

Introduction

Euklidian Norm:

$$\|x\|_2 = \sqrt{\sum_{i=1}^n x_i^2} = \sqrt{x^T x}$$

$$\|x\|_2^2 = x^T \cdot x$$

Weighting Eukledian Norm:

$$\|x\|_Q^2 = x^T Q \cdot x$$

Frobenius Norm:

$$\|x\|_F^2 = \text{trace}(AA^T) = \sum_{i=1}^n \sum_{j=1}^m A_{ij} A_{ij}$$

$$\nabla f(x) \quad \text{Jacobian}$$

$$\nabla^2 f(x) \quad \text{Hessian}$$

Error in variables

$$\hat{R}_{ev}(N) = \frac{\frac{1}{N} \sum_{k=1}^N u(k)}{\frac{1}{N} \sum_{k=1}^N i(k)}$$

Matrix derivatives

$$\frac{d(c^T x)}{dx} = c$$

$$\frac{d(x^T A x)}{dx} = (A^T + A)x$$

Linear and non-linear models:

linear if parameters go in linear i.e. $(\theta_1 x^2 + \theta_2 x + \theta_3)$,
nonlinear if i.e $(\sin(\theta_1)x + \theta_2)$ TODO check that...

Table of Derivatives:

$\mathbf{f}(\mathbf{x})$	$\mathbf{f}'(\mathbf{x})$
$g(x) \cdot h(x)$	$g'(x) \cdot h(x) + g(x) \cdot h'(x)$
$g(h(x))$	$g'(h(x)) \cdot h'(x)$
$\sin(x)$	$-\cos(x)$
$\cos(x)$	$\sin(x)$
$\tan(x)$	$-\ln(\cos(x))$
e^{kx}	$\frac{1}{k} e^{kx}$
$\ln(x)$	$\frac{1}{x}$
$\log_a x$	$\frac{1}{\ln a} (x \ln x - x)$

Probability and Statistics

Random Variables and Probability

$$P(A|B) \cdot P(B) = P(B|A) \cdot P(A)$$

$$P(A|B) = \frac{P(A, B)}{P(B)} \rightarrow \frac{P(B, A) \cdot P(A)}{P(B)}$$

$$P(X \in [a, b]) = \int_a^b p_X(x) dx$$

Mean

$$\mu_X = \mathbb{E}\{f(x)\} := \int_{-\infty}^{\infty} f(x) \cdot p_X(x) dx$$

$$\mathbb{E}\{a + bX\} := a + b\mathbb{E}\{X\}$$

Variance

$$\sigma_X^2 := \mathbb{E}\{(X - \mu_X)^2\} = \mathbb{E}\{X^2\} - \mu_X^2$$

$$\text{stddev } \sigma_X = \sqrt{\text{variance } \sigma_X^2}$$

Useful Definitions

Covariance and Correlation:

$$\sigma(Y, Z) := \mathbb{E}(Y - \mu_Y)(Z - \mu_Z) =$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} (y - \mu_Y)(z - \mu_Z) \cdot p_{Y,Z}(y, z) dy dz$$

Covariance Matrix:

$$\Sigma_x = \text{cov}(X) = \mathbb{E}\{XX^T\} - \mu_x \mu_x^T$$

Multidimensional Random Variables:

$$\mathbb{E}f(X) = \int_{\mathbb{R}^n} f(x) p_X(x) d^n x$$

$$\text{cov}(X) = \mathbb{E}\{(X - \mu_X)(X - \mu_X)^T\}$$

$$\text{cov}(X) = \mathbb{E}\{XX^T\} - \mu_X \mu_X^T$$

$$\text{cov}(Y) = \Sigma_y = A \Sigma_x A^T \quad \text{for } y = A \cdot x$$

$$\mathbb{E}\{AX\} = A \cdot \mathbb{E}\{X\}$$

Rules for variance:

$$\text{var}(X + Y) = \text{var}(X) + \text{var}(Y) + 2 \cdot \text{cov}(X, Y)$$

$$\text{var}(aX) = a^2 \cdot \text{var}(X)$$

Verschiebesatz:

$$\text{var}(X) = \mathbb{E}((X - \mathbb{E}(X))^2) = \mathbb{E}(X^2) - (\mathbb{E}(X))^2$$

unit Variance is variance = 1

Statistical estimators:

Biased- and unbiasedness → an estimator $\hat{\theta}_N$ is called unbiased iff $\mathbb{E}\{\hat{\theta}_N(y_N)\} = \theta_0$, where θ_0 is the true value of a parameter. Otherwise, is called biased.

Asymptotic Unbiasedness → An estimator $\hat{\theta}_N$ is called asymptotically unbiased iff $\lim_{n \rightarrow \infty} \mathbb{E}\{\hat{\theta}_N(y_N)\} = \theta_0$

Consistency → An estimator $\hat{\theta}_N(y_N)$ is called consistent if, for any $\epsilon > 0$, the probability $P(\hat{\theta}_N(y_N) \in [\theta_0 - \epsilon, \theta_0 + \epsilon])$ tends to one as $N \rightarrow \infty$.

Distributions

Uniform distribution:

$$P_y(x) = \begin{cases} \frac{1}{b-a} & \text{if } x \in [a, b] \\ 0 & \text{else} \end{cases}$$

Normal (Gaussian) distribution:

$$p(x) = \frac{1}{\sqrt{2\pi\sigma^2}} \cdot \exp\left(-\frac{(x - \mu)^2}{2\sigma^2}\right)$$

$$X \sim \mathcal{N}(\mu, \sigma^2)$$

Multidimensional Normal Distribution:

$$p(x) = \frac{1}{\sqrt{(2\pi)^n \cdot \det(\Sigma)}} \cdot \exp\left(-\frac{1}{2} \cdot (x - \mu)^T \cdot \Sigma^{-1} \cdot (x - \mu)\right)$$

Weibull distribution:

$$F(x) = 1 - e^{-(\lambda \cdot x)^k}$$

Laplace distribution:

$$f(x|\mu, b) = \frac{1}{2b} \cdot \exp\left(-\frac{|x - \mu|}{b}\right)$$

Unconstrained Optimization

Theorem 1: (First Order Necessary Conditions)

If $x^* \in D$ is local minimizer of $f : D \rightarrow \mathbb{R}$ and $f \in C^1$ then $\nabla f(x^*) = 0$ Definition (Stationary Point) A point \bar{x} with $\nabla f(\bar{x}) = 0$ is called a stationary point of f .

Theorem 2: (Second Order Necessary Conditions)

If $x^* \in D$ is local minimizer of $f : D \rightarrow \mathbb{R}$ and $f \in C^2$ then $\nabla^2 f(x^*) \succeq 0$

Theorem 3: (Second Order Sufficient Conditions and Stability under Perturbations)

Assume that $f : D \rightarrow \mathbb{R}$ is C^2 . If $x^* \in D$ is a stationary point and $\nabla^2 f(x^*) \succ 0$ then x^* is a strict local minimizer of f . In addition, this minimizer is locally unique and is stable against small perturbations of f , i.e. there exists a constant C such that for sufficiently small $p \in \mathbb{R}^n$ holds

$$\|x^* - \arg \min_x (f(x) + p^T x)\| \leq C \|p\|$$

Linear Least Squares Estimation

Preliminaries: I.I.D and gaussian noise

Overall Model

$$y(k) = \phi(k)^T \theta + \epsilon(k)$$

Least Squares cost function as sum

$$\sum_{k=1}^N (y(k) - \phi(k)^T \theta)^2$$

Least Squares cost function

$$f(\theta) = \|y_N - \Phi_N \theta\|_2^2$$

Unique minimizers

$$\hat{\theta}_{LS} = \arg \min_{\theta \in \mathbb{R}} f(\theta)$$

$$\theta^* = \underbrace{(\Phi^T \Phi)^{-1} \Phi^T y}_{\Phi^+}$$

Pseudo Inverse: $\Phi^+ = (\Phi^T \Phi)^{-1} \Phi^T$

Weighted Least Squares (unitless)

For I.I.D noise: Unweight Least Squares is optimal: $W=I$

$$\sum_{k=1}^N \frac{(y(k) - \phi(k)^T \theta)^2}{\sigma^2(k)}$$

$$f_{WLS}(\theta) = \|y_N - \Phi_N \theta\|_W^2 = (Y_N - \Phi \cdot \theta)^T \cdot W \cdot (Y_N - \Phi \cdot \theta)$$

Solution for WLS:

$$\hat{\theta}_{WLS} = \tilde{\Phi}^+ \tilde{y}$$

with $\tilde{\Phi} = W^{\frac{1}{2}} \Phi$ and $\tilde{y} = W^{\frac{1}{2}} y$.

$$\hat{\theta}_{WLS} = \arg \min_{\theta \in \mathbb{R}} f_{WLS}(\theta) = (\Phi^T W \Phi)^{-1} \Phi^T W y$$

Singular Value Decomposition

$$A = USV^T \quad \text{mit} \quad U \in \mathbb{R}^{m \times m}, \quad V \in \mathbb{R}^{n \times n} \text{ und } S \in \mathbb{R}^{m \times n}$$

Moore Penrose Pseudo Inverse

$$\Phi^+ = VS^+U^T$$

Regularization for Least Squares

$$\lim_{\alpha \rightarrow 0} (\Phi^T \Phi + \alpha I)^{-1} \Phi^T = \Phi^+ \text{ with } \Phi^+ \text{ MPPI}$$

$$\theta^*(\alpha) = \arg \min_{\theta \in \mathbb{R}} \frac{1}{2} \|y - \Phi \theta\|_2^2 + \frac{\alpha}{2} \|\theta\|_2^2$$

Expectation of Least Squares Estimator

$$E\{\hat{\theta}_{WLS}\} = E\{(\Phi_N^T W \Phi_N)^{-1} \Phi_N^T W y_N\} = \theta_0$$

Covariance of the least squares estimator

$$\text{cov}(\hat{\theta}_{WLS}) = (\Phi_N^T W \Phi_N)^{-1} = (\Phi_N^T \Sigma_{\epsilon_N}^{-1} \Phi_N)^{-1}$$

$$\text{cov}(\hat{\theta}_{WLS}) \succeq (\Phi_N^T W \Phi_N)^{-1}$$

Example LLS

$$\epsilon(1) \sim \mathcal{N}(0|\sigma_1^2) \quad \epsilon(2) \sim \mathcal{N}(0|\sigma_2^2)$$

$$N=2 \quad \Sigma_{\epsilon_N} = \begin{bmatrix} \sigma_1^2 & 0 \\ 0 & \sigma_2^2 \end{bmatrix}$$

$$W^{OPT} = \Sigma_{\epsilon_N}^{-1} \begin{bmatrix} \frac{1}{\sigma_1^2} & 0 \\ 0 & \frac{1}{\sigma_2^2} \end{bmatrix}$$

$$\text{cov}(\hat{\theta}_{WLS}) = (Y_N - \Phi_N \theta)^T \cdot W \cdot (Y_N - \Phi_N \theta) =$$

$$\sum_{k=1}^2 (y(k) - \phi(k)^T \theta) \cdot \frac{1}{\sigma_k^2} \cdot (y(k) - \phi(k)^T \theta)$$

Measuring the goodness of Fit using $R^2 \quad 0 \leq R^2 \leq 1$

$$R^2 = 1 - \frac{\|y_N - \Phi_N \hat{\theta}\|_2^2}{\|y_N\|_2^2} = 1 - \frac{\|\epsilon_N\|_2^2}{\|y_N\|_2^2} =$$

$$\frac{\|y_N\|_2^2 - \|\epsilon_N\|_2^2}{\|y_N\|_2^2} = \frac{\|\hat{y}_N\|_2^2}{\|y_N\|_2^2}$$

residual $\epsilon_N \uparrow \rightarrow R^2 \rightarrow 0 (= \text{bad})$

Estimating the Covariance with the Single Experiment

$$\hat{\sigma}_\epsilon^2 := \frac{1}{N-d} \sum_{k=1}^N (y(k) - \phi(k)^T \hat{\theta}_{LS})^2 = \frac{\|y_N - \phi_N \hat{\theta}_{LS}\|_2^2}{N-d}$$

$$\hat{\Sigma}_\theta := \hat{\sigma}_\epsilon^2 (\phi_N^T \phi_N)^{-1} = \frac{\|y_N - \phi_N \hat{\theta}_{LS}\|_2^2}{N-d} \cdot (\phi_N^T \phi_N)^{-1}$$

Maximum Likelihood Estimation

Maximum Likelihood Estimation (ML) L_2 Estimation:

Measurement Errors assumed to be Normally distributed

$$P(y|\theta) = C \prod_{i=1}^N \exp\left(\frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}\right)$$

Positive log-Likelihood. Logarithm makes from products a sum!

$$\log p(y|\theta) = \log(C) + \sum_{i=1}^N \frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}$$

Negative log-Likelihood:

$$\hat{\theta}_{ML} = \arg \max_{\theta \in \mathbb{R}^d} p(y|\theta) = \arg \min_{\theta \in \mathbb{R}^d} \sum_{i=1}^m \frac{(y_i - M_i(\theta))^2}{2 \sigma_i^2}$$

$$\arg \max_{\theta \in \mathbb{R}^d} p(y|\theta) = \arg \min_{\theta \in \mathbb{R}^d} \frac{1}{2} \| S^{-1} \cdot (y - M(\theta)) \|_2^2$$

L_1 Estimation:

Measurement Errors assumed to be Laplace distributed.

$$Median(x) = \lceil \frac{x+1}{2} \rceil$$

Robust against outliers

$$\begin{aligned} \min_{\theta} \| y - M(\theta) \|_1 &= \min_{\theta} \sum_{i=1}^N |y_i - M_i \theta| = \\ &= \text{median of } \{Y_1, \dots, Y_N\} \end{aligned}$$

$$P(y|\theta) = C \prod_{i=1}^N \exp\left(\frac{-|y_i - \theta|}{2 \cdot a_i}\right)$$

Bayesian Estimation and the Maximum a Posteriori Estimate

Assumptions: i.i.d noise and linear model

$$p(\theta|y_N) \cdot p(y_N) = p(y_N|\theta) \cdot p(\theta)$$

$$\hat{\theta}_{MAP} = \arg \min_{\theta \in \mathbb{R}} \{-\log(p(y_N|\theta)) - \log(p(\theta))\}$$

MAP Example: Regularised Least Squares

$$\theta = \bar{\theta} \pm \sigma_{\theta} \quad \text{with} \quad \bar{\theta} = \theta_{\text{apriori}}$$

$$\hat{\theta}_{MAP} = \arg \min_{\theta \in \mathbb{R}} \frac{1}{2} \cdot \frac{1}{\sigma_{\epsilon}^2} \cdot \| y_N - \Phi_N \cdot \theta \|_2^2 + \frac{1}{2} \cdot \frac{1}{\sigma_{\theta}^2} \cdot (\theta - \bar{\theta})^2$$

Recursive Linear Least Squares

$$\theta_{ML}(N) = \arg \min_{\theta \in \mathbb{R}^d} \frac{1}{2} \| y_N - \Phi_N \cdot \theta \|_2^2$$

$$\hat{\theta}_{ML}(N+1) = \arg \min_{\theta \in \mathbb{R}^d} \left(\alpha \cdot \frac{1}{2} \cdot \| \theta - \hat{\theta}_{ML}(N) \|_{Q_N}^2 + \right.$$

$$\left. \frac{1}{2} \cdot \| y(N+1) - \varphi(N+1)^T \cdot \theta \|_2^2 \right)$$

$$Q_0 \text{ given, and } \hat{\theta}_{ML}(0) \text{ given,}$$

$$Q_{N+1} = \alpha \cdot Q_N + \varphi(N+1) \cdot \varphi(N+1)^T,$$

$$\begin{aligned} \hat{\theta}_{ML}(N+1) &= \hat{\theta}_{ML}(N) + Q_{N+1}^{-1} \cdot \varphi(N+1) \cdot [y(N+1) - \\ &\quad \varphi(N+1)^T \cdot \hat{\theta}_{ML}(N)] \end{aligned}$$

Cramer-Rao-Inequality (Fisher information Matrix M)

$$\Sigma_{\theta} \succeq M^{-1} = (\Phi_N^T \cdot \Sigma^{-1} \cdot \Phi)^{-1}$$

$$L(\theta, y_N) = \frac{1}{2} \cdot (\Phi_N \cdot \theta - y_N)^T \cdot \Sigma^{-1} \cdot (\Phi_N \cdot \theta - y_N) = \log(p(y_N|\theta))$$

$$M = E\{\nabla_{\theta}^2 L(\theta, y_N)\} = \nabla_{\theta}^2 L(\theta, y_N) = \Phi_N^T \cdot \Sigma^{-1} \cdot \Phi_N$$

Confirms that $W = \Sigma^{-1}$ is the optimal weighting Matrix for WLS.

Dynamic Models

Linear Time Invariant (LTI) Systems

with A, B, C, D are matrices

$$\dot{x} = Ax + Bu \quad y = Cx + Du$$

$$G(s) = C(sI - A)^{-1}B + D$$

LTI sytems as Input-Output Models

$$G(S) = \frac{b_0 + b_1 s + \dots + b_n s^n}{a_0 + a_1 s + \dots + a_{n-1} s^{n-1} + s^n}$$

Different Models

Deterministic Model: $y(k) = M(k; U, x_{init}, p)$

Model with measurement Noise:

$$y(k) = M(k; U, x_{init}, p) + \varepsilon(k)$$

Model with Input and Output Errors:

$$y(k) = M(k; U + \varepsilon_N^u, x_{init}, p) + \varepsilon^y(k)$$

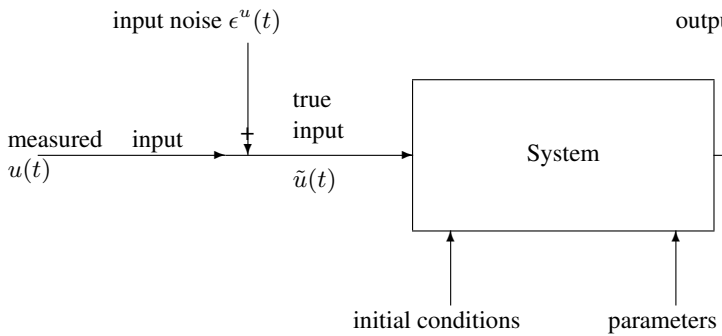
Pure Output Error (OE) Minimization

$$\theta_{ML} = \arg \min_{\theta} \sum_{k=1}^N (y(k) - M(k; U, x_{init}, p))^2$$

Output Error Minimization for FIR Models

$$y(k) = (u(k), u(k-1), \dots, u(k-n_b)) \cdot \theta + \varepsilon(k)$$

$$\min_{\theta} \sum_{k=n_b+1}^N (y(k) - (u(k), u(k-1), \dots, u(k-n_b)) \cdot \theta)^2$$



Models with Input and Output Errors

$$\arg \min_{\theta} \sum_{k=1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; U + \epsilon_N^u, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (\epsilon_u(k))^2$$

$$\arg \min_{\theta} \sum_{k=1}^N \frac{1}{\sigma_y^2} (y(k) - M(k; \tilde{U}, x_{init}, p))^2 + \frac{1}{\sigma_u^2} (u(k) - \tilde{u}(k))^2$$

Fourier Transformation

How to compute FT? By DFT, which solves the problem of finite time and discrete values.

Can we use an input with many frequencies to get many FRF (Frequency Response Function) values in a single experiment? So far only frequency sweeping (high comp. times due to repetition for each frequency). We should use multisines!

Aliasing and Leakage Errors

Aliasing Error due to sampling of continuous signal to discrete signal. Avoid with Nyquist Theorem:

$$f_{Nyquist} = \frac{1}{2\Delta t} [Hz] \quad \text{or} \quad \omega_{Nyquist} = \frac{2\pi}{2\Delta t} [rad/s]$$

Leakage Error due to windowing.

$$\omega_{base} := \frac{2\pi}{N \cdot \Delta t} = \frac{2\pi}{T} \rightarrow \omega = m \frac{2\pi}{N \cdot \Delta t}$$

Crest Factor = Scheitelfaktor

$$CrestFactor = \frac{u_{max}}{u_{rms}} \quad \text{with:}$$

$$u_{rms} := \sqrt{\frac{1}{T} \int_0^T u(t)^2 dt} \quad \text{and} \quad u_{max} := \max_{t \in [0, T]} |u(t)|$$

Optimising Multisine for optimal crest factor

Frequency: Choose frequencies in logarithmic manner as multiples of the base frequency. $\omega_{k+1}/\omega_k \approx 1.05$

Phase: To prevent high peaks (Crest Factor) in the Signal, the phases of the different frequencies are modulated accordingly. (Positive interference)

Multisine Identification Implementation procedure

Window Length integer multiple of sampling time: $T = N \cdot \delta t$

Harmonics of base frequency are contained in multisine
 $\omega_{base} = \frac{2\pi}{T}$

Highest contained Frequency is **half** of Nyquist frequency:
 $\omega_{Nyquist} = \frac{2\pi}{4\Delta T}$

Experiment and Analysis (step 2): Insert Multisine periodically. Drop first Periods (till transients died out). Record M Periods, each with N samples, of input and output data. Average all the M periods and make the DFT (or vice versa). Finally build transfer function.

$$\hat{G}_{j\omega_k} = \frac{\hat{Y}(k(p))}{\hat{U}(k(p))}$$

Nonparametric and Frequency Domain Identification Models

Impulse response and transfer function

$$y(t) = \int_0^\infty g(\tau)u(t-\tau)\delta t$$

$$Y(s) = G(s) \cdot U(s)$$

$$G(s) = \int_0^\infty e^{-st}g(t)dt$$

Bode diagram from frequency sweeps

$$u(t) = A \cdot \sin(\omega \cdot t), \quad y(t) = \| G(j \cdot \omega) \| A \cdot \sin(\omega \cdot t + \alpha)$$

Online estimation for dynamic systems

Recursive Least Squares

New Inverse Covariance:

$$Q_K = Q_{k-1} + \phi_K \phi_K^T$$

Innovation update:

$$\hat{\theta}_k = \hat{\theta}_{k-1} + \underbrace{Q_k^{-1} \phi_k (y_k - \phi_k^T \hat{\theta}_{k-1})}_{\text{"innovation"}}$$

General Optimization Problem:

$$\hat{\theta}_k = \arg \min_{\theta} (\theta - \hat{\theta}_0)^T \cdot Q_0 \cdot (\theta - \hat{\theta}_0) + \sum_{i=1}^k (y_k - \phi_k^T \cdot \theta)^2$$

Kalman Filter

Valid for Discrete and Linear!

(If recursive least squares: $x_{k+1} = A_k \cdot x_k$

$$x_{k+1} = A_k \cdot x_k + \omega_k \quad \text{and} \quad y_k = C_k \cdot x_k + v_k$$

Steps of Kalman Filter

1 Prediction $\hat{x}_{[k|k-1]} = A_{k-1} \cdot \hat{x}_{[k-1|k-1]}$
 $P_{[k|k-1]} = A_{k-1} \cdot P_{[k-1|k-1]} \cdot A_{k-1}^T \cdot W_{k-1}$
if recursive linear least squares without W_{k-1} .

2 Innovation update $P_{[k|k]} = (P_{[k|k-1]}^{-1} + C_k^T \cdot V^{-1} \cdot C_k)^{-1}$
 $\hat{x}_{[k|k]} = \hat{x}_{[k|k-1]} + P_{[k|k]} \cdot C_k^T \cdot V^{-1} \cdot (y_k - C_k \cdot \hat{x}_{[k|k-1]})$

Bode Diagram:

Magnitude = Amplitude $|G(j\omega)|$

Phase $\arg G(j\omega)$