ⁽³⁾	20
	t _{r(IO)out}	Output low to high level rise time	$C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 3.6 \text{ V}$	125 ⁽³⁾	ns
	f _{max(IO)out}	Maximum frequency ⁽²⁾	C _L = 50 pF, V _{DD} = 2 V to 3.6 V	10	MHz
01	t _{f(IO)out}	Output high to low level fall time	0 - 50 - 5 \ - 2 \ \ - 2 \ \ - 2 \ \ \ - 2 \ \ \ - 2 \ \ \ - 2 \ \ \ \	25 ⁽³⁾	
	t _{r(IO)out}	Output low to high level rise time	C _L = 50 pF, V _{DD} = 2 V to 3.6 V	25 ⁽³⁾	ns
	F _{max(IO)out}		C_L = 30 pF, V_{DD} = 2.7 V to 3.6 V	50	MHz
		Maximum Frequency ⁽²⁾	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	30	MHz
			$C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 2.7 \text{ V}$	20	MHz
			$C_L = 30 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	5 ⁽³⁾	
11	t _{f(IO)out}	Output high to low level fall time	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	8 ⁽³⁾	
			$C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 2.7 \text{ V}$	12 ⁽³⁾	
		Output low to high level rise	C _L = 30 pF, V _{DD} = 2.7 V to 3.6 V	5 ⁽³⁾	ns
	t _{r(IO)out}	time	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	['] 8 ⁽³⁾	
			C _L = 50 pF, V _{DD} = 2 V to 2.7 V	12 ⁽³⁾	
-	t _{EXTIpw}	Pulse width of external signals detected by the EXTI controller	-	10	ns

The I/O speed is configured using the MODEx[1:0] bits. Refer to the STM32F10xxx reference manual for a description of GPIO Port configuration register.

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^{2.} The maximum frequency is defined in Figure 43.

^{3.} Guaranteed by design, not tested in production.

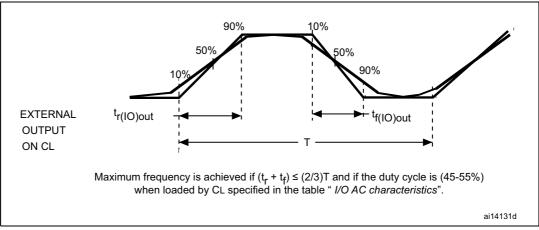


Figure 43. I/O AC characteristics definition

5.3.15 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R_{PU} (see *Table 46*).

Unless otherwise specified, the parameters given in *Table 49* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IL(NRST)} ⁽¹⁾	NRST Input low level voltage	-	-0.5	-	0.8	V
V _{IH(NRST)} ⁽¹⁾	NRST Input high level voltage	-	2	ı	V _{DD} +0.5	V
V _{hys(NRST)}	NRST Schmitt trigger voltage hysteresis	-	1	200	-	mV
R _{PU}	Weak pull-up equivalent resistor ⁽²⁾	$V_{IN} = V_{SS}$	30	40	50	kΩ
V _{F(NRST)} ⁽¹⁾	NRST Input filtered pulse	-	-	-	100	ns
V _{NF(NRST)} ⁽¹⁾	NRST Input not filtered pulse	-	300	-	-	ns

Table 49. NRST pin characteristics

^{1.} Guaranteed by design, not tested in production.

^{2.} The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance must be minimum (~10% order).

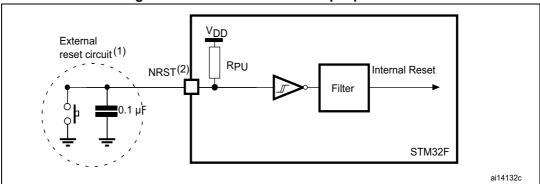


Figure 44. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- The user must ensure that the level on the NRST pin can go below the V_{IL(NRST)} max level specified in Table 49. Otherwise the reset will not be taken into account by the device.

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5.3.16 TIM timer characteristics

The parameters given in *Table 50* are guaranteed by design.

Refer to Section 5.3.13: I/O current injection characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Conditions Min **Symbol Parameter** Max Unit 1 t_{TIMxCLK} Timer resolution time t_{res(TIM)} $f_{TIMxCLK} = 36 MHz$ 27.8 ns 0 f_{TIMxCLK}/2 MHz Timer external clock f_{EXT} frequency on CH1 to CH4 $f_{TIMxCLK} = 36 MHz$ 0 18 MHz Res_{TIM} Timer resolution 16 bit 16-bit counter clock period 65536 1 t_{TIMxCLK} when internal clock is tCOUNTER f_{TIMxCLK} = 36 MHz | 0.0278 1820 selected μs 65536 × 65536 t_{TIMxCLK} t_{MAX_COUNT} | Maximum possible count $f_{TIMxCLK} = 36 \text{ MHz}$ 119.2 s

Table 50. TIMx⁽¹⁾ characteristics

5.3.17 Communications interfaces

I²C interface characteristics

The STM32F101xC, STM32F101xD and STM32F101xE access line I^2C interface meets the requirements of the standard I^2C communication protocol with the following restrictions: the I/O pins SDA and SCL are mapped to are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DD} is disabled, but is still present.

The I²C characteristics are described in *Table 51*. Refer also to *Section 5.3.13: I/O current injection characteristics* for more details on the input/output alternate function characteristics (SDA and SCL).

^{1.} TIMx is used as a general term to refer to the TIM1, TIM2, TIM3 and TIM4 timers.

Table 51. I²C characteristics

Symbol	Parameter	Standard mode I ² C ⁽¹⁾⁽²⁾		Fast mode I ² C ⁽¹⁾⁽²⁾		Unit
		Min	Max	Min	Max	
t _{w(SCLL)}	SCL clock low time	4.7	-	1.3	-	116
t _{w(SCLH)}	SCL clock high time	4.0	-	0.6	-	μs
t _{su(SDA)}	SDA setup time	250	-	100	-	
t _{h(SDA)}	SDA data hold time	-	3450 ⁽³⁾	-	900 ⁽³⁾	
$t_{r(SDA)} \ t_{r(SCL)}$	SDA and SCL rise time	-	1000	-	300	ns
$t_{f(SDA)} \ t_{f(SCL)}$	SDA and SCL fall time	-	300	-	300	
t _{h(STA)}	Start condition hold time	4.0	-	0.6	-	
t _{su(STA)}	Repeated Start condition setup time	4.7	-	0.6	-	μs
t _{su(STO)}	Stop condition setup time	4.0	-	0.6	-	μs
t _{w(STO:STA)}	Stop to Start condition time (bus free)	4.7	-	1.3	-	μs
C _b	Capacitive load for each bus line	-	400	-	400	pF
t _{SP}	Pulse width of the spikes that are suppressed by the analog filter for standard and fast mode	0	50 ⁽⁴⁾	0	50 ⁽⁴⁾	μs

^{1.} Guaranteed by design, not tested in production.



f_{PCLK1} must be at least 2 MHz to achieve standard mode I²C frequencies. It must be at least 4 MHz to achieve the fast mode I²C frequencies and it must be a multiple of 10 MHz in order to reach the I2C fast mode maximum clock speed of 400 kHz.

The maximum data hold time has only to be met if the interface does not stretch the low period of SCL signal.

^{4.} The minimum width of the spikes filtered by the analog filter is above $t_{SP}(max)$.

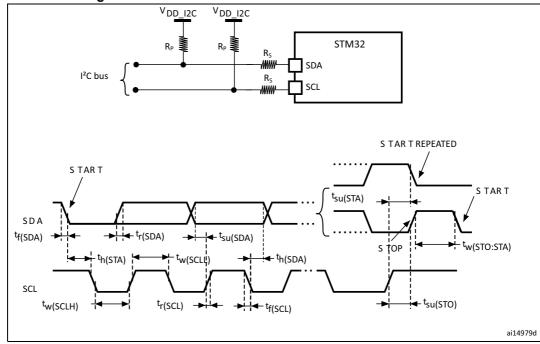


Figure 45. I²C bus AC waveforms and measurement circuit⁽¹⁾

- 1. Measurement points are done at CMOS levels: $\rm 0.3V_{DD}$ and $\rm 0.7V_{DD}$
- 1. R_S = series protection resistor.
- 2. R_P = external pull-up resistor.
- 3. V_{DD_I2C} is the I2C bus power supply.

Table 52. SCL frequency $(f_{PCLK1} = 36 \text{ MHz}, V_{DD} = V_{DD_I2C} = 3.3 \text{ V})^{(1)(2)}$

f _{SCL}	I2C_CCR value
(kHz)	$R_P = 4.7 \text{ k}\Omega$
400	0x801E
300	0x8028
200	0x803C
100	0x00B4
50	0x0168
20	0x0384

- 1. R_P = External pull-up resistance, f_{SCL} = I^2C speed.
- 2. For speeds around 200 kHz, the tolerance on the achieved speed is of $\pm 5\%$. For other speed ranges, the tolerance on the achieved speed $\pm 2\%$. These variations depend on the accuracy of the external components used to design the application.

SPI interface characteristics

Unless otherwise specified, the parameters given in *Table 53Table 54* are derived from tests performed under ambient temperature, f_{PCLKX} frequency and V_{DD} supply voltage conditions summarized in *Table 10*.

Refer to Section 5.3.13: I/O current injection characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Table 53. STM32F10xxx SPI characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
f _{SCK}	SDI alaak fraguanay	Master mode	-	10	MHz
1/t _{c(SCK)}	SPI clock frequency	Slave mode	-	10	- IVITZ
t _{r(SCK)}	SPI clock rise and fall time Capacitive load: C = 30 pF		-	8	
t _{su(NSS)} ⁽¹⁾	NSS setup time	Slave mode	4t _{PCLK}	-	
t _{h(NSS)} ⁽¹⁾	NSS hold time	Slave mode	73	-	
t _{w(SCKH)} (1) t _{w(SCKL)} (1)	SCK high and low time	Master mode, f _{PCLK} = 36 MHz, presc = 4	50	60	
		Master mode - SPI1	3	-	
$t_{su(MI)}^{(1)}$ $t_{su(SI)}^{(1)}$	Data input setup time	Master mode - SPI2	5	-	
Su(SI)		Slave mode	4	-	
. (1)		Master mode - SPI1	4	-	
t _{h(MI)} (1)	Data input hold time	Master mode - SPI2	6	-	
t _{h(SI)} ⁽¹⁾		Slave mode	5	-	ns
t _{a(SO)} ⁽¹⁾⁽²⁾	Data output access	Slave mode, f _{PCLK} = 36 MHz, presc = 4	0	55	
	time	Slave mode, f _{PCLK} = 20 MHz	-	4t _{PCLK}	
t _{dis(SO)} ⁽¹⁾⁽³⁾	Data output disable time	Slave mode	10	-	
t _{v(SO)} (1)	Data output valid time	Slave mode (after enable edge)	-	25	
t _{v(MO)} ⁽¹⁾	Data output valid time	Master mode (after enable edge)		6	
t _{h(SO)} ⁽¹⁾	Data output hold	Slave mode (after enable edge)	25	-	
t _{h(MO)} ⁽¹⁾	time	Master mode (after enable edge)	6	-	

 $^{1. \}quad \hbox{Guaranteed by characterization results, not tested in production.}$

^{2.} Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data

^{3.} Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z

Table 54. SPI characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
f _{SCK}	CDI alogic fraguency	Master mode	-	18	MHz
1/t _{c(SCK)}	SPI clock frequency	Slave mode	-	18	IVITZ
t _{r(SCK)}	SPI clock rise and fall time	Capacitive load: C = 30 pF	-	8	ns
DuCy(SCK)	SPI slave input clock duty cycle	Slave mode	30	70	%
t _{su(NSS)} ⁽¹⁾	NSS setup time	Slave mode	4t _{PCLK}	-	
t _{h(NSS)} ⁽¹⁾	NSS hold time	Slave mode	2t _{PCLK}	-	
t _{w(SCKH)} (1) t _{w(SCKL)} (1)	SCK high and low time	Master mode, f _{PCLK} = 36 MHz, presc = 4	50	60	
t _{su(MI)} (1)	Data input setup time	Master mode	5	-	
t _{su(MI)} (1) t _{su(SI)} (1)	Data input setup time	Slave mode	5	-	
t _{h(MI)} (1)	Data input hold time	Master mode	5	-	
t _{h(SI)} ⁽¹⁾	Data iriput riolu tirrie	Slave mode	4	-	ns
t _{a(SO)} ⁽¹⁾⁽²⁾	Data output access time	Slave mode, f _{PCLK} = 20 MHz	0	3t _{PCLK}	
t _{dis(SO)} (1)(3)	Data output disable time	Slave mode	2	10	
t _{v(SO)} (1)(1)	Data output valid time	Slave mode (after enable edge)	-	25	
t _{v(MO)} ⁽¹⁾⁽¹⁾	Data output valid time	Master mode (after enable edge)	-	5	
t _{h(SO)} ⁽¹⁾	Data output hold time	Slave mode (after enable edge)	15	-	
t _{h(MO)} ⁽¹⁾	Data output noid time	Master mode (after enable edge)	2	-	

^{1.} Guaranteed by characterization results not tested in production.

^{2.} Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data.

^{3.} Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z

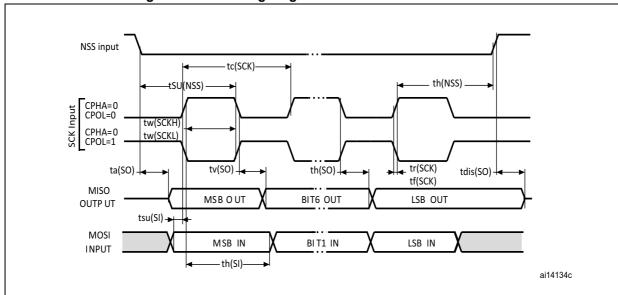
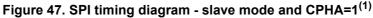
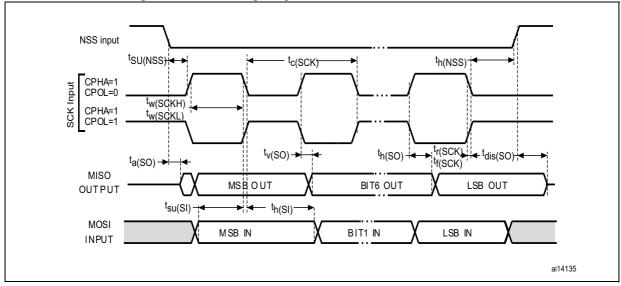


Figure 46. SPI timing diagram - slave mode and CPHA=0





1. Measurement points are done at CMOS levels: $0.3V_{DD}$ and $0.7V_{DD}$.

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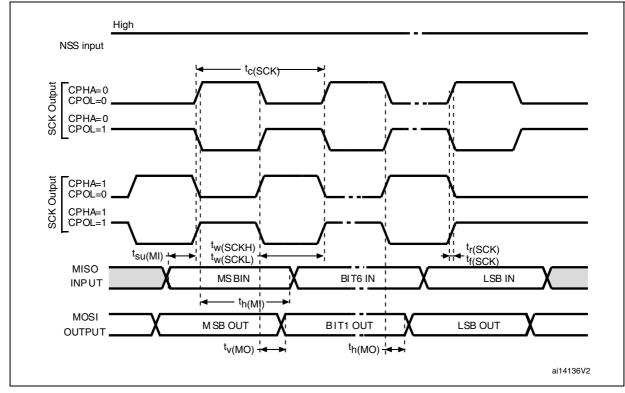


Figure 48. SPI timing diagram - master mode⁽¹⁾

1. Measurement points are done at CMOS levels: $0.3V_{DD}$ and $0.7V_{DD}$.

5.3.18 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 55* are valuesderived from tests performed under ambient temperature, f_{PCLK2} frequency and V_{DDA} supply voltage conditions summarized in *Table 10*.

Note: It is recommended to perform a calibration after each power-up.





The formula above (Equation 1) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. Here N = 12 (from 12-bit resolution).

Table 56. R_{AIN} max for $f_{ADC} = 14 \text{ MHz}^{(1)}$

T _s (cycles)	t _S (µs)	R_{AIN} max (kΩ)
1.5	0.11	0.4
7.5	0.54	5.9
13.5	0.96	11.4
28.5	2.04	25.2
41.5	2.96	37.2
55.5	3.96	50
71.5	5.11	NA
239.5	17.1	NA

^{1.} Guaranteed by design, not tested in production.

Table 57. ADC accuracy - limited test conditions⁽¹⁾⁽²⁾

Symbol	Parameter	Test conditions	Тур	Max ⁽³⁾	Unit
ET	Total unadjusted error	f _{PCLK2} = 28 MHz,	±1.3	±2	
EO	Offset error	f_{ADC} = 14 MHz, R_{AIN} < 10 kΩ, V_{DDA} = 3 V to 3.6 V, T_{A} = 25	±1	±1.5	
EG	Gain error	°C	±0.5	±1.5	LSB
ED	Differential linearity error	Measurements made after ADC calibration	±0.7	±1	
EL	Integral linearity error	V _{REF+} = V _{DDA}	±0.8	±1.5	

^{1.} ADC DC accuracy values are measured after internal calibration.



ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard analog
input pins should be avoided as this significantly reduces the accuracy of the conversion being performed
on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which
may potentially inject negative current.
Any positive injection current within the limits specified for I_{INJ(PIN)} and ΣI_{INJ(PIN)} in Section 5.3.13 does not
affect the ADC accuracy.

^{3.} Guaranteed by characterization results, not tested in production.

Symbol	Parameter	Test conditions	Тур	Max ⁽⁴⁾	Unit			
ET	Total unadjusted error	f - 20 MH-	±2	±5				
EO	Offset error	f _{PCLK2} = 28 MHz, f _{ADC} = 14 MHz, R _{AIN} < 10 kΩ,	±1.5	±2.5				
EG	Gain error	V _{DDA} = 2.4 V to 3.6 V	±1.5	±3	LSB			
ED	Differential linearity error	Measurements made after ADC calibration	±1	±2				
EL	Integral linearity error	7 LD O Gallibration	±1.5	±3				

Table 58. ADC accuracy⁽¹⁾ (2)(3)

- 1. ADC DC accuracy values are measured after internal calibration.
- 2. Better performance could be achieved in restricted V_{DD} , frequency, V_{REF} and temperature ranges.
- 3. ADC accuracy vs. negative injection current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current. Any positive injection current within the limits specified for I_{INJ(PIN)} and ΣI_{INJ(PIN)} in Section 5.3.13 does not affect the ADC accuracy.
- 4. Guaranteed by characterization results, not tested in production.

[1LSB IDEAL = depending on package) 4096 4096 (1) Example of an actual transfer curve 4095 (2) The ideal transfer curve 4094 (3) End point correlation line 4093 ET = Total unadjusted Error: maximum deviation between the actual and the ideal transfer curves. Eo = Offset Error: deviation between the first actual transition and the last actual one. 6 Eg = Gain Error: deviation between the last ideal 5 transition and the last actual one. ED = Differential Linearity Error: maximum deviation 4 between actual steps and the ideal one 3 EL = Integral Linearity Error: maximum deviation between any actual transition and the end-point correlation line. 0

Figure 49. ADC accuracy characteristics

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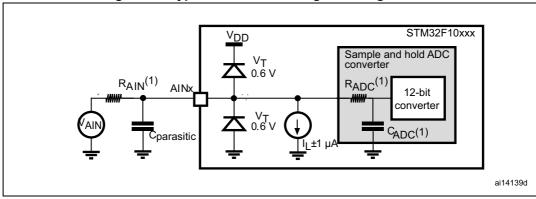


Figure 50. Typical connection diagram using the ADC

- Refer to *Table 55* for the values of R_{AIN}, R_{ADC} and C_{ADC}.
- $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high $C_{parasitic}$ value will downgrade conversion accuracy. To remedy this, f_{ADC} should be reduced.

General PCB design guidelines

Power supply decoupling should be performed as shown in Figure 51 or Figure 52, depending on whether $V_{\text{REF+}}$ is connected to V_{DDA} or not. The 10 nF capacitors should be ceramic (good quality). They should be placed them as close as possible to the chip.

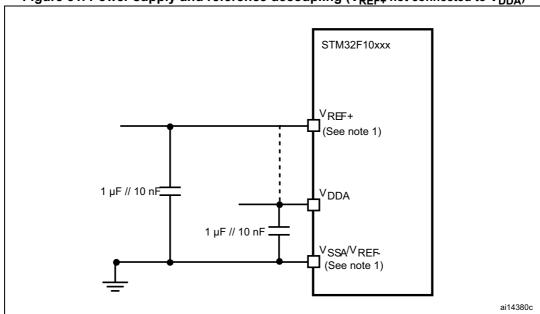


Figure 51. Power supply and reference decoupling (V_{REF+} not connected to V_{DDA})

1. V_{REF+} and V_{REF-} inputs are available only on 100-pin packages.



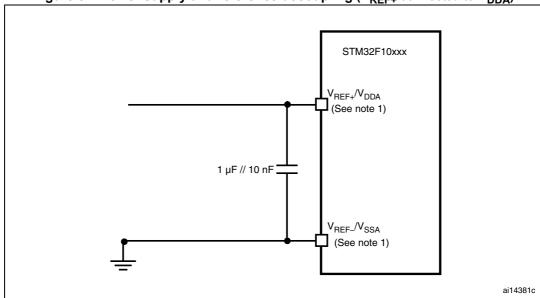


Figure 52. Power supply and reference decoupling (V_{REF+} connected to V_{DDA})

1. V_{REF+} and V_{REF-} inputs are available only on 100-pin packages.

5.3.19 DAC electrical specifications

Table 59. DAC characteristics

Symbol	Parameter	Min	Тур	Max ⁽¹⁾	Unit	Comments
V_{DDA}	Analog supply voltage	2.4	-	3.6	V	
V _{REF+}	Reference supply voltage	2.4	-	3.6	V	V _{REF+} must always be below V _{DDA}
V _{SSA}	Ground	0	-	0	V	
R _{LOAD} ⁽²⁾	Resistive load with buffer ON	5	-	-	kΩ	
R _O ⁽²⁾	Impedance output with buffer OFF	-	-	15	k()	When the buffer is OFF, the minimum resistive load between DAC_OUT and V_{SS} to have a 1% accuracy is 1.5 M Ω
C _{LOAD} ⁽²⁾	Capacitive load	-	-	50	рF	Maximum capacitive load at DAC_OUT pin (when the buffer is ON).
DAC_OUT min ⁽²⁾	Lower DAC_OUT voltage with buffer ON	0.2	-	-	· · ·	It gives the maximum output excursion of the DAC. It corresponds to 12-bit input code (0x0E0) to (0xF1C) at V _{RFF+} =
DAC_OUT max ⁽²⁾	Higher DAC_OUT voltage with buffer ON	-	-	V _{DDA} – 0.2	٧	3.6 V and (0x155) and (0xEAB) at V _{REF+} = 2.4 V.
DAC_OUT min ⁽²⁾	Lower DAC_OUT voltage with buffer OFF	-	0.5	-	mV	It gives the maximum output
DAC_OUT max ⁽²⁾	Higher DAC_OUT voltage with buffer OFF	-		V _{REF+} – 1LSB	٧	excursion of the DAC.

Table 59. DAC characteristics (continued)

Symbol	Parameter	Min	Тур	Max ⁽¹⁾	Unit	Comments
I _{DDVREF+}	DAC DC current consumption in quiescent mode (Standby mode)	-	-	220	μA	With no load, worst code (0xF1C) at V _{REF+} = 3.6 V in terms of DC consumption on the inputs.
	DAC DC current consumption	-	-	380	μA	With no load, middle code (0x800) on the inputs.
I _{DDA}	in quiescent mode ⁽³⁾	-	-	480	μA	With no load, worst code (0xF1C) at V _{REF+} = 3.6 V in terms of DC consumption on the inputs.
DNL ⁽¹⁾	Differential non linearity Difference between two	-	-	±0.5	LSB	Given for the DAC in 10-bit configuration.
	consecutive code-1LSB)	-	-	±2	LSB	Given for the DAC in 12-bit configuration.
(1)	Integral non linearity (difference between measured value at	-	-	±1	LSB	Given for the DAC in 10-bit configuration.
INL ⁽¹⁾	Code i and the value at Code i on a line drawn between Code 0 and last Code 1023)	-	-	±4	LSB	Given for the DAC in 12-bit configuration.
		-	-	±10	mV	
Offset ⁽¹⁾	Offset error (difference between measured value at Code (0x800) and the	-	-	±3	LSB	Given for the DAC in 10-bit at V _{REF+} = 3.6 V.
	ideal value = V _{REF+} /2)	-	-	±12	LSB	Given for the DAC in 12-bit at V _{REF+} = 3.6 V.
Gain error ⁽¹⁾	Gain error	-	-	±0.5	%	Given for the DAC in 12bit configuration.
^t settling ^v	Settling time (full scale: for a 10-bit input code transition between the lowest and the highest input codes when DAC_OUT reaches final value ±1LSB	-	3	4	μs	$C_{LOAD} \le 50 \text{ pF, } R_{LOAD} \ge 5 \text{ k}\Omega$
Update rate ⁽¹⁾	Max frequency for a correct DAC_OUT change when small variation in the input code (from code i to i+1LSB)	-	-	1	MS/s	$C_{LOAD} \le 50 \text{ pF, } R_{LOAD} \ge 5 \text{ k}\Omega$
t _{WAKEUP} (1)	Wakeup time from off state (Setting the ENx bit in the DAC Control register)	-	6.5	10	μs	$C_{LOAD} \le 50$ pF, $R_{LOAD} \ge 5$ k Ω input code between lowest and highest possible ones.
PSRR+ (2)	Power supply rejection ratio (to V _{DDA}) (static DC measurement	-	– 67	-40	dB	No R _{LOAD} , C _{LOAD} = 50 pF

^{1.} Guaranteed by characterization results, not tested in production.

^{3.} Quiescent mode refers to the state of the DAC when a steady value is kept on the output so that no dynamic consumption is involved.



^{2.} Guaranteed by design, not tested in production.

Buffer(1)

12-bit digital to analog converter

C L

ai17157V3

Figure 53. 12-bit buffered /non-buffered DAC

 The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC_CR register.

5.3.20 Temperature sensor characteristics

Table 60. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T _L ⁽¹⁾	V _{SENSE} linearity with temperature	-	±1	±2	°C
Avg_Slope ⁽¹⁾	Average slope	4.0	4.3	4.6	mV/°C
V ₂₅ ⁽¹⁾	Voltage at 25°C	1.34	1.43	1.52	V
t _{START} ⁽²⁾	Startup time	4	-	10	μs
T _{S_temp} (3)(2)	ADC sampling time when reading the temperature	-	-	17.1	μs

- 1. Guaranteed by characterization, not tested in production.
- 2. Guaranteed by design, not tested in production.
- 3. Shortest sampling time can be determined in the application by multiple iterations.

5//

Package information 6

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com. ECOPACK® is an ST trademark.

LQFP144 package information 6.1

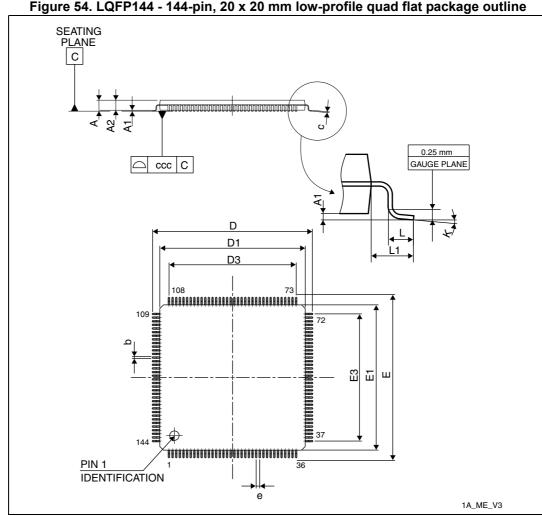


Figure 54. LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package outline

1. Drawing is not to scale.

Table 61. LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package mechanical data

Cumhal		millimeters			inches ⁽¹⁾	
Symbol	Min	Тур	Max	Min	Тур	Max
А	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	21.800	22.000	22.200	0.8583	0.8661	0.874
D1	19.800	20.000	20.200	0.7795	0.7874	0.7953
D3	-	17.500	-	-	0.689	-
E	21.800	22.000	22.200	0.8583	0.8661	0.874
E1	19.800	20.000	20.200	0.7795	0.7874	0.7953
E3	-	17.500	-		0.689	
е	-	0.500	-		0.0197	
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-		0.0394	
k	0°	3.5°	7°	0°	3.5°	7°
CCC	-	-	0.080	-	-	0.0031

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

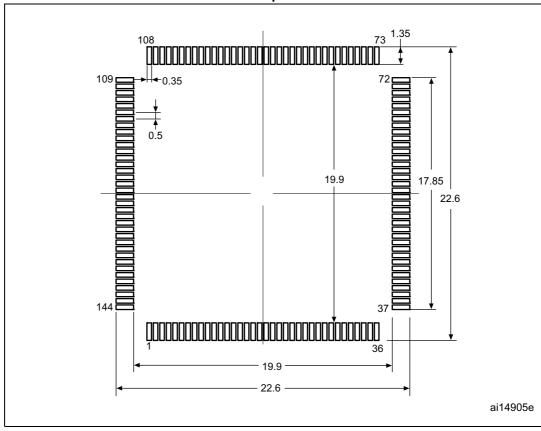


Figure 55. LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package footprint

1. Dimensions are expressed in millimeters.



Device marking for LQFP144

The following figure gives an example of topside marking and pin 1 position identifier location.

Product identification(1)

Revision code

R

STM32F101ZCTL

Date code

Y W W

MSv37277V1

Figure 56. LQFP144 marking (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering samples to run qualification activity.



6.2 LQFP100 package information

SEATING PLANE С 0.25 mm GAUGE PLANE ccc C D D1 의 대 PIN 1 **IDENTIFICATION** 1L_ME_V5

Figure 57. LQFP100 – 14 x 14 mm, 100-pin low-profile quad flat package outline

1. Drawing is not to scale.

Table 62. LQPF100 - 14 x 14 mm, 100-pin low-profile quad flat package mechanical data

Symbol		millimeters				
Symbol	Min	Тур	Max	Min	Тур	Max
А	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	15.800	16.000	16.200	0.622	0.6299	0.6378
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591
D3	-	12.000	-	-	0.4724	-
Е	15.800	16.000	16.200	0.622	0.6299	0.6378
E1	13.800	14.000	14.200	0.5433	0.5512	0.5591



inches⁽¹⁾ millimeters **Symbol** Тур Min Max Min Тур Max E3 12.000 0.4724 0.500 0.0197 е 0.600 0.0236 0.450 0.750 0.0177 0.0295 L 1.000 0.0394 L1 7° 7° k 0° 3.5° 0° 3.5° 0.080 0.0031 CCC

Table 62. LQPF100 – 14 x 14 mm, 100-pin low-profile quad flat package mechanical data (continued)

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

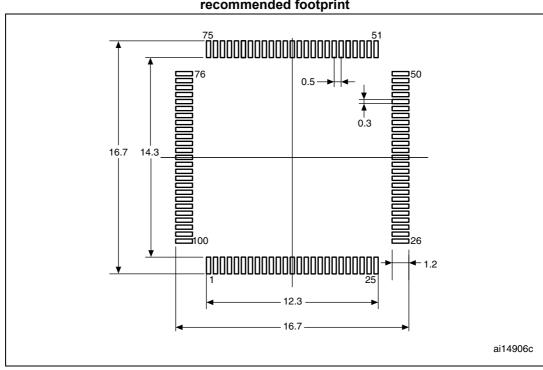


Figure 58. LQFP100 - 100-pin, 14 x 14 mm low-profile quad flat recommended footprint

1. Dimensions are in millimeters.

Device marking for LQFP100

The following figure gives an example of topside marking and pin 1 position identifier location.

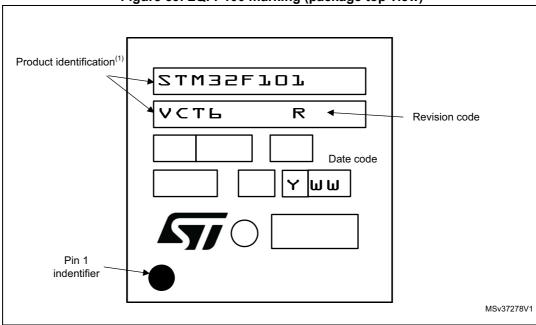


Figure 59. LQFP100 marking (package top view)

Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet
qualified and therefore not yet ready to be used in production and any consequences deriving from such
usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering
samples in production. ST Quality has to be contacted prior to any decision to use these Engineering
samples to run qualification activity.



6.3 LQFP64 information

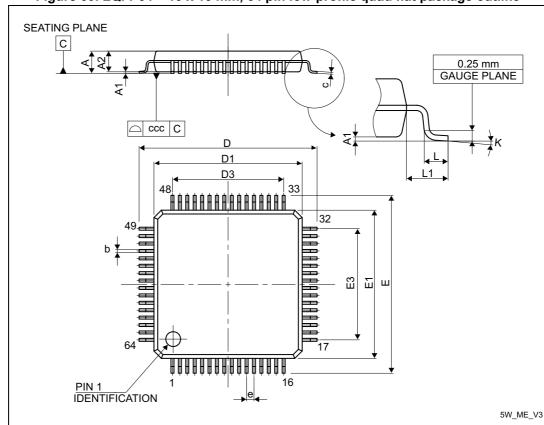


Figure 60. LQFP64 - 10 x 10 mm, 64 pin low-profile quad flat package outline

1. Drawing is not to scale.

Table 63. LQFP64 - 10 x 10 mm, 64 pin low-profile quad flat package mechanical data

0	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-
E3	-	7.500	-	-	0.2953	-

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Table 63. LQFP64 - 10 x 10 mm, 64 pin low-profile quad flat package mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Тур	Max	Min	Тур	Max
е	-	0.500	-	-	0.0197	-
θ	0°	3.5°	7°	0°	3.5°	7°
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
ccc	-	-	0.080	-	-	0.0031

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

48 12.7 10.3 10.3 10.3 10.3 10.3 10.3 10.3 10.3 10.3 10.3 10.3

1. Dimensions are in millimeters.

ai14909c

Device marking for LQFP64

The following figure gives an example of topside marking and pin 1 position identifier location.

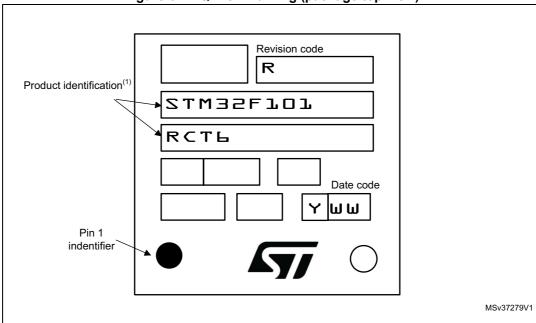


Figure 62. LQFP64 marking (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering samples to run qualification activity.



6.4 Thermal characteristics

The maximum chip junction temperature (T_Jmax) must never exceed the values given in *Table 10: General operating conditions on page 40*.

The maximum chip-junction temperature, T_J max, in degrees Celsius, may be calculated using the following equation:

$$T_J \max = T_A \max + (P_D \max \times \Theta_{JA})$$

Where:

- T_A max is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in °C/W,
- P_D max is the sum of P_{INT} max and $P_{I/O}$ max (P_D max = P_{INT} max + $P_{I/O}$ max),
- P_{INT} max is the product of I_{DD} and V_{DD}, expressed in Watts. This is the maximum chip internal power.

P_{I/O} max represents the maximum power dissipation on output pins where:

$$P_{I/O} \max = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

The contract of the contract o				
Symbol	Parameter	Value	Unit	
	Thermal resistance junction-ambient LQFP144 - 20 x 20 mm / 0.5 mm pitch	30		
Θ_{JA}	Thermal resistance junction-ambient LQFP100 - 14 x 14 mm / 0.5 mm pitch	46	°C/W	
	Thermal resistance junction-ambient LQFP64 - 10 x 10 mm / 0.5 mm pitch	45		

Table 64. Package thermal characteristics

6.4.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air), available from www.jedec.org.

6.4.2 Evaluating the maximum junction temperature for an application

When ordering the microcontroller, the temperature range is specified in the ordering information scheme shown in *Table 65: Ordering information scheme*.

Each temperature range suffix corresponds to a specific guaranteed ambient temperature at maximum dissipation and, to a specific maximum junction temperature. Here, only temperature range 6 is available (–40 to 85 °C).

The following example shows how to calculate the temperature range needed for a given application, making it possible to check whether the required temperature range is compatible with the STM32F10xxx junction temperature range.

Example: High-performance application

Assuming the following application conditions:

Maximum ambient temperature T_{Amax} = 82 °C (measured according to JESD51-2), I_{DDmax} = 50 mA, V_{DD} = 3.5 V, maximum 20 I/Os used at the same time in output at low level with I_{OL} = 8 mA, V_{OL} = 0.4 V and maximum 8 I/Os used at the same time in output mode at low level with I_{OL} = 20 mA, V_{OL} = 1.3 V

 $P_{INTmax} = 50 \text{ mA} \times 3.5 \text{ V} = 175 \text{ mW}$

 $P_{IOmax} = 20 \times 8 \text{ mA} \times 0.4 \text{ V} + 8 \times 20 \text{ mA} \times 1.3 \text{ V} = 272 \text{ mW}$

This gives: P_{INTmax} = 175 mW and P_{IOmax} = 272 mW

 $P_{Dmax} = 175 + 272 = 447 \text{ mW}$

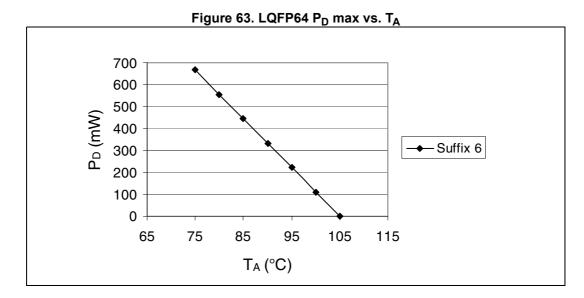
Thus: P_{Dmax} = 447 mW

Using the values obtained in *Table 65* T_{Jmax} is calculated as follows:

For LQFP64, 45 °C/W

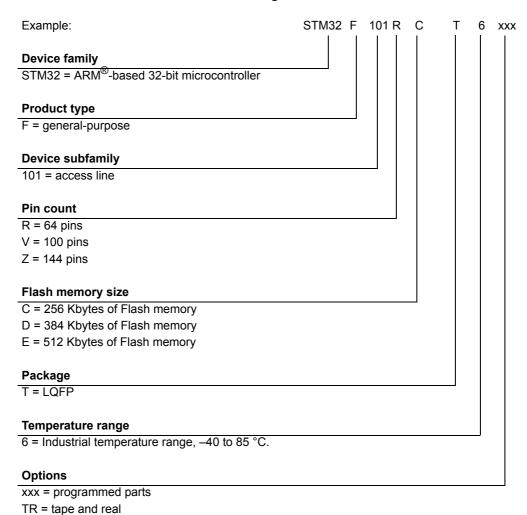
 T_{Jmax} = 82 °C + (45 °C/W × 447 mW) = 82 °C + 20.1 °C = 102.1 °C

This is within the junction temperature range of the STM32F10xxx ($-40 < T_J < 105$ °C).



7 Part numbering

Table 65. Ordering information scheme



For a list of available options (speed, package, etc..) or for further information on any aspect of this device, please contact your nearest ST sales office.

8 Revision history

Table 66. Document revision history

Date	Revision	Changes	
07-Apr-2008	1	Initial release.	
22-May-2008	2	Document status promoted from Target Specification to Preliminary Data. Section 1: Introduction and Section 2.2: Full compatibility throughout the family modified. Small text changes. Note 1 added in Table 2: STM32F101xC, STM32F101xD and STM32F101xE features and peripheral counts on page 11. LQPF100/BGA100 column added to Table 6: FSMC pin definition on page 32. Values added to Maximum current consumption on page 42 (see Table 14, Table 15, Table 16 and Table 17). Values added to Typical current consumption on page 48 (see Table 18, Table 19 and Table 20 and see Figure 11, Figure 12, Figure 14, Figure 15 and Figure 16), Table 19: Typical current consumption in Standby mode removed. Figure 55: LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package footprint on page 105 corrected. Equation 1 corrected. Section 6.4.2: Evaluating the maximum junction temperature for an application on page 114 added.	

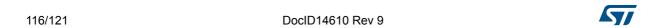


Table 66. Document revision history (continued)

Date	Revision	Changes
21-Jul-2008	3	Document status promoted from Preliminary Data to full datasheet. FSMC (flexible static memory controller) on page 15 modified. Power supply supervisor on page 17 modified and VDDA added to Table 10: General operating conditions on page 40. Table notes revised in Section 5: Electrical characteristics. Capacitance modified in Figure 9: Power supply scheme on page 37. Table 52: SCL frequency (fPCLK1= 36 MHz, VDD= VDD_IZC= 3.3 V) updated. Table 54: SPI characteristics modified, th(NSS) modified in Figure 46: SPI timing diagram - slave mode and CPHA=0 on page 94. Minimum SDA and SCL fall time value for Fast mode removed from Table 51: PC characteristics on page 90, note 1 modified. IDD_VBAT values added to Table 17: Typical and maximum current consumptions in Stop and Standby modes on page 45. Table 30: Flash memory endurance and data retention on page 59 updated. IHCLK corrected in Table 41: EMS characteristics. Lau(NSS) modified in Table 58: ADC accuracy on page 98, fPCLK2 corrected in Table 57: ADC accuracy - limited test conditions and Table 58: ADC accuracy. Figure 50: Typical connection diagram using the ADC on page 99 and note below corrected. Typical TS_temp value removed from Table 60: TS characteristics on page 102. Section 6.1: LQFP144 package information on page 103 updated, Small text changes.
12-Dec-2008	4	General-purpose timers (TIMx) on page 19 updated, Table 3: STM32F101xx family updated to show the low-density family, Table 4: Timer feature comparison added Figure 1: STM32F101xC, STM32F101xD and STM32F101xE access line block diagram updated. Note 9 added, main function after reset and Note 5 updated in Table 5: STM32F101xC/STM32F101xD/STM32F101xE pin definitions. Note 2 modified below Table 7: Voltage characteristics on page 38, $ \Delta V_{DDx} $ min and $ \Delta V_{DDx} $ min removed. Measurement conditions specified in Section 5.3.5: Supply current characteristics on page 42. General input/output characteristics on page 82 modified. Max values at T_A = 85 °C updated in Table 17: Typical and maximum current consumptions in Stop and Standby modes on page 45. Section 5.3.10: FSMC characteristics on page 59 revised. Values added to Table 42: EMI characteristics on page 80. I_{VREF} added to Table 55: ADC characteristics on page 96. Table 64: Package thermal characteristics on page 113 updated, Small text changes.



Table 66. Document revision history (continued)

Date	Revision	Changes	
		I/O information clarified on cover page, Number of ADC peripherals corrected in Table 2: STM32F101xC, STM32F101xD and STM32F101xE features and peripheral counts.	
		In Table 5: STM32F101xC/STM32F101xD/STM32F101xE pin definitions:	
		I/O level of pins PF11, PF12, PF13, PF14, PF15, G0, G1 and G15 updated	
	5	 PB4, PB13, PB14, PB15, PB3/TRACESWO moved from Default column to Remap column. 	
		PG14 pin description modified in <i>Table 6: FSMC pin definition</i> , <i>Figure 6: Memory map on page 35</i> modified.	
		Note modified in Table 14: Maximum current consumption in Run mode, code with data processing running from Flash and Table 16: Maximum current consumption in Sleep mode, code running from Flash or RAM.	
		Figure 14, Figure 15 and Figure 16 show typical curves (titles changed).	
20 Mar 2000		Table 21: High-speed external user clock characteristics and Table 22 Low-speed user external clock characteristics modified.	
30-Mar-2009		ACC _{HSI} max values modified in <i>Table 25: HSI oscillator characteristic</i> : FSMC configuration modified for <i>Asynchronous waveforms and timings</i> . Notes modified below <i>Figure 21: Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms</i> and <i>Figure 22: Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms</i> two values modified in <i>Table 31: Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings</i> and <i>Table 34: Asynchronous multiplexed NOR/PSRAM write timings</i> . tho timings to the value of the value o	
		In Table 36: Synchronous multiplexed PSRAM write timings and Table 38: Synchronous non-multiplexed PSRAM write timings:	
		t _{v(Data-CLK)} renamed as t _{d(CLKL-Data)} t _{d(CLKL-Data)} min value removed and max value added	
		- t _{h(CLKL-Data)} / t _{h(CLKL-ADV)} removed Figure 25: Synchronous multiplexed NOR/PSRAM read timings. Figure 26: Synchronous multiplexed PSRAM write timings and Figure 28: Synchronous non-multiplexed PSRAM write timings modified, Small text changes.	

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Table 66. Document revision history (continued)

Date	Revision	Changes	
21-Jul-2009	6	Figure 1: STM32F101xC, STM32F101xD and STM32F101xE access line block diagram modified. Note 5 updated and Note 4 added in Table 5: STM32F101xC/STM32F101xD/STM32F101xE pin definitions. V_RERINT and T_Coeff added to Table 13: Embedded internal reference voltage. f_HSE_ext min modified in Table 21: High-speed external user clock characteristics. Table 23: HSE 4-16 MHz oscillator characteristics modified. Note 1 modified below Figure 19: Typical application with an 8 MHz crystal. Figure 44: Recommended NRST pin protection modified. C_L1 and C_L2 replaced by C in Table 23: HSE 4-16 MHz oscillator characteristics and Table 24: LSE oscillator characteristics (f_LSE = 32.768 kHz), notes modified and moved below the tables. Table 25: HSI oscillator characteristics modified. Conditions removed from Table 27: Low-power mode wakeup timings. Jitter added to Table 28: PLL characteristics. In Table 31: Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings: th(BL_NOE) and th(A_NOE) modified. In Table 32: Asynchronous non-multiplexed NOR/PSRAM read timings: th(A_NWE) and th(Data_NWE) modified. In Table 33: Asynchronous multiplexed NOR/PSRAM read timings: th(A_NWE) and th(A_NOE) modified. In Table 34: Asynchronous multiplexed NOR/PSRAM read timings: th(A_NWE) modified. In Table 35: Synchronous multiplexed NOR/PSRAM read timings: th(CLKH-NWAITV) modified. In Table 36: Synchronous multiplexed NOR/PSRAM read timings: th(CLKH-NWAITV) modified. In Table 40: Switching characteristics for NAND Flash read and write cycles: th(NOE-D) modified. Table 54: SPI characteristics modified in Table 55: ADC characteristics. Rain max values modified in Table 56: Rain max for the ADC characteristics. Rain max values modified in Table 56: Rain max for the ADC characteristics modified. Figure 53: 12-bit buffered /non-buffered	
24-Sep-2009	7	Number of DACs corrected in <i>Table 3: STM32F101xx family</i> . I _{DD_VBAT} updated in <i>Table 17: Typical and maximum current consumptions in Stop and Standby modes</i> . Figure 13: Typical current consumption on V _{BAT} with RTC on vs. temperature at different V _{BAT} values added. IEC 1000 standard updated to IEC 61000 and SAE J1752/3 updated to IEC 61967-2 in <i>Section : on page 78</i> . <i>Table 59: DAC characteristics</i> modified. Small text changes.	



Table 66. Document revision history (continued)

Date	Revision	Changes
19-Apr-2011	8	Updated footnotes below Table 7: Voltage characteristics on page 38 and Table 8: Current characteristics on page 39 Updated tw min in Table 21: High-speed external user clock characteristics on page 51 Updated startup time in Table 24: LSE oscillator characteristics (f _{LSE} = 32.768 kHz) on page 55 Updated Table 31: Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings on page 60 Updated FSMC sync data latency in Figure 25 thru Figure 28 Updated Figure 38: NAND controller waveforms for common memory write access and Table 40: Switching characteristics for NAND Flash read and write cycles on page 78 Updated Figure 44: Recommended NRST pin protection Added Section 5.3.13: I/O current injection characteristics Updated Nection 5.3.13: I/O current injection characteristics Updated note 2 in Table 51: I ² C characteristics on page 90
15-May-2015	9	Added OSC_IN/OSC_OUT remap functions and updated PD0/PD1 in Table 5: STM32F101xC/STM32F101xD/STM32F101xE pin definitions. Modified Section 2.3.21: GPIOs (general-purpose inputs/outputs) on page 20. Updated notes related to parameters not tested in production in the whole document. Updated Table 20: Peripheral current consumption on page 50. Updated CDM standard and values in Section: Electrostatic discharge (ESD). Modified Section: Output driving current on page 84. Updated Figure 43: I/O AC characteristics definition. Updated conditions related to Section: PC interface characteristics. Modified Table 51: PC characteristics on page 90, updated Figure 45: PC bus AC waveforms and measurement circuit(1) and VDD/VDD_12C conditions in Table 52: SCL frequency (fPCLK1= 36 MHz, VDD = VDD_12C= 3.3 V) on page 91. Modified Figure 48: SPI timing diagram - master mode(1) on page 95. Modified note 3 in Table 58: ADC accuracy on page 98. Updated IDDA definition in Table 59: DAC characteristics on page 100 and removed comment related to the offset parameter for ±10 mV. Corrected "CLKL-NOEL" in Section 5.3.10: FSMC characteristics on page 59. Updated Section 6.1: LQFP144 package information on page 103 and added Section: Device marking for LQFP144 on page 106. Updated Section 6.2: LQFP100 package information on page 107 and added Section: Device marking for LQFP100 on page 110.

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