# THE UNIVERSITY of York

# Lattice Coding and its Applications in Communications

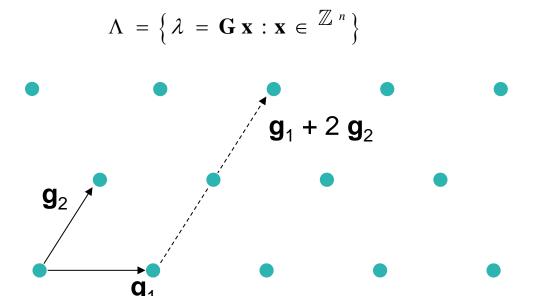
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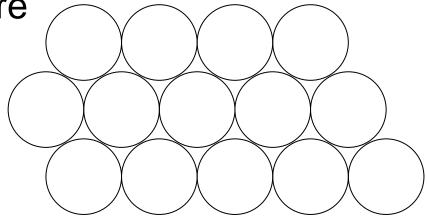
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- Introduction to lattices
  - Definition; Sphere packings; Basis vectors; Matrix description
- Codes and lattice codes
  - Shaping region; Nested lattices
- Lattice constructions
  - Construction A/D, LDLC codes; construction from Gaussian/Eisenstein integers
- Lattice encoding and decoding
  - Problems of shaping; LDLC decoding; Construction A decoding
- Lattices in multi-user networks: Compute and forward

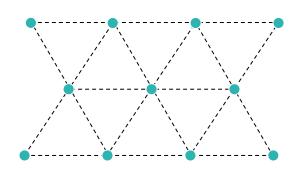
- A lattice is defined as:
  - the (infinite) set of points in an n-dimensional space given by all linear combinations with integer coefficients of a basis set of up to n linearly independent vectors
- It can be defined in terms of a generator matrix G, whose columns are the basis vectors:

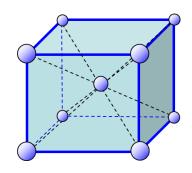


- A sphere packing is an arrangement of nonoverlapping hyperspheres of equal radius in Ndimensional space
- We are often interested in the *packing density*  $\eta$  or  $\delta_n$  of a packing
  - the proportion of space occupied by spheres
- Dense sphere packings are often lattice packings
  - have sphere centres on lattices

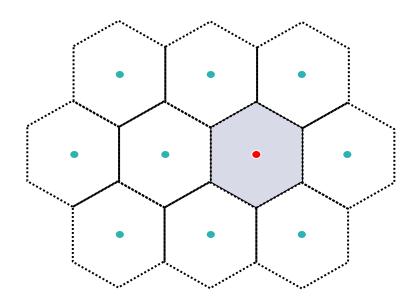


Dimensions	Lattice	Packing density	Kissing number
2	Hexagonal	$\frac{1}{6}\pi\sqrt{3}$ =0.91	6
3	BCC/FCC/HCP	$\frac{1}{6}\pi\sqrt{2}=0.74$	12
4	D4	$\frac{1}{16}\pi^2 = 0.62$	24
8	E8	$\frac{1}{384}\pi^4 = 0.25$	240
24	E24 (Leech)	$\frac{\pi^{12}}{12!} = 0.0019$	196 560





- The **Voronoi region** of a lattice point is the region of the *N*-dimensional space closer to that point than to all other lattice points
- Voronoi region of red point shown shaded



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- i.e. *forward error-correcting* (FEC) codes
- A code is a finite set of codewords of length n
  - Code contains M codewords encodes log<sub>2</sub>(M) bits
- where a codeword is a sequence of n symbols, usually drawn from a finite alphabet of size q
  - we will often assume the alphabet is a Galois field ( $\mathbb{F}_q$  or GF(q)) or a ring ( $\mathcal{R}(q)$ )
- In a communication system the codewords must be translated into signals of length nT
  - representing the variation in time of some quantity,
    such as electromagnetic field strength
- Each code symbol is typically modulated to some specific real or complex value of this variable

Message:

Encode

Codeword:

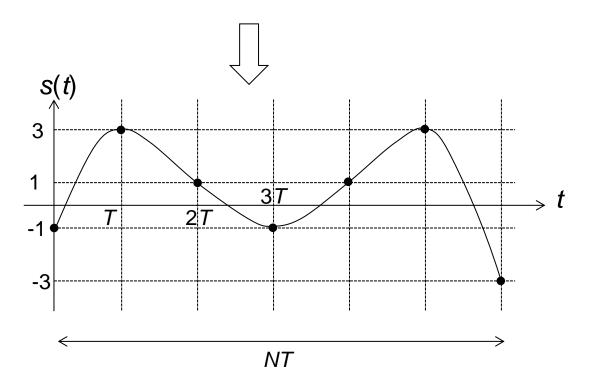
Modulate

Signal:

01111001

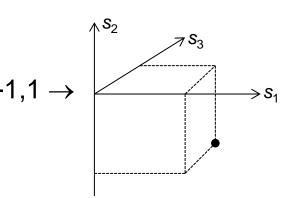


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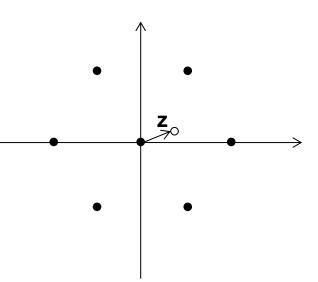


#### Geometric model

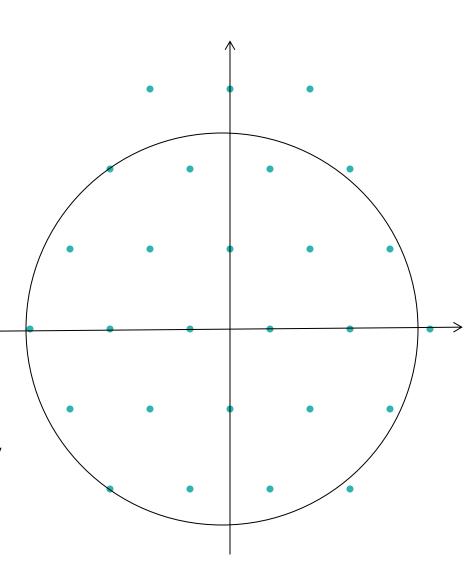
 Each coded signal can then be represented as a point in *N*-D *signal space*



- where modulated values of symbols provide the *n* coordinate values
- Code is represented by ensemble of points in signal space
- Noise on channel equivalent to vector z in signal space
- Decoder chooses closest point
- Error probability determined by minimum Euclidean distance between signal space points



- A lattice code is then defined by the (finite) set of lattice points within a certain region
  - the shaping region
  - ideally a hypersphere centred on the origin
  - this limits the maximum signal energy of the codewords
- Lattice may be offset by adding some vector

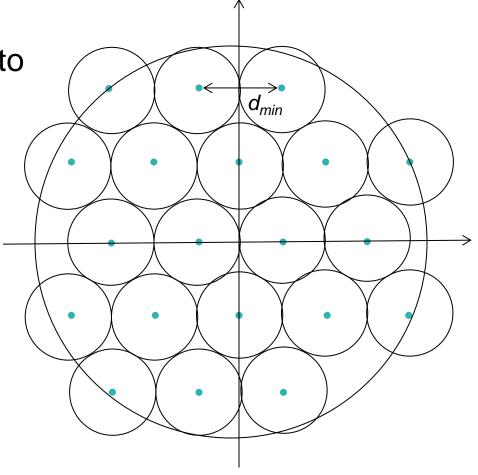


#### Minimum Euclidean distance

If the lattice is viewed as a sphere packing, then the minimum Euclidean distance must be twice the sphere radius

 Signal power S proportional to radius<sup>2</sup> of shaping region

- The greater the packing density, the greater M for given signal power
- Radius<sup>2</sup> of packed spheres proportional to maximum noise power



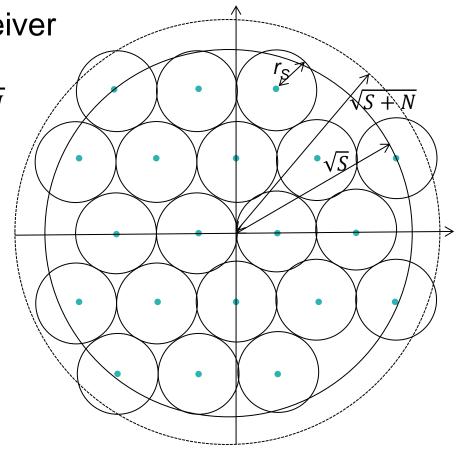
# Maximum signalling rate

- Hence for low error probability, noise power  $N \leq r_S^2$
- Radius of signal space at receiver containing signal plus noise is

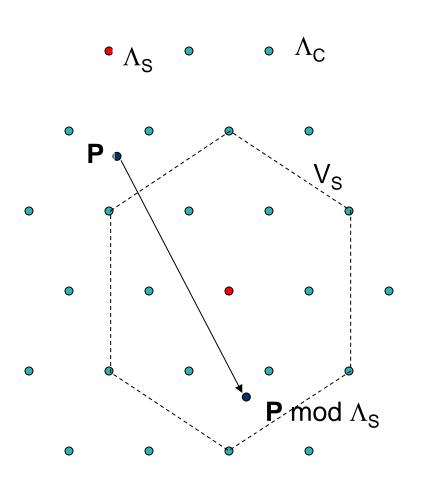
$$\sqrt{S+N}$$

- Volume of n-D sphere of radius r is  $V_n r^n$
- Hence max. no. of codewords in code

$$M \le \frac{V_n (S+N)^{n/2}}{V_N r_S^{N/2}} \le \left(\frac{S+N}{N}\right)^{n/2}$$
$$\frac{\log_2 M}{n} \le \frac{1}{2} \log_2 \left(1 + \frac{S}{N}\right)$$

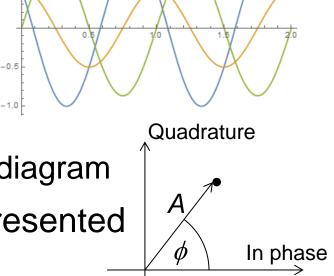


- Define fine lattice  $\Lambda_{C}$  for the code
  - plus a *coarse lattice*  $\Lambda_S$  which is a sub-lattice of  $\Lambda_C$
- Then use a Voronoi region V<sub>S</sub> of the coarse lattice as the shaping region
- Modulo- $\Lambda_{S}$  operation
  - for any point  $P \notin V_S$  find  $P (\lambda \in \Lambda_S) \in V_S$



# Complex signals

- Wireless signals consist of a sine wave carrier at the transmission frequency (MHz – GHz)
- Sine waves can be modulated in both amplitude and phase
  - hence the signal corresponding to each modulated symbol is 2-D
  - also conveniently represented as a complex value
  - typically represented on a *phasor* diagram
- Hence wireless signals can be represented in 2n dimensions
  - or *n* complex dimensions



- Introduction to lattices
- Codes and lattice codes
- Lattice constructions
  - Constructions A and D,
  - LDLC codes
  - Construction from Gaussian and Eisenstein integers
- Lattice encoding and decoding
- Lattices in multi-user networks: Compute and forward

#### THE UNIVERSITY Of York Constructions based on FEC codes

- For practical purposes in communications, we require lattices in very large numbers of dimensions
  - typically 1000, 10 000, 100 000...
- Lattices of this sort of dimension most easily constructed using FEC codes such as LDPC and turbocodes
- Most common constructions encountered are called Constructions A and D (Conway and Sloane)
  - Construction A based on a single code
  - Construction D is multilevel, based on a nested sequence of codes

- Start with a q-ary linear code C with generator matrix  $\mathbf{G}_C$
- The set of vectors  $\lambda$  such that  $\lambda \mod_q$  is a codeword of  $\mathcal{C}$  form a Construction A lattice from  $\mathcal{C}$ :

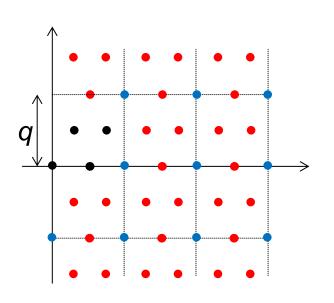
$$\Lambda = \left\{ \lambda : \lambda \, \operatorname{mod}_{q} \in \mathcal{C} \right\}$$

Alternatively we can write:

$$\Lambda = q^{\mathbb{Z}^n} + \mathcal{C}$$

The generator matrix of the lattice:

$$\mathbf{G} = \begin{bmatrix} \mathbf{G} & \mathbf{0} \\ \mathbf{G}_{C} & q \mathbf{I}_{n-k} \end{bmatrix}$$



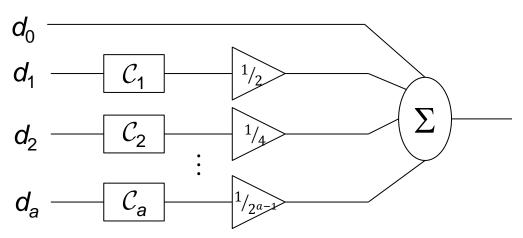
Note that minimum distance is limited by q

#### Construction D

- Let  $C_0 \subseteq C_1 \subseteq C_2 \subseteq C_a$  be a family of linear binary codes
  - where  $C_0$  is the (n, n) code and  $C_\ell$  is an  $(n, k_\ell)$  code
- Then the lattice is defined by:

$$\Lambda = \left\{ \lambda : \lambda = \mathbf{z} + \sum_{l=1}^{a} \sum_{j=1}^{k_l} d_j^l \frac{\mathbf{c}_{j,l}}{2^{l-1}} \right\}$$

■ where  $\mathbf{z} \in 2\mathbb{Z}^n$ ,  $\mathbf{c}_{j,\ell}$  is the  $j^{th}$  basis codeword of  $\mathcal{C}_{\ell}$ , and  $d_j^{\ell} \in \{0,1\}$  denotes the  $j^{th}$  data bit for the  $\ell^{th}$  code



# Low density lattice codes

- Uses the principle of LDPC codes:
  - Define generator matrix such that its inverse H = G<sup>-1</sup> is sparse
  - Then decode using sum-product algorithm (message passing) as in LDPC decoder
- However elements of H and G are reals (or complex) rather than binary
  - Messages are no longer simple log-likelihood ratios
- Ideally use nested lattice code
  - i.e. shaping region is Voronoi region of a coarse lattice

#### Gaussian and Eisenstein integers

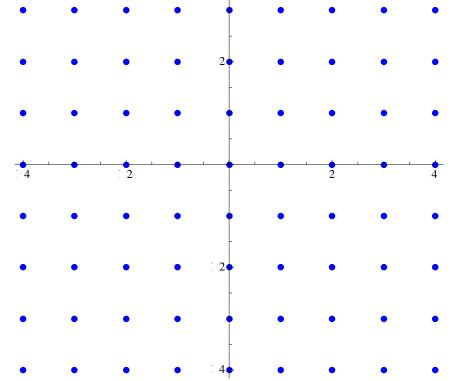
- Construction A/D and LDLC result in real lattices
  - can exploit Gaussian/Eisenstein integers to construct complex lattices
- Gaussian and Eisenstein integers form the algebraic equivalent in complex domain of the ring of integers
- Can construct complex constellations from them which form complex lattices

# Gaussian Integers

 Gaussian integers are the set of complex numbers with integer real and imaginary parts, denoted

$$\mathbb{Z}\left[\mathfrak{i}\right] = a + b\mathfrak{i}, a, b \in \mathbb{Z}$$

- They form a ring on ordinary complex arithmetic
- Hence operations in the ring exactly mirror operations in signal space
- Also form a lattice



#### Nested lattice of Gaussian integers

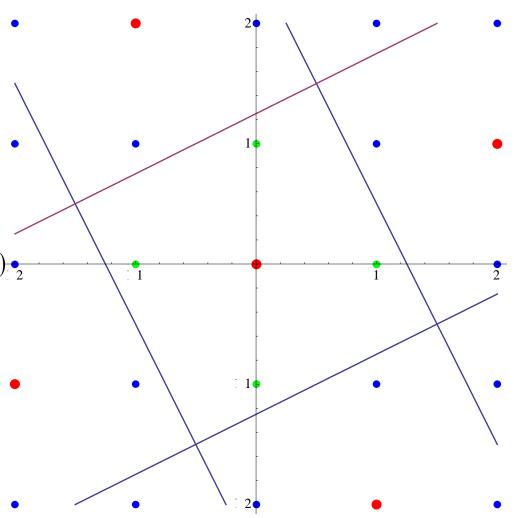
■ Consider *fine* and *coarse* lattices,  $\Lambda_f$  and  $\Lambda_c$ , both based on Gaussian integers

$$\Lambda_c \subset \Lambda_f$$

- Here we assume that each point in the coarse lattice is a point in the fine multiplied by some Gaussian integer q
  - i.e. the coarse is a scaled and rotated version of the fine
  - and the fine is just the Gaussian integers
- We then define our constellation as consisting of those Gaussian integers which fall in the Voronoi region of the coarse lattice

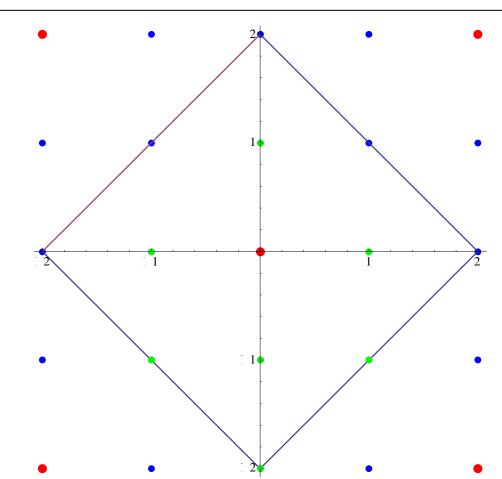
7 May, 2016

- **e.g.** q = 2 + i
- Blue points are fine lattice
- Red points are coarse lattice
- Fundamental region  $V_c(0)$  is region closer to origin than any other coarse lattice point
- Hence constellation is green points, inc origin



#### **Boundary points**

- The fundamental region is surrounded by regions corresponding to q, qi, -q and -qi
- We treat the boundaries of the latter two as belonging to the fundamental region
  - use this to allocate certain boundary points to constellation
- This also leads to an alternative definition of the fundamental region:



$$= \left\{ \begin{array}{l} \lambda \in \mathbb{C} : \left( -\frac{\left|q\right|^{2}}{2} \leq \Re\left[\lambda\right] \Re\left[q\right] + \Im\left[\lambda\right] \Im\left[q\right] < \frac{\left|q\right|^{2}}{2} \right) \\ \& \left( -\frac{\left|q\right|^{2}}{2} \leq -\Re\left[\lambda\right] \Im\left[q\right] + \Im\left[\lambda\right] \Re\left[q\right] < \frac{\left|q\right|^{2}}{2} \right) \end{array} \right\}$$

- We can establish isomorphisms between these constellations and either fields or rings
- An isomorphism is a one-to-one (or *bijective*, and hence invertible) mapping between the constellation  $\mathcal{C}$  and the ring  $\mathcal{R}$   $\lambda = \mathcal{M}(s), \lambda \in \mathcal{C}, s \in \mathcal{R}$   $s = \mathcal{M}^{-1}(\lambda), \lambda \in \mathcal{C}, s \in \mathcal{R}$
- such that the operations on the ring are equivalent to those on the constellation

$$\mathcal{M}\left(s_{1}\otimes s_{2}\right) = \mathcal{M}\left(s_{1}\right)\mathcal{M}\left(s_{2}\right) \quad \mathcal{M}\left(s_{1}\oplus s_{2}\right) = \mathcal{M}\left(s_{1}\right) + \mathcal{M}\left(s_{2}\right)$$

- It turns out that if *q* is a *Gaussian prime*, then the constellation is isomorphic to a field, otherwise it is isomorphic to a ring
- Size of field/ring is  $|q|^2$

#### Lattice construction

- This isomorphism can be used to construct a complex lattice from a code based on the field or ring
  - in a manner equivalent to Construction A

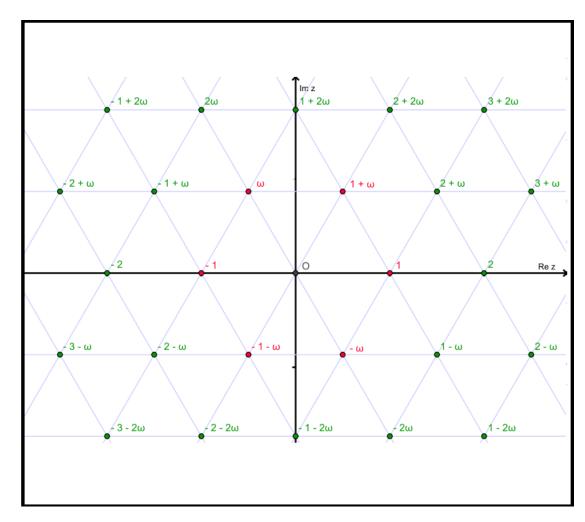
$$\Lambda = \left\{ \lambda : \lambda = \mathbf{z} + \mathcal{M} \left( \mathbf{c} \right), \mathbf{z} \in q^{\mathbb{Z}} \left[ i \right]^{n}, \mathbf{c} \in \mathcal{C} \left( \mathbb{F}_{\left| q \right|^{2}} \right) \right\}$$

- $\blacksquare$  that is, we encode a data sequence in the field  $\mathbb{F}_q$  using the code  $\mathcal{C}$  (over  $^{\mathbb{F}}_{|_q|^2}$ )
- then map the resulting symbols to the complex constellation using the mapping based on the isomorphism
- then combine with a lattice of Gaussian integers scaled by q

# Eisenstein integers

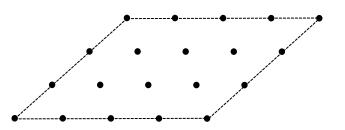
- Set of complex values with similar properties to Gaussian integers
- Hexagonal structure may result in denser lattices
- Note:

$$\omega = \frac{1 + i\sqrt{3}}{2} = e^{2\pi i/3}$$



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- Ideally the shaping region should be as close as possible to a hypersphere
  - provides shaping gain up to 1.5 dB compared to hypercube shaping
- Nested lattice shaping gives a good approximation to this
- First multiply data vector by generator matrix
  - this may generate region of lattice of arbitrary shape
- Then apply modulo-lattice operation:
  - decode to coarse lattice, and subtract resulting coarse lattice vector



- In practice this decoding operation may be difficult
  - may use hypercube shaping as simpler alternative

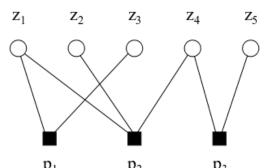
# Construction A decoding

- Generally can be carried out with decoder for underlying code C
- Applying  $mod_q$  operation regenerates codeword of C
  - then decode this codeword
  - can then recover specific point in  $\mathbb{Z}^n$
- Note that in practice we use non-binary codes (q > 2)
  - because q = 2 limits minimum distance and hence coding gain
- Typically use LDPC or turbocodes to achieve good performance
  - hence need non-binary sum-product or BCJR decoder
  - messages are probability distribution of q symbol values

# Construction D decoding

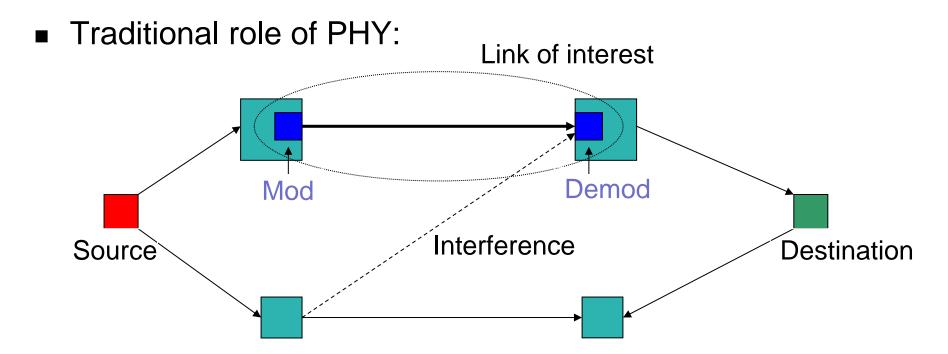
- Use multilevel decoding approach based on component codes
  - decode codes  $C_a$ ,  $C_{a-1}$ , ...  $C_1$  in succession
- Component codes may usually be binary
- May require iterative approach
  - c.f. multilevel coded modulation

- Code structure designed for sum-product decoding, cf LDPC
  - using factor graph
- However symbol values are now continuous variables (reals)
  - hence messages should be probability density functions
  - requires compact means of representing PDF in decoder
- May use Fourier or Karhunen-Loeve basis representation
  - or Gaussian mixture model



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  - Wireless physical-layer network coding
  - Compute and forward

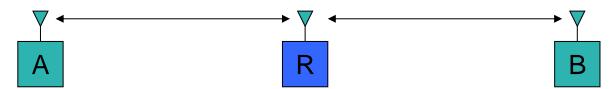
#### THE UNIVERSITY of York Physical layer in multi-user networks



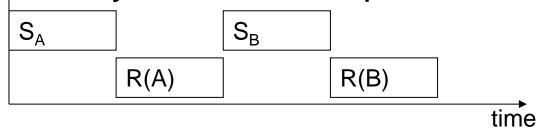
- signals from elsewhere in network treated as harmful interference
- however they may carry related information that can be exploited

#### Two-way relay channel

Two terminals want to exchange data via a relay:

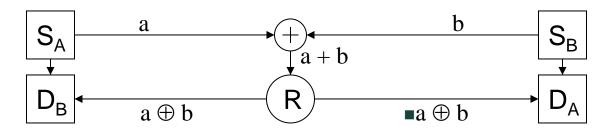


Conventionally this would require 4 time-slots:

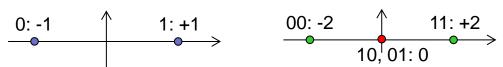


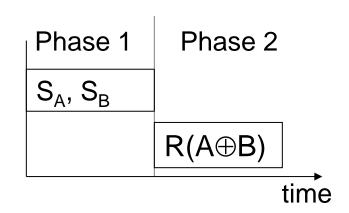
13th February 2012





- We can do better using Wireless Physical-layer Network Coding
  - using two phases
- Assume both sources transmit BPSK:
  - map data symbol '1' to signal +1;'0' to -1
- At relay, map signals +2 and -2 to '0'; 0 to '1'



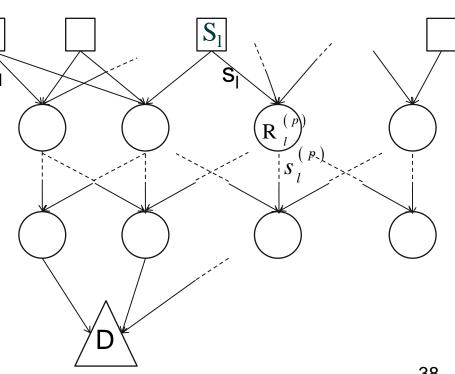


а	b	a+b	a⊕b
0	0	-2	0
0	1	0	1
1	0	0	1
1	1	+2	0



#### A general network model

- Model a network with P layers of relays
- In general all nodes in a layer transmit simultaneously
- Each relay decodes a (linear) function of symbols from previous layer  $s_{1}^{(p)} = a_{11} s_{1}^{(p-1)} + a_{21} s_{2}^{(p-1)} + \cdots + a_{11} s_{1}^{(p-1)} = \mathbf{a}_{11} \mathbf{s}_{1}^{(p-1)}$
- based on the combined signals they receive
- Destination extracts symbol it is interested in from outputs of functions
- Lattices provide useful signal sets

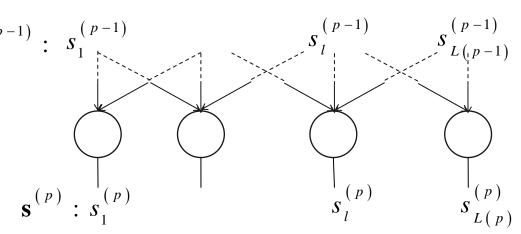




#### Network coding model of network

■ We can relate the  $s^{(p-1)}: s_1^{(p-1)}$  vector of outputs of each layer to its inputs via the matrix **A**:

$$\mathbf{s}^{(p)} = \mathbf{A}^{(p)} \mathbf{s}^{(p-1)}$$



We can combine these in cascade, so that:

$$\mathbf{s}^{(p)} = \mathbf{A}^{(p)} \mathbf{A}^{(p-1)} \cdots \mathbf{A}^{(1)} \mathbf{s}$$

We can write this as a single matrix relating the vector of symbol s<sup>D</sup> at relays connected to the destination:

$$\mathbf{s}^{D} = \mathbf{B} \mathbf{s}$$

 We assume that the destination can (in principle) decode all symbols in its connection set

#### Lattice signal sets

 $h_{B}$ 

 $h_{A}$ 

 $S_B$ 

- Consider relay receiving from two sources via channel  $h_A$ ,  $h_B$
- Sources transmit codewords  $\mathbf{c}_A$ ,  $\mathbf{c}_B$  from the same fine lattice  $\Lambda_C$
- Received signal at relay is then:

$$\mathbf{x} = h_A \mathbf{c}_A + h_B \mathbf{c}_B + \mathbf{w}$$

- Now the sum of any integer multiples of two lattice points is another lattice point
  - hence if  $h_A$ ,  $h_B$  were integers we could decode at the relay using the same lattice decoder
- Key idea is to scale received signal by scaling factor  $\alpha$  so that  $\alpha h_A$  and  $\alpha h_B$  are approximately integers

# Optimum approximation

■ Then:

$$\alpha \mathbf{x} = \alpha h_A \mathbf{c}_A + \alpha h_B \mathbf{c}_B + \alpha \mathbf{w} \approx a_A \mathbf{c}_A + a_B \mathbf{c}_B$$

- where  $a_A$  and  $a_B$  are integers
- Approximation error is:

$$(\alpha h_A - a_A) \mathbf{c}_A + (\alpha h_B - a_B) \mathbf{c}_B + \alpha \mathbf{w}$$

• We can minimise this by choosing  $\alpha$ :

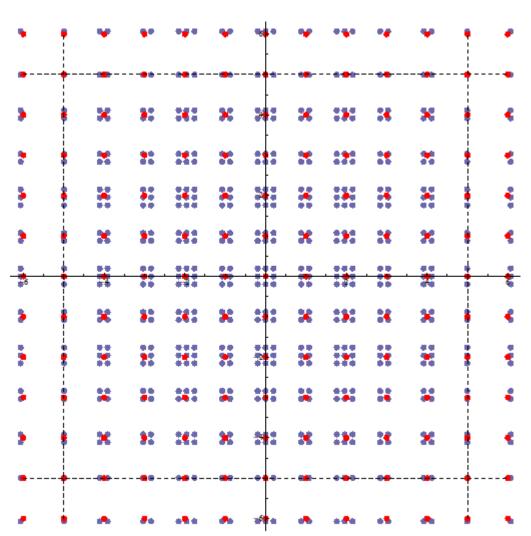
$$\alpha_{\text{MMSE}} = \frac{P \sum_{i} h_{i} a_{i}}{N + P \sum_{i} |h_{i}|^{2}}$$

- where *P* is signal power
- Also need to choose  $a_A$  and  $a_B$ 
  - could choose such that  $a_A/a_B = h_A/h_B$
  - but might require large  $\alpha$ , and hence increase noise

- $h_A = 0.55$ ;  $h_B = 1.0$
- Choose:

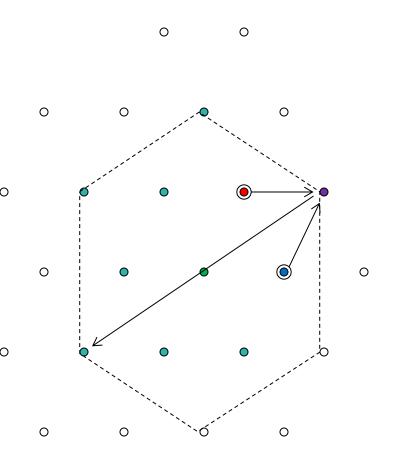
$$a_A = 1$$
;  $a_B = 2$ ;  $\alpha = 1.95$ 

- Blue points are received signal
- Red are approximated lattice



#### Modulo-Λ operation

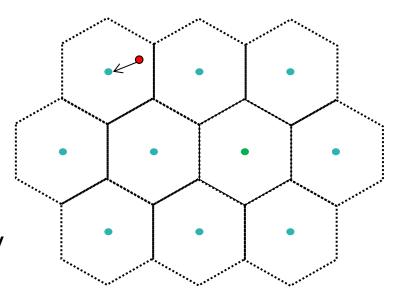
- Sum of two points from a lattice code may in general result in point outside shaping region
- Hence we apply modulo-lattice operation
  - returns a point in the original lattice code
  - so we can use the same decoder to recover sum point
- For lattice constellations isomorphic with field this operation can always be inverted

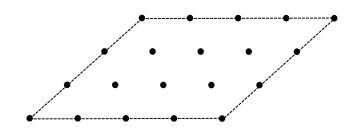


- Lattices can be extensively used in communications
  - especially for *lattice coding*
- Can be shown to achieve capacity, as lattice dimension tends to infinity
- Practical lattice constructions are based on FEC codes
  - can provide high dimension lattices
  - with practical decoding algorithms
- For wireless channels use complex lattice constellations based on Gaussian/Eisenstein integers
- Important application is compute and forward
  - applies to relay networks

#### More lattice applications

- Lattice quantisation:
  - quantising signals to lattice points in high dimension can reduce mean square error
  - Applying modulo-lattice operation also allows Wyner-Ziv compression of correlated sources
- Lattice reduction aided MIMO detection
  - MIMO channel may distort received signal:
  - LRA treats as a different lattice





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