

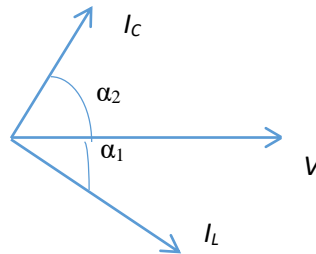
SOLUTIONS

Solution of pre-tutorial:

For the same source, the impedance of LHS figure is $Z_1 = 4 + j2\omega$,
 whereas that of RHS is either $Z_2 = j8\omega L / (8 + j\omega L)$ or $Z_2 = 8 / (j8\omega C + 1)$.
 The circuits are equivalent to each other when the source is subjected to the same impedance i.e. $Z_1 = Z_2$.
 In the case of an inductor, $4 + j2\omega = j8\omega L / (8 + j\omega L)$(1)
 Cross-multiplication of (1) yields $j8\omega L = 32 - 2\omega^2 L + j(16\omega + 4\omega L)$(2)
 Equating real and imaginary part of (2) gives $\omega = 2$ rad/s and $L = 4$ H.
 In the case of a capacitor, $4 + j2\omega = 8 / (j8\omega C + 1)$(3)
 Solution of (3) yields a negative value for C.
 Therefore, the equivalent circuit will have a sinusoidal source with $\omega = 2$ rad/s and the 8Ω resistor in parallel with the inductor of 4 H.

Solution of problem 2:

Approximate Phasor diagram can be given as follows:



$$\alpha_1 = \tan^{-1}(\omega L / R_1) \text{ and } \alpha_2 = \tan^{-1}(1 / (\omega R_2 C))$$

For the two currents to be in quadrature, $\alpha_1 + \alpha_2 = \pi/2 \Rightarrow \tan^{-1}(\omega L / R_1) + \tan^{-1}(1 / (\omega R_2 C)) = \pi/2$ (1)

$$\Rightarrow \tan[\tan^{-1}(\omega L / R_1) + \tan^{-1}(1 / (\omega R_2 C))] = \tan(\pi/2)$$

$$\Rightarrow [(\omega L / R_1) + 1 / (\omega R_2 C)] / [1 - L / (R_1 R_2 C)] = 1/0$$
(2)

Cross-multiplication and further simplification of (2) gives the condition: $R_1 R_2 = L / C$.

Solution of problem 3:

Given, $V = 240\sqrt{2} \angle 0^\circ = V_m \angle \theta$ and $Z = (75 \parallel j32\pi) = 48 + j36 = 60 \angle 36.87^\circ$ (approx.)

Then, $I = V/Z = 4\sqrt{2} \angle -36.87^\circ = I_m \angle \phi$ and $(\theta - \phi) = 36.87^\circ$.

The power factor is $\cos(\theta - \phi) = \cos(36.87^\circ) = \mathbf{0.8 \text{ lagging}}$ (as the pf angle is +ve)

$$(1) \text{ Real power} = P = (1/2) V_m I_m \cos(\theta - \phi) = (1/2) 240\sqrt{2} \times 4\sqrt{2} \times \cos(36.87^\circ) = 768 \text{ W}$$

Also, using RMS values of voltage and current, $P = V_{\text{rms}} \times I_{\text{rms}} \times \cos(36.87^\circ) = 240 \times 4 \times 0.8 = 768 \text{ W}$

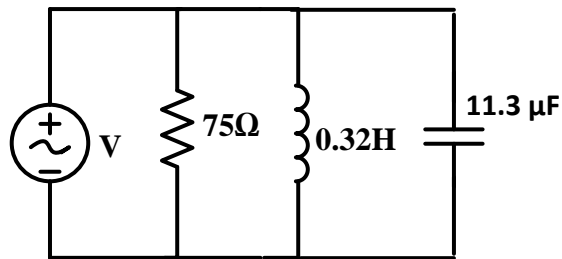
Also, $P = V_{\text{rms}}^2 / R = 240^2 / 75 = 768 \text{ W}$

$$(2) \text{ Reactive power} = Q = V_{\text{rms}}^2 / X_L = V_{\text{rms}} \times I_{\text{rms}} \times \sin(36.87^\circ) = 576 \text{ VAR}$$

(Note: $X_L = 100 \Omega$, approx)

$$\text{Also, reactive power} = P \tan(\theta - \phi) = 576 \text{ VAR}$$

In the following circuit, a capacitor of $C = 11.3 \mu\text{F}$ is connected in parallel with the source.



$$\text{Reactive power of the capacitor, } Q_C = -V_{\text{rms}}^2 / X_C = -204.04 \text{ VAR}$$

$$\text{Net reactive power experienced by the source} = 576 - 204.04 = 371.96 \text{ VAR}$$

$$\text{Net real power experienced by the source} = 768 \text{ W}$$

Using the power triangle,

$$\tan(\theta - \phi) = 371.96 / 768 \Rightarrow (\theta - \phi) = 25.842^\circ \quad (\text{pf angle is positive})$$

$$\text{Therefore, power factor} = \cos(25.842^\circ) = 0.9 \text{ lagging}$$

Solution of problem 4:

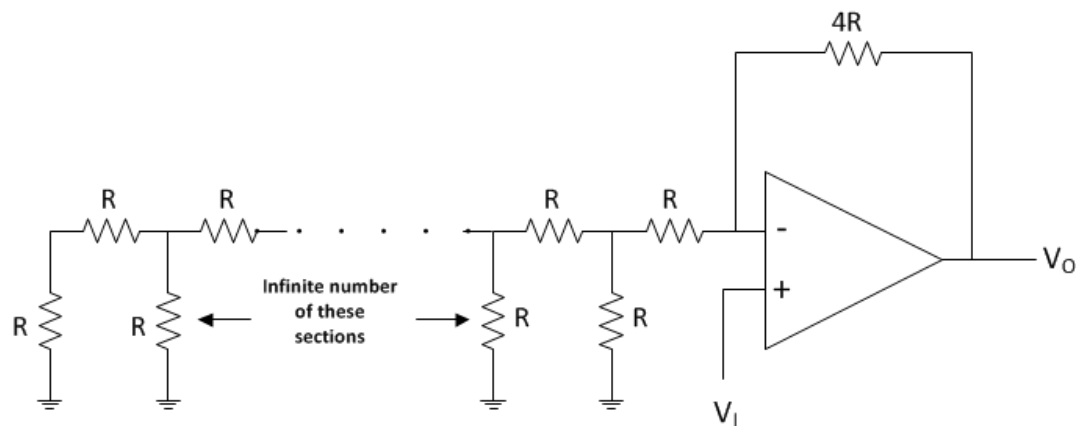
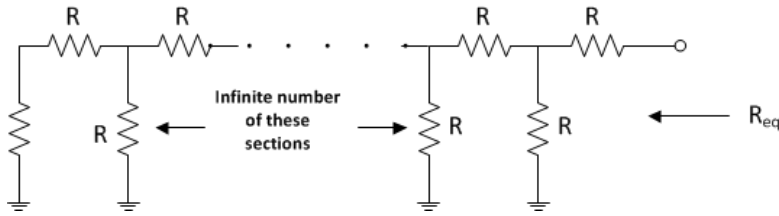
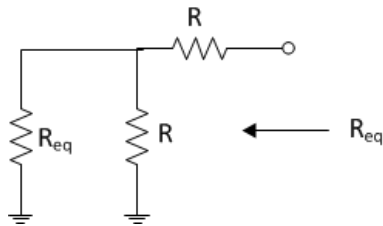


Fig. 3



or, equivalently



$$\left\{ \begin{array}{l} R_{eq} = R + \frac{RR_{eq}}{R + R_{eq}} \quad \text{or} \quad R_{eq}^2 + RR_{eq} = R^2 + RR_{eq} + RR_{eq} \\ R_{eq}^2 - RR_{eq} - R^2 = 0 \\ \text{Solving, } R_{eq} = \frac{R + \sqrt{R^2 + 4R^2}}{2} = \left(\frac{1 + \sqrt{5}}{2} \right) R = 1.618R \end{array} \right.$$

$$R_{eq} = (R_{eq} \parallel R) + R$$

$$\frac{V_I}{R_{eq}} = \frac{V_O - V_I}{4R}$$

$$\frac{V_O}{V_I} = 1 + \frac{4R}{R_{eq}} = 1 + \frac{4}{1.618} = 3.472$$

$$A_V = 3.472$$

Solution of problem 5:

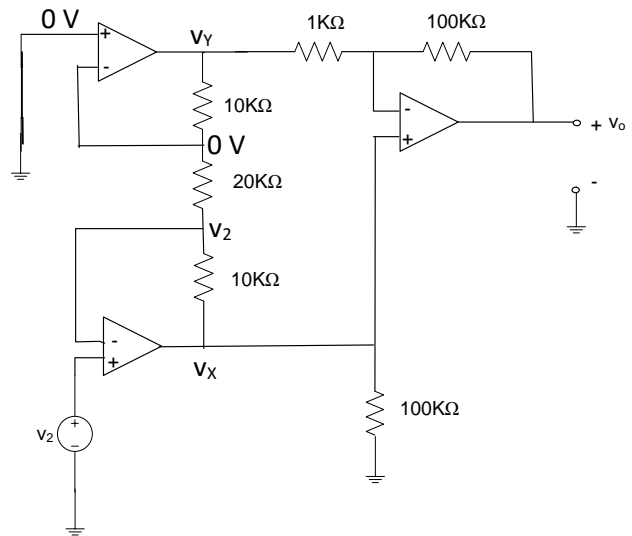
We can solve this circuit directly but it would be interesting to consider an unusual way of solving it through superposition.

Setting $v_1=0$ (i.e. grounding that source), the circuit becomes as shown.

$$\frac{v_x - v_2}{10} = \frac{v_2}{20} \Rightarrow v_x = 1.5v_2$$

$$\frac{v_y}{10} = -\frac{v_2}{20} \Rightarrow v_y = -0.5v_2$$

$$v_o \text{ (for } v_1 = 0) = -100v_y + (1+100)v_x = 201.5v_2$$



Now setting $v_2=0$ (i.e. grounding that source), the circuit becomes as shown

$$\frac{v_x}{10} = -\frac{v_1}{20} \Rightarrow v_x = -0.5v_1$$

$$\frac{v_y - v_1}{10} = \frac{v_1}{20} \Rightarrow v_y = 1.5v_1$$

$$v_o = -100v_y + (1+100)v_x = -200.5v_1$$

Combining the two, we get –

$$\mathbf{v_o = -200.5v_1 + 201.5v_2}$$

