Introduction

Euklidian Norm: $\|x\|_2 = \sqrt{\sum_{i=1}^{n} x_i^2} = \sqrt{x^T x}$

$$\|x\|_2^2 = x^T \cdot x$$

Weighting Eukl. Norm: $\|x\|_Q^2 = x^T Q \cdot x$

Frobenius Norm: $\|x\|_F^2 = trace(AA^T) = \sum_{i=1}^{n} \sum_{j=1}^{m} A_{ij} A_{ij}$

Jacobian: $\nabla f(x)$ \qquad Hessian: $\nabla^2 f(x)$

- -

Error in variables: $\hat{R}_{EV}(N) = \dfrac{\frac{1}{N} \sum_{k=1}^{N} u(k)}{\frac{1}{N} \sum_{k=1}^{N} i(k)}$

Simple Approach: $\hat{R}_{SA}(N) = \dfrac{1}{N} \cdot \sum_{k=1}^{N} \dfrac{u(k)}{i(k)}$

Least Squares: $\hat{R}_{LS}(N) = \underset{R \in \mathbb{R}}{arg\,min} \sum_{k=1}^{N} (R \cdot i(k) - u(k))^2$

$$= \dfrac{\frac{1}{N} \sum_{k=1}^{N} u(k) \cdot i(k)}{\frac{1}{N} \sum_{k=1}^{N} i(k)^2}$$

Matrix derivates: $\dfrac{d(c^T x)}{dx} = c$ \qquad $\dfrac{d(x^T A x)}{dx} = (A^T + A)x$

Linear and non-linear models:
- linear if parameters linear i.e. $(\theta_1 x^2 + \theta_2 x + \theta_3)$
- nonliniar if i.e $(\sin(\theta_1)x + \theta_2)$

TODO check that... and the derivative of tan(x) please too.

Table of Derivatives:

$\mathbf{f(x)}$	$\mathbf{f'(x)}$
$g(x) \cdot h(x)$	$g'(x) \cdot h(x) + g(x) \cdot h'(x)$
$g(h(x))$	$g'(h(x)) \cdot h'(x)$
$\sin(x)$	$-\cos(x)$
$\cos(x)$	$\sin(x)$
$\tan(x)$	$-\ln(|\cos(x)|)$
e^{kx}	$\frac{1}{k} e^{kx}$
$\ln(x)$	$\frac{1}{x}$
$\log_a x$	$\frac{1}{\ln a}(x \ln x - x)$

Random Variables and Probability

Dependent Probability: $P(A \vee B) = P(A) + P(B)$

Independent Prob.: $P(A, B) = P(A \wedge B) = P(A) \cdot P(B)$

Conditional Prob.: $P(A|B) = \dfrac{P(A|B)}{P(B)} = \dfrac{P(B|A) \cdot P(A)}{P(B)}$

$$P(X \in [a,b]) = \int_a^b p_X(x)dx \qquad p(x|y) = \dfrac{p(x,y)}{p(y)}$$

Mean/Expectation value: $\mathbb{E}\{\mu_X\} := \mu_X = \int_{-\infty}^{\infty} x \cdot p_X(x)dx$

$$\mathbb{E}\{a + bX\} := a + b\mathbb{E}\{X\}$$

Variance: $\sigma_X^2 := \mathbb{E}\{(X - \mu_X)^2\} = \mathbb{E}\{X^2\} - \mu_X^2$

Standard deviation: $\sigma_X = \sqrt{\sigma_X^2}$

Distributions

Uniform distribution: $P_y(x) = \begin{cases} \frac{1}{b-a} & if \quad x \in [a,b] \\ 0 & else \end{cases}$

Mean: $\int_{-\infty}^{\infty} x \, p_X(x)dx = \int_a^b \dfrac{1}{b-a} \cdot x dx = \dfrac{a+b}{2} =: \mu_X$

Normal distribution: $X \sim \mathcal{N}(\mu, \sigma^2)$

$$p(x) = \dfrac{1}{\sqrt{2\pi\sigma^2}} \cdot exp(-\dfrac{(x-\mu)^2}{2\sigma^2})$$

Multidimensional Normal Distribution:

$$p(x) = \dfrac{1}{\sqrt{(2\pi)^n \cdot det(\Sigma)}} \cdot exp(-\dfrac{1}{2} \cdot (x-\mu)^T \cdot \Sigma^{-1} \cdot (x-\mu))$$

Weibull distribuation: $F(x) = 1 - e^{-(\lambda \cdot x)^k}$

Laplace distribuation: $f(x|\mu, b) = \dfrac{1}{2b} \cdot exp\left(-\dfrac{|x-\mu|}{b}\right)$

Useful statistic definitions

Covariance and Correlaton: $\sigma(X, Y) := \mathbb{E}(X - \mu_X)(Y - \mu_Y)$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} (X - \mu_X)(y - \mu_Y) \cdot p_{X,Y}(x,y) \, dx \, dy$$

Covariance Matrix: $\Sigma_x = cov(X) = \mathbb{E}\{XX^T\} - \mu_x \mu_x^T$

Multidimensional Random Variables:

$$\mathbb{E}f(X) = \int_{\mathbb{R}^n} f(x)p_X(x)d^n x$$

$$cov(X) = \mathbb{E}\{(X - \mu_X)(X - \mu_X)^T\}$$

$$cov(X) = \mathbb{E}\{XX^T\} - \mu_X \mu_X^T$$

$$cov(Y) = \Sigma_y = A\Sigma_x A^T \quad for \quad y = A \cdot x$$

$$\mathbb{E}\{AX\} = A \cdot \mathbb{E}\{X\}$$

Rules for variance:

$$var(aX) = a^2 \cdot var(X)$$

$$var(X + Y) = var(X) + var(Y) + 2 \cdot cov(X, Y)$$

Verschiebesatz: $var(X) = \mathbb{E}((X - \mathbb{E}(X))^2) = \mathbb{E}(X^2) - (\mathbb{E}(X))^2$

Unit Variance is variance = 1

Was sagt mir das?

- -

Statistical estimators:

Biased- and Unbiasedness An estimator $\hat{\theta}_N$ is called unbiased iff $\mathbb{E}\{\hat{\theta}_N(y_N)\} = \theta_0$, where θ_0 is the true value of a parameter. Otherwise, is called biased.

Asymptotic Unbiasedness An estimator $\hat{\theta}_N$ is called asymptotically unbiased iff $\lim\limits_{n \to \infty} \mathbb{E}\{\hat{\theta}_N(y_N)\} = \theta_0$

Consistency An estimator $\hat{\theta}_N(y_N)$ is called consistent if, for any $\epsilon > 0$, the probability $P(\hat{\theta}_N(y_N) \in [\theta_0 - \epsilon, \theta_0 + \epsilon])$ tends to one as $N \to \infty$.

Unconstrainded Optimization

Theorem 1: (First Order Necessary Conditions)
If $x^* \in D$ is local minimizer of $f : D \to \mathbb{R}$ and $f \in C^1$ then $\nabla f(x^*) = 0$ Definition (Stationary Point) A point $\bar{x}$ with $\nabla f(\bar{x}) = 0$ is called a stationary point of f.

Theorem 2: (Second Order Necessary Conditions)
If $x^* \in D$ is local minimizer of $f : D \to R$ and $f \in C^2$ then $\nabla^2 f(x^*) \succeq 0$

Theorem 3: (Second Order Sufficient Conditions and Stability under Perturbations)
Assume that $f : D \to R$ is C^2. If $x^* \in D$ is a stationary point and $\nabla^2 f(x^*) \succ 0$ then x^* is a strict local minimizer of f. In addition, this minimizer is locally unique and is stable against small perturbations of f, i.e. there exists a constant C such that for sufficiently small $p \in \mathbb{R}^n$ holds

$$\|x^* - \underset{x}{arg\,min}(f(x) + p^T x)\| \leq C\|p\|$$

Linear Least Squares Estimation

Preliminaries: i.i.d. and Gaussian noise

Overall Model: $y(k) = \phi(k)^T \theta + \epsilon(k)$

LS cost function as sum: $\sum_{k=1}^{N} \left(y(k) - \phi(k)^T \theta\right)^2$

LS cost function: $f(\theta) = \|y_N - \Phi_N \theta\|_2^2$

Unique minimizers: $\hat{\theta}_{LS} = \underset{\theta \in \mathbb{R}}{arg\,min}\, f(\theta) \quad \theta^* = \underbrace{(\Phi^T\Phi)^{-1}\Phi^T}_{\Phi^+} y$

Pseudo Inverse: $\Phi^+ = (\Phi^T\Phi)^{-1}\Phi^T$

Weighted Least Squares (unitless)

For i.i.d noise: Unweight Least Squares is optimal: $W = I$

$$f_{WLS}(\theta) = \sum_{k=1}^{N} \frac{(y(k) - \phi(k)^T\theta)^2}{\sigma_\epsilon^2(k)} = \|y_N - \Phi_N\theta\|_W^2$$

$$= (Y_N - \Phi \cdot \theta)^T \cdot W \cdot (Y_N - \Phi \cdot \theta)$$

Solution for WLS:
$$\hat{\theta}_{WLS} = \tilde{\Phi}^+ \tilde{y} \qquad \text{mit } \tilde{\Phi} = W^{\frac{1}{2}}\Phi \text{ und } \tilde{y} = W^{\frac{1}{2}}y$$

$$= \underset{\theta \in \mathbb{R}}{arg\,min}\, f_{WLS}(\theta) = (\Phi^T W \Phi)^{-1}\Phi^T W y$$

Ill-Posed Least Squares

Singular Value Decomposition: $A = USV^T \in \mathbb{R}^{mxn}$
with $U \in \mathbb{R}^{mxm}$, $V \in \mathbb{R}^{nxn}$ and $S \in \mathbb{R}^{mxn}$ where S is a Matrix with non-negative elements $(\sigma_1, \ldots, \sigma_r, 0, \ldots, 0)$ on the diagonal and 0 everywhere else.

Moore Penrose Pseudi Inverse:
$$\Phi^+ = VS^+U^T = V(S^TS + \alpha I)^{-1}S^TU^T$$

Φ^+ therefore selects $\theta^* \in S^*$ with minimal norm.

Regularization for Least Squares:
$$\lim_{a \to 0}(\Phi^T\Phi + \alpha I)^{-1}\Phi^T = \Phi^+ \quad \text{with } \Phi^+ \, MPPI$$

$$\theta^*(\alpha) = \underset{\theta \in \mathbb{R}}{arg\,min}\, \frac{1}{2}\|y - \Phi\theta\|_2^2 + \frac{\alpha}{2}\|\theta\|_2^2$$

Stimmt das auch so mit dem α? Ich find das nirgends so.

Statistical Analysis of WLS

Expectation of Least Squares Estimator:
$$E\{\hat{\theta}_{WLS}\} = E\{(\Phi_N^T W \Phi_N)^{-1}\Phi_N^T W y_N\} = \theta_0$$

Covariance of the least squares estimator:
$$cov(\hat{\theta}_{WLS}) = (\Phi_N^T W \Phi_N)^{-1} = (\Phi_N^T \Sigma_{\in N}^{-1} \Phi_N)^{-1}$$
$$cov(\hat{\theta}_{WLS}) \succeq (\Phi_N^T W \Phi_N)^{-1}$$

Example LLS

Example of the Linear Least Square Estimator for:
$N = 2$
$$\varepsilon(1) \sim \mathcal{N}(0|\sigma_1^2) \qquad\qquad \varepsilon(2) \sim \mathcal{N}(0|\sigma_2^2)$$

$$N = 2; \quad \Sigma_{\varepsilon_N} = \begin{bmatrix} \sigma_1^2 & 0 \\ 0 & \sigma_2^2 \end{bmatrix} \qquad W^{OPT} = \Sigma_{\varepsilon_N}^{-1} = \begin{bmatrix} \frac{1}{\sigma_1^2} & 0 \\ 0 & \frac{1}{\sigma_2^2} \end{bmatrix}$$

$$cov(\hat{\theta}_{WLS}) = (Y_N - \Phi_N\theta)^T \cdot W \cdot (Y_N - \Phi_N\theta)$$
$$= \sum_{k=1}^{2}(y(k) - \phi(k)^T\theta) \cdot \frac{1}{\sigma_k^2} \cdot (y(k) - \phi(k)^T\theta)$$

Measuring the goodness of Fit using: $R^2 \quad (0 \le R^2 \le 1)$

$$R^2 = 1 - \frac{\|y_N - \Phi_N\hat{\theta}\|_2^2}{\|y_N\|_2^2} = 1 - \frac{\|\epsilon_N\|_2^2}{\|y_N\|_2^2}$$

$$= \frac{\|y_N\|_2^2 - \|\epsilon_N\|_2^2}{\|y_N\|_2^2} = \frac{\|\hat{y}_N\|_2^2}{\|y_N\|_2^2}$$

Residual: $\epsilon_N \uparrow \to \quad R^2 \to 0 \ (\Rightarrow bad)$

- -

Estimating the Covariance with the Single Experiment:

$$\hat{\sigma}_\varepsilon^2 := \frac{1}{N-d}\sum_{k=1}^{N}(y(k) - \phi(k)^T\hat{\theta}_{LS})^2 = \frac{\|y_N - \phi_N\hat{\theta}_{LS}\|_2^2}{N-d}$$

$$\hat{\Sigma}_{\hat{\theta}} := \hat{\sigma}_\varepsilon^2(\phi_N^T\phi_N)^{-1} = \frac{\|y_N - \phi_N\hat{\theta}_{LS}\|_2^2}{N-d} \cdot (\phi_N^T\phi_N)^{-1}$$

Maximum Likelihood Estimation

Maximum Likelihood Estimation (ML) L_2 **Estimation:**
- Measurement Errors assumed to be Normally distributed

  Ist das wirklich so ?

- Model described by a non-linear function $M(\theta)$
- Every unbiased estimator needs to satisfy the Cramer-Rau inequality, which gives a lower bound on the covariance matrix

  Model: $\quad y = M(\theta) + \epsilon$

$$P(y|\theta) = C\prod_{i=1}^{N}\exp(\frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}) \qquad C = \prod_{i=1}^{N}\frac{1}{\sqrt{2 \cdot \pi\sigma_i^2}}$$

Positive log-Likelihood. Logarithm makes from products a sum!

$$\log p(y|\theta) = \log(C) + \sum_{i=1}^{N}-\frac{(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}$$

Negative log-Likelihood:

$$\hat{\theta}_{ML} = \underset{\theta \in \mathbb{R}^d}{arg\,max}\, p(y|\theta) = \underset{\theta \in \mathbb{R}^d}{arg\,min}\sum_{i=1}^{N}\frac{(y_i - M_i(\theta))^2}{2\,\sigma_i^2}$$

$$= \underset{\theta}{arg\,min}\,\frac{1}{2}\sum_{i=1}^{N}\left(\frac{y_i - M_i(\theta)}{\sigma_i}\right)^2 = \underset{\theta \in \mathbb{R}^d}{arg\,min}\,\frac{1}{2}\|S^{-1}(y - M(\theta))\|_2^2$$

$$\text{mit:} \quad S = \begin{bmatrix} \sigma_1 & & \\ & \ddots & \\ & & \sigma_N \end{bmatrix}$$

L_1 **Estimation:**
Measurement Errors assumed to be Laplace distributed and more robust against outliers.

$$\underset{\theta}{min}\,\| y - M(\theta) \|_1 = \underset{\theta}{min}\sum_{i=1}^{N}|y_i - M_i(\theta)| = \text{ median of } \{Y_1, ..., Y$$

Bayesian Estimation and the Maximum a Posteriori Estimate

Assumptions:
- Measurement $\quad y_N \in \mathbb{R}^N \quad$ has i.i.d noise
- Linear Model $\quad M(\theta) = \phi_N \cdot \theta \quad$ and $\theta \in \mathbb{R}$

$$p(\theta|y_N) = \frac{p(y_N, \theta)}{p(y_N)} = \frac{p(y_N|\theta) \cdot p(\theta)}{p(y_N)}$$

$$\hat{\theta}_{MAP} = \underset{\theta \in \mathbb{R}}{arg\,min}\{-\log(p(y_N|\theta)) - \log(p(\theta))\}$$

MAP Example: Regularised Least Squares
$$\theta = \overline{\theta} \pm \sigma_\theta \quad \text{with} \quad \overline{\theta} = \theta_{\text{a-priori}}$$

$$\hat{\theta}_{MAP} = \underset{\theta \in R}{arg\,min}\,\frac{1}{2} \cdot \frac{1}{\sigma_{\epsilon^2}} \cdot \| y_N - \Phi_N \cdot \theta \|_2^2 + \frac{1}{2} \cdot \frac{1}{\sigma_\theta^2} \cdot (\theta - \overline{\theta})^2$$

Recursive Linear Least Squares

$$\theta_{ML}(N) = \arg\min_{\theta \in \mathbb{R}} \frac{1}{2} \parallel y_N - \Phi_N \cdot \theta \parallel_2^2$$

$$\hat{\theta}_{ML}(N+1) = \arg\min_{\theta \in R^d} \left(\alpha \cdot \frac{1}{2} \cdot \parallel \theta - \hat{\theta}_{ML}(N) \parallel_{Q_N}^2 \right.$$

$$\left. + \frac{1}{2} \cdot \parallel y(N+1) - \varphi(N+1)^T \cdot \theta \parallel_2^2 \right)$$

Q_0 given, and $\hat{\theta}_{ML}(0)$ given

$Q_{N+1} = \alpha \cdot Q_N + \varphi(N+1) \cdot \varphi(N+1)^T,$

$\hat{\theta}_{ML}(N+1) = \hat{\theta}_{ML}(N) + Q_{N+1}^{-1} \cdot \varphi(N+1) \cdot [y(N+1) -$

$\varphi(N+1)^T \cdot \hat{\theta}_{ML}(N)]$

Cramer-Rao-Inequality (Fisher information Matrix M)

$$\Sigma_{\hat{\theta}} \succeq M^{-1} = (\Phi_N^T \cdot \Sigma^{-1} \cdot \Phi)^{-1}$$

Assumptions: - Minimising a Linear Model

         - Gaussian Noise: $X \sim \mathcal{N}(0, \Sigma)$

$L(\theta, y_N) = -\log(p(y_N|\theta))$

$L(\theta, y_N) = \frac{1}{2} \cdot (\Phi_N \cdot \theta - y_N)^T \cdot \Sigma^{-1} \cdot (\Phi_N \cdot \theta - y_N)$

$M = \mathbb{E}\{\nabla_\theta^2 L(\theta, y_N)\} = \nabla_\theta^2 L(\theta, y_N) = \Phi_N^T \cdot \Sigma^{-1} \cdot \Phi_N$

Confirms that $W = \Sigma^{-1}$ is the optimal weighting Matrix for WLS. ∎

Linear Time Invariant (LTI) Systems

With A, B, C, D are matrices

$$\dot{x} = Ax + Bu \quad y = Cx + Du$$

$$G(s) = C(sI - A)^{-1} B + D$$

LTI sytems as Input-Output Models:

$$G(S) = \frac{b_0 + b_1 s + \dots + b_n s^n}{a_0 + a_1 s + \dots + a_{n-1} s^{n-1} + s^n}$$

Deterministic Models

> Hier fehlt noch etwas

$$y(k) = M(k; U, x_{init}, p)$$

Finite Impulse Response (FIR):

$$G(z) = b_0 + b_1 z^{-1} + \dots + b_{n_b} z^{-n_b} = \frac{b_0 z^{n_b} + b_1 z^{n_b-1} + \dots + b_{n_b}}{z^{n_b} + 0 + \dots + 0}$$

Auto Regressive Models with Exogenus Inputs (ARX):

$$G(z) = \frac{b_0 z^n + b_1 z^{n-1} + \dots + b_n}{a_0 z^n + a_1 z^{n-1} + \dots + a_n}$$

Stochastik Models

Model with measurement Noise:
$y(k) = M(k; U, x_{init}, p) + \varepsilon(k)$

Liniar In the Parameters models (LIP):

$$y(k) = \sum_{i=1}^d \theta_i \phi_i(u(k)..., y(k-1), ...) + \epsilon(k)$$

$\to y(k) = \varphi(k)^T \theta + \epsilon(k)$ for $\varphi = (phi_1(\cdot), ..., \phi_d(\cdot))$

LIP-LTI Models with Equation Errors (ARX)
combining best of two worlds (LTI and LIP)

$$a_0 y(k) + \dots + a_{n_a} y(k - n_a) = b_0 u(k) + \dots + b_{n_b} u(k - n_b) + \epsilon(k)$$

Model with Input and Output Errors

$$y(k) = M(k; U + \varepsilon_N^u, x_{init}, p) + \varepsilon^y(k)$$

```
model.pdf
```

Pure Output Error (OE) Minimization

$$\theta_{ML} = \min_\theta \sum_{k=1}^N (y(k) - M(k; U, x_{init} p))^2$$

Output Error Minimization for FIR Models:
$y(k) = (u(k), u(k-1), \cdots, u(k - n_b)) \cdot \theta + \varepsilon(k)$

$$= \min_\theta \sum_{k=n_b+1}^N (y(k) - (u(k), u(k-1), \cdots, u(k - n_b)) \cdot \theta)^2$$

Models with Input and Output Errors:

$$arg\min_\theta \sum_{k-1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; U + \epsilon_N^u, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (\epsilon_u(k))^2$$

$$arg\min_\theta \sum_{k-1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; \tilde{U}, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (u(k) - \tilde{u}(k))^2$$

Fourier Transformation

> Wirklich notwendig?

How to compute FT? By DFT, which solves the problem of finite time and discrete values.

Can we use an input with many frequencies to get many FRF (Frequency Response Function) values in a single experiment? So far only frequency sweeping (high comp. times due to repetition for each frequency). We should use multisines!

Aliasing and Leakage Errors

Aliasing Error: Due to sampling of continous signal to discrete signal. Avoid with Nyquist Theoreme:

$$f_{Nyquist} = \frac{1}{2\Delta t} [\text{Hz}] \quad or \quad \omega_{Nyquist} = \frac{2\pi}{2\Delta t} [\text{rad/s}]$$

Leackage Error: Due to windowing.

$$\omega_{base} := \frac{2\pi}{N \cdot \Delta t} = \frac{2\pi}{T} \to \omega = m \frac{2\pi}{N \cdot \Delta t}$$

Crest Factor = Scheitelfaktor

$$\text{Crest Factor} = \frac{u_{max}}{u_{rms}} \quad \text{with} : u_{rms} := \sqrt{\frac{1}{T} \int_0^T u(t)^2 \, dt}$$

$$\text{and} \quad u_{max} := \max_{t \in [0,T]} |u(t)|$$
∎

Optimising Multisine for optimal crest factor

Frequency: Choose frequencies in logarithmic manner as multiples of the base frequency. $\omega_{k+1}/\omega_k \approx 1.05$

Phase: To prevent high peaks (Crest Factor) in the Signal, the phases of the different frequencies are modulated accordingly. (Positive interference)

Multisine Identification Implementation procedure

Window Length: Integer multiple of sampling time: $T = N \cdot \delta t$

Harmonics of base frequency: Are contained in multisine
$\omega_{base} = \frac{2\pi}{T}$

Highest contained Frequency: Is **half** of Nyquist frequency:
$\omega_{Nyquist} = \frac{2\pi}{4\Delta T}$

Experiment and Analysis: (Step 2): Insert Multisine periodically. Drop first Periods (till transients died out). Record M Periods, each with N samples, of input and output data. Average all the M periods and make the DFT (or vice versa). Finally build transfer function: $\hat{G}_{j\omega_k} = \frac{\hat{Y}(k(p))}{\hat{U}(k(p))}$

Nonparametric and Frequency Domain Identification Models

Impulse response and transfer function:

$$y(t) = \int_0^\infty g(\tau) u(t - \tau) \, \delta t$$

$$Y(s) = G(s) \cdot U(s)$$

$$G(s) = \int_0^\infty \epsilon^{-st} g(t) \, dt$$

Bode diagram from frequency sweeps:

$$u(t) = A \cdot sin(\omega \cdot t), \quad y(t) = \|G(j \cdot \omega)\| A \cdot sin(\omega \cdot t + \alpha)$$

Bode Diagramm

Magnitude = Amplitude $|G(j\omega)|$
Phase $arg\, G(j\omega)$

Hier sollten wir glaub noch bisschen was rein machen.

Recursive Least Squares

New Inverse Covariance: $Q_K = Q_{k-1} + \phi_K \phi_K^T$

Innovation update: $\hat{\theta}_k = \hat{\theta}_{k-1} + \underbrace{Q_k^{-1} \phi_k (y_k - \phi_k^T \hat{\theta}_{k-1})}_{\text{„innovation"}}$

General Optimization Problem:

$$\hat{\theta}_k = \underset{\theta}{arg\, min}(\theta - \hat{\theta}_0)^T \cdot Q_0 \cdot (\theta - \hat{\theta}_0) + \sum_{i=1}^{k}(y_k - \phi_k^T \cdot \theta)^2$$

Kalman Filter

Valid for Discrete and Linear!

If recursive least squares: $x_{k+1} = A_k \cdot x_k$

$x_{k+1} = A_k \cdot x_k + \omega_k$ and $y_k = C_k \cdot x_k + v_k$

Steps of Kalman Filter

1 Prediction

$$\hat{x}_{[k|k-1]} = A_{k-1} \cdot \hat{x}_{[k-1|k-1]}$$

$$P_{[k|k-1]} = A_{k-1} \cdot P_{[k-1|k-1]} \cdot A_{k-1}^T \cdot W_{k-1}$$

If RLS, without: W_{k-1}

2 Innovation update

$$P_{[k|k]} = (P_{[k|k-1]}^{-1} + C_k^T \cdot V^{-1} \cdot C_k)^{-1}$$

$$\hat{x}_{[k|k]} = \hat{x}_{[k|k-1]} + P_{[k|k]} \cdot C_k^T \cdot V^{-1} \cdot (y_k - C_k \cdot \hat{x}_{[k|k-1]})$$

Ein Beispiel einfuegen.