

Introduction

Euklidische Norm:

|| x ||_2 = sqrt(sum_{i=1}^n x_i^2) = sqrt(x^T x)

|| x ||_2^2 = x^T . x

Weighting Eukledian Norm:

|| x ||_Q^2 = x^T Q . x

Frobenius Norm:

|| x ||_F^2 = trace(AA^T) = sum_{i=1}^n sum_{j=1}^m A_{ij} A_{ij}

∇ f(x) Jacobian

∇^2 f(x) Hessian

Error in variables

Ŕ_{ev}(N) = (1/N * sum_{k=1}^N u(k)) / (1/N * sum_{k=1}^N i(k))

Matrix derivatives

d(c^T x) / dx = c

d(x^T A x) / dx = (A^T + A) x

Linear and non-linear equations: TODO; polynomial etc.

Probablility and Statistics

Random Variables and Probability

P(A|B) · P(B) = P(B|A) · P(A)

P(A|B) = P(A, B) / P(B) → P(B, A) · P(A) / P(B)

P(X ∈ [a, b]) = ∫_a^b p_X(x) dx

Mean

μ_X = E{f(x)} := ∫_{-∞}^∞ f(x) · p_X(x) dx

E{a + bX} := a + bE{X}

Variance

σ_X^2 := E{(X - μ_X)^2} = E{X^2} - μ_X^2

stddev σ_X = sqrt(variance σ_X^2)

Distributions

Uniform distribution:

P_y(x) = { 1/(b-a) if x ∈ [a, b], 0 else }

Normal (Gaussian) distribution:

p(x) = 1 / sqrt(2πσ^2) · exp(-(x - μ)^2 / 2σ^2)

X ~ N(μ, σ^2)

Multidimensional Normal Distribution:

p(x) = 1 / sqrt((2π)^n · det(Σ)) · exp(-1/2 · (x - μ)^T · Σ^{-1} · (x - μ))

Weibull distribution:

F(x) = 1 - e^{-(λ · x)^k}

Laplace distribution:

f(x|μ, b) = 1/(2b) · exp(-|x - μ|/b)

Covariance and Correlaton:

σ(Y, Z) := E(Y - μ_Y)(Z - μ_Z) =

= ∫_{-∞}^∞ ∫_{-∞}^∞ (y - μ_Y)(z - μ_Z) · p_{Y,Z}(y, z) dy dz

Multidimensional Random Variables:

E f(X) = ∫_{R^n} f(x) p_X(x) d^n x

cov(X) = E{(X - μ_X)(X - μ_X)^T}

cov(X) = E{X X^T} - μ_X μ_X^T

cov(Y) = Σ_y = A Σ_x A^T for y = A · x

E{AX} = A · E{X}

Rules for variance:

var(X + Y) = var(X) + var(Y) + 2 · cov(X, Y)

var(aX) = a^2 · var(X)

Verschiebesatz:

var(X) = E((X - E(X))^2) = E(X^2) - (E(X))^2

unit Variance is variance = 1

Statistical estimators:
Biased- and unbiasedness → an estimator $\hat{\theta}_N$ is called unbiased iff $E\{\hat{\theta}_N(y_N)\} = \theta_0$, where θ_0 is the true value of a parameter. Otherwise, is called biased.

Asymptotic Unbiasedness → An estimator $\hat{\theta}_N$ is called asymptotically unbiased iff $\lim_{n \rightarrow \infty} E\{\hat{\theta}_N(y_N)\} = \theta_0$

Consistency → An estimator $\hat{\theta}_N(y_N)$ is called consistent if, for any $\epsilon > 0$, the probability $P(\hat{\theta}_N(y_N) \in [\theta_0 - \epsilon, \theta_0 + \epsilon])$ tends to one as $N \rightarrow \infty$.

Unconstrained Optimization

Theorem 1 (First Order Necessary Conditions)
If $x^* \in D$ is local minimizer of $f : D \rightarrow \mathbb{R}$ and $f \in C^1$ then $\nabla f(x^*) = 0$
Definition (Stationary Point) A point \bar{x} with $\nabla f(\bar{x}) = 0$ is called a stationary point of f.

Theorem 2 (Second Order Necessary Conditions)
If $x^* \in D$ is local minimizer of $f : D \rightarrow R$ and $f \in C^2$ then $\nabla^2 f(x^*) \succeq 0$

Theorem 3 (Second Order Sufficient Conditions and Stability under Perturbations)
Assume that $f : D \rightarrow R$ is C^2 . If $x^* \in D$ is a stationary point and $\nabla^2 f(x^*) \succ 0$ then x^* is a strict local minimizer of f. In addition, this minimizer is locally unique and is stable against small perturbations of f, i.e. there exists a constant C such that for sufficiently small $p \in \mathbb{R}^n$ holds

|| x* - arg min_x (f(x) + p^T x) || ≤ C || p ||

Linear Least Squares Estimation

Preliminaries: I.I.D and gaussian noise

Overall Model

y(k) = ϕ(k)^T θ + ϵ(k)

Least Squares cost function as sum

sum_{k=1}^N (y(k) - ϕ(k)^T θ)^2

Least Squares cost function

f(θ) = || y_N - Φ_N θ ||_2^2

Unique minimizers

θ_LS = arg min_{θ ∈ ℝ} f(θ)

θ* = (Φ^T Φ)^{-1} Φ^T y

Pseudo Inverse: Φ+ = (Φ^T Φ)^{-1} Φ^T

Weighted Least Squares (unitless)
For I.I.D noise: Unweight Least Squares is optimal: W=I

sum_{k=1}^N (y(k) - ϕ(k)^T θ)^2 / σ_ϵ^2(k)

f_WLS(θ) = || y_N - Φ_N θ ||_W^2 = (Y_N - Φ · θ)^T · W · (Y_N - Φ · θ)

θ_WLS = arg min_{θ ∈ ℝ} f_WLS(θ) = (Φ^T W Φ)^{-1} Φ^T W y

Singular Value Decomposition

A = U S V^T mit U ∈ ℝ^{m × m}, V ∈ ℝ^{n × n} und S ∈ ℝ^{m × n}

Moore Penrose Pseudo Inverse

Φ+ = V S+ U^T

Regularization for Least Squares
lim_{α → 0} (Φ^T Φ + α I)^{-1} Φ^T = Φ+ with Φ+ MPPI

θ*(α) = arg min_{θ ∈ ℝ} 1/2 || y - Φ θ ||_2^2 + α/2 || θ ||_2^2

Expectation of Least Squares Estimator

E{θ_WLS} = E{(Φ_N^T W Φ_N)^{-1} Φ_N^T W y_N} = θ_0

Covariance of the least squares estimator

cov(θ_WLS) = (Φ_N^T W Φ_N)^{-1} = (Φ_N^T Σ_{ε_N}^{-1} Φ_N)^{-1}

cov(θ_WLS) ⪰ (Φ_N^T W Φ_N)^{-1}

Example:

ε(1) ~ N(0|σ_1^2) ε(2) ~ N(0|σ_2^2)

N = 2 Σ_{ε_N} = [σ_1^2 0; 0 σ_2^2]

W^{OPT} = Σ_{ε_N}^{-1} [1/σ_1^2 0; 0 1/σ_2^2]

cov(θ_WLS) = (Y_N - Φ_N θ)^T · W · (Y_N - Φ_N θ) =

sum_{k=1}^2 (y(k) - ϕ(k)^T θ) · 1/σ_k^2 · (y(k) - ϕ(k)^T θ)

Measuring the goodness of Fit using R^2 0 ≤ R^2 ≤ 1

R^2 = 1 - || y_N - Φ_N θ̂ ||_2^2 / || y_N ||_2^2 = 1 - || ε_N ||_2^2 / || y_N ||_2^2 = (|| y_N ||_2^2 - || ε_N ||_2^2) / || y_N ||_2^2 = || ŷ_N ||_2^2 / || y_N ||_2^2

residual ε_N ↑ → R^2 → 0 (= bad)

Estimating the Covariance with the Single Experiment

σ_ε^2 := 1/(N - d) sum_{k=1}^N (y(k) - ϕ(k)^T θ_LS)^2 = || y_N - Φ_N θ_LS ||_2^2 / (N - d)

Σ_θ̂ := σ_ε^2 (Φ_N^T Φ_N)^{-1} = || y_N - Φ_N θ_LS ||_2^2 / (N - d) · (Φ_N^T Φ_N)^{-1}

Maximum Likelihood Estimation

Maximum Likelihood Estimation (ML) L_2 Estimation:
Measurement Errors assumed to be Normally distributed

$$P(y|\theta) = C \prod_{i=1}^N \exp\left(\frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}\right)$$

Positive log-Likelihood. Logarithm makes from products a sum!

$$\log p(y|\theta) = \log(C) + \sum_{i=1}^N \frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}$$

Negative log-Likelihood:

$$\hat{\theta}_{ML} = \arg \max_{\theta \in \mathbb{R}^d} p(y|\theta) = \arg \min_{\theta \in \mathbb{R}^d} \sum_{i=1}^m \frac{(y_i - M_i(\theta))^2}{2 \sigma_i^2}$$

$$\arg \max_{\theta \in \mathbb{R}^d} p(y|\theta) = \arg \min_{\theta \in \mathbb{R}^d} \frac{1}{2} \|S^{-1} \cdot (y - M(\theta))\|_2^2$$

L_1 Estimation:

Measurement Errors assumed to be Laplace distributed.

$Median(x) = \lceil \frac{x+1}{2} \rceil$

Robust against outliers

$$\begin{aligned} \min_{\theta} \|y - M(\theta)\|_1 &= \min_{\theta} \sum_{i=1}^N |y_i - M_i(\theta)| = \\ &= median\{Y_1, \dots, Y_N\} \end{aligned}$$

$$P(y|\theta) = C \prod_{i=1}^N \exp\left(\frac{-|y_i - \theta|}{2 \cdot a_i}\right)$$

Bayesian Estimation and the Maximum a Posteriori Estimate
Assumptions: i.i.d noise and linear model

$$p(\theta|y_N) \cdot p(y_N) = p(y_N|\theta) \cdot p(\theta)$$

$$\hat{\theta}_{MAP} = \arg \min_{\theta \in \mathbb{R}} \{-\log(p(y_N|\theta)) - \log(p(\theta))\}$$

MAP Example: Regularised Least Squares

$$\theta = \bar{\theta} \pm \sigma_{\theta} \quad \text{with} \quad \bar{\theta} = \theta_{\text{apriori}}$$

$$\hat{\theta}_{MAP} = \arg \min_{\theta \in \mathbb{R}} \frac{1}{2} \cdot \frac{1}{\sigma_{\epsilon}^2} \cdot \|y_N - \Phi_N \cdot \theta\|_2^2 + \frac{1}{2} \cdot \frac{1}{\sigma_{\theta}^2} \cdot (\theta - \bar{\theta})^2$$

Recursive Linear Least Squares

$$\theta_{ML}(N) = \arg \min_{\theta \in \mathbb{R}} \frac{1}{2} \|y_N - \Phi_N \cdot \theta\|_2^2$$

$$\begin{aligned} \hat{\theta}_{ML}(N+1) &= \arg \min_{\theta \in \mathbb{R}^d} \left(\alpha \cdot \frac{1}{2} \cdot \|\theta - \hat{\theta}_{ML}(N)\|_{Q_N}^2 + \right. \\ &\quad \left. \frac{1}{2} \cdot \|y(N+1) - \varphi(N+1)^T \cdot \theta\|_2^2 \right) \end{aligned}$$

Q_0 given, and $\hat{\theta}_{ML}(0)$ given,

$$Q_{N+1} = \alpha \cdot Q_N + \varphi(N+1) \cdot \varphi(N+1)^T,$$

$$\begin{aligned} \hat{\theta}_{ML}(N+1) &= \hat{\theta}_{ML}(N) + Q_{N+1}^{-1} \cdot \varphi(N+1) \cdot [y(N+1) - \\ &\quad \varphi(N+1)^T \cdot \hat{\theta}_{ML}(N)] \end{aligned}$$

Cramer-Rao-Inequality

$$\Sigma_{\theta} \succeq M^{-1} = (\Phi_N^T \cdot \Sigma^{-1} \cdot \Phi)^{-1}$$

$$L(\theta, y_N) = \frac{1}{2} \cdot (\Phi_N \cdot \theta - y_N)^T \cdot \Sigma^{-1} \cdot (\Phi_N \cdot \theta - y_N) = \log(p(y_N|\theta))$$

$$M = E\{\nabla_{\theta}^2 L(\theta, y_N)\} = \nabla_{\theta}^2 L(\theta, y_N) = \Phi_N^T \cdot \Sigma^{-1} \cdot \Phi_N$$

Dynamic Models

Linear Time Invariant (LTI) Systems

with A, B, C, D are matrices

$$\dot{x} = Ax + Bu \quad y = Cx + Du$$

$$G(s) = C(sI - A)^{-1}B + D$$

LTI sytems as Input-Output Models

$$G(S) = \frac{b_0 + b_1 s + \dots + b_n s^n}{a_0 + a_1 s + \dots + a_{n-1} s^{n-1} + s^n}$$

Different Models

Deterministic Model: $y(k) = M(k; U, x_{init}, p)$

Model with measurement Noise:

$$y(k) = M(k; U, x_{init}, p) + \varepsilon(k)$$

Model with Input and Output Errors:

$$y(k) = M(k; U + \varepsilon_N^u, x_{init}, p) + \varepsilon^y(k)$$

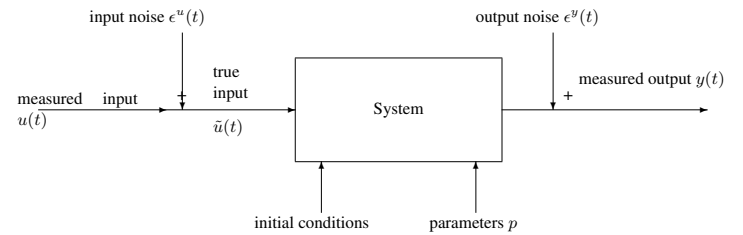
Pure Output Error (OE) Minimization

$$\theta_{ML} = \arg \min_{\theta} \sum_{k=1}^N (y(k) - M(k; U, x_{init}, p))^2$$

Output Error Minimization for FIR Models

$$y(k) = (u(k), u(k-1), \dots, u(k-n_b)) \cdot \theta + \varepsilon(k)$$

$$\min_{\theta} \sum_{k=n_b+1}^N (y(k) - (u(k), u(k-1), \dots, u(k-n_b)) \cdot \theta)^2$$



Models with Input and Output Errors

$$\arg \min_{\theta} \sum_{k=1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; U + \epsilon_N^u, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (\epsilon_u(k))^2$$

$$\arg \min_{\theta} \sum_{k=1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; \tilde{U}, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (u(k) - \tilde{u}(k))^2$$

Fourier Transformation

How to compute FT? By DFT, which solves the problem of finite time and discrete values.

Can we use an input with many frequencies to get many FRF (Frequency Response Function) values in a single experiment? So far only frequency sweeping (high comp. times due to repetition for each frequency). We should use multisines!

Aliasing and Leakage Errors

Aliasing Error due to sampling of continuous signal to discrete signal. Avoid with Nyquist Theorem:

$$f_{Nyquist} = \frac{1}{2\Delta t} [Hz] \quad \text{or} \quad \omega_{Nyquist} = \frac{2\pi}{2\Delta t} [rad/s]$$

Leakage Error due to windowing.

$$\omega_{base} := \frac{2\pi}{N \cdot \Delta t} = \frac{2\pi}{T} \rightarrow \omega = m \frac{2\pi}{N \cdot \Delta t}$$

Crest Factor = Scheitelfaktor

$$CrestFactor = \frac{u_{max}}{u_{rms}} \quad \text{with :}$$

$$u_{rms} := \sqrt{\frac{1}{T} \int_0^T u(t)^2 dt} \quad \text{and} \quad u_{max} := \max_{t \in [0, T]} |u(t)|$$

Optimising Multisine for optimal crest factor

Frequency: Choose frequencies in logarithmic manner as multiples of the base frequency. $\omega_{k+1}/\omega_k \approx 1.05$

Phase: To prevent high peaks (Crest Factor) in the Signal, the phases of the different frequencies are modulated accordingly. (Positive interference)

Multisine Identification Implementation procedure

Window Length integer multiple of sampling time: $T = N \cdot \delta t$

Harmonics of base frequency are contained in multisine

$$\omega_{base} = \frac{2\pi}{T}$$

Highest contained Frequency is half of Nyquist frequency: $\omega_{Nyquist} = \frac{2\pi}{4\Delta T}$

Experiment and Analysis (step 2): Insert Multisine periodically. Drop first Periods (till transients died out). Record M Periods, each with N samples, of input and output data. Average all the M periods and make the DFT (or vice versa). Finally build transfer function.

$$\hat{G}_{j\omega_k} = \frac{\hat{Y}(k(p))}{\hat{U}(k(p))}$$

Nonparametric and Frequency Domain Identification Models

Impulse response and transfer function

$$y(t) = \int_0^\infty g(\tau) u(t - \tau) \delta t$$

$$Y(s) = G(s) \cdot U(s)$$

$$G(s) = \int_0^\infty e^{-st} g(t) dt$$

Bode diagram from frequency sweeps

$$u(t) = A \cdot \sin(\omega \cdot t), \quad y(t) = \| G(j \cdot \omega) \| A \cdot \sin(\omega \cdot t + \alpha)$$

Online estimation for dynamic systems

Recursive Least Squares

New Inverse Covariance:

$$Q_K = Q_{k-1} + \phi_K \phi_K^T$$

Innovation update:

$$\hat{\theta}_k = \hat{\theta}_{k-1} + \underbrace{Q_k^{-1} \phi_k (y_k - \phi_k^T \hat{\theta}_{k-1})}_{\text{"innovation"}}$$

General Optimization Problem:

$$\hat{\theta}_k = \arg \min_{\theta} (\theta - \hat{\theta}_0)^T \cdot Q_0 \cdot (\theta - \hat{\theta}_0) + \sum_{i=1}^k (y_i - \phi_i^T \cdot \theta)^2$$

Kalman Filter

Valid for Discrete and Linear!

(If recursive least squares: $x_{k+1} = A_k \cdot x_k$

$$x_{k+1} = A_k \cdot x_k + \omega_k \quad \text{and} \quad y_k = C_k \cdot x_k + v_k$$

Steps of Kalman Filter

1 Prediction $\hat{x}_{[k|k-1]} = A_{k-1} \cdot \hat{x}_{[k-1|k-1]}$

$$P_{[k|k-1]} = A_{k-1} \cdot P_{[k-1|k-1]} \cdot A_{k-1}^T + W_{k-1}$$

if recursive linear least squares without W_{k-1} .

2 Innovation update $P_{[k|k]} = (P_{[k|k-1]}^{-1} + C_k^T \cdot V^{-1} \cdot C_k)^{-1}$

$$\hat{x}_{[k|k]} = \hat{x}_{[k|k-1]} + P_{[k|k]} \cdot C_k^T \cdot V^{-1} \cdot (y_k - C_k \cdot \hat{x}_{[k|k-1]})$$

Bode Diagram:

Magnitude = Amplitude $|G(j\omega)|$

Phase $\arg G(j\omega)$