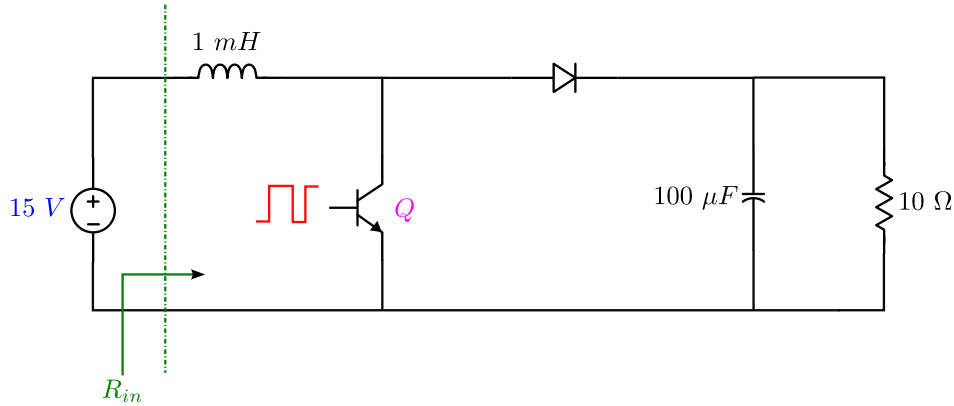
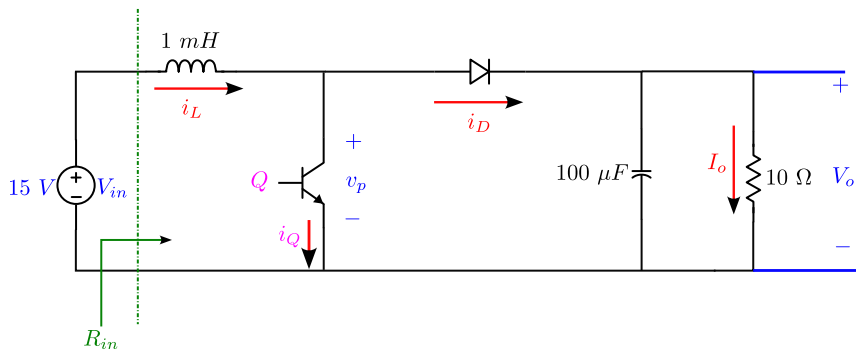


**Question 1: Numerical type**

Consider a boost converter shown in the circuit schematic. Switch  $Q$  is operating at 25 kHz with a duty cycle of 0.6. Assume the diode and switch to be ideal. Under steady-state conditions, the average resistance,  $R_{in}$ , as seen by the source is \_\_\_\_\_  $\Omega$ . (Round off to two decimal places)



**Solution: Answer Range (1.55 - 1.65)**



The output voltage of the boost converter can be calculated as

$$V_o = \frac{1}{1-D} \times V_{in} = \frac{15}{1-0.6} = 37.5 \text{ V}$$

Thus, the load current,  $I_o$ , is

$$I_o = \frac{V_o}{10} = 3.75 \text{ V}$$

By small ripple approximation we can derive

$$I_o = (1-D)I_L$$

Where,  $I_L$  is the average value of the inductor current,  $i_L$ . Thus, we can calculate  $I_L$  as

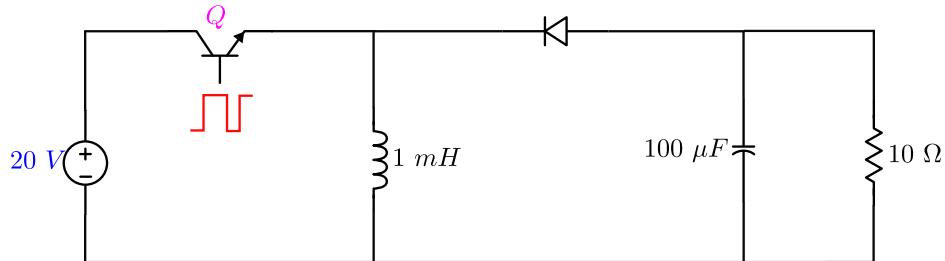
$$I_L = \frac{I_o}{(1-D)} = 9.375 \text{ A}$$

$I_L$  is the average current as seen by the source  $V_{in}$ . Thus, the average value of resistance as seen by the source is

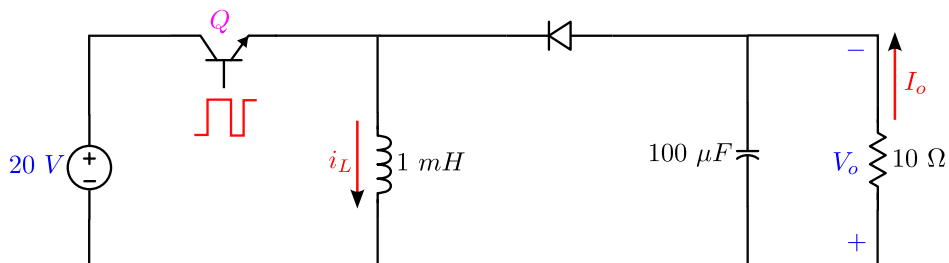
$$R_{in} = \frac{V_{in}}{I_L} = 1.6 \Omega$$

**Question 2: Numerical type**

Consider the buck-boost converter as shown in the circuit schematic below. Switch  $Q$  is operating at 25 kHz and 0.75 duty-cycle. Assume diode and switch to be ideal. Under steady-state conditions, the average current flowing through the inductor is \_\_\_\_\_ A.



**Solution: Answer range (24-24)**



The output voltage of a buck-boost converter is given by

$$V_o = \frac{D}{1-D} V_{in} = \frac{0.75}{1-0.75} \times 20 = 60 \text{ V}$$

The output current,  $I_o$ , can then be calculated as

$$I_o = \frac{V_o}{10} = 6 \text{ A}$$

Now, by small ripple approximation we can write

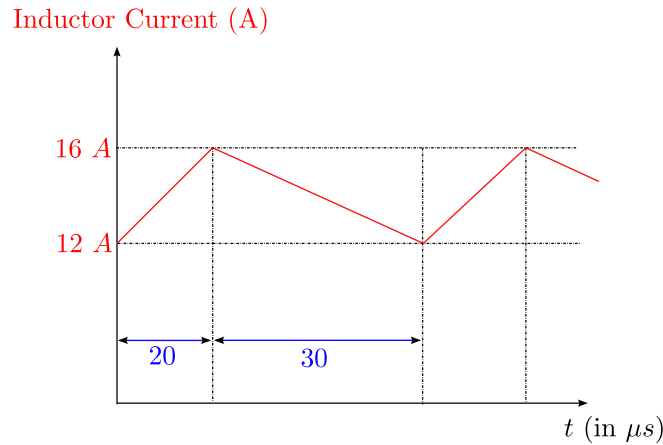
$$I_o = (1-D)I_L$$

Where,  $I_L$  is the average value of the inductor current  $i_L$ . Thus, we can calculate the average value of the inductor current as

$$I_L = \frac{I_o}{(1-D)} = 24 \text{ A}$$

**Question 3: Numerical type**

The steady-state current flowing through the inductor of a DC-DC buck-boost converter is shown below. If the peak-to-peak ripple in output voltage of converter is 1 V, then the value of the output capacitor, in  $\mu F$ , is \_\_\_\_\_ (round off to nearest integer).



**Solution: Answer range (167-169)**

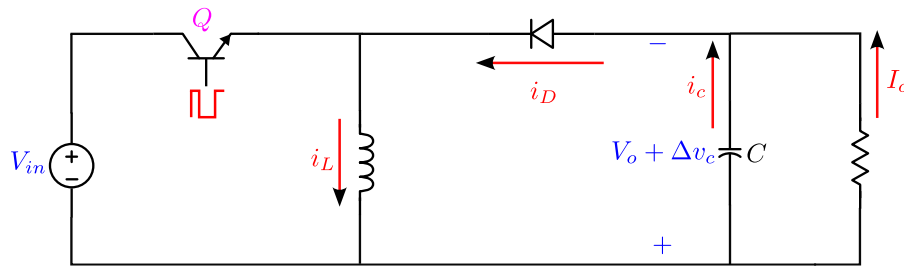


Fig. a

The waveform of the inductor current,  $i_L$ , diode current,  $i_D$ , and capacitor ripple current,  $i_c$  are shown in the Fig. b.

The average value of  $i_D$  is the load current  $I_o$  and is indicated by a dotted blue line in the Fig. b. Thus, we can calculate  $I_o$  as

$$I_o = \frac{1}{50} \left( 12 \times 30 + \frac{1}{2} \times 30 \times 4 \right) = 8.4 \text{ A}$$

The ripple current flowing through the capacitor can be obtained by subtracting the load current  $I_o$  from the diode current as shown in Fig. b. (Since net charge stored in the capacitor over a cycle is 0 Coulomb, or in other words, average current through the capacitor is 0 A). The charge stored in the capacitor ( $\Delta Q_c$ ) during the on time of the switch is indicated by the shaded yellow area in Fig. b.

We can determine ( $\Delta Q_c$ ) as

$$\Delta Q_c = C(\Delta v_c) = 8.4 \times 20 \mu C$$

Thus, output capacitor filter requirement is

$$C = \frac{8.4 \times 20}{\Delta v_c} = 168 \mu F$$

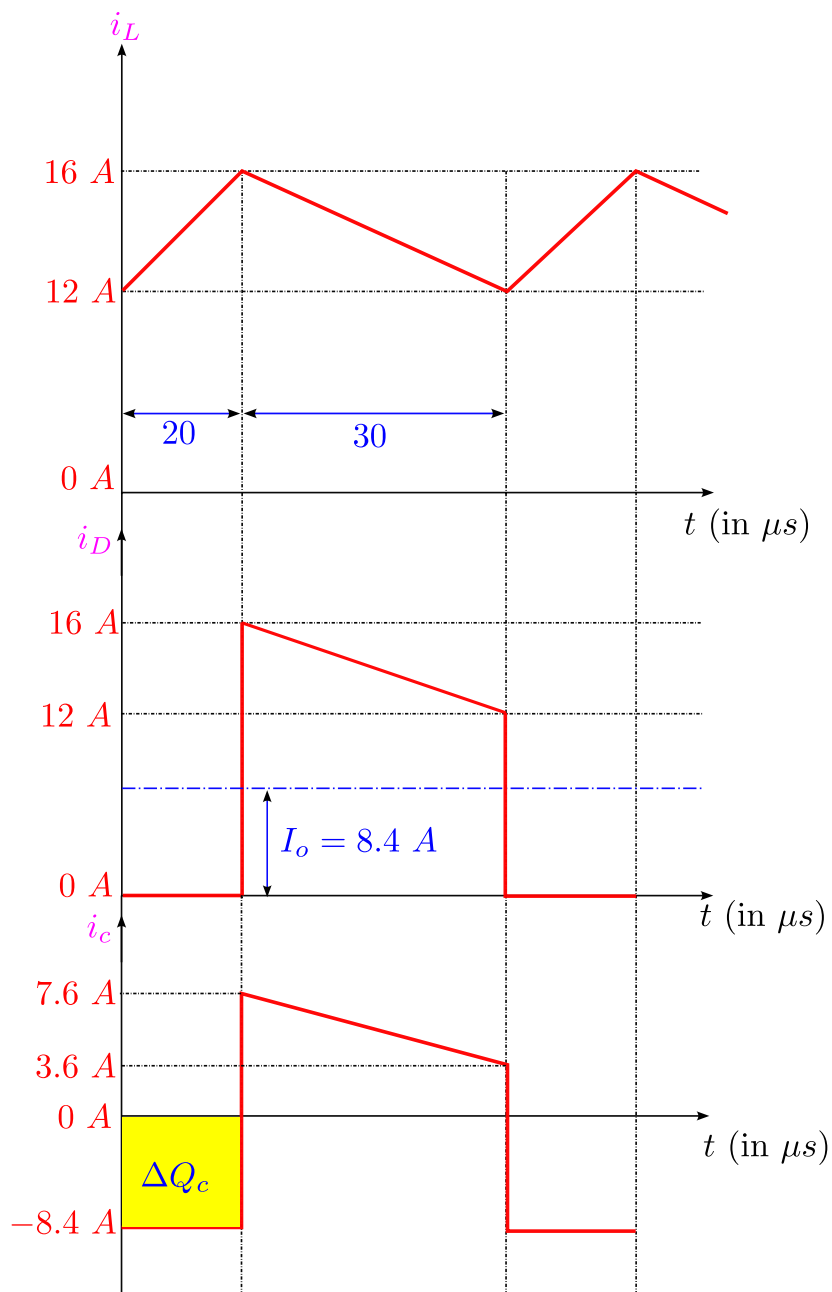
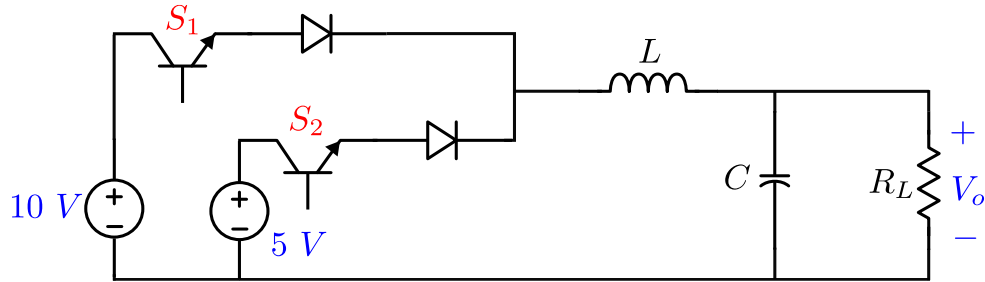


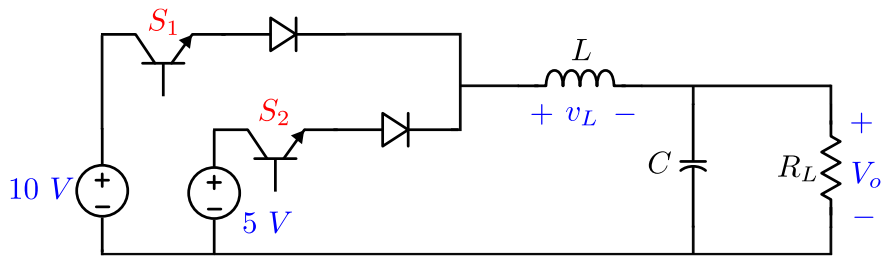
Fig. b

**Question 4: Numerical type**

The circuit shown in the figure is meant to supply a resistive load  $R_L$  from two separate DC voltage sources. The switches  $S_1$  and  $S_2$  are controlled so that only one of them is ON at any instant.  $S_1$  is turned on for 0.2 ms and  $S_2$  is turned on for 0.3 ms in a 0.5ms switching cycle time period. Assuming continuous conduction of the inductor current and negligible ripple on the capacitor voltage, the output voltage  $V_o$  (in volts) across  $R_L$  is \_\_\_\_\_.



**Solution: Answer range (6.99 - 7)**



During  $0 < t < 0.2 \text{ ms}$ , the switch  $S_1$  is on. Thus, the voltage across the inductor during this period is

$$v_L = 10 - V_o$$

During  $0.2 \text{ ms} < t < 0.5 \text{ ms}$ , the switch  $S_2$  is on. Thus, the voltage across the inductor during this period is

$$v_L = 5 - V_o$$

Now, applying volt-sec balance across the inductor we can write

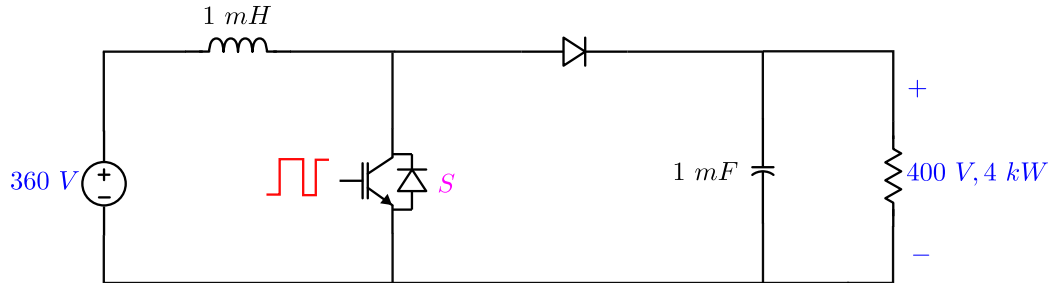
$$(10 - V_o) \times 0.2 + (5 - V_o) \times 0.3 = 0$$

Solving the above equation, we get

$$V_o = 7 \text{ V}$$

**Question 5: Numerical type**

A DC-DC boost converter as shown in the figure below is used to boost 360 V to 400 V at a power of 4kW. All devices are ideal considering continuous inductor current, the rms current in the solid-state switch (S), (in Amperes) is \_\_\_\_\_. (Neglect the ripple in Inductor Current)



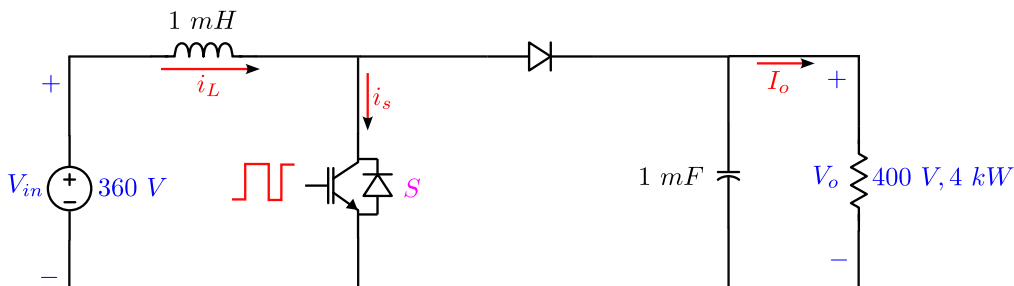
**Solution: Answer range (3.0 – 4.0)**

The output power is given as

$$P_o = V_o I_o = 4 \times 10^3 \text{ W}$$

From the input and output voltages of the boost converter we can calculate the duty ratio,  $D$ , as

$$D = 1 - \frac{V_{in}}{V_o} = 0.1$$



The average value of the load current can then be calculated as

$$I_o = \frac{P_o}{V_o} = \frac{4 \times 10^3}{4 \times 10^2} = 10 \text{ A}$$

Now, as the semiconductor switches are considered ideal, we can neglect the losses in these devices. Thus, we can equate the input power to output power as shown below

$$P_{in} = P_o = 4 \times 10^3 \text{ W}$$

The average value of inductor current can be written as

$$I_L = \langle i_L \rangle_{avg} = \frac{P_{in}}{V_{in}} = 11.11 \text{ A}$$

The rms value of the current flowing through the switch 'S' can be calculated as follows

$I_{S,rms} = \sqrt{D} I_L = 3.51 \text{ A}$  (Neglecting Ripple in Inductor current. You can draw the waveform for the switch current and then calculate rms value and you will get the same expression)

*Question 6:*

*A chopper is employed to charge a battery as shown in the figure below. The charging current is 5 A. The duty ratio is 0.2. The chopper output voltage is also shown in figure. The peak-to-peak ripple current in the charging current is*

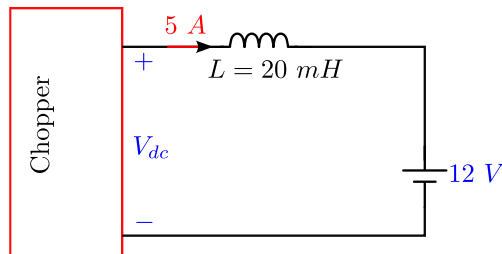


Fig. (a)

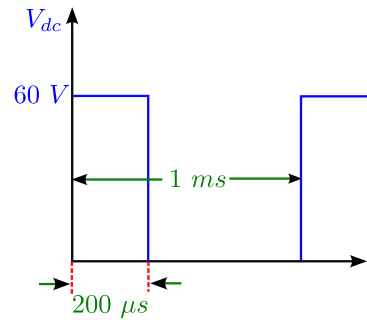


Fig. (b)

- (a)  $0.48 \text{ A}$
- (b)  $1.2 \text{ A}$
- (c)  $0.24 \text{ A}$
- (d)  $0.96 \text{ A}$

Solution: Correct option is (a)

For the time interval  $0 < t < 200 \mu\text{s}$ , the voltage across the inductor can be calculated as

$$v_L = 60 - 12 = L \frac{\Delta i_L}{\Delta t}$$

Therefore, the ripple current can be calculated as

$$\Delta i_L = \frac{48}{L} \times \Delta t = \frac{48}{20 \times 10^{-3}} \times 200 \times 10^{-6} = 0.48 \text{ A}$$

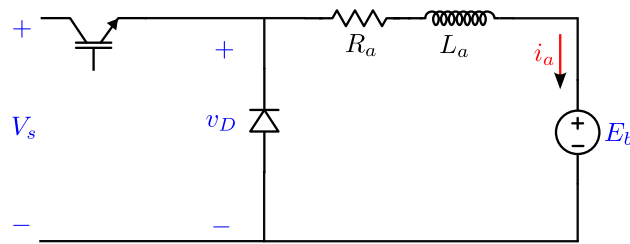
**Question 7:**

A step-down chopper operates from a dc voltage source,  $V_s$ , feeds a dc motor armature with a back emf of  $E_b$ . From oscilloscope traces, it is found that the armature current increases for time,  $t_r$ , falls to zero over the time,  $t_f$ , and remains zero for the time,  $t_o$ , in every chopping cycle, then the average dc voltage across the freewheeling diode is: [Assume Diode forward drop to be 0V]

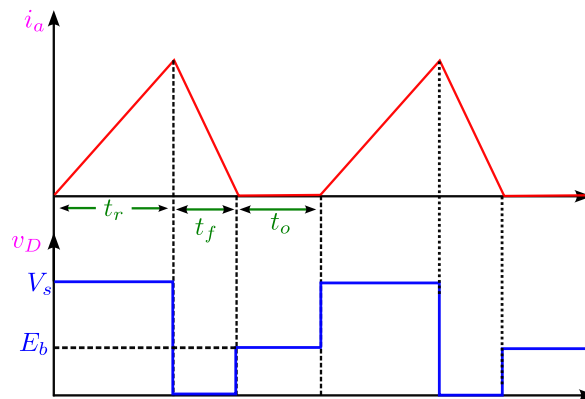
- (a)  $\frac{V_s t_r}{t_r + t_f + t_o}$
- (b)  $\frac{V_s t_r + E_b t_f}{t_r + t_f + t_o}$
- (c)  $\frac{V_s t_r + E_b t_o}{t_r + t_f + t_o}$
- (d)  $\frac{V_s t_r + E_b (t_f + t_o)}{t_r + t_f + t_o}$

**Solution:** Correct option is (c)

A dc motor armature is excited from a step-down chopper. The equivalent circuit is shown in the schematic below



The waveforms of the armature current,  $i_a$ , and the diode voltage,  $v_D$ , is shown in the figure below:



The average value of voltage across the diode can be calculated from the waveform of  $v_D$  as

$$V_D = \langle v_D \rangle_T = \frac{V_s t_r + E_b t_o}{t_r + t_f + t_o}$$



**Question 8: Numerical type**

The semiconductor switch shown in the Fig. (a) is operated at a frequency of 20 kHz and duty ratio of 0.5. The circuit operates in steady state,  $V_1 = 100\text{ V}$  and  $V_2 = 275\text{ V}$ . The diagram of inductor current  $i_L$  is shown in Fig. (b). The value of  $\beta T$  is \_\_\_\_\_  $\mu\text{s}$ .

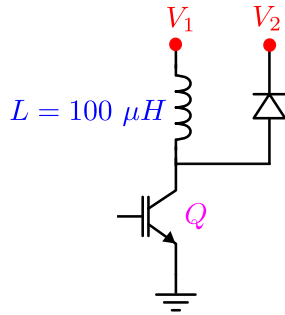


Fig. (a)

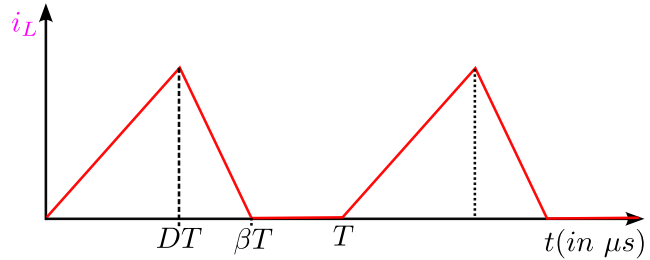
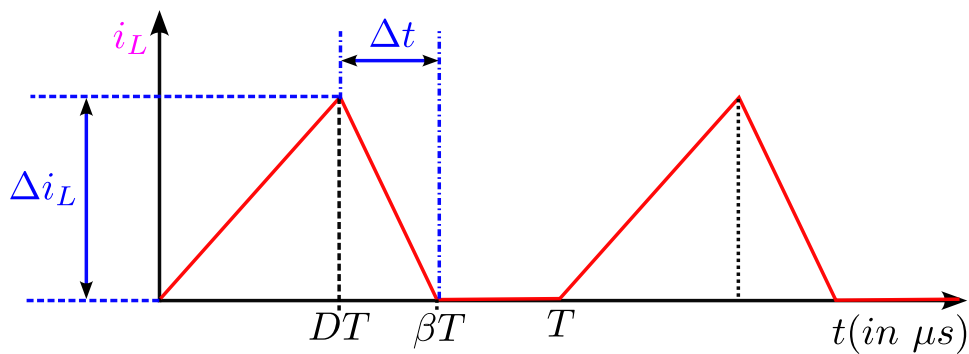


Fig. (b)

**Solution: Answer range (38 - 40)**



When the switch  $Q$  is on the voltage across the inductor can be written as

$$v_L = V_1 = 100 = L \times \frac{\Delta i_L}{DT}$$

Thus, the ripple current in the inductor,  $\Delta i_L$ , can be evaluated as

$$\Delta i_L = \frac{100}{L} \times DT = 25\text{ A}$$

Now, as the switch  $Q$  turns off, the diode starts conducting and the voltage across the inductor becomes

$$v_L = V_1 - V_2 = -175 = -L \times \frac{\Delta i_L}{\Delta t}$$

Negative sign is taken as the slope is negative. Now,  $\Delta t$  can be calculated as

$$\Delta t = \frac{L \times \Delta i_L}{175} = 14.29\text{ } \mu\text{s}$$

Now, we have

$$\beta T = \Delta t + DT = 39.28\text{ } \mu\text{s}$$

**Question 9: Numerical type**

*In continuation with previous question, if  $V_2 = 300\text{ V}$ , what is the power transferred (in Watts) from the dc voltage source  $V_1$  to dc voltage source  $V_2$ . Assume the controlled and uncontrolled switches to be ideal.*

**Solution: Answer range (936-938)**

In the previous question we have found that

$$\Delta i_L = \frac{100}{L} \times DT = 25\text{ A}$$

Now, when the switch  $Q$  is turned off,

$$v_L = V_1 - V_2 = -200 = -L \times \frac{\Delta i_L}{\Delta t}$$

Thus,  $\Delta t$  can be calculated as

$$\Delta t = \frac{L \times \Delta i_L}{200} = 12.5\text{ }\mu\text{s}$$

The switch is switched at 20 kHz. Thus, the time-period ( $T$ ) can be calculated as

$$T = \frac{1}{20 \times 10^3} = 50\text{ }\mu\text{s}$$

Thus, the average value of the inductor current can be calculated as

$$I_L = \langle i_L \rangle_{avg} = \frac{1}{50} \left( \frac{1}{2} \times (DT + \Delta t) \times \Delta i_L \right) = 9.375\text{ A}$$

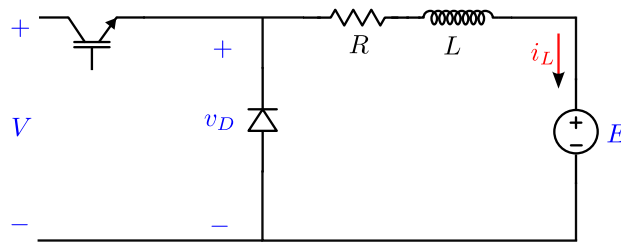
Thus, power transferred from voltage source  $V_1$  to  $V_2$  can be calculated as

$$P = I_L \times V_1 = 9.375 \times 100 = 937.5\text{ W}$$

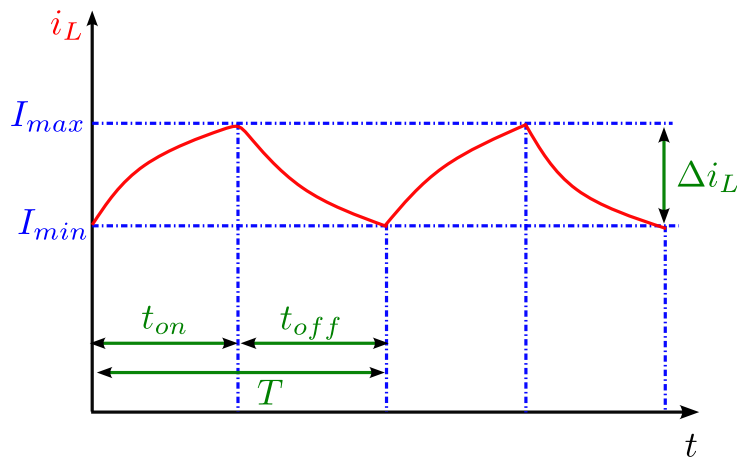
Question 10:

A DC chopper feeds an RLE load as shown in the figure below. The value of  $E$  is increased by 20%, the current ripple

- (a) Increases by 20%
- (b) Decreases by 20%
- (c) Increases by 10%
- (d) Remains same



Solution: Correct option is (d)



The expression for  $I_{max}$  can be derived as

$$I_{max} = \frac{V}{R} \left[ \frac{1 - e^{-t_{on}/t_a}}{1 - e^{-T/t_a}} \right] - \frac{E}{R}$$

Where  $t_a = \frac{L}{R}$  is the time constant of the circuit.

The expression for  $I_{min}$  can be derived as

$$I_{min} = \frac{V}{R} \left[ \frac{e^{t_{on}/t_a} - 1}{e^{T/t_a} - 1} \right] - \frac{E}{R}$$

Thus, the inductor current ripple can be calculated as

$$\Delta i_L = I_{max} - I_{min} = \frac{V}{R} \left[ \frac{1 - e^{-\frac{t_{on}}{t_a}}}{1 - e^{-\frac{T}{t_a}}} \right] - \frac{V}{R} \left[ \frac{e^{\frac{t_{on}}{t_a}} - 1}{e^{\frac{T}{t_a}} - 1} \right]$$

**Thus, the ripple current is independent of 'E'.**

[Hint for Derivation: For  $T_{ON}$  and  $T_{OFF}$  draw the 2 circuits. First consider the  $T_{ON}$  circuit (which is an RL circuit with 2 sources  $V$  and  $E$ ), solve for  $i_L$ , with initial condition as  $I_{min}$ . In the expression

obtained for  $i_L$  substitute  $t=t_a$  to obtain  $I_{max}$ . Then consider the  $T_{OFF}$  circuit (which is an RL circuit with just  $E$  as source. Ignore diode drop), solve for  $i_L$ , with initial condition as  $I_{max}$ . In the expression obtained for  $i_L$ , substitute  $t=T$  to obtain  $I_{min}$ . Then you can solve for  $I_{max}$  and  $I_{min}$ . ]