

## Chapter 6: Controllability and Observability

- Consider the linear time-varying system

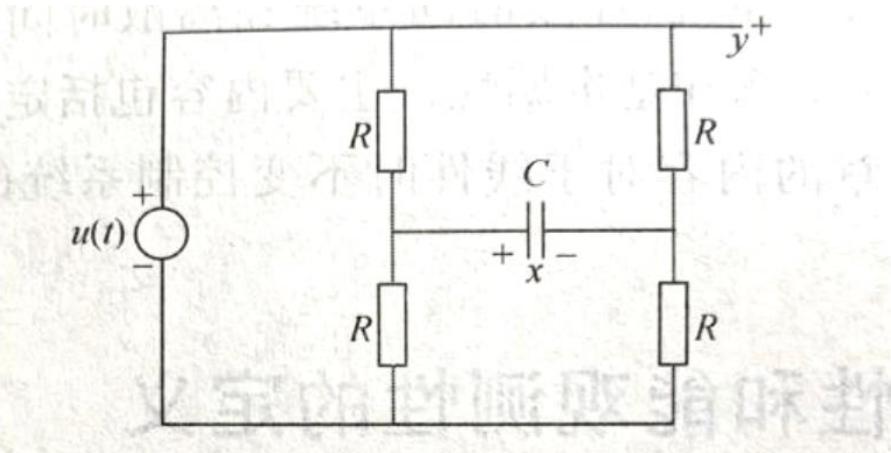
$$\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t) \quad (1)$$

$$\mathbf{y}(t) = C(t)\mathbf{x}(t) + D(t)\mathbf{u}(t) \quad (2)$$

where  $\mathbf{x} \in \mathcal{R}^n$  is the state vector,  $\mathbf{u} \in \mathcal{R}^m$  is the input vector and  $\mathbf{y} \in \mathcal{R}^p$  is the output vector.

- The controllability is concerned with the influence of input on the state whereas the observability examines the influence of state on the output.
- Since term  $D(t)\mathbf{u}(t)$  does not affect the above link, it can be excluded when discussing controllability and observability.

Consider the following example:



It is clear that  $x = 0$  no matter what is the input. So, the state is not controllable. On the other hand, if  $u(t) = 0$ , then  $y(t) = 0$  regardless of the initial charge of the capacitor. That is,  $x$  cannot be reflected from  $y$ . So, the system is not observable.

## Controllability

- The system (1)-(2) is called controllable on  $[t_0, t_f]$  if given any initial state  $\mathbf{x}(t_0) = \mathbf{x}_0$  and any final state  $\mathbf{x}(t_f) = \mathbf{x}_f$ , there exists a continuous input signal  $\mathbf{u}(t)$  that transfers  $\mathbf{x}_0$  to  $\mathbf{x}_f$  in a finite time.
- Note that there is no requirement for the system to stay at the origin after  $t_f$ .

**Theorem 6.1** *The linear system (1)-(2) is controllable on  $[t_0, t_f]$  if and only if the  $n \times n$  matrix*

$$W(t_0, t_f) = \int_{t_0}^{t_f} \Phi(t_f, t) B(t) B^T(t) \Phi(t_f, t) dt \quad (3)$$

*is invertible. (Observe that  $W(t_0, t_f)$  is non-negative definite and is called controllability Gramian).*

*Proof.* '= $\Rightarrow$ ': Suppose  $W(t_0, t_f)$  is invertible. Choose

$$\mathbf{u}(t) = -B^T(t)\Phi^T(t_f, t)W^{-1}(t_0, t_f)[\Phi(t_f, t_0)\mathbf{x}_0 - \mathbf{x}_f], \quad t \in [t_0, t_f] \quad (4)$$

Then,

$$\begin{aligned} \mathbf{x}(t_f) &= \Phi(t_f, t_0)\mathbf{x}_0 + \int_{t_0}^{t_f} \Phi(t_f, \sigma)B(\sigma)\mathbf{u}(\sigma)d\sigma \\ &= \Phi(t_f, t_0)\mathbf{x}_0 - \int_{t_0}^{t_f} \Phi(t_f, \sigma)B(\sigma)B^T(\sigma)\Phi^T(t_f, \sigma)W^{-1}(t_0, t_f) \\ &\quad \times [\Phi(t_f, t_0)\mathbf{x}_0 - \mathbf{x}_f]d\sigma \\ &= \Phi(t_f, t_0)\mathbf{x}_0 - W(t_0, t_f)W^{-1}(t_0, t_f)[\Phi(t_f, t_0)\mathbf{x}_0 - \mathbf{x}_f] \\ &= \mathbf{x}_f \end{aligned}$$

That is, the system is controllable on  $[t_0, t_f]$ .

' $\implies$ ' By contradiction, suppose the system (1)-(2) is controllable but  $W(t_0, t_f)$  is not invertible. Then, there exists a non-zero  $\mathbf{x}_f$  such that

$$0 = \mathbf{x}_f^T W(t_0, t_f) \mathbf{x}_f = \int_{t_0}^{t_f} \mathbf{x}_f^T \Phi(t_f, t) B(t) B^T(t) \Phi^T(t_f, t) \mathbf{x}_f dt \quad (5)$$

which implies

$$\mathbf{x}_f^T \Phi(t_f, t) B(t) = 0, \quad t \in [t_0, t_f] \quad (6)$$

On the other hand, since the system is controllable on  $[t_0, t_f]$ , choosing  $\mathbf{x}_0 = 0$  and  $\mathbf{x}(t_f) = \mathbf{x}_f$ , there exists an input  $u(t)$  such that

$$\mathbf{x}_f = \int_{t_0}^{t_f} \Phi(t_f, \sigma) B(\sigma) u(\sigma) d\sigma$$

Multiplying on the left by  $\mathbf{x}_f^T$  and applying (6), we obtain

$$\mathbf{x}_f^T \mathbf{x}_f = \int_{t_0}^{t_f} \mathbf{x}_f^T \Phi(t_0, \sigma) B(\sigma) u(\sigma) d\sigma = 0$$

which contradicts  $\mathbf{x}_f \neq 0$ .

## Linear Time-invariant Case:

$$A(t) = A, \quad B(t) = B, \quad C(t) = C, \quad D(t) = D$$

**Theorem 6.2** *The following are equivalent:*

- (i) *The system or pair  $(A, B)$  is controllable;*
- (ii) *The matrix*

$$W_c(t) = \int_0^t e^{A\tau} BB^T e^{A^T\tau} d\tau$$

*is positive definite for any  $t > 0$ ;*

- (iii) *The controllability matrix*

$$\mathcal{C} = [B \ AB \ A^2B \ \dots \ A^{n-1}B]$$

*has full row rank  $n$ ;*

- (iv) *The matrix  $[A - \lambda I, \ B]$  has full row rank (rank  $n$ ) for all  $\lambda$  in complex domain or  $\text{rank}[\lambda_i I - A, B] = n$ ,  $i = 1, 2, \dots, n$ , where  $\lambda_i$  is an eigenvalue of  $A$ .*

- (v) *Let  $\lambda$  and  $\mathbf{x}$  be any eigenvalue and any corresponding left eigenvector of  $A$ , i.e.  $\mathbf{x}^* A = \mathbf{x}^* \lambda$ , then  $\mathbf{x}^* B \neq 0$ ; (iv)-(v) Popov-Belevitch-Hautus (PHB) tests*

- (vi) *The eigenvalues of  $A + BF$  can be freely assigned by a suitable choice of  $F$ .*

*Proof.* (i)  $\iff$  (ii) is similar to the time-varying case.

(ii)  $\iff$  (iii): Suppose  $W_c(t) > 0$  but the matrix  $\mathcal{C}$  does not have full row rank. Then, there exists a  $\mathbf{v} \in \mathbb{R}^n$  such that

$$\mathbf{v}^* A^i B = 0, \quad i = 0, 1, \dots, n-1$$

In fact, the above equality holds for all  $i \geq 0$  by the Cayley-Hamilton Theorem. Hence,

$$\mathbf{v}^* e^{At} B = 0$$

for any  $t$ , then  $\mathbf{v}^* W_c(t) = 0$ , which is a contradiction as  $W_c(t) > 0$ .

Conversely, suppose  $\mathcal{C}$  has full row rank but  $W_c(t)$  is singular for some  $t_1$ . Then, there exists a non-zero vector  $\mathbf{v}$  such that  $\mathbf{v}^* e^{At} B = 0$  for all  $t \in [0, t_1]$ . By setting  $t = 0$ , we know  $\mathbf{v}^* B = 0$ . Further, by evaluating the  $i$ -th derivative of  $\mathbf{v}^* e^{At} B = 0$  at  $t = 0$  gives

$$\mathbf{v}^* A^i B = 0, \quad i > 0$$

Hence,

$$\mathbf{v}^* [B \ AB \ A^2 B \ \dots \ A^{n-1} B] = 0$$

which contradicts with the full rank of  $\mathcal{C}$ .

(iii)  $\implies$  (iv): Suppose, on the contrary, that the matrix

$$[A - \lambda I \quad B]$$

does not have full row rank for some  $\lambda$ . Then, there exists a vector  $\mathbf{x}$  such that

$$\mathbf{x}^*[A - \lambda I \quad B] = 0$$

i.e.

$$\mathbf{x}^*A = \lambda\mathbf{x}^*, \quad \mathbf{x}^*B = 0$$

This results in

$$\mathbf{x}^*[B \quad AB \cdots A^{n-1}B] = [\mathbf{x}^*B \quad \lambda\mathbf{x}^*B \cdots \lambda^{n-1}\mathbf{x}^*B] = 0$$

which contradicts with the fact that  $\mathcal{C}$  has full row rank.

(iv)  $\implies$  (v): It is obvious from the above proof.

The rest of the proof can be known later.

**Remark 6.1** For time-invariant systems, the controllability matrix  $\mathcal{C}$  is independent of  $t_0$  and  $t_f$ . Thus, the controllability of the system is not tied up to a particular time interval.

**Remark 6.2** If  $A$  is asymptotically stable and  $(A, B)$  is controllable, then the unique solution of

$$AW_c + W_c A^T = -BB^T$$

is positive definite and the solution is the controllability Gramian:

$$W_c = \int_0^\infty e^{A\tau} BB^T e^{A^T\tau} d\tau$$

## Example 6.1

$$A = \begin{bmatrix} -2 & -2 & 0 \\ 0 & 0 & 1 \\ 0 & -3 & -4 \end{bmatrix}, \quad B = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 1 & 1 \end{bmatrix}$$

*Then, the controllability matrix*

$$\mathcal{C} = [B \ AB \ A^2B] = \left[ \begin{array}{cc|cc|cc} 1 & 0 & -2 & -2 & 2 & 2 \\ 0 & 1 & 1 & 1 & -4 & -7 \\ 1 & 1 & -4 & -7 & 13 & 25 \end{array} \right]$$

*It is easy to check that  $\text{rank}(\mathcal{C}) = 3$ . Hence, the system is controllable.*

## Example 6.2

$$A = \begin{bmatrix} \lambda_1 & 0 & 0 \\ 0 & \lambda_2 & 1 \\ 0 & 0 & \lambda_2 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

*The controllability matrix*

$$\mathcal{C} = [B \ AB \ A^2B] = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 1 & \lambda_2 + \lambda_2 \\ 1 & \lambda_2 & \lambda_2^2 \end{bmatrix}$$

*It is easy to see that  $\text{rank}(\mathcal{C}) = 2$ . The system is not controllable.*

*Alternatively,*

$$[A - \lambda_1 I, \quad B] = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & \lambda_2 - \lambda_1 & 1 & 0 \\ 0 & 0 & \lambda_2 - \lambda_1 & 1 \end{bmatrix}$$

*does not have full row rank. So, the system is not controllable. In fact,  $\dot{x}_1 = \lambda_1 x_1$  is not influenced by the control input.*

### Example 6.3

$$A = \begin{bmatrix} a_1 & 0 \\ 0 & a_2 \end{bmatrix}, \quad B = \begin{bmatrix} b_1(t) \\ b_2(t) \end{bmatrix}$$

where  $a_1 \neq a_2$ .

In the case of  $b_1(t) = b_1$  and  $b_2(t) = b_2$ , the system is controllable if and only if  $b_1 \neq 0$  and  $b_2 \neq 0$ .

However, for time-varying case

$$b_1(t) = e^{a_1 t}, \quad b_2(t) = e^{a_2 t}$$

we have

$$W(t_0, t_f) = (t_f - t_0) \begin{bmatrix} e^{2a_1 t_0} & e^{(a_1 + a_2)t_0} \\ e^{(a_1 + a_2)t_0} & e^{2a_2 t_0} \end{bmatrix}$$

Since  $\det W(t_0, t_f) = 0$ , the system is not controllable.

Hence, point-wise interpretation of controllability property can be misleading.

## Example 6.4

$$\dot{\mathbf{x}} = \begin{bmatrix} -0.5 & 0 \\ 0 & -1 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0.5 \\ 1 \end{bmatrix} u$$

If  $x_1(0) = 10$  and  $x_2(0) = -1$ , can we apply a control input to bring the system state to 0 in 2 seconds?

Note that

$$\mathcal{C} = [B \ AB] = \begin{bmatrix} 0.5 & -0.25 \\ 1 & -1 \end{bmatrix} \implies \text{rank}(\mathcal{C}) = 2$$

Now,

$$W_c(2) = \int_0^2 \begin{bmatrix} e^{-0.5\tau} & 0 \\ 0 & e^{-\tau} \end{bmatrix} \begin{bmatrix} 0.5 \\ 1 \end{bmatrix} [0.5 \ 1] \begin{bmatrix} e^{-0.5\tau} & 0 \\ 0 & e^{-\tau} \end{bmatrix} d\tau = \begin{bmatrix} 0.2162 & 0.3167 \\ 0.3167 & 0.4908 \end{bmatrix}$$

$$\begin{aligned} u(t) &= -[0.5 \ 1] \begin{bmatrix} e^{-0.5(2-t)} & 0 \\ 0 & e^{-(2-t)} \end{bmatrix} W_c^{-1}(2) \begin{bmatrix} e^{-1} & 0 \\ 0 & e^{-2} \end{bmatrix} \begin{bmatrix} 10 \\ -1 \end{bmatrix} \\ &= -58.82e^{0.5t} + 27.96e^t \end{aligned}$$

**Remark 6.3** If you allow a longer time duration, say 4 seconds, to transfer the state from  $[10 \ -1]^T$  to 0, the control signal would be smaller.

**Example 6.5** Controllability of CCF form

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

The controllability matrix is

$$\mathcal{C} = \begin{bmatrix} 0 & 0 & 1 \\ 0 & 1 & -a_2 \\ 1 & -a_2 & -a_1 + a_2^2 \end{bmatrix}$$

has full row rank. So, the system in CCF form is controllable.

## Controllability Indices

- Let  $A \in \mathcal{R}^{n \times n}$  and  $B \in \mathcal{R}^{n \times p}$  with  $B$  full column rank. Note that if  $B$  is not of full column rank, there exist redundant inputs.
- Denote  $B = [\mathbf{b}_1 \ \cdots \ \mathbf{b}_p]$ . Then, the controllability matrix:

$$\mathcal{C} = [\mathbf{b}_1 \ \cdots \ \mathbf{b}_p : A\mathbf{b}_1 \ \cdots \ A\mathbf{b}_p : \cdots : A^{n-1}\mathbf{b}_1 \ \cdots \ A^{n-1}\mathbf{b}_p]$$

- To search linearly independent columns of  $\mathcal{C}$  from left to right, note that if  $A^i\mathbf{b}_m$  is linearly dependent on its left hand side columns, so is  $A^{i+k}\mathbf{b}_m$  ( $k > 0$ ).
- Let  $\mu_m$ ,  $m = 1, 2, \dots, p$  be the number of linearly independent columns in  $\mathcal{C}$  associated with  $\mathbf{b}_m$ . Then, if  $(A, B)$  is controllable,

$$\mu_1 + \mu_2 + \cdots + \mu_p = n$$

- The set  $\{\mu_1, \mu_2, \dots, \mu_p\}$  is called *the controllability indices* and

$$\mu = \max\{\mu_1, \mu_2, \dots, \mu_p\}$$

is the *controllability index* of  $(A, B)$ .

- $\mu$  is the least integer such that

$$\text{rank}(\mathcal{C}_1) = \text{rank}[B \ AB \ \dots \ A^{\mu-1}B] = n$$

- It is clear that

$$n/p \leq \mu \leq n - p + 1$$

**Corollary 6.1** *Let  $B$  be of full column rank.  $(A, B)$  is controllable if and only if*

$$\text{rank}[B \ AB \ \dots \ A^{n-p}B] = n$$

## Example 6.6

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 3 & 0 & 0 & 2 \\ 0 & 0 & 0 & 1 \\ 0 & -2 & 0 & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 0 & 0 \\ 1 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix}$$

Since  $\text{rank}(B) = 2 = p$  and  $n - p = 2$ ,

$$[B \ AB \ A^2B] = \begin{bmatrix} 0 & 0 & 1 & 0 & 0 & 2 \\ 1 & 0 & 0 & 2 & -1 & 0 \\ 0 & 0 & 0 & 1 & -2 & 0 \\ 0 & 1 & -2 & 0 & 0 & -4 \end{bmatrix}$$

is of rank 4. The system is controllable.

**Theorem 6.3** *The controllability property is invariant under any equivalence transformation.*

*Proof* Consider  $(A, B)$  and its equivalent pair  $(\bar{A}, \bar{B})$ , where  $\bar{A} = TAT^{-1}$ ,  $\bar{B} = TB$ . Note that  $\bar{A}^k = TA^kT^{-1}$  and  $\bar{A}^k\bar{B} = TA^kB$ . Then, the controllability matrix of  $(\bar{A}, \bar{B})$  is

$$\begin{aligned}\bar{\mathcal{C}} &= [\bar{B} \ \bar{A}\bar{B} \ \dots \ \bar{A}^{n-1}\bar{B}] \\ &= [TB \ TAB \ \dots \ TA^{n-1}B] \\ &= T[B \ AB \ \dots \ A^{n-1}B] \\ &= T\mathcal{C}\end{aligned}$$

Since  $T$  is non-singular,

$$\text{rank}(\bar{\mathcal{C}}) = \text{rank}(\mathcal{C})$$

The result follows.

**Theorem 6.4** *The set of the controllability indices of  $(A, B)$  is invariant under any equivalence transformation and any reordering of the columns of  $B$ .*

*Proof* Define

$$\mathcal{C}_k = [B \ AB \ \cdots \ A^{k-1}B]$$

Then,  $\text{rank}(\mathcal{C}_k) = \text{rank}(\bar{\mathcal{C}}_k)$ . Hence, the controllability indices is invariant under any equivalence transformation.

Next, the reordering of the columns of  $B$  can be expressed as

$$\hat{B} = BM$$

where  $M \in \mathbb{R}^{p \times p}$  is an elementary matrix which is non-singular. Then,

$$\hat{\mathcal{C}}_k = [\hat{B} \ A\hat{B} \ \cdots \ A^{k-1}\hat{B}] = \mathcal{C}_k \text{diag}\{M, M, M, M\}$$

Since  $M$  is non-singular,

$$\text{rank}(\hat{\mathcal{C}}_k) = \text{rank}(\mathcal{C}_k)$$

That is, the reordering of the columns of  $B$  does not change the controllability indices

## Observability

- The observability studies the effect of the state on the output. Hence, WLOG, we consider the system (1)-(2) with zero-input, i.e.

$$\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t), \quad \mathbf{x}(t_0) = \mathbf{x}_0 \quad (7)$$

$$\mathbf{y}(t) = C(t)\mathbf{x}(t) \quad (8)$$

- Definition.** The above system is called *observable* on  $[t_0, t_f]$  if any initial state  $\mathbf{x}(t_0) = \mathbf{x}_0$  is uniquely determined by the output  $\mathbf{y}(t)$  for  $t \in [t_0, t_f]$ .

**Theorem 6.5** *The above system is observable on  $[t_0, t_f]$  if and only if the  $n \times n$  matrix*

$$M(t_0, t_f) = \int_{t_0}^{t_f} \Phi^T(t, t_0)C^T(t)C(t)\Phi(t, t_0)dt \quad (9)$$

*is invertible.*

*Proof.* Note the solution

$$\mathbf{y}(t) = C(t)\Phi(t, t_0)\mathbf{x}_0$$

Multiplying the above from the left by  $\Phi^T(t, t_0)C^T(t)$  and integrating yields

$$\int_{t_0}^{t_f} \Phi^T(t, t_0)C^T(t)\mathbf{y}(t)dt = M(t_0, t_f)\mathbf{x}_0 \quad (10)$$

It is clear from (10) that if  $M(t_0, t_f)$  is invertible,  $\mathbf{x}_0$  is uniquely determined. On the other hand, if  $M(t_0, t_f)$  is not invertible, then there exists a non-zero vector  $\mathbf{x}_a$  such that  $M(t_0, t_f)\mathbf{x}_a = 0$ , implying  $\mathbf{x}_a^T M(t_0, t_f)\mathbf{x}_a = 0$ . Using (10), we know

$$\int_{t_0}^{t_f} \mathbf{x}_a^T \Phi^T(t, t_0)C^T(t)C(t)\Phi(t, t_0)\mathbf{x}_a dt = 0$$

which implies

$$C(t)\Phi(t, t_0)\mathbf{x}_a = 0, \quad t \in [t_0, t_f]$$

Thus,  $\mathbf{x}(t_0) = \mathbf{x}_0 + \mathbf{x}_a$  gives the same zero input response on  $[t_0, t_f]$  as  $\mathbf{x}_0$ . Hence, the system is not observable.

**Remark 6.4** *The matrix  $M(t_0, t_f)$  is called observable Gramian. It is always symmetric and positive semidefinite and it is positive definite if and only if the system is observable.*

Observability has important applications in the design of observer and estimator. That system is observable allows us to design an observer/estimator to estimate the state  $\mathbf{x}(t)$  based on the output  $\mathbf{y}(t)$ .

## Linear Time-invariant Systems

$$A(t) = A, \quad C(t) = C$$

**Theorem 6.6** *The following are equivalent:*

- (i) *The system or pair  $(C, A)$  is observable;*
- (ii) *The matrix*

$$W_o(t) = \int_0^t e^{A^T \tau} C^T C e^{A\tau} d\tau$$

*is positive definite for any  $t > 0$ ;*

- (iii) *The observability matrix*

$$\mathcal{O} = \begin{bmatrix} C \\ CA \\ CA^2 \\ \vdots \\ CA^{n-1} \end{bmatrix}$$

*has full column rank  $n$ ;*

- (iv) *The matrix  $\begin{bmatrix} A - \lambda I \\ C \end{bmatrix}$  has full column rank (rank  $n$ ) for all  $\lambda$  in complex domain;*

- (v) Let  $\lambda$  and  $\mathbf{y}$  be any eigenvalue and any corresponding right eigenvector of  $A$ , i.e.  $A\mathbf{y} = \lambda\mathbf{y}$ , then  $C\mathbf{y} \neq 0$ ;
- (vi) The eigenvalues of  $A + LC$  can be freely assigned by a suitable choice of  $F$ .
- (vi)  $(A^T, C^T)$  is controllable.

*Proof.* (i)  $\iff$  (iii): Under zero input,

$$\mathbf{y}(t) = Ce^{At}\mathbf{x}(0), \quad t \in [0, t_f]$$

from which we have

$$\begin{bmatrix} \mathbf{y}(0) \\ \dot{\mathbf{y}}(0) \\ \vdots \\ \mathbf{y}^{(n-1)}(0) \end{bmatrix} = \begin{bmatrix} C \\ CA \\ CA^2 \\ \vdots \\ CA^{n-1} \end{bmatrix} \mathbf{x}(0)$$

Since  $\mathcal{O}$  is of full column rank,  $\mathbf{x}(0)$  can be uniquely determined. Hence, the system is observable.

(iii)  $\iff$  (i): By contradiction, assume that  $(C, A)$  is observable but  $\mathcal{O}$  does not have full column rank. Then, there exists a vector  $\mathbf{x}_0$  such that  $\mathcal{O}\mathbf{x}_0 = 0$  or  $CA^i\mathbf{x}_0 = 0, \forall i \geq 0$  (by Cayley-Hamilton Theorem). Choose  $\mathbf{x}(t_0) = \mathbf{x}_0$ . Then

$$\mathbf{y}(t) = Ce^{At}\mathbf{x}(t_0) = 0$$

This implies that the system is not observable as  $\mathbf{x}(t_0)$  cannot be determined from  $\mathbf{y}(t) = 0$ .

The rest of the proof follows by the duality.

**Remark 6.5** If  $A$  is asymptotically stable and  $(C, A)$  is observable, then the unique solution of

$$W_o A + A^T W_o = -C^T C$$

is positive definite and the solution is the observability Gramian:

$$W_o = \int_0^\infty e^{A^T \tau} C^T C e^{A \tau} d\tau$$

**Example 6.7 Observability of OCF form**

$$A = \begin{bmatrix} -a_2 & 1 & 0 \\ -a_1 & 0 & 1 \\ -a_0 & 0 & 0 \end{bmatrix}, \quad C = [1 \ 0 \ 0]$$

*Observability matrix is*

$$\mathcal{O} = \begin{bmatrix} C \\ CA \\ CA^2 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ -a_2 & 1 & 0 \\ a_2^2 - a_1 & -a_2 & 1 \end{bmatrix}, \quad \text{rank}(\mathcal{O}) = 3$$

*So, OCF form is always observable.*

**Example 6.8** Consider the system

$$A = \begin{bmatrix} -2 & 0 & 0 \\ 1 & 0 & 2 \\ 0 & 0 & 0 \end{bmatrix}, \quad C = [1 \quad 0 \quad 1]$$

The observability matrix is

$$\mathcal{O} = \begin{bmatrix} 1 & 0 & 1 \\ -2 & 0 & 0 \\ 4 & 0 & 0 \end{bmatrix}, \quad \text{rank}(\mathcal{O}) = 2$$

The system is not observable.

In fact, by similarity transformation  $\bar{x} = Tx = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 1 \\ 0 & 1 & 0 \end{bmatrix} x$ , we obtain

$$\bar{A} = T^{-1}AT = \begin{bmatrix} -2 & 0 & 0 \\ 0 & 0 & 0 \\ 1 & 2 & 0 \end{bmatrix}, \quad \bar{C} = CT = [1 \quad 1 \quad 0]$$

It can be seen that  $\bar{x}_3$  is not observable.

**Example 6.9** Verify if the system below is observable.

$$\begin{aligned}\dot{x} &= \begin{bmatrix} -1 & -4 & -2 \\ 0 & 6 & -1 \\ 1 & 7 & -1 \end{bmatrix} x, \quad n = 3 \\ y &= \begin{bmatrix} 0 & 2 & 1 \\ 1 & 1 & 0 \end{bmatrix} x\end{aligned}$$

The observability matrix is

$$\text{rank}(\mathcal{O}) = \text{rank} \begin{bmatrix} C \\ CA \\ CA^2 \end{bmatrix} = \text{rank} \begin{bmatrix} 0 & 2 & 1 \\ 1 & 1 & 0 \\ 1 & 19 & -3 \\ * & * & * \\ * & * & * \\ * & * & * \end{bmatrix} = 3 = n$$

The system is observable.

**Example 6.10** Determine if the following system is observable using PBH test.

$$\begin{aligned}\dot{x} &= \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & -1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 5 & 0 \end{bmatrix} x, \quad n = 4 \\ y &= \begin{bmatrix} 0 & 1 & 0 & -2 \\ 1 & 0 & 1 & 0 \end{bmatrix} x\end{aligned}$$

Note that

$$\begin{bmatrix} \lambda I - A \\ C \end{bmatrix} = \begin{bmatrix} \lambda & -1 & 0 & 0 \\ 0 & \lambda & 1 & 0 \\ 0 & 0 & \lambda & -1 \\ 0 & 0 & -5 & \lambda \\ 0 & 1 & 0 & -2 \\ 1 & 0 & 1 & 0 \end{bmatrix}$$

The eigenvalues of  $A$ :  $\lambda_1 = \lambda_2 = 0$ ,  $\lambda_3 = \sqrt{5}$ ,  $\lambda_4 = -\sqrt{5}$ .

$$\text{rank} \begin{bmatrix} \lambda_{1,2}I - A \\ C \end{bmatrix} = \text{rank} \begin{bmatrix} 0 & -1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & -1 \\ 0 & 0 & -5 & 0 \\ 0 & 1 & 0 & -2 \\ 1 & 0 & 1 & 0 \end{bmatrix} = 4 = n$$

$$\text{rank} \begin{bmatrix} \lambda_3I - A \\ C \end{bmatrix} = \text{rank} \begin{bmatrix} \sqrt{5} & -1 & 0 & 0 \\ 0 & \sqrt{5} & 1 & 0 \\ 0 & 0 & \sqrt{5} & -1 \\ 0 & 0 & -5 & \sqrt{5} \\ 0 & 1 & 0 & -2 \\ 1 & 0 & 1 & 0 \end{bmatrix} = 4 = n$$

$$\text{rank} \begin{bmatrix} \lambda_3I - A \\ C \end{bmatrix} = \text{rank} \begin{bmatrix} -\sqrt{5} & -1 & 0 & 0 \\ 0 & -\sqrt{5} & 1 & 0 \\ 0 & 0 & -\sqrt{5} & -1 \\ 0 & 0 & -5 & -\sqrt{5} \\ 0 & 1 & 0 & -2 \\ 1 & 0 & 1 & 0 \end{bmatrix} = 4 = n$$

So, the system is observable.

## Observability Indices

- $C \in \mathcal{R}^{q \times n}$  with  $C$  full row rank.
- Let  $\mathbf{c}_i$  be the  $i$ th row of  $C$ . To search linearly independent rows of  $\mathcal{O}$  from top to bottom, note that if  $\mathbf{c}_m A^i$  is linearly dependent on its rows above, so is  $\mathbf{c}_m A^{i+k}$  ( $k > 0$ ).
- Let  $\nu_m$ ,  $m = 1, 2, \dots, q$  be the number of linearly independent rows in  $\mathcal{O}$  associated with  $\mathbf{c}_m$ . Then, if  $(A, C)$  is observable,

$$\nu_1 + \nu_2 + \cdots + \nu_q = n$$

- The set  $\{\nu_1, \nu_2, \dots, \nu_q\}$  is called *the observability indices* and

$$\nu = \max\{\nu_1, \nu_2, \dots, \nu_q\}$$

is the *observability index* of  $(A, C)$ .

- $\nu$  is the least integer such that

$$\text{rank}(\mathcal{O}_1) = \text{rank} \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{\nu-1} \end{bmatrix} = n$$

- It is clear that

$$n/q \leq \nu \leq n - q + 1$$

**Corollary 6.2** Let  $C$  be of full row rank.  $(A, C)$  is observable if and only if

$$\text{rank} \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-q} \end{bmatrix} = n$$

**Theorem 6.7** The observability property is invariant under any equivalence transformation.

**Theorem 6.8** The set of the observability indices of  $(A, C)$  is invariant under any equivalence transformation and any reordering of the rows of  $C$ .

**Definition.** Let  $\lambda$  be an eigenvalue of  $A$ , or equivalently, a mode of the system.

The mode  $\lambda$  is said to be controllable (observable) if for all left (right) eigenvectors associated with  $\lambda$ , i.e.  $\mathbf{x}^* A = \lambda \mathbf{x}^*$  ( $A\mathbf{x} = \lambda \mathbf{x}$ ),  $\mathbf{x}^* B \neq 0$  ( $C\mathbf{x} \neq 0$ ).

Otherwise, the mode is said to be uncontrollable (unobservable).

### Example 6.9

$$A = \begin{bmatrix} \lambda_1 & 0 \\ 0 & \lambda_2 \end{bmatrix}, \quad B = \begin{bmatrix} 1 \\ \alpha \end{bmatrix}, \quad C = [1 \ \beta]$$

If  $\alpha = 0$ , the mode  $\lambda_2$  is not controllable; if  $\beta = 0$ , the mode  $\lambda_2$  is not observable.

## Stabilizability (Time-invariant Case)

- **Definition.** The system (1) or the pair  $(A, B)$  is said to be stabilizable if there exists a state feedback  $\mathbf{u} = F\mathbf{x}$  such that the system is stable, i.e.  $A + BF$  is stable.
- Obviously, that a system is controllable implies that it is stabilizable. The converse is not necessarily true. Stabilizability basically implies that the unstable modes of the system are controllable.

**Theorem 6.9** *The following are equivalent:*

- $(A, B)$  is stabilizable;*
- The matrix  $[A - \lambda I, B]$  has full row rank for all  $Re\lambda \geq 0$ ;*
- For all  $\lambda$  and  $\mathbf{x}$  such that  $\mathbf{x}^* A = \mathbf{x}^* \lambda$  and  $Re\lambda \geq 0$ ,  $\mathbf{x}^* B \neq 0$ ;*
- There exists a matrix  $F$  such that  $A + BF$  is Hurwitz.*

## Detectability (Time-invariant Case)

- **Definition.** The system (1)-(2) or the pair  $(C, A)$  is said to be *detectable* if  $A + LC$  is stable for some  $L$ .
- Obviously, that a system is observable implies that it is detectable. The converse is not necessarily true. Detectability basically implies that the unstable modes of the system are observable.

**Theorem 6.10** *The following are equivalent:*

- (i)  $(C, A)$  is detectable;
- (ii) The matrix  $\begin{bmatrix} A - \lambda I \\ C \end{bmatrix}$  has full column rank for all  $\text{Re}\lambda \geq 0$ ;
- (iii) For all  $\lambda$  and  $\mathbf{x}$  such that  $A\mathbf{x} = \lambda\mathbf{x}$  and  $\text{Re}\lambda \geq 0$ ,  $C\mathbf{x} \neq 0$ ;
- (iv) There exists a matrix  $L$  such that  $A + LC$  is Hurwitz;
- (v)  $(A^T, C^T)$  is stabilizable.

## Kalman Canonical Decomposition

- For a given system, there are infinite number of state-space realizations. In fact, let a state-space representation of a system be

$$\dot{\mathbf{x}}(t) = A\mathbf{x}(t) + B\mathbf{u}(t) \quad (11)$$

$$\mathbf{y}(t) = C\mathbf{x}(t) + D\mathbf{u}(t) \quad (12)$$

A nonsingular transformation  $\mathbf{z} = T\mathbf{x}$  will define another state-space realization of the same system:

$$\dot{\mathbf{z}}(t) = TAT^{-1}\mathbf{z}(t) + TB\mathbf{u}(t) \quad (13)$$

$$\mathbf{y}(t) = CT^{-1}\mathbf{z}(t) + D\mathbf{u}(t) \quad (14)$$

- It is easy to check that they have the same transfer function.

$$G(s) = C(sI - A)^{-1}B + D = CT^{-1}(sI - TAT^{-1})^{-1}TB + D$$

- Although in many applications, state-space variables will be chosen to have some physical meanings such as position, velocity, voltage, current and etc., there are some state-space realizations which make system analysis and synthesis much easier.

- The controllability and observability matrices of the systems (11)-(12) and (13)-(14) are related by

$$\mathcal{C}_z = T\mathcal{C}, \quad \mathcal{O}_z = OT^{-1}$$

- The controllability (or stabilizability) and observability (or detectability) are invariant under similarity transformations.
- Theorem.** If  $\text{rank}(\mathcal{C}) = k_1 < n$ , then there exists a similarity transformation

$$\mathbf{z} = \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \end{bmatrix} = T\mathbf{x}$$

such that

$$\begin{bmatrix} \dot{\mathbf{z}}_1 \\ \dot{\mathbf{z}}_2 \end{bmatrix} = \begin{bmatrix} \bar{A}_{11} & \bar{A}_{12} \\ 0 & \bar{A}_{22} \end{bmatrix} \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \end{bmatrix} + \begin{bmatrix} \bar{B}_1 \\ 0 \end{bmatrix} \mathbf{u} \quad (15)$$

$$\mathbf{y} = [\bar{C}_1 \ \bar{C}_2] \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \end{bmatrix} + D\mathbf{u} \quad (16)$$

with  $\bar{A}_{11} \in \mathcal{R}^{k_1 \times k_1}$  and  $(\bar{A}_{11}, \bar{B}_1)$  controllable. Moreover,

$$G(s) = C(sI - A)^{-1}B + D = \bar{C}_1(sI - \bar{A}_{11})^{-1}\bar{B}_1 + D$$

*Proof.* Since  $\text{rank}(\mathcal{C}) = k_1 < n$ , there exist  $k_1$  linearly independent columns  $q_i$ ,  $i = 1, 2, \dots, k_1$  in  $\mathcal{C}$ . Let  $q_i$ ,  $i = k_1 + 1, \dots, n$  be any  $n - k_1$  linearly independent vectors such that the matrix

$$Q = [q_1 \ \cdots \ q_{k_1} \ q_{k_1+1} \ \cdots \ q_n]$$

is nonsingular. Then, the transformation  $T = Q^{-1}$  will be the desired one.

To show this, since  $Aq_i$ ,  $i = 1, 2, \dots, k_1$  is a linear combination of the columns of  $\mathcal{C}$  by Cayley-Hamilton Theorem,  $Aq_i$  can be written as a linear combination of  $q_i$ ,  $i = 1, 2, \dots, k_1$ . Hence,

$$\begin{aligned} AT^{-1} &= [Aq_1 \ \cdots \ Aq_{k_1} \ Aq_{k_1+1} \ \cdots \ Aq_n] \\ &= [q_1 \ \cdots \ q_{k_1} \ q_{k_1+1} \ \cdots \ q_n] \begin{bmatrix} \bar{A}_{11} & \bar{A}_{12} \\ 0 & \bar{A}_{22} \end{bmatrix} \\ &= T^{-1} \begin{bmatrix} \bar{A}_{11} & \bar{A}_{12} \\ 0 & \bar{A}_{22} \end{bmatrix} \end{aligned}$$

Similarly, each column of the matrix  $B$  is a linear combination of  $q_i$ ,  $i = 1, 2, \dots, k_1$ , hence

$$B = Q \begin{bmatrix} \bar{B}_1 \\ 0 \end{bmatrix} = T^{-1} \begin{bmatrix} \bar{B}_1 \\ 0 \end{bmatrix}$$

To show  $(\bar{A}_{11}, \bar{B}_1)$  is controllable, note that

$$T\mathcal{C} = \begin{bmatrix} \bar{B}_1 & \bar{A}_{11}\bar{B}_1 & \cdots & \bar{A}_{11}^{k_1-1}\bar{B}_1 & \cdots & \bar{A}_{11}^{n-1}\bar{B}_1 \\ 0 & 0 & \cdots & 0 & \cdots & 0 \end{bmatrix}$$

Again, by Cayley Hamilton theorem,  $\bar{A}_{11}^j$ ,  $j \geq k_1$  is a linear combination of  $\bar{A}_{11}^i$ ,  $i = 0, 1, \dots, k_1 - 1$ . Hence,

$$\text{rank}[\bar{B}_1 \ \bar{A}_{11}\bar{B}_1 \ \cdots \ \bar{A}_{11}^{k_1-1}\bar{B}_1] = \text{rank}(\mathcal{C}) = k_1$$

i.e.  $(\bar{A}_{11}, \bar{B}_1)$  is controllable.

- The above decomposition clearly gives the controllable and uncontrollable modes of the system. The uncontrollable modes, i.e. the eigenvalues of  $\bar{A}_{22}$  will be cancelled with zeros when computing the transfer function  $G(s)$ .
- The space  $\mathbf{z}$  is partitioned into the controllable subspace  $\begin{bmatrix} \mathbf{z}_1 \\ 0 \end{bmatrix}$  and uncontrollable subspace  $\begin{bmatrix} 0 \\ \mathbf{z}_2 \end{bmatrix}$ .
- A numerical reliable way of computing  $T$  is the QR factorization. In fact, let  $\mathcal{C} = QR$ , then  $T = Q^{-1}$ .
- If the system is stabilizable and  $\text{rank}(\mathcal{C}) = k_1 < n$ , then the decomposed system will have stable  $\bar{A}_{22}$ .

### Example 6.10

$$\begin{aligned}\dot{\mathbf{x}} &= \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 0 \\ 0 & 1 & 1 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ 0 & 1 \end{bmatrix} \mathbf{u} \\ \mathbf{y} &= [1 \ 1 \ 1] \mathbf{x}\end{aligned}$$

Since  $\text{rank}(B) = 2$  and

$$[B \ AB] = \begin{bmatrix} 0 & 1 & 1 & 1 \\ 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 \end{bmatrix}$$

is of rank  $2 < 3$ , the system is uncontrollable.

Let  $T^{-1} = \begin{bmatrix} 0 & 1 & 1 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}$ , where the first two columns are the two first linearly independent columns of  $[B \ AB]$  and the last column is arbitrarily chosen to make  $T$  non-singular.

Then,

$$\bar{A} = TAT^{-1} = \begin{bmatrix} 1 & 0 & \vdots & 0 \\ 1 & 1 & \vdots & 0 \\ \dots & \dots & \dots & \dots \\ 0 & 0 & \vdots & 1 \end{bmatrix}$$

$$\bar{B} = TB = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ \dots & \dots \\ 0 & 0 \end{bmatrix}$$

$$\bar{C} = CT^{-1} = [1 \ 2 : 1]$$

Note that  $A_{12}$  happens to be zero in this example. Thus, the system can be reduced to

$$\dot{\mathbf{z}}_1 = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} \mathbf{z}_1 + \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \mathbf{u}, \quad \mathbf{y} = [1 \ 2] \mathbf{z}_1$$

**Theorem 6.11** If the observability matrix with  $\text{rank}(\mathcal{O}) = k_2 < n$ , then there exists a similarity transformation

$$\mathbf{z} = \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \end{bmatrix} = T\mathbf{x}$$

such that

$$\begin{bmatrix} \dot{\mathbf{z}}_1 \\ \dot{\mathbf{z}}_2 \end{bmatrix} = \begin{bmatrix} \bar{A}_{11} & 0 \\ \bar{A}_{21} & \bar{A}_{22} \end{bmatrix} \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \end{bmatrix} + \begin{bmatrix} \bar{B}_1 \\ \bar{B}_2 \end{bmatrix} \mathbf{u} \quad (17)$$

$$\mathbf{y} = [\bar{C}_1 \ 0] \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \end{bmatrix} + Du \quad (18)$$

with  $\bar{A}_{11} \in \mathcal{R}^{k_2 \times k_2}$  and  $(\bar{C}_1, \bar{A}_{11})$  observable. Moreover,

$$G(s) = C(sI - A)^{-1}B + D = \bar{C}_1(sI - \bar{A}_{11})^{-1}\bar{B}_1 + D$$

If the system is detectable and  $\text{rank}(\mathcal{O}) = k_2 < n$ , then the decomposed system will have stable  $\bar{A}_{22}$ .

Since  $\text{rank}(\mathcal{O}) = k_2 < n$ , there exist  $k_2$  linearly independent rows  $p_1, \dots, p_{k_2}$  in  $\mathcal{O}$ . Let  $p_i$ ,  $i = k_2 + 1, \dots, n$  be any  $n - k_2$  linearly independent row vectors such that the matrix

$$T = \begin{bmatrix} p_1 \\ p_2 \\ \vdots \\ p_n \end{bmatrix}$$

is invertible. Then,  $\bar{A} = TAT^{-1}$ ,  $\bar{B} = TB$ ,  $\bar{C} = CT^{-1}$  are of the form in the theorem.

### Example 6.11

$$\begin{aligned}\dot{\mathbf{x}} &= \begin{bmatrix} -2 & 0 & 0 \\ 1 & -2 & 0 \\ 0 & 0 & -2 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ 0 & 1 \end{bmatrix} \mathbf{u} \\ \mathbf{y} &= [1 \ 0 \ 1] \mathbf{x}\end{aligned}$$

The observability matrix

$$\mathcal{O} = \begin{bmatrix} 1 & 0 & 1 \\ -2 & 0 & -2 \\ 4 & 0 & 4 \end{bmatrix}, \quad \text{rank}(\mathcal{O}) = 1 < 3$$

The system is not observable.

Let  $T = \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$ . Then,

$$T^{-1} = \begin{bmatrix} 1 & 0 & -1 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

Thus,

$$\bar{A} = TAT^{-1} = \begin{bmatrix} -2 & 0 & 0 \\ 1 & -2 & -1 \\ 0 & 0 & -2 \end{bmatrix}, \quad \bar{B} = TB = \begin{bmatrix} 0 & 2 \\ 1 & 0 \\ 0 & 1 \end{bmatrix},$$

$$\bar{C} = CT^{-1} = [1 \ 0 \ 0]$$

## Kalman Canonical Decomposition

**Theorem 6.12** For a LTI system (11)-(12), there exists a nonsingular transformation  $\mathbf{z} = T\mathbf{x}$  such that

$$\begin{bmatrix} \dot{\mathbf{z}}_1 \\ \dot{\mathbf{z}}_2 \\ \dot{\mathbf{z}}_3 \\ \dot{\mathbf{z}}_4 \end{bmatrix} = \begin{bmatrix} \bar{A}_{co} & 0 & \bar{A}_{13} & 0 \\ \bar{A}_{21} & \bar{A}_{c\bar{o}} & \bar{A}_{23} & \bar{A}_{24} \\ 0 & 0 & \bar{A}_{\bar{c}o} & 0 \\ 0 & 0 & \bar{A}_{43} & \bar{A}_{\bar{c}\bar{o}} \end{bmatrix} \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \\ \mathbf{z}_3 \\ \mathbf{z}_4 \end{bmatrix} + \begin{bmatrix} \bar{B}_1 \\ \bar{B}_2 \\ 0 \\ 0 \end{bmatrix} \mathbf{u} \quad (19)$$

$$\mathbf{y} = [\bar{C}_1 \ 0 \ \bar{C}_3 \ 0] \begin{bmatrix} \mathbf{z}_1 \\ \mathbf{z}_2 \\ \mathbf{z}_3 \\ \mathbf{z}_4 \end{bmatrix} + D\mathbf{u} \quad (20)$$

where  $\mathbf{z}_1$  is both controllable and observable,  $\mathbf{z}_2$  is controllable but unobservable,  $\mathbf{z}_3$  is observable but uncontrollable and  $\mathbf{z}_4$  is both uncontrollable and unobservable. Further,  $G(s) = \bar{C}_1(sI - \bar{A}_{co})^{-1}\bar{B}_1 + D$ .

**Remark 6.6** Although the input and output relationship only depends on the subsystem which is both controllable and observable, the internal behaviour including stability depends on all the subsystems.

## Discrete-time Systems

Consider the system

$$\mathbf{x}(k+1) = A\mathbf{x}(k) + B\mathbf{u}(k), \quad \mathbf{x}(k_0) = \mathbf{x}_0 \quad (21)$$

$$\mathbf{y}(k) = C\mathbf{x}(k) \quad (22)$$

where  $A$ ,  $B$ ,  $C$  are respectively  $n \times n$ ,  $n \times p$  and  $q \times n$  real constant matrices.

**Definition 6.1** *The discrete-time system (21)-(22) or the pair  $(A, B)$  is said to be controllable if for any initial state  $\mathbf{x}(0) = \mathbf{x}_0$  and any final state  $\mathbf{x}(k_f) = \mathbf{x}_f$ , there exists an input sequence of finite length that transfers  $\mathbf{x}_0$  to  $\mathbf{x}_f$ .*

**Theorem 6.13** *The following are equivalent:*

- *The pair  $(A, B)$  is controllable.*

- *The matrix*

$$W_{dc}(n-1) = \sum_{m=0}^{n-1} A^m B B^T (A^T)^m$$

*is non-singular.*

- *The  $n \times np$  controllability matrix*

$$\mathcal{C}_d = [B \ AB \ \cdots \ A^{n-1}B]$$

*has rank  $n$  (full row rank).*

- *$[A - \lambda I \ B]$  has full row rank at every eigenvalue of  $A$ .*
- *If, in addition, all eigenvalues of  $A$  have magnitudes less than 1, then the unique solution of*

$$W_{dc} - AW_{dc}A^T = BB^T$$

*is positive definite.  $W_{dc}$  is the discrete Controllability Gramian and is given by*

$$W_{dc} = \sum_{m=0}^{\infty} A^m B B^T (A^T)^m$$

Note that the solution at  $k = n$ ,

$$\mathbf{x}(n) = A^n \mathbf{x}(0) + \sum_{m=0}^{n-1} A^{n-1-m} B \mathbf{u}(m)$$

which can be expressed as

$$\mathbf{x}(n) - A^n \mathbf{x}(0) = [B \ AB \ \dots \ A^{n-1}B] \begin{bmatrix} \mathbf{u}(n-1) \\ \mathbf{u}(n-2) \\ \vdots \\ \mathbf{u}(0) \end{bmatrix}$$

It is well known that there exists a solution  $\{\mathbf{u}(0), \mathbf{u}(1), \dots, \mathbf{u}(n-1)\}$  to the above linear system of equations if and only if the controllability matrix is of full row rank ( $n$ ).

**Remark 6.7** *A linear time-invariant discrete-time system might fail to be reachable simply because the time-interval  $[0, k_f]$  is too short - This does not happen for continuous LTI systems. A SISO  $n$ -dimensional system may require  $n$  steps to reach a specified state.*

## Observability

**Definition 6.2** *The discrete-time system (21)-(22) or the pair  $(A, C)$  is said to be observable if for any unknown initial state  $\mathbf{x}(0) = \mathbf{x}_0$ , there exists a finite integer  $k_f$  such that the knowledge of the input sequence  $\mathbf{u}(k)$  and the output sequence  $\mathbf{y}(k)$  from  $k = 0$  to  $k = k_f$  suffices to determine uniquely the initial state  $\mathbf{x}(0)$ .*

**Duality** The pair  $(A, C)$  is observable if and only if  $(A^T, C^T)$  is controllable.

**Theorem 6.14** *The following are equivalent:*

- *The pair  $(A, C)$  is observable.*
- *The matrix*

$$W_{do}(n-1) = \sum_{m=0}^{n-1} (A^T)^m C^T C A^m$$

*is non-singular.*

- The  $nq \times n$  observability matrix

$$\mathcal{O}_d = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix}$$

has rank  $n$  (full column rank).

- The  $(n + q) \times n$  matrix

$$\begin{bmatrix} A - \lambda I \\ B \end{bmatrix}$$

has full column rank at every eigenvalue of  $A$ .

- If, in addition, all eigenvalues of  $A$  have magnitudes less than 1, then the unique solution of

$$W_{do} - A^T W_{do} A = C^T C$$

is positive definite.  $W_{do}$  is the discrete Observability Gramian and is given by

$$W_{do} = \sum_{m=0}^{\infty} (A^T)^m C^T C A^m$$

- In literature, there are three different controllability definitions:
  - 1 Transfer any state to any other state.
  - 2 Transfer any state to the origin, called controllability to the origin.
  - 3 Transfer zero state to any state, called controllability from the origin, more often, *reachability*.

**Remark 6.8** • For continuous-time systems, since  $e^{At}$  is non-singular, the above definitions are equivalent.

- For discrete-time systems, if  $A$  is non-singular, the three definitions are again equivalent.
- However, if  $A$  is singular, (1) and (3) are equivalent but not (2). For example,

$$\mathbf{x}(k+1) = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix} \mathbf{u}(k)$$

whose controllability matrix has rank 0. But for any  $k \geq 3$ ,  $A^k = 0$ . Thus,  $\mathbf{x}(3) = A^3\mathbf{x}(0) = 0$ .

## Controllability after Sampling

Given a continuous-time system

$$\dot{\mathbf{x}}(t) = A\mathbf{x}(t) + B\mathbf{u}(t)$$

and piecewise constant input:

$$\mathbf{u}(k) = \mathbf{u}(kT) = \mathbf{u}(t), \quad kT \leq t \leq (k+1)T$$

the discretized system is

$$\bar{\mathbf{x}}(k+1) = \bar{A}\bar{\mathbf{x}}(k) + \bar{B}\mathbf{u}(k)$$

where

$$\bar{A} = e^{AT}, \quad \bar{B} = \left( \int_0^T e^{At} dt \right) B := MB$$

Question: What is the relationship of controllability between the original and discretized systems?

Apparently, if the discretized system is controllable, the original continuous-time system must be controllable. The converse is not necessarily true.

**Theorem 6.15** Suppose  $(A, B)$  is controllable. A sufficient condition for  $(\bar{A}, \bar{B})$  with sampling period  $T$  is controllable is that

$$|Im(\lambda_i - \lambda_j)| \neq 2\pi m/T, \quad m = 1, 2, \dots$$

whenever  $Re(\lambda_i - \lambda_j) = 0$ . For the single-input case, the condition is also necessary.

**Remark 6.9** • If  $A$  has only real eigenvalues, then the discretized system with any sampling period  $T$  is always controllable if the original continuous-time system is.

- Suppose  $A$  has complex conjugate eigenvalues  $\alpha \pm j\beta$ . If  $T \neq m\pi/\beta$ ,  $m = 1, 2, \dots$ , then the discretized system is controllable. If  $T = m\pi/\beta$ ,  $m = 1, 2, \dots$ , then the discretized system may be uncontrollable. This is because for  $\lambda_1 = \alpha + j\beta$  and  $\lambda_2 = \alpha - j\beta$ , the corresponding eigenvalues of  $\bar{A}$  become a repeated eigenvalue of  $-e^{\alpha T}$  or  $e^{\alpha T}$  which causes the discretized system uncontrollable.

**Example 6.12** Consider the continuous-time system

$$\begin{aligned}\dot{\mathbf{x}} &= \begin{bmatrix} -3 & -7 & -5 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} \mathbf{u} \\ y &= [0 \ 1 \ 2] \mathbf{x}\end{aligned}$$

The eigenvalues of  $A$  are  $-1, -1 \pm j2$ . All three have the same real part. Thus, the discretized system is controllable if and only if

$$T \neq \frac{2\pi m}{2} = \pi m \quad \text{and} \quad T \neq \frac{2\pi m}{4} = 0.5\pi m$$

for  $m = 1, 2, \dots$ . For example, if setting  $T = 0.5\pi$  ( $m = 1$ ). Then, by Matlab:  
`[ad,bd]=c2d(a,b,pi/2)`  
we get

$$\bar{A} = \begin{bmatrix} -0.1039 & 0.2079 & 0.5197 \\ -0.1390 & -0.4158 & -0.5197 \\ 0.1039 & 0.2079 & 0.3118 \end{bmatrix}, \quad \bar{B} = \begin{bmatrix} -0.1039 \\ 0.1039 \\ 0.1376 \end{bmatrix}$$

It is easy to check that the controllability matrix is of rank 2. Hence, the discretized system is uncontrollable.

## Summary

- Concepts of controllability, observability, stabilizability, and detectability
- Determine if a LTI system is controllable and observable by checking the rank of controllability and observability
- Kalman decomposition
- Controllability of sampled systems

## Exercises

1. Check the controllability and observability of the system with

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 2 & -1 \end{bmatrix}, \quad B = \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ 0 & 0 \end{bmatrix}, \quad C = [1 \ 0 \ 1]$$

2. Show that the state equation

$$\dot{x} = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix}x + \begin{bmatrix} B_1 \\ 0 \end{bmatrix}u$$

is controllable if and only if the pair  $(A_{22}, A_{21})$  is controllable.

3. Reduce the state equation

$$\begin{aligned} \dot{x} &= \begin{bmatrix} \lambda_1 & 1 & 0 & 0 & 0 \\ 0 & \lambda_1 & 1 & 0 & 0 \\ 0 & 0 & \lambda_1 & 0 & 0 \\ 0 & 0 & 0 & \lambda_2 & 1 \\ 0 & 0 & 0 & 0 & \lambda_2 \end{bmatrix}x + \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \\ 1 \end{bmatrix}u \\ y &= [0 \ 1 \ 1 \ 1 \ 0 \ 1]x \end{aligned}$$

to a controllable and observable equation.

4. Consider the system

$$\begin{aligned}\dot{\mathbf{x}} &= \begin{bmatrix} \lambda_1 & 0 & 0 & 0 & 0 \\ 0 & \alpha_1 & \beta_1 & 0 & 0 \\ 0 & -\beta & \alpha_1 & 0 & 0 \\ 0 & 0 & 0 & \alpha_2 & \beta_2 \\ 0 & 0 & 0 & -\beta_2 & \alpha_2 \end{bmatrix} \mathbf{x} + \begin{bmatrix} b_1 \\ b_{11} \\ b_{12} \\ b_{21} \\ b_{22} \end{bmatrix} \mathbf{u} \\ y &= [c_1 \ c_{11} \ c_{12} \ c_{21} \ c_{22}] \mathbf{x}\end{aligned}$$

The system has one real eigenvalue and two pairs of complex conjugate eigenvalues. It is assumed that they are distinct. Show that the system is controllable if and only if  $b_1 \neq 0$ ;  $b_{i1} \neq 0$  or  $b_{i2} \neq 0$  for  $i = 1, 2$ . It is observable if and only if  $c_1 \neq 0$ ;  $c_{i1} \neq 0$  or  $c_{i2} \neq 0$  for  $i = 1, 2$ .