## EE4341/EE6341 Advanced Analog Circuits

## Tutorial 7

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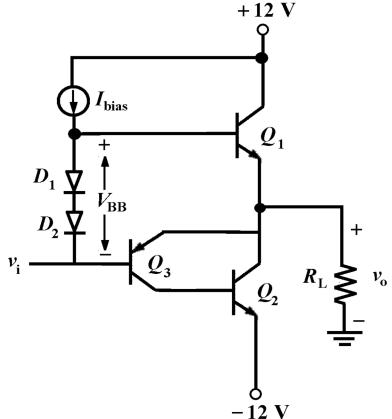
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Q1. Fig. 1 shows a Class AB power amplifier biased by a constant current source  $I_{\text{bias}} = 5 \text{ mA}$ . The load  $R_{\text{L}} = 100 \Omega$  and the transistor and the diode parameters are:  $I_S = 10^{-13} \text{ A}$  and  $V_T = 26 \text{ mV}$ . Current gains  $\beta_n = 100$  and  $\beta_p = 20$  for npn and pnp transistors, respectively.

- (a) Determine  $V_{BB}$ ,  $I_C$  and  $V_{BE}$  for each transistor when  $v_o = 0$ .
- (b) Repeat part (a) when  $v_o = +10 \text{ V}$ .
- (c) What is the instantaneous powers delivered to the load and dissipated in each transistor when  $v_o = +10 \text{ V}$ ?



(a) When 
$$v_0 = 0 \text{ V}$$
,  $i_L = 0 \text{ mA}$ .

$$i_{E1} = i_{E3} + i_{C2}$$

$$\left(\frac{\beta_n+1}{\beta_n}\right)i_{C1} = \left(\frac{\beta_p+1}{\beta_p}\right)i_{C3} + \beta_n i_{C3}$$

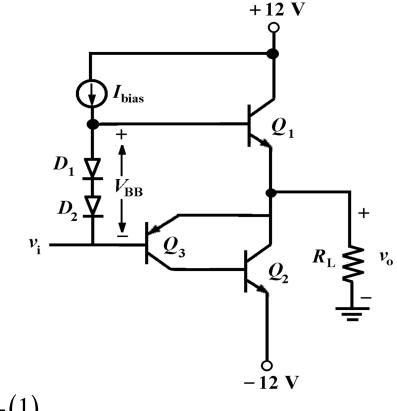
$$\left(\frac{101}{100}\right)i_{C1} = \left(\frac{21}{20}\right)i_{C3} + 100i_{C3}$$

$$1.01i_{C1} = 1.05i_{C3} + 100i_{C3}$$

$$i_{C1} = \left(\frac{1.05 + 100}{1.01}\right) i_{C3} = 100.05 i_{C3}$$
 ----(1)

$$V_{BB} = v_{BE1} + v_{EB3} = 2V_D$$

$$V_T \ln \frac{i_{C1}}{I_S} + V_T \ln \frac{i_{C3}}{I_S} = 2V_T \ln \frac{i_D}{I_S}$$



$$V_T \ln \frac{i_{C1}i_{C3}}{I_S^2} = V_T \ln \frac{i_D^2}{I_S^2}$$

$$i_{C1}i_{C3} = i_D^2$$
 ----(2)



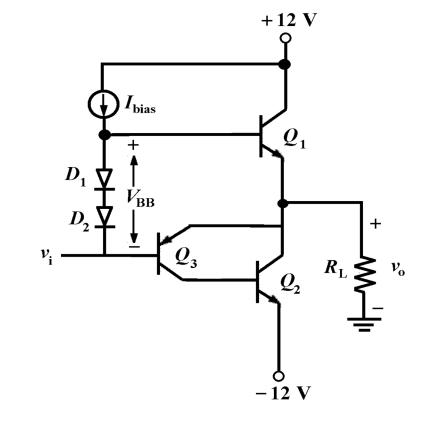
Substitute (1) into (2):

$$i_{C1} \left(\frac{i_{C1}}{100.05}\right) = i_D^2 = \left(I_{Bias} - i_{B1}\right)^2$$

$$\frac{i_{C1}^2}{100.05} = \left(5 - \frac{i_{C1}}{100}\right)^2$$

$$i_{C1}^2 = 100.05 \left(25 - 0.1i_{C1} + \frac{i_{C1}^2}{100^2}\right)$$

$$0.99i_{C1}^2 + 10.005i_{C1} - 2501.25 = 0$$



$$i_{C1} = \frac{-10.005 \pm \sqrt{10.005^2 + 4(0.99)(2501.25)}}{2(0.99)}$$

$$i_{C1} = 45.465 \text{ mA (take + ve value)}$$



$$i_{B1} = \frac{i_{C1}}{100} = 454.65 \,\mu\text{A}$$

$$i_D = 5 \text{ mA} - 454.65 \ \mu\text{A} = 4.54535 \text{ mA}$$

$$V_{BB} = 2V_T \ln \frac{i_D}{I_S} = 2(26 \times 10^{-3}) \ln \frac{4.54535 \times 10^{-3}}{10^{-13}} = 1.2761 \text{ V}$$

$$v_{BE1} = V_T ln \frac{i_{C1}}{I_S} = (26 \times 10^{-3}) ln \frac{45.465 \times 10^{-3}}{10^{-13}} = 0.6979 \text{ V}$$

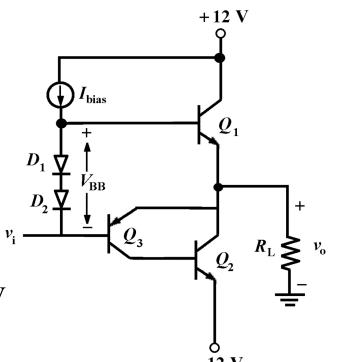
$$v_{EB3} = V_{BB} - v_{BE1} = 1.2761 - 0.6979 = 0.5782 \text{ V}$$

$$i_{C3} = I_S e^{\frac{v_{EB3}}{V_T}} = 10^{-13} e^{\frac{0.5782}{26 \times 10^{-3}}} = 0.455 \text{ mA}$$

$$i_{C2} = 100i_{C3} = 45.5 \text{ mA}$$

$$v_{BE2} = V_T \ln \frac{i_{C2}}{I_S} = (26 \times 10^{-3}) \ln \frac{45.5 \times 10^{-3}}{10^{-13}} = 0.6979 \text{ V}$$





(b) When 
$$v_o = 10 \text{ V}$$
,  $i_L = v_o / R_L = 100 \text{ mA}$ .

$$i_{E1} = i_{E3} + i_{C2} + i_{L}$$

$$1.01i_{C1} = 1.05i_{C3} + 100i_{C3} + 100$$

$$i_{C3} = \frac{i_{C1} - 99.01}{100.05}$$
 ----(3)

Substitute (3) into (2):

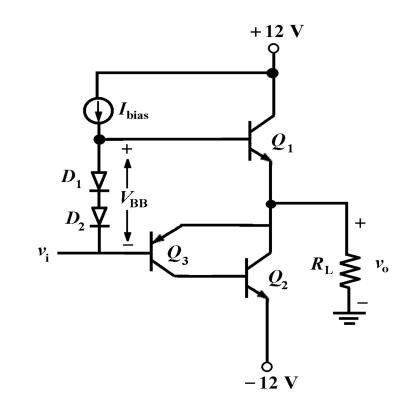
$$i_{C1} \left( \frac{i_{C1} - 99.01}{100.05} \right) = i_D^2 = \left( 5 - \frac{i_{C1}}{100} \right)^2$$

$$0.0099i_{C1}^{2} - 0.89i_{C1} - 25 = 0$$

$$i_{C1} = \frac{0.89 \pm \sqrt{0.89^2 + 4(0.0099)(25)}}{2(0.0099)}$$

$$i_{C1} = 112.37 \text{ mA (take + ve value)}$$





$$i_{B1} = \frac{i_{C1}}{100} = 1.1237 \text{ mA}$$
 $i_D = 5 \text{ mA} - 1.1237 \text{ mA} = 3.8763 \text{ mA}$ 

$$V_{BB} = 2V_T \ln \frac{i_D}{I_S} = 2(26 \times 10^{-3}) \ln \frac{3.8763 \times 10^{-3}}{10^{-13}} = 1.2678 \text{ V} \frac{D_1}{D_2} \nabla$$

$$v_{BE1} = V_T \ln \frac{i_{C1}}{I_S} = (26 \times 10^{-3}) \ln \frac{112.37 \times 10^{-3}}{10^{-13}} = 0.7214 \text{ V}$$

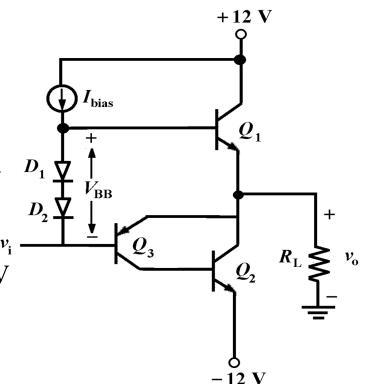
$$v_{EB3} = V_{BB} - v_{BE1} = 1.2678 - 0.7214 = 0.5464 \text{ V}$$

$$i_{C3} = I_S e^{\frac{v_{EB3}}{V_T}} = 10^{-13} e^{\frac{0.5464}{26 \times 10^{-3}}} = 133.9 \ \mu A$$

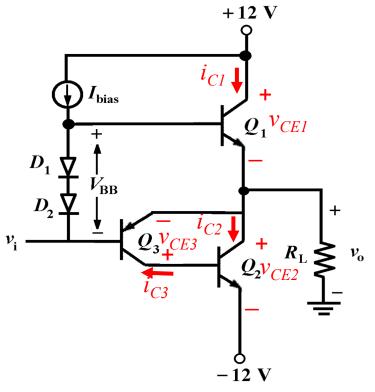
$$i_{C2} = 100i_{C3} = 13.39 \text{ mA}$$

$$v_{BE2} = V_T ln \frac{i_{C2}}{I_S} = (26 \times 10^{-3}) ln \frac{13.39 \times 10^{-3}}{10^{-13}} = 0.6661 \text{ V}$$





The instantaneous powers to the load and the power dissipation in the transistors:



$$p_{Q1} = i_{C1}v_{CE1}$$
  
= 112.37 mA × (12 – 10)V  
= 225 mW

$$p_{Q2} = i_{C2}v_{CE2}$$
  
= 13.39 mA × [10 – (-12)]V  
= 295 mW

$$p_{Q3} = i_{C3}v_{EC3}$$
  
= 133.9  $\mu$ A×[10-(-12+0.6661)]V  
= 2.86 mW

$$p_L = \frac{{v_o}^2}{R_L} = \frac{10^2}{100} = 1 \text{ W}$$



- 2. Fig. 2 shows a Class D power amplifier to drive a load  $R_L = 8 \Omega$ . Assume  $V_{\text{CE(sat)}} = 0.3 \text{ V}$  and the LC low pass filter is lossless.
- (a) What is the maximum power that can be delivered to the load?
- (b) If each switching transition of the transistor is about 5% of the period of switching frequency, what is the conversion efficiency at maximum output power?
- (c) Repeat part (b) if faster transistors are used with the switching transition reduced to 2.5%.

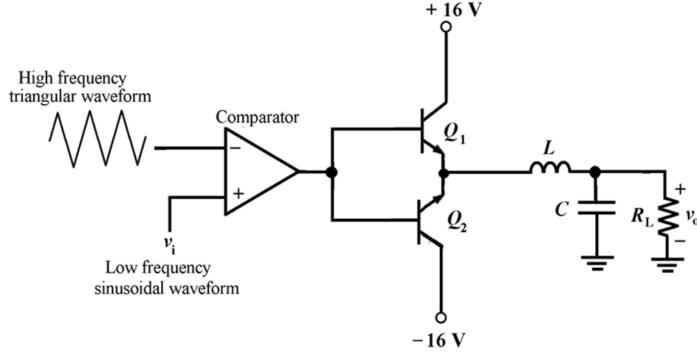
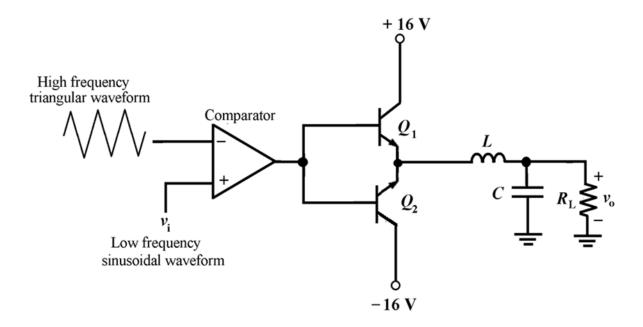




Figure 2



(a) Maximum power that can be delivered to the load:

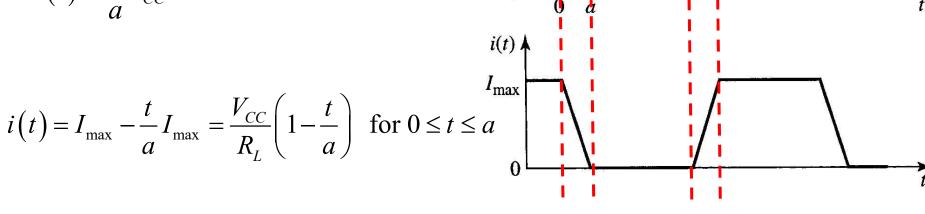
$$V_{o,\text{max}} = V_{CC} - V_{CE(sat)} = 16 - 0.3 = 15.7 \text{ V}$$

$$P_{L,\text{max}} = \frac{V_{o,\text{max}}^2}{2R_L} = \frac{(15.7)^2}{2 \times 8} = 15.4 \text{ W}$$



(b) Each transistor has two switching transitions and the power dissipated in each transistor:

$$v(t) = \frac{t}{a} V_{CC}$$
 for  $0 \le t \le a$ 



$$p(t) = v(t)i(t) = \frac{V_{CC}^{2}}{R_{L}} \left(\frac{t}{a} - \frac{t^{2}}{a^{2}}\right)$$

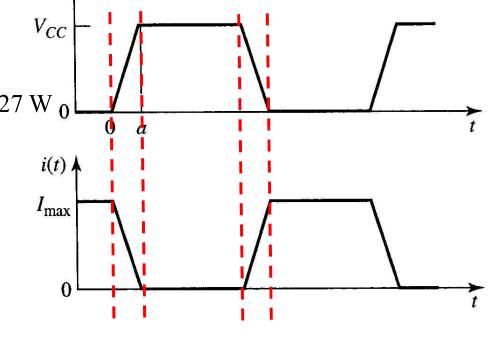
$$P_{T} = 2 \times \frac{1}{T} \int_{0}^{a} \frac{V_{CC}^{2}}{R_{L}} \left( \frac{t}{a} - \frac{t^{2}}{a^{2}} \right) dt = \frac{2}{T} \left( \frac{V_{CC}^{2}}{R_{L}} \right) \left[ \frac{t^{2}}{2a} - \frac{t^{3}}{3a^{2}} \right]_{0}^{a} = \frac{a}{T} \left( \frac{V_{CC}^{2}}{3R_{L}} \right)$$



$$P_T = \frac{a}{T} \left( \frac{V_{CC}^2}{3R_L} \right)$$

$$2P_T = \frac{a}{T} \left( \frac{2V_{o,max}^2}{3R_L} \right) = (0.05) \left( \frac{2 \times 15.7^2}{3 \times 8} \right) = 1.027 \text{ W }_0$$

$$I_{max} = \frac{V_{max}}{R_I} = \frac{15.7}{8} = 1.96 \text{ A}$$



$$\eta = \frac{P_{L,\text{max}}}{P_{L,\text{max}} + 2P_T + 0.9I_{\text{max}}V_{CE(sat)}} = \frac{15.4}{15.4 + 1.027 + 0.9 \times 1.96 \times 0.3} = 90.8\%$$



$$2P_T = \frac{a}{T} \left( \frac{2V_{o,\text{max}}^2}{3R_L} \right) = (0.025) \left( \frac{2 \times 15.7^2}{3 \times 8} \right) = 0.514 \text{ W}$$

$$I_{\text{max}} = 1.96 \text{ A}$$

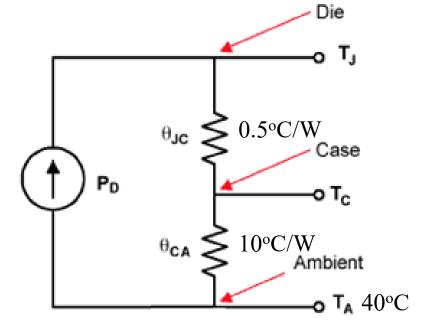
$$\eta = \frac{P_{L,\text{max}}}{P_{L,\text{max}} + 2P_T + 0.95I_{\text{max}}V_{CE(sat)}} = \frac{15.4}{15.4 + 0.514 + 0.95 \times 1.96 \times 0.3} = 93.5\%$$



- 3. The maximum permissible junction temperature of a power transistor is 150°C. It is desired to operate the transistor with a power dissipation of 15 W in an ambient temperature of 40°C. The thermal resistances of the transistor:
- $\theta_{JC}$  = 0.5°C/W (junction to case) and  $\theta_{CA}$  = 10°C/W (case to ambient).
- (a) Determine whether a heat sink is required for this application.
- (b) If a heat sink is needed, determine the required thermal resistance. To mount the heat sink, there is a mica washer between the transistor case and the heat sink. The thermal resistance of the mica washer  $\theta_W = 0.5$ °C/W.



(a)



$$\theta_T = \theta_{JC} + \theta_{CA} = 0.5^{\circ} \text{C/W} + 10^{\circ} \text{C/W} = 10.5^{\circ} \text{C/W}$$

$$T_J - T_A = \theta_T P_D = 10.5 \, ^{\circ}\text{C/W} \times 15 \, \text{W} = 157.5 \, ^{\circ}\text{C}$$

$$T_J = 157.5^{\circ} \text{C} + T_A = 157.5^{\circ} \text{C} + 40^{\circ} \text{C} = 197.5^{\circ} \text{C}$$

The junction temperature has exceeded the maximum permissible junction temperature (150°C). Hence, a heat sink is necessary to reduce the total thermal resistance.



(b) Let set the  $T_J$  to its maximum permissible value of 150°C, then we can solve for the required total thermal resistance.

$$T_{J} - T_{A} = \theta_{T}P$$

$$150 - 40 = \theta_{T} \times 15$$

$$\theta_{T} = \frac{110}{15} = 7.33^{\circ} \text{C/W}$$

$$\theta_{T} = \theta_{JC} + \theta_{W} + \theta_{SA} = 7.33^{\circ} \text{C/W}$$

$$\theta_{SA} = \theta_{T} - (\theta_{JC} + \theta_{W})$$

$$\theta_{SA} = 7.33 - (0.5 + 0.5)$$

$$\theta_{SA} = 6.33 \text{ °C/W}$$
Die

$$0.5^{\circ}\text{C/W}$$
Case
$$0.5^{\circ}\text{C/W}$$
Heat Sink
$$\theta_{SA} = 7.33 - (0.5 + 0.5)$$

$$\theta_{SA} = 7.33 - (0.5 + 0.5)$$

