

Homework 3

Answers:

1. (a) It is an impulse train with a period of $T = 4$

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk \frac{2\pi}{T} t} dt = \frac{1}{4}$$

We can find that a_k is a constant.

$$x(t) = \sum_{k=-\infty}^{\infty} \frac{1}{4} e^{jk\omega_0 t} = \frac{1}{4} + \sum_{k=1}^{\infty} \frac{1}{2} \cos(k \frac{\pi}{2} t)$$

- (b) Sketch $x(t)$ and we find it can be expressed as

$$\begin{aligned} x(t) &= 1 + \sum_{n=-\infty}^{\infty} \text{rect}(t - 5n - \frac{1}{2}) \\ x(t) &= 1 + \sum_{k=-\infty}^{\infty} \frac{1}{5} \text{sinc}(\frac{k}{5}) e^{jk(2\pi/5)(1/2)} e^{jk(2\pi/5)t} \\ x(t) &= \frac{6}{5} + \sum_{k=1}^{\infty} \frac{1}{5} \text{sinc}(\frac{k}{5}) \cos(\frac{2\pi(t - 1/2)k}{5}) \end{aligned}$$

- (c) For this signal $T = 2$, $\omega_0 = \pi$

$$\begin{aligned} a_k &= \frac{1}{2} \int_0^1 e^{-t} e^{-jk\omega_0 t} dt \\ a_k &= \frac{1}{2} \int_0^1 e^{-(1+jk\pi)t} dt \\ a_k &= \frac{1 - e^{-(1+jk\pi)}}{2(1+jk\pi)} \\ x(t) &= \frac{1 - e^{-1}}{2} + \sum_{k=1}^{\infty} \left[\frac{1 - e^{-1} \cos(k\pi)}{1 + k^2 \pi^2} \cos(k\pi t) + \frac{k\pi(1 - e^{-1} \cos(k\pi))}{1 + k^2 \pi^2} \sin(k\pi t) \right] \end{aligned}$$

- 2.

$$\begin{aligned} x(t) &= a_0 + \sum_{k=1}^{\infty} (a_k e^{jk\omega_0 t} + a_{-k} e^{-jk\omega_0 t}) \\ x(t) &= a_0 + \sum_{k=1}^{\infty} [(a_k + a_{-k}) \cos(k\omega_0 t) + (a_k - a_{-k}) j \sin(k\omega_0 t)] \\ B[0] &= a_0 = \frac{1}{T} \int_0^T x(t) dt \\ B[k] &= a_k + a_{-k} = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt + \frac{1}{T} \int_0^T x(t) e^{jk\omega_0 t} dt = \frac{1}{T} \int_0^T x(t) (e^{-jk\omega_0 t} + e^{jk\omega_0 t}) dt \\ B[k] &= \frac{2}{T} \int_0^T x(t) \cos(k\omega_0 t) dt \end{aligned}$$

For the same reason,

$$A[k] = (a_k - a_{-k})j = \frac{2}{T} \int_0^T x(t) \sin(k\omega_0 t) dt$$

3. (a) If $x(t)$ is real. Then $x(t) = x(t)^*$, which implies that $a_k = a_{-k}^*$. It is not true in this case, so $x(t)$ is not real.
- (b) If $x(t)$ is even, then $x(t) = x(-t)$ and $a_k = a_{-k}$. It is true in this case, so $x(t)$ is even.
- (c) We have

$$g(t) = \frac{dx(t)}{dt} \xrightarrow{FS} b_k = jk \frac{2\pi}{T_0} a_k$$

Therefore,

$$b_k = \begin{cases} 0 & , k = 0 \\ -k(\frac{1}{2})^{|k|} \frac{2\pi}{T_0}, & otherwise \end{cases}$$

Since b_k is not even, $\frac{dx(t)}{dt}$ is not even.

4. (a) we have

$$x(t) = \sum_{\text{odd } k} a_k e^{jk(2\pi/T)t}$$

Therefore,

$$x(t + \frac{T}{2}) = \sum_{\text{odd } k} a_k e^{jk\pi} e^{jk(2\pi/T)t}$$

Since $e^{jk\pi} = -1$ for odd k .

$$x(t + \frac{T}{2}) = -x(t)$$

- (b) The Fourier series coefficients of $x(t)$ are

$$a_k = \frac{1}{T} \int_0^{T/2} x(t) e^{-jk\omega_0 t} dt + \frac{1}{T} \int_{T/2}^T x(t) e^{-jk\omega_0 t} dt$$

$$a_k = \frac{1}{T} \int_0^{T/2} [x(t) + x(t + \frac{T}{2})] e^{-jk\omega_0 t} dt$$

Note that a_k evaluate to zero for even k if we have $x(t) = -x(t + \frac{T}{2})$. So it is odd harmonious.

5. (a) $x(t) = \text{LC} \frac{d^2 y(t)}{dt^2} + \text{RC} \frac{dy(t)}{dt} + y(t)$, so the diffeq is given by $\boxed{\frac{d^2 y(t)}{dt^2} + \frac{dy(t)}{dt} + y(t) = x(t)}$.
- (b) The output of $e^{j\omega t}$ should be in the form $H(j\omega) e^{j\omega t}$, where $H(j\omega)$ have nothing to do with time t . Hence $\frac{d^2}{dt^2}(H(j\omega) e^{j\omega t}) + \frac{d}{dt}(H(j\omega) e^{j\omega t}) + H(j\omega) e^{j\omega t} = e^{j\omega t}$. From which we have $H(j\omega) = \boxed{\frac{1}{-\omega^2 + j\omega + 1}}$.
- (c) $H(s) = \boxed{\frac{1}{s^2 + s + 1}}$.
- (d) $|H(j\omega)| = \left(\frac{1}{(1-\omega^2)^2 + \omega^2} \right)^{1/2}$. Use Matlab `ezplot('1/((1-x^2)^2 + x^2)^(1/2)', [0 6])` to plot this function of ω :
- (e) Sample code:

```
num = [1];
den = [1 1 1];
x = linspace(0, 6);
H = freqs(num, den, x);
plot(x, abs(H));
xlabel('Angular Freqs w (rad/s)');
ylabel('Magnitude Response |H(jw)|');
title('Hw6-3e')
```

And the output figure:

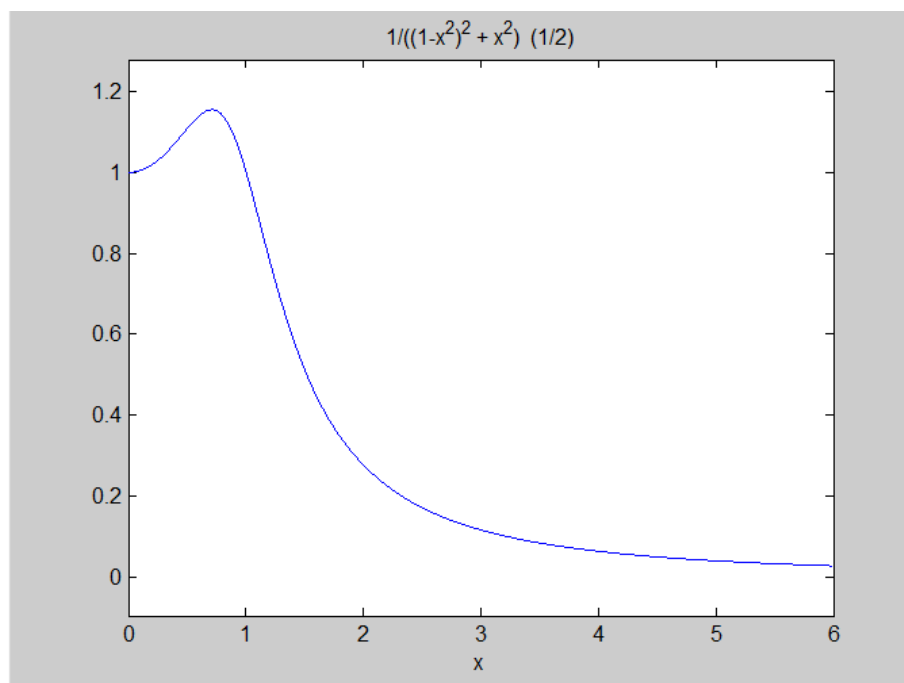


Figure 1: HW3-5d

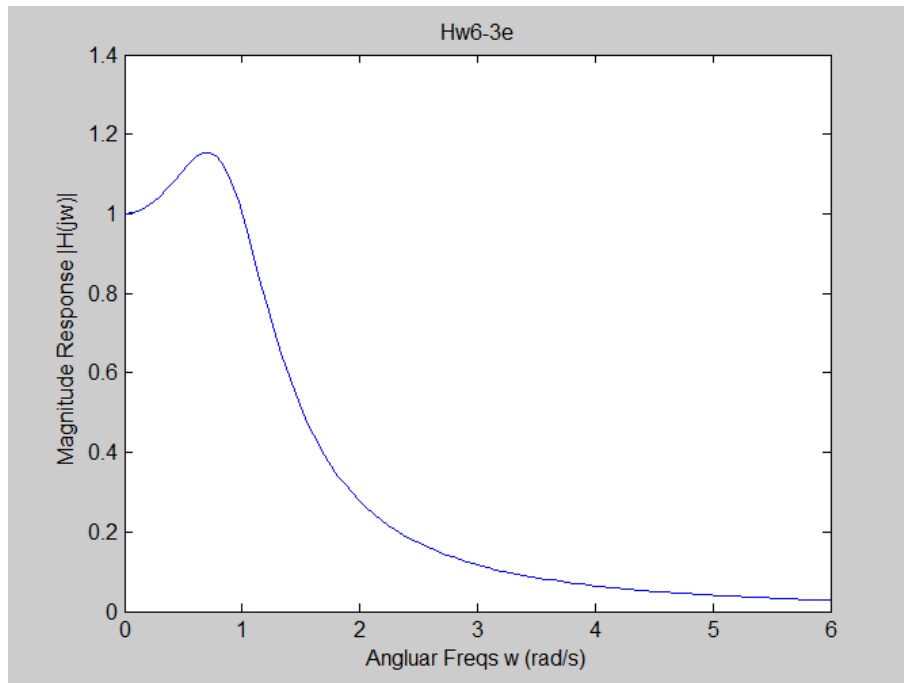


Figure 2: HW3-5e

We can see that the figure is identical to part (d), as expected.

- (f) $T_0 = 2\pi$ and $\omega_0 = 1$. And $x(t) = \boxed{1 + \frac{1}{2j}(e^{jt} - e^{-jt}) + \frac{1}{2j}(e^{4jt} - e^{-4jt})}$. Note that this already implies $c_1 = c_4 = 1/(2j)$, $c_{-1} = c_{-4} = -1/(2j)$ and $c_0 = 1$. After calculating the $|c_k|^2$'s, we use the following code to generate a PDS:

```
x = [-4 -1 0 1 4];
y = [0.25 0.25 1 0.25 0.25];
stem(x,y,'filled');
axis([-5 5 0 1.5]);
xlabel('Angular frequency \omega (k\omega_0)'),
ylabel('Component power |c_k|^2'),
title('Power Density Spectrum (Hw6-3f)')
```

- (g) By using computer, we get $\frac{1}{2\pi} \int_0^{2\pi} |x(t)|^2 = 2$. And we have $\sum |c_k|^2 = 1 + 0.25 + 0.25 + 0.25 + 0.25 = 2$.

Hence the Parseval's relation is verified.

- (h) $y(t) = 1 * H(0) + \frac{1}{2j}(H(1)e^{jt} - H(-1)e^{-jt} + H(4)e^{j4t} - H(-4)e^{-j4t}) = 1 + \frac{1}{2j}(\frac{1}{j}e^{jt} - \frac{1}{-j}e^{-jt} + \frac{1}{-15+j4}e^{j4t} - \frac{1}{-15-j4}e^{-j4t}) = \boxed{1 - \cos(t) - \frac{4}{241} \cos(4t) - \frac{15}{241} \sin(4t)}$. The code for PDS is very similar to part (f), and the figure is shown below: (Note that the stem at ± 4 should have height $1/964$.)

- (i) The high frequency component (4 rad/s sine wave) is attenuated. The circuit serves as a lowpass filter.

6. The sample code and the three plots are given below:

```
close all;
clear all;

[num, den] = besself(5,1e4);
f = linspace(0, 2e4, 1e5);
H = freqs(num, den, f);
```

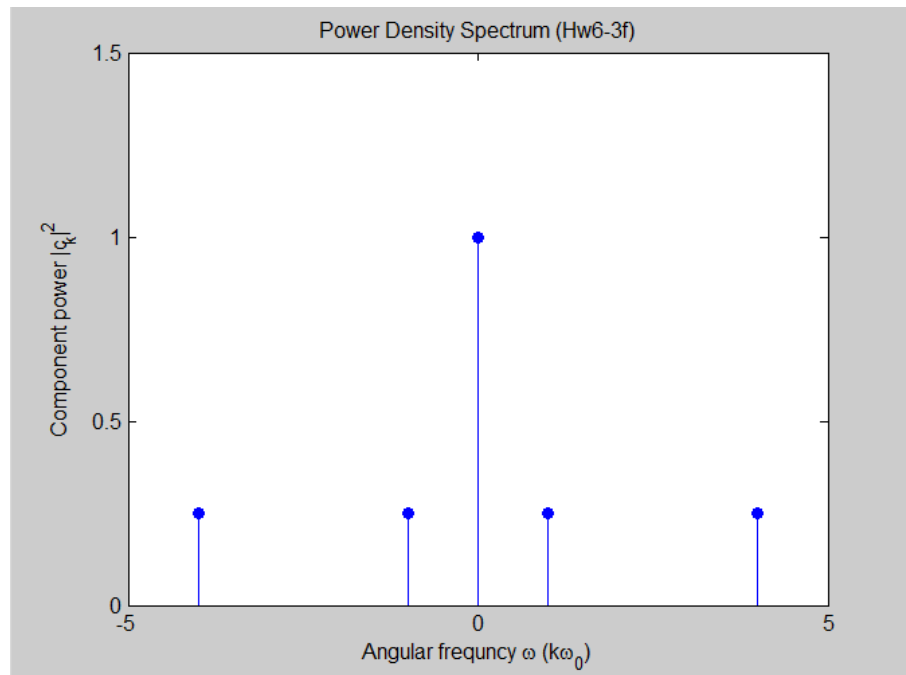


Figure 3: HW3-5f

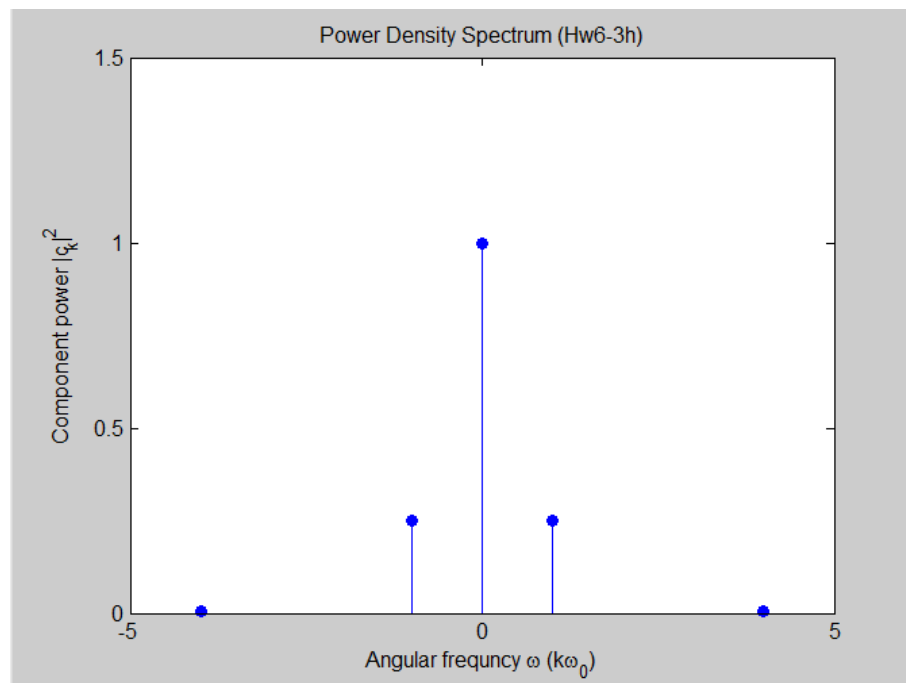


Figure 4: HW3-5h

```
figure, plot(f, abs(H));

[u, t] = gensig('square',0.002, 1, 1e-4);
figure, plot(t,u);
axis([0 0.01 -1 2]);

sys = tf(num, den);
y = lsim(sys, u, t);
```

```
figure, plot(t, y);
axis([0 0.01 -1 2]);
```

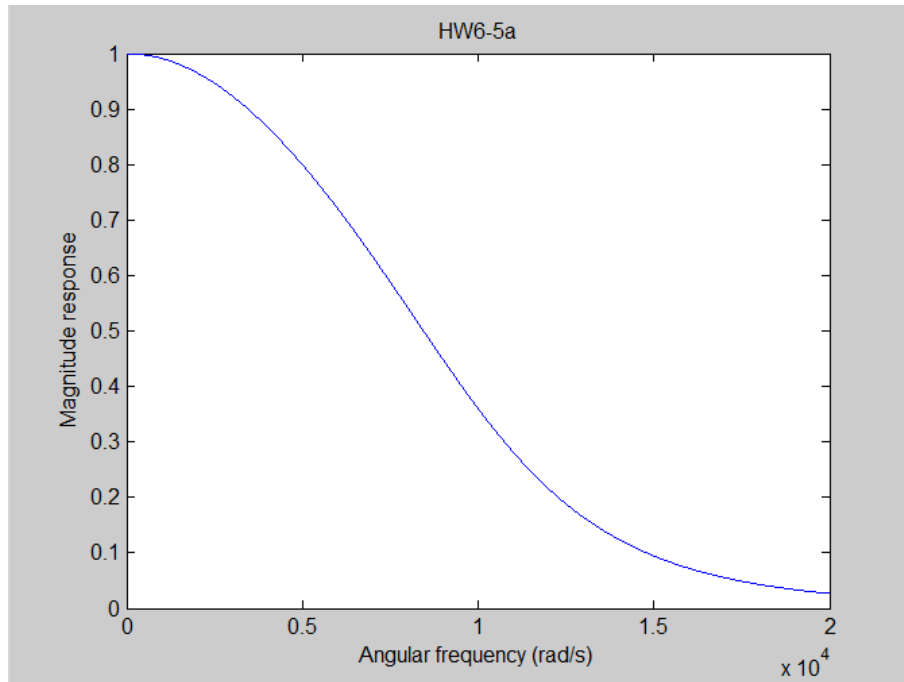


Figure 5: HW3-6a

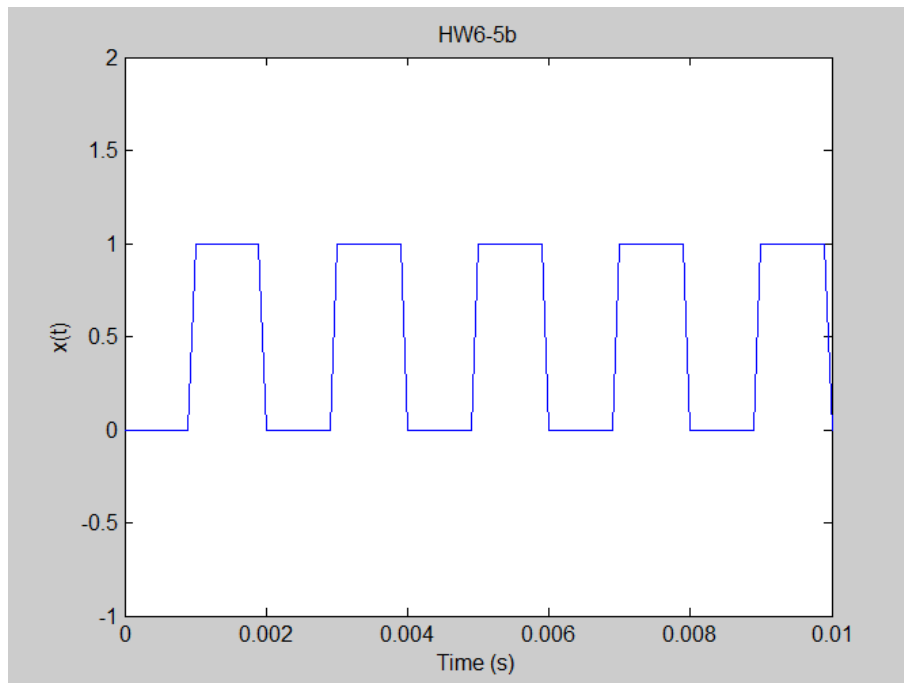


Figure 6: HW3-6b

7. If the FS coefficients of $x(t)$ are periodic with period N , then $a_k = a_{k-N}$ for all k . This implies that $x(t) = x(t)e^{j(2\pi/T)Nt}$ for all t . This is possible only if $x(t)$ is zero for all t other than when $(2\pi/T)Nt = 2\pi k$, where $k \in \mathbb{Z}$. Therefore, $x(t)$ must be in the form $x(t) = \sum_{k=-\infty}^{\infty} g[k]\delta(t - kT/N)$. On the other hand,

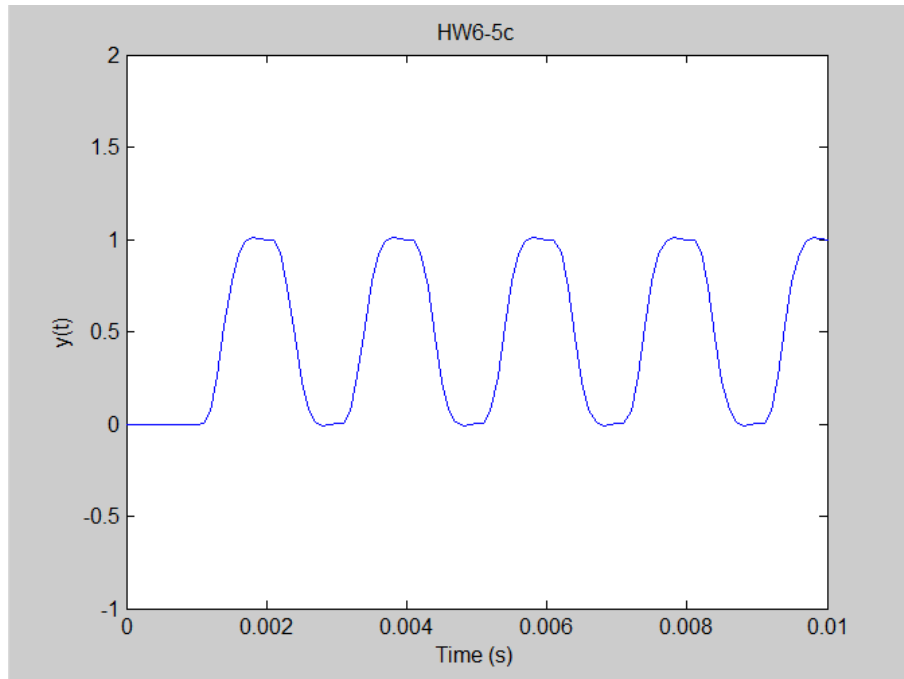


Figure 7: HW3-6c

because $x(t + NT/N) = x(t + T) = x(t)$ from the periodicity of $x(t)$, we easily obtain $g[k + N] = g[k]$ for all k .

8. .

$$\sin^5 x = (1/(2j))^5 (e^{jx} - e^{-jx})^5 = \frac{1}{32j} (e^{j5x} - 5e^{j3x} + 10e^{jx} - 10e^{-jx} + 5e^{-j3x} - e^{-j5x}) = \frac{10 \sin x - 5 \sin(3x) + \sin(5x)}{16}.$$

Thus $y(t) = 7 \left[\sin 3t + \frac{10b}{16} \sin 3t - \frac{5b}{16} \sin 9t + \frac{b}{16} \sin 15t \right]$. So $c_1 = 7(1 + 10b/16)$.

Thus THD = $1 - \frac{(1+10b/16)^2}{(1+10b/16)^2 + (5b/16)^2 + (b/16)^2} 100\% = \boxed{0.0899\%}$ for $b = 0.10$.