Vv256 Lecture 16

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A crucial step of the Laplace method is finding the inverse Laplace transform

$$Y(s) \longrightarrow y(t)$$

Definition:

$$\mathcal{L}\Big[y(t)\Big] = \int_0^\infty e^{-st} y(t) = Y(s)$$

Linear property:

$$Y(s) = c_1 Y_1(s) + c_2 Y_2(s)$$

Transform of a derivative:

$$\mathcal{L}[f^{(n)}(t)] = s^n F(s) - s^{n-1} f(0) - s^{n-2} f'(0) - \dots - s^0 f^{(n-1)}(0),$$

Translation Theorems:

$$\mathcal{L}\left[e^{at}f(t)\right] = F(s-a)$$
 and $\mathcal{L}\left[f(t-a)u(t-a)\right] = e^{-as}F(s)$

Q: What is the inverse Laplace transform of a product?

$$H(s) = F(s)G(s) \qquad \text{where} \qquad \frac{\mathcal{L}\big[f(t)\big] = F(s)}{\mathcal{L}\big[g(t)\big] = G(s)}$$

• If one of those function is

$$e^{-as}$$

then we can use the second translation theorem

$$\mathcal{L}[f(t-a)u(t-a)] = e^{-as}F(s)$$

Q: What happens otherwise? For example,

$$H(s) = \frac{1}{s^2(s-1)} = \frac{1}{s-1} - \frac{1}{s} - \frac{1}{s^2}$$
 or $R(s) = \frac{s}{(s^2 + k^2)^2}$

ullet We could use a PFD for H. However, it is not always possible. e.g. R(s).

Firstly notice

$$\mathcal{L}^{-1}\Big[H(s)\Big] \neq \mathcal{L}^{-1}\Big[F(s)\Big] \cdot \mathcal{L}^{-1}\Big[G(s)\Big]$$

For that, consider

$$\mathcal{L}^{-1}\left[\frac{1}{s^2(s-1)}\right] = \mathcal{L}^{-1}\left[\frac{1}{s-1} - \frac{1}{s} - \frac{1}{s^2}\right] = e^t - 1 - t$$

whereas

$$\mathcal{L}^{-1} \left[\frac{1}{s^2} \right] \cdot \mathcal{L}^{-1} \left[\frac{1}{s-1} \right] = te^t$$

Hence

$$\mathcal{L}^{-1}\left[\frac{1}{s^2(s-1)}\right] \neq \mathcal{L}^{-1}\left[\frac{1}{s^2}\right] \cdot \mathcal{L}^{-1}\left[\frac{1}{s-1}\right]$$

Convolution Theorem

If f(t) and g(t) are piecewise continuous on $[0,\infty)$ and of exponential order, then

$$\int_0^\infty e^{-st} \left(\int_0^t f(\tau)g(t-\tau) d\tau \right) dt = \mathcal{L}\Big[f(t)\Big] \mathcal{L}\Big[g(t)\Big] = F(s)G(s)$$

Proof

Let

$$F(s) = \int_0^\infty e^{-s\tau} f(\tau) \, d\tau, \quad \text{and} \quad G(s) = \int_0^\infty e^{-s\mu} g(\mu) \, d\mu$$

• Take the right-hand side, we have

$$F(s)G(s) = \left(\int_0^\infty e^{-s\tau} f(\tau) d\tau\right) \left(\int_0^\infty e^{-s\mu} g(\mu) d\mu\right)$$
$$= \int_0^\infty \int_0^\infty e^{-s(\tau+\mu)} f(\tau)g(\mu) d\mu d\tau$$

Proof

$$F(s)G(s) = \int_0^\infty \left(\int_0^\infty e^{-s(\tau+\mu)} f(\tau)g(\mu) d\mu \right) d\tau$$
$$= \int_0^\infty f(\tau) \left(\int_0^\infty e^{-s(\tau+\mu)} g(\mu) d\mu \right) d\tau$$

ullet To compute the inner integral, we hold au as fixed, and integrate w.r.t. μ ,

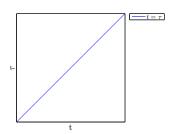
$$\int_0^\infty e^{-s(\tau+\mu)}g(\mu)\,d\mu$$

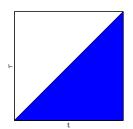
$$F(s)G(s) = \int_0^\infty f(\tau) \left(\int_0^\infty e^{-s(\tau+\mu)} g(\mu) d\mu \right) d\tau$$
$$= \int_0^\infty f(\tau) \left(\int_{t-\tau}^\infty e^{-st} g(t-\tau) dt \right) d\tau$$

Proof

• Consider what region we are integrating over,

$$\int_0^\infty \int_{t=\tau}^\infty dt \, d\tau = \iint dA = \int_0^\infty \int_0^t d\tau \, dt$$





 $F(s)G(s) = \iint e^{-st} f(\tau)g(t-\tau) dA = \int_0^\infty e^{-st} \left(\int_0^t f(\tau)g(t-\tau) d\tau \right) dt$

Definition

If f(t) and g(t) are piecewise continuous on the interval $[0,\infty)$, then the function

$$\int_0^t f(\tau)g(t-\tau)\,d\tau$$

is known as the convolution of f and g. This special function is denoted as

$$(f * g)(t) = \int_0^t f(\tau)g(t - \tau) d\tau$$

• The convolution of f and g, which is a function of t, is often considered as a product between f and g since it has some properties of a typical product.

$$f*g=g*f \qquad \text{commutative}$$

$$f*(g*h)=(f*g)*h \qquad \text{associative}$$

$$f*(g+h)=f*g+f*h \qquad \text{distributive}$$

Recall we had

$$\mathcal{L}^{-1}\left[\frac{1}{s^2(s-1)}\right] = \mathcal{L}^{-1}\left[\frac{1}{s-1} - \frac{1}{s} - \frac{1}{s^2}\right] = e^t - 1 - t$$

• Let us apply the convolution theorem

$$\mathcal{L}^{-1}\left[\frac{1}{s^2(s-1)}\right] = \mathcal{L}^{-1}\left[\frac{1}{s^2}\right] * \mathcal{L}^{-1}\left[\frac{1}{(s-1)}\right]$$

$$= (t) * (e^t) = (e^t) * (t) \qquad = \int_0^t f(\tau)g(t-\tau) d\tau$$

$$= \int_0^t e^{\tau}(t-\tau) d\tau$$

$$= e^t - t - 1$$

Exercise

Find the inverse Laplace transform of $\frac{1}{(s^2+k^2)^2}$.

• Let
$$F(s) = G(s) = \frac{k}{s^2 + k^2}$$
, so

$$f(t) = g(t) = \mathcal{L}^{-1} \left[\frac{k}{s^2 + k^2} \right] = \sin kt$$

 \bullet The convolution theorem states that $\mathcal{L}\Big[f*g\Big] = F(s)G(s)$

$$\mathcal{L}^{-1}\Big[F(s)G(s)\Big] = f * g = \int_0^t \sin k\tau \sin k(t-\tau) d\tau$$
$$= \frac{1}{2k} \sin kt - \frac{t}{2} \cos kt$$

•
$$\mathcal{L}^{-1} \left\{ \frac{1}{(s^2 + k^2)^2} \right\} = \frac{1}{k^2} \mathcal{L}^{-1} \left\{ \frac{k^2}{(s^2 + k^2)^2} \right\}$$

= $\frac{1}{k^2} \left(\frac{1}{2k} \sin kt - \frac{t}{2} \cos kt \right) = \frac{\sin kt - kt \cos kt}{2k^3}$

Series Circuits

• In a series circuit, Kirchhoff's second law states that the sum of the voltage drops across an inductor, resistor, and capacitor is equal to the voltage

• It is known that voltage drops across an inductor, resistor, and capacitor are,

$$\Delta V_L = L \frac{di}{dt}, \qquad \Delta V_R = Ri(t), \qquad \text{and} \qquad \Delta V_C = \frac{1}{C} \int_0^t i(\tau) \, d\tau,$$

respectively, where i(t) is the current and L, R, and C are constants.

 \bullet So the current in a LRC series circuit is governed by the following equation,

$$L\frac{di}{dt} + Ri(t) + \frac{1}{C} \int_0^t i(\tau) d\tau = E(t)$$

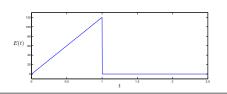
Exercise

Determine the current i(t) in a series LRC circuit when

$$L=0.1, \qquad R=2, \qquad C=0.1, \qquad \text{and} \quad i(0)=0,$$

and voltage is given by

$$E(t) = 120t - 120tu(t-1).$$



Solution

ullet To find the current i(t), we need to solve the following

$$\frac{1}{10}\frac{di}{dt} + 2i + 10\int_0^t i(\tau) d\tau = 120t - 120tu(t-1)$$

• This is known as integrodifferential equation.

• First notice that, in general,

$$f * g = \int_0^t f(\tau)g(t-\tau) d\tau = \int_0^t f(\tau) d\tau \quad \text{for} \quad g(t-\tau) = 1$$

According to the convolution theorem,

$$\mathcal{L}\Big[f*g\Big] = \mathcal{L}\Big[f\Big] \cdot \mathcal{L}\Big[g\Big] = \frac{F(s)}{s}, \quad \text{where} \quad \mathcal{L}\Big[f\Big] = F(s)$$

for the case when $g(t - \tau) = 1$.

• Back to the current i(t),

$$\mathcal{L}\left[\int_0^t f(\tau) \, d\tau\right] = \frac{F(s)}{s} \implies \mathcal{L}\left[\int_0^t i(\tau) \, d\tau\right] = \frac{I(s)}{s}$$

where
$$I(s) = \mathcal{L}[i(t)]$$
.

• Thus the Laplace transform of the integrodifferential equation is

$$\mathcal{L}\left[\frac{1}{10}\frac{di}{dt} + 2i + 10\int_{0}^{t} i(\tau) d\tau\right] = \mathcal{L}\left[120t - 120tu(t-1)\right]$$

$$\implies \frac{1}{10}sI(s) + 2I(s) + 10\frac{I(s)}{s} = 120\left(\frac{1}{s^{2}} - e^{-s}\mathcal{L}\left[t+1\right]\right)$$

$$\implies \frac{1}{10}sI(s) + 2I(s) + 10\frac{I(s)}{s} = 120\left(\frac{1}{s^{2}} - e^{-s}\frac{1}{s^{2}} - e^{-s}\frac{1}{s}\right)$$

• Rearranging and collecting I(s),

$$I(s) = 1200 \left[\frac{1}{s(s+10)^2} - \frac{1}{s(s+10)^2} e^{-s} - \frac{1}{(s+10)^2} e^{-s} \right]$$

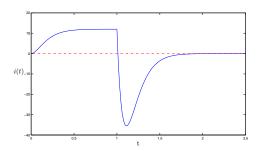
• By partial fractions,
$$I(s)=1200\left[\overbrace{\frac{1/100}{s}}^{\stackrel{(1)}{\underbrace{1/100}}}-\overbrace{\frac{1/100}{(s+10)^2}}^{\stackrel{(2)}{\underbrace{(3)}}}-\overbrace{\frac{1/100}{s}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}}^{\stackrel{(4)}{\underbrace{(4)}}}-\underbrace{\frac{1/100}{s+10}$$

Transforming back to the time domain,

$$\begin{split} i(t) &= 1200 \Bigg[\overbrace{\frac{1}{100}}^{(1)} - \overbrace{\frac{1}{100}}^{(2)} - \overbrace{\frac{1}{10}}^{(3)} - \overbrace{\frac{1}{100}}^{(4)} - \overbrace{\frac{1}{100}}^{(4)} - \underbrace{\frac{1}{100}}_{(6)} u(t-1) \\ &+ \underbrace{\frac{1}{100}}_{(5)} e^{-10(t-1)} u(t-1) - \underbrace{\frac{9}{10}}_{(6)} (t-1) e^{-10(t-1)} u(t-1) \Bigg] \end{split}$$

• Written as a piecewise-defined function,

$$i(t) = \begin{cases} 12 - 12e^{-10t} - 120te^{-10t}, & 0 \le t < 1 \\ -12e^{-10t} + 12e^{-10(t-1)} - 120te^{-10t} - 1080(t-1)e^{-10(t-1)} & 1 \le t. \end{cases}$$



ullet The input E(t) is discontinuous, however, the output i(t) is a continuous

• Consider a periodic function $\mathcal{L}[f(t)]$ where f(t+T) = f(t).

Theorem

Let f(t) be piecewise continuous on $[0,\infty)$ and of exponential order, and periodic with period T, then

$$\mathcal{L}\Big[f(t)\Big] = \frac{1}{1 - e^{-sT}} \int_0^T e^{-st} f(t) dt$$

Proof

• Break the interval into two parts,

$$\mathcal{L}\Big[f(t)\Big] = \int_0^T e^{-st} f(t) dt + \int_T^\infty e^{-st} f(t) dt$$

• When we let t = u + T, then the second integral becomes,

$$\int_0^\infty e^{-s(u+T)} f(u+T) du = e^{-sT} \int_0^\infty e^{-su} f(u) du = e^{-sT} \mathcal{L} \Big[f(t) \Big]$$

Exercise

The current i(t) in a single-loop LR series circuit is governed by

$$L\frac{di}{dt} + Ri = E(t)$$

Find the current i(t) when i(0) = 0 and E(t) is a square wave function

$$E(t) = \begin{cases} 1, & 0 \le t \le 1 \\ 0, & 1 \le t \le 2 \end{cases}, \quad E(t) = E(t+2)$$



Solution

• Find the Laplace transform of input function E(t),

$$\mathcal{L}\Big[E(t)\Big] = \frac{1}{1 - e^{-2s}} \int_0^2 e^{-st} E(t) dt$$
$$= \frac{1}{1 - e^{-2s}} \int_0^1 e^{-st} \cdot 1 dt = \frac{1}{1 - e^{-2s}} \frac{1 - e^{-s}}{s} = \frac{1}{s(1 + e^{-s})}$$

Applying the Laplace transform to the equation, we have

$$LsI(s) + RI(s) = \frac{1}{s(1 + e^{-s})}$$

• Rearranging and collecting I(s), we have

$$I(s) = \frac{1/L}{s(s+R/L)} \cdot \frac{1}{1+e^{-s}}$$

ullet To find the current in the t-domain, we first make use of power series,

$$1 - r + r^{2} - r^{3} \cdots = \frac{1}{1 - (-r)} \implies \frac{1}{1 + e^{-s}} = 1 - e^{-s} + e^{-2s} - e^{-3s} \cdots$$

• By partial fraction,

$$\frac{1/L}{s(s+R/L)} = \frac{1}{L} \left(\frac{L/R}{s} - \frac{L/R}{s+R/L} \right)$$

Together, we have

$$\begin{split} I(s) &= \frac{1}{R} \left(\frac{1}{s} - \frac{1}{(s+R/L)} \right) \left(1 - e^{-s} + e^{-2s} - e^{-3s} + \cdots \right) \\ &= \frac{1}{R} \left(\frac{1}{s} - \frac{e^{-s}}{s} + \frac{e^{-2s}}{s} - \frac{e^{-3s}}{s} + \cdots \right) \\ &- \frac{1}{R} \left(\frac{1}{s+R/L} - \frac{e^{-s}}{s+R/L} + \frac{e^{-2s}}{s+R/L} - \frac{e^{-3s}}{s+R/L} + \cdots \right) \end{split}$$

Now we are ready to back transform!!!

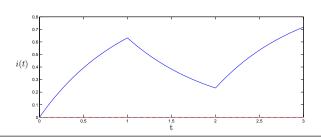
$$i(s) = \frac{1}{R} \left(1 - u(t-1) + u(t-2) - u(t-3) + \cdots \right)$$
$$- \frac{1}{R} \left(e^{-Rt/L} - e^{-R(t-1)/L} u(t-1) + e^{-R(t-2)/L} u(t-2) - \cdots \right)$$

• Tidy it up, we have

$$i(t) = \frac{1}{R} \left(1 - e^{-Rt/L} \right) + \frac{1}{R} \sum_{k=1}^{\infty} (-1)^k \left(1 - e^{-R(t-k)/L} \right) u(t-k)$$

ullet For R=1, L=1, and let's consider $0\leq t\leq 3$, then

$$i(t) = 1 - e^{-t} - (1 - e^{-(t-1)})u(t-1) + (1 - e^{-(t-2)})u(t-2)$$



Recall we had our theorem on the Laplace transform of a derivative function

$$\mathcal{L}\left[\frac{df}{dt}\right] = s\mathcal{L}\left[f(t)\right] - f(0) = sF(s) - f(0)$$

Q: What will be the inverse Laplace transform of

$$\frac{d}{ds}F(s) \qquad \text{where} \qquad F(s) = \mathcal{L}\Big[f(t)\Big]$$

 \bullet Suppose f(t) is well-behaved, thus we can interchange the order of

integration and differentiation

$$\frac{d}{ds}F(s) = \frac{d}{ds} \int_0^\infty e^{-st} f(t) dt = \int_0^\infty \frac{d}{ds} \left(e^{-st} f(t) \right) dt$$
$$= \int_0^\infty -t e^{-st} f(t) dt = -\mathcal{L} \left[t f(t) \right]$$

Exercise

Find the Laplace transform of te^{3t} using the last theorem.

Solution

We can treat it as a translation

$$\mathcal{L}\left[e^{at}f(t)\right] = F(s - a)$$

Then

$$\mathcal{L}[te^{3t}] = \frac{n!}{(s-a)^{n+1}}$$
$$= \frac{1}{(s-3)^2}$$

for s > 3.

We can treat it as a derivative

$$\mathcal{L}\Big[tg(t)\Big] = -\frac{d}{ds}F(s)$$

Then

$$\mathcal{L}[te^{3t}] = -\frac{d}{ds} \left(\frac{1}{s-a}\right)$$
$$= -\frac{d}{ds} \left(\frac{1}{s-3}\right)$$
$$= \frac{1}{(s-3)^2}$$

for s > 3.

• We can use the last result to find the Laplace transform of

$$\begin{split} \mathcal{L}\Big[t^2f(t)\Big] &= \mathcal{L}\Big[t\cdot tf(t)\Big] = -\frac{d}{ds}\mathcal{L}\Big[tf(t)\Big] = -\frac{d}{ds}\left(-\frac{d}{ds}\mathcal{L}\Big[f(t)\Big]\right) \\ &= \frac{d^2}{ds^2}\left(\mathcal{L}\Big[f(t)\Big]\right) \end{split}$$

• The preceding two cases suggest the following general result:

Theorem

If
$$F(s) = \mathcal{L}\Big[f(t)\Big]$$
 and ${\color{red}n}=1,2,3,\ldots$, then

$$\mathcal{L}\left[t^{n}f(t)\right] = (-1)^{n} \frac{d^{n}}{ds^{n}} F(s)$$

Q: When is this theorem useful?

Exercise

Solve
$$t\ddot{y}-t\dot{y}+y=2,\quad y(0)=2,\quad \dot{y}(0)=-4$$
 using the method of Laplace.

Solution

• Take the Laplace transform and apply the last theorem, we have

$$-\frac{d}{ds}\mathcal{L}\Big[\ddot{y}\Big] + \frac{d}{ds}\mathcal{L}\Big[\dot{y}\Big] + \mathcal{L}\Big[y\Big] = \mathcal{L}\Big[2\Big]$$

$$-\frac{d}{ds}\Big(s^{2}Y(s) - sy(0) - y'(0)\Big) + \frac{d}{ds}\Big(sY(s) - y(0)\Big) + Y(s) = \frac{2}{s}$$

$$-2sY(s) - s^{2}Y'(s) + y(0) + Y(s) + sY'(s) + Y(s) = \frac{2}{s}$$

$$Y'(s) + \frac{2}{s}Y(s) = \frac{2}{s^{2}}$$

• From this first order linear equation, and the initial condition $\dot{y}(0)=-4$

$$Y(s) = 2s^{-1} + Cs^{-2} \implies y(t) = 2 + ct \implies y(t) = 2 - 4t$$

 A linear equation with variable coefficients can be solved by using the method of Laplace transform, but it is not a universal method.

Using series is still the standard method for variable coefficients

- However, a second approach offers a way to compute various things, e.g.
 the Laplace transform of special functions
- ${\sf Q}$: How can we find the closed-form for the Laplace transform of J_0 .

$$\mathcal{L}\left[J_0\right] = \mathcal{L}\left[\sum_{n=0}^{\infty} \frac{(-1)^n}{n!\Gamma(1+\lambda+n)} \left(\frac{t}{2}\right)^{2n+\lambda}\right]$$
$$= \sum_{n=0}^{\infty} \frac{(-1)^n}{n!\Gamma(1+n)} \left(\frac{1}{2}\right)^{2n} \mathcal{L}\left[t^{2n}\right]$$
$$= \sum_{n=0}^{\infty} \frac{(-1)^n}{n!\Gamma(1+n)} \left(\frac{1}{2}\right)^{2n} \frac{(2n)!}{s^{2n+1}}$$

Q: The closed form is not obvious, what shall we do next?

• Let us consider the values of J_0 and J'_0 at t=0.

$$J_0(0) = \sum_{n=0}^{\infty} \frac{(-1)^n}{n!\Gamma(1+n)} \left(\frac{t}{2}\right)^{2n} \bigg|_{t=0} = 1 + 0 + 0 + \dots = 1$$

and

$$J'(0) = \sum_{n=1}^{\infty} \frac{(-1)^n}{n!\Gamma(1+n)} \frac{2n}{2} \left(\frac{t}{2}\right)^{2n-1} \bigg|_{t=0} = 0$$

• Thus the following initial-value problem defines J_0 .

$$t^2\ddot{y} + t\dot{y} + (t^2 - \lambda^2)y = 0$$
 $y(0) = 1$ $\dot{y}(0) = 0$

where $\lambda = 0$.

Q: Why is this useful? What shall we do next?

Take the Laplace transform, we have

$$\mathcal{L}\Big[t\ddot{y}+\dot{y}+ty\Big]=\mathcal{L}\Big[0\Big]$$

Apply the last theorem, we have

$$-\frac{d}{ds}\mathcal{L}\Big[\ddot{y}\Big] + \mathcal{L}\Big[\dot{y}\Big] - \frac{d}{ds}\mathcal{L}\Big[y\Big] = \mathcal{L}\Big[0\Big]$$

• Apply the theorem of the Laplace transform of a derivative,

$$-\frac{d}{ds}\left(s^2Y(s) - sy(t_0) - y'(t_0)\right) + sY(s) - y(t_0) - \frac{dY}{ds} = 0$$
$$-2sY(s) - s^2Y'(s) + y(t_0) + sY(s) - y(t_0) - Y'(s) = 0$$
$$-(s^2 + 1)Y'(s) - sY(s) = 0$$

Solve this first order linear equation, we have

$$Y(s) = \frac{C}{\sqrt{s^2 + 1}}$$
 where C is an arbitrary constant.

Q: How can we determine the arbitrary constant C?

$$Y(s) = \frac{C}{\sqrt{s^2 + 1}}$$

ullet We need some connection between initial values in t-domain and s-domain.

Theorem

If f(t) and f'(t) are continuous, and of exponential order, then

$$\lim_{t\to 0} f(t) = \lim_{s\to \infty} sF(s) \qquad \text{where} \qquad F(s) = \int_0^\infty e^{-st} f(t) \, dt$$

Furthermore,

$$\lim_{t\to\infty} f(t) = \lim_{s\to 0} sF(s) \qquad \text{where} \qquad F(s) = \int_0^\infty e^{-st} f(t) \, dt$$

provided the limit of f(t) exists when $t \to \infty$.

• This is often known as the initial-value/final-value theorem.

ullet Since J_0 converges for $t\in\mathbb{R}$, and it is a solution to a second-order equation

$$\lim_{t \to 0} J_0 = J_0(0) = 1$$

• So the initial-value theorem implies

$$1 = \lim_{s \to \infty} s\mathcal{L}[J_0] = \lim_{s \to \infty} s \frac{C}{\sqrt{s^2 + 1}} = \lim_{s \to \infty} \frac{C}{\sqrt{1 + \frac{1}{s^2}}} = C$$

Hence the Laplace transform of Bessel function of the first kind of order 0 is

$$\mathcal{L}\left[J_0\right] = \frac{1}{\sqrt{s^2 + 1}}$$

Exercise

Solve $t\ddot{y}+\dot{y}+4ty=0, \quad y(0)=3, \quad \dot{y}(0)=0$ using the method of Laplace.

• Taking the Laplace transform, we obtain

$$-\frac{d}{ds}\left(s^2Y(s) - sy(0) - y'(0)\right) + sY(s) - y(0) - 4\frac{dY}{ds} = 0$$

Use the initial conditions and simplify, we have

$$(s^2 + 4)Y'(s) + sY(s) = 0$$

Solve this first-order linear equation, and use the following theorem, we have

$$Y = \frac{C}{\sqrt{s^2 + 4}} = \frac{C}{2\sqrt{\left(\frac{s}{2}\right)^2 + 1}} \implies y(t) = CJ_0(2t) \implies y(t) = 3J_0(2t)$$

Theorem

Let
$$F(s) = \int_{0}^{\infty} e^{-st} f(t) dt$$
, then $\frac{1}{a} F\left(\frac{s}{a}\right) = \int_{0}^{\infty} e^{-st} f(at) dt$, where $a > 0$.