

20/07/23

SNS

- * classification of signals
- * Diff. b/w signals & systems

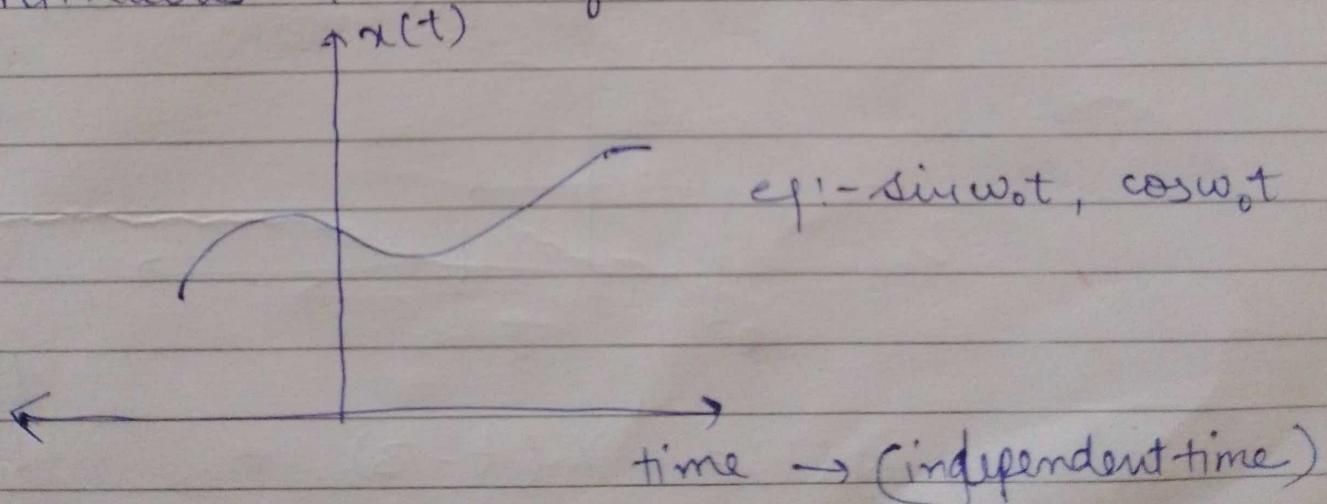
Unit-3 Imp.

[Book :- Oppenheim]

Signal :- Any physical quantity that contains information in a pattern of variation of some functions is known as signal.

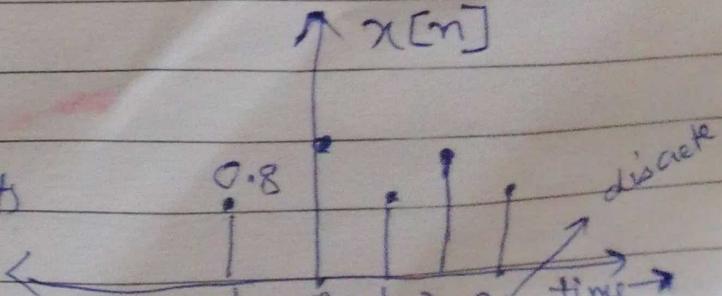
Classification of signal :-

1. Continuous Time Signal :-



2. Discrete Time Signal :-

eg:- crime rates, budget, no. of objects



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- ① Time is discrete in Discrete Time signal
 - ② Digital signal \rightarrow Quantized & coded
 ↳ levels me divide
 ↳ values either (0, 1)

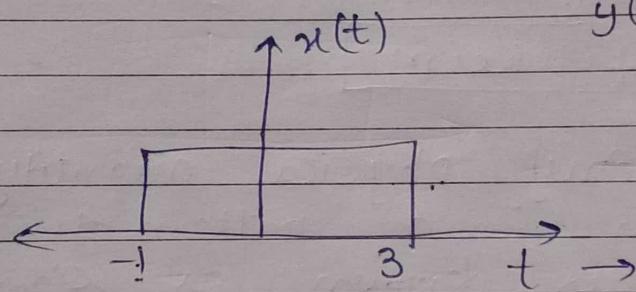
Basic Operations on continuous Time signals :-

- ① Time Shifting \rightarrow Time ke respect me signal ko shift ki rhe h.

- ② Time reversing :- mirror image

- ③ Time Scaling :- Compress or expand signals

Graphical Representation Mathematical Rep.



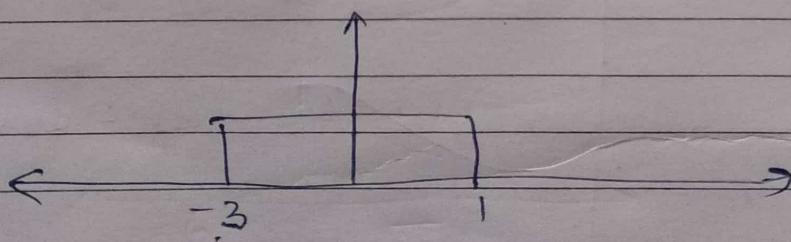
$$y(t) = x(t + \alpha)$$

Time shifting
 +ve, advancement
 -ve \rightarrow delay

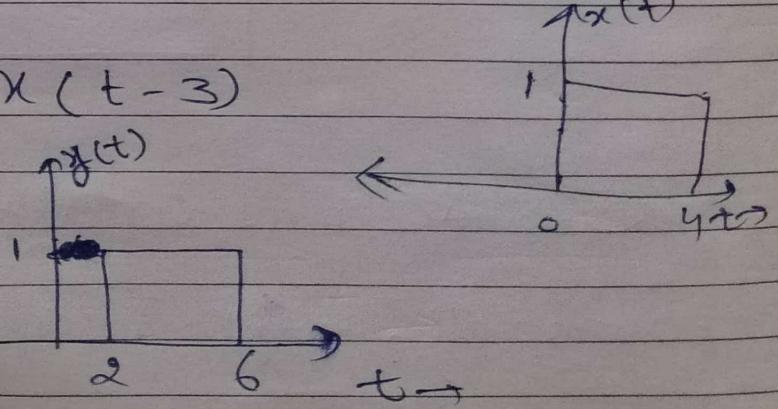
Time reversing

$$y(t) = x(t + 2)$$

$$y(t) = x(t - 2)$$



$$y(t) = x(t - 3)$$



* Bandwidth will not be impacted by operations

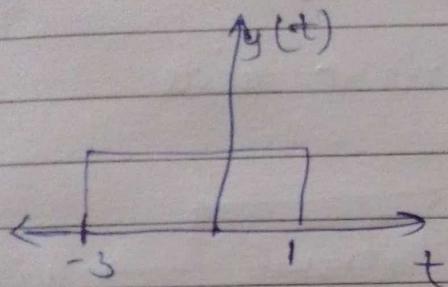
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② Time Reversal :- $y(t)$

Along y-axis mirror image

$$y(t) = x(-t)$$



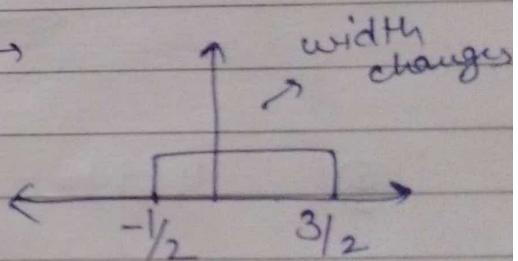
(not time reversal)
 $-x(t) \neq x(t)$
 (mirror image along y-axis)

③ Time Scaling :-

$$y(t) = x(\alpha t)$$

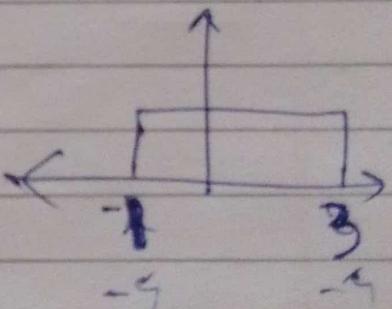
① ~~$\alpha \rightarrow +\infty$ if $\alpha > 1$, compression~~

$$y(t) = x(2t) \rightarrow$$

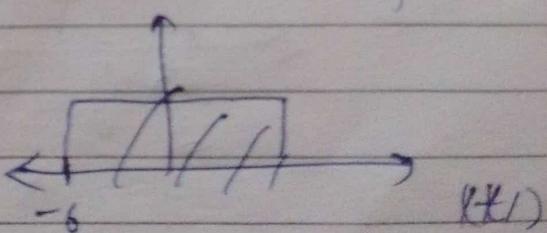
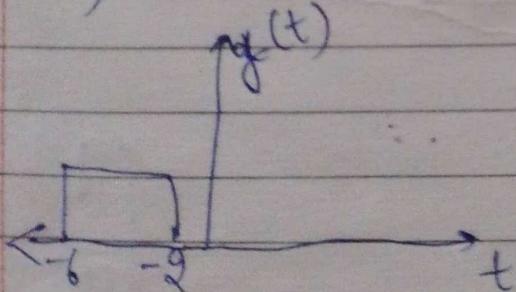


② $0 < \alpha < 1$, expand

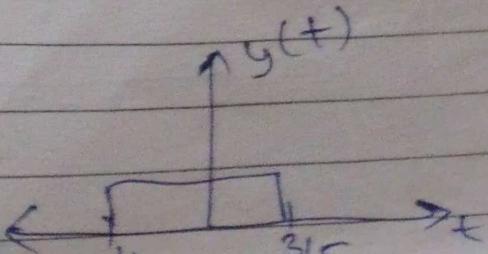
$$y(t) = x\left(\frac{t}{\alpha}\right)$$



$$\textcircled{1} \quad y(t) = x(t+5)$$



$$\textcircled{2} \quad y(t) = x(5t)$$



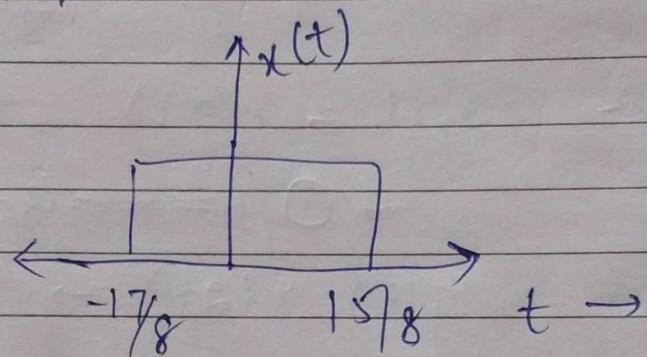
$$\textcircled{2} \quad y(t) = x\left(\frac{7}{8} - t\right)$$

\textcircled{1} convert it into std. form

$$y(t) = x(-t + \frac{7}{8})$$

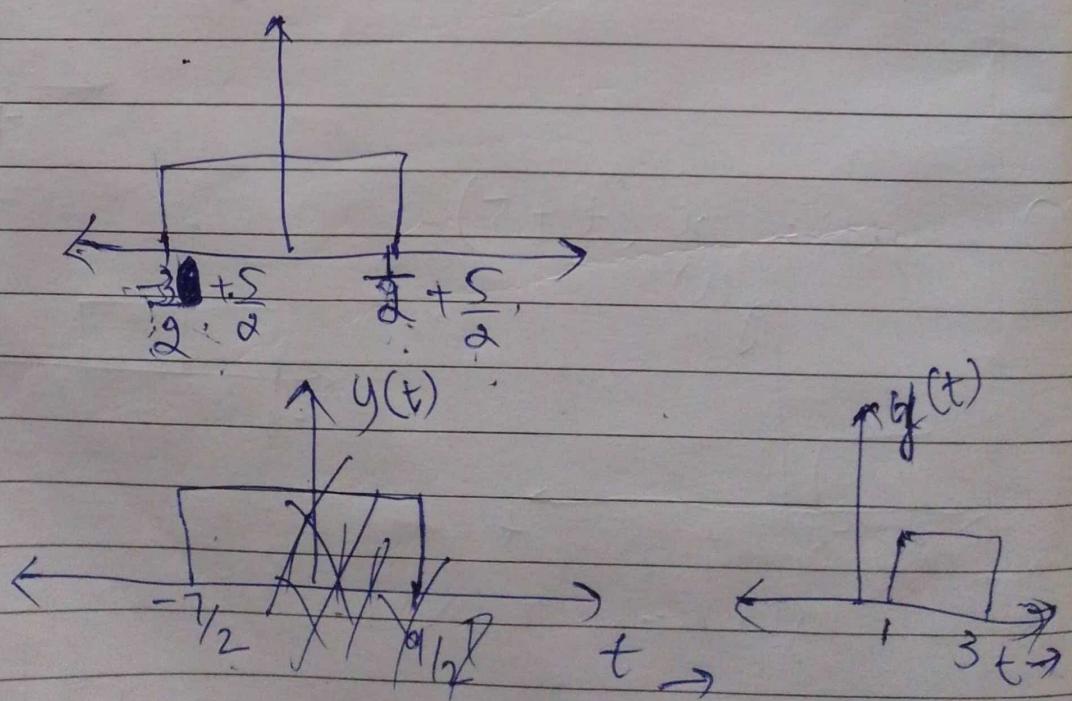
\textcircled{2} start solving from right side

$$x(t) \rightarrow \begin{cases} \text{Time adv.} \\ \frac{7}{8} \end{cases} \rightarrow \begin{cases} \text{Time reversal} \end{cases} \rightarrow y(t)$$



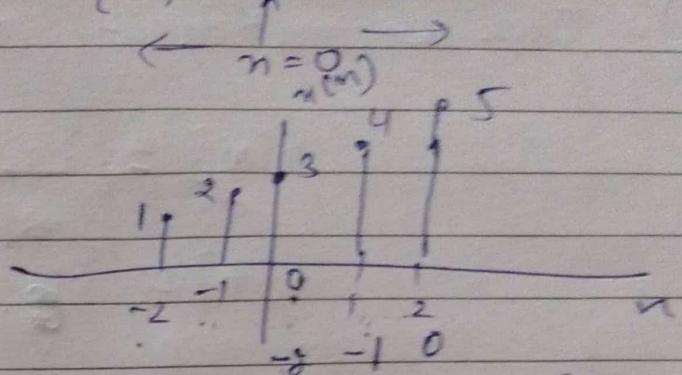
$$\textcircled{4} \quad y(t) = x(5 - 2t)$$

$$y(t) = x\left(-2\left(\frac{-5}{2} + t\right)\right)$$



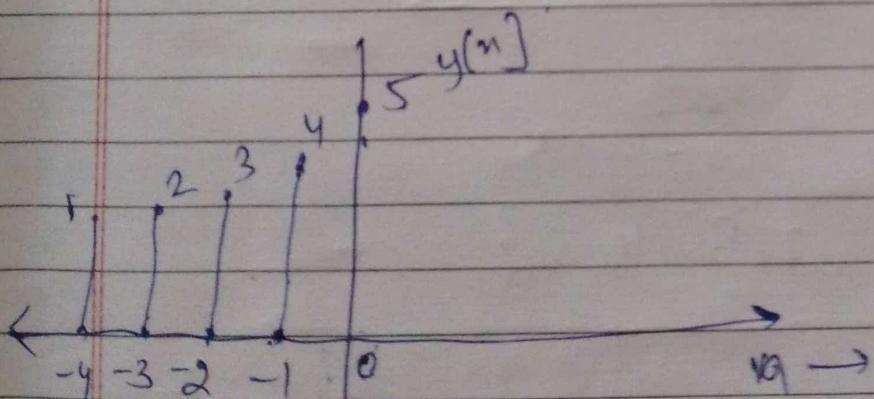
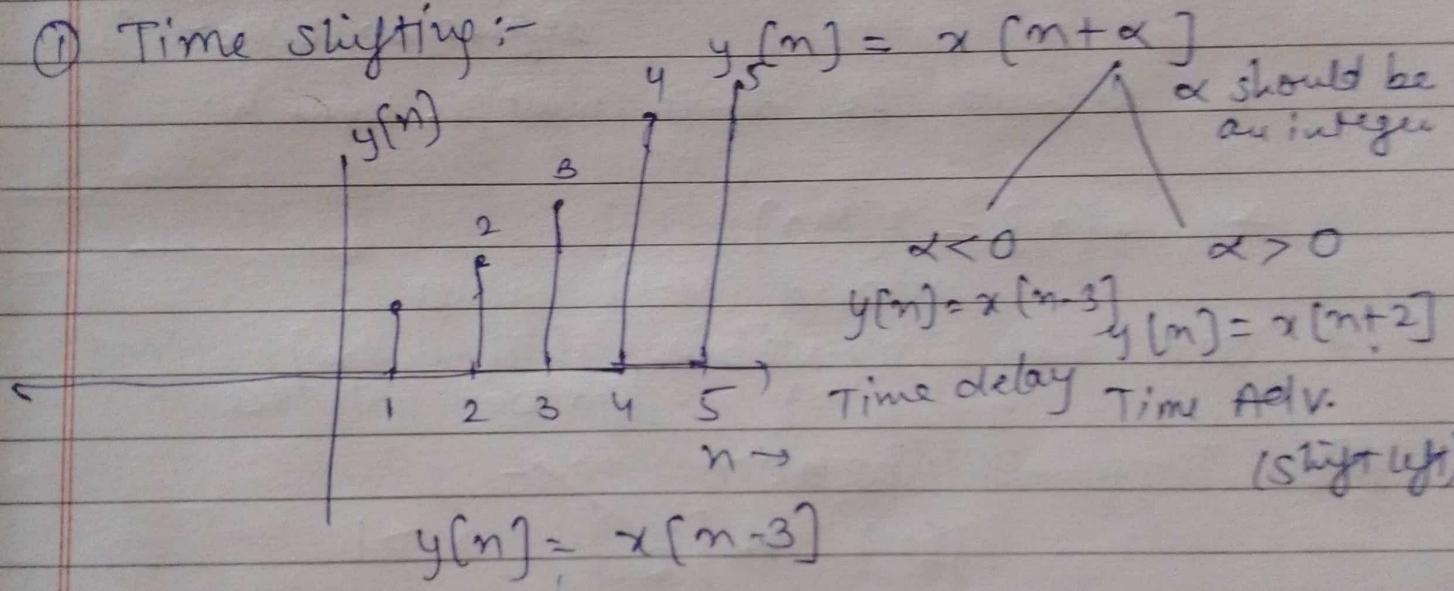
Discrete signals:-

$$x[n] = \{1, 2, 3, 4, 5\}$$



If arrow is not given then left most will be considered.

① Time shifting:-

② Time Reversal:- $y[n] = x[-n]$

$$x[n] = \{1, 2, 3, 4, 5\} \rightarrow y[n] = \{5, 4, 3, 2, 1\}$$

③ Time Scaling:- $y[n] = x[\alpha n]$ α should be integer

$\alpha > 1$

* Time Decimation :- $y[n] = x[2n]$

$$x[n] = \{1, 2, 3, 4, 5, 6\}$$

-2 -1 0 1 2

$$y[n] = \left\{ \begin{array}{c} -1, -1, 0, 1, 0 \\ x \quad x \quad x \quad x \end{array} \right\}$$

$(-1, 0, 1)$ → discard these values, it's not defined in discrete]

n ki value pe divide, multiples kru h.

* $y[n] = x[Mn]$

discard $(M-1)$ values

* It's not invertible (Time decimation) ↳ Data will lose

* Time Interpolation of p :-

$$y[n] = x\left[\frac{n}{2}\right] \quad \alpha > 1$$

$$y[n] = x\left[\frac{n}{2}\right]$$

$$x[n] = \{1, 2, 3, 4, 5\}$$

↑

$$\begin{cases} -2 & -1 & 0 & 1 & 2 \\ -4 & -2 & 0 & 2 & 4 \end{cases}$$

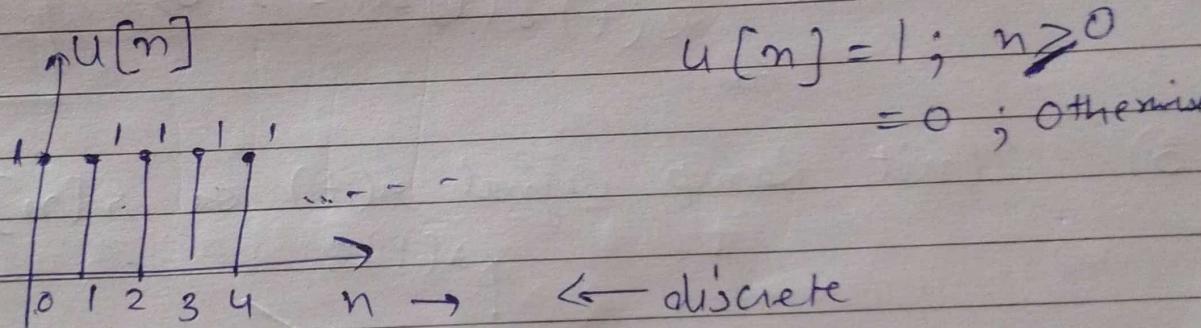
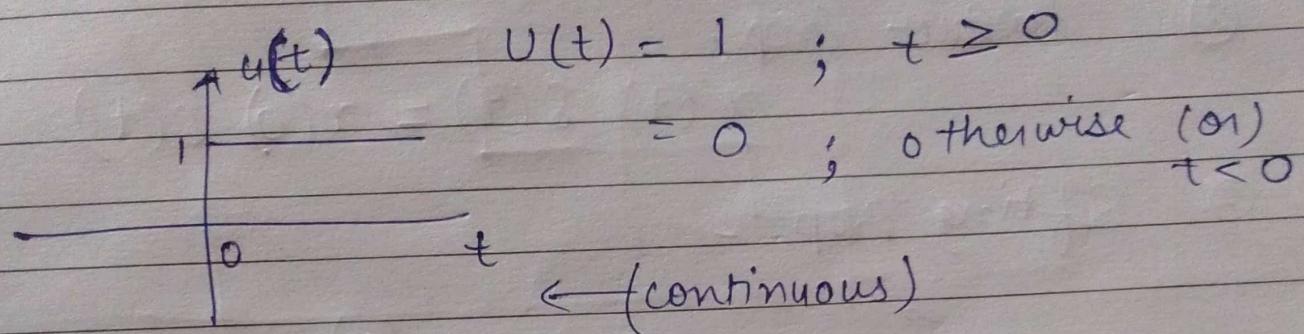
$$y[n] = [1, 0, 2, 0, 3, 0, 4, 0, 5]$$

$$y[n] = x\left[\frac{n}{M}\right]$$

Add $(M-1)$ zeros in mid of two signals (even)

* Basic Elementary ^{signals} Operation :-

① Unit step signal, $u(t)$, $u[n]$:-



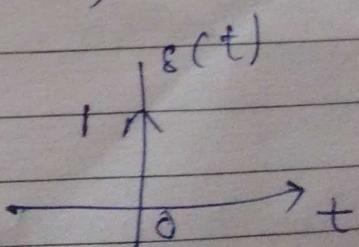
Just take any. $t \geq 0$, $u(t) = \frac{0+1}{2} = \frac{1}{2}$ (Practically)

② Unit Impulse Signal :- $\delta(t)$ (or) $\delta(n)$

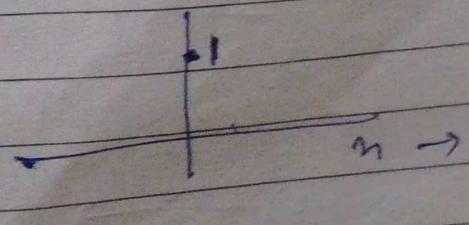
$$\delta(t) = \begin{cases} 1 & ; t = 0 \\ 0 & ; \text{otherwise} \end{cases}$$

Dirac delta fun.

$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$



$$\delta[n] = \begin{cases} 1 & ; n = 0 \\ 0 & ; \text{otherwise} \end{cases}$$



* Properties of Impulse signals :-

① Time scaling

$$\delta(\alpha t) = \frac{1}{|\alpha|} \delta(t)$$

② Product Prop.

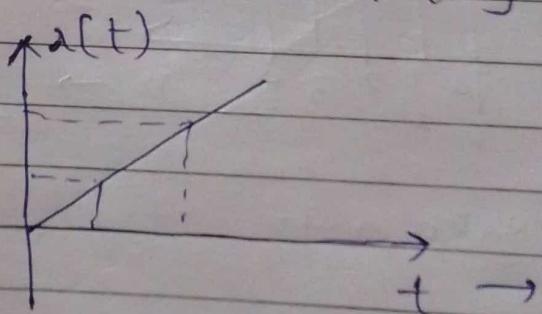
$$x(t) \delta(t) = x(0) \delta(t)$$

③ Shifting Prop.:-

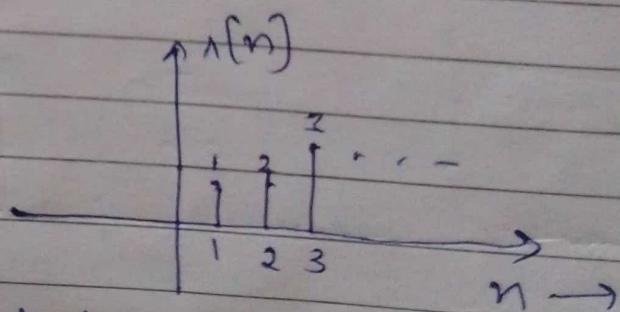
$$\int_{-\infty}^{\infty} x(t) \delta(t) dt = x(0)$$

3) Unit Ramp Signal $r(t)$ or $r[n]$

$$r(t) = t, \quad t \geq 0 \\ = 0, \text{ otherwise}$$

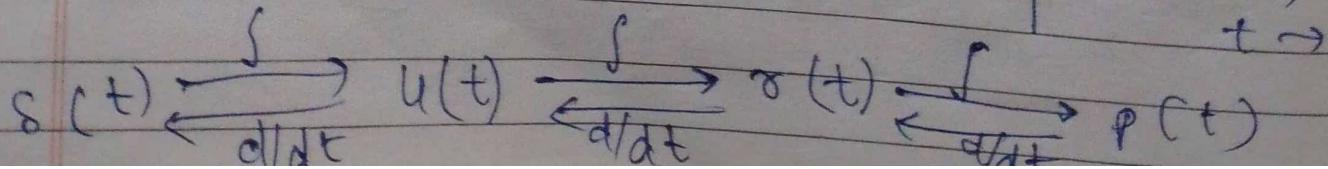


$$r[n] = n \quad n \geq 0 \\ = 0 \quad \text{otherwise}$$



4.) Unit parabolic symbol signal :- $p(t)$

$$p(t) = \frac{t^2}{2}, \quad t \geq 0 \\ = 0, \text{ otherwise}$$



Relation b/w ramp signal & unit step signal:-

$$r(t) = t u(t)$$

$$r[n] = t u[n]$$

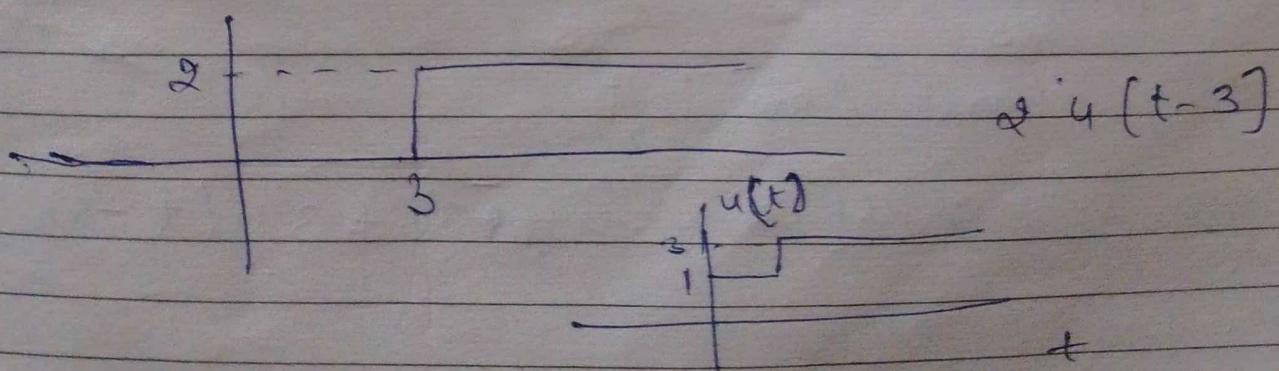
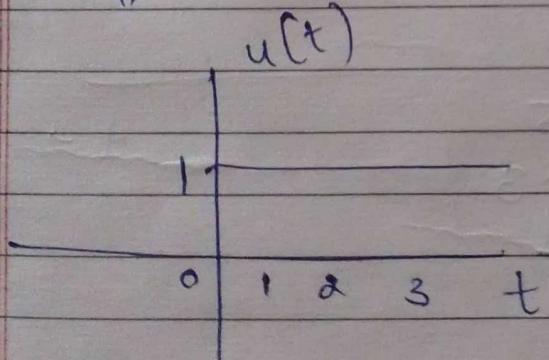
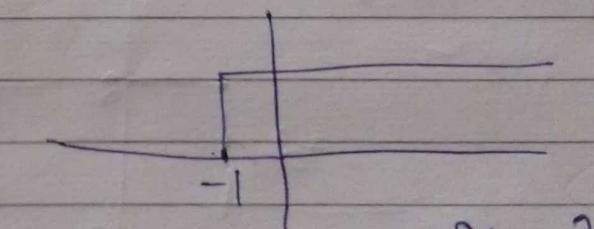
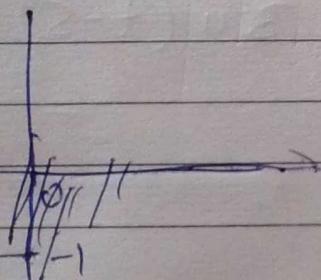
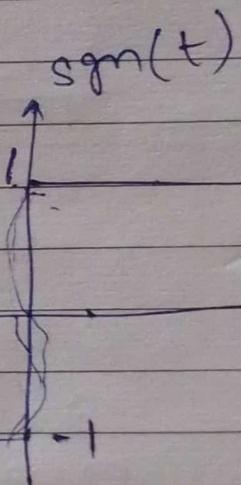
* Signum Signal :-

$$s(t) = 1 ; \quad t \geq 0$$

$$= -1 ; \quad t < 0$$

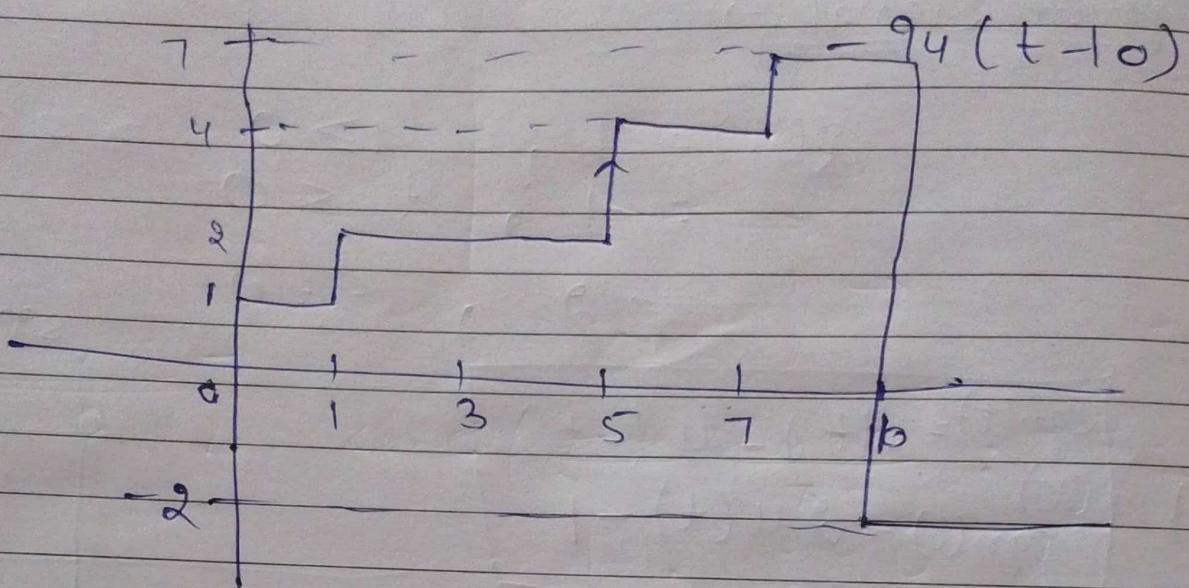
$$= 2 , \quad t=0$$

$$\left. \begin{aligned} sgn(t) &= u(t) - u(-t) \\ sgn(t) &= 2u(t) - 1 \end{aligned} \right\}$$



$$+ \underbrace{① u(t)}_{t=0} + \underbrace{② u(t-3)}_{t=3} \rightarrow \text{amplitude}$$

$$u(t) = u(t) + u(t-1) + 2u(t-5) + 3u(t-7)$$

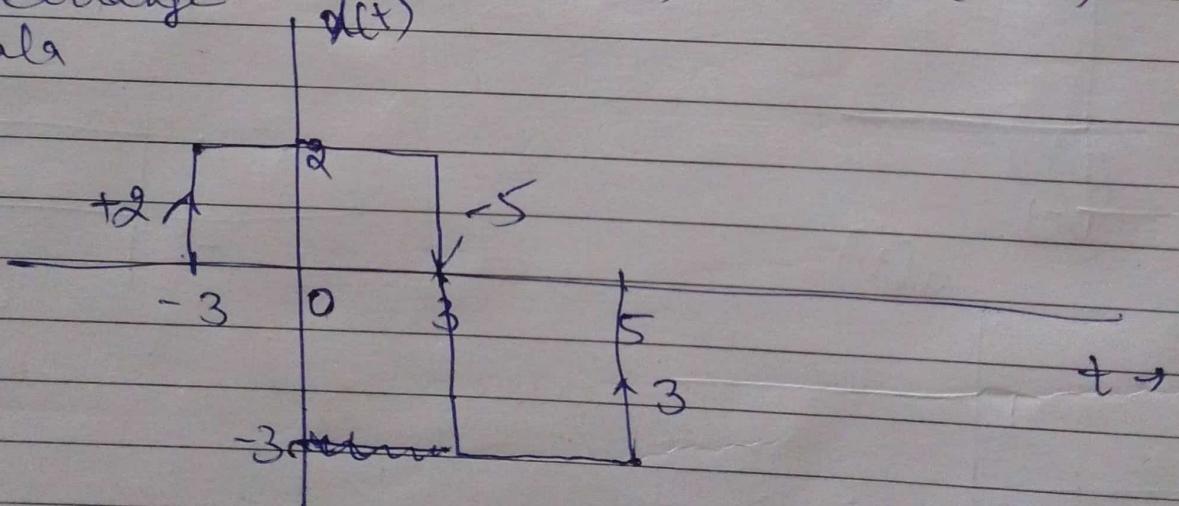


Q: $2u(t+3) - 5u(t-3) + 3u(t-5)$

= first rearrange

Adv. vala

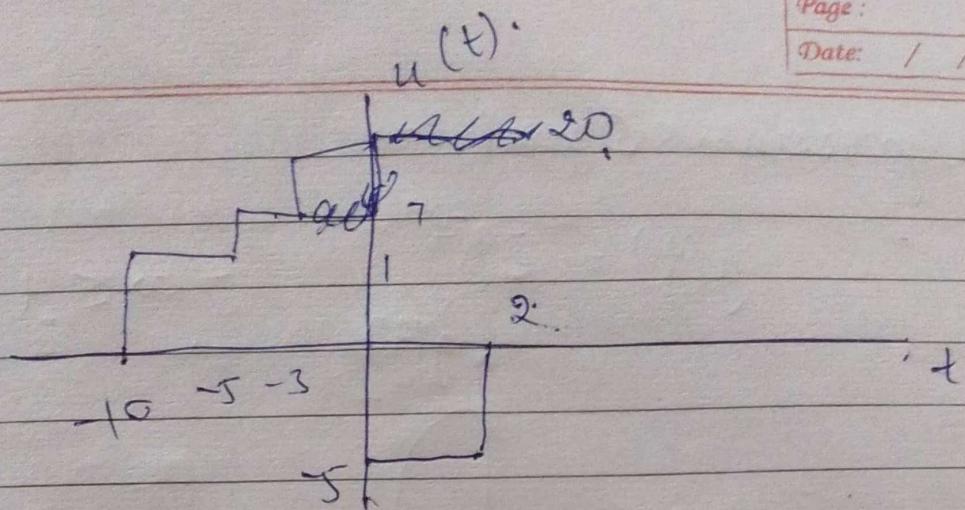
phle



Graph to ep 1-

Start from left

$$2u(t+3) - 5u(t-3) + 3u(t-5)$$



$$u(t+10) + 6u(t+5) + 13u(t+3) \\ - 25u(t) \quad \cancel{+ 5u(t-2)}$$

* Rectangular pulse :-

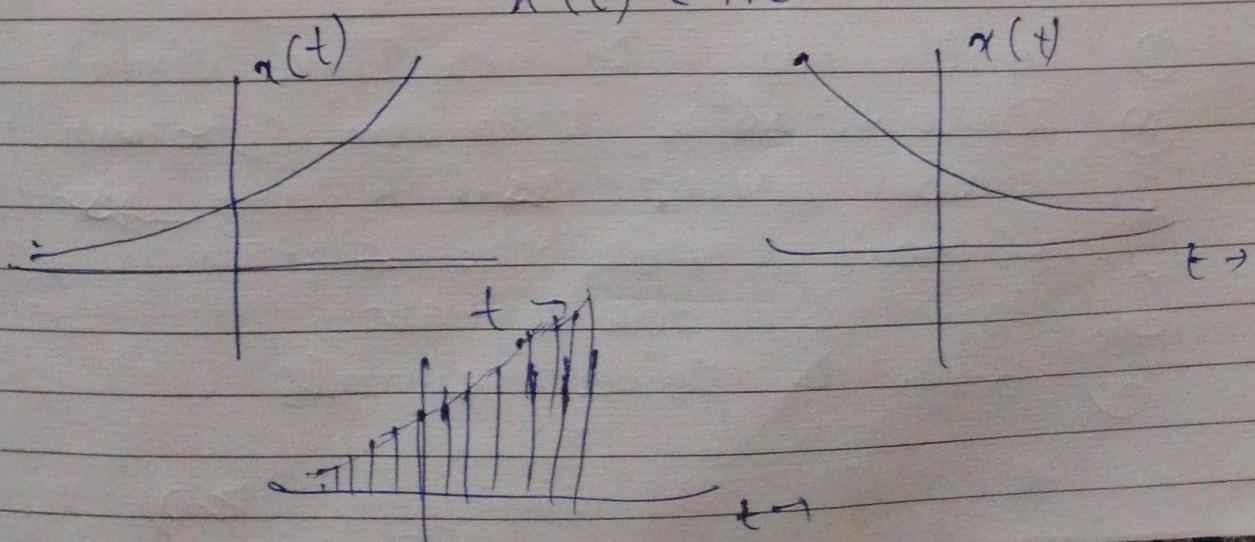
$$\text{rect}\left(\frac{t}{T}\right) \quad \begin{matrix} \uparrow \\ \text{total duration of sec} \end{matrix}$$

$$\text{rect}\left(\frac{t}{T}\right) = 1 ; \quad -\frac{T}{2} \leq t \leq \frac{T}{2}$$

$$= 0 ; \quad \text{otherwise}$$

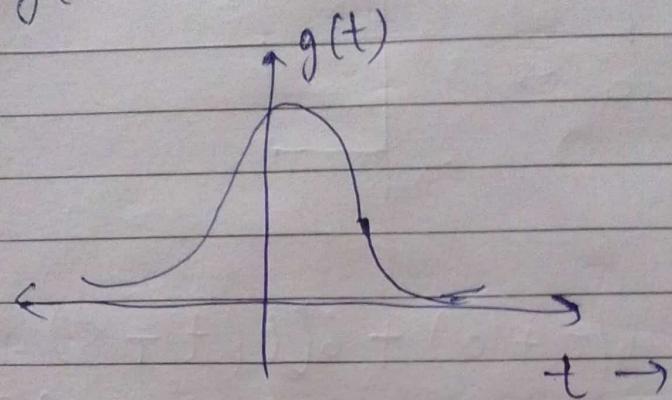
* Exponential Signal :-

$$u(t) = Ae^{bt}$$



* Gaussian Signal :-

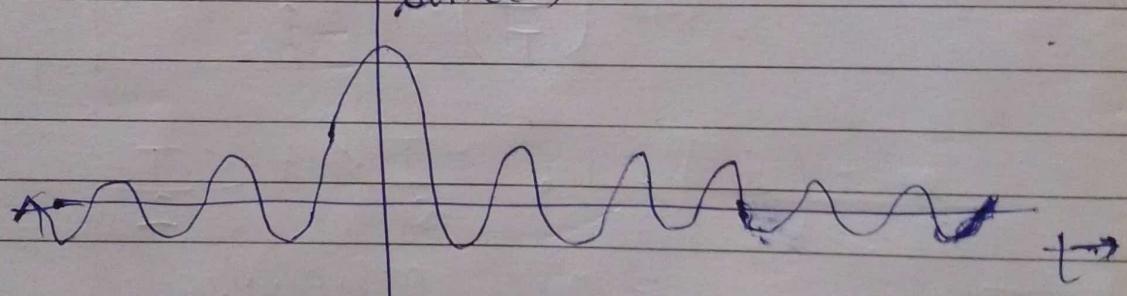
$$g(t) = e^{-\alpha^2 t^2} \quad -\infty \leq t \leq \infty$$



* Sinc. fxn. :-

$$\text{sinc}(t) = \frac{\sin(\pi t)}{\pi t}; \quad t \neq 0$$

$$= 1, \quad t = 0$$



v.v Imp:

Classification of Signals:-

- ① Continuous & Discrete Time signal
- ② Deterministic & Random (Stochastic) signal
- ③ Causal, Non-causal & anti causal signal
- ④ Periodic & Aperiodic signals
- ⑤ Even & odd signals
- ⑥ Energy & power signals.

* Causal (only defined for $t \geq 0$)
Anti causal ($t < 0$ only) $u(-t)$

Multiply with $u(t)$ to make causal signals!

* Even & Odd Signal:-

* If $x(-t) = x(t)$; Even; sym. abt y-axis

$x(t) = -x(t)$; odd; Anti-sym.
 $\sin(t), \text{signum}$

$$* E + E = E$$

$$0 + E = N \in NO \leftarrow \sin t + \cos t$$

$$0 + 0 = 0$$

$$E \times E = E$$

$$E \times 0 = 0$$

$$0 \times 0 = E$$

$$x_e(t) = \frac{x(t) + x(-t)}{2}$$

$$x_o(t) = \frac{x(t) - x(-t)}{2}$$

* Periodic & Aperiodic signals:-

$$x(t+T) = x(t) \rightarrow \text{Periodic}$$

↳ period

Eg:- $\sin t, \cos t$

$u(t) \rightarrow \text{Aperiodic}$

$$x(t) = 5 \sin \frac{\omega}{T} t$$

$\omega = \frac{\pi}{T}$

$\omega = \frac{2\pi}{T} = \frac{\pi}{T}$

$T = \frac{2\pi}{\omega}$

Periodic

$$x(t) = 5 \sin 2\pi t + 3 \cos 8\pi t$$

Method:-
 find ω_1 & ω_2 --

$$\omega_0 = \text{HCF}(\omega_1, \omega_2, \dots)$$

$$T = \frac{2\pi}{\omega_0}$$

(2) Find T_1, T_2, T_3, \dots
 $= \text{LCM}(T_1, T_2, \dots)$

How to check periodic

$$= 5 \sin 2\pi t + 3 \cos 8\pi t$$

$$\frac{\omega_1}{\omega_2} \text{ or } \frac{T_1}{T_2} = \frac{2}{8} = \text{rational}$$

then
periodic

- * Either all terms contains π otherwise no one.

$$5 \sin 2\pi t + 3 \cos 8\pi t$$

$$\text{HCF}(2\pi, 8\pi)$$

$$\Rightarrow 2\pi$$

$$T = \frac{2\pi}{\omega} = 1$$

$$\begin{array}{r} 10 \\ 2 | 5 - 3 \\ \hline 1 \end{array}$$

$$\cancel{\frac{3}{2}}, \cancel{1}, \cancel{1}$$

$$N = \frac{2\pi k}{\omega_0}$$

$$x[n] = \cos 5\pi n$$

$$\cos n = 5\pi$$

$$N = \frac{2\pi k}{\omega_0}$$

$$\text{Perf } x = \frac{2\pi k}{5\pi}$$

$$\boxed{k=5} \quad \boxed{N=2}$$

Q. $x[n] = \cos 5n$

$$\omega_0 = 5$$

$$N = \frac{2\pi k}{5}$$

k must be an integer

it's not integer.

Aperiodic

$$\text{so, } k = \frac{5}{\pi}$$

03/08/23

$$\textcircled{1} \quad x[n] = e^{j\frac{4\pi}{7}n} + e^{j\frac{2\pi}{5}n}$$

$$\omega_0 = \frac{4\pi}{7}, \quad \omega_1 = \frac{2\pi}{5}$$

$$\omega_0 = \text{HCF} \left(\frac{4\pi}{7}, \frac{2\pi}{5} \right)$$

$$\omega_0 = \frac{2\pi}{35}$$

$$N = \frac{2\pi k}{\omega_0}$$

$$N = \frac{2\pi}{2\pi} \times 35k$$

$$N = 35 \text{ sec.}$$

$$\textcircled{2} \quad x[n] = \cos 7\pi n + \cos 5\pi n$$

$$\omega_0 = \text{HCF}(7\pi, 5\pi)$$

$$\omega_0 = \pi$$

$$N = \frac{2\pi k}{\pi}$$

$$N = 2 \text{ sec.}$$

$$\textcircled{3} \quad x[n] = \cos 7\pi n + \cos 5n \rightarrow \text{Aperiodic.}$$

$$E \rightarrow \text{if } E = \text{finite} \& P=0$$

Power ~~if~~ if $P = \text{finite} \& E = \infty$

$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt \quad \text{or} \quad \sum_{n=-\infty}^{\infty} |x(n)|^2$$

$$P = \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt \quad \text{or} \quad \frac{1}{2N+1} \sum_{n=-N}^N |x(n)|^2$$

Just to verify.

- * If $t \rightarrow \infty$, signal value = 0; Energy
 - * If $t \rightarrow \infty$, _____ = finite, Power
 - * If $t \rightarrow \infty$, Neither energy nor power
- $x[t] = A \rightarrow$ Power

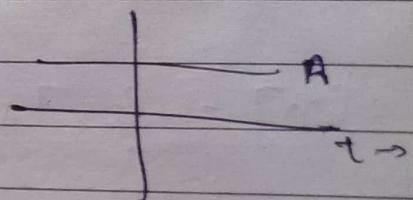
Q: $x(t) = e^{-2t} u(t)$

$$t \rightarrow \infty$$

$$x(t) = 0$$

↳ energy

Q: $x[t] = A \rightarrow$ Power



$$x[t] = e^{-2t} u(t)$$

$$\int_{-\infty}^{\infty} |x[t]|^2$$

$$= \int_{-\infty}^{\infty} (e^{-2t} u(t))^2 dt$$

$$\Rightarrow \int_{-\infty}^0 0 + \int_0^{\infty} e^{-4t} dt = \left[\frac{e^{-4t}}{-4} \right]_0^{\infty}$$

$$\text{Energy} = \frac{1}{4} \text{ Joules}$$

$$P = 0 \text{ W}$$

$$P = \frac{1}{2T} \int_{-T}^T (x(t))^2 dt$$

$$x[t] = A$$

Power = A^2

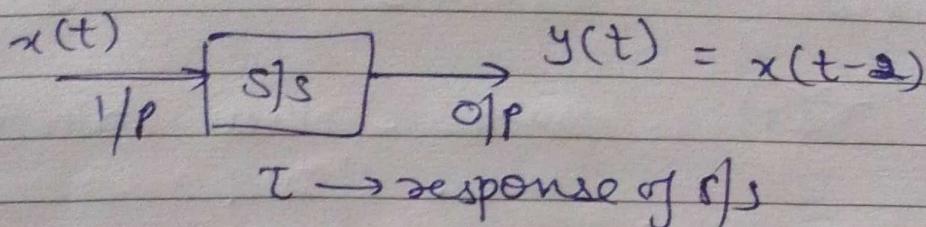
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SNS

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* Classification of System :-



$$y(t) = T[x(t)]$$

$$y(t) = x(t^2)$$

$$y(t) = 2x(t)$$

$$y(t) = x(-t)$$

$$y(t) = \cos t x(t)$$

All these are systems

V.V Imp:→ Qs

Classification of System:-

Type of system to identify

One numerical

graph → $e^{j\omega}$ or $e^{j\omega}$ to graph

① static & Dynamic (Memory or Memoryless) S/s :-
↓
memoryless memory

② Causal & Non causal S/s (Reliable & Non-reliable)

Static :- A system is said to be static if the response at any instance of time depends on i/p only at that instant of time (at present instant of time) Also known as memory-less system.

$t \rightarrow 0$
 $t \rightarrow +ve$
 $t \rightarrow -ve$
(To check)

$$\left. \begin{aligned} y(t) &= 2x(t) \\ y(0) &= 2x(0) \\ y(2) &= 2x(2) \\ y(-3) &= 2x(-3) \end{aligned} \right\} \rightarrow \text{static}$$

(response)

* $y(t) = \cos t x(t) \rightarrow$ static

$y(t) = x^2(t) \rightarrow$ static

$y(t) = x(2t) \rightarrow$ dynamic

Q: $y(t) = x^2(t)$ check whether static, dyn
L|NL, V|IV, -
Dynamic Static (Memoryless)

② Causal & Non-Causal :-

A system is said to be a causal system if the response at any instant of time depends on past & present i/p only.

If it depends upon (past or present) & future i/p. • non-causal.

If it depends upon future i/p only is called Anti-causal.

$y(t) = x^2(t)$ Reliable

$y(0) = x^2(0)$

$y(\alpha) = x^2(\alpha)$

$y(3) = x^2(3)$

$y(t) = x(2t) \rightarrow$ Non-causal

$y(t) = x\left(\frac{t}{2}\right) =$ Non-reliable
N.C.

$y(t) = \cos t x(t) \rightarrow c$

$y(t) = y(t-2) = c$

$y(t) = y(t+5) = N.c$

* ③ Linear & Non-Linear S/I S:-

write all steps in numerical:-

$$\textcircled{1} \quad y(t) = x^2(t)$$

$$\textcircled{2} \quad x_1(t) \xrightarrow{\tau} y_1(t) = x_1^2(t)$$

$$\textcircled{3} \quad x_2(t) \xrightarrow{\tau} y_2(t) = x_2^2(t)$$

$$\textcircled{4} \quad \text{if } y_1(t) + y_2(t) = \tau [x_1(t) + x_2(t)]$$

↑
(Response)

$$\text{Also, } \underset{\substack{\uparrow \\ \text{(homogeneity prop.)}}}{a x_1(t)} \xrightarrow{\tau} a y_1(t) = a x_1^2(t)$$

$$a x_1(t) + b x_2(t) \xrightarrow{\tau} a y_1(t) + b y_2(t)$$

$$\text{if } x_1^2(t) + x_2^2(t) = [x_1(t) + x_2(t)]^2$$

\Rightarrow false \Rightarrow N.L.

If equal then Linear

$$y_1(t) + y_2(t) = \tau [x_1(t) + x_2(t)]$$

$$\text{Q: } y(t) = \pi x(t)$$

$$\pi x_1(t) + \pi x_2(t) = \pi [x_1(t) + x_2(t)]$$

\Rightarrow linear

$$\text{Q: } y(t) = \pi x(t) + c$$

$$\pi x_1(t) + c + \pi x_2(t) + c = \pi [x_1(t) + x_2(t)] + c$$

\Rightarrow N.Linear

(4) Time Variant & Invariant system:-

If response of system is changing with time it is variant system.

If delayed response = response of S/s to delayed i/p

$$y(t-t_0) = \tau [x(t-t_0)]$$

Then time
Invariant

(Shift Invariant)

Q:-

$$y(t) = 2t x(t)$$

$$y_1(t-t_0) x_1(t) = y_1(t) = 2t x_1(t)$$

$$x_1(t-t_0) = y_1(t-t_0) = 2(t-t_0) \\ x_1(t-t_0)$$

$$2(t-t_0) x_1(t-t_0) = 2(t_0) [x(t-t_0)]$$

Invariant

Q:- $y(t) = 3x^2(t)$

$$y(t-t_0) = 3x^2(t-t_0)$$

$$x(t) \xrightarrow{t_0} x(t-t_0) \xrightarrow{\tau} 3x^2(t-t_0)$$

$$3x^2(t-t_0) = 3(x(t-t_0))^2 \\ = 3x^2(t-t_0)$$

Invariant

Q:- $y(t) = 3x(t^2)$

$$y(t-t_0) = 3x((t-t_0)^2) \\ \Rightarrow 3x((t-t_0)^2)$$

$$x(t) \xrightarrow{t_0} x(t-t_0) \xrightarrow{T} 3x((t^2 - t_0^2))$$

\Rightarrow Invariant

Q: $y(t) = x(-t)$

$$y(t-t_0) = x(-t+t_0)$$

$$x(t) \xrightarrow{t_0} x(t+t_0) \xrightarrow{T} x(t_0-t_0)$$

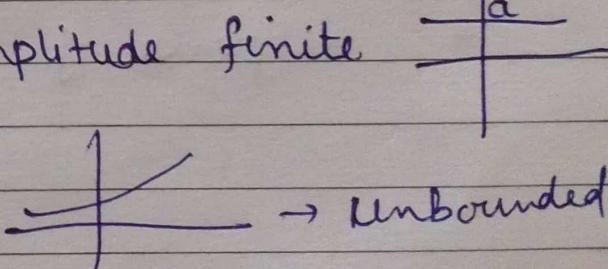
\Rightarrow Invariant

Shortcut! - t p1 operation no shg h \rightarrow variant
 $x(t) \xrightarrow{\quad}$ Invariant

* $y(t) = x(t) \rightarrow$ variant

(5) Stable & Unstable system:- A system is said to be a stable system, if bounded I/p results in bounded o/p.
If BIBO \rightarrow stable

bounded \rightarrow signal ka amplitude finite



Example :-

$$y(t) = \cos(x(t))$$

\downarrow
-1 to 1

Put $x(t) = u(t)$

If bounded o/p \rightarrow stable

\Rightarrow bounded \rightarrow stable

Unbounded \rightarrow unstable

$$y(t) = \cos t x(t)$$

\hookrightarrow bounded \rightarrow stable

$$y(t) = \frac{x(t)}{\cos t} = \text{unbounded} \rightarrow \text{unstable}$$

$$y(t) = + (x(t)) = + (u(t)) \Rightarrow y(t) = \text{unstable}$$

$$y(t) = x^2(t) = \text{stable}$$

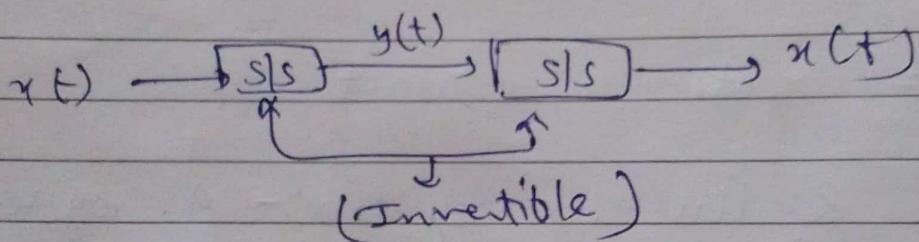
⑥ Invertible or Non-Invertible:-

If ~~one~~ inverse distinct i/p results in distinct o/p then the system is said to be invertible.

And if distinct i/p result in ~~the~~ same o/p then the system is non-invertible

- * $y(t) = x^2(t)$

Non-Invertible



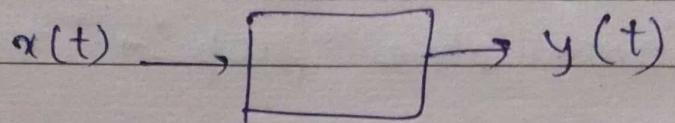
- * Response of unit Impulse system is $h(t)$ - impulse response

- * Response of step Impulse system is $s(t)$ step response.

07/08/23

SNS

Convolution Integral:- is a mathematical tool for finding the response of any LTI system for any arbitrary i/p $x(t)$ provided the impulse response is known.



$$y(t) = \mathcal{Z}[x(t)]$$

$$\delta(t) \rightarrow [s/s] \rightarrow h(t)$$

$$u(t) \rightarrow [s/s] \rightarrow S(t)$$

$$y(t) = x(t) * h(t) \quad (\text{convolution})$$

$$y(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$$

Proof :-

$$x(t) = \int_{-\infty}^{\infty} x(z) \delta(t-z) dz$$

(Prop. of imp. signal)

$$y(t) = \mathcal{Z}[x(t)] = \mathcal{Z} \left[\int_{-\infty}^{\infty} x(z) \delta(t-z) dz \right]$$

$$\Rightarrow \int_{-\infty}^{\infty} \mathcal{Z} \left[x(z) \delta(t-z) dz \right]$$

$$= \int_{-\infty}^{\infty} x(z) \mathcal{Z} [\delta(t-z)] dz$$

$$y(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz = x(t) * h(t)$$

If

$$y[n] = x[n] * h[n] = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$

* Properties of Convolution :-

① Commutative prop : $x(t) * h(t) = h(t) * x(t)$

L.H.S $x(t) * h(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$

let $t-z=p$ $dz=$

$$= \int_{-\infty}^{\infty} x(t-p) h(p) dp$$

$$= \int_{-\infty}^{\infty} h(p) x(t-p) dp$$

$$= h(t) * x(t)$$

② Associative Prop :- * connecting in cascade
 Cascade Interconnection

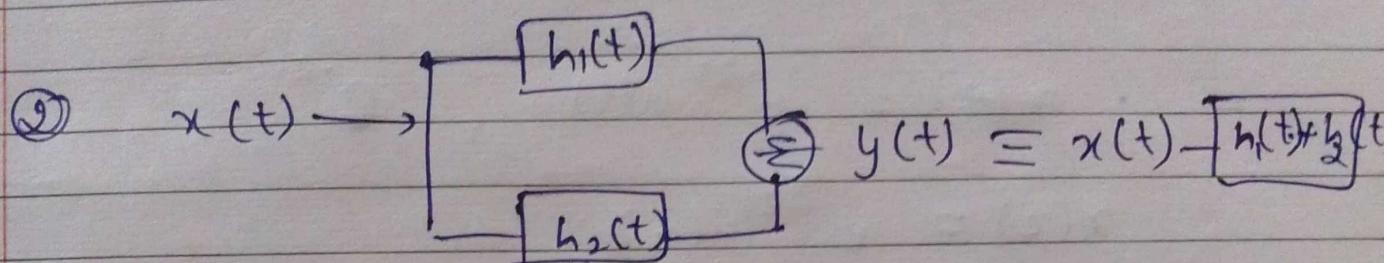
$$[x_1(t) * x_2(t)] * h(t) = x_1(t) * [x_2(t) * h(t)]$$

③ Distributive Prop :- Parallel Interconnection.

$$x(t) * [h_1(t) + h_2(t)] = [x(t) * h_1(t)] + [x(t) * h_2(t)]$$

① $x(t) \rightarrow [h_1(t)] \rightarrow [h_2(t)] \rightarrow y(t)$

$\equiv x(t) \rightarrow [h_1(t) * h_2(t)] \rightarrow y(t)$



Steps to solve numericals :-

Step 1:- change of time index

① $t \rightarrow z$ $x(z)$
 $h(z)$

② $h(z) \rightarrow h(-z)$

③ shifting $\rightarrow h(t-z)$

④ Multiplication $x(z) h(t-z)$

⑤ Integration

Q:- $x(t) = 2u(t)$
 $h(t) = u(t)$

find response?

" convolution?

$u(t) \rightarrow$ causal

If both causal, then it changes to \int_0^t

$x(t) = 2u(t) = 2, t \geq 0$

$h(t) = u(t) = 1, t \geq 0$

$y(t) = \int_0^t x(z) h(t-z) dz = \int_0^t 2 \cdot 1 dz = \int_0^t 2 dz$

$$= 2t, t \geq 0 = 2 + u(t)$$

Q: $x(t) = e^{-2t} u(t)$
 $h(t) = e^{-5t} u(t)$

$$\therefore x(t) = e^{-2t} u(t), e^{-2t}$$

$$x(z) = e^{-2z} u(z)$$

$$h(t-z) = e^{-5(t-z)} u(t-z)$$

$$\Rightarrow x(t) = e^{-2t}, t \geq 0$$

$$h(t) = e^{-5t}, t \geq 0$$

$$\Rightarrow y(t) = \int_0^t x(z) h(t-z) dz$$

$$\Rightarrow y(t) = \int_0^t e^{-2z} e^{-5(t-z)} dz$$

$$\Rightarrow y(t) = \int_0^t e^{-2z} e^{-5t} e^{5z} dz$$

$$= e^{-5t} \int_0^t e^{3z} dz$$

$$\Rightarrow e^{-5t} \left[\frac{e^{3z}}{3} \right]_0^t$$

$$\Rightarrow \frac{e^{-5t}}{3} \cdot (e^{3t} - 1)$$

$$= \frac{1}{3} [e^{-2t} - e^{-5t}] u(t)$$

* Std. form answer is must

11/08/23

Graphical Method \rightarrow (Always when not mentioned)

$$x(t) = u(t)$$

$$h(t) = u(t)$$

$$y(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$$

$$x(z) = u(z)$$

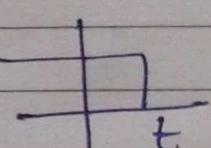
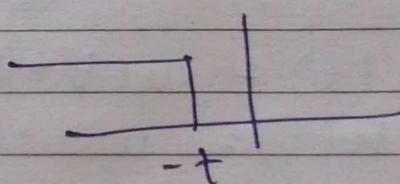
$$h(z) = u(z), \quad h(t-z) = u(t-z)$$

$$= \int_0^{\infty} u(t-z) dz = \int_0^{\infty} u(-z-t) dz$$

$$\begin{cases} t < 0 \\ u(-z-t) \end{cases}$$

$$\begin{cases} t > 0 \\ u(-z+t) \end{cases}$$

$$y(t) = \int_0^{\infty} dz, t \geq 0$$



$$= 0$$

$$y(t) = t, t \geq 0$$

$$= 0, t < 0$$

$$y(t) = t u(t)$$

$$= \delta(t)$$

Note:- If we convolve unit step signal with unit step signal the resultant signal will be same signal.

$$x(t) = u(t+3)$$

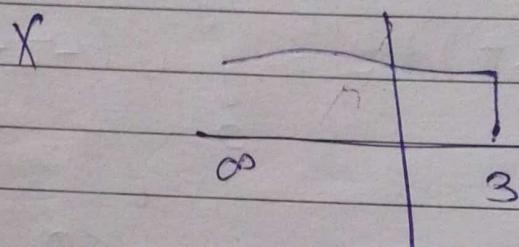
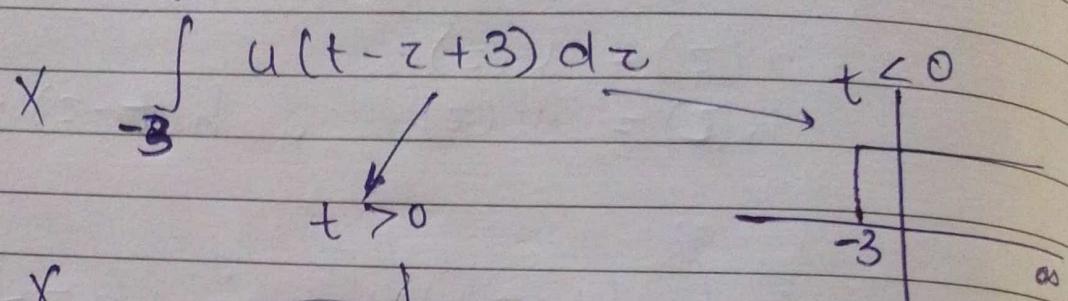
$$h(t) = u(t+3)$$

$$y(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$$

$$x(z) = u(z+3), h(z) = u(z+3)$$

$$X \quad h(t-z) = u(t-z+3)$$

$$\int_0^\infty u(z+3) u(t-z+3) dz$$



$$y(t) =$$

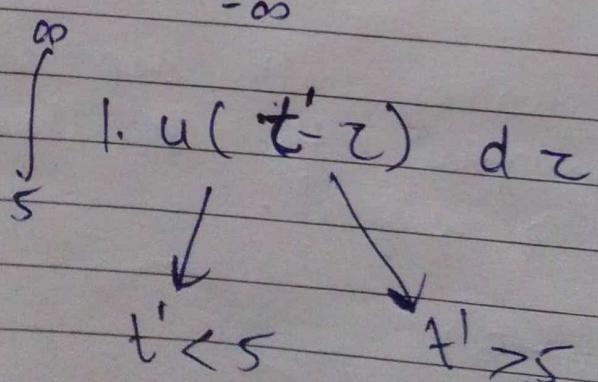
Q.

$$x(t) = u(t+3)$$

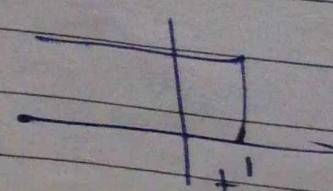
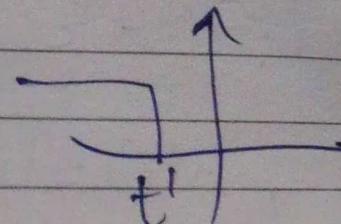
$$h(t) = u(t-5)$$

$$y(t) = \int_{-\infty}^t x(z) h(t-z) dz$$

$$= \int_{-\infty}^{\infty} u(z-5) u(-z+t+3) dz$$



$$y(t) = \int_s^{t'} 1 \cdot dz, \\ = 0,$$



$$y(t) = t^1 - 5 \quad t' \geq 5
= 0 \quad \text{otherwise}$$

$$y(t) = t^1 - 5, \quad t + 3 \geq 5
= t + 3 - 5, \quad t \geq 2
= t - 2$$

$$y(t) = t - 2 \quad \text{for } t \geq 2
= 0 \quad \text{otherwise}$$

Q. $x(t) = u(t+7)$
 ~~$h(t) = u(t+2)$~~

~~NOTE~~
 $u(t) * u(t) = \gamma(t)$
 $u(t \pm \alpha) * u(t \pm \beta) = \gamma(t \pm \alpha \pm \beta)$

$$y(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$$

$$\int_{-\infty}^{\infty} u(t+2) u(-z+t+7) dz$$

In discrete \Rightarrow $x[n] = u[n+2]$
 $h[n] = u[n-3]$
 $x[n] = \cancel{x}[n-1]$

* $[x[n] \delta[n]] = x[n]$

* $[x[n+2] \delta[n-1]] = x[n+1]$

$$\Rightarrow \int_{-2}^{\infty} 1 \cdot u(t'-z) dz$$

$t' < -2$ $t' \geq -2$

\Rightarrow graphical representation

$$y(t) = \begin{cases} 1, & t \geq -2 \\ 0, & \text{otherwise} \end{cases}, \quad t \geq -2$$

$$y(t) = \begin{cases} t+2, & t+7 \cancel{\geq} -2 \\ t+7+2 \\ t+9, & t \geq -9 \end{cases}$$

$$y(t) = \begin{cases} t+9, & \text{for } t \geq -9 \\ 0, & \text{otherwise} \end{cases}$$

$$\sum_{-1}^t 1 = 3 \quad (1+1+1)$$

[Upper Lt - lower Lt]

* Convolution Proof } ②

* Numerical
* Discrete value } (***)

$$x[n] = u[n+2]$$

$$h[n] = u[n-3]$$

(***)

$$x[n] = [1, 2, 3, 2]$$

$$h[n], [1, 0, 1, 2]$$

$$y[n] = ?$$

(Arrow is given other left most)

$$x[n] = \delta[n+1] + 2\delta[n] + 3\delta[n-1] + 2\delta[n-2]$$

$$h[n] = \delta[n+2] + \delta[n] + 2\delta[n-1]$$

$$y[n] = x[n] * h[n]$$

$$y[n] = \delta[n+1] + 2\delta[n] + 3\delta[n-1] + 2\delta[n-2] *$$

$$\delta[n+2] + \delta[n] + 2\delta[n-1]$$

$$x[n] * f[n] = x[n]$$

$$\sum_n [n+\alpha] * f[n] = x[n+\alpha]$$

$$\sum_n [n+\alpha] * \delta[n+\beta] = x[n+\alpha \pm \beta]$$

$$y(n) = 8[n+3] + 6\delta[n+1] + 2\delta[n] + 2\delta[n+2] \\ + 2\delta[n] + 4\delta[n+1] + 3\delta[n+1] + 3\delta[n-1] \\ + 6\delta[n-2] + 2\delta[n] + 2\delta[n-2] + 4\delta[n-3]$$

$$\Rightarrow \underbrace{8[n+3]}_{y(0)}, \underbrace{4\delta[n+1]}_{y(1)}, \underbrace{6\delta[n]}_{y(2)}, \dots$$

$$y(n) = [1, 2, 4, 6, 7, 8, 4]$$

* Tabular method :-

$$* x[n] = [1, 2, 3, 2] = m \text{ terms}$$

$$h[n] = [1, 0, 1, 2] = n \text{ terms}$$

	1	2	3	2
1	1	2	3	2
0	0	0	0	0
1	1	2	3	2
2	2	4	6	4

$$y(n) = [1, 2, 4, 6, 7, 8, 4]$$

$$\text{No. of terms in } y(n) = m - 1 + n$$

$$4 + 4 - 1 = 7$$

$$Q. \quad x[n] = [1, 2, 4, 3, 2]$$

$$h[n] = [1, 0, 1]$$

Tabular method :-

$$y(n) = [1, 2, 5, 5, 6, 3, 2]$$

	1	2	4	3	2
1	1	2	4	3	2
0	0	0	0	0	0
1	1	2	4	3	2
1	2	4	3	2	1

No. of terms :-

$$M + N - 1$$

$$= 5 + 3 - 1 = 7$$

conditions 1.5 marks)
② cont.

Static | Dynamic LTI system:- A discrete sys. is said to be static if $h[n]=0$ for $n \neq 0$ (memoryless). Such a system has impulse response $h[n] = k\delta[n]$ where $k = h[0]$ is a constant & $y[n] = kx[n]$ where $k = h[0]$ is a const. & $y[n] = kx[n]$.

* Causality:- $y(t) = x(t) * h(t) = h[t] \star x[t]$

$$= \int_{-\infty}^{\infty} h(z)x(t-z)dz$$

$$\Rightarrow h(\infty)x(t-\infty) + \dots h(1)x(t-1) + h(0)x(t) + \underbrace{h(-1)x(t+1)}_{\text{future values}} - \dots$$

If $h[n]=0$ for $n < 0$
if $h(t)=0$ " $t < 0$
then our system is called causal
we can't make 0 to i/p, we can
make 0 response.

A sys. is said to be causal if its impulse response is 0 for -ve values of t or n.

* Stability for LTI:-

A system is said to be stable if its impulse response is absolutely integrable. Then system is stable.
* LTI hoge hi hoga.

18/08/23

Page:

Date:

$$x[n] = [1, 2, 1, 1]$$

$$h[n] = [1, 2, 1, -1]$$

$$y[n] = \sum_{k=-1}^3 x[k] h[n-k] = 1 \cdot 1 + 2 \cdot 2 + 1 \cdot 1 - 1 \cdot (-1) = 6$$

By Tabular method,

$$y[n] = \{1, 4, 6, 4, 1, 0, -1\}$$

↑
-1 0 1 2 3 4 5

$$y[n] = \sum_{n=-\infty}^{\infty} x[n] h[n-k]$$

$$y[n] = \sum_{n=-1}^5 x[k] h[n-k]$$

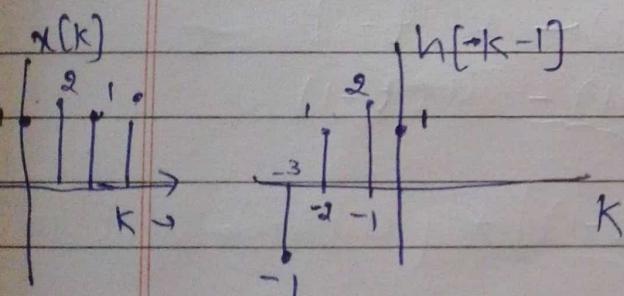
	1	2	1	1
1	X	2	X X	
2	2	4	2	2
1	X	2	1	Y
-1	-1	-2	1	-X

Put $n = -1$

$$y[-1] = x[k] h[-1-k]$$

$$= x[k] h[-k-1]$$

$$h[k-1] = [1, 2, 1, -1]$$

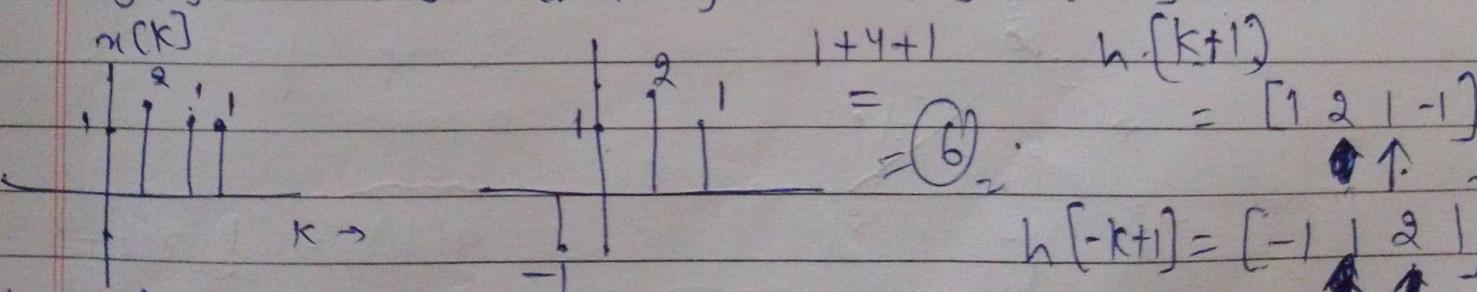


$$h[-k-1] = [-1, 1, 2, 1]$$

(Swap)

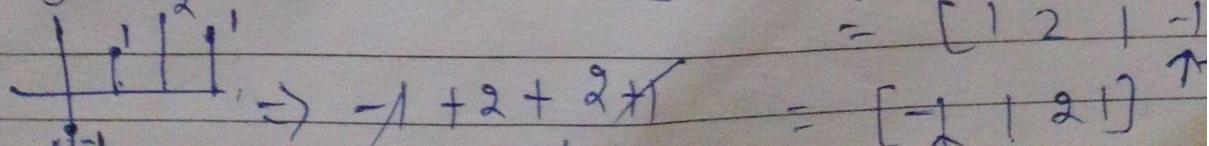
Put $n = 1$

$$y[1] = x[k] h[1-k] = x[k] h[-k+1] =$$



Put $n = 2$

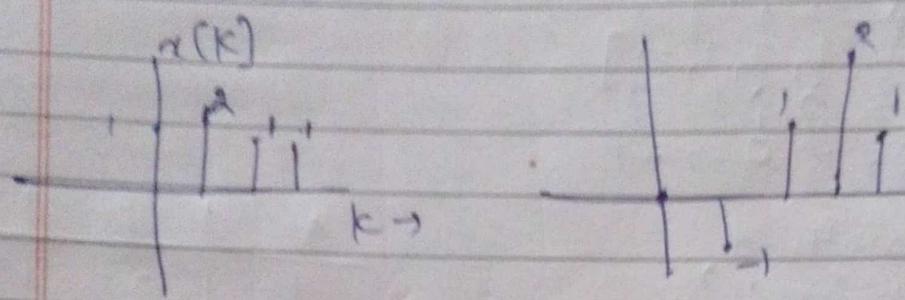
$$y[2] = x[k] h[2-k] = h[k+2]$$



$$y(3) = h[3-k]$$

$$n=3$$

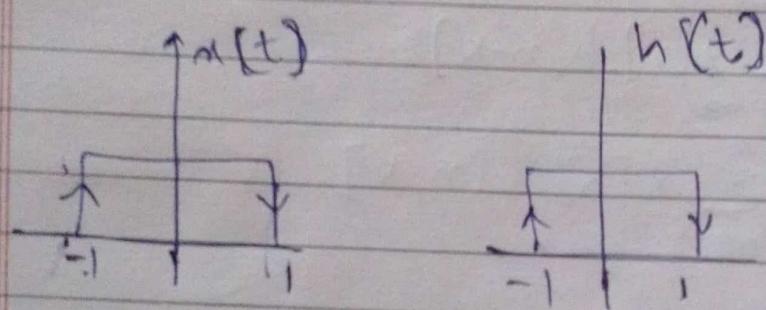
$$x(k)$$



$$-2 + 1 + 2 = 1$$

Verify by $y(n)$

Q:



$$y(t) = ?$$

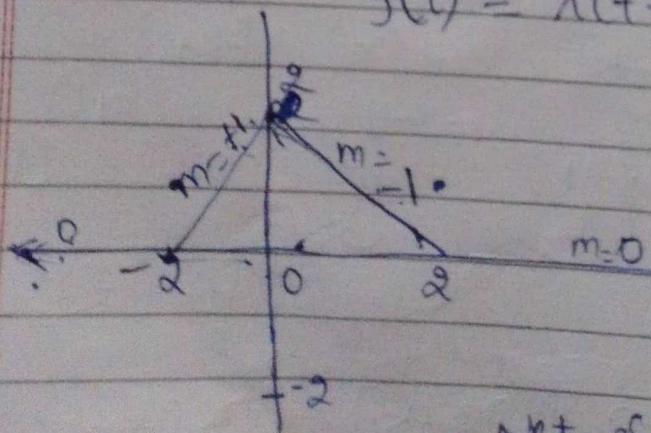
$$x(t) = u(t+1) - u(t-1)$$

$$h(t) = u(t+1) - u(t-1)$$

$$y(t) = [u(t+1) - u(t-1)] * [u(t+1) - u(t-1)]$$

$$y(t) = r(t+2) - r(t) - r(t) + r(t-2)$$

Ramp func me Slope mention.

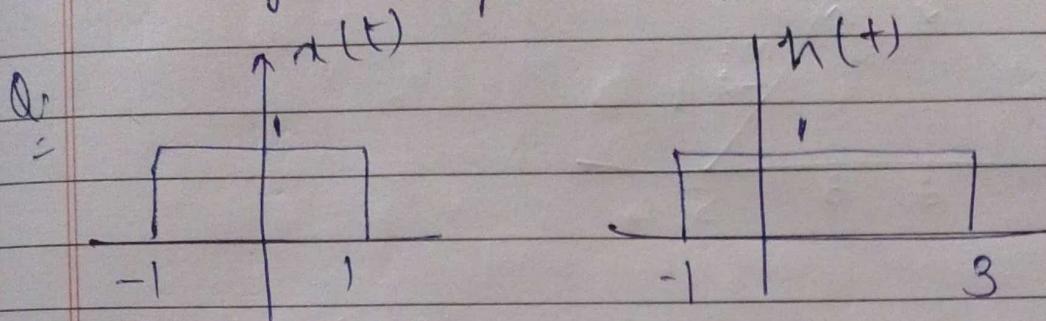


$$y(t) = \frac{1}{2}r(t+2)$$

Net slope = final slope - initial slope

$$y(t) = \alpha(t+2) - 2\alpha(t) + \gamma(t-2)$$

* If two rec. pulses of duration t are convolved then the resultant signal is a triangular pulse of duration $(2t)$



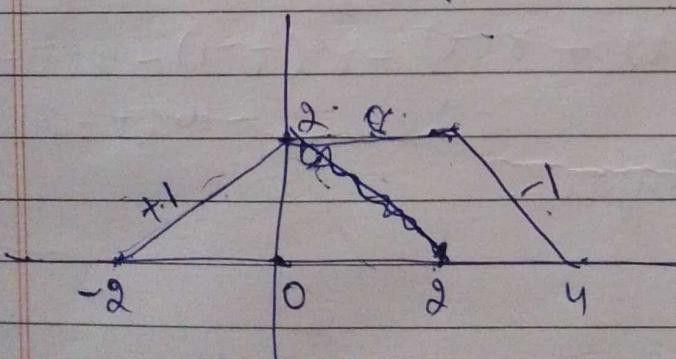
$$x(t) = u(t+1) - u(t-1)$$

$$h(t) = u(t+1) - u(t-3)$$

$$y(t) = x(t) * h(t)$$

$$y(t) = [u(t+1) - u(t-1)] * [u(t+1) - u(t-3)]$$

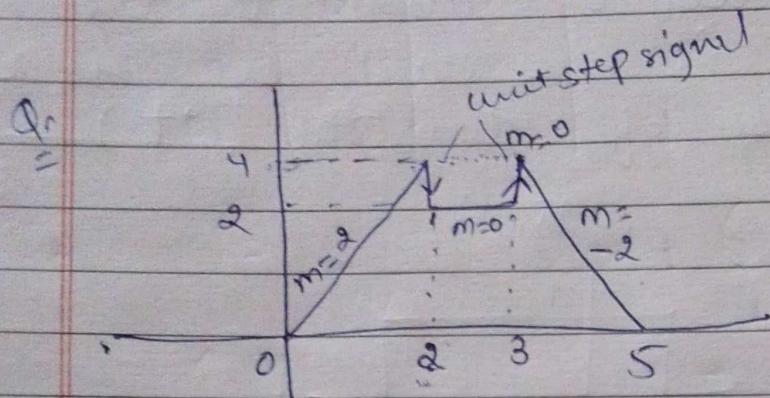
$$y(t) = \gamma(t+2) - \gamma(t-2) - \gamma(t) + \gamma(t-4)$$



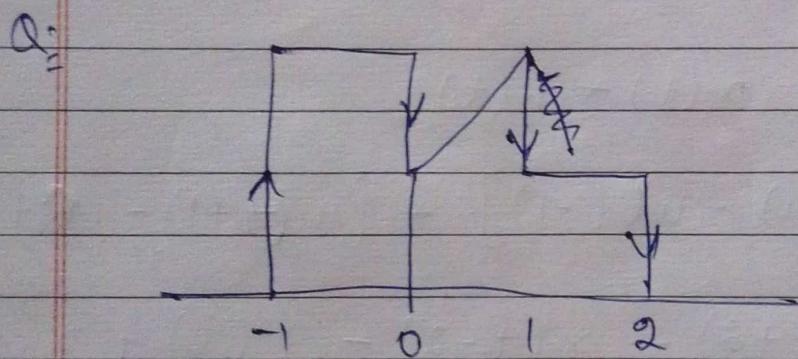
* If we convolve 2 rec. pulses of duration t_1 & t_2 then, the resultant signal will be a trapezoidal pulse of duration $(t_1 + t_2)$

* The starting pt. of resultant signal $y(t)$ is equal to the sum of starting pt. of $x(t)$ & $h(t)$.

* The ending of $y(t)$ is equal to the sum of constant ending pt. of $x(t)$ & $h(t)$.



$$\Rightarrow 2x(t) - 2x(t-2) - 2u(t-2) + 2u(t-3) - 2x(t-3) + 2x(t-5)$$



$$\Rightarrow 2u(t+1) - u(t) + \gamma(t) - \gamma(t-1) - u(t-1) - u(t-2)$$

Book:- offerman , A. Nagur .

21/08/93

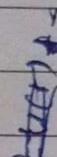
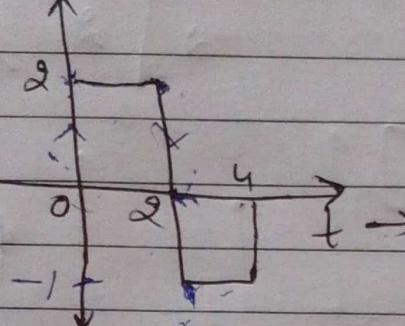
$$s(t) = \int h(t) dt$$

$$u(t) = \int \delta(t)$$

✓ V.V. Imp

Q: Laplace transform of $\delta(t)$, $u(t)$, $x(t)$.

Q:

 $x(t)$  $\delta(t-0)$ $2u(t) - 3u(t-2)$ $+ u(t-4)$

⇒

Q: $x(t) = 5 \sin 2\pi t + 8 \cos 7\pi t$ is periodic

It's periodic! -

$$T_1 = \frac{2\pi}{w}$$

$$T_2 = \frac{2\pi}{7\pi}$$

$$T_1 = \frac{2\pi}{2\pi} = 1, T_2 = \frac{2\pi}{7\pi}$$

$$T_0 = \text{LCM}(T_1, T_2)$$

$$\text{LCM}(1, \frac{2}{7}) = \frac{2}{7} \approx$$

Q: $y(t) = \cos t x(t)$ = ~~non causal~~

⇒ Static

⇒ ~~Non linear~~ ~~linear~~

⇒ Stable

⇒ $T \in V$ (+ by operation ^{has to} _{various})linear - $y_1(t) = \cos t x_1(t)$ $y_2(t) = \cos t x_2(t)$

$$\cos t x_1(t) + \cos t x_2(t) = \cos(t_1 + t_2)$$

$$(x_1 + x_2) = y_1(t) + y_2(t)$$

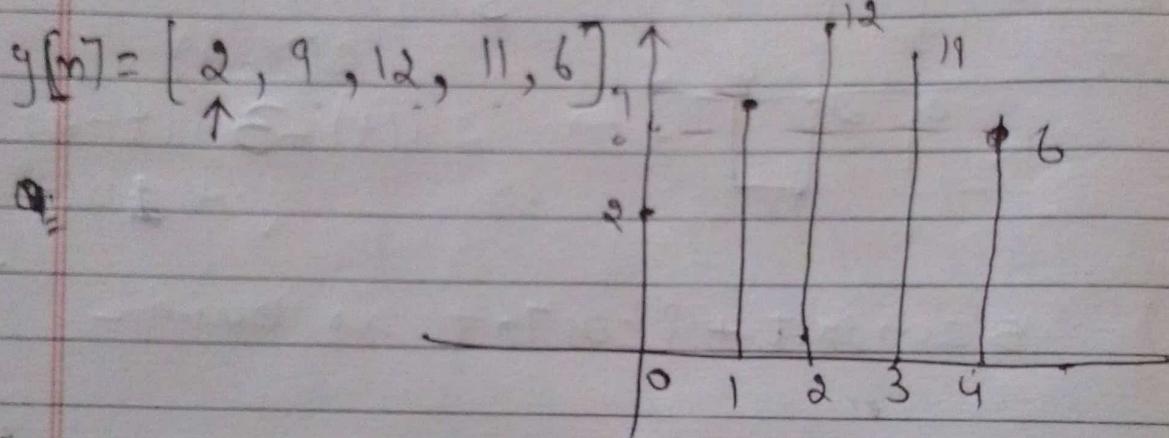
$$\cos [x_1(t) + x_2(t)] = \cos t x_1(t) + \cos t x_2(t) \quad (\text{linear})$$

$$\text{Q: } x[n] = [1, 4, 3]$$

$$h[n] = [2, \underset{\uparrow}{1}, 2]$$

(By Tabular)

	1	4	3
2	8	6	
1	4	3	
2	8	6	



$$\text{Q: } y(t) = \frac{d}{dt} x(t)$$

$$\text{Q: } y(t) = e^{x(t)}$$

$$\text{Q: } y(t) = \log x(t)$$

$$\text{Q: } y(t) = \int x(t)$$

$$\text{Q: } y(t) = x(-t)$$

25/08/23

Laplace Transform: is used to transform a time domain signal into complex frequency domain.

Continuous Time signal $\xrightarrow{\text{Laplace Transform}}$ Complex freq. domain

Discrete Time signal $\xrightarrow{\text{Z-transform}}$ freq.

$$s = \sigma + j\omega$$

σ freq. in rad/sec
 $j\omega$ Laplace freq. in nepes/sec.

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

Laplace [x(t)]

V.V.V.V.V Imf

If this integration has finite value then it exists.

$$\text{Q1: } x(t) = e^{-at} u(t)$$

$$\text{Q2: } n(t) = -e^{-at} u(t)$$

$$\textcircled{1}: x(s) = \int_{-\infty}^{\infty} e^{-at} u(t) e^{-st} dt$$

$$\Rightarrow \int_{-\infty}^{\infty} e^{-at} u(t) e^{-st} dt + \int_{-\infty}^{\infty} e^{-at} u(t) e^{-st} dt$$

$$0 + \int_{-\infty}^{\infty} e^{-at} e^{-st} dt$$

Unit 1 + Unit 2

+ Laplace

Date:

+ Test Unit

$$\Rightarrow \int_0^\infty e^{-(a+s)t} dt$$

$$\Rightarrow \left[\frac{e^{-(a+s)t}}{-(a+s)} \right]_0^\infty$$

$$\Rightarrow \frac{1}{s+a}$$

Q:

$$x(t) = -e^{-at} u(-t)$$

$$X(s) = \int_{-\infty}^0 -e^{-at} u(-t) e^{-st} dt + 0$$

$$\Rightarrow \int_{-\infty}^0 -e^{-at} e^{-st} dt$$

$$= - \int_{-\infty}^0 e^{-(a+s)t} dt$$

$$\Rightarrow - \left[\frac{e^{-(a+s)t}}{-(a+s)} \right]_{-\infty}^0$$

$$\Rightarrow - \left[\frac{1}{-s-a} + 0 \right] = \frac{1}{s+a}$$

Laplace ~~function~~ exist.

R.O.C: Region of convergence:- It includes those values of sigma $\operatorname{Re}[s]$ for which the Laplace transform converges.
↳ (finite value)

If $u(t) \in$ to put

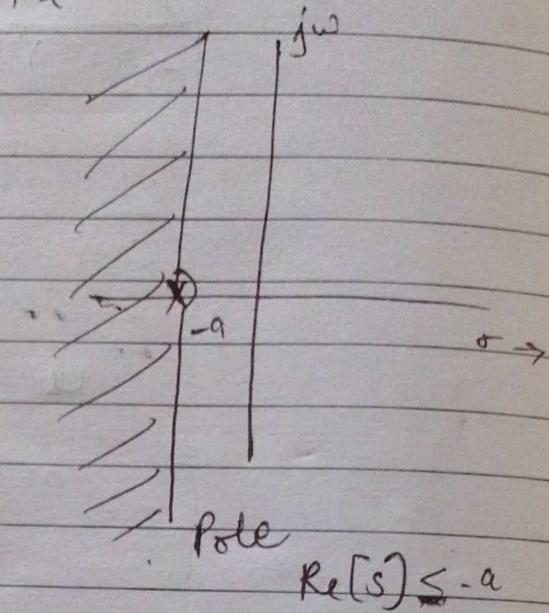
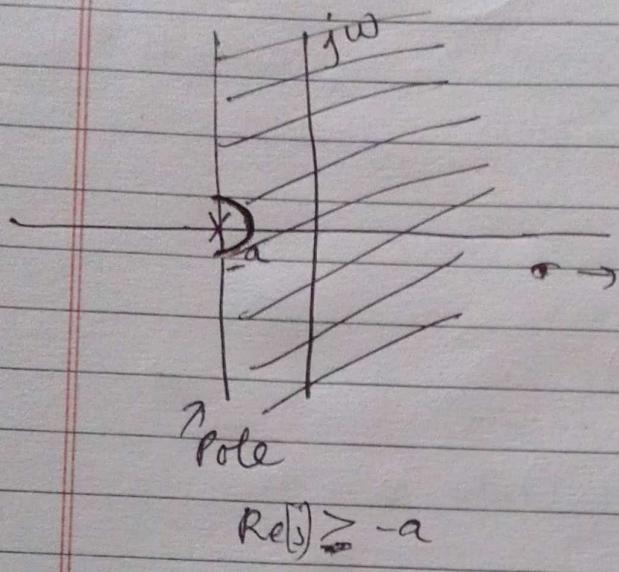
$$\text{den} \geq 0$$

$$s+a < 0$$

If $u(-t) \in$ to put den ≤ 0

* LT of $e^{-at} u(t) = \frac{1}{s+a}$, $\text{Re}[s] \geq -a$

* LT of $-e^{-at} u(-t) = \frac{1}{s+a}$, $\text{Re}[s] \leq -a$



$$X(s) = \frac{p}{q} = \frac{0}{0} \rightarrow \text{zero}$$

pole

$$X(s) = \frac{(s+2)(s+3)}{s(s^2 + 6s + 13)}$$

* $(\text{Put num} = 0)$ = $(s+2)(s+3) = 0$
 for zeros

$$s = -2, -3$$

* $(\text{Put den.} = 0)$ = $s(s^2 + 6s + 13) = 0$
 for pole

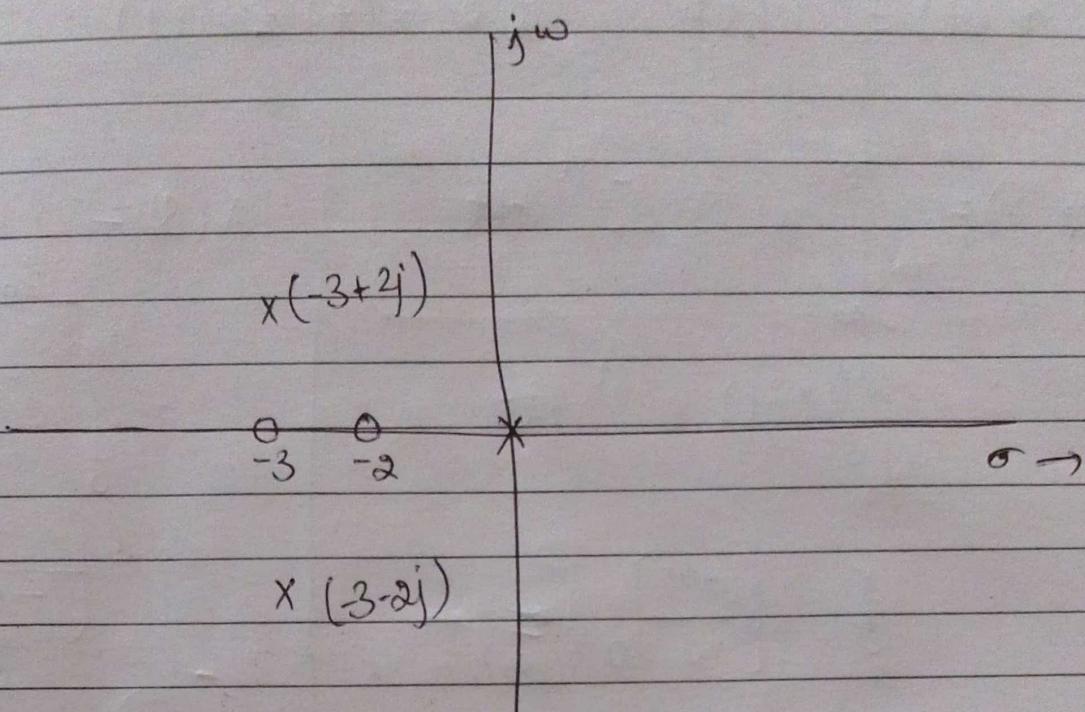
$$\frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$$

$$= \frac{-6 \pm \sqrt{36 - 52}}{2}$$

$$= -3 + \frac{2i}{2}, -3 - \frac{2i}{2}, 0$$

zero $\rightarrow 0$

pole $\rightarrow \infty$

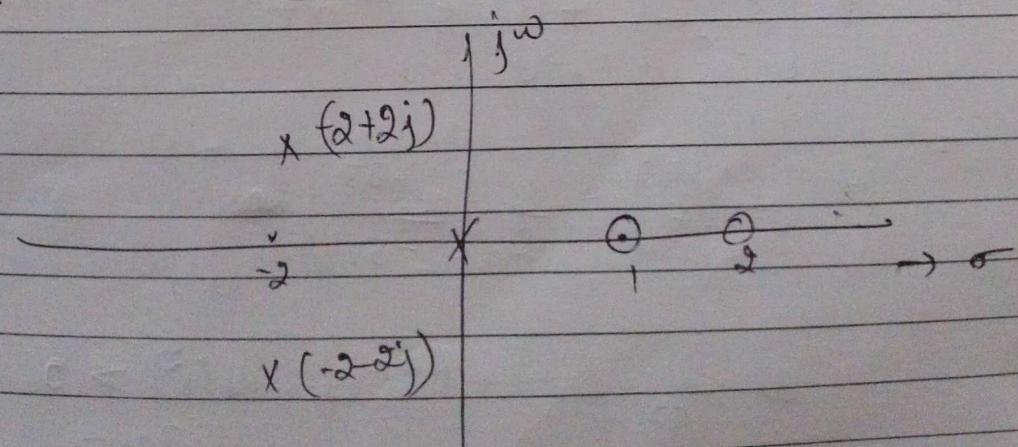


$$Q_2: X(s) = \frac{s^2 - 3s + 2}{s(s^2 + 4s + 8)}$$

$$\text{Num} \Rightarrow s^2 - 2s - 2 + 2 \\ s(s-2) - 1(s-2) \\ \text{zeros} = 1, 2$$

$$\text{Den} \Rightarrow s(s^2 + 4s + 8) \\ -4 \pm \sqrt{16-32} = -\frac{4 \pm 4j}{2} \\ = -2 \pm 2j$$

Poles $\rightarrow [0, -2 \pm 2j]$



$$\text{All transform} = 1$$

$$Q.1 \quad x(t) = s(t) \quad Q.2 \quad x(t) = u(t) \quad Q.3 \quad x(t) = \delta(t)$$

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

$$\int_{-\infty}^{\infty} s(t) e^{-st} dt$$

$$\left[\frac{e^{-st}}{-s} \right]_0^\infty = 1$$

$$X(s) = \int_{-\infty}^{\infty} u(t) e^{-st} dt$$

$$= \int_0^{\infty} e^{-st} dt$$

$$\Rightarrow \left[\frac{e^{-st}}{-s} \right]_0^\infty$$

$$\Rightarrow \frac{1}{s}$$

ROC = Entire s plane

ROC $\Rightarrow s > 0$

$$Q.3 \quad x(t) = \delta(t)$$

$$\int_{-\infty}^{\infty} t u(t) e^{-st} dt$$

$$\int_0^{\infty} t e^{-st} dt$$

$$= \left(t \int e^{-st} dt - \int \left[\int e^{-st} dt \right] dt \right)_0^\infty$$

$$= t \left[\frac{e^{-st}}{-s} \right] + \frac{1}{s} \int e^{-st} dt$$

$$= \left(t \left[\frac{e^{-st}}{-s} \right] + \frac{1}{s} \left[\frac{e^{-st}}{-s} \right] \right)_0^\infty$$

$$= \frac{1}{s^2}$$

$$\text{ROC} = \text{Re}[s] > 0$$

$$\text{Re}[s] > 0$$

* Right sided signal - i.e. signal for which $x(t) = 0$ before some finite time.

Page:

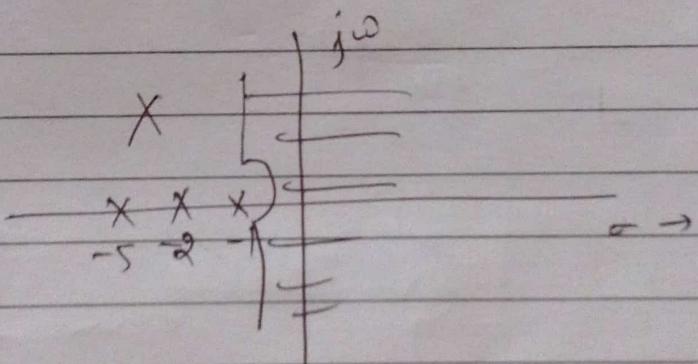
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Properties :- (ROC)

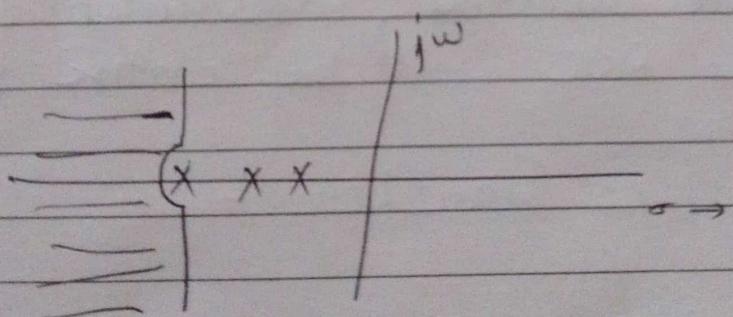
1. ROC never contain any pole.
2. ROC of laplace transform $X(s)$ consists of lines parallel to $j\omega$ axis.: ROC is
3. only dependent upon real part of s .
3. For a right sided signal the ROC will be right to the right most pole.

$$\text{Ex } s = -2, -5, -1$$

$$x(t) = u(t)$$



4. for a left sided signal, ROC will be left to the left most pole.



5. If $x(t)$ is a two sided signal then ROC will be b/w the poles but never contains any pole.

$$x(t) = e^{-2|t|}$$

$$X(s) = \int_{-\infty}^{\infty} e^{-st} x(t) dt$$

~~XX~~ $L[t^n u(t)] = \frac{n!}{s^{n+1}}$

$$x(t) = e^{-2t}, t \geq 0$$

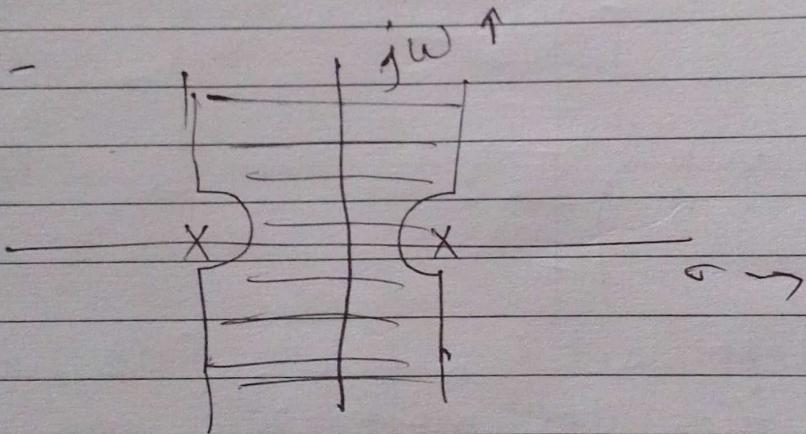
$$= e^{2t}, t < 0$$

$$X(s) = \int_{-\infty}^{\infty} e^{-s(2+y)t} dt$$

$$e^{-2t} u(t), t \geq 0 \rightarrow \frac{1}{s+2} \quad s > -2$$

$$e^{2t} u(-t), t < 0 \rightarrow \left(\frac{-1}{s-2}\right) \quad s < 2$$

ROC:-



6. for a finite duration, absolutely integrable signal, the ROC will be entire s-plane.

Properties of Laplace Transform:-

1. Linearity prop:-

If $x_1(t) \rightarrow X_1(s)$, ROC=R₁
 $x_2(t) \rightarrow X_2(s)$, ROC=R₂

Acc to linearity prop:-

$$a x_1(t) + b x_2(t) \rightarrow a X_1(s) + b X_2(s),$$

ROC containing R₁ & R₂

$$\text{Q: } x_1(t) = e^{-2t} u(t)$$

$$x_2(t) = e^{-4t} u(t)$$

$$X_1(s) = \frac{1}{s+2}, \quad s > -2$$

$$X_2(s) = \frac{1}{s+4}, \quad s > -4$$

$$X(s) = \frac{1}{s+2} + \frac{1}{s+4} \quad [s > -2]$$

$$\text{Q: } X_1(s) = \frac{1}{s+1} \quad s > -1$$

$$X_2(s) = \frac{1}{(s+1)(s+2)} \quad s > -1$$

$$\frac{1}{s+1} - \frac{1}{(s+1)(s+2)}$$

$$\frac{s+1}{(s+1)(s+2)} = \left(\frac{1}{s+2}\right) \quad \text{ROC: } s > -2$$

2. Time Shifting prop! -

If $x(t) \rightarrow X(s)$ $\xrightarrow{\text{ROC=R}}$
 then $x(t \pm t_0) \rightarrow e^{\pm st_0} X(s)$ $\xrightarrow{\text{ROC=R}}$

v. Imp Proof 1 →

$$L(x(t+t_0)) = X(s) = \int_{-\infty}^{\infty} x(t+t_0) e^{-st} dt$$

$$\text{let } t+t_0 = u.$$

$$= \int_{-\infty}^{\infty} x(u) e^{-s(u-t_0)} du$$

$$= \int_{-\infty}^{\infty} x(u) e^{-su} e^{stu} du$$

$$= e^{stu} x(s)$$

Q:

$$x(t) = u(t)$$

$$\mathcal{L}[x(t-t_0)] = ?$$

$$x = \int_{-\infty}^{\infty} x(t) e^{-s(t+t_0)} du$$

$$\Rightarrow e^{-stu} x(s)$$

04/09/23

SNS

- * If laplace transform of $x(t)$ is $X(s)$ then according to time reversal property laplace transform of $x(-t)$ is $X(-s)$

 Time reversal property

If $x(t) \rightarrow X(s)$; ROC = R

then $x(-t) \rightarrow X(-s)$, ROC = R

- * freq. shifting prop.

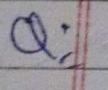
e. $x(t) \xrightarrow{\text{freq. shift}} X(s + s_0)$; ROC
 $= R + R_0(s)$

 $x(t) = e^{-2t} u(t) \leftrightarrow \frac{1}{s+2}$

* Time Scaling prop :-

$x(t) \rightarrow X(s)$, ROC = R

$x(\alpha t) \rightarrow \frac{1}{|\alpha|} X\left(\frac{s}{\alpha}\right)$, ROC = αR

 $x(t) = e^{-2t} u(t)$

find L.T. of $x(3t)$

$$= \frac{1}{3} \cancel{*} \left(\frac{1}{s/3 + 2} \right)$$

Q: If $x(t) = r(t)$

$$x(-2t) \rightarrow ?$$

$$\Rightarrow \frac{1}{s^2}$$

$$= \frac{1}{2} \left[\frac{1}{\left(\frac{s^2}{2}\right)} \right]$$

convolution prop:-

convolution prop is ~~one~~ multiplication
in time domain.

If $x_1(t) \rightarrow X_1(s)$; ROC = R_1 ,
 $x_2(t) \rightarrow X_2(s)$; ROC = R_2

then $x_1(t) * x_2(t) \rightarrow X_1(s) \cdot X_2(s)$, ROC =
cond. $R_1 \cap R_2$

$$X_1(s) = \int_{-\infty}^{\infty} \left[\int_{-\infty}^{\infty} x_1(z) x_2(t-z) dz \right] e^{-st} dt$$

$$= \int_{-\infty}^{\infty} x_1(z) \underbrace{\left[\int_{-\infty}^{\infty} x_2(t-z) e^{-st} dt \right]}_{X_2(s)} dz$$

$$= \int_{-\infty}^{\infty} (x_1(z) e^{-sz}) \underbrace{x_2(s)}_{X_2(s)} dz$$

$$= X_2(s) \cdot X_1(s)$$

Q: If $x_1(t) = e^{-2t} u(t)$ & $x_2(t) = e^{-4t} u(t)$

find the L.T. of $x_1(t) * x_2(t)$.

$\Rightarrow \frac{1}{s+2} \times \frac{1}{s+4}$ (If graphical method
then use that,
convert it into + form. otherwise this)

$$\frac{1}{(s+2)(s+4)} = \frac{A}{s+2} + \frac{B}{s+4}$$

$$= Ae^{-2t}u(t) + Be^{-4t}u(t)$$

* Multiplication Prop:-

$$\text{If } x_1(t) \rightarrow X_1(s) ; \text{ ROC} = R_1 \\ x_2(t) \rightarrow X_2(s) ; \text{ ROC} = R_2$$

$$\text{then } x_1(t) \cdot x_2(t) \rightarrow \frac{1}{2\pi j} [X_1(s) * X_2(s)]$$

[Unit 1 (~~test~~ place) → Friday Test]

* Differentiation in Time Domain :-

$$\text{If } x(t) \rightarrow X(s) ; \text{ ROC} = R$$

$$\text{Proof: } \frac{d}{dt} x(t) = s(X(s)) , \text{ ROC - containing } R$$

$$\frac{d^2}{dt^2} x(t) = s^2 X(s) - s x(0) - x'(0)$$

$$\frac{d^3}{dt^3} x(t) = s^3 X(s) - s^2 x(0) - s x'(0) - x''(0)$$

* Multiplication by ' t ' in time domain (or)
Differentiation in S-domain.

$$\text{if } x(t) \rightarrow X(s) \\ \text{then } t x(t) \rightarrow -\frac{d}{ds} X(s)$$