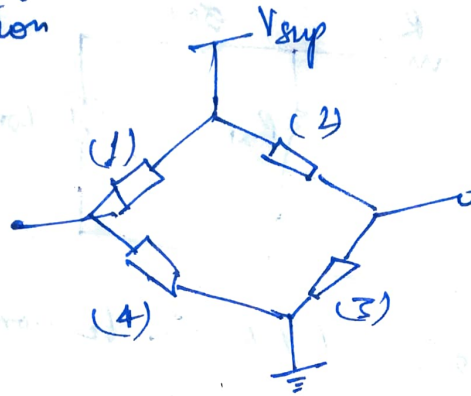


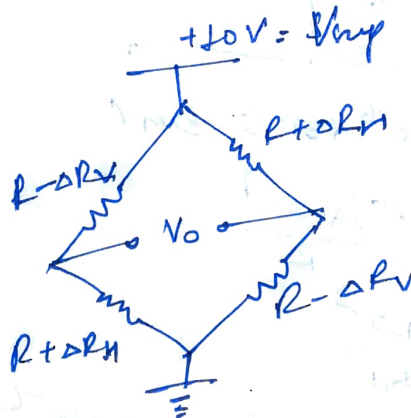
Industrial Instrumentation Assignment-2

Pratyush Jainwal
18EE35014

Q1. (a) Connection



(b) Let applied force be F , area of cross-section
 $A = \frac{\pi d^2}{4} [d = 0.3 \text{ mm}]$



$$V_0 = \left[\frac{R + \Delta R_H}{2R + \Delta R_H - \Delta R_V} - \frac{R - \Delta R_V}{2R + \Delta R_H - \Delta R_V} \right] V_{sup}$$

$$V_0 = 10 \times \frac{2 \Delta R_H + \Delta R_V}{2R + \Delta R_H - \Delta R_V} \approx 10 \left[\frac{\Delta R_V + \Delta R_H}{2R} \right]$$

$$|\epsilon_{\text{vertical}}| = \frac{F}{AY} \Rightarrow \Delta R_V = \frac{\alpha F R}{AY}$$

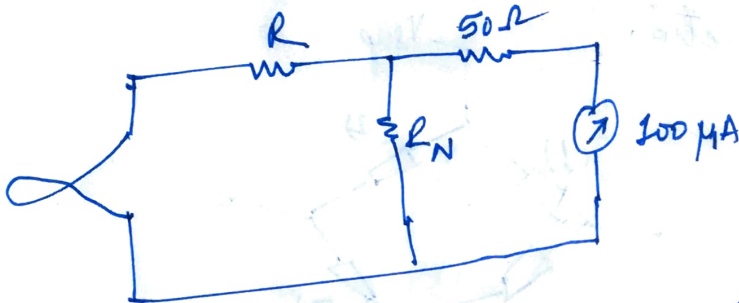
$\alpha \Rightarrow$ Gauge's ~~constant~~ Factor
 $Y \Rightarrow$ Young's Modulus

$$|\epsilon_{\text{horizontal}}| = \frac{F \mu}{AY} \Rightarrow \Delta R_H = \frac{\alpha F R \mu}{AY} \quad \mu \Rightarrow \text{Poisson Ratio}$$

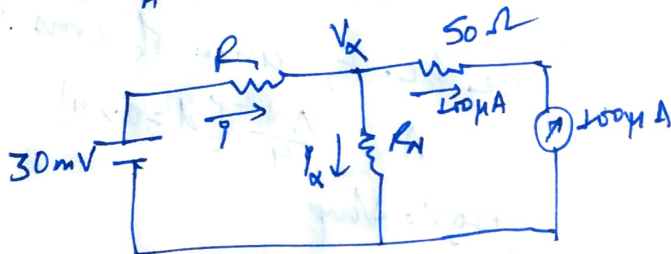
$$V_o = \frac{I_o \times F R_\alpha}{A_T \times 2R} [1 + M]$$

$$\frac{V_o}{F} = \frac{5\alpha}{A_T} (1 + M) = \frac{5\alpha}{\frac{\pi d^2}{4}} (1 + M) = 9.1956 \times 10^{-4} \frac{V}{\mu A}$$

Q. 3)



At $T_A = 25^\circ\text{C}$ (let 25°C be normal temperature)



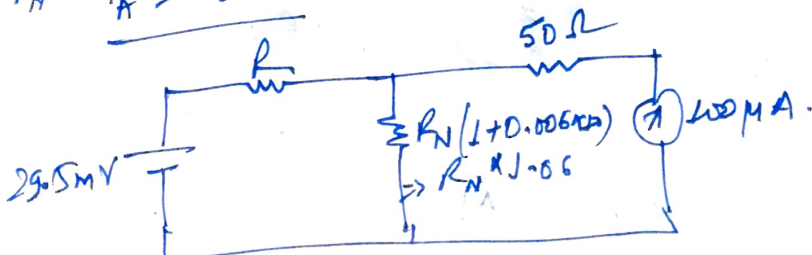
$$V_x = 100 \times 10^{-6} \times 50 = 5 \text{ mV}$$

$$I = \frac{(30 - 5) \times 10^{-3}}{R} =$$

$$P_x = \frac{5 \times 10^{-3}}{R_N}$$

$$\frac{25 \times 10^{-3}}{R} = \frac{5 \times 10^{-3}}{R_N} + 100 \times 10^{-6}$$

At $T_A = 30^\circ\text{C}$



$$P = \frac{24.5 \times 10^{-3}}{R} ; P_x = \frac{5 \times 10^{-3}}{1.06 R_N}$$

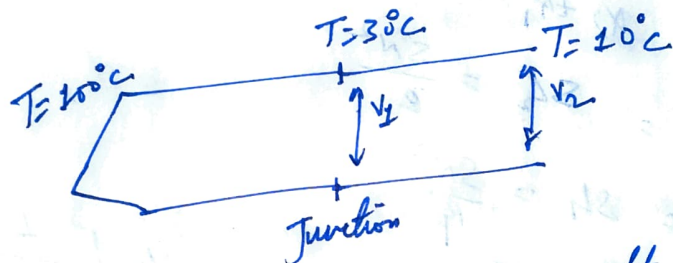
$$\frac{24.5 \times 10^{-3}}{R} - \frac{5 \times 10^{-3}}{1.06 \text{ k}\Omega} = 100 \times 10^{-6}$$

Solving the above two equations,

$$\frac{1}{R} = 6.185567 \times 10^{-3}, \quad \frac{1}{R_N} = 0.0109278$$

$$\Rightarrow R = 161.67 \Omega, \quad R_N = 91.5094 \Omega$$

Q4.



At the Junction, voltage across the terminal is

Type E ($T_2 = 10^\circ\text{C}$) - Type E ($T = 30^\circ\text{C}$)

$$V_1 = 6.317 \text{ mV} - 1.801 \text{ mV} = 4.516 \text{ mV}$$

Voltage due to copper-constantan blade

$$V_2 - V_1 = \text{Type T} (T = 30^\circ\text{C}) - \text{Type T} (T_2 = 0^\circ\text{C})$$

$$= 1.916 \text{ mV}$$

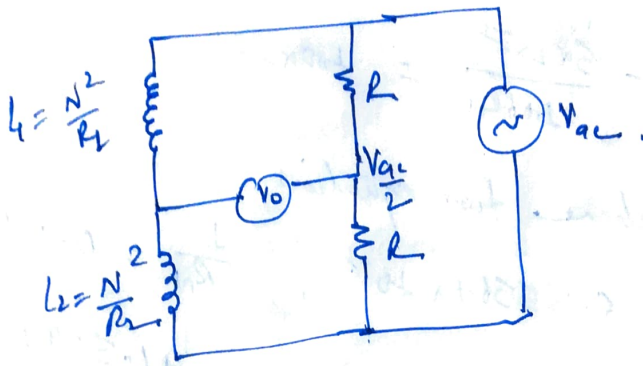
$$\therefore V_2 = 5.712 \text{ V}$$

Measured Temperature (assuming chromel-constantan thermocouple)

$$T = \frac{10 \times (5.712 - 5.646)}{(6.312 - 5.646)} + 30$$

$$= 90.99^\circ\text{C}$$

Q6.



$$V_0 = \frac{V_{ac} X_2}{X_2 + R_1} - \frac{V_{ac}}{2}$$

$$X_2 = S L_2 = \frac{S N^2}{R_2}$$

$$X_1 = S L_1 = \frac{S N^2}{R_1}$$

$$V_0 = \frac{V_{ac} \times \frac{1}{R_2}}{\frac{1}{R_1} + \frac{1}{R_2}} - \frac{V_{ac}}{2} = \frac{V_{ac}}{2} \left[\frac{1}{\frac{R_2 + 1}{R_1}} - \frac{1}{2} \right]$$

$$\frac{R_2}{R_1} = \frac{R_0 + K(\theta + d)}{R_0 + K(\theta - d)}$$

$$V_0 = V_{ac} \left[\frac{R_0 + K(\theta - d)}{R_0 + K(\theta + d) + R_0 + K(\theta - d)} - \frac{1}{2} \right]$$

$$V_0 = V_{ac} \left[\frac{-kd}{2(R_0 + K\theta)} \right]$$

$$V_0 \propto d$$

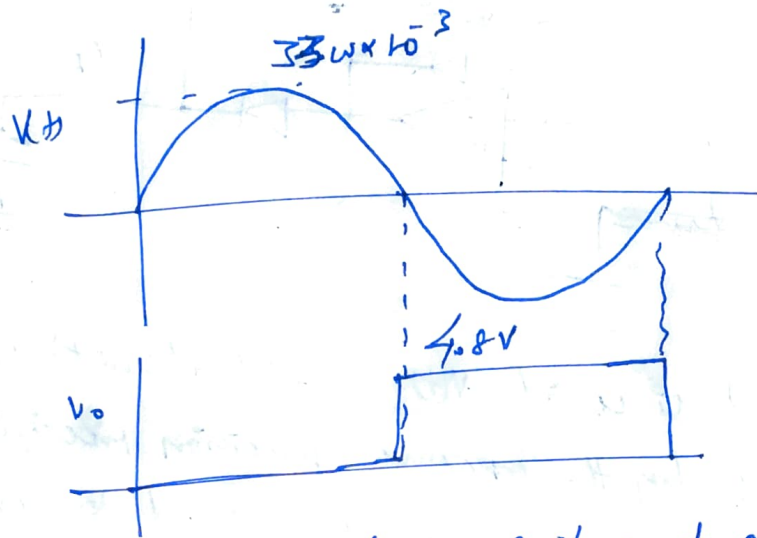
Q7.

$$N(\phi) = 4.0 + j0.5 \cos(220) \text{ mWb.}$$

(a) Number of teeth = 22 (Coeff. of 0)

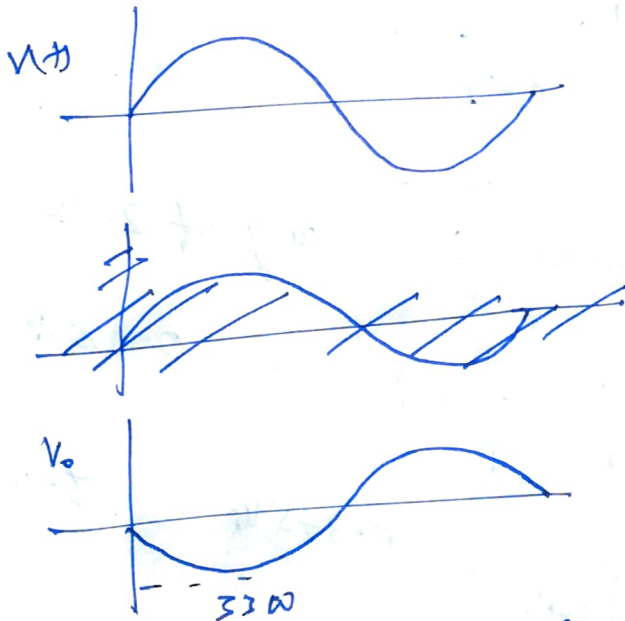
$$(b) V(\phi) = -\frac{dN(\phi)}{dt} = 3.3 \omega \sin(220) \times 10^{-3} \text{ V} \quad \omega = \frac{d\phi}{dt}$$

when $V(t)$ is negative, $V_o = 4.8V$
 when $V(t)$ is positive, $V_o = 0V$



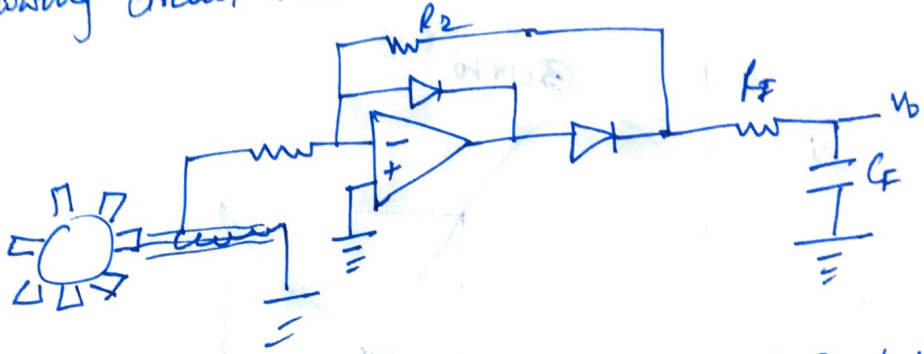
This waveform looks similar at all frequencies.

(c) When $R_F = 1k\Omega$, $V_o = -V(t)$



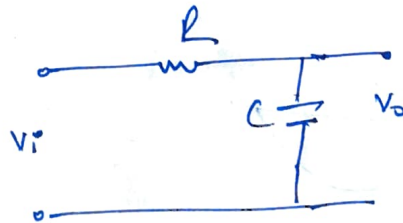
(d) The zener diode output is a square wave which makes measurement of frequency easy (using a digitized counter). Moreover, the output amplitude is constant.

(e) To obtain a DC voltage proportional to ω , the following circuit can be used.



The input sinusoid $V(t) = 330 \sin(2\pi \omega t)$ is full wave rectified by the op-amp-precision rectifier. The low pass filter circuit is used to remove the ripple to obtain a DC voltage proportional to ω .

Q80



for a maximum inaccuracy of 2.5% at 100 rad/s

$$\left| \frac{\frac{1}{j\omega C}}{\frac{1}{j\omega C} + R} \right| \times V_i = V_o = 0.975 V_i \text{ at } \omega = 100$$

$$\frac{V_o}{V_i} = 0.975 V_i, \quad \tau = RC$$

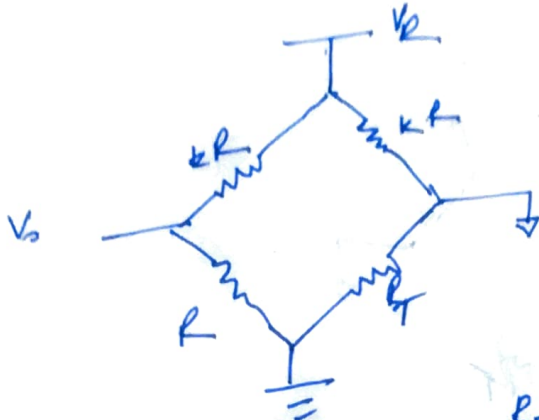
$$\frac{1}{\sqrt{1 + \omega^2 \tau^2}} = 0.975$$

$$\sqrt{1 + \omega^2 \tau^2} = \frac{1}{0.975}$$

$$1 + 100^2 \tau^2 = 1.0519392513$$

$$\tau = 2.279 \text{ ms}$$

Q90 (a)



$$V_o = V_R \left[\frac{1}{k+1} - \frac{R_T}{R_T + kR} \right]$$

Let the three points be:

V_o	0	2.5	5.0
ΔT	0	25	50

(i) $\Delta T = 0$ ($T = 273 \text{ K}$)

$$R_T = 1.68 \text{ e}$$

$$3050 \left(\frac{1}{273} - \frac{1}{298} \right) = 4.289 \text{ k}\Omega$$

$$\therefore V_R \left[\frac{1}{k+1} - \frac{4.289}{4.289 + kR} \right] = 0 \text{ V}$$

(ii) $\Delta T = 25$ ($T = 298$)

$$R_T = 1.68 \text{ e}$$

$$3050 \left(\frac{1}{298} - \frac{1}{298} \right) = 1.68 \text{ k}\Omega$$

$$\therefore V_R \left[\frac{1}{k+1} - \frac{1.68}{1.68 + kR} \right] = 2.5 \text{ V}$$

(iii) $\Delta T = 50$ ($T = 323 \text{ K}$)

$$R_T = 1.68 \text{ e}$$

$$3050 \left(\frac{1}{323} - \frac{1}{298} \right) = 0.7608 \text{ k}\Omega$$

$$\therefore V_R \left[\frac{1}{k+1} - \frac{0.7608}{0.7608 + kR} \right] = 5.0 \text{ V}$$

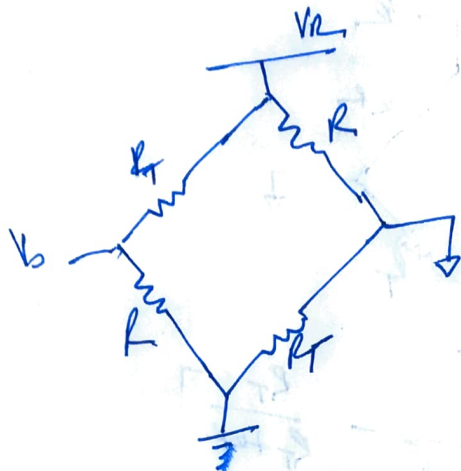
Solving these equations yields

$$V_R = 12.7898 \text{ V}$$

$$R = 4.289 \text{ k}\Omega$$

$$k = 0.271$$

(b)



$$V_o = V_R \left[\frac{R}{R_T + R} - \frac{R_T}{R + R_T} \right]$$

$$V_o = V_R \left[\frac{R - R_T}{R + R_T} \right]$$

Since there are only 2 degrees of freedom, i.e., R, V_R . This configuration cannot be fit to 3-points.

$$\therefore \Delta T = 0 \quad (T = 273K)$$

$$V_R \left[\frac{R - 4.289}{R + 4.289} \right] = 0$$

$$R = 4.289k\Omega$$

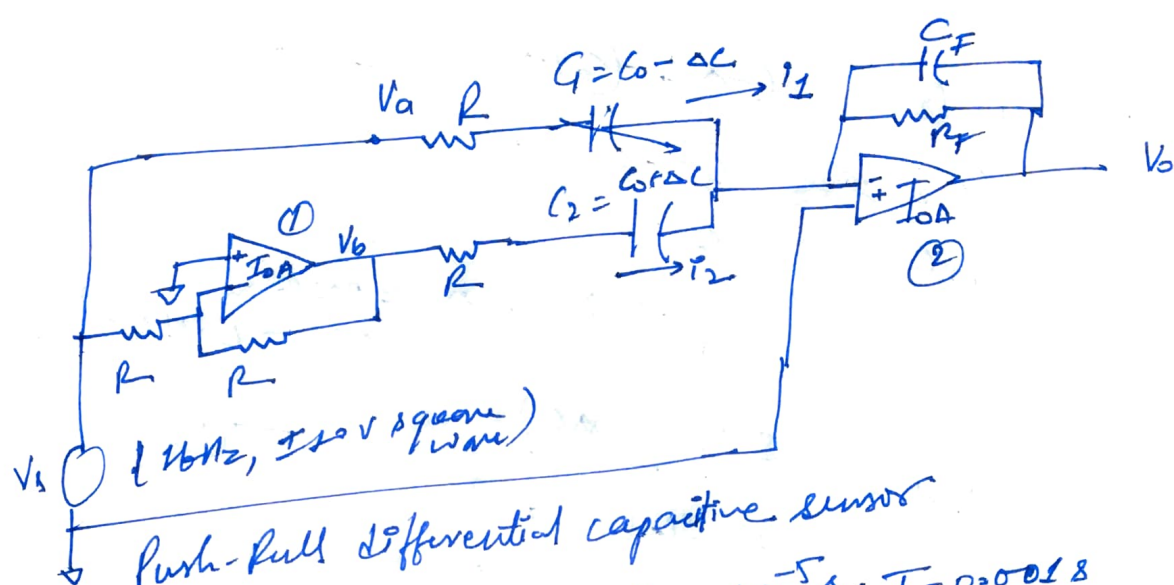
$$\Delta T = 50 \quad (T = 323K)$$

$$V_R \left[\frac{R - 0.7608}{R + 0.7608} \right] = 5$$

$$V_R = 7.1563V$$

In this configuration, non-linearity is decreased whereas sensitivity remains approximately constant.

11)



$$K_G \ll T \ll R_F C_F; R_G = 10^{-5} s; T = 0.0018$$

$$R_F C_F = 0.18.$$

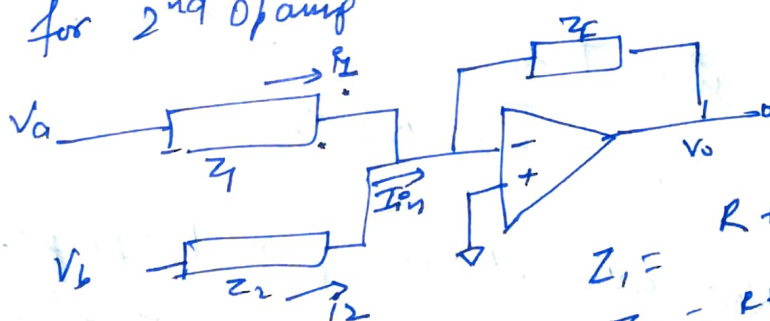
Assumption = Ideal Opamps.

Here, $V_a = V_s$; K_{CL} at (-) of ①:

$$\frac{V_s - 0}{R} = \frac{0 - V_b}{R} \Rightarrow \boxed{V_b = -V_s}$$

(a)

for 2nd Opamp



$$Z_1 = R + \frac{1}{sC_1}$$

$$Z_2 = R + \frac{1}{sC_2}$$

$$\Rightarrow Z_F = (R_F \parallel \frac{1}{sC_F})$$

$$I_{in} = I_1 + I_2$$

$$= \frac{V_a}{Z_1} + \frac{V_b}{Z_2}$$

$$I_{in} = V_s \left(\frac{1}{Z_1} - \frac{1}{Z_2} \right)$$

$$I_{in} = V_s \left(\frac{sC_1}{(R s C_1 + 1)} - \frac{sC_2}{(R s C_2 + 1)} \right)$$

$$I_{in} = \frac{s V_s (C_1 - C_2)}{1 + sR(C_1 + C_2) + s^2 R^2 C_1 C_2}$$

$$i_1 - i_2 = (-\Delta C - (C_0 + \Delta C)) = -2\Delta C$$

$$i_1 i_2 = C_0^2 - \Delta C^2 \approx C_0^2$$

$$i_1 + i_2 = 2C_0$$

$$\left\{ \begin{array}{l} \therefore \text{ gives } \Delta C \approx 0.1 C_0 \Rightarrow \\ \Delta C^2 = 0.01 C_0^2 \end{array} \right.$$

$$i_{IN} = \frac{S V_s (-2\Delta C)}{1 + s R_C (2C_0) + s^2 R_C^2 C_0^2} \quad \text{--- (1)}$$

$$i_{IN}(s) = V(s) \left\{ \frac{\left(\frac{-2\Delta C}{R_C^2 C_0^2} \right) s}{s^2 + \frac{2s}{R_C} + \frac{1}{R_C^2 C_0^2}} \right\}$$

Putting $s = j\omega$ in (1) $\left(T = \frac{2\pi}{\omega} \right)$

$$i_{IN} = V_s \left(\frac{-2j\omega \Delta C}{T + 2j\omega R_C C_0 + (j\omega)^2 R_C^2 C_0^2} \right)$$

given $T \gg R_C C_0 \Rightarrow \omega R_C C_0 \ll 1$

$$i_{IN} = -2j\omega \Delta C \cdot V_s$$

(b)

Now aVL at (-) of 2nd Opamp:-

$$\frac{0 - V_o}{Z_F} = i_{IN} \Rightarrow V_o = -Z_F i_{IN}$$

Substituting i_{IN} , $V_o = (2j\omega \Delta C V_s) Z_F$

$$\Rightarrow V_o = \frac{2j\omega \Delta C V_s R_F}{1 + j\omega R_F C_F}$$

$$T \ll R_F C_F$$

$$\omega R_F C_F \gg 1$$

$$\approx \frac{2\Delta C R_F V_s j\omega}{j\omega R_F C_F}$$

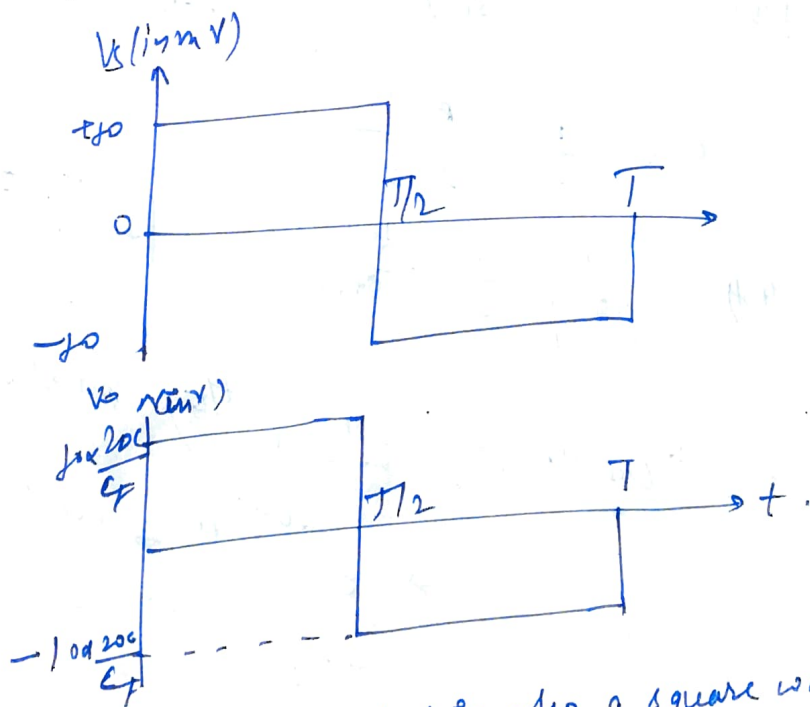
$$\Rightarrow V_o = \left(\frac{2\Delta C}{C_F} \right) V_s$$

(c) If, $\Delta C = 0.1 G \sin(2\pi f t)$
 here, $f = \frac{10}{100} \times \frac{1}{T} = \frac{1}{10T}$
 $\Delta C = 0.1 G \sin\left(2\pi \cdot \frac{1}{10T} \cdot t\right)$

(d) $\Rightarrow V_o = \frac{2 \times 0.1 G}{C_f} \times \sin\left(\frac{2\pi t}{10T}\right) \propto V_s$ 11c

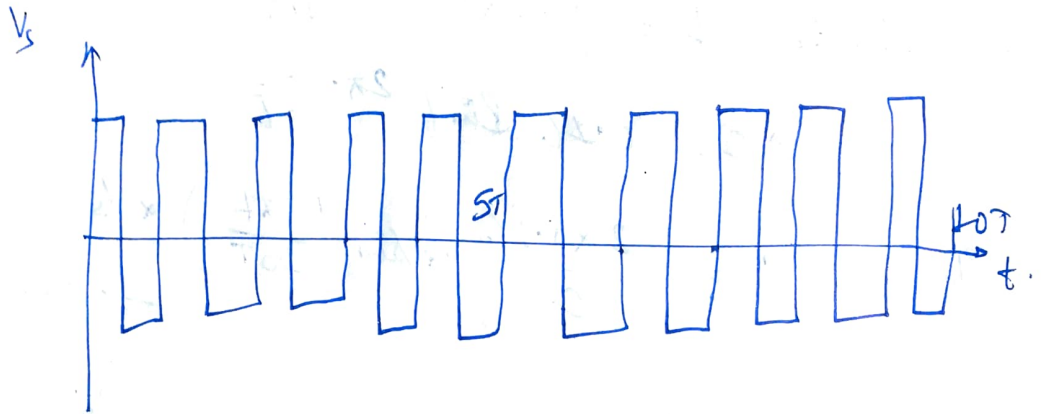
11) a) from (11a):
 $I_{IN} = -g_m \Delta C V_s$; $\omega = \frac{2\pi}{T}$
 $T = 0.001 s.$

11b) from 11(b): given constant ΔC ,
 $\Rightarrow V_o$ will also be square wave output.
 here $T = 0.001 s.$



\Rightarrow Output is also a square wave.
 which is in phase with V_s but of
 different amplitude.

$$14) c) \text{ (H.C.) } \Rightarrow V_o = \frac{0.2 C_o}{C_f} \sin\left(\frac{2\pi t}{10T}\right) \cdot V_s$$



Also \equiv As it is of first order: $H(s) = \frac{k}{sT+1}$

Step Input $x(s) = \frac{1}{s} \Rightarrow y(s) = \frac{k}{s(sT+1)}$

$$Y(s) = \frac{A}{s} + \frac{B}{sT+1} \Rightarrow A = k$$

$$B = -AT \Rightarrow -Ak$$

$$Y(s) = \frac{k}{s} - \frac{AkT}{sT+1}$$

$$y(t) = \mathcal{L}^{-1} \left[\frac{k}{s} - \frac{kT}{sT+1} \right] = k \left(1 - e^{-t/T} \right)$$

from 1st Condition: $k \left(1 - e^{-t/T} \right) \geq 0.95k$

$$\Rightarrow e^{-0.05/T} \leq 0.05$$

$$\Rightarrow T \leq \frac{0.05}{\ln 20} = 0.016698$$

$$= 16.69 \text{ ms}$$

Ramp Input of 100 PST/s

$$x(s) = \frac{100}{s^2} \Rightarrow y(s) = \frac{100k}{s^2(sT+1)}$$

$$= \frac{A}{s} + \frac{B}{s^2} + \frac{C}{1+sT}$$

$$As(sT+1) + B(1+sT) + A^2C = 100k.$$

$$\boxed{B = 100k}$$

$$AT + C = 0 \Rightarrow C = -AT$$

$$A + BT = 0$$

$$\Rightarrow A = -BT = -100kT$$

$$AC = +100kT^2$$

$$Y(s) = \frac{-100kT}{s} + \frac{100k}{s^2} + \frac{100kT^2}{1+sT}$$

$$\boxed{y(t) = 100k \left(t - T + Te^{-t/T} \right)}$$

$$t \rightarrow \infty: y(t) = 100kT - 100kT; \quad x(t) = 100kT$$

$$s \text{ error: } \lim_{t \rightarrow \infty} (t x(t) - y(t)) = 100kT \leq 2$$

$$\boxed{kT \leq 0.02}$$

c) for a first order T.F & sine wave input,

$$A \sin \omega t \Rightarrow H(s) = \frac{K}{1+sT}$$

$$y(t) = Ak \left[\frac{\omega_0 t}{1+\omega_0^2 T^2} e^{-t/T} + \frac{1}{\sqrt{1+\omega_0^2 T^2}} \sin(\omega_0 t + \phi) \right]$$

$$\phi = \tan^{-1}(\omega_0 T)$$

$$\text{as } t \rightarrow \infty \Rightarrow y(t) = \frac{Ak}{\sqrt{1+\omega_0^2 T^2}} \sin(\omega_0 t + \phi)$$

Given from 3rd Condition:

$$\frac{Ak}{\sqrt{1+(40\pi T)^2}} \geq 0.9 Ak$$

$$1 + 40^2 \pi^2 T^2 \leq (0.9)^{-2}$$

$$T \leq \frac{\sqrt{19}}{360\pi} \text{ s}$$

$$T \leq 3.854 \text{ ms}$$

from above,

$$T \leq 3.854 \text{ ms}$$

$$RT \leq 0.020 \text{ PSI}$$

$$= 20 \text{ mPSI}$$

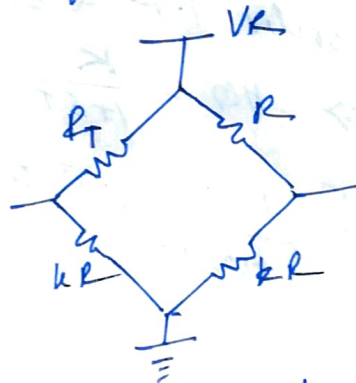
$k \rightarrow$ Steady State Gain
 $T \rightarrow$ Time Constant

Q5

Extreme Scenario:-

$$T_a = 0^\circ \text{C}$$

Bridge Voltage should be 0V.



$$R_T = R_0 (1 + \alpha \Delta T), \text{ where } \Delta T = 0$$

$$\Rightarrow R_T = R_0 = 100 \Omega$$

$$\text{for } \Delta V = 0 \Rightarrow R = R_0 = 100 \Omega$$

$$T_a = 100^\circ\text{C}$$

Bridge Voltage should be $\Delta V = 5.268 \text{ mV}$

$$\Delta V = V_R \left[\frac{kR}{(k+1)R} - \frac{kR}{kR + R(1 + \Delta T \alpha)} \right] = 5.268 \times 10^{-3}$$

$$\Rightarrow V_R \left[\frac{k}{k+1} - \frac{k}{k+1 + 0.00392 \times 100} \right] = 5.268 \times 10^{-3}$$

$$\Rightarrow \frac{k \times V_R \times [0.0392]}{(k+1)(k+1.392)} = 5.268 \times 10^{-3}$$

Let $V_R = 10 \text{ V} \Rightarrow k = 0.0018767$

$$\therefore kR = 0.18767 \Omega$$

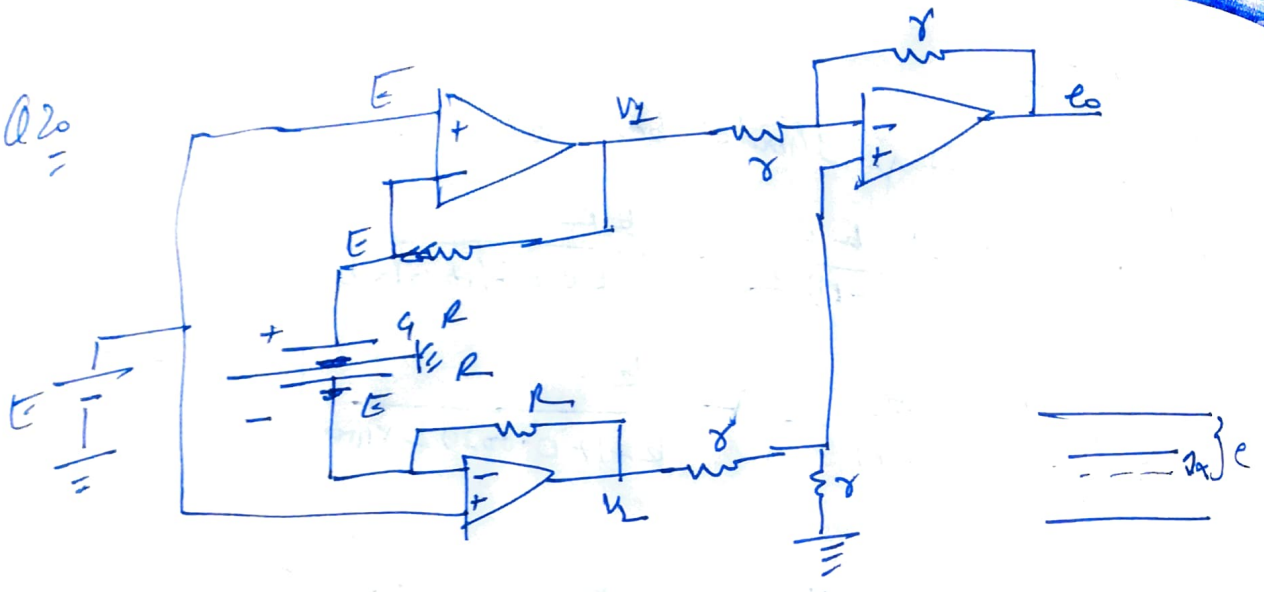
The error due to non-linearity can be defined as:-
 $\text{error} = V_{\text{compensated}}|_{T_a=T} - 5.268 \times 10^{-3}$

where $V_{\text{compensated}}|_{T_a=T} = -V_{\text{thermocouple}}|_{T_a=T}$

Temperature (Ambient $^\circ\text{C}$)	Error (mV)
0	0
10	0.1983
20	0.3404
30	0.4318
40	0.4769
50	0.4800
60	0.4464
70	0.3784
80	0.2795
90	0.1534
100	0

$$\therefore \text{Non-linearity} = \frac{0.4800}{5.268} \times 100\% = 9.11\%$$

Q20



$$\frac{3e^{-n} G}{\{3e+n\} C_2}$$

$$G = \frac{EA}{d_1} = \frac{EA}{e-n}$$

$$C_2 = \frac{EA}{e+n}$$

$$v_1 = E + ER \frac{dG}{dt}$$

$$v_1 = E + ER \frac{dC_2}{dt}$$

Then there is a difference amplifier stage

$$e_o = v_2 - v_1$$

$$= ER \left\{ \frac{dC_2}{dt} - \frac{dG}{dt} \right\}$$

$$e_o = ER \left\{ \frac{dC_2}{dt} - \frac{dG}{dt} \right\}$$

$$\frac{dC_2}{dt} = \frac{-EA}{(e+n)^2} \frac{dn}{dt}$$

$$\frac{dG}{dt} = + ER \frac{EA}{(e-n)^2} \frac{dn}{dt}$$

$$\begin{aligned}
 l_0 &= -EREA \left\{ \frac{(c+n)^2 + (c-n)^2}{(c+n)^2 (c-n)^2} \right\} \frac{dn}{dt} \\
 &= -EAE R \left\{ \frac{2(c^2 + n^2)}{(c^2 - n^2)^2} \right\} \frac{dn}{dt} \\
 &\quad e^2 \gg n^2 \left\{ \begin{array}{l} \text{Assumption} \\ \text{for linearity} \end{array} \right\}.
 \end{aligned}$$

$$l_0 = -EAE R \frac{2c^2}{c^4} \times \frac{dn}{dt}$$

$$l_0 = \frac{-2EAE R}{c^2} \frac{dn}{dt}$$

Taking L.T

$$\begin{aligned}
 E_0(s) &= \frac{-2EAE R s X(s)}{c^2} \\
 \left| \frac{E_0(s)}{X(s)} \right| &= \frac{-2EAE R s}{c^2}
 \end{aligned}$$