

# **TTT4120 Digital Signal Processing**

## **Problem Set 4**

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## Problem 1 (2 points)

Given a filter with transfer function

$$H(z) = \frac{1}{1 - az^{-1}}$$

(a) Draw the pole-zero plot for the filter given  $a = 0.9$  and  $a = -0.9$ . Determine the filter type for two filters? Explain using the pole-zero plot.

We can rewrite the transfer function as

$$H(z) = \frac{z}{z - a}$$

This gives us 1 zero when  $z=0$  for both case  $a = 0.9$  and  $a = -0.9$ , we however get a pole at  $z = 0.9$  when  $a = 0.9$  and a pole at  $z = -0.9$  when  $a = -0.9$

this gives us the pole zero plots:

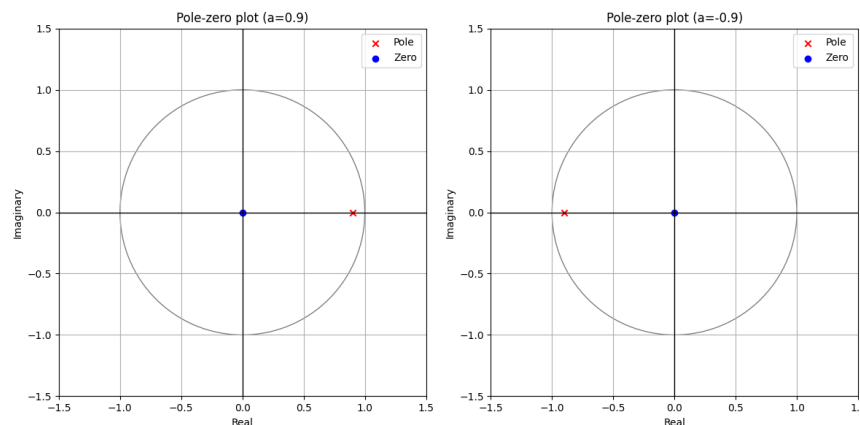


Figure 1: pole-zero plot for the filter given  $a = 0.9$  and  $a = -0.9$ .

Here we can see that when  $a = 0.9$ ,  $H(z)$  will increase as  $z$  increase, this indicates that the it is a low pass filter. The opposite happens when  $a = -0.9$  as the pole is on the negative side of the imaginary axis. We can also se this in the magnitude plots

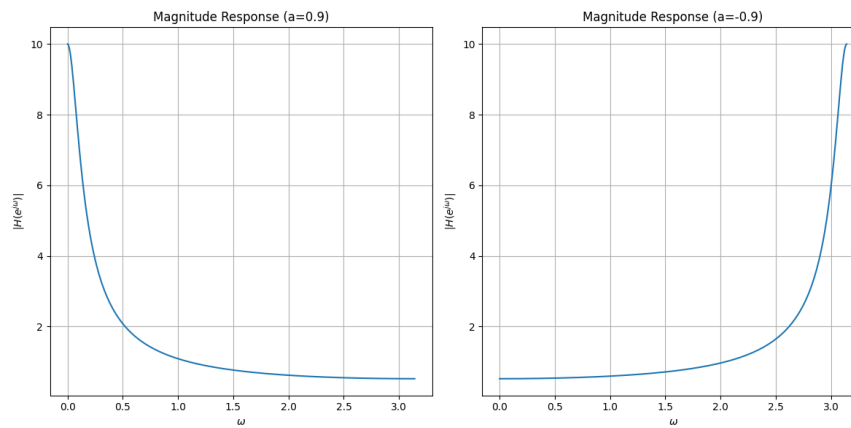


Figure 2: Magnitude plot for the filter given  $a = 0.9$  and  $a = -0.9$ .

**(b) Verify the results in 1(a) with pezdemo. The demo can be downloaded from the course home page.**

## Problem 2 (2 points)

Consider a causal digital filter with transfer function

$$H(z) = \frac{1}{\left(1 - \frac{1}{2}z^{-1}\right)\left(1 + \frac{1}{2}z^{-1}\right)}$$

**(a) Find the transfer function of the inverse filter of  $H(z)$**

As  $H(z) = \frac{B(z)}{A(z)}$ , we can find transfer function of the inverse filter of  $H(z)$  by switching the numerator with the denominator  $H^{-1}(z) = \frac{A(z)}{B(z)}$ . This gives us

$$H^{-1}(z) = \frac{\left(1 - \frac{1}{2}z^{-1}\right)\left(1 + \frac{1}{2}z^{-1}\right)}{1}$$

We can further rewrite too

$$H^{-1}(z) = \frac{\left(z - \frac{1}{2}\right)\left(z + \frac{1}{2}\right)}{z}$$

**(b) Is the inverse filter stable? Justify the answer.**

In this filter we can see that we get a pole when  $z = 0$ , as the pole is inside the unit circle we know that the system is stable. However this implies that the system gets an infinite gain at  $z = 0$  or at DC this is not realizable. This makes the filter not stable in a practical sense.

**(c) Is the inverse filter a minimum-phase filter?**

The definition for a minimum-phase system is whenever all zeros and poles are inside the unit circle. This stands true for this filter as the zeros are at  $z = \frac{1}{2}$  and  $z = -\frac{1}{2}$ .

**(d) Does the inverse filter have a linear phase characteristics? Justify your answer.**

As the zeros don't have a corresponding symmetric counterpart, and the 2 we have aren't symmetric with respect to the unit circle, the inverse filter does not have a linear phase characteristics.

### Problem 3 (2 points)

In the recording/mastering of sound signals or during playback, it is often desired to alter the characteristics of the sound at different frequencies. For example, we may wish to highlight the lower/middle frequencies, while we may wish to reduce the presence of high frequencies. This can be done by using so-called 'shelving' filters. Figure 3 shows a low-frequency shelving filter implementation. The filter  $A(z)$  is :

$$A(z) = \frac{\alpha - z^{-1}}{1 - \alpha z^{-1}}$$

The parameters  $\alpha$  and  $K$  are used to *tune* the filter.

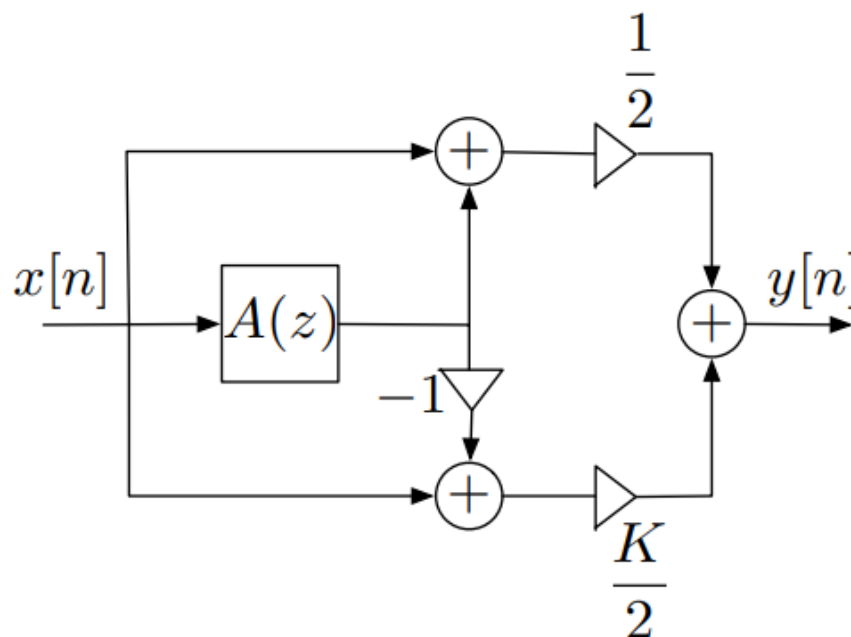


Figure 3: Low-frequency shelving filter

title

## Problem 4 (4 points)

## Appendix