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$$R = \begin{bmatrix} 1 & 0.5 \\ 0.5 & 1 \end{bmatrix}$$

$$P = \begin{bmatrix} 0.5 \\ 0.25 \end{bmatrix}$$

a) from the Wiener-Hopf equation, we have — (1)

$$W = R^{-1}P$$

Inverse of  $R$  is

$$R^{-1} = \begin{bmatrix} 1 & 0.5 \\ 0.5 & 1 \end{bmatrix}^{-1}$$

$$= \frac{1}{(1-0.5^2)} \begin{bmatrix} 1 & -0.5 \\ -0.5 & 1 \end{bmatrix}$$

Using equation (1),

$$W = \frac{1}{0.75} \begin{bmatrix} 1 & -0.5 \\ -0.5 & 1 \end{bmatrix} \begin{bmatrix} 0.5 \\ 0.25 \end{bmatrix}$$

$$= \frac{1}{0.75} \begin{bmatrix} 0.275 \\ 0 \end{bmatrix} = \begin{bmatrix} 0.5 \\ 0 \end{bmatrix}$$

(b) The minimum mean-square error is

$$\begin{aligned} J_{\min} &= \sigma_d^2 - \rho^2 \omega \\ &= \sigma_d^2 - [0.5 \ 0.25] \begin{bmatrix} 0.5 \\ 0 \end{bmatrix} \\ &= \sigma_d^2 - 0.25 \end{aligned}$$

$$\text{If } \sigma_d^2 = 1 \text{ (assume)}$$

$$J_{\min} = 0.75$$

(c) The eigen-values of the matrix  $R$  are the roots of the characteristic equation:

$$(1-d)^2 - 0.25^2 = 0 \Rightarrow (0.5-d)(1.5-d) = 0$$

That is, the two roots are:

$$d_1 = 0.5 \text{ and } d_2 = 1.5$$

The associated eigenvectors are defined by

$$Rq = dq$$

for  $d_1 = 0.5$ , we have

$$\begin{bmatrix} 1 & 0.5 \\ 0.5 & 1 \end{bmatrix} \begin{bmatrix} q_{11} \\ q_{12} \end{bmatrix} = 0.5 \begin{bmatrix} q_{11} \\ q_{12} \end{bmatrix}$$

$$q_{11} + 0.5q_{12} = 0.5q_{11} \quad \text{--- (i)}$$

$$0.5q_{11} + q_{12} = 0.5q_{12} \quad \text{--- (ii)}$$

From (1) & (2)

$$q_{11} = -q_{12}$$

Normalizing the eigenvectors  $q_1$  to unit length, we therefore have

$$q_1 = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 \\ -1 \end{bmatrix}$$

Similarly, for the eigenvalue  $d_2 = 1.5$ , we have.

$$q_2 = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

Accordingly we express the Wiener filter in terms of its eigenvalues & eigenvectors as follows:

$$\begin{aligned} W &= \left( \sum_{i=1}^n \frac{1}{d_i} q_i q_i^H \right) P \\ &= \left( \frac{1}{d_1} q_1 q_1^H + \frac{1}{d_2} q_2 q_2^H \right) P \\ &= \left[ \frac{1}{0.5} \times \left( \frac{1}{\sqrt{2}} \begin{bmatrix} 1 \\ -1 \end{bmatrix} \right) \left( \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & -1 \end{bmatrix} \right) + \frac{1}{1.5} \times \left( \frac{1}{\sqrt{2}} \begin{bmatrix} 1 \\ 1 \end{bmatrix} \right) \left( \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1 \end{bmatrix} \right) \right] \begin{bmatrix} 0.5 \\ 0.25 \end{bmatrix} \\ &= \frac{1}{2 \times 0.5} \left[ \begin{bmatrix} 1 \\ -1 \end{bmatrix} \begin{bmatrix} 1 & -1 \end{bmatrix} + \frac{1}{3} \begin{bmatrix} 1 \\ 1 \end{bmatrix} \begin{bmatrix} 1 & 1 \end{bmatrix} \right] \begin{bmatrix} 0.5 \\ 0.25 \end{bmatrix} \\ &= \frac{1 \times 2}{2 \times 2} \left[ \begin{bmatrix} 1 & -1 \\ -1 & 1 \end{bmatrix} + \begin{bmatrix} \frac{1}{3} & \frac{1}{3} \\ \frac{1}{3} & \frac{1}{3} \end{bmatrix} \right] \begin{bmatrix} 0.5 \\ 0.25 \end{bmatrix} \end{aligned}$$

$$= \begin{bmatrix} \frac{4}{3} & -\frac{2}{3} \\ -\frac{2}{3} & \frac{4}{3} \end{bmatrix} \begin{bmatrix} 0.5 \\ 0.25 \end{bmatrix}$$

$$= \begin{bmatrix} \frac{4}{3} - \frac{1}{3} \\ -\frac{1}{3} + \frac{1}{3} \end{bmatrix}$$

$$= \begin{bmatrix} \frac{1}{2} \\ 0 \end{bmatrix}$$

$$\therefore \begin{bmatrix} 0.5 \\ 0 \end{bmatrix} \Rightarrow \text{Verified with parts (a) \& (b)}$$

~~So, if we don't consider the normalized version,~~  
~~it is equivalent to.~~

$$\begin{bmatrix} 0.5 \\ 0 \end{bmatrix}$$

So, it is hereby verified with (a) \& (b).

Q6

(a) Filter impulse response =  $\omega$ ,  
let  $x$  be the input vector,  $y$  be the output vector.

$$\text{Average power} = \frac{1}{N} \sum_{n=-\infty}^{\infty} |y|^2$$

$$= E[yy^*]$$

$$= E[\omega^H x (\omega^H x)^*]$$

we know that  $(a^H b)^* = b^H a$

$$\text{Hence, } E[yy^*] = E[\omega^H x x^H \omega]$$

$$= \omega^H E[xx^H] \omega$$

By definition,  $E[xx^H] = R$

$$E[yy^*] = \omega^H R \omega$$

Proved.

(b) If the input  $x$  is a zero mean white gaussian noise with variance with  $\sigma^2$  variance we know,

$$R = \sigma^2 I$$

$$\text{Hence, } E[yy^*] = E[y(n)^2] = \omega^H \sigma^2 I \omega$$

$$= \sigma^2 \omega^H \omega$$

Q5: (a) The summarized version of the LMS algorithm:

$$y(n) = \hat{w}^H(n) u(n)$$

$$e(n) = d(n) - y(n)$$

$$w(n+1) = w(n) + \mu u(n) e^*(n)$$

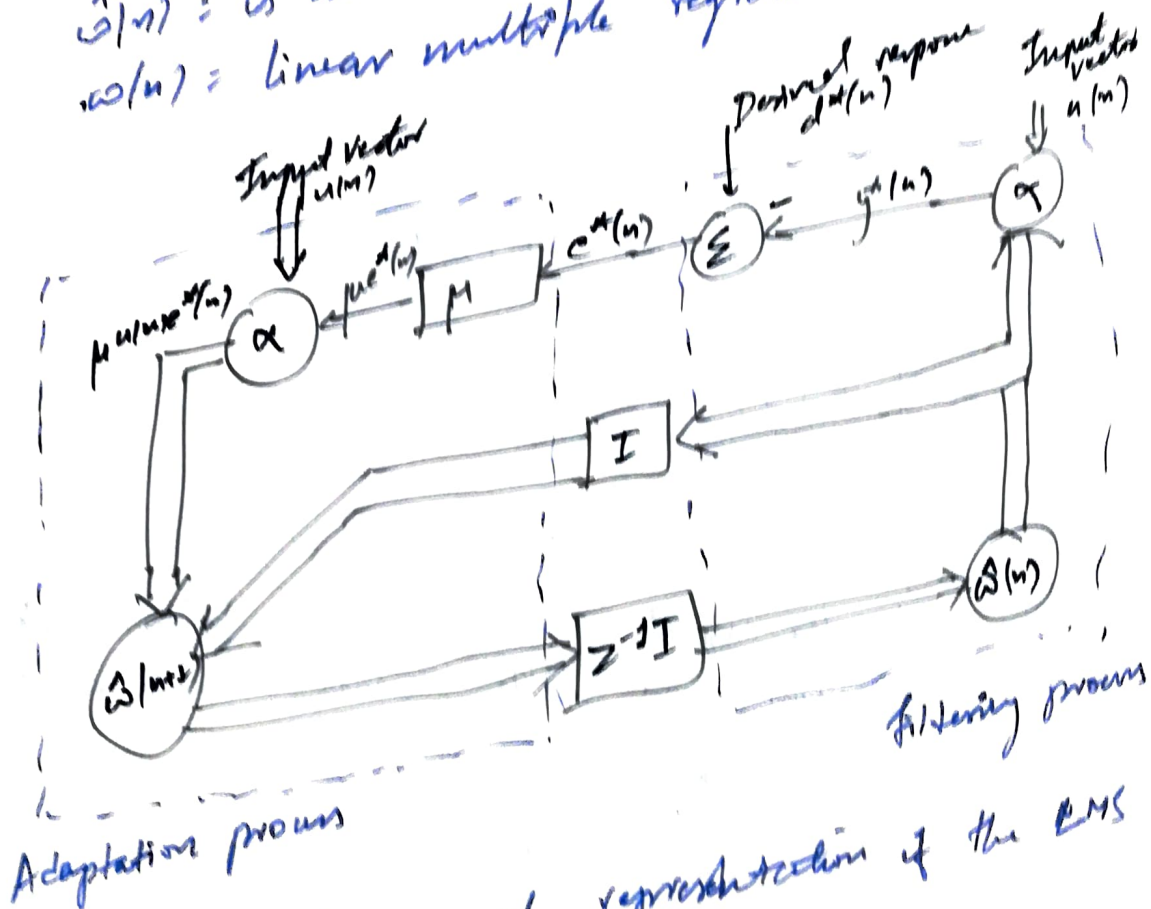
Where:

$u(n)$  is the input vector

$d(n)$  is the corresponding desired response

$\hat{w}(n)$  is an estimate of the tap-weight vector

$w(n)$  = linear multiple regression model



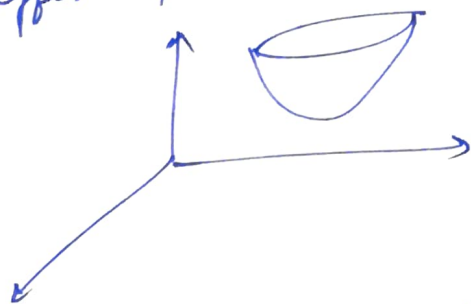
Signal flow graph representation of the LMS algorithm.



b) The differences are the following:

→ Comparing the control mechanism for the LMS algorithm with the method of steepest descent, we observe that the LMS algorithm uses the product  $u(n-k)e^*(n)$  as an estimate of descent is the gradient vector  $\nabla J(n)$  that characterizes the method of steepest descent.

— Now there is no expectation operator in all the paths. Therefore the recursive computation of each tap weight in the LMS algorithm suffers from a gradient noise.



For such an environment, we know that the method of steepest descent computes a tap vector  $\hat{w}(n)$  that moves down the ensemble average error performance surface along a deterministic trajectory that terminates on the Wiener solution  $w_0$ .

Q2. (9)  $x(n) = 0.5x(n-1) + w(n).$

hence,

$$w_1 = 0.5$$

and the AR parameters equal

$$a_1 = -0.5.$$

Accordingly we write Yule-Walker Equations,

$$[r(0)] [w_1] = [r(1)]$$

$$r(1) = 0.5[r(0)]$$

Since the noise  $w(n)$  has zero mean, so, will the AR process  $x(n)$ . Hence,

$$\begin{aligned} \text{Var}[x(n)] &= E[(x^2(n))] \\ &= r(0) \end{aligned}$$

Also,

$$\begin{aligned} \sigma_w^2 &= \sum_{k=0}^{\infty} a_k r(k) \\ &= r(0) + a_1 r(1) \end{aligned}$$

$$1 = r(0) - 0.5 r(1)$$

$$1 = r(0) - 0.5 \times (0.5 r(0))$$

$$1 = 0.75 r(0).$$



$$r(0) = \frac{4}{3} = 1.33.$$

$$r(1) = \frac{2}{3} = 0.667$$

(b).  $r(n) = 1.5n/(n-1) - 0.6n/(n-2) + u(n)$

hence,

$$\omega_1 = 1.5$$

$$\omega_2 = -0.6.$$

& then AR parameter equal.

$$a_1 = -1.5$$

$$a_2 = 0.6.$$

We write the Yule-Walker Equation, as.

$$\begin{bmatrix} r(0) & r(1) \\ r(1) & r(0) \end{bmatrix} \begin{bmatrix} -1.5 \\ -0.6 \end{bmatrix} = \begin{bmatrix} r(1) \\ r(2) \end{bmatrix}$$

$$1.5r(0) - 0.6r(1) = r(1)$$

$$1.5r(1) - 0.6r(0) = r(2)$$

from here,

$$r(1) = \frac{1.5}{1.6} r(0)$$

$$r(2) = \frac{129}{160} r(0).$$

$$\text{var}(y/n) = E[y^2/n] \\ = r(0).$$

$$\sigma_u^2 = \sum_{k=0}^{\infty} a_k r(k)$$

$$1 = r(0) + \frac{15}{16} r(0) + \frac{129}{160} r(0)$$

$$r(0) = \cancel{0.8744}$$

$$1 = r(0) - 1.5 \times \frac{15}{16} r(0) + 0.6 \times \frac{129}{160} r(0).$$

$$r(0) = 12.903.$$

$$r(1) = 12.096$$

$$r(2) = 10.403$$

Q4:

$$r_y(l) = a^{|l|} \quad -1 < a < 1$$

$$x(n) = y(n) + v(n)$$

$$\begin{bmatrix} r_{xx}(0) & r_{xy}(1) \\ r_{xy}(1) & r_{xx}(0) \end{bmatrix} \begin{bmatrix} \omega(0) \\ \omega(1) \end{bmatrix} = \begin{bmatrix} r_{yx}(0) \\ r_{yn}(1) \end{bmatrix}$$

$$\begin{aligned} r_{dx}(l) &= E[y(n) x^*(n-l)] \\ &= r_y(l) + 0 = a^{|l|} \end{aligned}$$

$$\begin{aligned} r_x(l) &= r_y(l) + r_v(l) \\ &= a^{|l|} + \sigma_v^2 \delta(l) \end{aligned}$$

$$\begin{bmatrix} 1 + \sigma_v^2 & a \\ a & 1 + \sigma_v^2 \end{bmatrix} \begin{bmatrix} \omega(0) \\ \omega(1) \end{bmatrix} = \begin{bmatrix} 1 \\ a \end{bmatrix}$$

$$\begin{bmatrix} \omega(0) \\ \omega(1) \end{bmatrix} = \frac{1}{(1 + \sigma_v^2)^2 - a^2} \begin{bmatrix} 1 + \sigma_v^2 - a^2 \\ a \sigma_v^2 \end{bmatrix}$$

$$\frac{MMSE}{J_{min}} = \sigma_y^2 - p^H \omega_0$$

$$= r_y(0) - \omega(0) r_{yn}(0) - \omega(1) r_{yn}(1)$$

$$= 1 - \frac{(1 - a^2) + \sigma_v^2 / (1 + a^2)}{(1 + \sigma_v^2)^2 - a^2} = \frac{\sigma_v^2 \sqrt{1 + \sigma_v^2 - a^2}}{(1 + \sigma_v^2)^2 - a^2}$$

$$= \sigma_v^2 \left[ \frac{1}{1 + \sigma_v^2 + a^2} \right]$$