

1.3.3. System realisation

Testsignal: Dirac-Impuls $\delta(t)$

→ enthält alle Frequenzen mit Amplitude = 1

$$\delta(t) \longleftrightarrow X(f) = 1$$

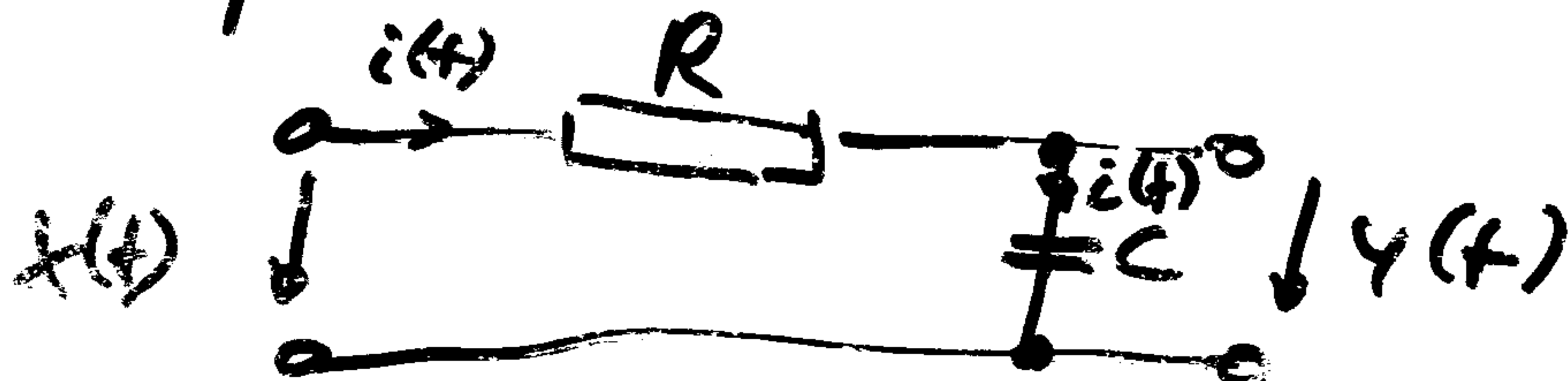
Für lineare Systeme gilt:

$$y(t) = \sum_{n=0}^{\infty} x_n(t), \quad x_n(t) = a_n \cdot \sin(\omega_n t)$$

3.1

1.3.1. Impulssantwort

Beispiel: RC-Glied



DGL = ?

Kirchhoff: $x(t) = i(t) \cdot R + y(t)$

$$i(t) = C \cdot \frac{dy}{dt}$$

$$\Rightarrow x(t) = C \cdot \frac{dy}{dt} \cdot R + y(t)$$

$$= RC \dot{y}(t) + y(t)$$

$$\Rightarrow \dot{y}(t) + \frac{1}{RC} y(t) = \frac{1}{RC} x(t)$$

$$\dot{y}(t) + \frac{1}{T} y(t) = \frac{1}{T} x(t), \quad T = RC$$

Lösung:
$$y(t) = e^{-\int \frac{1}{T} dt} \left(\int \frac{1}{T} x(t) e^{\int \frac{1}{T} dt} dt + C \right)$$

$$= e^{-t/T} \left(\frac{1}{T} \int x(t) e^{t/T} dt + C \right)$$

$$y(0) = 0 \quad \underline{\text{kausal!}} \quad (\text{für } x(t) = 0 \text{ für } t \leq 0)$$

$$\Rightarrow \underline{c=0}$$

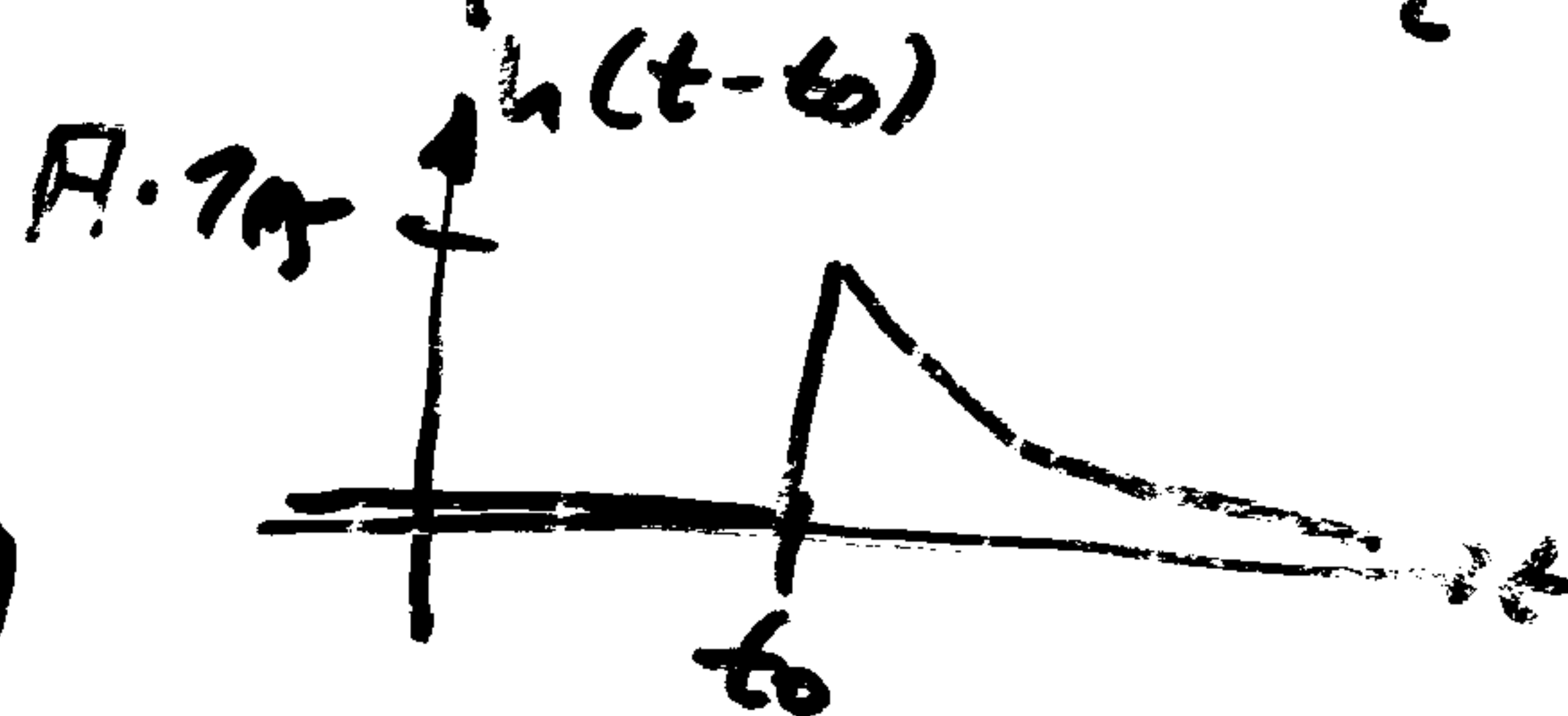
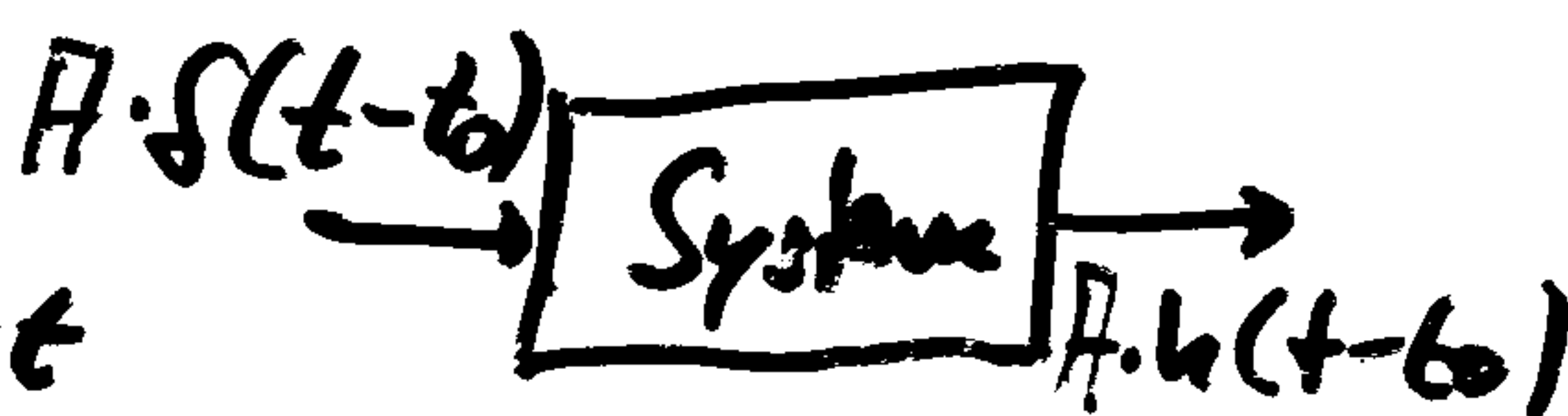
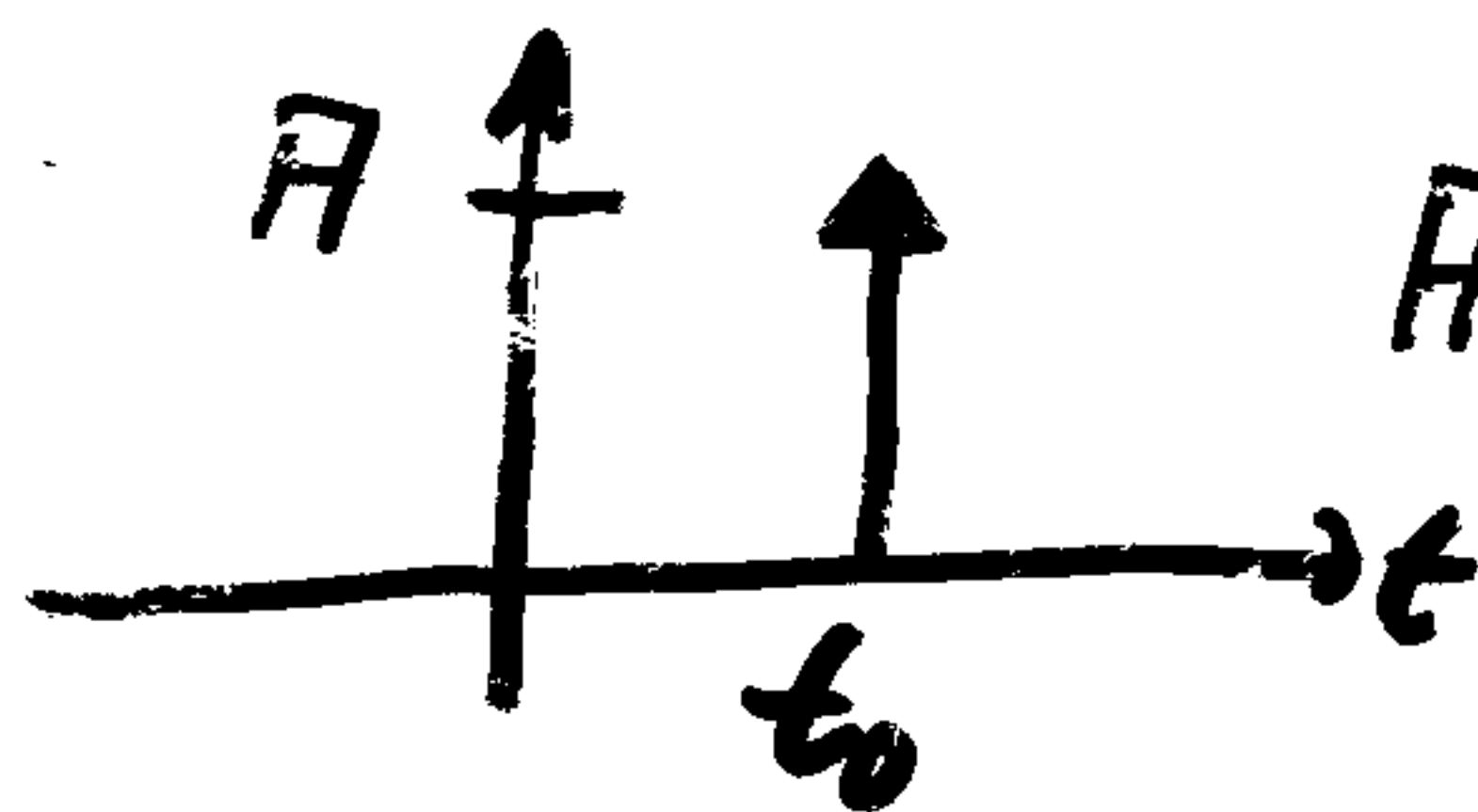
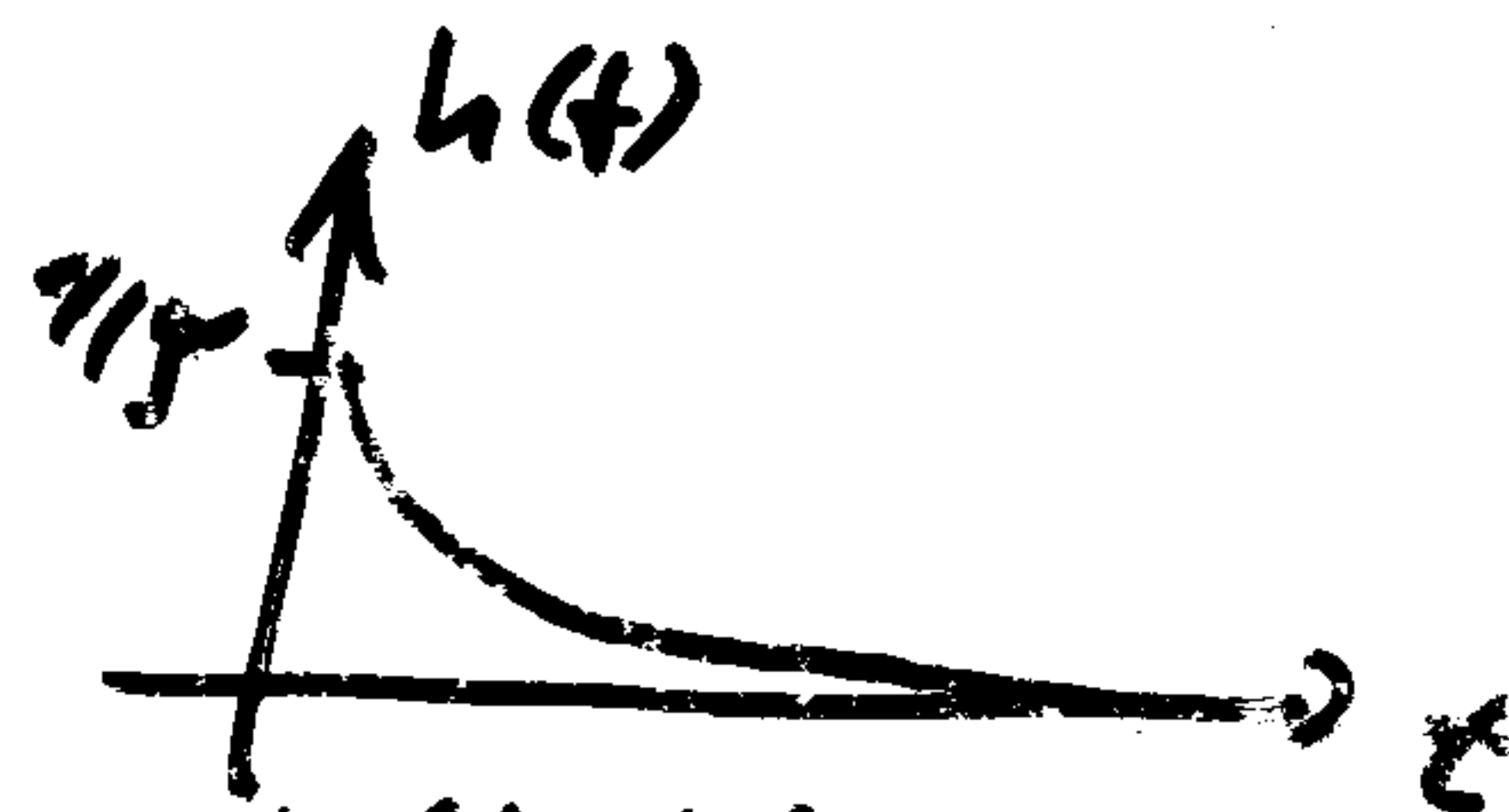
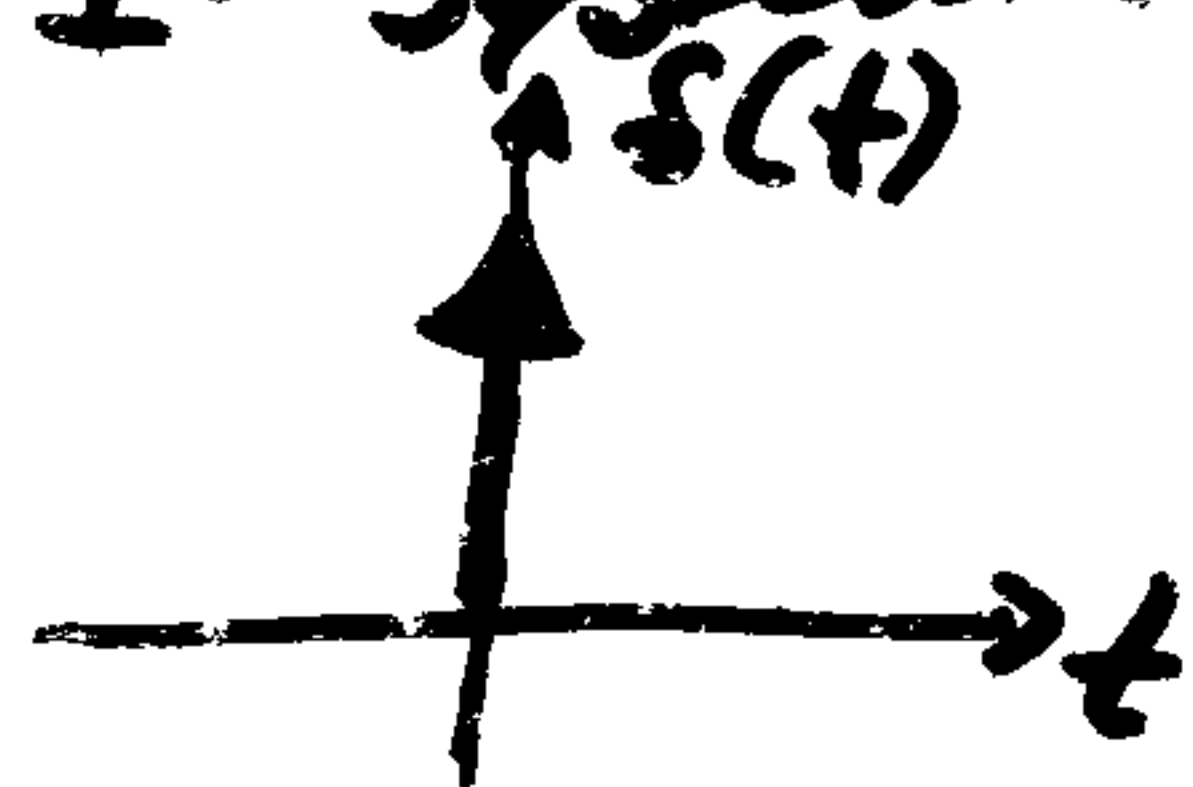
$$\Rightarrow y(t) = \frac{1}{\tau} e^{-t/\tau} \int x(t) e^{t/\tau} dt$$

Für $x(t) = \delta(t)$ Impuls-Eregung
 $y(t) = \underline{h(t)}$ Impulsantwort

$$\Rightarrow h(t) = \frac{1}{\tau} e^{-t/\tau} \int \delta(t) e^{t/\tau} dt$$

= 1

LTI-System:

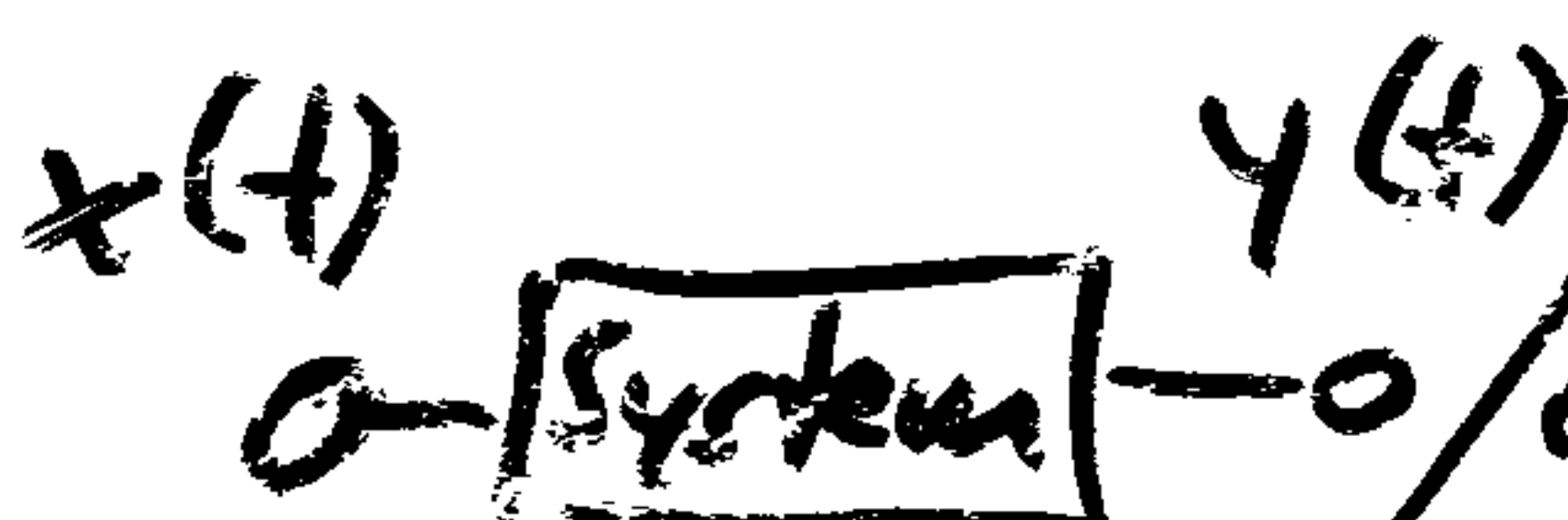


$$k_1 \delta(t-t_1) + k_2 \delta(t-t_2) \rightarrow \boxed{} \rightarrow k_1 h(t-t_1) + k_2 h(t-t_2)$$

$$\text{allg.: } \sum_{i=-\infty}^{\infty} k_i \delta(t-t_i) \rightarrow \boxed{} \rightarrow \sum_{i=-\infty}^{\infty} k_i h(t-t_i)$$

↓
Grenzübergang:

$$x(t) = \int_{-\infty}^{\infty} x(\tau) \cdot \delta(t-\tau) d\tau$$



Faltung

$$\int_{-\infty}^{\infty} x(\tau) \cdot h(t-\tau) d\tau = y(t)$$

Faltung: 3.2

$$y(t) = \int_{-\infty}^{\infty} x(\tau) \cdot h(t-\tau) d\tau$$

$$= x(t) * h(t)$$

↑
Faltung(soperation)
(integral.)

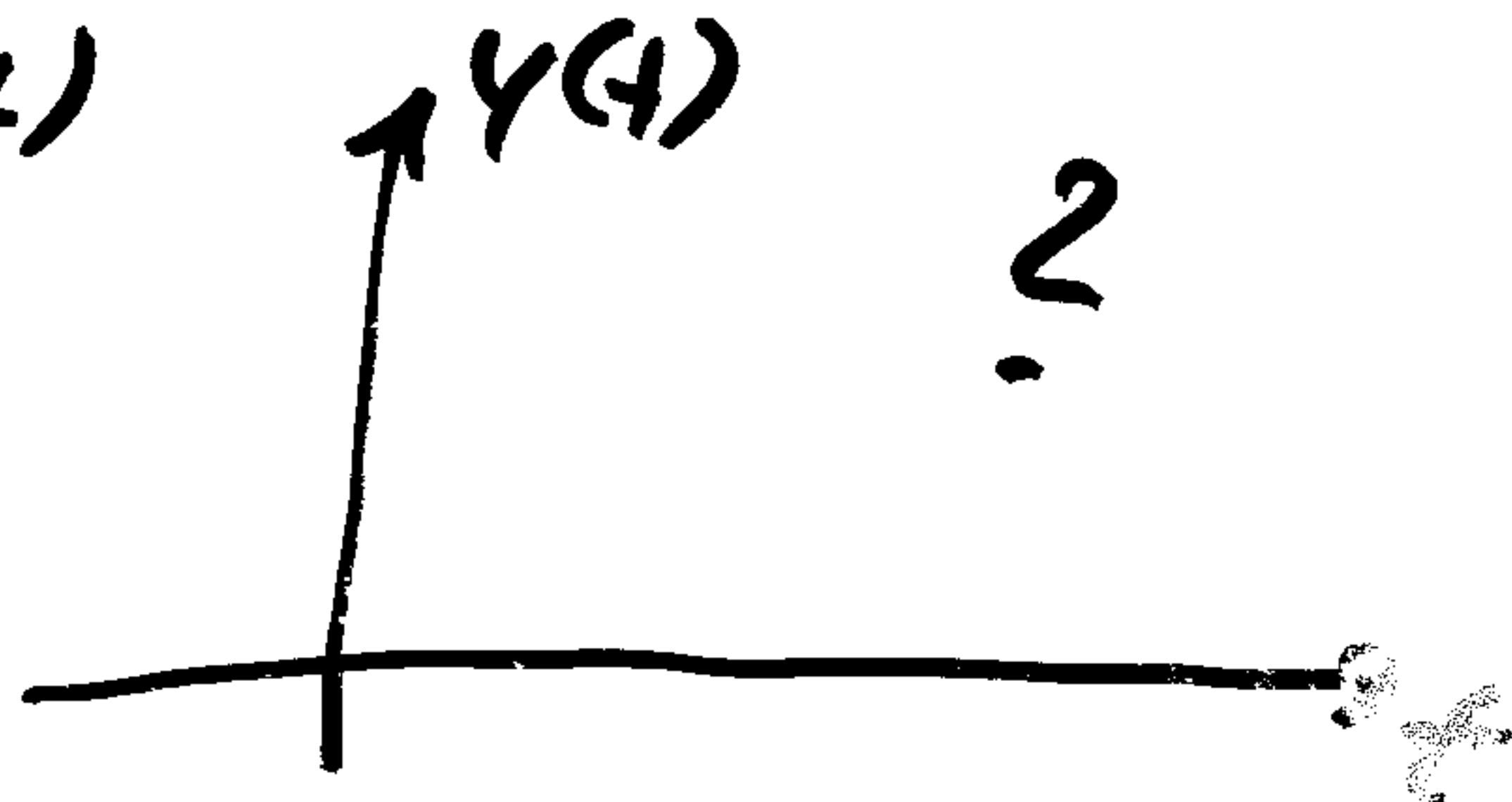
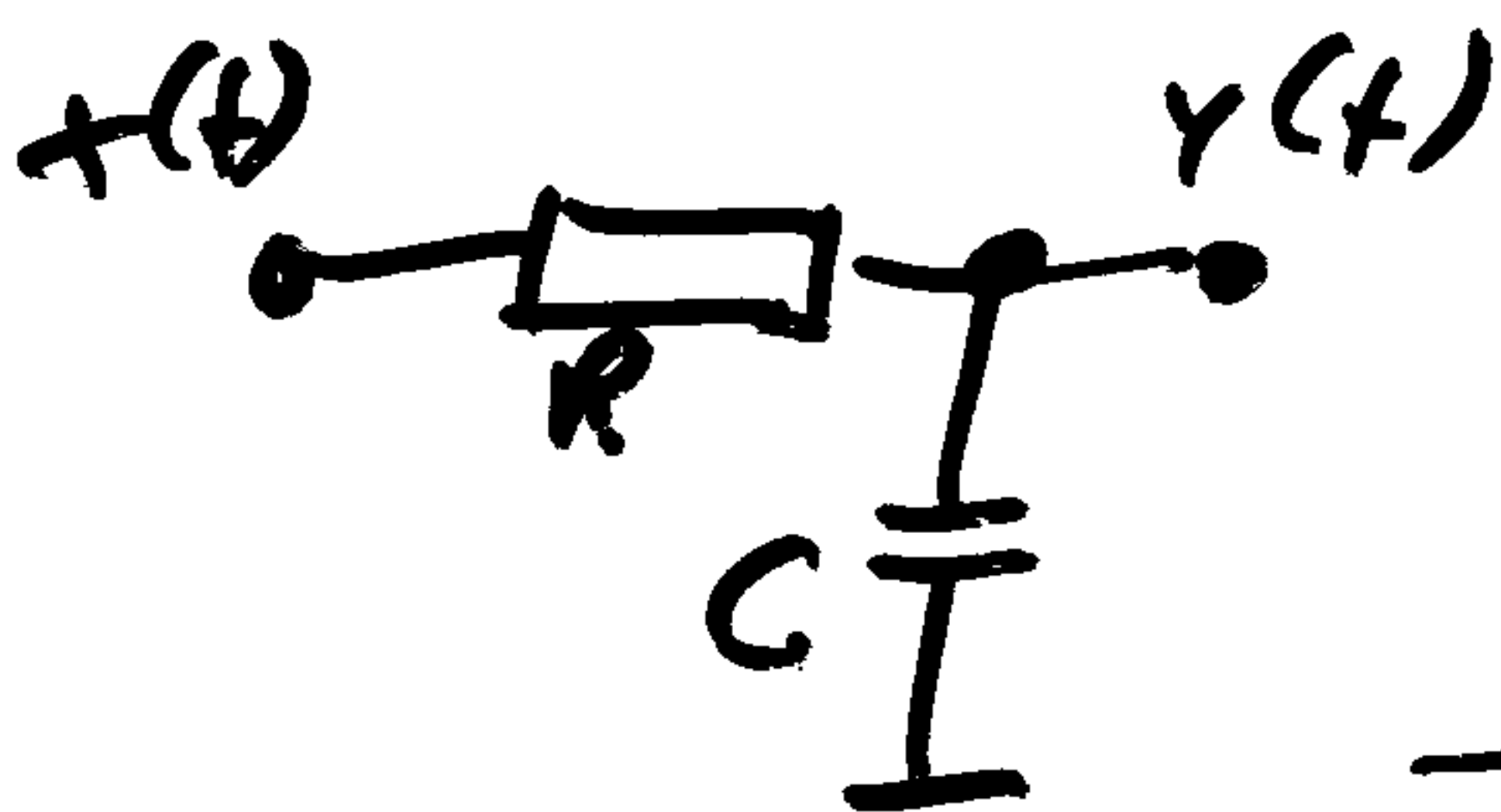
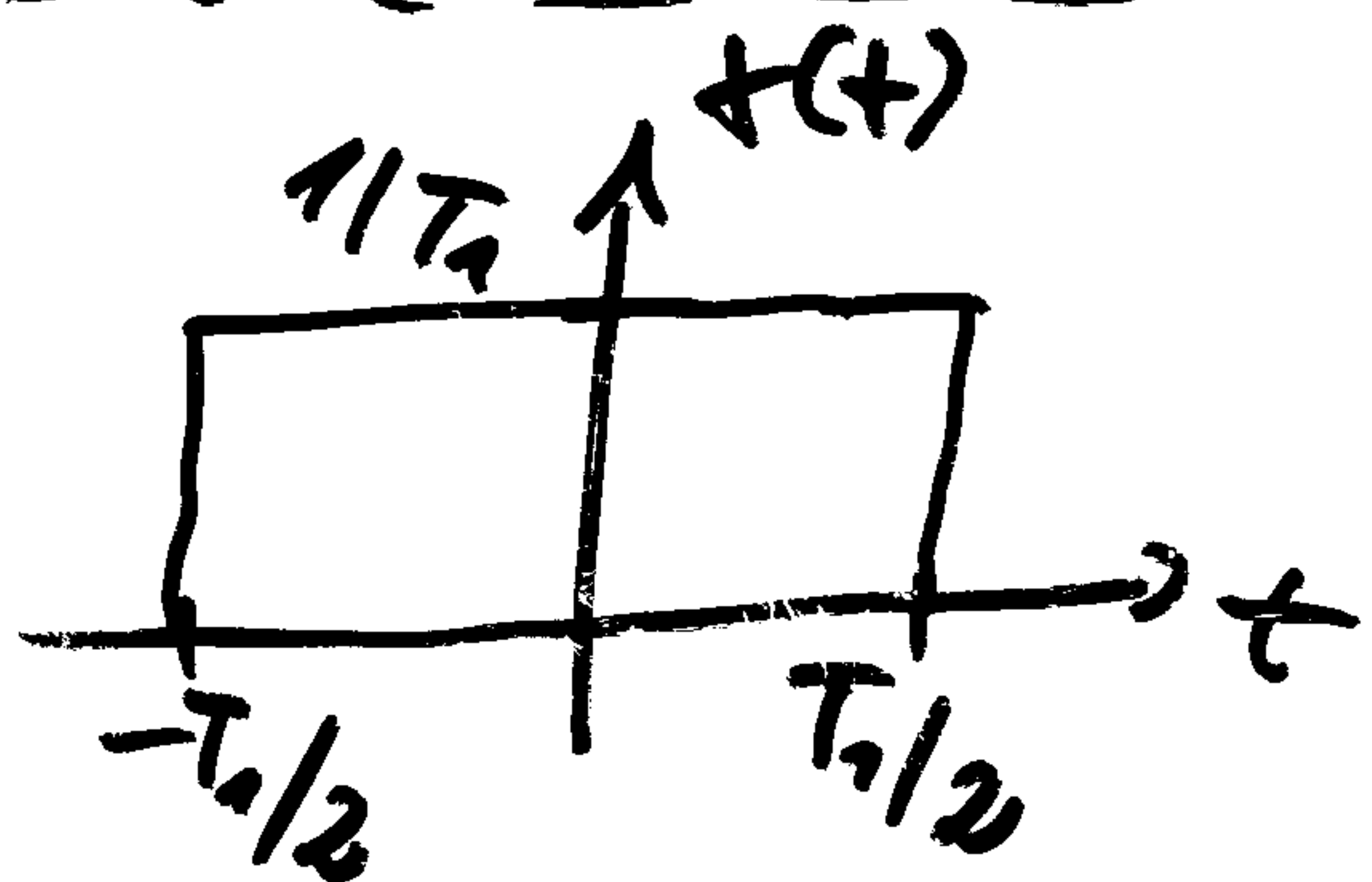
Für kausale Systeme:

$$y(t) = \int_0^{\infty} x(\tau) \cdot h(t-\tau) d\tau$$

Kommutativität der Faltungsoperation:

$$x(t) * h(t) = h(t) * x(t)$$

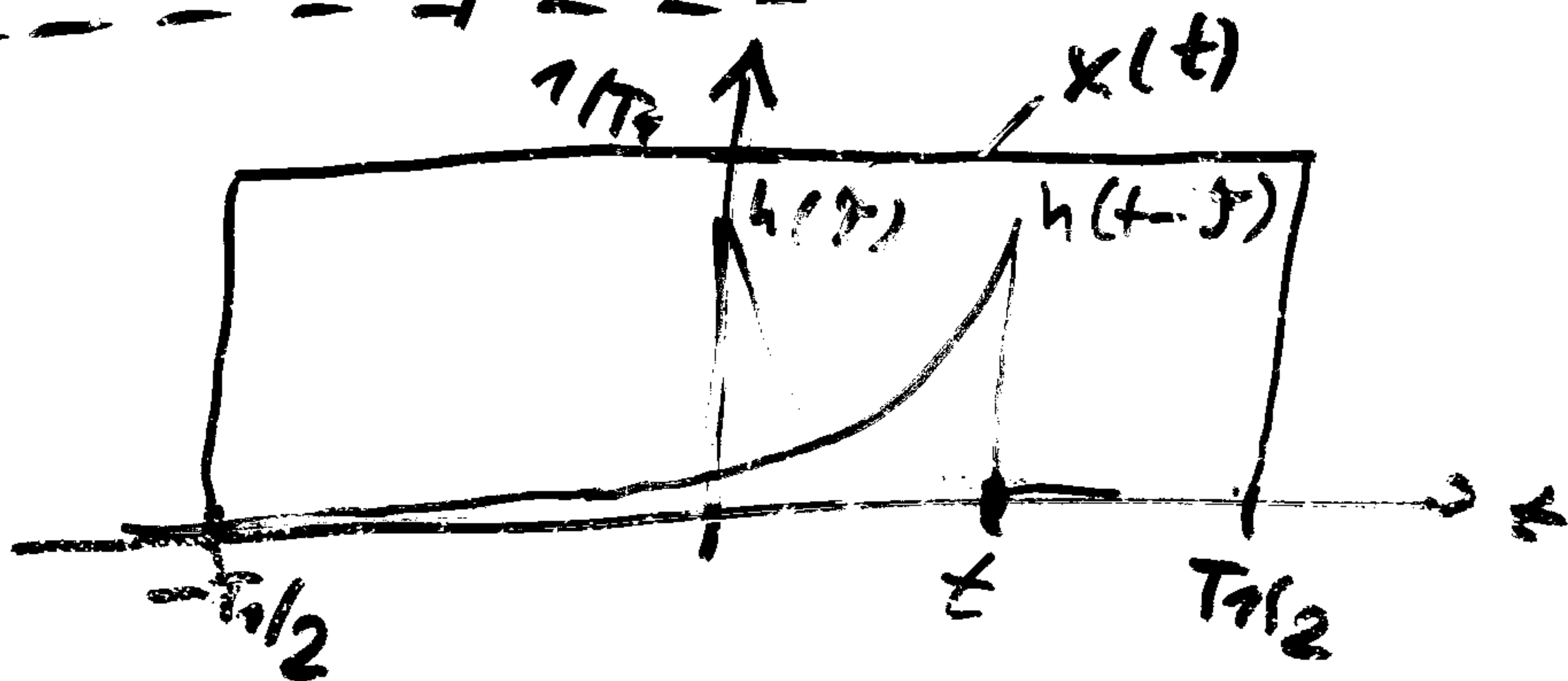
Beispiel: Rechteck-Erregung auf RL-Glied



$$x(t) = \frac{1}{T_1} \text{rect}\left(\frac{t}{T_1}\right) \quad h(t) = \frac{1}{T_{RL}} e^{-t/T_{RL}}$$

$$y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

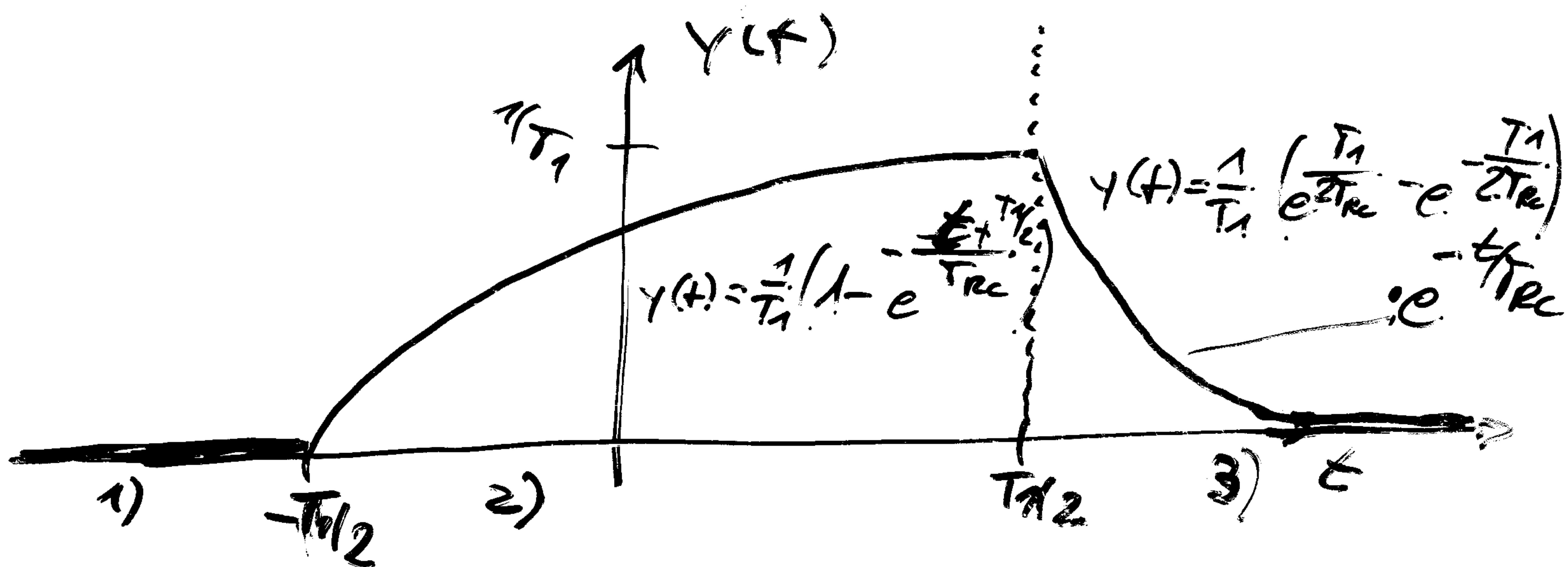
graphische Interpretation:



1) $t < -T_1/2$: $y(t) = 0$, da keine gemeinsame Fläche

$$\begin{aligned} 2) -T_1/2 \leq t \leq T_1/2 : y(t) &= \int_{-T_1/2}^t \frac{1}{T_1} \cdot \frac{1}{T_{RL}} e^{-\frac{t-\tau}{T_{RL}}} d\tau = \frac{1}{T_1} e^{-\frac{t-\tau}{T_{RL}}} \Big|_{-T_1/2}^t \\ &= \frac{1}{T_1} \left(1 - e^{-\frac{t+T_1/2}{T_{RL}}} \right) \end{aligned}$$

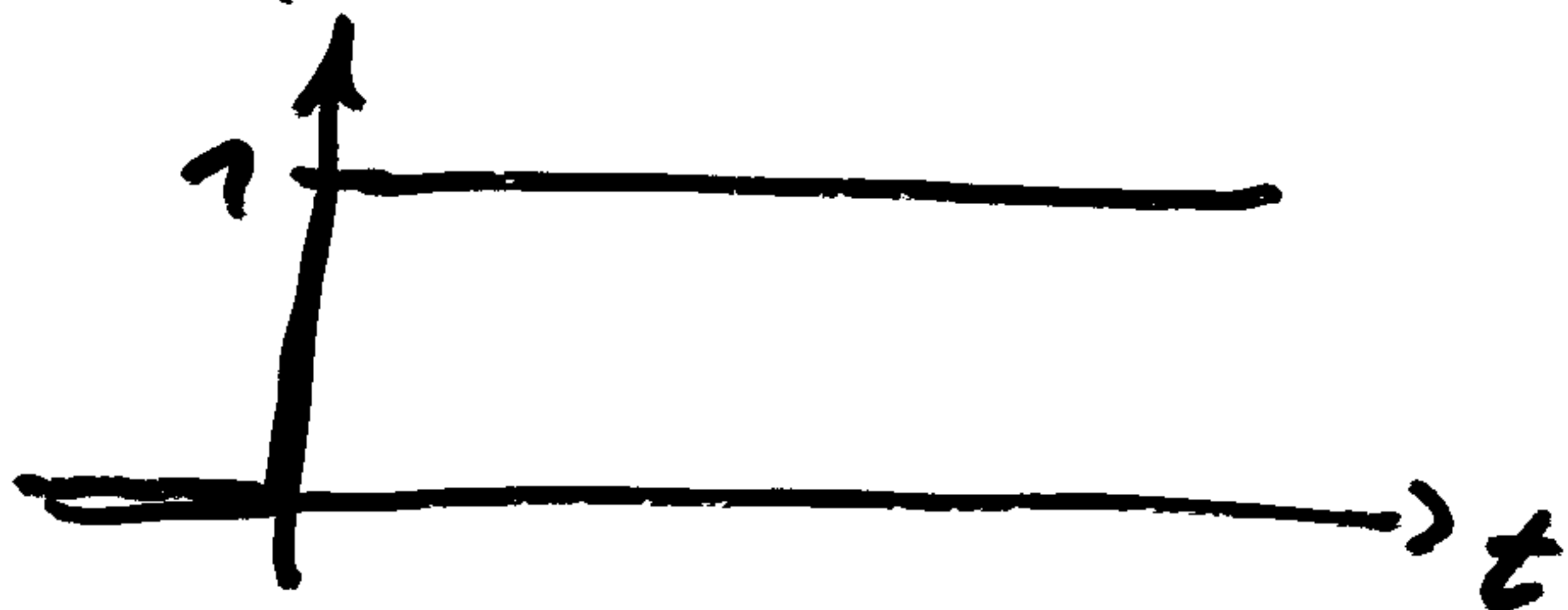
$$\begin{aligned} 3) t > T_1/2 : y(t) &= \int_{-T_1/2}^{T_1/2} \frac{1}{T_1} \cdot \frac{1}{T_{RL}} e^{-\frac{t-\tau}{T_{RL}}} d\tau = \frac{1}{T_1} \left(e^{-\frac{t-T_1/2}{T_{RL}}} - e^{-\frac{t+T_1/2}{T_{RL}}} \right) \\ &= \frac{1}{T_1} \left(e^{-\frac{t}{2T_{RL}}} - e^{-\frac{t}{2T_{RL}}} \right) e^{-\frac{t}{2T_{RL}}} \end{aligned}$$



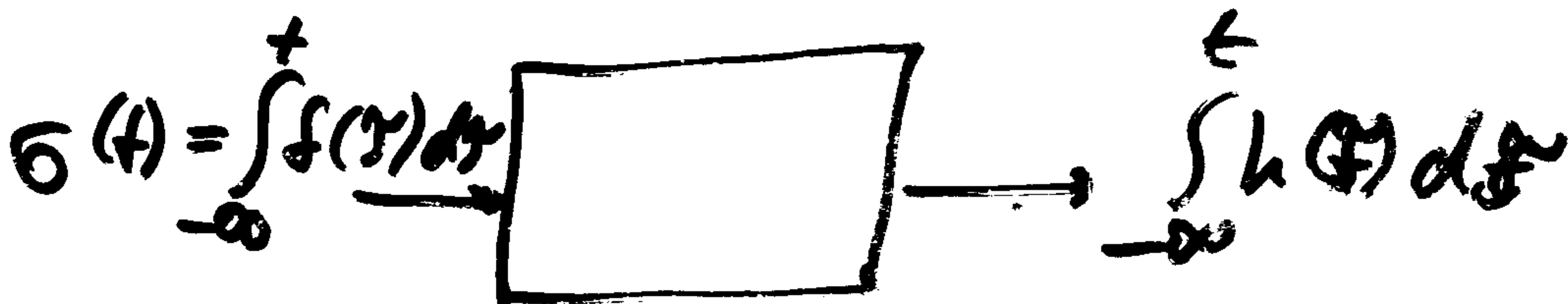
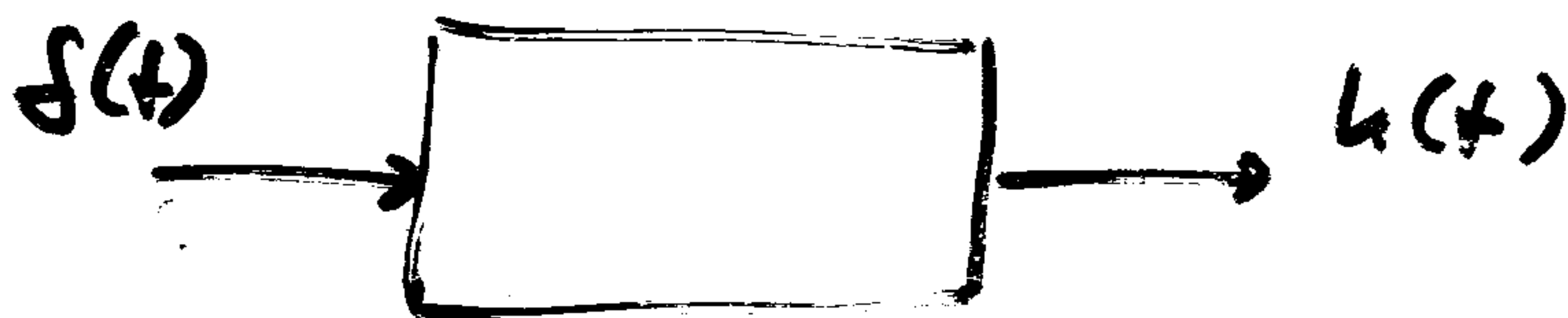
Beispiel 1.3.2 Sprungantwort

RC-Glied: $h(t) = \frac{1}{T_{RC}} e^{-t/T_{RC}}$

$x(t) = \delta(t)$



$$y(t) = \int_{-\infty}^t \delta(\tau) d\tau$$



$$y(t) = \int_{-\infty}^t h(\tau) d\tau = \frac{1}{T_{RC}} \int_{-\infty}^t e^{-\tau/T_{RC}} d\tau$$

$$= \frac{1}{T_{RC}} \cdot (-T_{RC}) e^{-\tau/T_{RC}} \Big|_{-\infty}^t$$

$$= - \left[e^{-t/T_{RC}} - 1 \right] = \frac{1 - e^{-t/T_{RC}}}{\text{Sprungantwort}}$$

Beispiel: Frequenzgang RC-Glied

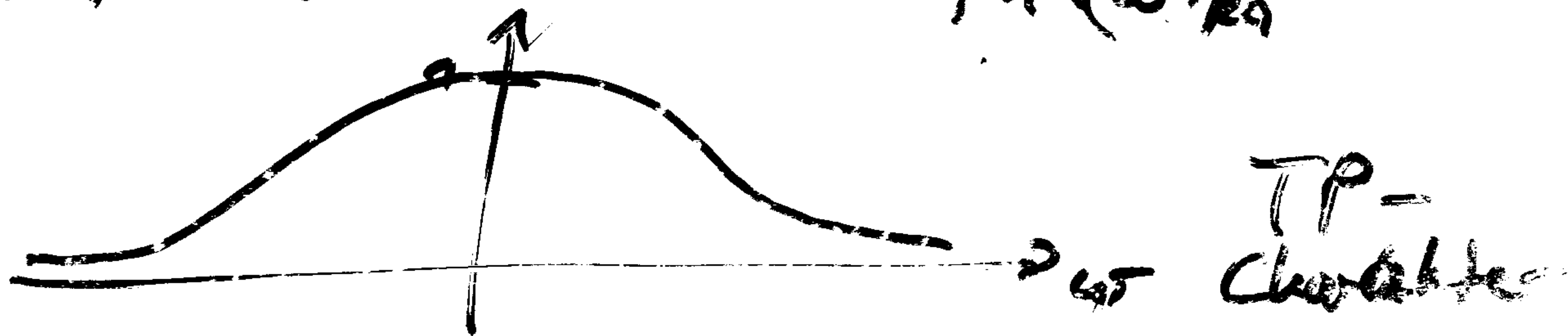
$$h(t) = \frac{1}{T_{RC}} e^{-t/T_{RC}} \Rightarrow H(\omega) = \int_{-\infty}^{\infty} h(t) e^{-j\omega t} dt$$

$$\Rightarrow H(\omega) = \int_0^{\infty} \frac{1}{T_{RC}} e^{-t/T_{RC}} e^{-j\omega t} dt = \frac{1}{T_{RC}} \int_0^{\infty} e^{-(\frac{1}{T_{RC}} + j\omega)t} dt$$

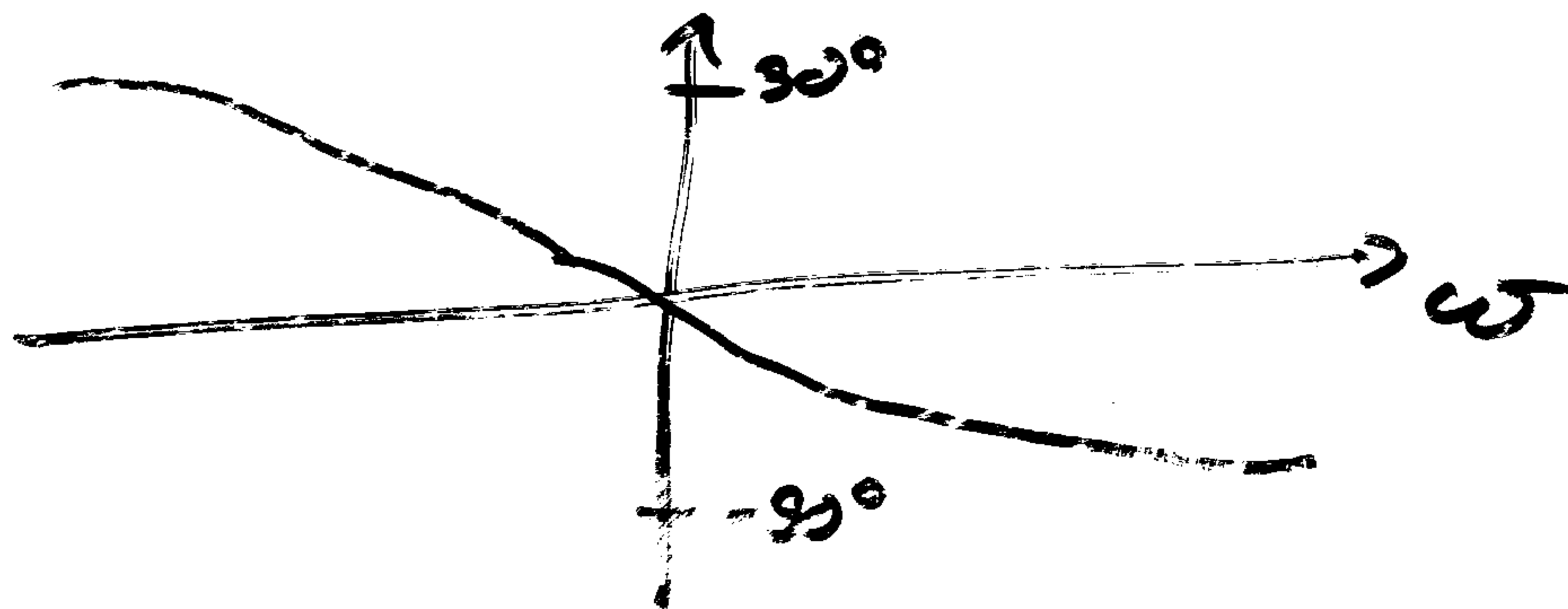
$$= -\frac{\frac{1}{T_{RC}}}{\frac{1}{T_{RC}} + j\omega} e^{-(\frac{1}{T_{RC}} + j\omega)t} \Big|_0^{\infty} = 0 - \left(-\frac{1}{1 + j\omega T_{RC}} \right)$$

$$= \underline{\underline{\frac{1}{1 + j\omega T_{RC}}}}$$

Betragsfrequenzgang: $|H(\omega)| = \frac{1}{\sqrt{1 + (\omega T_{RC})^2}}$



Phasengang: $\varphi(\omega) = -\arctan \frac{\omega T_{RC}}{1} = -\arctan(\omega T_{RC})$



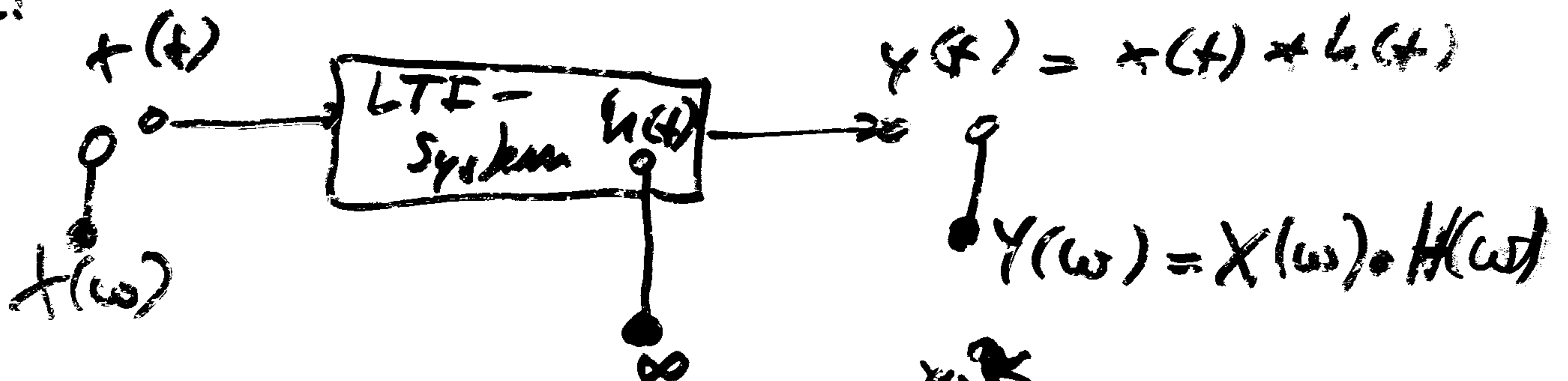
3.3. Eigenfunktion und Frequenzgang

Komplexe Exp. fkt.: $x(t) = \underline{U} \cdot e^{j\omega t}$
= Eigenfunktion von
LTI-Systemen

$$\begin{aligned} y(t) &= x(t) * h(t) = h(t) * x(t) \\ &= \int_{-\infty}^{\infty} h(\tau) \cdot x(t-\tau) d\tau \\ &= \int_{-\infty}^{\infty} h(\tau) \cdot \underline{U} \cdot e^{j\omega(t-\tau)} d\tau \\ &= \underbrace{\left[\int_{-\infty}^{\infty} h(\tau) \cdot e^{-j\omega\tau} d\tau \right]}_{F\{h(t)\} = H(\omega)} \cdot \frac{\underline{U} \cdot e^{j\omega t}}{x(t)} \end{aligned}$$

Frequenzgang
Übertragungsfunktion

Blockdiagramm:



$$H(\omega) = F\{h(t)\} = \int_{-\infty}^{\infty} h(t) \cdot e^{-j\omega t} dt$$

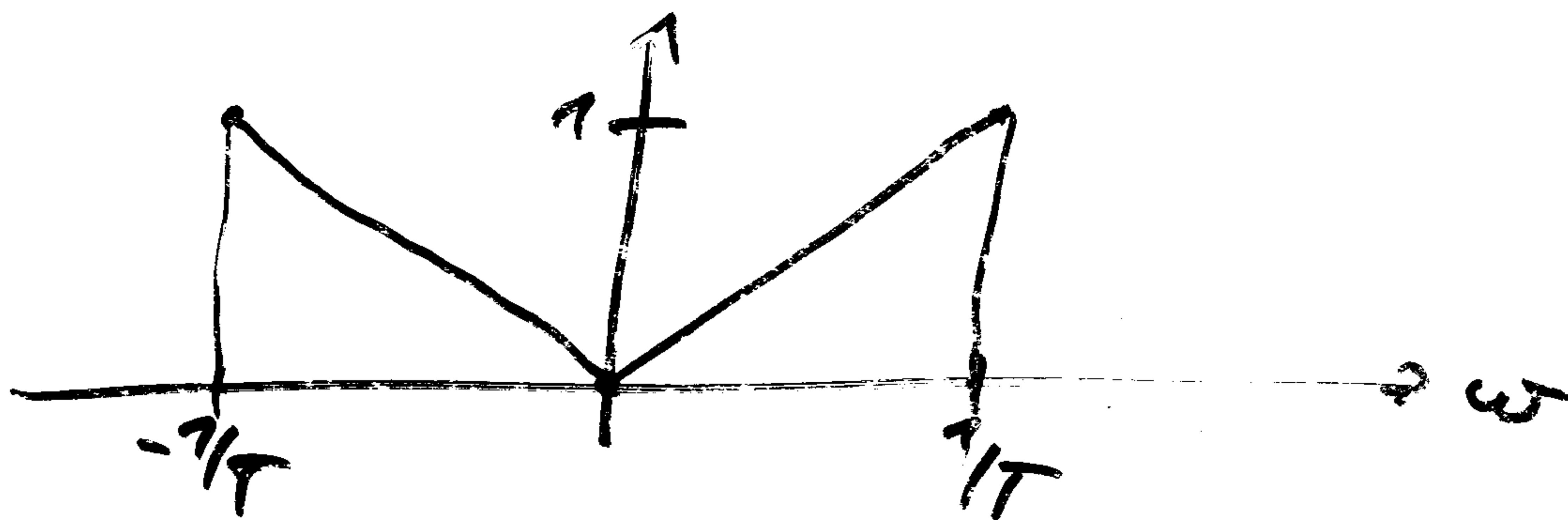
$|H(\omega)|$: Betragsfrequenzgang / $\varphi(\omega) = \arg\{H(\omega)\}$: Phasengang

Beispiel: Differenzierer

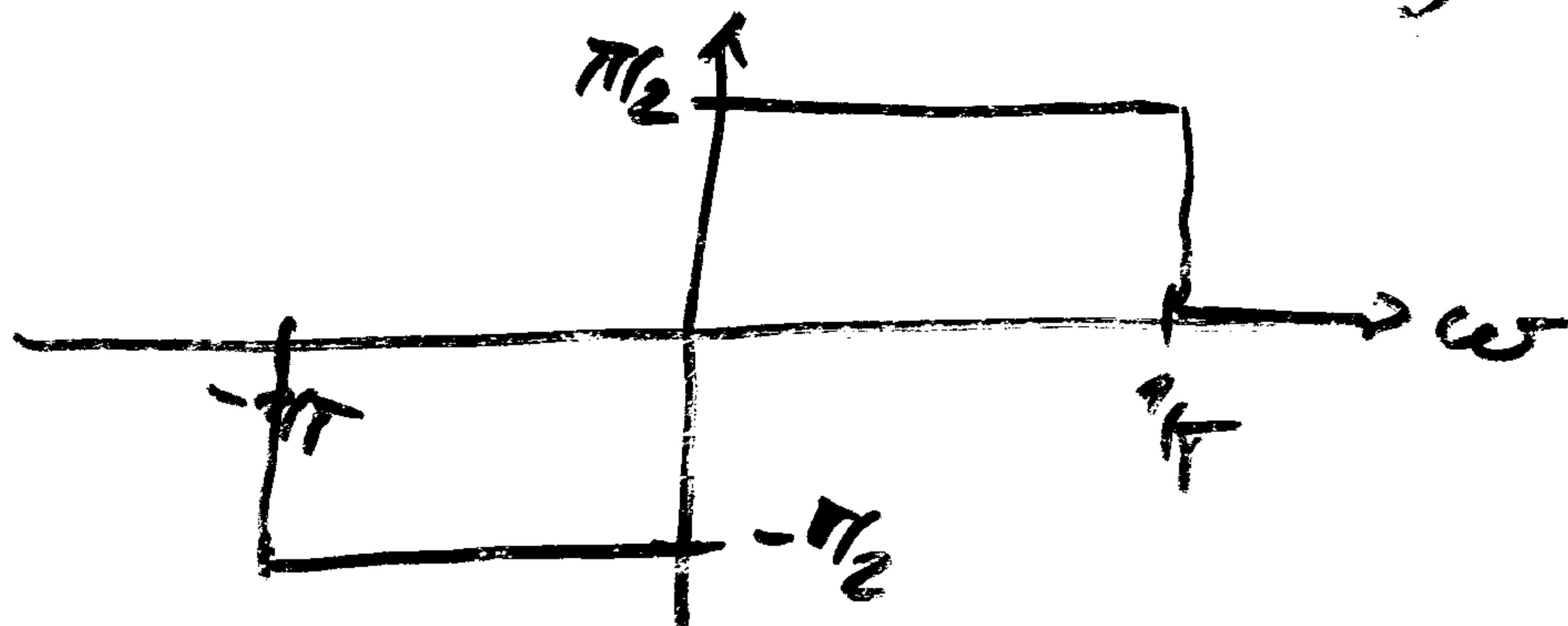
$$H(\omega) = \begin{cases} j\omega T & \text{f. } -\frac{1}{T} \leq \omega \leq \frac{1}{T} \\ 0 & \text{sonst} \end{cases}$$

$$x(t) = \cos(2\pi f_0 t + \varphi) \quad , \quad f_0 = \frac{1}{2T}$$

a) Betrag: $|H(\omega)| = |j\omega T|$



b) Phase: $\varphi(\omega) = \arctan\left\{\frac{\omega T}{0}\right\} = \arctan(\pm\infty) = \pm \pi/2$



→ 90°-Phasenverschiebung

→ Hilberttransformator

$$Y(\omega) = X(\omega) \cdot H(\omega)$$

$$X(\omega) = F \left\{ \underbrace{\frac{1}{2} e^{j\omega t} e^{j\tau} + \frac{1}{2} e^{j\omega t} e^{-j\tau}}_{\cos(\omega t + \tau)} \right\}$$

$$= \frac{1}{2} e^{j\tau} \delta(f - f_0) + \frac{1}{2} e^{-j\tau} \delta(f + f_0)$$

$$Y(\omega) = j\omega T \cdot \frac{1}{2} \cdot \{ e^{j\tau} \delta(f - f_0) + e^{-j\tau} \delta(f + f_0) \}$$

Ausblendabhängigkeit $\delta(f)$

$$Y(t) = j \underbrace{2\pi f_0}_{\omega_0} T \cdot \frac{1}{2} \cdot \{ e^{j\tau} e^{j2\pi f_0 t} - e^{-j\tau} e^{j2\pi f_0 t} \}$$

$$= \underbrace{2\pi f_0}_{\omega_0} T \cdot \sin(\underbrace{2\pi f_0 t}_{\omega_0} + \tau)$$