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Digital Signal Processing

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CONTENTS

Software Installation

-	Software instanation	1
2	Digital Filter	1
3	Difference Equation	1
4	Z-transform	2
5	Impulse Response	4
6	DFT and FFT	6

Abstract—This manual provides a simple introduction to digital signal processing.

1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

wget https://raw.githubusercontent.com/ gadepall/ EE1310/master/filter/codes/Sound Noise.wav

- 2.2 You will find a spectrogram at https: //academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find? **Solution:** There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the
 - **Solution:** There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.
- 2.3 Write the python code for removal of out of band noise and execute the code.

Solution: We can get code from below link-

https://github.com/PradeepMundlik/ EE3900/tree/master/Assignment1/ codes/q2/Cancel-noise.py

2.4 The of output the python script 2.3 Problem is the audio file Sound With ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 Difference Equation

3.1 Let

$$x(n) = \left\{ \frac{1}{1}, 2, 3, 4, 2, 1 \right\}$$
 (3.1)

Sketch x(n).

Solution:

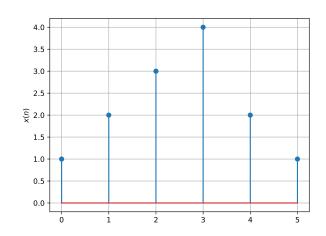


Fig. 3.1

https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q3/xn.py 3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The c code for yn:

https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q3/yn.c

Data file for Y(n):

https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q3/yn.dat

The following code yields Fig. 3.2.

https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q3/yn.py

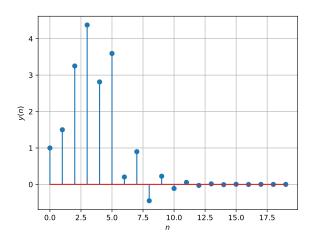


Fig. 3.2

4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

Solution: From (4.1),

$$Z\{x(n-1)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)
$$(4.5)$$

$$= z^{-1}X(z) \tag{4.6}$$

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\lbrace x(n-k)\rbrace = \sum_{n=-\infty}^{\infty} x(n-k)z^{-n}$$
 (4.7)

$$=\sum_{n=-\infty}^{\infty}x(n)z^{-n-k} \tag{4.8}$$

$$= z^{-k} \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$
 (4.9)

$$= z^{-k}X(z) \tag{4.10}$$

4.2 Obtain X(z) for x(n) defined in problem 3.1 **Solution:** from 3.1

$$x(n) = \{1, 2, 3, 4, 2, 1\} \tag{4.11}$$

$$X(z) = \mathbb{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.12)

$$X(z) = 1z^{0} + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + 1z^{-5}$$
(4.13)

$$X(z) = 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + 1z^{-5}$$
(4.14)

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.15}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.7) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.16)

$$\implies \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \tag{4.17}$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.18)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.19)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.20}$$

Solution:

$$\mathcal{Z}\{\delta(n)\} = \sum_{n=-\infty}^{\infty} \delta(n) z^{-n}$$
 (4.21)

$$= 1 \tag{4.22}$$

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.23}$$

and from (4.19),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.24)

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.25}$$

using the fomula for the sum of an infinite geometric progression.

4.5 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a| \tag{4.26}$$

Solution:

$$\mathcal{Z}\lbrace a^n u(n)\rbrace = \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n}$$
 (4.27)

$$=\sum_{n=0}^{\infty} a^n z^{-n}$$
 (4.28)

$$=\sum_{n=0}^{\infty} \left(az^{-1}\right)^n \tag{4.29}$$

$$= \frac{1}{1 - az^{-1}}, \quad \left| az^{-1} \right| < 1 \quad (4.30)$$

using the fomula for the sum of an infinite geometric progression.

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.31)

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of x(n).

Solution:

$$H(e^{j\omega}) = \frac{1 + e^{-2j\omega}}{1 + \frac{1}{2}e^{-j\omega}}$$
(4.32)

$$\implies |H(e^{j\omega})| = \frac{\left|1 + \cos 2\omega - j\sin 2\omega\right|}{\left|1 + \frac{1}{2}\cos \omega - \frac{1}{2}\sin \omega\right|}$$

$$(4.33)$$

$$= \sqrt{\frac{(1+\cos 2\omega)^2 + (\sin 2\omega)^2}{(1+\frac{1}{2}\cos \omega)^2 + (\frac{1}{2}\sin \omega)^2}}$$
(4.34)

$$=\sqrt{\frac{2+2\cos 2\omega}{\frac{5}{4}+\cos \omega}}\tag{4.35}$$

$$= \sqrt{\frac{2(2\cos^2\omega)4}{5 + 4\cos\omega}}$$
 (4.36)

$$|H(e^{j\omega})| = \frac{4|\cos\omega|}{\sqrt{5 + 4\cos\omega}} \tag{4.37}$$

So,

$$\frac{4\left|\cos(\omega+2\pi)\right|}{\sqrt{5+4\cos(\omega+2\pi)}} = \frac{4\left|\cos\omega\right|}{\sqrt{5+4\cos\omega}} \quad (4.38)$$

It is clear that $|H(e^{j\omega})|$ is periodic with period 2π .

The following code plots Fig. 4.6.

https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q4/dtft.py

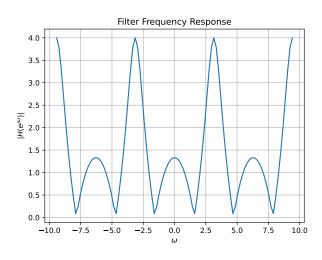


Fig. 4.6: $|H(e^{j\omega})|$

5 Impulse Response

5.1 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z) \tag{5.1}$$

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse response* of the system defined by (3.2).

Solution: From (4.17),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.2)

$$\implies h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$
(5.3)

using (4.26) and (4.7).

5.2 Sketch h(n). Is it bounded? Convergent? **Solution:** The following code plots Fig. 5.2.

https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q5/hn.py

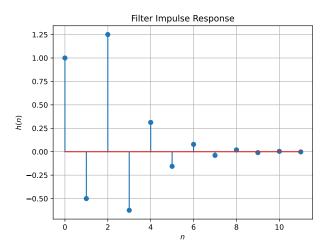


Fig. 5.2: h(n) as the inverse of H(z)

Fron fig.5.2 it is clear that, h(n) converges to 0 and it is bounded as well.

5.3 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.4}$$

Is the system defined by (3.2) stable for the impulse response in (5.1)?

Solution:

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2} \right)^n u(n) + \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2} \right)^{n-2} u(n-2) \quad (5.5)$$

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=0}^{\infty} \left(-\frac{1}{2} \right)^n + \sum_{n=2}^{\infty} \left(-\frac{1}{2} \right)^{n-2}$$
 (5.6)

These are both sums of infinite geometric progressions with first terms 1 and common ratios $-\frac{1}{2}$

$$\sum_{n=-\infty}^{\infty} h(n) = \frac{1}{1 - \left(-\frac{1}{2}\right)} + \frac{1}{1 - \left(-\frac{1}{2}\right)}$$
 (5.7)

$$=\frac{4}{3}<\infty\tag{5.8}$$

Therefore, the system is stable.

5.4 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.9)$$

This is the definition of h(n).

Solution:

$$h(0) = 1 \tag{5.10}$$

Now, for n = 1,

$$h(1) + \frac{1}{2}h(0) = \delta(1) + \delta(-1) = 0$$
 (5.11)

$$\implies h(1) = -\frac{1}{2}h(0) = -\frac{1}{2} \tag{5.12}$$

For n = 2.

$$h(2) + \frac{1}{2}h(1) = \delta(2) + \delta(0) = 1$$
 (5.13)

$$\implies h(2) = 1 - \frac{1}{2}h(1) = \frac{3}{2} \tag{5.14}$$

For n > 2, the right hand side of the equation

is always zero. Thus,

$$h(n) = -\frac{1}{2}h(n-1) \qquad n > 2$$
 (5.15)

$$h(3) = \frac{3}{2} \left(-\frac{1}{2} \right) \tag{5.16}$$

$$h(4) = \frac{3}{2} \left(-\frac{1}{2} \right)^2 \tag{5.17}$$

$$\vdots (5.18)$$

$$h(n) = \frac{3}{2} \left(-\frac{1}{2} \right)^{n-2} \tag{5.19}$$

Therefore.

$$h(n) = \begin{cases} 1 & n = 0\\ -\frac{1}{2} & n = 1\\ \frac{3}{2} \left(-\frac{1}{2}\right)^{n-2} & n \ge 2 \end{cases}$$
 (5.20)

Thus, it is bounded and convergent to 0

$$\lim_{n \to \infty} h(n) = 0 \tag{5.21}$$

The following code plots Fig. 5.4. Note that this is the same as Fig. 5.2.

> https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q5/hndef.py

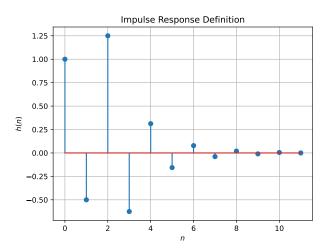


Fig. 5.4: h(n) from the definition

5.5 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.22)

Comment. The operation in (5.22) is known as convolution.

Solution:

$$x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.23)

$$= \sum_{k=0}^{5} x(k)h(n-k)$$
 (5.24)

The following code plots Fig. 5.5. Note that this is the same as y(n) in Fig. 3.2.

> https://github.com/PradeepMundlik/ EE3900/blob/master/Assignment1/ codes/q5/ynconv.py

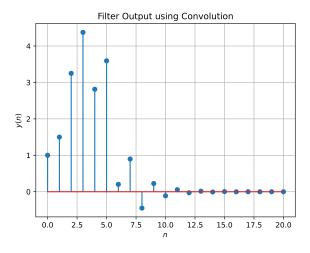


Fig. 5.5: y(n) from the definition of convolution

5.6 Show that

$$y(n) = \sum_{n=-\infty}^{\infty} x(n-k)h(k)$$
 (5.25)

Solution:

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (5.22)$$

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (5.26)$$

Substitute k = n - i

$$\sum_{k=-\infty}^{\infty} x(k)h(n-k) = \sum_{n-i=-\infty}^{\infty} x(n-i)h(n-(n-i))$$
(5.27)

$$=\sum_{i=-\infty}^{\infty}x(n-i)h(i)$$
 (5.28)

$$=\sum_{i=-\infty}^{\infty}x(n-i)h(i) \qquad (5.29)$$

since the order of limits does not matter for a summation. Thus,

$$\sum_{k=-\infty}^{\infty} x(k)h(n-k) = \sum_{k=-\infty}^{\infty} x(n-k)h(k) \quad (5.30)$$

$$\implies x(n) * h(n) = h(n) * x(n)$$
 (5.31)

Therefore, convolution is commutative.

6 DFT AND FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

Solution: The following code plots Fig.6.1

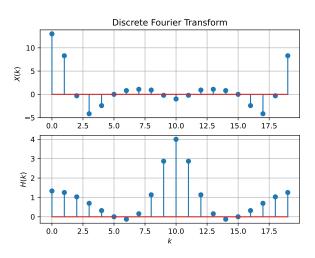


Fig. 6.1: Discret Fourier Transform

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

Solution: The following code plots Fig.6.2

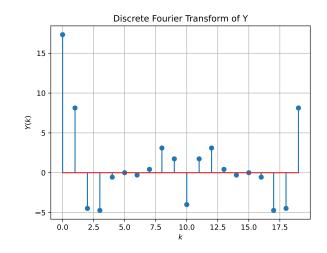


Fig. 6.2: Discret Fourier Transform of Y(k)

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

Solution: The following code plots Fig. 5.5. Note that this is the same as y(n) in Fig. 3.2.

wget https://raw.githubusercontent.com/ gadepall/EE1310/master/**filter**/codes/yndft. py

Fig. 6.3: y(n) from the DFT

- 6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT.
- 6.5 Wherever possible, express all the above equations as matrix equations.