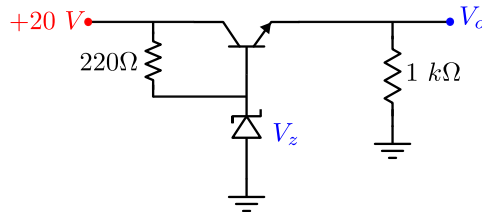
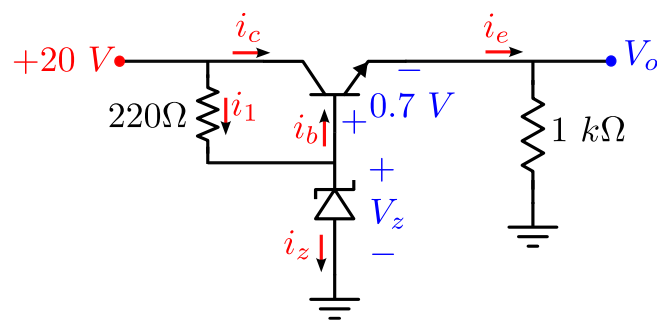


**Question 1: Numerical type**

In the regulator circuit shown below  $V_Z = 12\text{ V}$ ,  $\beta = 50$ ,  $V_{BE} = 0.7\text{ V}$ . The Zener current (in mA) is \_\_\_\_\_.



**Solution: Answer range (36.0 - 36.2)**



Applying KVL we can obtain the value of the output voltage as

$$V_o = V_Z - 0.7 = 11.3\text{ V}$$

Now, the emitter current can be calculated as

$$i_e = \frac{V_o}{1k} = 11.3\text{ mA}$$

As the BJT is operating in linear region and hence the base current can be calculated as

$$i_b = \frac{i_e}{1 + \beta} = 0.22\text{ mA}$$

We can calculate  $i_1$  as

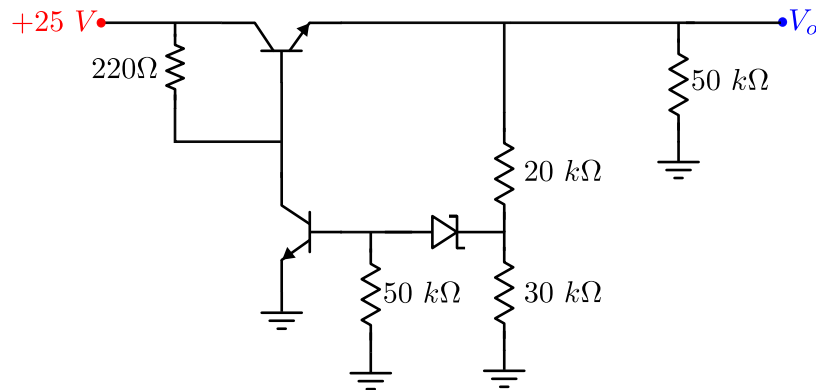
$$i_1 = \frac{20 - V_Z}{0.22k} = \frac{20 - 12}{0.22} = 36.36\text{ mA}$$

Now applying KCL we can write

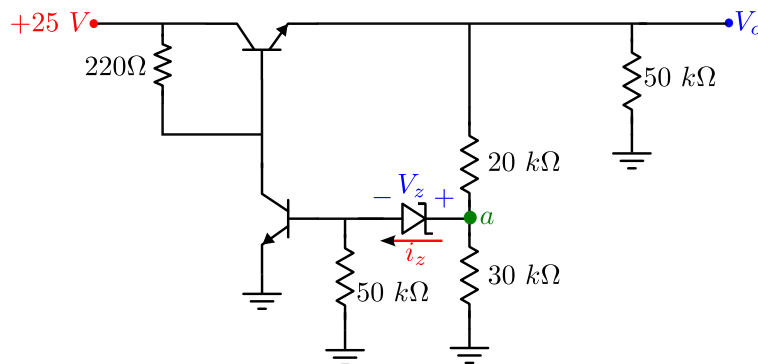
$$i_z = i_1 - i_b = 36.14\text{ mA}$$

**Question 2: Numerical type**

In the series voltage regulator circuit shown below assume the Zener diode to be ideal. Given that:  $V_Z = 8.3\text{ V}$ ,  $\beta = 50$ ,  $V_{BE} = 0.7\text{ V}$ , What would be the output voltage,  $V_o$ , (in Volts) \_\_\_\_\_.



**Solution: Answer range (14.8-15.2)**



In this problem, we must assume first whether the transistors are ON or OFF and whether the Zener is in reverse breakdown region or not. Based on the assumption, we solve the circuit to see whether our assumptions are right or wrong. Based on this method, we have found out that both the transistors are ON, and the Zener is in reverse breakdown. **Please note that “ON” here doesn’t mean it is in saturation. It means that the transistors are in linear region, and it basically behaves like a resistor. For series regulator circuits, normally the transistors operate in linear region.**

The potential at the point ‘a’ can be calculated as

$$V_a = V_{BE} + V_Z = 9\text{ V}$$

As Zener diode is assumed to ideal, then we have

$$i_Z \approx 0\text{ A}$$

Now, applying potential division we have

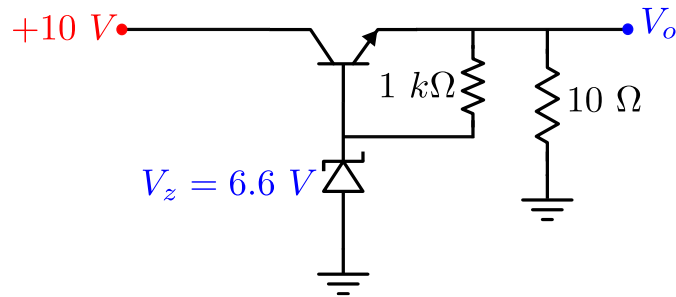
$$V_a = \frac{30}{30 + 20} V_o = \frac{3}{5} V_o$$

Thus, the output voltage  $V_o$  can be calculated as

$$V_o = \frac{5}{3} V_a = 15\text{ V}$$

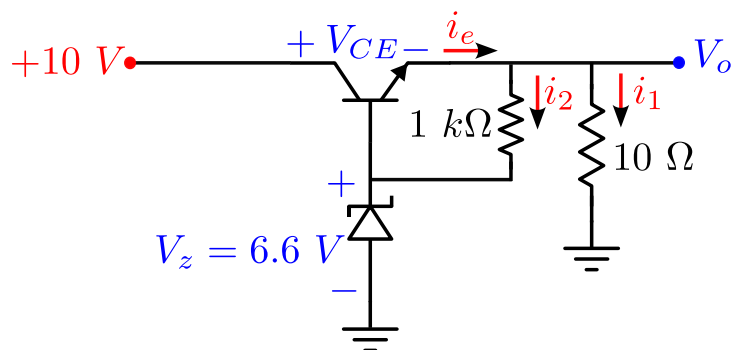
**Question 3:**

The three terminal linear voltage regulator is connected to a  $10\ \Omega$  load resistor as shown in the figure below. If  $V_{in}$  is  $10\text{ V}$ , what is the power dissipated in the transistor. (Assume  $V_{BE} = 0.6\text{ V}$ )



- (a)  $0.6\text{ W}$
- (b)  $4.2\text{ W}$
- (c)  $2.4\text{ W}$
- (d)  $5.4\text{ W}$

Solution: Correct option is (c)



The output voltage of the linear voltage regulator can be calculated as

$$V_o = V_Z - V_{BE} = 6.6 - 0.6 = 6\text{ V}$$

$i_1$  can be calculated as follows

$$i_1 = \frac{V_o}{10} = 0.6\text{ A}$$

$i_2$  can be calculated as follows

$$i_2 = \frac{V_o - V_Z}{1k} = -0.6\text{ mA}$$

Applying KCL we get the value of emitter current,  $i_e$ , as

$$i_e = i_1 + i_2 = 0.599\text{ A}$$

Now, the emitter-collector voltage can be calculated as

$$V_{CE} = 10 - V_o = 10 - 6 = 4 \text{ V}$$

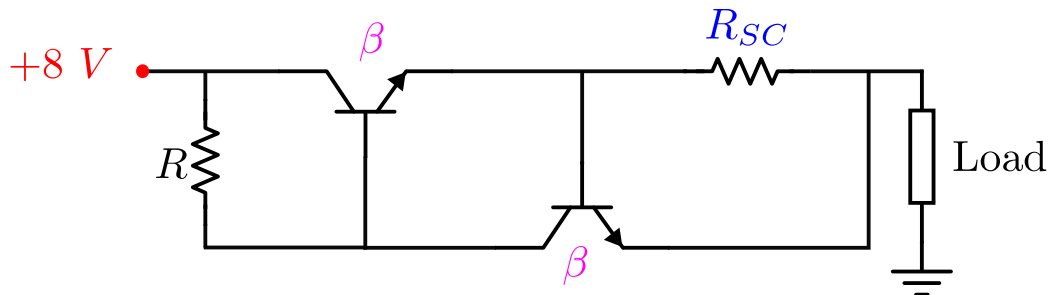
Thus, the power dissipated in the transistor can now be calculated as

$$P = V_{CE} \times i_e = 2.39 \text{ W}$$

Question 4:

A 5V regulator uses a 12 W series pass transistor. Find  $R_{SC}$  for output short circuit protection.

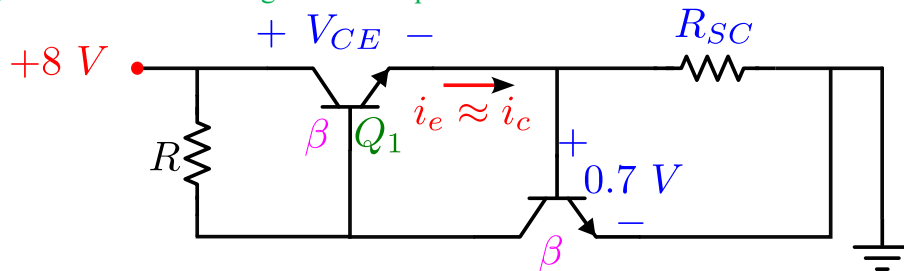
(Assume  $V_{BE} = 0.7 \text{ V}$ ,  $\beta \rightarrow \infty$ )



- (a)  $0.13 \Omega$
- (b)  $0.23 \Omega$
- (c)  $0.33 \Omega$
- (d)  $0.43 \Omega$

Solution: Correct option is (d)

If the load gets shorted the circuit given in the question becomes as shown below



The transistor is rated for 12 W. Thus,  $R_{SC}$  should be chosen such that the current flowing through the transistor  $Q_1$  during short circuit conditions does not cause a power dissipation more than the rated power of  $Q_1$ .

The collector to emitter voltage can be calculated as

$$V_{CE} = 8 - 0.7 = 7.3 \text{ V}$$

The rated power of transistor  $Q_1$  is 12 W. Thus, the maximum permissible current through  $Q_1$  can be calculated as

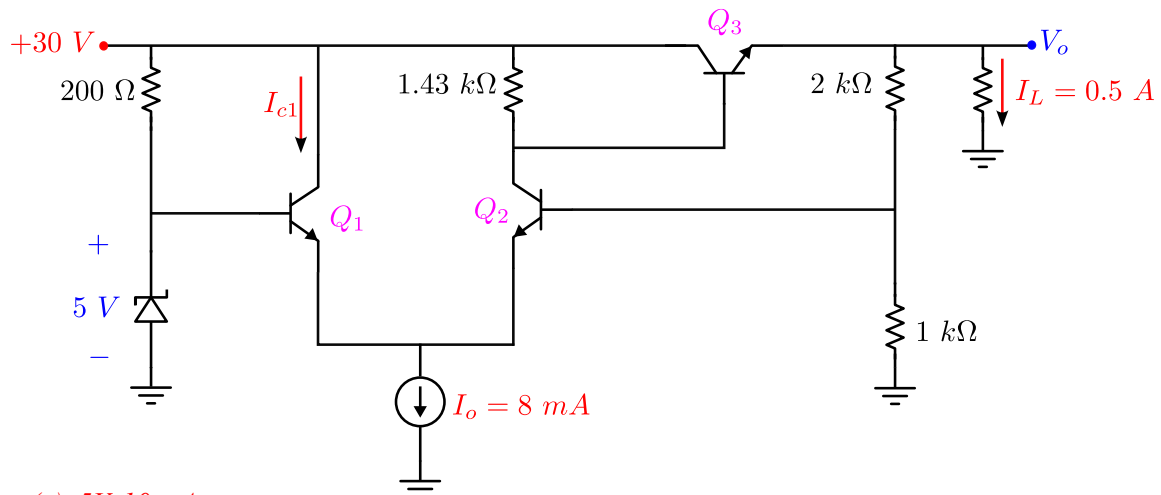
$$i_{c,max} \approx i_{e,max} = \frac{12}{V_{CE}} = 1.64 \text{ A}$$

As  $\beta \rightarrow \infty$ , the base current of the transistors can be neglected, and we can assume that  $i_{e,max}$  is flowing through  $R_{SC}$ . Thus,  $R_{SC}$  can be calculated as

$$R_{sc} = \frac{0.7}{1.64} \approx 0.43 \Omega$$

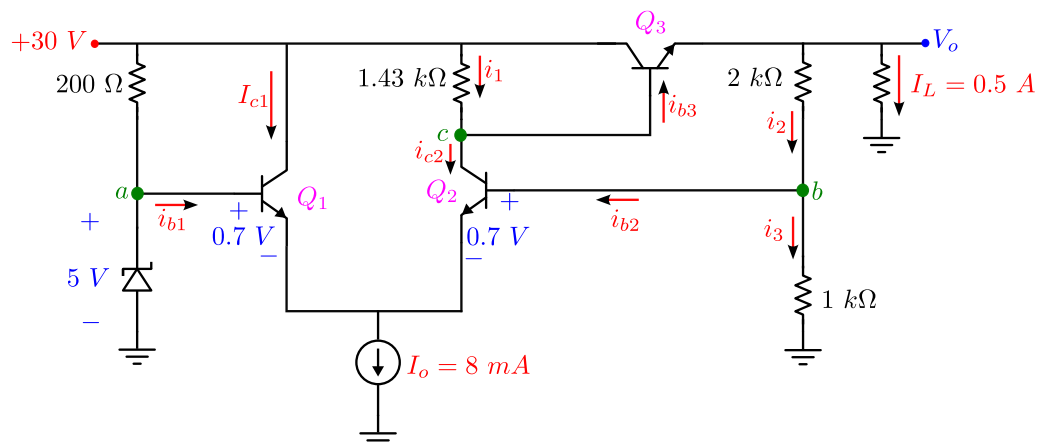
**Question 5:**

In the circuit diagram given below assume that  $Q_1$  and  $Q_2$  has infinite current gain (i.e.  $\beta_1 \approx \beta_2 \rightarrow \infty$ ) The current gain of transistor  $Q_3$ ,  $\beta_3 = 99$ . Calculate  $V_o$  (in V),  $I_{c1}$  (in mA) if  $I_L = 0.5$  A. (Assume  $V_{BE} = 0.7$  V)



- (a) 5V, 10 mA
- (b) 5V, 4.95 mA
- (c) 15V, 4.95 mA
- (d) 15V, 3.05 mA

**Solution:** Correct option is (d)



From the above circuit diagram, we can conclude that

$$V_a = V_b = 5 \text{ V}$$

As  $\beta_2 \rightarrow \infty$  and  $\beta_1 \rightarrow \infty$ , we have

$$i_{b1} = i_{b2} = 0$$

Thus, applying potential divider we can write

$$V_b = \frac{1}{3} V_o$$

Thus, the output voltage,  $V_o$ , can be calculated as

$$V_o = 3 \times V_b = 15 \text{ V}$$

The potential at point 'c' can be calculated as

$$V_c = V_o + 0.7 = 15.7 \text{ V}$$

Thus,  $i_1$  can be calculated as

$$i_1 = \frac{30 - V_c}{1.43} \text{ mA} = 10 \text{ mA}$$

Since  $i_{b2} = 0$ , the current flowing through  $2 \text{ k}\Omega$  can be calculated as

$$i_2 = \frac{V_o}{3} = 5 \text{ mA}$$

Applying KCL, the emitter current of  $Q_3$  can be calculated as

$$i_{e3} = I_L + i_2 = 505 \text{ mA}$$

Thus, the base current of  $Q_3$  is

$$i_{b3} = \frac{i_{e3}}{1 + \beta_3} = 5.05 \text{ mA}$$

Thus, the collector current of  $Q_2$  is

$$i_{c2} = i_1 - i_{b3} = 4.95 \text{ mA}$$

Since  $i_{b2} = 0$ , we can write

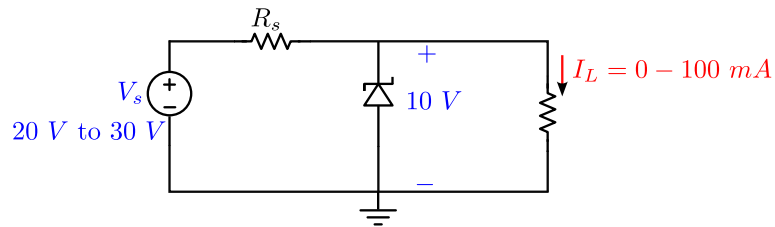
$$i_{c2} = i_{e2} = 4.95 \text{ mA}$$

Thus, the collector current of  $Q_1$  is

$$I_{c1} = I_0 - i_{e2} = 8 - 4.95 = 3.05 \text{ mA}$$

**Question 6: Numerical type**

Find the value of  $R_s$  (in  $\Omega$ ) as a worst-case design, if  $V_s$  ranges from 20 V to 30 V and  $I_L$  ranges from 0 to 100 mA and  $I_{Z(knee)} = 1$  mA.



**Solution: Answer range (98.5-99.5)**

$I_{Z(knee)}$  is the minimum current that must flow through the Zener diode so that it can maintain a voltage of  $V_Z = 10$  V.

Worst case condition would be:

$$V_s = 20 \text{ V} \text{ and } I_L = 100 \text{ mA}$$

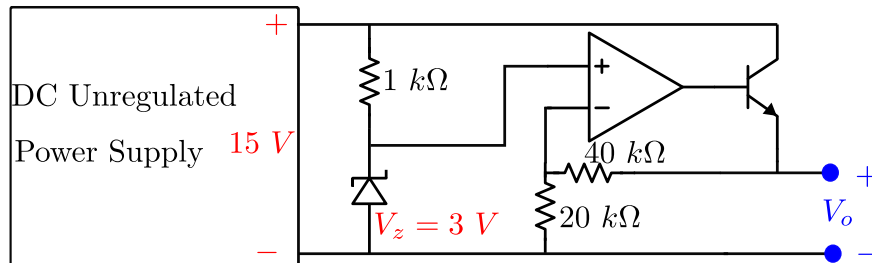
Because it would be the case when the voltage drop across  $R_s$  would be maximum. The value of  $R_s$  can then be calculated as

$$R_s = \frac{20 - 10}{I_{Z(knee)} + I_L} = \frac{10}{101} \times 10^3 \Omega = 99 \Omega$$

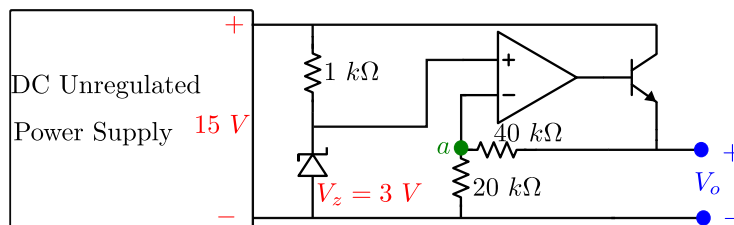


**Question 7: Numerical type**

Find the output voltage,  $V_o$  (in volts), of the regulated power supply shown in the figure. Assume the op-amp to be ideal.



**Solution: Answer range (8.9-9)**



As the op-Amp is ideal, the gain of op-Amp is infinite. And hence due to virtual grounding

$$V_a = V_z = 3 V$$

Again, we have

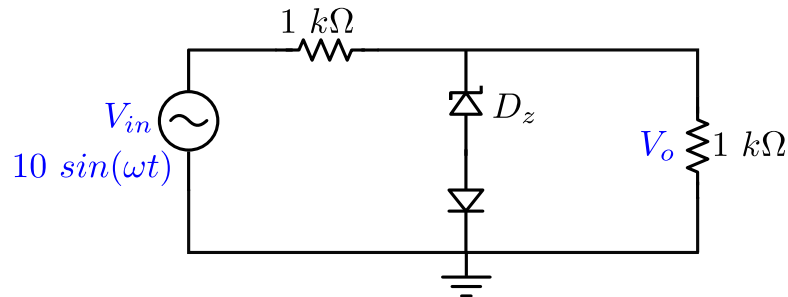
$$V_a = \frac{20}{20 + 40} V_o = \frac{1}{3} V_o$$

The output voltage of regulated power supply is

$$V_o = 3V_a = 9 V$$

**Question 8:**

The cut in voltage of both Zener diode  $D_z$  and  $D$  shown in the figure below is  $0.7\text{ V}$  while breakdown voltage of the Zener is  $3.3\text{ V}$  and reverse break down of  $D$  is  $50\text{ V}$ . The other parameters can be assumed to be the same as those of an ideal diode. The values of the peak output voltage ( $V_o$ ) are



- (a)  $3.3\text{ V}$  in the positive half cycle and  $1.4\text{ V}$  in the negative half cycle
- (b)  $4\text{ V}$  in the positive half cycle and  $5\text{ V}$  in the negative half cycle
- (c)  $3.3\text{ V}$  in both positive and negative half cycle
- (d)  $5\text{ V}$  in both positive and negative half cycle

Solution: Correct option is (b)

In the positive half cycle,

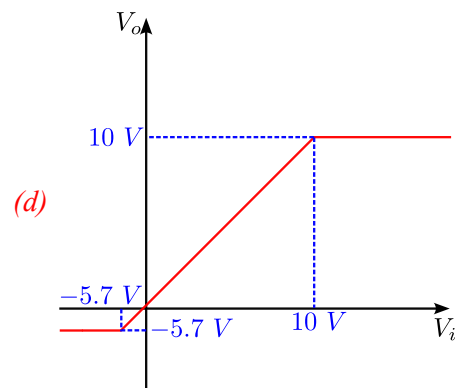
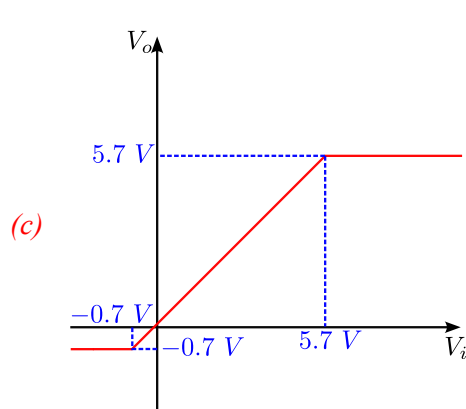
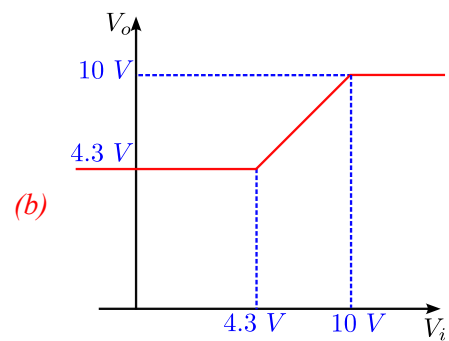
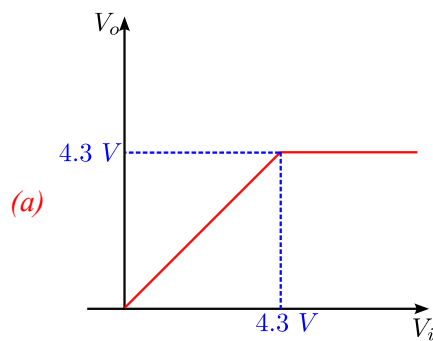
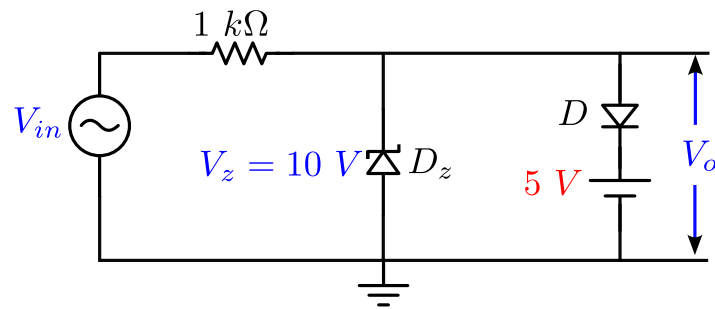
When  $V_o$  exceeds  $(V_z + 0.7\text{ V} = 4\text{ V})$ , it gets clamped to  $4\text{ V}$

In the negative half cycle,

The diode  $D$  remains reverse biased and hence  $V_o = \frac{V_{in}}{2}$ . Hence, the peak value of  $V_o$  in the negative half cycle is  $5\text{ V}$ .

**Question 9:**

Assuming forward voltage drop of the diodes to be  $0.7\text{ V}$ , the input-output transfer characteristics of the circuit is



Solution: Correct option is (c)

When  $-0.7\text{ V} \leq V_{in} \leq 5.7\text{ V}$ ,

Diode  $D$  remains reverse biased. The Zener diode acts as open circuit as  $V_{in}$  has not exceeded its breakdown voltage ( $V_z = 10\text{ V}$ ). Thus. During this period  $V_o = V_{in}$

When  $V_{in} > 5.7\text{ V}$ ,

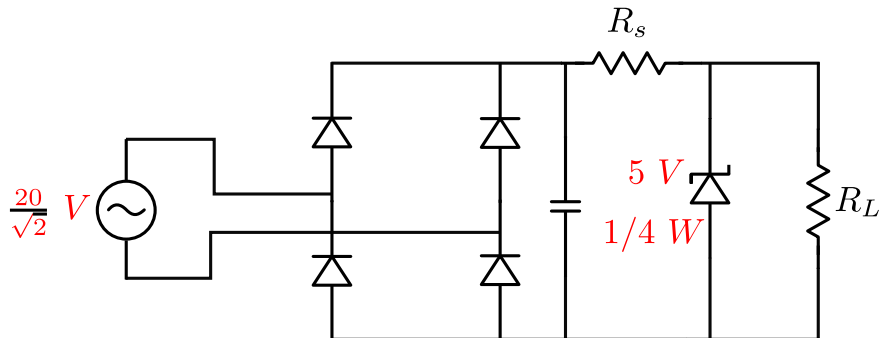
The output voltage gets clamped at  $5.7\text{ V}$ .

When  $V_{in} \leq -0.7\text{ V}$ ,

The Zener diode starts acting as a forward biased diode and hence the output voltage gets clamped at  $-0.7\text{ V}$ .

Question 10: Numerical Type

The sinusoidal ac source in the figure has rms value of  $\frac{20}{\sqrt{2}}$  V. Considering all possible values of  $R_L$ , the minimum value of  $R_s$  in  $\Omega$  to avoid burnout of the Zener diode is \_\_\_\_\_.



Solution: Answer range (299-300)

If  $R_L \rightarrow \infty$  (i.e. open circuit condition),

The load current would flow through the Zener diode. Thus, we should ensure that the power dissipation during this condition does not exceed the rated value of 0.25 W. Therefore,

$$i_{Z,max} = \frac{0.25}{5} = 0.05 \text{ A}$$

Considering the worst-case condition, the voltage across the capacitor will be almost close to the peak value of the source voltage. Thus, the value of resistance  $R_s$  should be

$$R_s = \frac{20 - 5}{0.05} = 300 \Omega$$