

# K. J. SOMAIYA COLLEGE OF ENGINEERING

(AFFILIATED TO THE UNIVERSITY OF MUMBAI)

Candidate Roll No. \_\_\_\_\_ (In figures)

Name : Basis of Digital Communication Test Exam.

Date : \_\_\_\_\_ 200

Examination : \_\_\_\_\_ Branch/Semester \_\_\_\_\_

Subject : \_\_\_\_\_

Junior Supervisor's full  
Signature with Date

Question No.	1	2	3	4	5	6	7	8	9	10	11	12	Total
Marks Obtained													

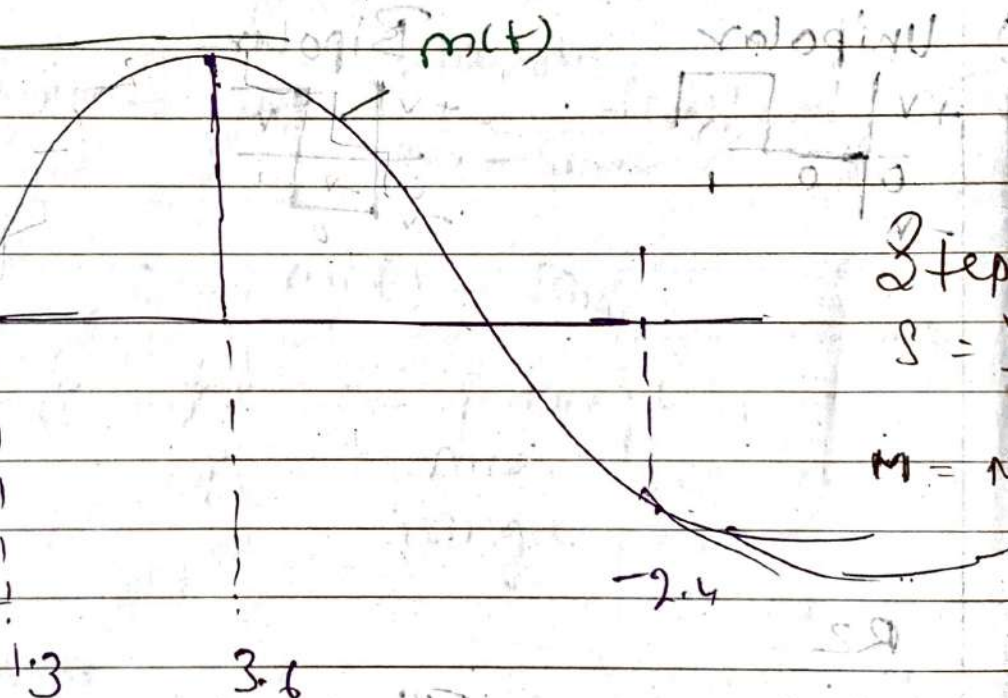
PCM:

## Digital Representation of Analog Signal:

Sampling  
Quantization  
Encoding

Code No.

Quant. level	max	min
7	3.5	3
6	2.5	2
5	1.5	1
4	0.5	0
3	-0.5	-1
2	-1.5	-2
1	-2.5	-3
0	-3.5	-4
	$V_{min}$	



Step Size

$$S = \frac{V_H - V_L}{2^M - 1}$$

$M = \text{No. of levels}$

a Sample Value - 1.3

b Nearest Quantization level - 1.5

c Code No. - 5

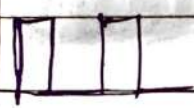
Binary representation - 101

Step Size

$$S = \frac{V_H - V_L}{\text{Number of levels}}$$

$$S = \frac{V_H - V_L}{2^M} = 256$$

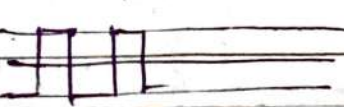
## Electrical Representation of Binary Digits



101

+V

-V



Pulse Representation

Representation for Voltage level



## Line Coding

### Issues in Digital Transmission

- (i) Power & B.W. required for transmission
- (ii) ability to Extract timing information
- (iii) Error monitoring ability
- (iv) Presence of low freq. or dc Component which is unsuitable for ac Coupled Ckt

(i) Unipolar

(ii) Bipolar

(iii) Biphasic

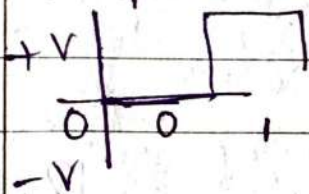
(i) NRZ  $\leftarrow$  NRZ-L  
Mark - M  
Space - S

RZ  $\leftarrow$

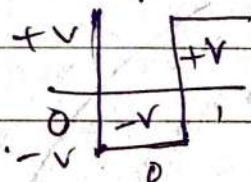
Split Phase or Manchester Coding

### NRZ

(i) Unipolar



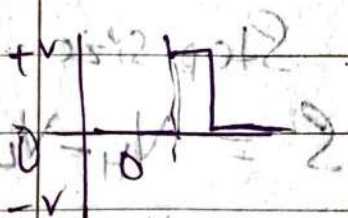
Bipolar



(ii)

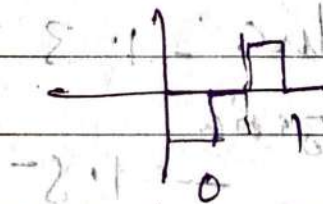
RZ

Unipolar



AMZ

Bipolar RZ



(iii)

Split Phase (Manchester)





## Quantization error

Quantized Signal and the Original Signal from which it was derived different from one another in a random manner. This difference or error may be viewed as a noise due to the quantization process and is called quantization error.

∴ Mean Square quantization error  
 $\overline{e^2}$  Where  $e = \text{diff. between the Original \& Quantized Signal Voltage}$

$$\overline{e^2} = \int_{m_1 - S/2}^{m_1 + S/2} f(m) (m - m_1)^2 dm + \int_{m_2 - S/2}^{m_2 + S/2} f(m) (m - m_2)^2 dm + \dots$$

(Mean Square Quantization error)

Where  $e = m(t) - m_k$

&  $f(m) dm = \text{probability that } m(t) \text{ lies in the voltage range } m - dm/2 \text{ to } m + dm/2$

Substitute  $x = m - m_k$

$$\overline{e^2} = (f^{(1)} + f^{(2)} + \dots) \int_{-S/2}^{S/2} x^2 dx$$

$$= (f^{(1)} + f^{(2)} + \dots) \frac{S^3}{12}$$

$$(f^{(1)}(S) + f^{(2)}(S) + \dots) \frac{S^2}{12}$$

$$\boxed{\overline{e^2} = \frac{S^2}{12}}$$

$f^{(1)}(S)$  is the Prob. that the signal voltage  $m(t)$  will be



In first quantization range,  $f^{(1)}s$  is the probability that  $m$  is in the second quantization range. Hence the sum of terms in the parentheses has a value of unity.

$\overline{e^2}$  = mean square quantization

error is  $\boxed{\overline{e^2} = \frac{s^2}{12}}$

No. of Quantization level

$$Q = 2^N$$

where  $N$  = No. of bits/word

Each sample convert to  $N$  bit  
Code word

No. of bits/sec.

$$\boxed{\text{Signaling rate}} = \text{No. of samples/sec.} \times \text{No. of bits/sample}$$

$$= f_s \times N$$

$$= \boxed{f_s \times N}$$

$$\boxed{2f_s N}$$

Transmitted B.W. of PCM

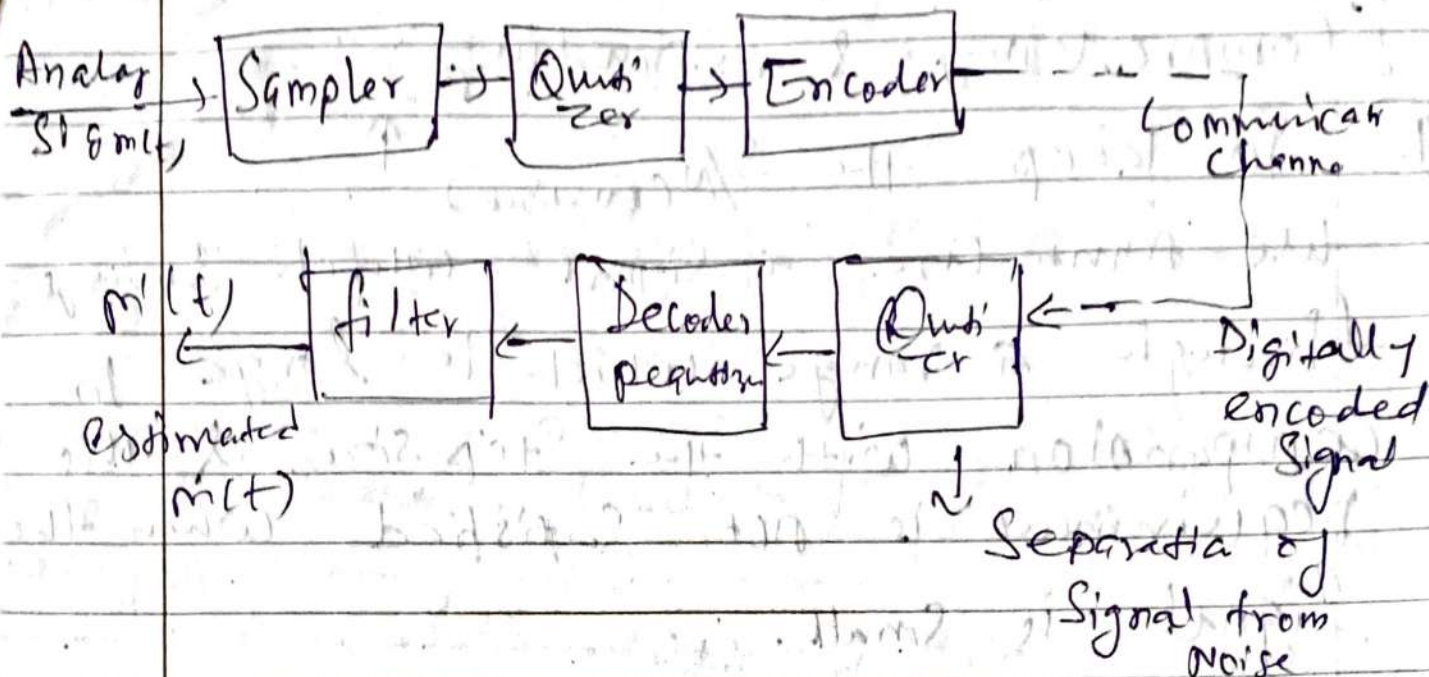
$$\boxed{B.W. = \frac{1}{2} N f_s}$$



# PCM

Encoder

Decoder



## Companding

In quantization voltage range from  $V_L$  to  $V_H$

① If the signal  $m(t)$  should make excursion beyond the bounds  $V_H$  and  $V_L$ , within these bounds, the instantaneous quantization error never exceeds  $\pm S/2$  while outside these bounds the error is larger.

② Similarly consider a case in which  $m(t)$  has a p.p.p. voltage which is less than  $S$  and never crosses one of the transition level. In such a case  $m(t)$  will be a fixed dc voltage & will bear no relationship to  $m(t)$ . Reconstruction is not possible.

$$\text{Noise } \overline{e^2} = \frac{S^2}{12}$$

$S = \frac{V_H - V_L}{2}$

If no. of quantization level is  $M$   
then  $MS = 2V$   $V = \frac{MS - S}{2}$



The dynamic range can be materially improved by a process called Compressing [Compressing & expanding]

(iii) To keep the  $S/N$  constant  $\uparrow$   
 we must use a signal which swings through a range which is large in comparison with the step size. & this requirement is not satisfied when the signal is small.

(iv) Before applying the signal to the quantizer we pass it through a  $N/w$  which has an input O/p characteristic as shown

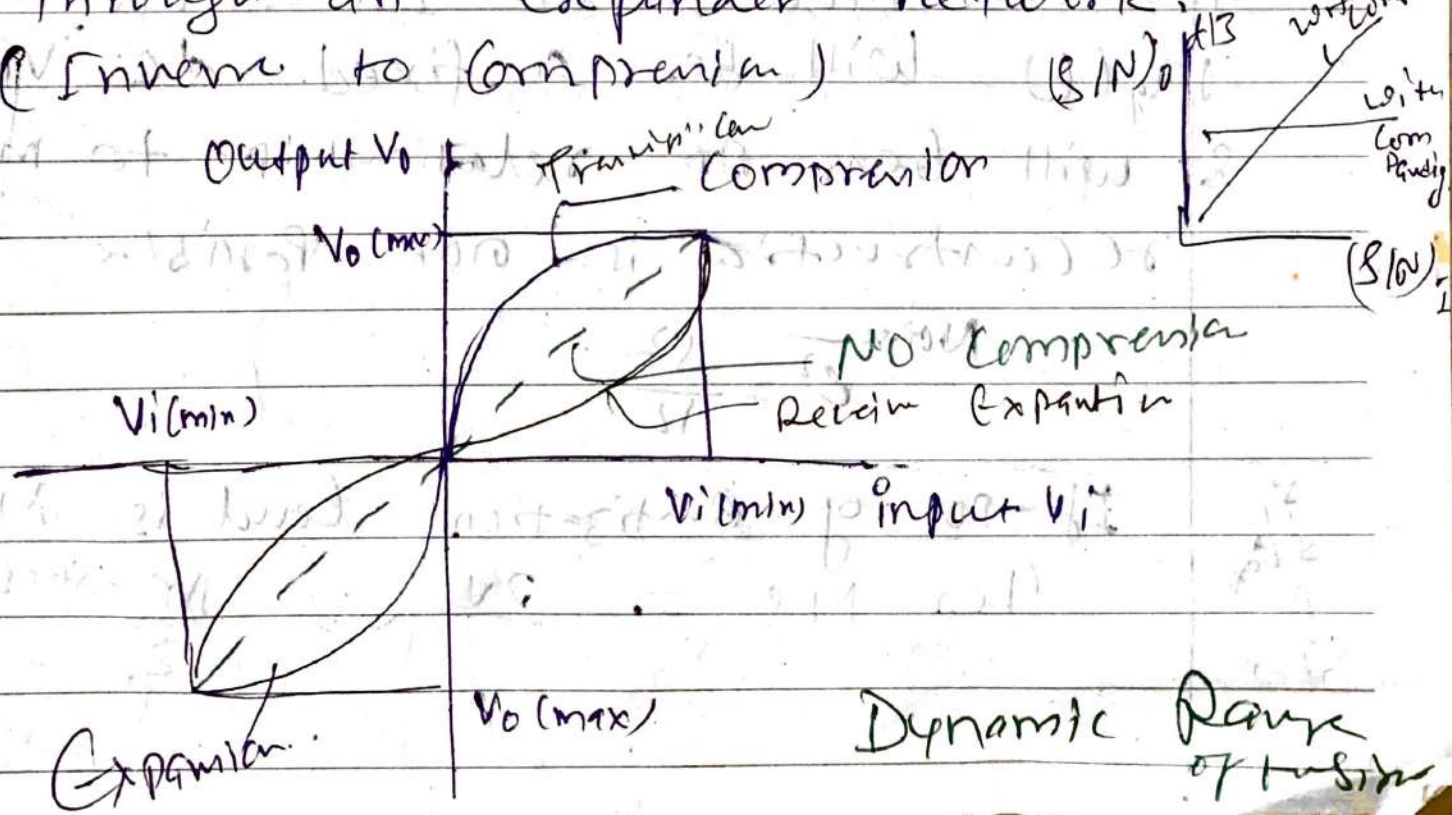
[at low amplitude the slope is large than at large amplitude]

A signal transmitted through such a  $N/w$  will have the extremities of its waveform compressed.

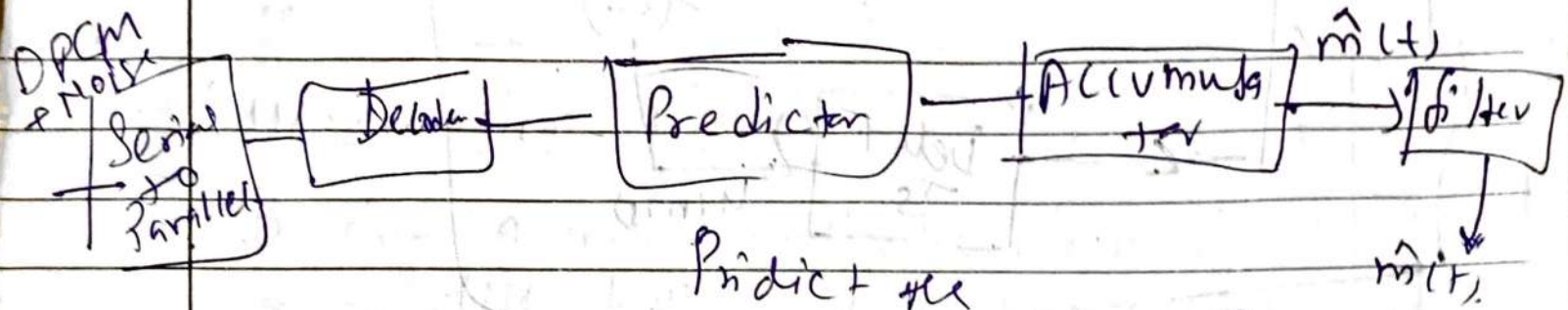
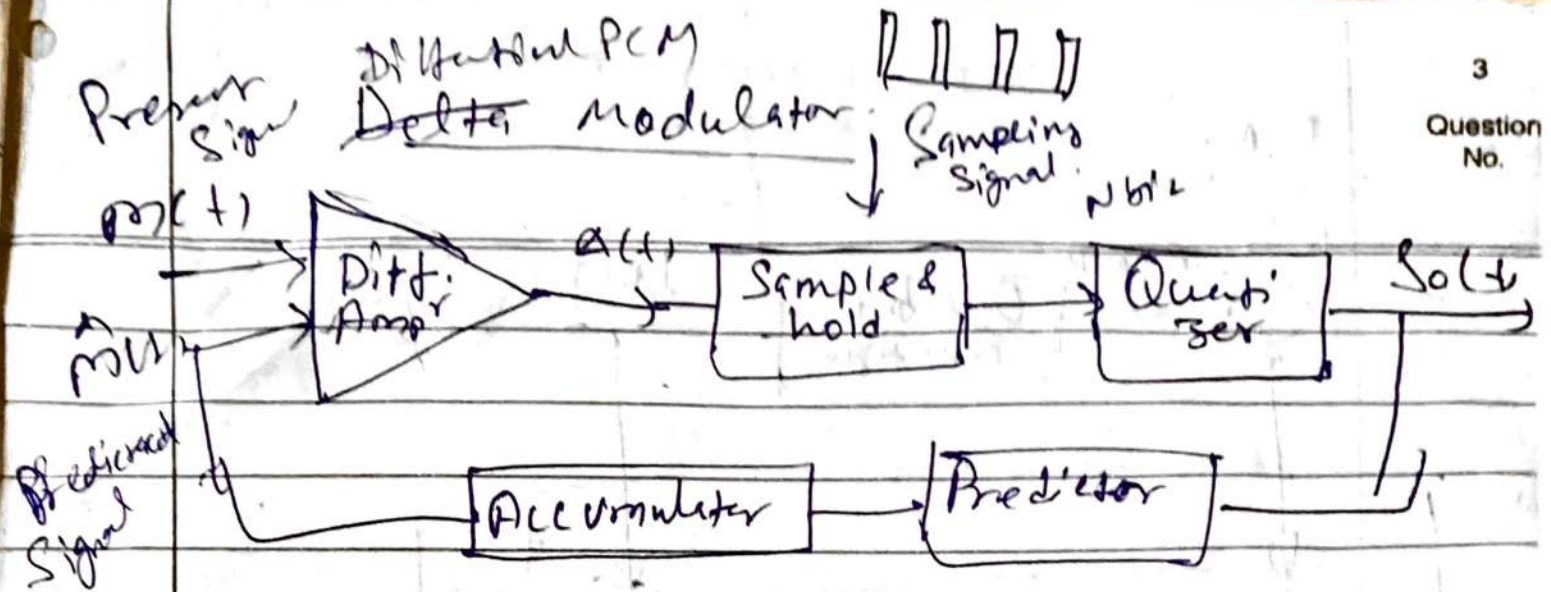
The compression produces signal distortion. To undo the distortion, at the

receiver we pass the received signal through an expander network.

(Inverse to Compression)







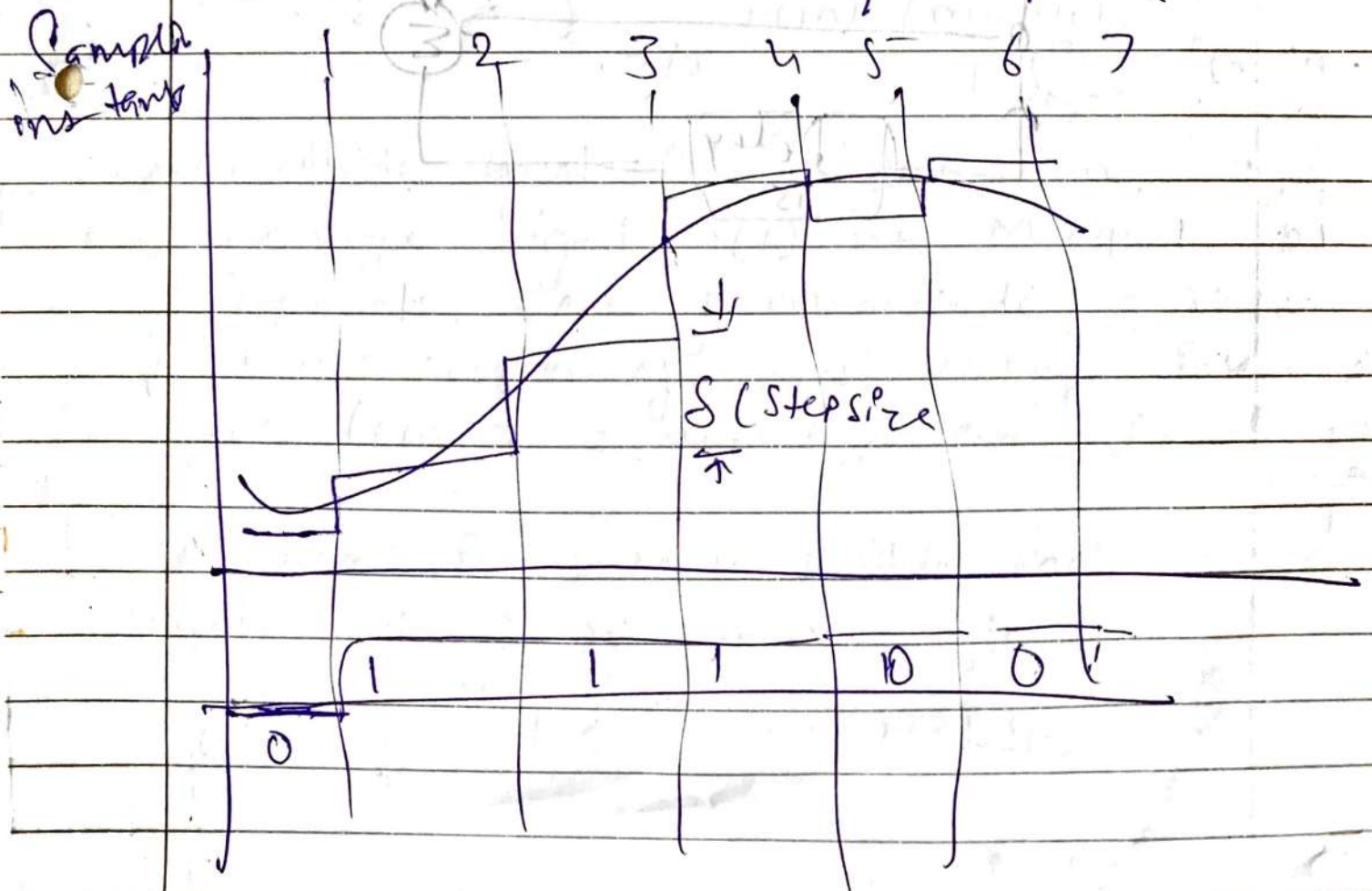
This helps in reducing redundancy

if  $f_s \geq \text{Nyquist rate}$

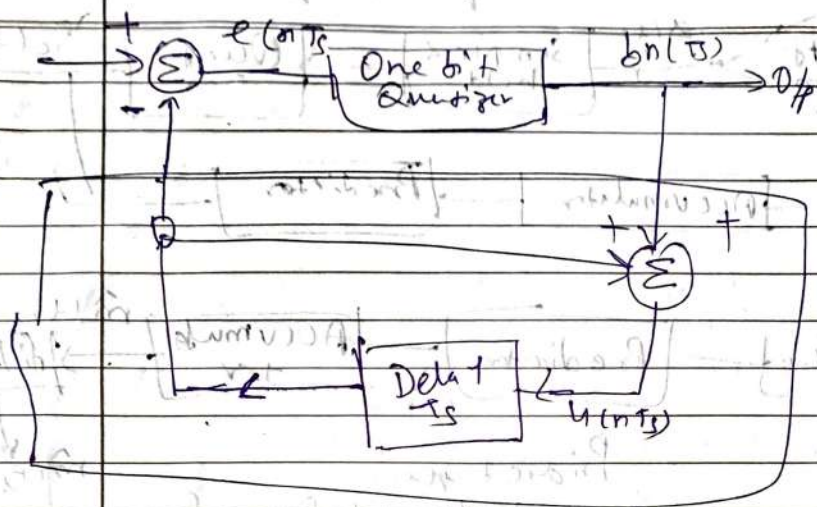
then this is possible

N No. of bits transmitted per sample

Delta Modu One bit / Sample

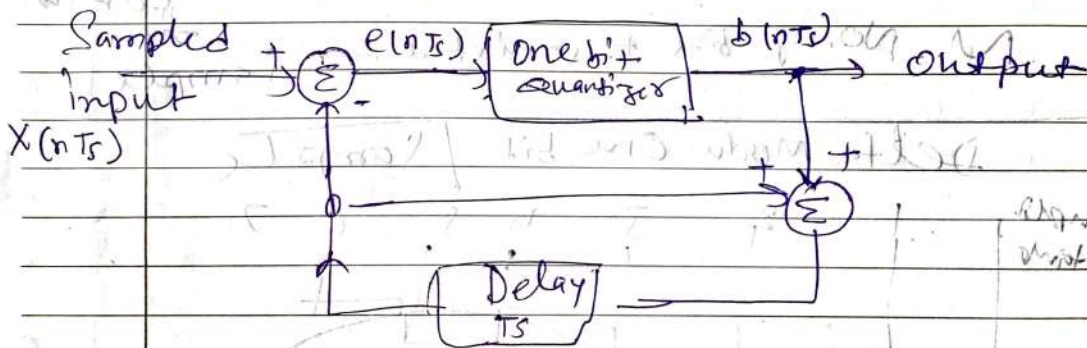


D.M.



Accumulator

D.M. Receiver





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Test Exam.

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Junior Supervisor's full  
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## Quantization Error.

Mean-Square quantization error  $\overline{e^2}$

$e$  = diff. bet<sup>n</sup> the Original and Quantized Signal Voltages.

Quantizer o/p =  $m_k$ .

$$e = m(t) - m_k$$

$m(t)$  = Message Signal.

Let  $f(m) dm$  be the probability that  $m(t)$  lies in the voltage range  $m - dm/2$  to  $m + dm/2$

then the mean-square quantization error is

$$\overline{e^2} = \int_{m_1 - S/2}^{m_1 + S/2} f(m) (m - m_1)^2 dm + \int_{m_2 - S/2}^{m_2 + S/2} f(m) (m - m_2)^2 dm + \dots$$

Let's divide total Peak to Peak range of the Message Signal  $m(t)$  into  $M$  equal voltage intervals each of magnitude  $S$  Volts at the center of each voltage interval we locate a quantization level  $m_1, m_2, \dots, m_M$

Let's  $f(m)$  is constant within each quantization level. then  $x = m - m_k$

$$\overline{e^2} = (f^{(1)} + f^{(2)} + \dots) \int_{-S/2}^{S/2} x^2 dx$$

$$f^{(1)}(S) \text{ is the probability that signal voltage will be in 1st quantization level}$$

$$\overline{e^2} = (f^{(1)}S + f^{(2)}S + \dots) \frac{S^3}{12} = \frac{S^2}{12} (f^{(1)} + f^{(2)} + \dots)$$



So sum of total probabilities

$$\left[ \overline{\sigma^2} = \frac{\sigma^2}{T_2} \right].$$

Companding -

Signal

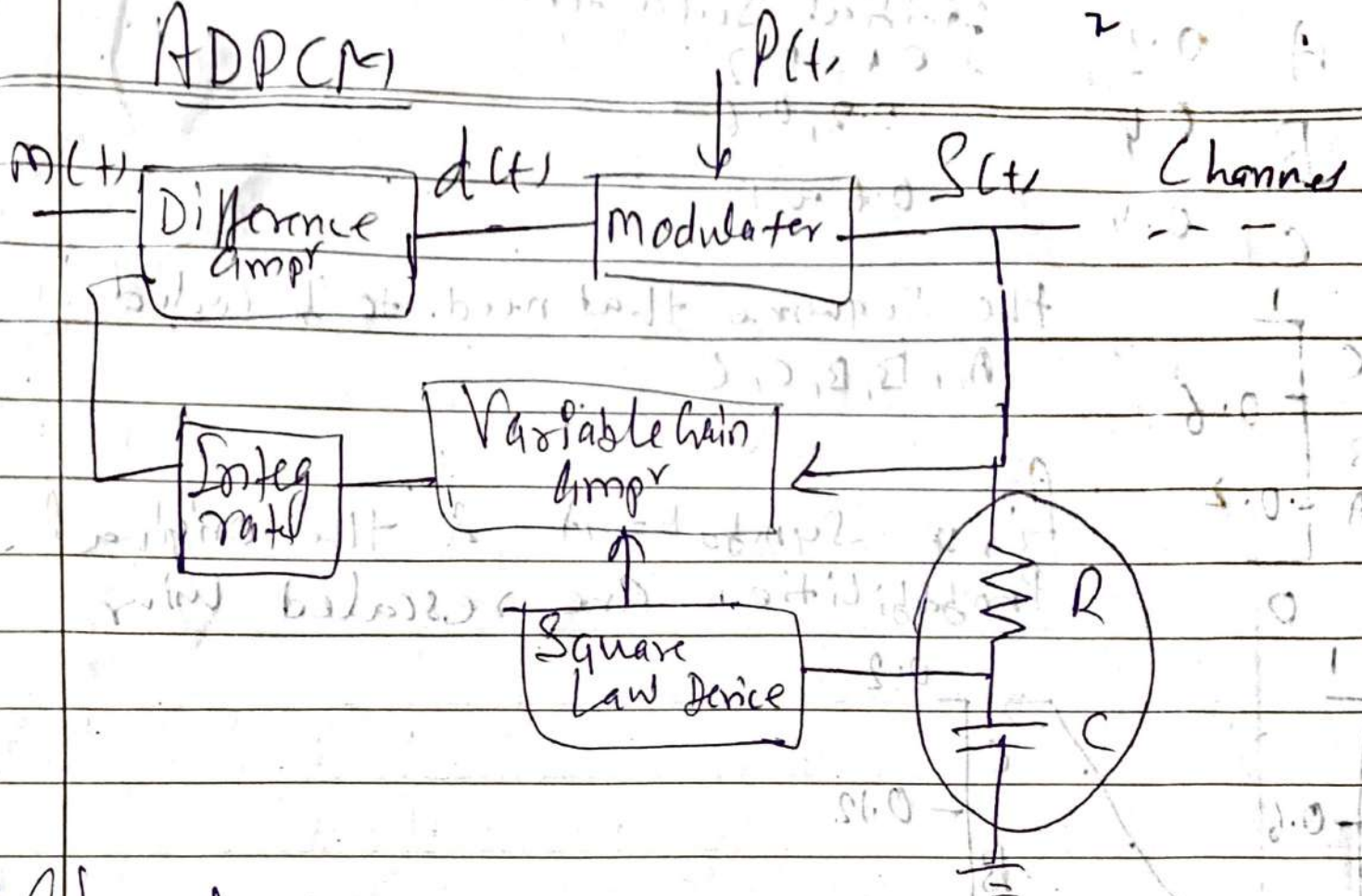
Quantization Noise

Quantization Error

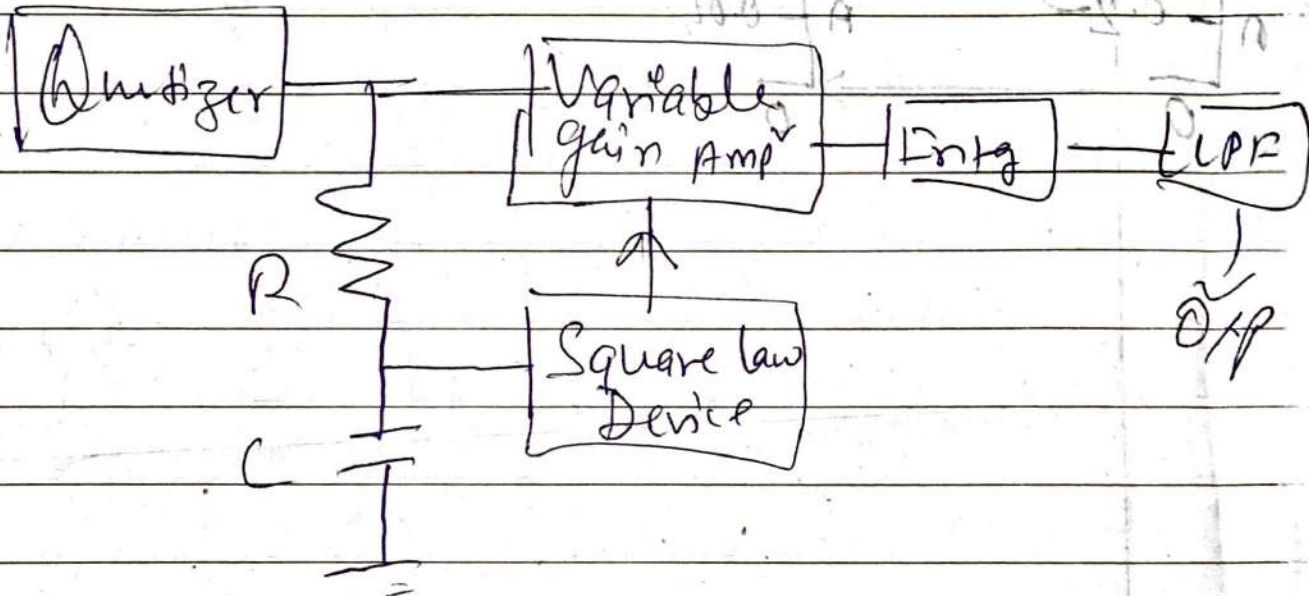
Mean-Square Quantization Error



# ADPCM



Channel



## Delta modulation

Sampling Signal

