

AC to DC Converter



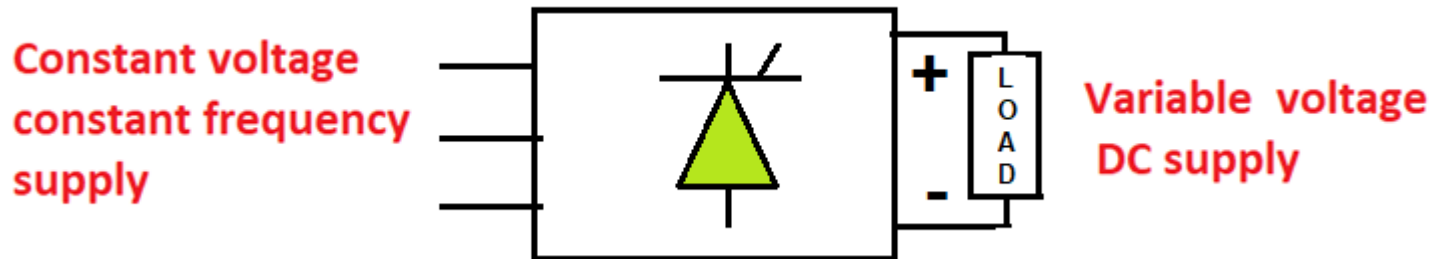
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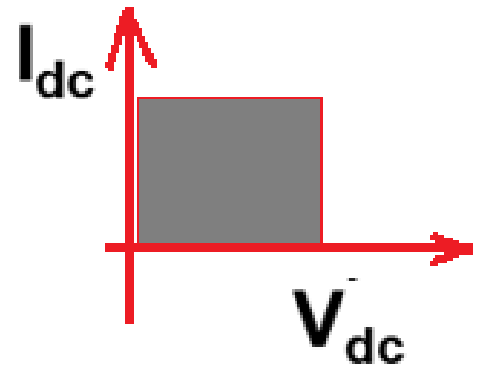
- Introduction
- Classification of rectifiers
- Single phase rectifiers
- Three phase rectifiers
- Input side performance
- Output side performance

Rectifiers

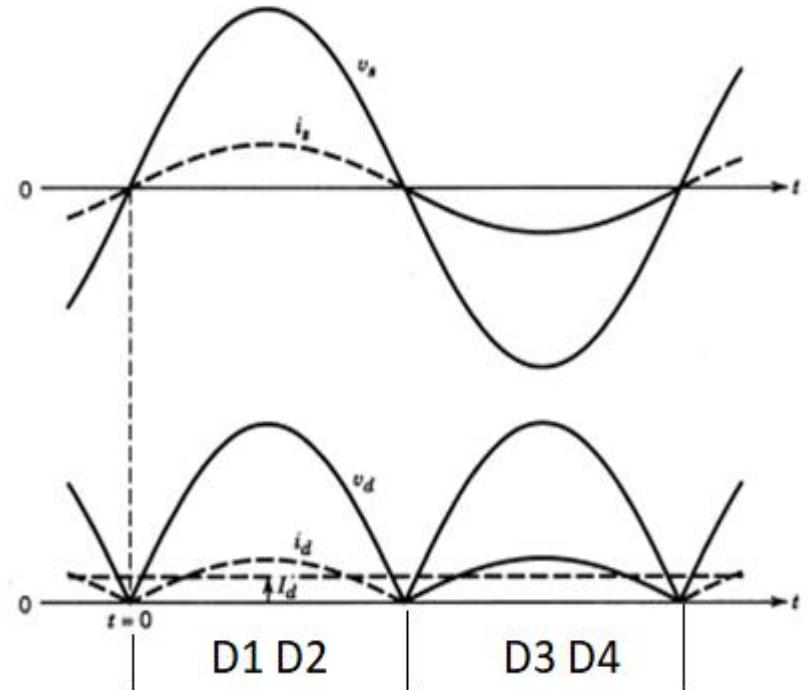
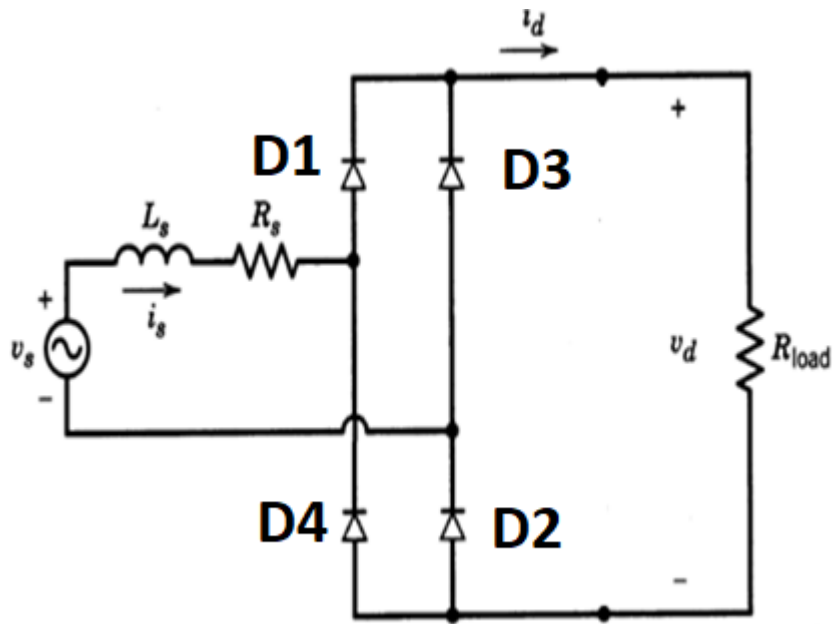
- Block diagram



- Classification
- No of phases
- Half wave/ full wave rectification
- Pulse number
- Quadrant operation



1 ϕ Diode bridge with R load



Each diode conducts for 180°

Average DC O/P voltage

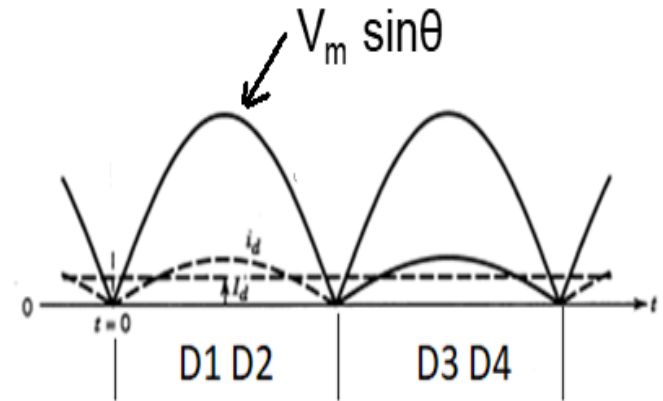
$$V_{dc} = \frac{2}{2\pi} \int_0^\pi V_m \sin\theta d\theta$$

1 ϕ Diode bridge with R load

- $V_{dc} = V_m/\pi (-\cos\theta)|_0^\pi$
- $V_{dc} = 2 V_m/\pi$
- Average DC Voltage = $2 V_m/\pi$
- When D1 D2 conducts, supply current is +ve
- And waveshape is same as that of I_{dc}
- When D3 D4 conducts, supply current is -ve
- And waveshape is same as that of I_{dc}
- Source current is sinusoidal and in phase with voltage

V_{dc} (RMS) Value

$$\begin{aligned}
 V_{dc}(rms) &= \left[\frac{2}{2\pi} \int_0^\pi V_m^2 \sin^2 \theta \cdot d\theta \right]^{1/2} \\
 &= \left[\frac{V_m^2}{\pi} \int_0^\pi \sin^2 \theta \cdot d\theta \right]^{1/2} \\
 &= \left[\frac{V_m^2}{\pi} \int_0^\pi \left(\frac{1 - \cos 2\theta}{2} \right) \cdot d\theta \right]^{1/2} \\
 &= \left[\frac{V_m^2}{2\pi} \int_0^\pi (1 - \cos 2\theta) \cdot d\theta \right]^{1/2} \\
 &= \left[\frac{V_m^2}{2\pi} \left\{ \int_0^\pi 1 \cdot d\theta - \int_0^\pi \cos 2\theta \cdot d\theta \right\} \right]^{1/2}
 \end{aligned}$$

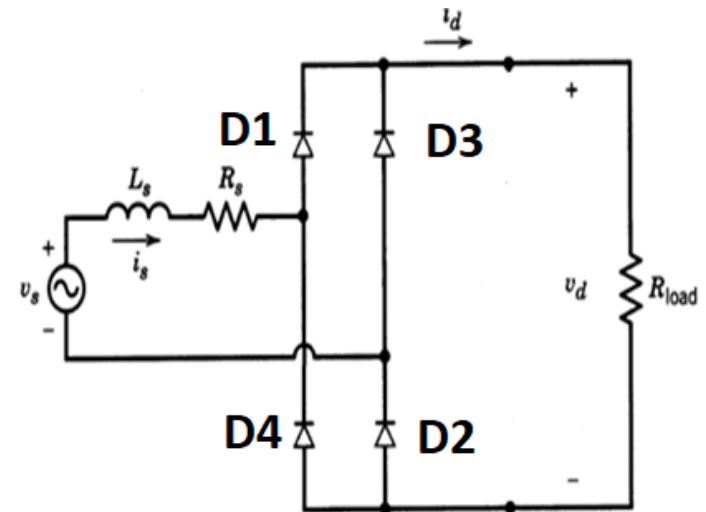


V_{dc} (RMS) Value

$$\begin{aligned} V_{dc}(\text{rms}) &= \left[\frac{V_m^2}{2\pi} \left\{ \left. 0 \right|_0^\pi - \frac{\sin 2\theta}{2} \right|_0^\pi \right\} \right]^{1/2} \\ &= \left[\frac{V_m^2}{2\pi} \{ (\pi) - 0 \} \right]^{1/2} \\ &= \left[\frac{V_m^2}{2\pi} \cdot \pi \right]^{1/2} = \left[\frac{V_m^2}{2} \right]^{1/2} \\ &= \frac{V_m}{\sqrt{2}} \end{aligned}$$

1 ϕ Diode bridge with R load

- Power dissipated in R load
- $P = V_{dc}(rms) \times I_{dc}(rms)$
- Input voltage 230 V (rms)
- $R = 23 \text{ ohm}$
- $I_{dc}(rms) = 10A$
- Power in R = 2300 W

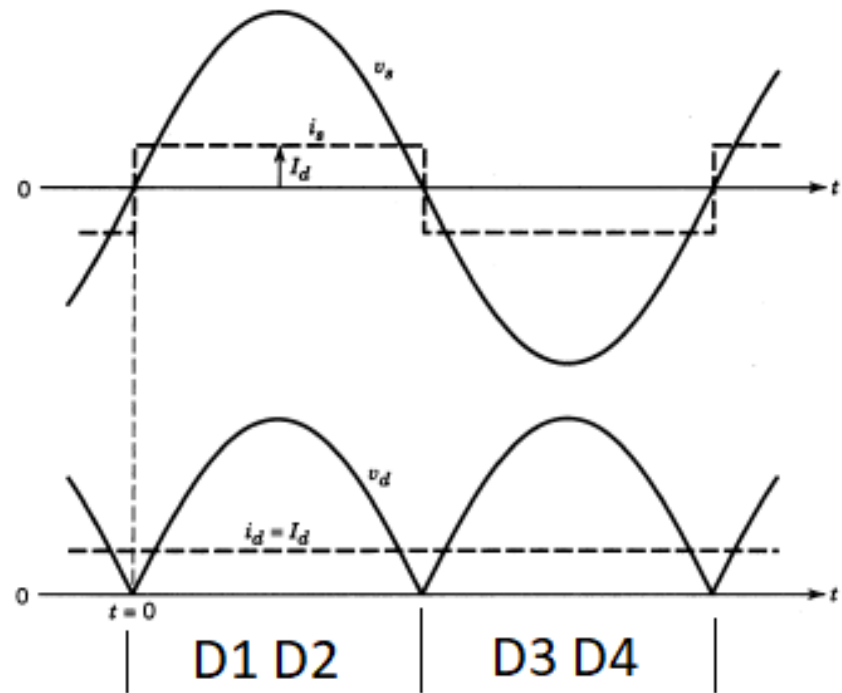
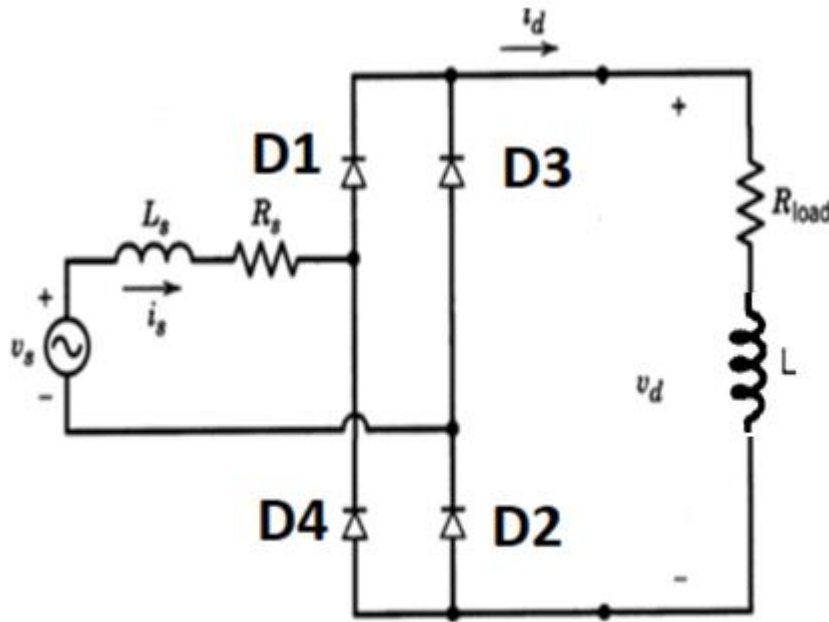


1 ϕ Diode bridge with R load

- Procedure to determine the ripple content
- Determine the average and RMS value of the waveform
- Form factor (FF) = RMS Value/ Average Value
- % ripple = $\sqrt{FF^2 - 1}$
- Diode bridge with R load % ripple in o/p voltage
- RMS value = $V_m/\sqrt{2}$
- Average value = $2 V_m/\pi$
- FF factor = $\pi/2\sqrt{2} = 1.11$
- Ripple factor = 48.2%

Single phase full wave rectifier

■ RL load

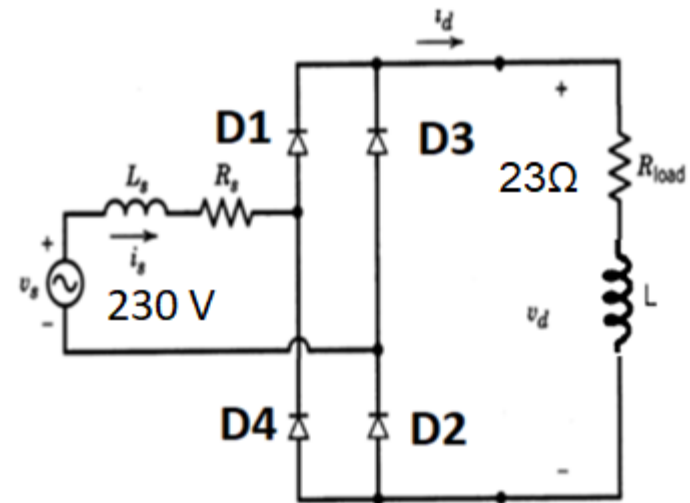


Single phase full wave rectifier: RL load

- O/P voltage waveform is same as resistive load
- DC current (I_d) is smooth and ripple free due to inductance .
- When D1 D2 conducts, supply current is +ve
And waveshape is same as that of I_{dc}
- When D3 D4 conducts, supply current is -ve
And waveshape is same as that of I_{dc}
- The source current waveform is square wave

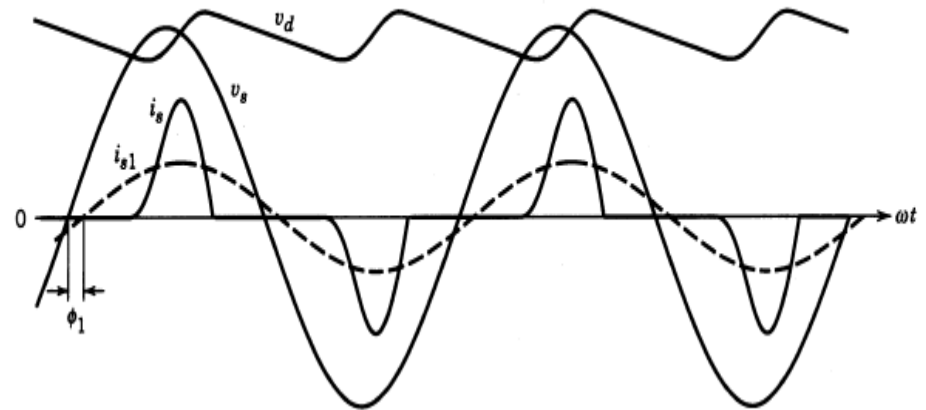
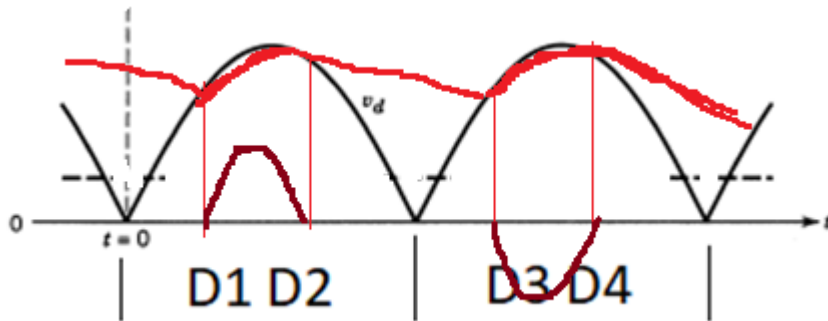
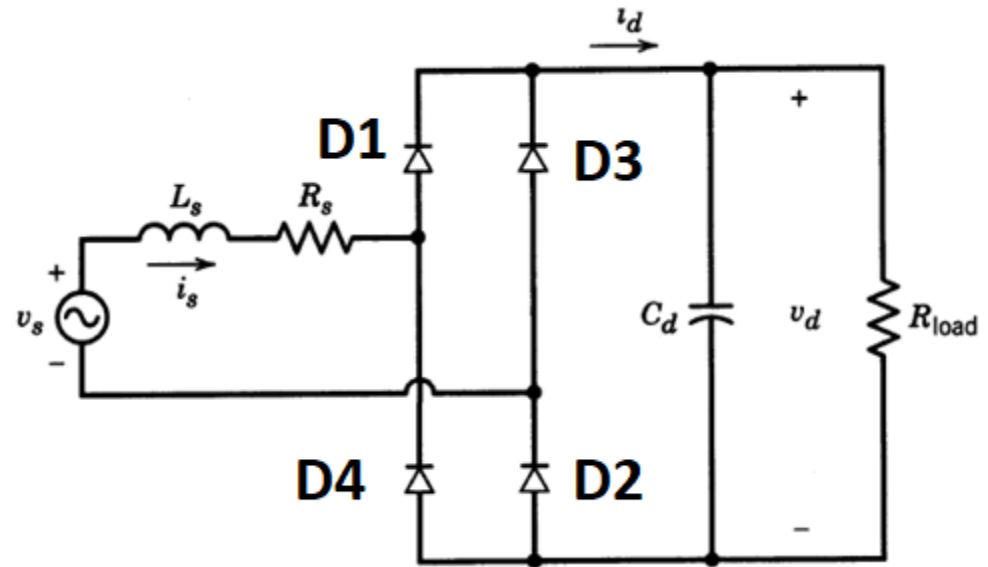
Single phase full wave rectifier: RL load

- Power dissipated in $R=23\Omega$ & $V_s=230V$
- I_{dc} is smooth & ripple free
- Voltage across R = smooth and ripple free and magnitude $\Rightarrow V_{dc}(avg)$
- $P = V_{dc}(avg) \times I_{dc}(avg)$
- $V_{dc}(avg) = 207.07 V$
- $I_{dc}(avg) = 9.0 A$
- $P = 1863.63 W$



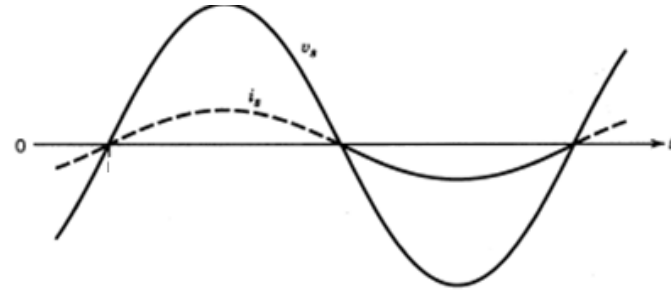
1 ϕ Diode bridge with R C load

- Diode conduction period
- Source current

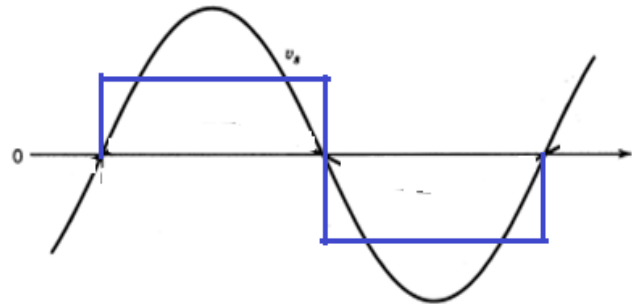


1 ϕ Diode bridge: Source current Waveform

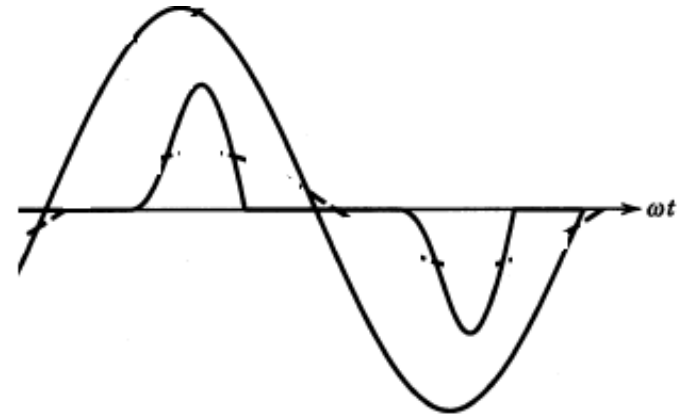
- R load



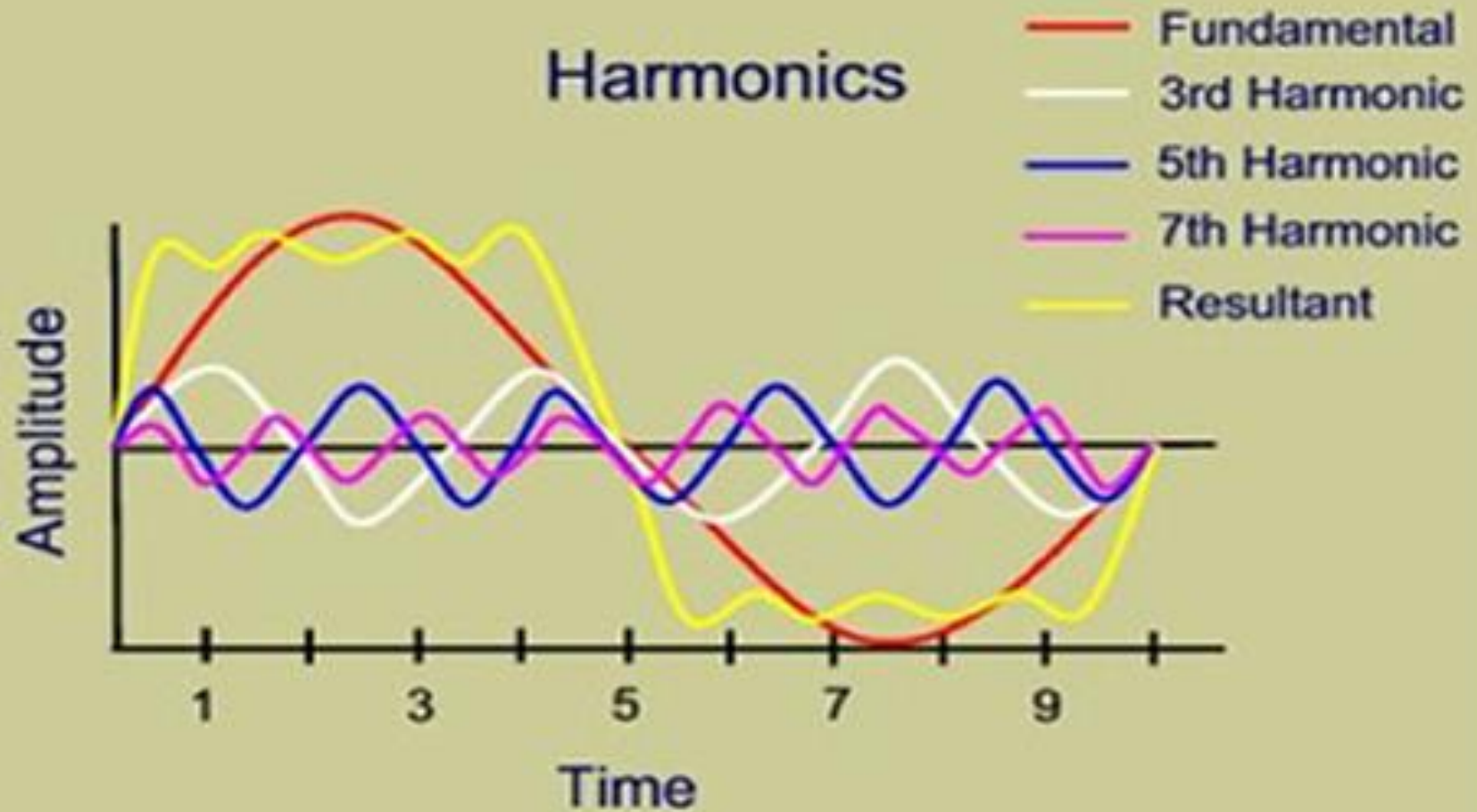
- RL Load



- RC parallel load



Harmonics



Harmonics

- **Harmonics** are voltages or currents that operate at a frequency that is an integer (whole-number) multiple of the fundamental frequency.
- For 50Hz fundamental waveform, this means a 2nd **harmonic** frequency would be 100Hz ($2 \times 50\text{Hz}$), a 3rd **harmonic** would be 150Hz ($3 \times 50\text{Hz}$), a 5th at 250Hz, a 7th at 350Hz and so on.

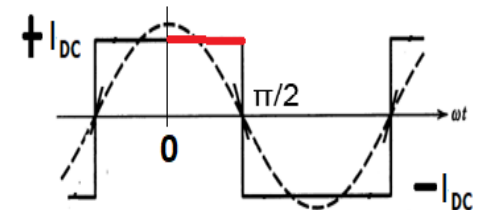
Harmonics

- The relationship between rms current, fundamental current and harmonic current
- $I_{rms}^2 = I_{1rms}^2 + I_{2rms}^2 + I_{3rms}^2 + I_{4rms}^2 + I_{5rms}^2 + \dots$
- $I_{rms}^2 = I_{1rms}^2 + I_{hrms}^2$
- Where,
- Total harmonic current can be
- $I_{hrms}^2 = I_{2rms}^2 + I_{3rms}^2 + I_{4rms}^2 + I_{5rms}^2 + \dots$
- For the given waveform I_{rms}^2 and I_{1rms}^2 is computed
- $\%THD = (I_{hrms} / I_{1rms}) \times 100$

Harmonic spectrum of a waveform

■ Procedure

- Determine the RMS value of the waveform. = I
- Determine the peak amplitude of nth harmonic
- Considering quarter wave symmetry
- $A_n = \frac{8}{2\pi} \int_0^{\pi/2} F(\theta) \cos(n\theta) d\theta$
- Where A_n = peak amplitude of nth Harmonic.
- Determine peak amplitude of fundamental and the RMS value of fundamental = I_1
- $I^2 = I_1^2 + I_h^2$
- $\%THD = (I_h/I_1) \times 100$



Harmonic spectrum of square wave

- Harmonic spectrum

- RMS value $I = I_{dc}$

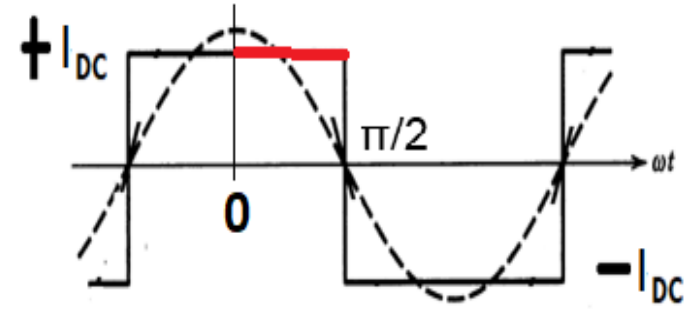
- $A_n = \frac{8}{2\pi} \int_0^{\pi/2} I_{dc} \cos(n\theta) d\theta$

- $A_n = (4/n\pi) I_{dc} \sin(n\pi/2)$

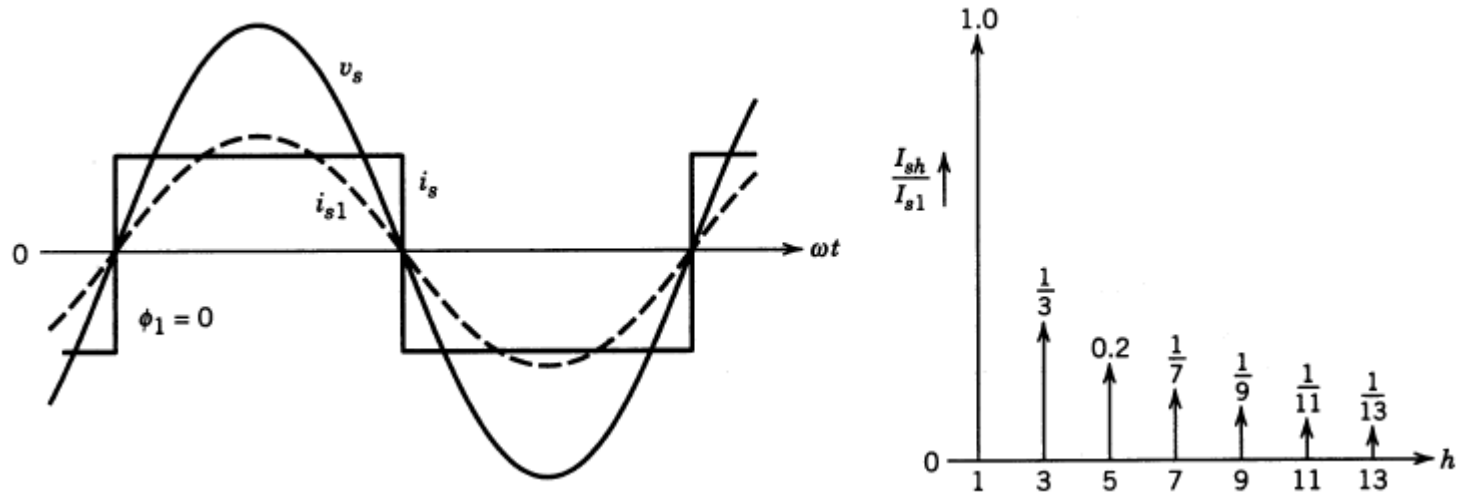
- $A_1 = 4I_{dc}/\pi$ (Peak amplitude of fundamental)

- $I_1 = \frac{2\sqrt{2}}{\pi} I_{dc}$ and $I_h = I_{dc} \sqrt{1 - \frac{8}{\pi^2}}$

- $I_h/I_1 = 0.482$



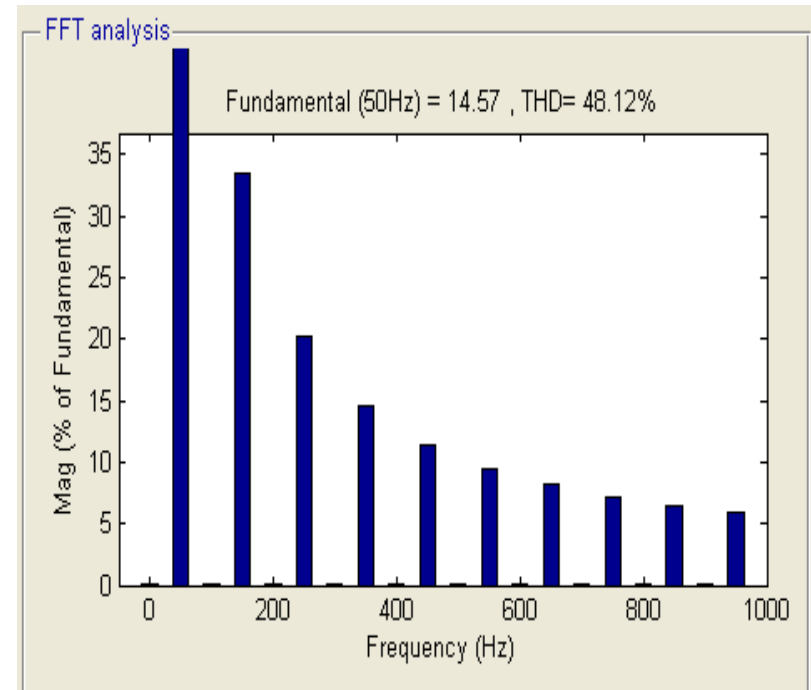
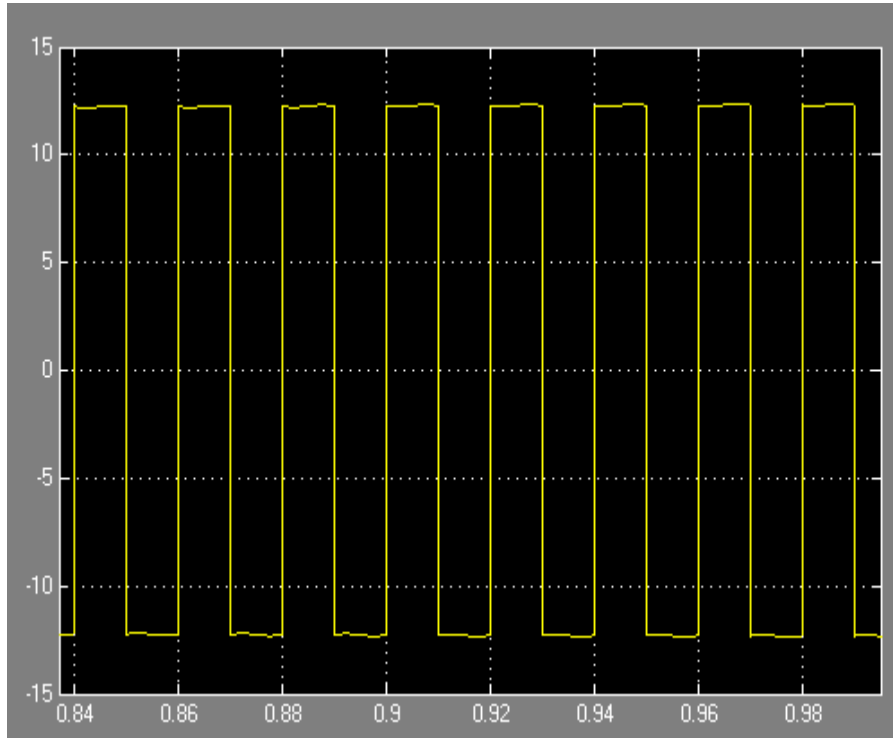
Harmonic spectrum of 1 ϕ diode bridge -RL load



Peak value of harmonic component

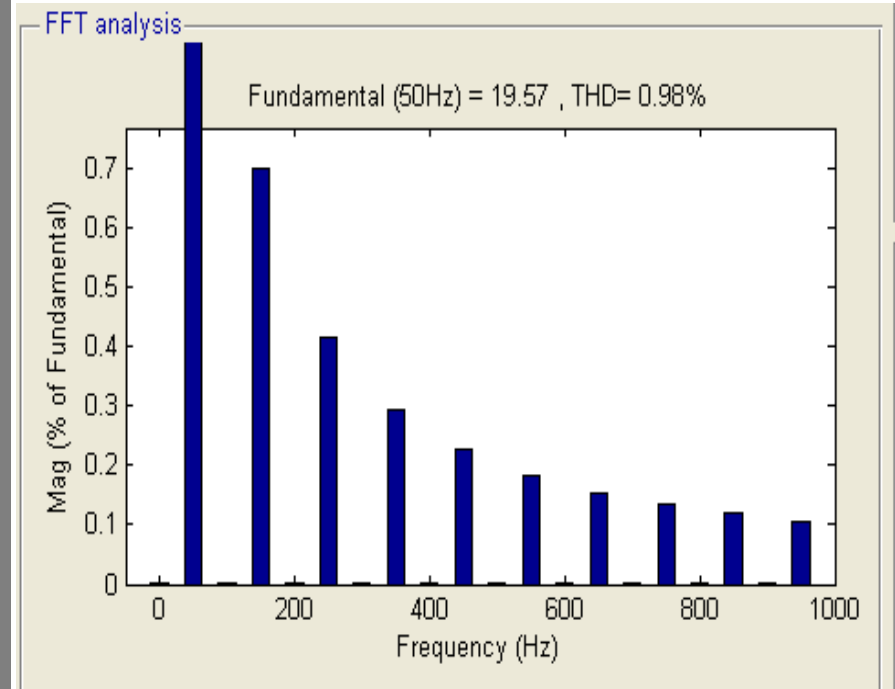
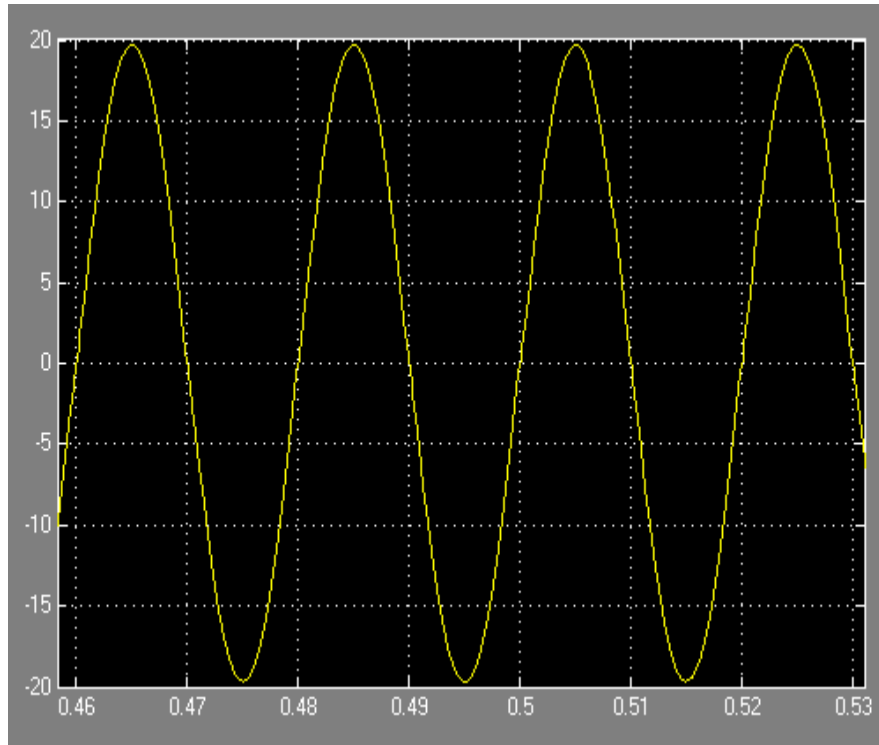
$$A_n = (4/n\pi) I_{dc} \sin(n\pi/2)$$

1 ϕ Diode bridge with RL load



Harmonic spectrum depends upon type of load on the converter
THD in source current is 48.12%

1 ϕ Diode bridge with R load



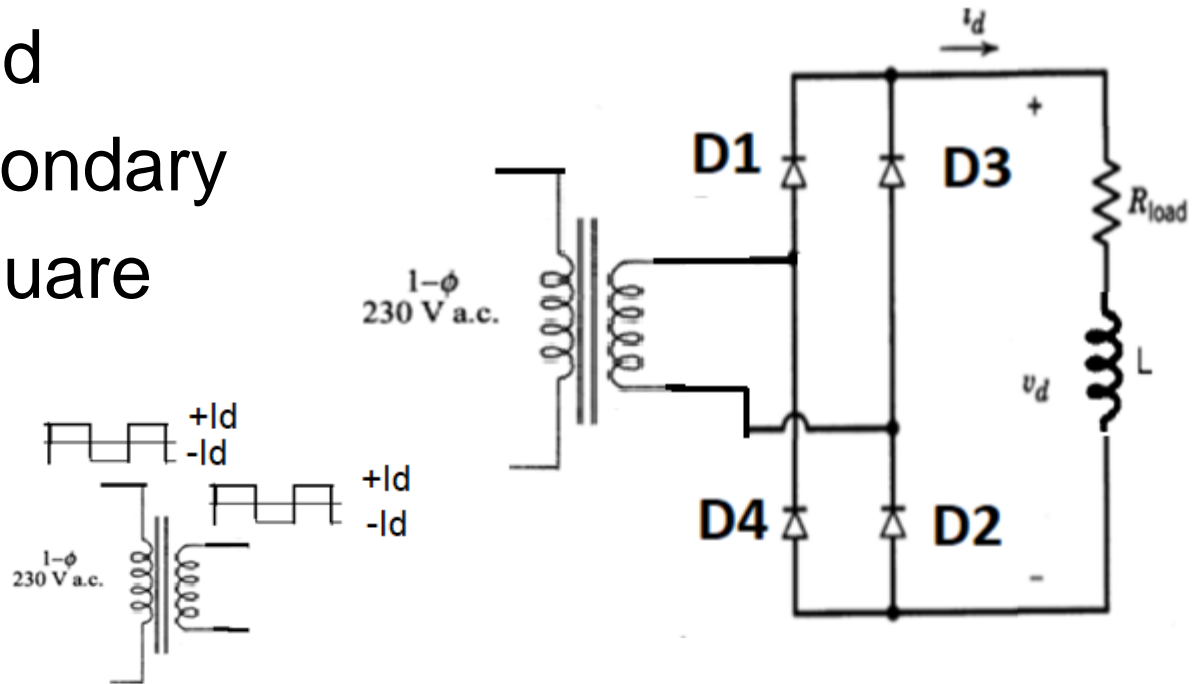
$$I = \sqrt{\frac{1}{T} \int_0^T i^2(t) dt} = \sqrt{\sum_{n=1}^{\infty} I_n^2}$$

$$I^2 = I_1^2 + I_h^2$$

$$\% \text{ Current THD} = (I_h / I_1) \times 100$$

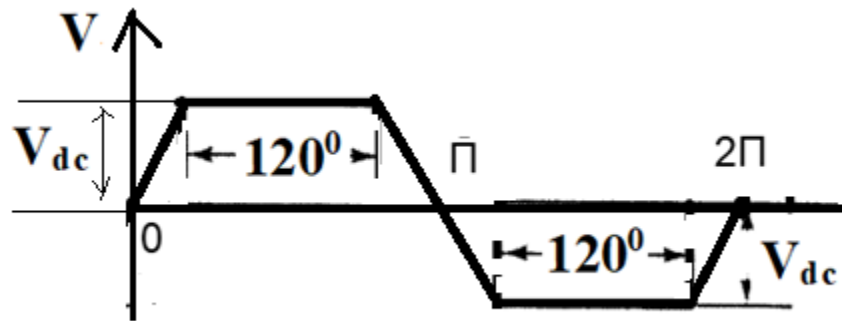
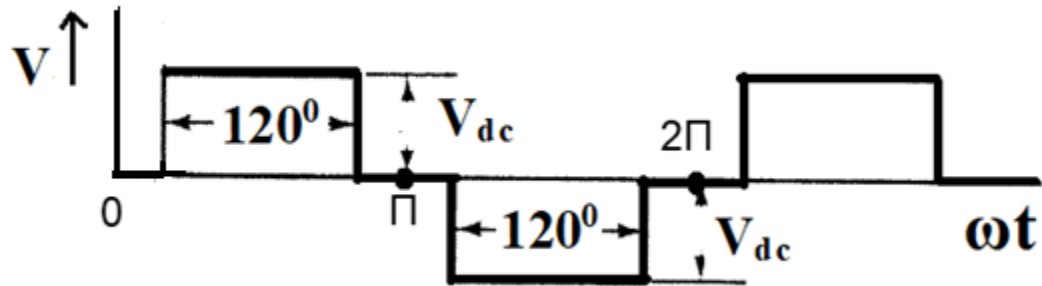
1 ϕ Diode bridge : Tx rating

- Why transformer between supply & rectifier
- Adjustment of O/P DC voltage.
- Isolation between AC & DC system
- Load = RL load
- Primary & secondary currents are square wave



Exercise : Harmonic Spectrum

- Determine the THD of the following wave forms



1 ϕ Diode bridge : Tx rating

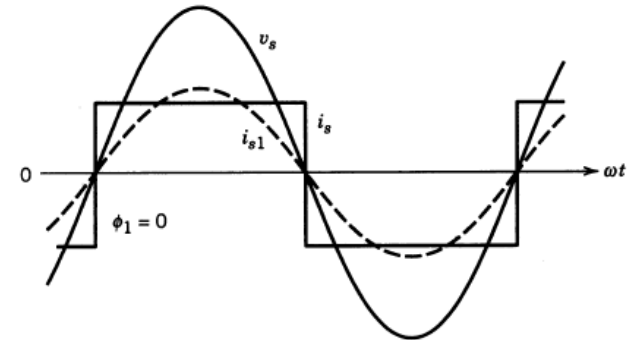
- tx turn ratio =1
- Dc o/p power = $V_{dc} \times I_{dc}$
- Transformer current $I_{rms} = I_{dc}$
- Transformer voltage = $V_m/\sqrt{2}$
- Transformer VA rating = $I_{dc} \times V_m/\sqrt{2}$
- $V_{dc} = 2 V_m/\pi \Rightarrow V_m = (V_{dc} \times \pi)/2$
- VA rating of tx = $I_{dc} \times (V_{dc} \times \pi)/ 2\sqrt{2}$
- $1.11 V_{dc} \times I_{dc}$

1 ϕ Diode bridge : Tx rating

- Transformer utilization factor = DC Power/VA rating
- $TUF = V_{dc} \times I_{dc} / (1.11 V_{dc} \times I_{dc})$
- $TUF = 0.9$
- Transformer utilization factor = 0.9
- For 1000 W DC power the transformer VA = 1100VA

1 ϕ Diode bridge : supply power factor

- Consider diode bridge with RL Load
- When current or voltage is non sinusoidal
- Displacement angle (θ)
- Is angle between fundamental component voltage & current
- diode bridge \Rightarrow displacement angle $\Rightarrow 0^\circ$
- Displacement factor = $\cos\theta$
- displacement factor = 1



1 ϕ Diode bridge : supply power factor

- When current or voltage is non sinusoidal

$$\text{Power factor} = \frac{\text{mean i/p power}}{\text{rms i/p VA}}$$

- Only fundamental component will contribute to mean power
- Mean power = $V_1 I_1 \cos \theta$
- Rms i/p VA = $V_{\text{rms}} I_{\text{rms}}$ $V_{\text{rms}} = V_1$

$$\text{Power factor} = \frac{I_1 \times \cos \theta}{I_{\text{rms}}}$$

1 ϕ Diode bridge : supply power factor

- Distortion factor (DF) = $\frac{I_1}{I_{rms}}$
- $I_1 = \frac{2\sqrt{2}}{\pi} I_{dc}$ and $I_{rms} = I_{dc} \Rightarrow DF = \frac{2\sqrt{2}}{\pi}$

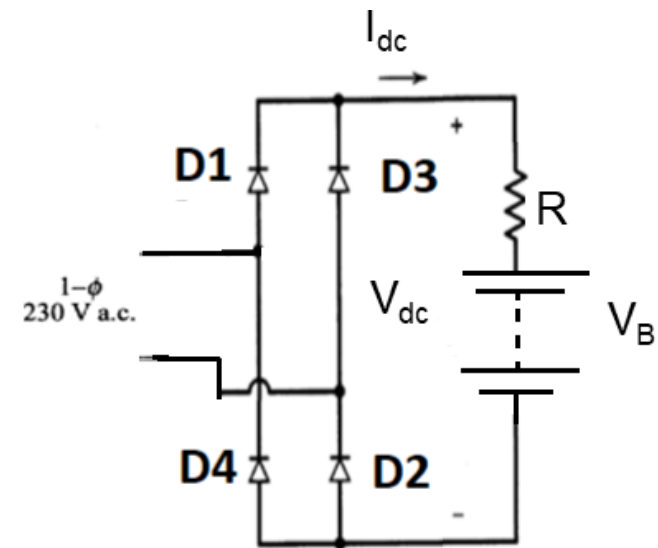
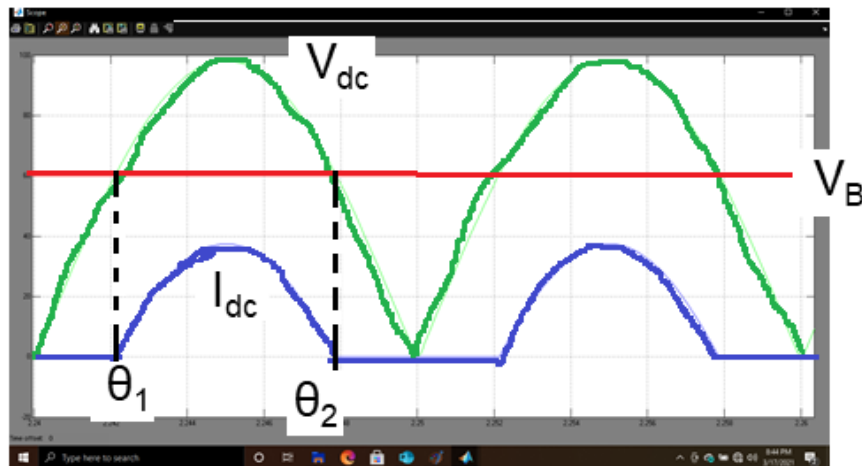
PF = Distortion factor X displacement factor

$$\cos \phi = DF \times \cos \theta$$

- For single phase diode bridge DF = 0.9
- $\cos \theta = 1$
- Hence power factor of 1 ϕ diode bridge = 0.9

1 ϕ Diode bridge : battery charging

- Battery charging application
- Supply voltage = 230 V 50 Hz
- Battery voltage = 150 V
- Internal resistance $R=2\Omega$
- Determine average charging current



Average battery charging current

$$V_m \sin \theta_1 = V_B \quad \text{Putting } V_m = \sqrt{2} \times 230, V_B = 150V$$

$$\therefore \sin \theta_1 = \frac{V_B}{V_m} = 0.4611 \quad \therefore \theta_1 = 27.46^\circ$$

$$\theta_2 = 152.54^\circ$$

$$\text{Inst. current } i = \frac{V_m \sin \theta - V_B}{R}$$

$$I_{av} = \frac{2}{2\pi} \int_{27.46}^{152.54} \left(\frac{V_m \sin \theta - V_B}{R} \right) \cdot d\theta$$

$$I_{av} = \frac{1}{\pi R} \left[\int_{27.46}^{152.54} V_m \sin \theta \cdot d\theta - \int_{27.46}^{152.54} V_B \cdot d\theta \right]$$

Average battery charging current

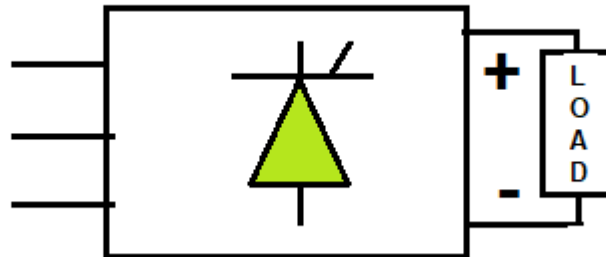
$$\begin{aligned} I_{av} &= \frac{1}{\pi R} \left[-V_m \cos \theta \right]_{27.46}^{152.54} - V_B \left(\frac{\pi}{180} \times 152.54 - \frac{\pi}{180} \times 27.46 \right) \\ I_{av} &= \frac{1}{\pi R} \left[-V_m \cos 152.54 + V_m \cos 27.46 - V_B \times 2.183 \right] \\ &= \frac{1}{\pi \times 2} \left[-\sqrt{2} \times 230 \cos 152.54 + \sqrt{2} \times 230 \cos 27.46 \right. \\ &\quad \left. - 150 \times 2.183 \right] \\ &= \frac{1}{2\pi} \left[288.62 + 288.62 - 327.45 \right] \\ &= \frac{249.79}{2\pi} = 39.755 \text{ A} \end{aligned}$$

1 ϕ Controlled converter

- Classification of controlled converter
- Half wave /Full wave converter
- Half controlled/ full controlled
- Pulse number \Rightarrow 1, 2, 3, 6. 12. 24
- Quadrant of operation
- Single quadrant
- Two quadrant
- Four quadrant

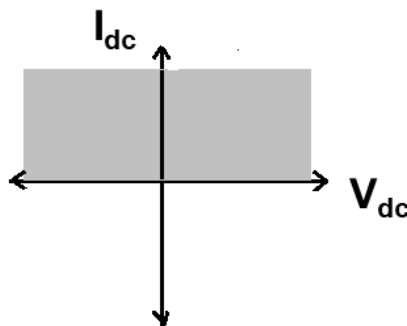
Quadrant operation of converter

Constant voltage
constant frequency
supply

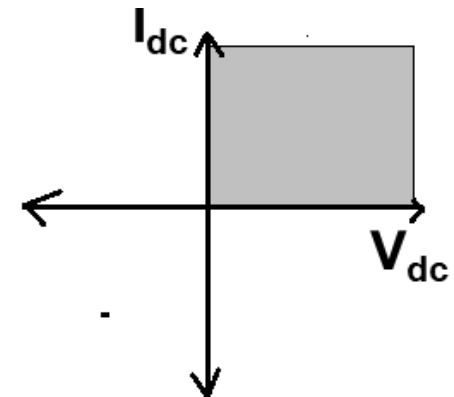


Variable voltage
DC supply

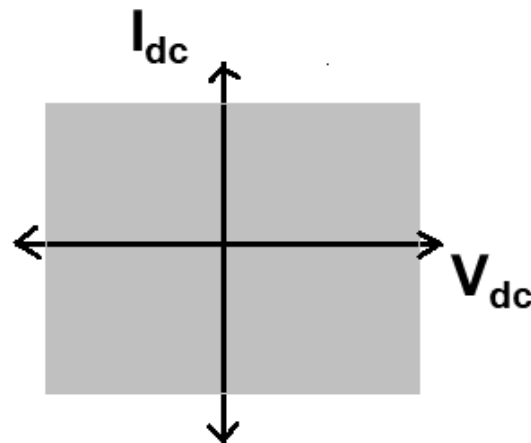
Single quadrant =>



<= Two quadrant

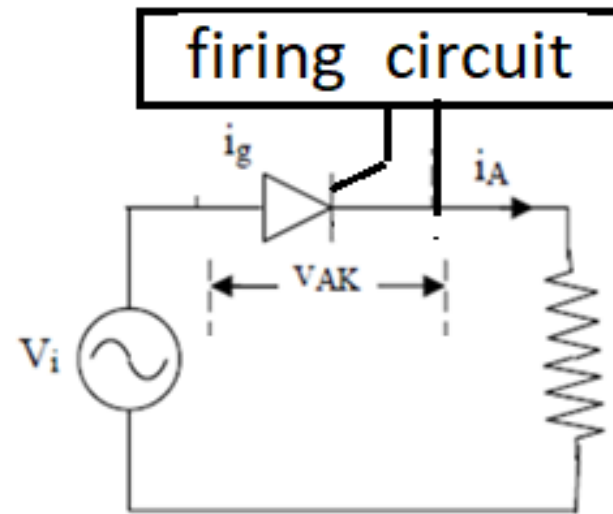
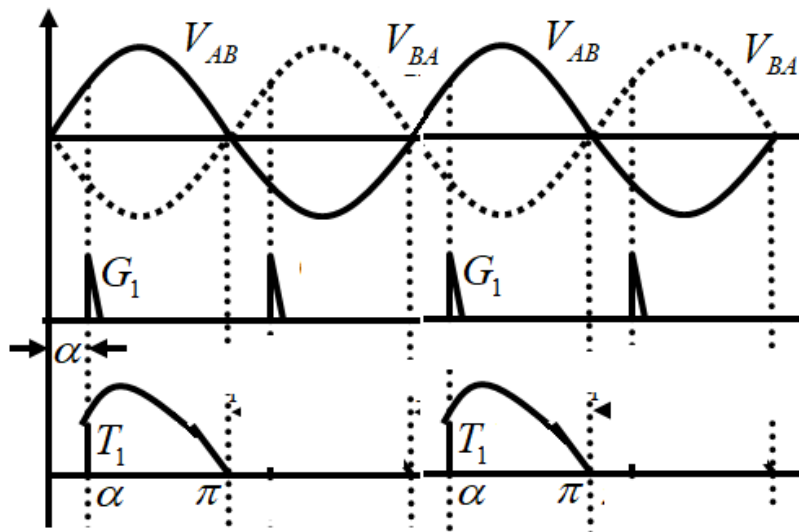


Four quadrant =>



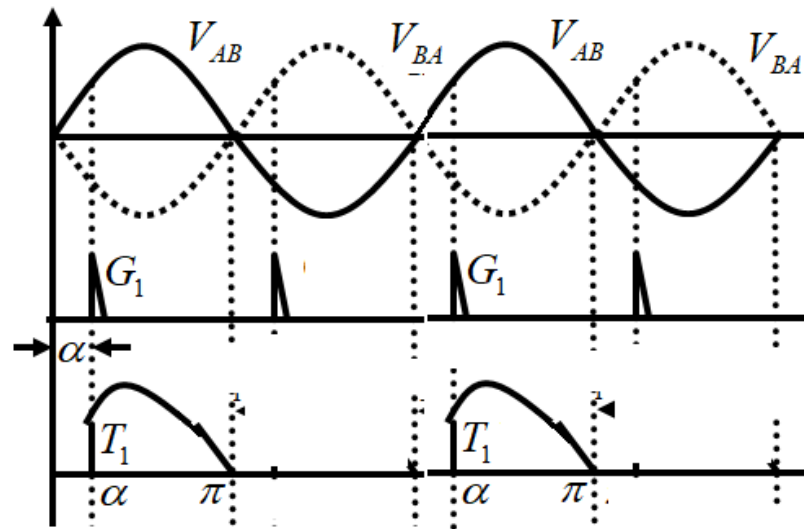
1 ϕ Half wave-controlled converter

- Circuit diagram
- Half wave rectification



1 ϕ Half wave-controlled converter

- O/P voltage waveform
- Supply voltage
- $V_m \sin \theta$

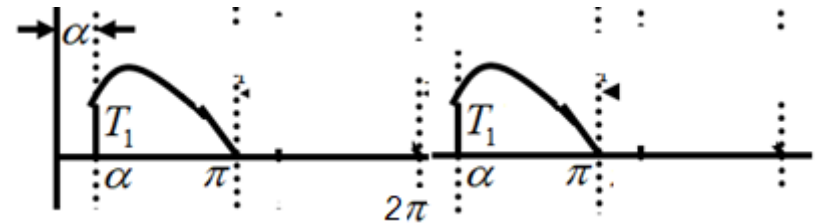


- Average and RMS value of the O/P voltage
- Form factor and ripple factor of the waveform

1 ϕ Half wave-controlled converter

■ Average DC voltage

$$\begin{aligned} V_{dc}(\text{avg}) &= \frac{1}{2\pi} \int_{\alpha}^{\pi} V_m \cdot \sin \theta \cdot d\theta \\ &= \frac{V_m}{2\pi} \left[-\cos \theta \right]_{\alpha}^{\pi} \\ &= \frac{V_m}{2\pi} \left[1 + \cos \alpha \right] \end{aligned}$$



■ $V_m \Rightarrow$ Peak amplitude of supply voltage

1 ϕ Half wave-controlled converter V_{dc} (rms)

$$V_{dc}(\text{rms}) = \left[\frac{1}{2\pi} \int_{\alpha}^{\pi} V_m^2 \sin^2 \theta \cdot d\theta \right]^{1/2}$$
$$= \left[\frac{V_m^2}{2\pi} \int_{\alpha}^{\pi} \sin^2 \theta \cdot d\theta \right]^{1/2}$$

$$= \left[\frac{V_m^2}{2\pi} \int_{\alpha}^{\pi} \left(\frac{1 - \cos 2\theta}{2} \right) \cdot d\theta \right]^{1/2}$$

$$= \left[\frac{V_m^2}{4\pi} \left\{ \int_{\alpha}^{\pi} d\theta - \int_{\alpha}^{\pi} \cos 2\theta \cdot d\theta \right\} \right]^{1/2}$$

$$= \left[\frac{V_m^2}{4\pi} \left\{ \theta \Big|_{\alpha}^{\pi} - \frac{\sin 2\theta}{2} \Big|_{\alpha}^{\pi} \right\} \right]^{1/2}$$

1 ϕ Half wave-controlled converter V_{dc} (rms)

$$V_{dc(rms)} = \left[\frac{V_m^2}{4\pi} \left\{ (\pi - \alpha) - \frac{1}{2} (0 - \sin 2\alpha) \right\} \right]^{1/2}$$

$$= \left[\frac{V_m^2}{4\pi} \left\{ (\pi - \alpha) + \frac{1}{2} \sin 2\alpha \right\} \right]^{1/2}$$

$$= \frac{V_m}{2\sqrt{\pi}} \left[(\pi - \alpha) + \frac{1}{2} \sin 2\alpha \right]^{1/2}$$

$$= \frac{V_m}{2} \sqrt{\frac{(\pi - \alpha) + \frac{1}{2} \sin 2\alpha}{\pi}}$$

Form factor & ripple factor

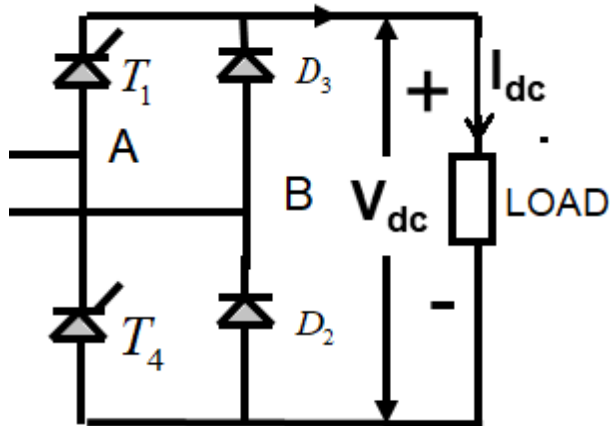
$$\text{Form Factor} = \frac{V_{dc(\text{rms})}}{V_{dc(\text{avg})}}$$

$$FF = \frac{\frac{V_m}{2} \sqrt{\frac{(\pi - \alpha) + \frac{1}{2} \sin 2\alpha}{\pi}}}{\frac{V_m}{2\pi} [1 + \cos \alpha]}$$

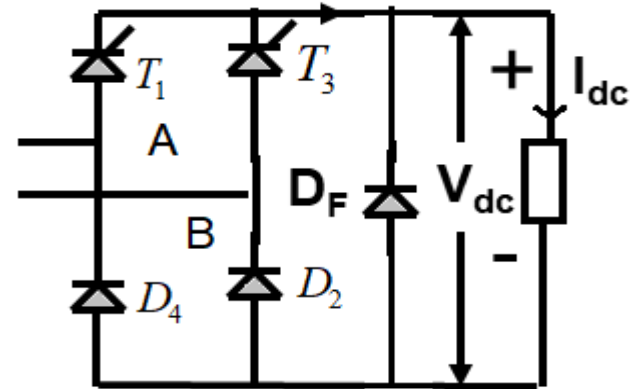
$$\text{RIPPLE FACTOR} = \sqrt{FF^2 - 1}$$

1 ϕ full wave half-controlled converter

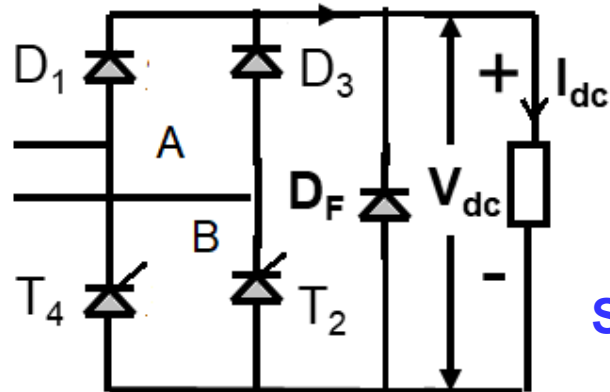
■ Circuit configurations



Asymmetric



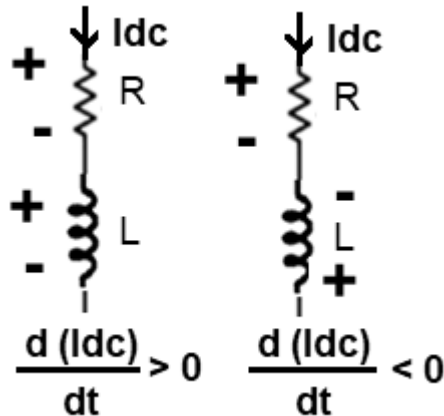
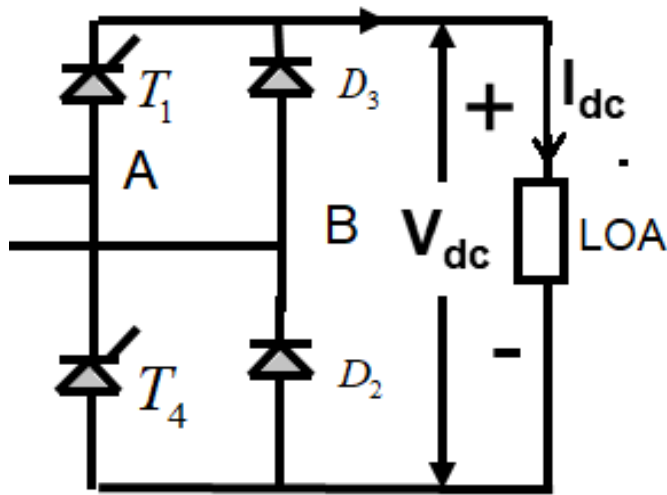
Symmetric



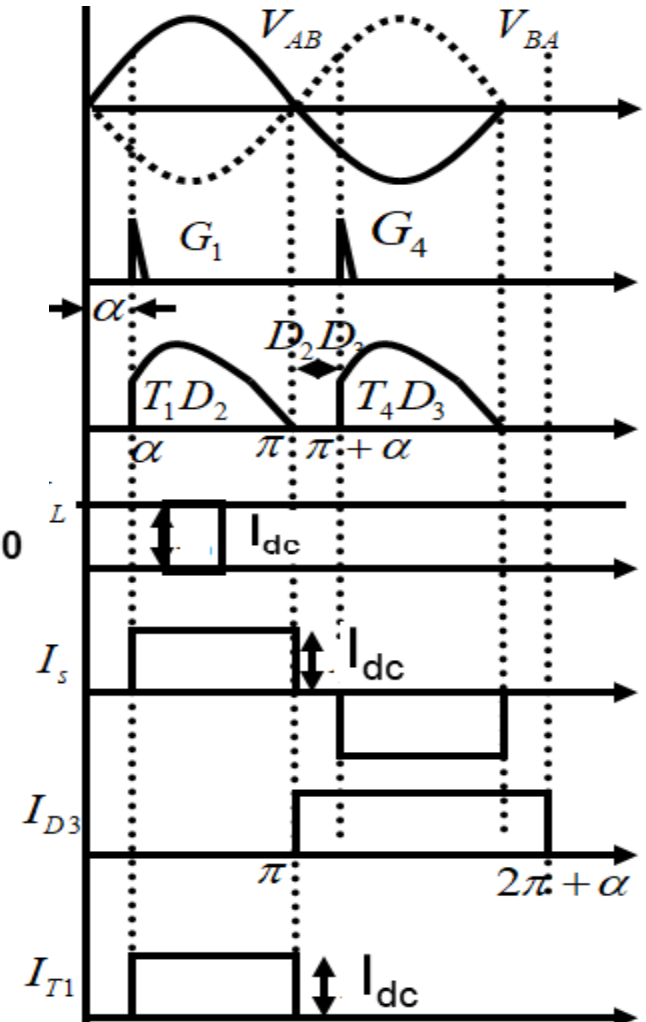
Symmetric

1 ϕ full wave half-controlled converter

■ Asymmetric configuration RL Load

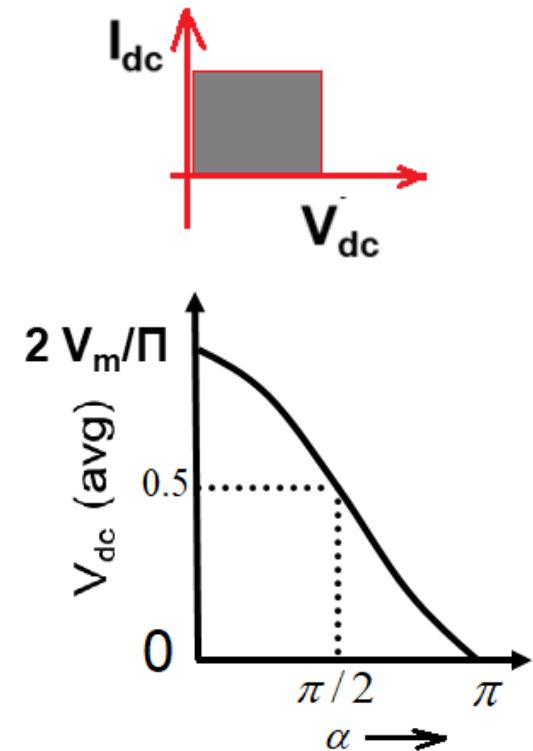


■ Conduction period of diode is more than thyristor

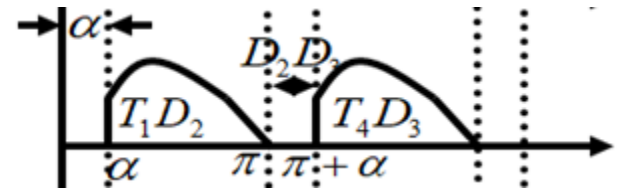


1 ϕ full wave half-controlled converter

- Average value of DC O/P voltage
- $V_{dc} (avg) = (2/2\pi) \int_{\alpha}^{\pi} V_m \sin \theta d\theta$
- $V_{dc} (avg) = (V_m/\pi) (-\cos \theta)$
- $V_{dc} (avg) = (V_m/\pi) (-\cos \pi + \cos \alpha)$
- $V_{dc} (avg) = (V_m/\pi) (1 + \cos \alpha)$
- V_{dc} and I_{dc} are always +ve
- Single quadrant operation
- Unidirectional power flow
- Power flow from AC side to DC side



1 ϕ full wave half-controlled converter : V_{dc} (rms)



$$V_{dc}(rms) = \left[\frac{2}{2\pi} \int_{\alpha}^{\pi} V_m^2 \sin^2 \theta \cdot d\theta \right]^{1/2}$$

$$= \left[\frac{V_m^2}{\pi} \int_{\alpha}^{\pi} \sin^2 \theta \cdot d\theta \right]^{1/2}$$

$$= \left[\frac{V_m^2}{\pi} \int_{\alpha}^{\pi} \frac{(1 - \cos 2\theta)}{2} \cdot d\theta \right]^{1/2}$$

$$= \left[\frac{V_m^2}{2\pi} \int_{\alpha}^{\pi} (1 - \cos 2\theta) \cdot d\theta \right]^{1/2}$$

$$= \left[\frac{V_m^2}{2\pi} \left\{ \int_{\alpha}^{\pi} d\theta - \int_{\alpha}^{\pi} \cos 2\theta \cdot d\theta \right\} \right]^{1/2}$$

1 ϕ full wave half-controlled converter : V_{dc} (rms)

$$\begin{aligned} V_{dc}(\text{rms}) &= \left[\frac{V_m^2}{2\pi} \left\{ 0 \Big|_{\alpha}^{\pi} - \frac{\sin 2\theta}{2} \Big|_{\alpha}^{\pi} \right\} \right]^{1/2} \\ &= \left[\frac{V_m^2}{2\pi} \left\{ (\pi - \alpha) - \frac{1}{2} (0 - \sin 2\alpha) \right\} \right]^{1/2} \\ &= \left[\frac{V_m^2}{2\pi} \left\{ (\pi - \alpha) + \frac{1}{2} \sin 2\alpha \right\} \right]^{1/2} \\ &= \frac{V_m}{\sqrt{2\pi}} \left\{ (\pi - \alpha) + \frac{1}{2} \sin 2\alpha \right\}^{1/2} \\ &= \frac{V_m}{\sqrt{2\pi}} \sqrt{(\pi - \alpha) + \frac{1}{2} \sin 2\alpha} \end{aligned}$$

Form factor & ripple factor

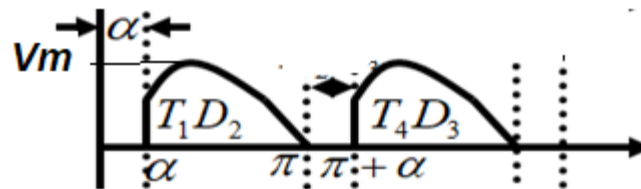
$$\text{Form factor} = \frac{V_{dc}(\text{rms})}{V_{dc}(\text{avg})}$$

$$FF = \frac{\frac{V_m}{\sqrt{2\pi}} \sqrt{(\pi - \alpha) + \frac{1}{2} \sin 2\alpha}}{\frac{V_m}{\pi} (1 + \cos \alpha)}$$

$$\text{Ripple factor} = \sqrt{FF^2 - 1}$$

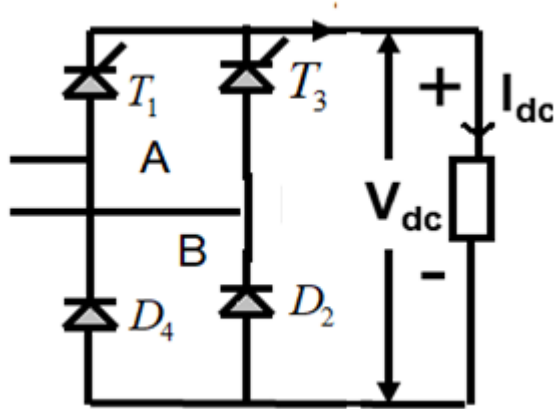
1 ϕ half controlled converter

- Resistive load
- o/p voltage waveform is same as that of RL Load.
- No freewheeling action
- Dc current and source current are different
- Power dissipated in R load
- $P = V_{dc}^2(\text{rms})/R$

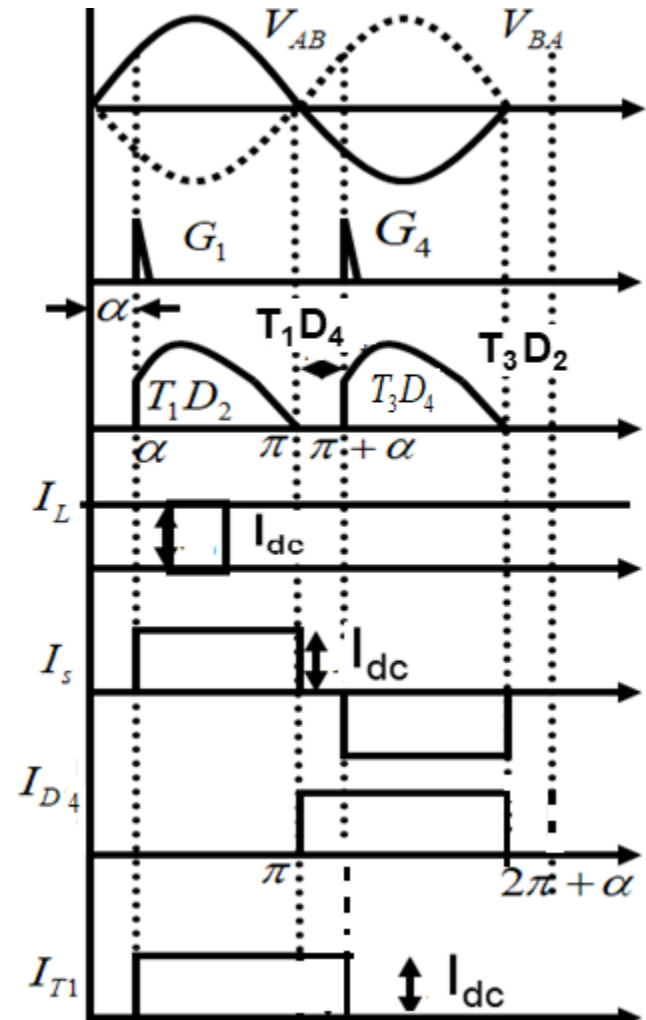


1 ϕ full wave half controlled converter

- Symmetric configuration
- without Df



- RL load \Rightarrow each device
- Conducts for 180°

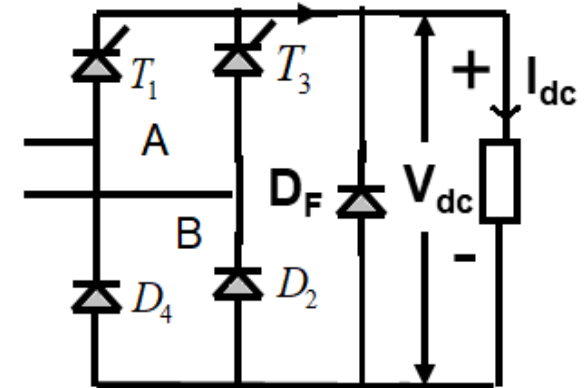
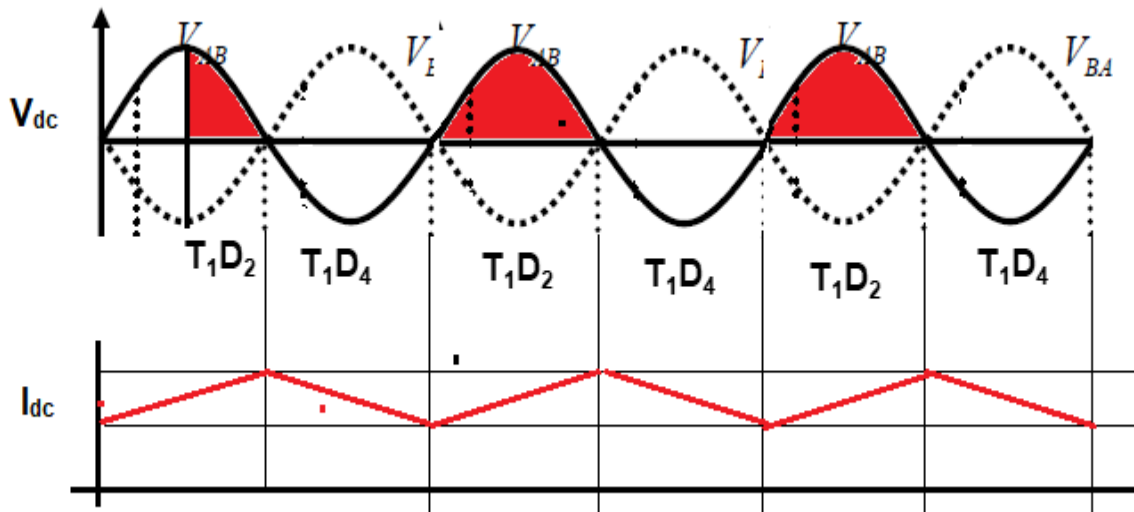
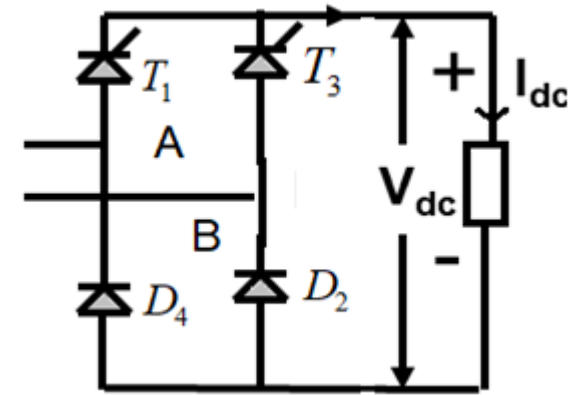


1 ϕ full wave half-controlled converter => Symmetric conf.

- Expressions for average and rms voltage, Form factor and ripple factor same as Asymmetric configuration
- T1 conducts till T3 turns on.
- If T3 fails to turn on, T1 and D4 will conduct for –ve half cycle => o/p voltage =0
- In positive half cycle T1 and D2 will conduct. T1 never turns off. Uncontrolled half wave DC O/P
- Half waving effect

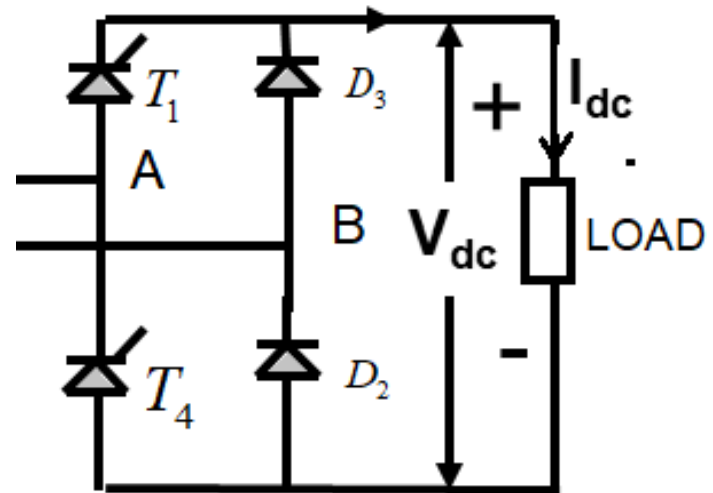
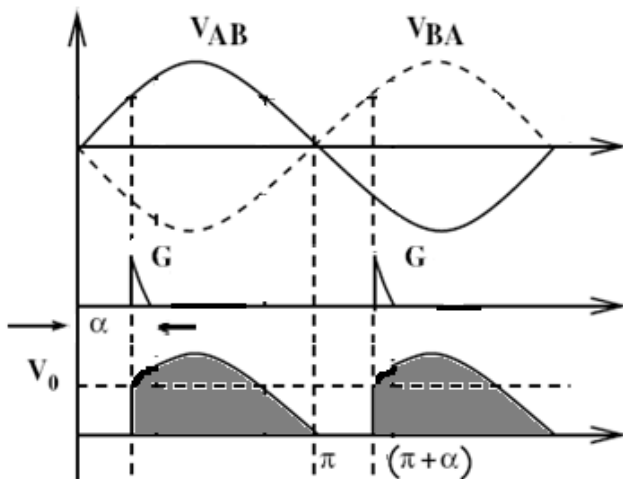
Half waving effect

- T3 fails to turn on
- T1 will not turn off
- T1 will remain continuous on



1 ϕ full wave half controlled converter \Rightarrow R load

- Power dissipated in R load
- Supply voltage = 230V 1 ϕ 50 Hz
- $\alpha = 30^\circ$ & $R = 10\ \Omega$
- Find out V_{dc} (rms)



1 ϕ full wave half-controlled converter \Rightarrow R load

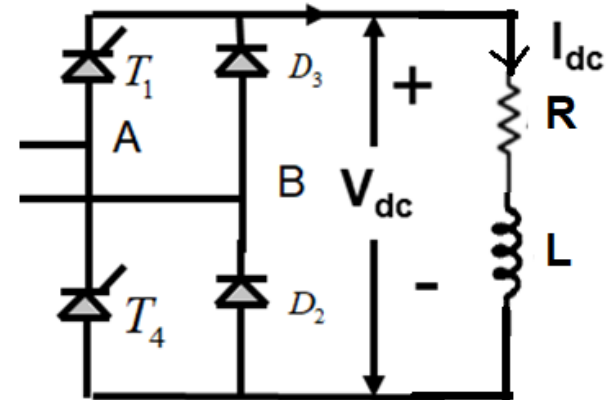
- Vdc(rms) is given by the expression

$$V_{dc(rms)} = \frac{V_m}{\sqrt{2\pi}} \sqrt{(\pi - \alpha) + \frac{1}{2} \sin 2\alpha}$$

- Substituting $V_m \Rightarrow \sqrt{2} \times 230$ and $\alpha = 30^\circ$
- $V_{dc(rms)} = 226.56 \text{ V}$
- Power dissipated $= (226.56)^2 / 10 = 5132.94 \text{ W}$

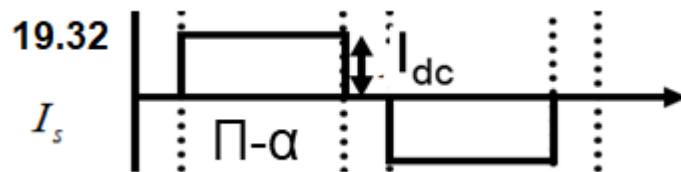
1 ϕ full wave half controlled converter => RL load

- Supply voltage = 230V , R= 10 Ω
- $\alpha = 30^\circ$ and large value of L
- Find the power dissipated in R Load
- I_{dc} is smooth & ripple free
- V_R = Cons. Ripple free DC
- Power dissipated = $V_{dc} (avg)^2 / R$
- $V_{dc} (avg) = V_m / \pi (1 + \cos \alpha)$
- $V_{dc} (avg) = 193.2 \text{ V}$
- $P_{dc} = 3732.68 \text{ W}$

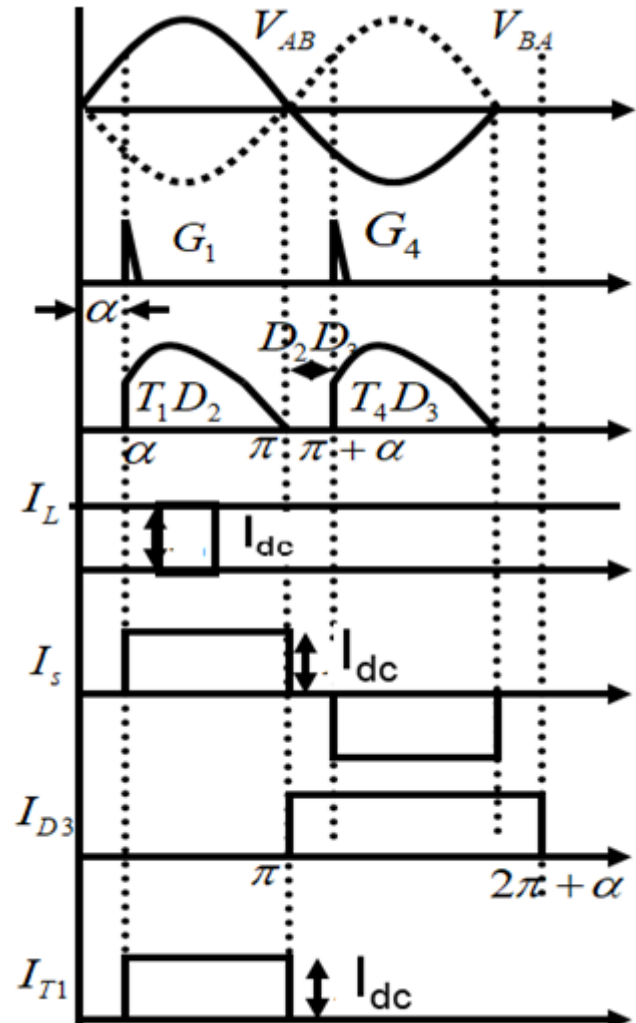


1 ϕ full wave half controlled converter => RL load

- Source current & pf
- $I_{dc} = 19.32A$



- $I_s (\text{rms}) = [(I_{dc}^2 \times 150 \times 2) / 360]^{1/2}$
- $I_s (\text{rms}) = 17.63A$



1 ϕ full wave half controlled converter \Rightarrow RL load

- AC Power = DC power
- $V_{dc}(avg) \times I_{dc} = V_s I_s \times \cos\phi$
- Putting values
- $V_s = 230 \text{ V}$ $I_s = 17.63 \text{ A}$ $V_{dc}(avg) = 193.2 \text{ V}$
- $I_{dc} = 19.32 \text{ A}$
- Power factor = $\cos\phi = 0.9179$ lag

1 ϕ full wave half controlled converter => RE load

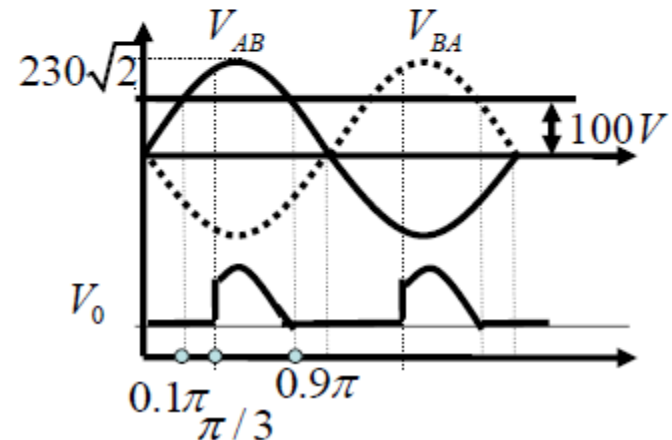
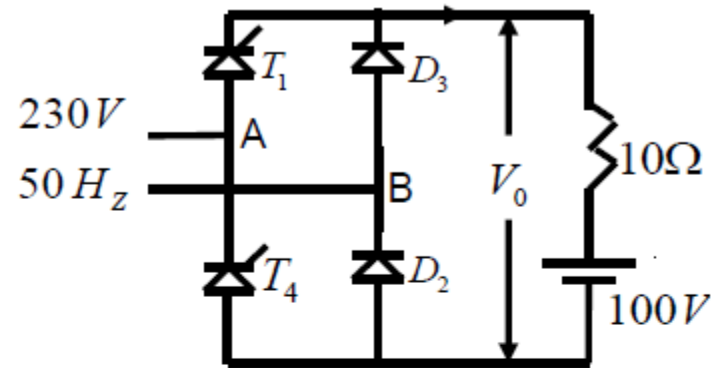
For the circuit shown in fig.
determine average value of load
current for $\alpha = 60^\circ$

What is the new value of average
current flowing through load if a
large 'L' is connected in series with
the load

Neglect the device drop

a) R-E Load:

$$\alpha_{\min} = \sin^{-1} \left(\frac{100}{230 * \sqrt{2}} \right) = 18^\circ = 0.1\pi^c$$



1 ϕ full wave half controlled converter \Rightarrow RE load

$$E = 100 \text{ V}$$

$$\sqrt{2} V_m \sin \theta_1 = 100$$

$$\therefore \sin \theta_1 = \frac{100}{V_m} = \frac{100}{\sqrt{2} \times 230} \therefore \sin \theta_1 = 0.30743$$

$$\theta_1 = \sin^{-1} 0.307437 = 17.90^\circ \approx 18^\circ$$

But $\alpha = 60^\circ$ \therefore conduction starts at 60°

and conduction ends at $180 - 18 = 162^\circ$

1 ϕ full wave half controlled converter \Rightarrow RE load

$$\therefore i = \frac{V_m \sin \alpha - E}{R}$$

$$\therefore I_{av} = \frac{2}{2\pi} \int_{60}^{162} \left(\frac{V_m \sin \alpha - E}{R} \right) \cdot d\alpha$$

$$\therefore I_{av} = \frac{1}{\pi R} \left[\int_{60}^{162} V_m \sin \alpha \cdot d\alpha - \int_{60}^{162} E \cdot d\alpha \right]$$

$$\therefore I_{av} = \frac{1}{10\pi} \left[V_m \left(-\cos \alpha \right)_{60}^{162} - E \left(\frac{162}{180} \pi - \frac{60}{180} \times \pi \right) \right]$$

1 ϕ full wave half controlled converter => RE load

$$\begin{aligned}\therefore I_{AV} &= \frac{1}{10\pi} \left[\sqrt{2} \times 230 (-\cos 162 + \cos 60) - 100 (1.780) \right] \\ &= \frac{1}{10\pi} \left[\sqrt{2} \times 230 (0.9510 + 0.5) - 178 \right] \\ &= \frac{1}{10\pi} \left[471.96 - 178 \right] = 9.35 \text{ A}\end{aligned}$$

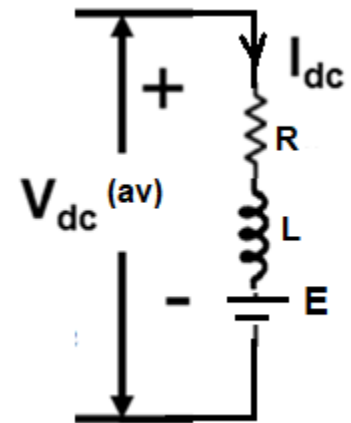
1 ϕ full wave half controlled converter => RLE load

- b **With large inductance in series with the load current becomes continuous.**

$$\alpha = 60^\circ$$

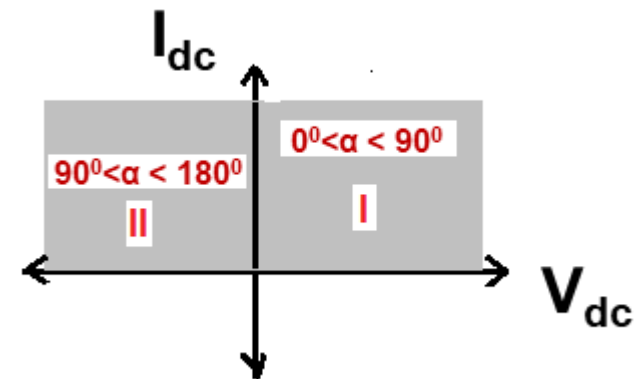
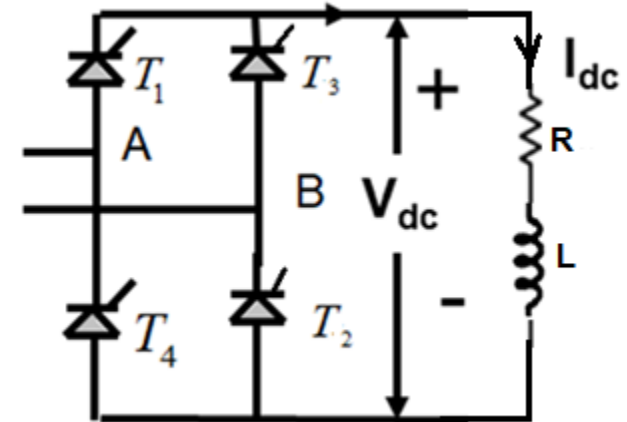
$$\begin{aligned} \text{Av. value of o/p Voltage} &= \frac{230\sqrt{2}}{\pi} (1 + \cos\alpha) \\ &= 155 \text{ V} \end{aligned}$$

$$\text{Av. value of } I = \frac{155 - 100}{10} = 5.5 \text{ A}$$



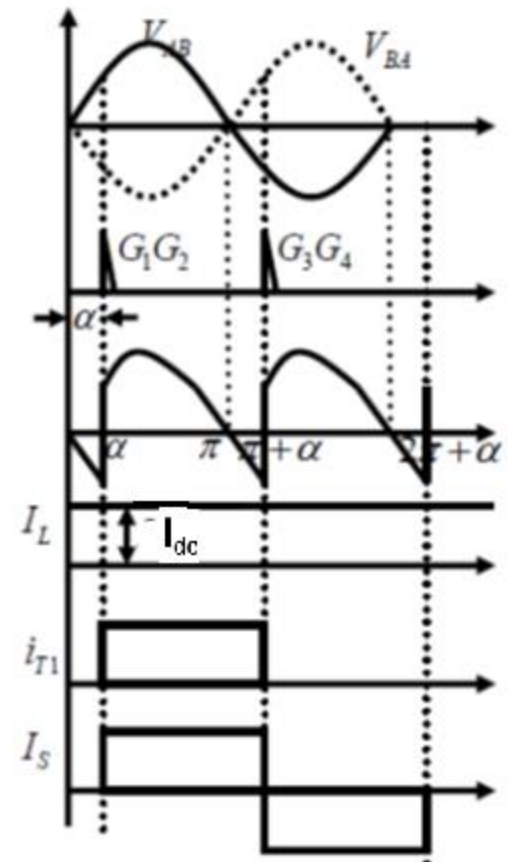
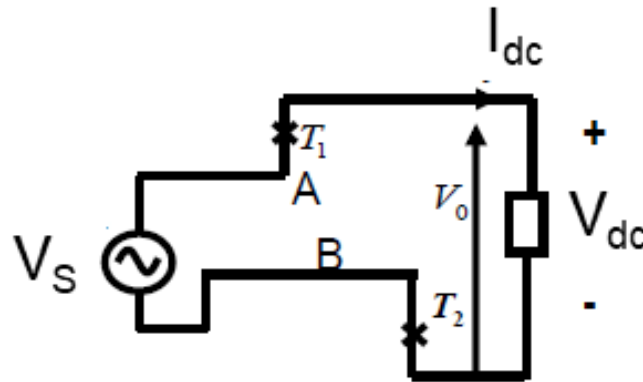
1 ϕ full wave full controlled converter

- With RL load & continuous conduction
- All devices are thyristors
- Range of $\alpha \Rightarrow 0$ to 180
- Two quadrant converter
- Rectifier operation
- Inverter operation
- Current direction remains same
- Polarity of O/P voltages changes



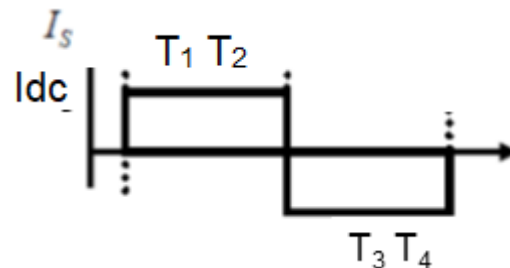
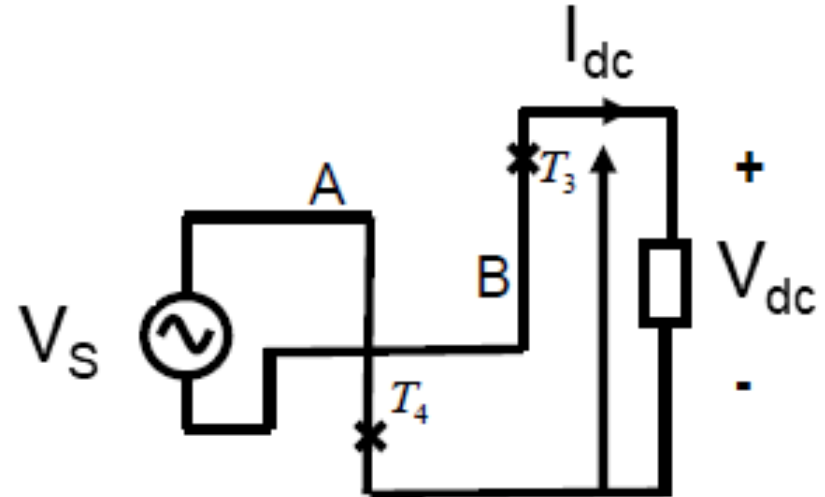
1 ϕ full wave full controlled converter

- operation
- Assume $I_L(I_{dc})$ is continuous, smooth and ripple free
- As continuous conduction $T_3 T_4$ are conducting before turn on of $T_1 T_2$
- Positive half cycle V_{AB} +ve $T_1 T_2$ are fired at α



ϕ full wave full controlled converter

- operation
- At $\pi + \alpha$, $T_3 T_4$ are turned on
- Each device Conducts for 180°
- Source current is square wave

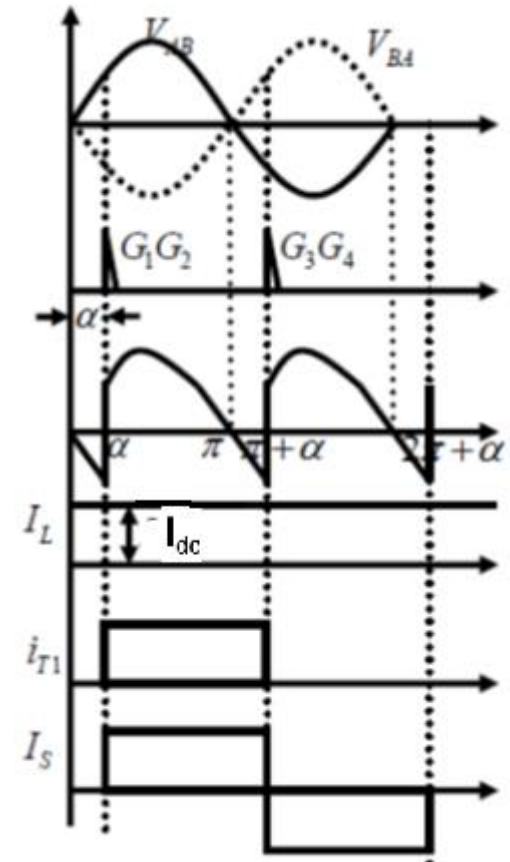


1 ϕ full wave full controlled converter

■ Rectifier operation

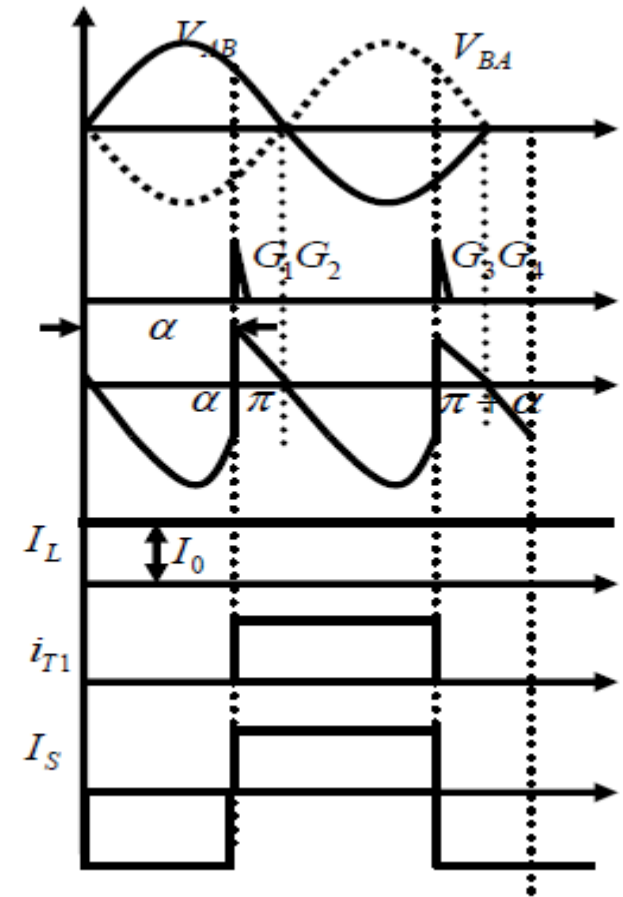
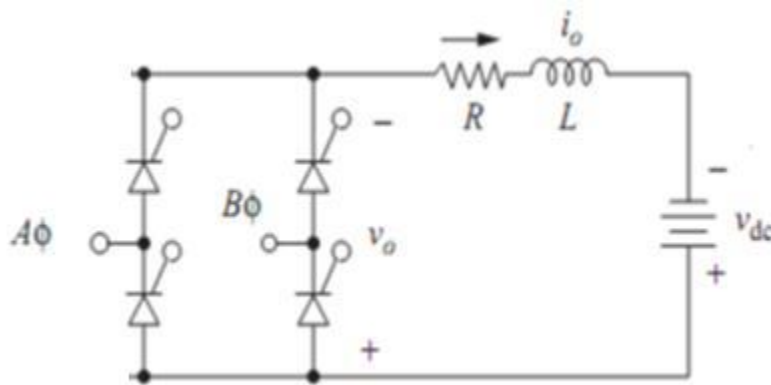
- $0^\circ < \alpha < 90^\circ$
- Each device conducts for 180°
- T1 and T2 will turn off when T3
- T4 are turned ON
- $V_{dc}(\text{avg}) = \frac{2}{2\pi} \int_{\alpha}^{\pi+\alpha} V_m \sin\theta d\theta$
- $V_{dc}(\text{avg}) = \frac{V_m}{\pi} \int_{\alpha}^{\pi+\alpha} \sin\theta d\theta$
- $V_{dc}(\text{avg}) = \frac{V_m}{\pi} (-\cos\theta d\theta)$
- $V_{dc}(\text{avg}) = 2 (V_m / \pi) \cos\alpha$

- $$V_{dc}(\text{avg}) = \frac{2V_m}{\pi} \cos\alpha$$



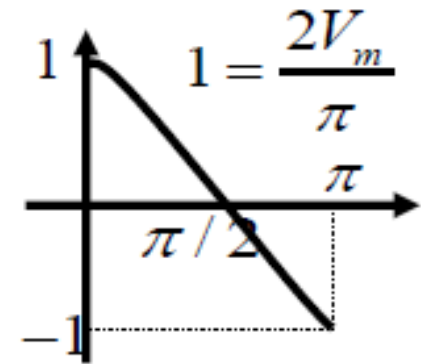
1 ϕ full wave full controlled converter

- Inverter operation
- Essential conditions for Inverter operation
 - 1) $90^\circ < \alpha < 180^\circ$
 - 2) continuous conduction
 - 3) DC source at the DC side



1 ϕ full wave full controlled converter

- Voltage control
- source current \Rightarrow square wave
- $I_s(\text{rms}) = I_{\text{dc}}$
- $I_1(\text{rms}) = \frac{2\sqrt{2}}{\pi} I_{\text{dc}}$
- $V_{\text{dc}}(\text{rms}) = V_m/\sqrt{2}$
- Form factor (FF) = $V_{\text{dc}}(\text{rms}) / V_{\text{dc}}(\text{avg})$



- $$\text{FF} = \frac{V_m/\sqrt{2}}{2 (V_m/\pi) \cos\alpha} = \frac{\pi}{2\sqrt{2}\cos\alpha}$$

1 ϕ full wave full controlled converter

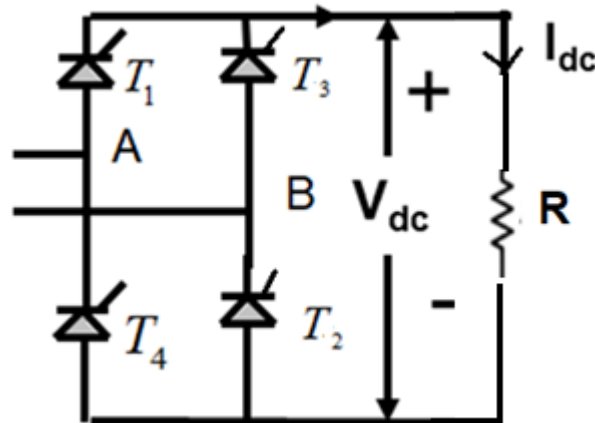
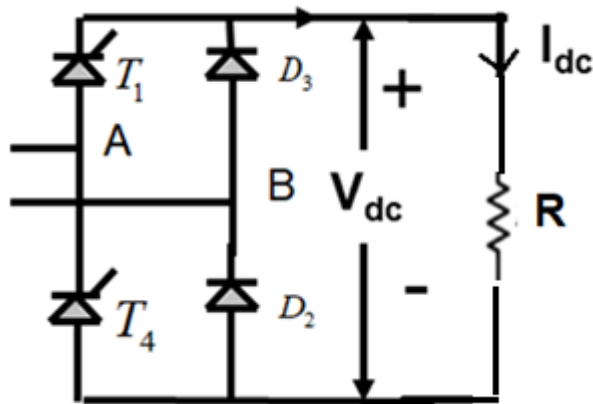
- Ripple factor = $\sqrt{(FF^2 - 1)}$
- Power factor =
- Distortion factor X Displacement factor
- Distortion factor = $(2\sqrt{2})/\pi$
- Displacement factor = $\cos \alpha$
- Power factor = $\frac{2\sqrt{2}}{\pi} \cos \alpha$

1 ϕ full wave full controlled converter

R load

■ Half controlled

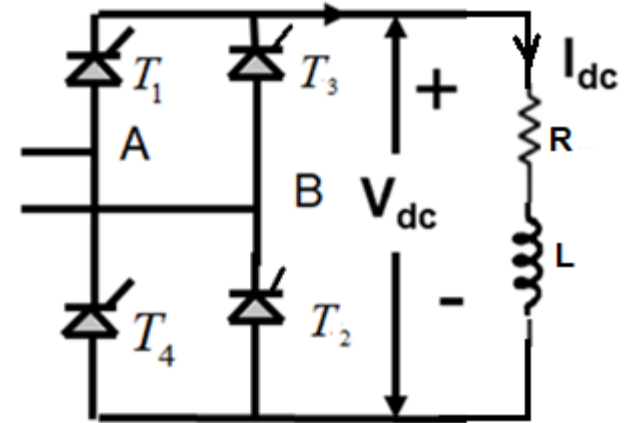
Full controlled



■ Operation and performance is same for both converters

1 ϕ full wave full controlled converter

- RL load continuous conduction
- Supply voltage = 230 V
- $\alpha = 45^\circ$ $R = 20\ \Omega$ & $L = \text{large}$
- Power dissipated in R load
- Source current & supply pf
- % ripple in DC O/P voltage
- Power dissipated in R load = $V_{dc}(\text{avg}) \times I_{dc}(\text{avg})$
- $V_{dc}(\text{avg}) = 2 (V_m / \pi) \cos \alpha$
- $V_{dc}(\text{avg}) = 146.42\text{ V}$
- I_{dc} is smooth and ripple free



RL load continuous conduction

- $I_{dc} = V_{dc(av)}/R = 146.42/20 = 7.321\text{ A}$
- Power dissipated in R load $V_{dc(av)} \times I_{dc}$
- $P_{dc} = 146.42 \times 7.321 = 1071.94\text{ W}$
- Source current is square wave with 7.321 amplitude
- $I_s(\text{rms}) = 7.321\text{ A}$
- $V_s I_s \cos\phi = P_{dc}$, putting values $\cos\phi = 0.6366$

$$P_f = \frac{2\sqrt{2}}{\pi} \cos\alpha$$

RL load continuous conduction

- form factor= $v_{dc}(\text{rms}) / V_{dc}(\text{av})$
- $FF = 230/146.42 = 1.5708$
- Ripple factor = 1.211
- %ripple = 121.1%