

Introduction

Euklidische Norm:

|| x ||_2 = sqrt(sum_{i=1}^n x_i^2) = sqrt(x^T x)

|| x ||_2^2 = x^T · x

Weighting Eukledian Norm:

|| x ||_Q^2 = x^T Q · x

Frobenius Norm:

|| x ||_F^2 = trace(AA^T) = sum_{i=1}^n sum_{j=1}^m A_{ij} A_{ij}

∇ f(x) Jacobian

∇^2 f(x) Hessian

Error in variables

R_hat_ev(N) = (1/N * sum_{k=1}^N u(k)) / (1/N * sum_{k=1}^N i(k))

Matrix derivatives

d(c^T x) / dx = c

d(x^T A x) / dx = (A^T + A) x

Linear and non-linear models:
linear if parameters go in linear i.e. (θ₁x² + θ₂x + θ₃), nonlinear if i.e (sin(θ₁)x + θ₂) TODO check that...

Table of Derivatives:

| f(x) | f'(x) |
|--------------------|-----------------------------|
| g(x) · h(x) | g'(x) · h(x) + g(x) · h'(x) |
| g(h(x)) | g'(h(x)) · h'(x) |
| sin(x) | − cos(x) |
| cos(x) | sin(x) |
| tan(x) | − ln(cos(x)) |
| e ^{kx} | 1/k e ^{kx} |
| ln(x) | 1/x |
| log _a x | 1/(ln a) (x ln x − x) |

Probability and Statistics

Random Variables and Probability

P(A|B) · P(B) = P(B|A) · P(A)

P(A|B) = P(A, B) / P(B) → P(B, A) · P(A) / P(B)

P(X ∈ [a, b]) = ∫_a^b p_X(x) dx

Mean

μ_X = E{f(x)} := ∫_{-∞}^∞ f(x) · p_X(x) dx

E{a + bX} := a + bE{X}

Variance

σ_X^2 := E{(X − μ_X)^2} = E{X^2} − μ_X^2

stddev σ_X = sqrt(variance σ_X^2)

Distributions

Uniform distribution:

P_y(x) = { 1/(b-a) if x ∈ [a, b]
0 else

Normal (Gaussian) distribution:

p(x) = 1 / (sqrt(2πσ^2) · exp(-(x − μ)^2 / (2σ^2)))

X ~ N(μ, σ^2)

Multidimensional Normal Distribution:

p(x) = 1 / (sqrt((2π)^n · det(Σ)) · exp(-1/2 · (x − μ)^T · Σ^-1 · (x − μ)))

Weibull distribution:

F(x) = 1 − e^-(λ·x)^k

Laplace distribution:

f(x|μ, b) = 1 / (2b) · exp(-|x − μ| / b)

Covariance and Correlaton:

σ(Y, Z) := E(Y − μ_Y)(Z − μ_Z) =

= ∫_{-∞}^∞ ∫_{-∞}^∞ (y − μ_Y)(z − μ_Z) · p_{Y,Z}(y, z) dy dz

Covariance Matrix:

Σ_x = cov(X) = E{X X^T} − μ_x μ_x^T

Multidimensional Random Variables:

E f(X) = ∫_{R^n} f(x) p_X(x) d^n x

cov(X) = E{(X − μ_X)(X − μ_X)^T}

cov(X) = E{X X^T} − μ_X μ_X^T

cov(Y) = Σ_y = A Σ_x A^T for y = A · x

E{AX} = A · E{X}

Rules for variance:

var(X + Y) = var(X) + var(Y) + 2 · cov(X, Y)

var(aX) = a^2 · var(X)

Verschiebesatz:

var(X) = E((X − E(X))^2) = E(X^2) − (E(X))^2

unit Variance is variance = 1

Statistical estimators:
Biased- and unbiasedness → an estimator θ̂_N is called unbiased iff E{θ̂_N(y_N)} = θ_0, where θ_0 is the true value of a parameter. Otherwise, is called biased.

Asymptotic Unbiasedness → An estimator θ̂_N is called asymptotically unbiased iff lim_{n→∞} E{θ̂_N(y_N)} = θ_0

Consistency → An estimator θ̂_N(y_N) is called consistent if, for any ε > 0, the probability P(θ̂_N(y_N) ∈ [θ_0 − ε, θ_0 + ε]) tends to one as N → ∞.

Unconstrained Optimization

Theorem 1 (First Order Necessary Conditions)
If $x^* \in D$ is local minimizer of $f : D \rightarrow \mathbb{R}$ and $f \in C^1$ then $\nabla f(x^*) = 0$
Definition (Stationary Point) A point \bar{x} with $\nabla f(\bar{x}) = 0$ is called a stationary point of f.

Theorem 2 (Second Order Necessary Conditions)
If $x^* \in D$ is local minimizer of $f : D \rightarrow \mathbb{R}$ and $f \in C^2$ then $\nabla^2 f(x^*) \succeq 0$

Theorem 3 (Second Order Sufficient Conditions and Stability under Perturbations)
Assume that $f : D \rightarrow \mathbb{R}$ is C^2 . If $x^* \in D$ is a stationary point and $\nabla^2 f(x^*) \succ 0$ then x^* is a strict local minimizer of f. In addition, this minimizer is locally unique and is stable against small perturbations of f, i.e. there exists a constant C such that for sufficiently small $p \in \mathbb{R}^n$ holds

$$\| x^* - argmin_x (f(x) + p^T x) \| \leq C \| p \|$$

Linear Least Squares Estimation

Preliminaries: I.I.D and gaussian noise

Overall Model

$$y(k) = \phi(k)^T \theta + \epsilon(k)$$

Least Squares cost function as sum

$$\sum_{k=1}^N (y(k) - \phi(k)^T \theta)^2$$

Least Squares cost function

$$f(\theta) = \| y_N - \Phi_N \theta \|_2^2$$

Unique minimizers

$$\hat{\theta}_{LS} = argmin_{\theta \in \mathbb{R}} f(\theta)$$

$$\theta^* = \underbrace{(\Phi^T \Phi)^{-1} \Phi^T}_{\Phi^+} y$$

Pseudo Inverse: $\Phi^+ = (\Phi^T \Phi)^{-1} \Phi^T$

Weighted Least Squares (unitless)
For I.I.D noise: Unweight Least Squares is optimal: W=I

$$\sum_{k=1}^N \frac{(y(k) - \phi(k)^T \theta)^2}{\sigma_{\epsilon}^2(k)}$$

$$f_{WLS}(\theta) = \| y_N - \Phi_N \theta \|_W^2 = (Y_N - \Phi \cdot \theta)^T \cdot W \cdot (Y_N - \Phi \cdot \theta)$$

Solution for WLS:

$$\hat{\theta}_{WLS} = \tilde{\Phi}^+ \tilde{y}$$

with $\tilde{\Phi} = W^{\frac{1}{2}} \Phi$ and $\tilde{y} = W^{\frac{1}{2}} y$.

$$\hat{\theta}_{WLS} = argmin_{\theta \in \mathbb{R}} f_{WLS}(\theta) = (\Phi^T W \Phi)^{-1} \Phi^T W y$$

Singular Value Decomposition

$$A = U S V^T \quad mit \quad U \in \mathbb{R}^{m \times m}, \quad V \in \mathbb{R}^{n \times n} \text{ und } S \in \mathbb{R}^{m \times n}$$

Moore Penrose Pseudo Inverse

$$\Phi^+ = V S^+ U^T$$

Regularization for Least Squares
 $lim_{\alpha \rightarrow 0} (\Phi^T \Phi + \alpha I)^{-1} \Phi^T = \Phi^+$ with Φ^+ MPPI

$$\theta^*(\alpha) = argmin_{\theta \in \mathbb{R}} \frac{1}{2} \| y - \Phi \theta \|_2^2 + \frac{\alpha}{2} \| \theta \|_2^2$$

Expectation of Least Squares Estimator

$$E\{\hat{\theta}_{WLS}\} = E\{(\Phi_N^T W \Phi_N)^{-1} \Phi_N^T W y_N\} = \theta_0$$

Covariance of the least squares estimator

$$cov(\hat{\theta}_{WLS}) = (\Phi_N^T W \Phi_N)^{-1} = (\Phi_N^T \Sigma_{\epsilon N}^{-1} \Phi_N)^{-1}$$

$$cov(\hat{\theta}_{WLS}) \succeq (\Phi_N^T W \Phi_N)^{-1}$$

Example:

$$\epsilon(1) \sim \mathcal{N}(0 | \sigma_1^2) \quad \epsilon(2) \sim \mathcal{N}(0 | \sigma_2^2)$$

$$N = 2 \quad \Sigma_{\epsilon N} = \begin{bmatrix} \sigma_1^2 & 0 \\ 0 & \sigma_2^2 \end{bmatrix}$$

$$W^{OPT} = \Sigma_{\epsilon N}^{-1} \begin{bmatrix} \frac{1}{\sigma_1^2} & 0 \\ 0 & \frac{1}{\sigma_2^2} \end{bmatrix}$$

$$cov(\hat{\theta}_{WLS}) = (Y_N - \Phi_N \theta)^T \cdot W \cdot (Y_N - \Phi_N \theta) =$$

$$\sum_{k=1}^2 (y(k) - \phi(k)^T \theta) \cdot \frac{1}{\sigma_k^2} \cdot (y(k) - \phi(k)^T \theta)$$

Measuring the goodness of Fit using $R^2 \quad 0 \leq R^2 \leq 1$

$$R^2 = 1 - \frac{\| y_N - \Phi_N \hat{\theta} \|_2^2}{\| y_N \|_2^2} = 1 - \frac{\| \epsilon_N \|_2^2}{\| y_N \|_2^2} =$$

$$\frac{\| y_N \|_2^2 - \| \epsilon_N \|_2^2}{\| y_N \|_2^2} = \frac{\| \hat{y}_N \|_2^2}{\| y_N \|_2^2}$$

residual $\epsilon_N \uparrow \rightarrow R^2 \rightarrow 0$ (= bad)

Estimating the Covariance with the Single Experiment

$$\hat{\sigma}_{\epsilon}^2 := \frac{1}{N - d} \sum_{k=1}^N (y(k) - \phi(k)^T \hat{\theta}_{LS})^2 = \frac{\| y_N - \phi_N \hat{\theta}_{LS} \|_2^2}{N - d}$$

$$\hat{\Sigma}_{\hat{\theta}} := \hat{\sigma}_{\epsilon}^2 (\phi_N^T \phi_N)^{-1} = \frac{\| y_N - \phi_N \hat{\theta}_{LS} \|_2^2}{N - d} \cdot (\phi_N^T \phi_N)^{-1}$$

Maximum Likelihood Estimation

Maximum Likelihood Estimation (ML) L_2 Estimation:
Measurement Errors assumed to be Normally distributed

$$P(y|\theta) = C \prod_{i=1}^N \exp\left(\frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}\right)$$

Positive log-Likelihood. Logarithm makes from products a sum!

$$\log p(y|\theta) = \log(C) + \sum_{i=1}^N \frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}$$

Negative log-Likelihood:

$$\hat{\theta}_{ML} = \arg \max_{\theta \in \mathbb{R}^d} p(y|\theta) = \arg \min_{\theta \in \mathbb{R}^d} \sum_{i=1}^m \frac{(y_i - M_i(\theta))^2}{2 \sigma_i^2}$$

$$\arg \max_{\theta \in \mathbb{R}^d} p(y|\theta) = \arg \min_{\theta \in \mathbb{R}^d} \frac{1}{2} \|S^{-1} \cdot (y - M(\theta))\|_2^2$$

L_1 Estimation:

Measurement Errors assumed to be Laplace distributed.

$$\text{Median}(x) = \left\lceil \frac{x+1}{2} \right\rceil$$

Robust against outliers

$$\begin{aligned} \min_{\theta} \|y - M(\theta)\|_1 &= \min_{\theta} \sum_{i=1}^N |y_i - M_i(\theta)| = \\ &= \text{median of } \{Y_1, \dots, Y_N\} \end{aligned}$$

$$P(y|\theta) = C \prod_{i=1}^N \exp\left(\frac{-|y_i - \theta|}{2 \cdot a_i}\right)$$

Bayesian Estimation and the Maximum a Posteriori Estimate
Assumptions: i.i.d noise and linear model

$$p(\theta|y_N) \cdot p(y_N) = p(y_N|\theta) \cdot p(\theta)$$

$$\hat{\theta}_{MAP} = \arg \min_{\theta \in \mathbb{R}} \{-\log(p(y_N|\theta)) - \log(p(\theta))\}$$

MAP Example: Regularised Least Squares

$$\theta = \bar{\theta} \pm \sigma_{\theta} \quad \text{with} \quad \bar{\theta} = \theta_{\text{apriori}}$$

$$\hat{\theta}_{MAP} = \arg \min_{\theta \in \mathbb{R}} \frac{1}{2} \cdot \frac{1}{\sigma_{\epsilon}^2} \cdot \|y_N - \Phi_N \cdot \theta\|_2^2 + \frac{1}{2} \cdot \frac{1}{\sigma_{\theta}^2} \cdot (\theta - \bar{\theta})^2$$

Recursive Linear Least Squares

$$\theta_{ML}(N) = \arg \min_{\theta \in \mathbb{R}} \frac{1}{2} \|y_N - \Phi_N \cdot \theta\|_2^2$$

$$\begin{aligned} \hat{\theta}_{ML}(N+1) &= \arg \min_{\theta \in \mathbb{R}^d} \left(\alpha \cdot \frac{1}{2} \cdot \|\theta - \hat{\theta}_{ML}(N)\|_{Q_N}^2 + \right. \\ &\quad \left. \frac{1}{2} \cdot \|y(N+1) - \varphi(N+1)^T \cdot \theta\|_2^2 \right) \end{aligned}$$

Q_0 given, and $\hat{\theta}_{ML}(0)$ given,

$$Q_{N+1} = \alpha \cdot Q_N + \varphi(N+1) \cdot \varphi(N+1)^T,$$

$$\begin{aligned} \hat{\theta}_{ML}(N+1) &= \hat{\theta}_{ML}(N) + Q_{N+1}^{-1} \cdot \varphi(N+1) \cdot [y(N+1) - \\ &\quad \varphi(N+1)^T \cdot \hat{\theta}_{ML}(N)] \end{aligned}$$

Cramer-Rao-Inequality (Fisher information Matrix M)

$$\Sigma_{\theta} \succeq M^{-1} = (\Phi_N^T \cdot \Sigma^{-1} \cdot \Phi_N)^{-1}$$

$$L(\theta, y_N) = \frac{1}{2} \cdot (\Phi_N \cdot \theta - y_N)^T \cdot \Sigma^{-1} \cdot (\Phi_N \cdot \theta - y_N) = \log(p(y_N|\theta))$$

$$M = E\{\nabla_{\theta}^2 L(\theta, y_N)\} = \nabla_{\theta}^2 L(\theta, y_N) = \Phi_N^T \cdot \Sigma^{-1} \cdot \Phi_N$$

Confirms that $W = \Sigma^{-1}$ is the optimal weighting Matrix for WLS.

Dynamic Models

Linear Time Invariant (LTI) Systems

with A, B, C, D are matrices

$$\dot{x} = Ax + Bu \quad y = Cx + Du$$

$$G(s) = C(sI - A)^{-1}B + D$$

LTI systems as Input-Output Models

$$G(S) = \frac{b_0 + b_1 s + \dots + b_n s^n}{a_0 + a_1 s + \dots + a_{n-1} s^{n-1} + s^n}$$

Different Models

Deterministic Model: $y(k) = M(k; U, x_{init}, p)$

Model with measurement Noise:

$$y(k) = M(k; U, x_{init}, p) + \varepsilon(k)$$

Model with Input and Output Errors:

$$y(k) = M(k; U + \varepsilon_N^u, x_{init}, p) + \varepsilon^y(k)$$

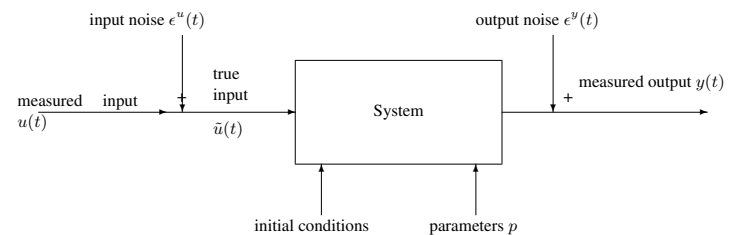
Pure Output Error (OE) Minimization

$$\theta_{ML} = \arg \min_{\theta} \sum_{k=1}^N (y(k) - M(k; U, x_{init}, p))^2$$

Output Error Minimization for FIR Models

$$y(k) = (u(k), u(k-1), \dots, u(k-n_b)) \cdot \theta + \varepsilon(k)$$

$$\min_{\theta} \sum_{k=n_b+1}^N (y(k) - (u(k), u(k-1), \dots, u(k-n_b)) \cdot \theta)^2$$



Models with Input and Output Errors

$$\arg \min_{\theta} \sum_{k=1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; U + \epsilon_N^u, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (\epsilon_u(k))^2$$

$$\arg \min_{\theta} \sum_{k=1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; \tilde{U}, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (u(k) - \tilde{u}(k))^2$$

Fourier Transformation

How to compute FT? By DFT, which solves the problem of finite time and discrete values.

Can we use an input with many frequencies to get many FRF (Frequency Response Function) values in a single experiment? So far only frequency sweeping (high comp. times due to repetition for each frequency). We should use multisines!

Aliasing and Leakage Errors

Aliasing Error due to sampling of continous signal to discrete signal. Avoid with Nyquist Theoreme:

f_{Nyquist} = \frac{1}{2\Delta t} [Hz] \quad or \quad \omega_{Nyquist} = \frac{2\pi}{2\Delta t} [rad/s]

Leakage Error due to windowing.

\omega_{base} := \frac{2\pi}{N \cdot \Delta t} = \frac{2\pi}{T} \rightarrow \omega = m \frac{2\pi}{N \cdot \Delta t}

Crest Factor = Scheitelfaktor

CrestFactor = \frac{u_{max}}{u_{rms}} \quad with :
u_{rms} := \sqrt{\frac{1}{T} \int_0^T u(t)^2 dt} \quad and \quad u_{max} := \max_{t \in [0,T]} |u(t)|

Optimising Multisine for optimal crest factor

Frequency: Choose frequencies in logarithmic manner as multiples of the base frequency. \omega_{k+1}/\omega_k \approx 1.05

Phase: To prevent high peaks (Crest Factor) in the Signal, the phases of the different frequencies are modulated accordingly. (Positive interference)

Multisine Identification Implementation procedure

Window Length integer multiple of sampling time: T = N \cdot \delta t

Harmonics of base frequency are contained in multisine
\omega_{base} = \frac{2\pi}{T}

Highest contained Frequency is half of Nyquist frequency: \omega_{Nyquist} = \frac{2\pi}{4\Delta T}

Experiment and Analysis (step 2): Insert Multisine periodically. Drop first Periods (till transients died out). Record M Periods, each with N samples, of input and output data. Average all the M periods and make the DFT (or vice versa). Finally build transfer function.

\hat{G}_{j\omega_k} = \frac{\hat{Y}(k(p))}{\hat{U}(k(p))}

Nonparametric and Frequency Domain Identification Models

Impulse response and transfer function

y(t) = \int_0^\infty g(\tau) u(t - \tau) \delta t

Y(s) = G(s) \cdot U(s)

G(s) = \int_0^\infty e^{-st} g(t) dt

Bode diagram from frequency sweeps

u(t) = A \cdot \sin(\omega \cdot t), \quad y(t) = \| G(j \cdot \omega) \| A \cdot \sin(\omega \cdot t + \alpha)

Online estimation for dynamic systems

Recursive Least Squares

New Inverse Covariance:

Q_K = Q_{k-1} + \phi_K \phi_K^T

Innovation update:

\hat{\theta}_k = \hat{\theta}_{k-1} + \underbrace{Q_k^{-1} \phi_k (y_k - \phi_k^T \hat{\theta}_{k-1})}_{\text{“innovation”}}

General Optimization Problem:

\hat{\theta}_k = \argmin_{\theta} (\theta - \hat{\theta}_0)^T \cdot Q_0 \cdot (\theta - \hat{\theta}_0) + \sum_{i=1}^k (y_i - \phi_i^T \cdot \theta)^2

Kalman Filter

Valid for Discrete and Linear!
(If recursive least squares: x_{k+1} = A_k \cdot x_k

x_{k+1} = A_k \cdot x_k + \omega_k \quad and \quad y_k = C_k \cdot x_k + v_k

Steps of Kalman Filter

- 1 Prediction** \hat{x}_{[k|k-1]} = A_{k-1} \cdot \hat{x}_{[k-1|k-1]}
P_{[k|k-1]} = A_{k-1} \cdot P_{[k-1|k-1]} \cdot A_{k-1}^T \cdot W_{k-1}
if recursive linear least squares without W_{k-1}.
- 2 Innovation update** P_{[k|k]} = (P_{[k|k-1]}^{-1} + C_k^T \cdot V^{-1} \cdot C_k)^{-1}
\hat{x}_{[k|k]} = \hat{x}_{[k|k-1]} + P_{[k|k]} \cdot C_k^T \cdot V^{-1} \cdot (y_k - C_k \cdot \hat{x}_{[k|k-1]})

Bode Diagram:

Magnitude = Amplitude |G(j\omega)|
Phase arg G(j\omega)