## Introduction

Euklidian Norm: 
$$\|x\|_2 = \sqrt{\sum_{i=1}^n x_i^2} = \sqrt{x^T x}$$
 
$$\|x\|_2^2 = x^T \cdot x$$

Weighting Eukl. Norm:  $\|x\|_Q^2 = x^T Q \cdot x$ 

Frobenius Norm: 
$$\|x\|_F^2 = trace(AA^T) = \sum_{i=1}^n \sum_{j=1}^m A_{ij} A_{ij}$$

Jacobian: 
$$\nabla f(x) = \frac{\partial f}{\partial x}(x)$$
 in  $\mathbb{R}^{n \times m}$  Hessian:  $\nabla^2 f(x)$ 

Error in variables: 
$$\hat{R}_{EV}(N) = \frac{\frac{1}{N}\sum_{k=1}^{N}u(k)}{\frac{1}{N}\sum_{k=1}^{N}i(k)}$$

Simple Approach: 
$$\hat{R}_{SA}(N) = \frac{1}{N} \cdot \sum_{k=1}^{N} \frac{u(k)}{i(k)}$$

$$\begin{split} \textbf{Least Squares:} \quad \hat{R}_{LS}(N) &= \underset{R \in \mathbb{R}}{\operatorname{argmin}} \sum_{k=1}^{N} (R \cdot i(k) - u(k))^2 \\ &= \frac{\frac{1}{N} \sum_{k=1}^{N} u(k) \cdot i(k)}{\frac{1}{N} \sum_{k=1}^{N} i(k)^2} \end{split}$$

 $\label{eq:matrix derivates:} \quad \frac{\mathrm{d}(c^Tx)}{\mathrm{d}x} = c \qquad \frac{\mathrm{d}(x^TAx)}{\mathrm{d}x} = (A^T+A)x$ 

Linear and non-linear models:

- linear if parameters linear i.e.  $(\theta_1 x^2 + \theta_2 x + \theta_3)$
- nonliniar if i.e  $(\sin(\theta_1)x + \theta_2)$  or derivatives in other orders than 1 Table of Derivatives:

f(x)	$\mathbf{f'}(\mathbf{x})$
$g(x) \cdot h(x)$	$g'(x) \cdot h(x) + g(x) \cdot h'(x)$
g(h(x))	$g'(h(x)) \cdot h'(x)$
$\sin(x)$	$-\cos(x)$
$\cos(x)$	$\sin(x)$
$\tan(x) = \frac{\sin(x)}{\cos(x)}$	$\frac{1}{\cos^2(x)} = \sec^2(x)$
$e^{kx}$	$\frac{1}{k}e^{kx}$
ln(x)	$\frac{1}{x}$
$\log_a x$	$\frac{1}{\ln a}(x\ln x - x)$
Ax	A
$x^T A$	$A^T$
$x^T B x$	$x^T(B^T+B)$

## Random Variables and Probability

Dependent Probability:  $P(A \lor B) = P(A) + P(B)$ 

Independent Prob.:  $P(A, B) = P(A \wedge B) = P(A) \cdot P(B)$ 

Conditional Prob.:  $P(A|B) = \frac{P(A|B)}{P(B)} = \frac{P(B|A) \cdot P(A)}{P(B)}$  (Bayes' theorem)

$$P(X \in [a, b]) = \int_{a}^{b} p_X(x) dx \qquad p(x|y) = \frac{p(x, y)}{p(y)}$$

 $\mathbf{Mean/Expectation\ value:}\ \mathbb{E}\{\mu_X\} := \mu_X = \int_{-\infty}^{\infty} x \cdot p_X(x) dx$ 

$$\mathbb{E}\{a+bX\}:=a+b\mathbb{E}\{X\}$$

Variance:  $\sigma_X^2 := \mathbb{E}\{\left(X - \mu_X\right)^2\} = \mathbb{E}\{X^2\} - \mu_X^2$ 

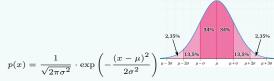
Standard deviation:  $\sigma_X = \sqrt{\sigma_X^2}$ 

#### Distributions

$$\textbf{Uniform distribution:} P_y(x) = \begin{cases} \frac{1}{b-a} & \text{if} \quad x \in [a,b] \\ 0 & \text{else} \end{cases}$$

$$\mathbf{Mean:}\ \mu_X = \int_{-\infty}^{\infty} x \, p_X(x) \mathrm{d}x = \int_a^b \frac{1}{b-a} \cdot x \mathrm{d}x = \frac{a+b}{2} =: \mu_X$$

Normal distribution:  $X \sim \mathcal{N}(\mu, \sigma^2)$   $\hat{\theta}_{LS} \sim \mathcal{N}(\theta_0, \Sigma_{\hat{\theta}})$ 



Multidimensional Normal Distribution

$$p(x) = \frac{1}{\sqrt{(2\pi)^n \cdot det(\Sigma)}} \cdot \exp\left(-\frac{1}{2} \cdot (x-\mu)^T \cdot \Sigma^{-1} \cdot (x-\mu)\right)$$

Weibull distribution:  $F(x) = 1 - \exp(-(\lambda \cdot x)^k)$ 

Laplace distribution:  $f(x|\mu, b) = \frac{1}{2b} \cdot \exp\left(-\frac{|x - \mu|}{b}\right)$ 

## Useful statistic definitions

Covariance and Correlaton:  $\sigma(X,Y) := \mathbb{E}(X-\mu_X)(Y-\mu_Y)$ 

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} (x - \mu_X)(y - \mu_Y) \cdot p_{X,Y}(x, y) \, \mathrm{d}x \, \mathrm{d}y$$

Covariance Matrix:  $\Sigma_x = cov(X) = \mathbb{E}\{XX^T\} - \mu_x \mu_x^T \text{ is PSD}$ 

$$\Sigma = \begin{bmatrix} \sigma_x^2 & \sigma_{yx} \\ \sigma_{xy} & \sigma_y^2 \end{bmatrix} \quad \sigma_{xy} = \sigma yx = \rho_{xy} \cdot \sigma_x \cdot \sigma_y \text{ where } \rho \text{ is correlation}$$

and u are i.i.d.  $\Rightarrow \Sigma$  is diagonal

Multidimensional Random Variables:

$$\mathbb{E}f(X) = \int_{\mathbb{R}^n} f(x) p_X(x) d^n x$$

$$cov(X) = \mathbb{E}\{(X - \mu_X)(X - \mu_X)^T\}$$

$$cov(X) = \mathbb{E}\{XX^T\} - \mu_X \mu_X^T$$

$$cov(Y) = \Sigma_y = A\Sigma_x A^T \quad for \quad y = A \cdot x$$

$$\mathbb{E}\{AX\} = A \cdot \mathbb{E}\{X\}$$

Rules for variance:

$$var(aX) = a^2 \cdot var(X)$$

$$\operatorname{var}(X+Y) = \operatorname{var}(X) + \operatorname{var}(Y) + 2 \cdot \operatorname{cov}(X,Y)$$

Formula for variance:  $var(X) = \mathbb{E}((X - \mathbb{E}(X))^2) = \mathbb{E}(X^2) - (\mathbb{E}(X))^2$ Correlation:

uncorrelated if  $\rho(X,Y) = 0$ ,  $\rho(X,Y) := \frac{\text{cov}(X,Y)}{\sigma_x \sigma_y}$ 

#### Statistical estimators

Biased- and Unbiasedness An estimator  $\hat{\theta}_N$  is unbiased  $\Leftrightarrow \mathbb{E}\{\hat{\theta}_N(y_N)\} = \theta_0$ , where  $\theta_0 \equiv$  "true" value of  $\theta$ . Otherwise: biased.

Asymptotic Unbiasedness An estimator  $\hat{\theta}_N$  is called asymptotically unbiased  $\Leftrightarrow \lim_{n \to \infty} \mathbb{E}\{\hat{\theta}_N(y_N)\} = \theta_0$ 

Consistency An estimator  $\hat{\theta}_N(y_N)$  is called consistent if, for any  $\epsilon > 0$ , the probability  $P(\hat{\theta}_N(y_N) \in [\theta_0 - \epsilon, \theta_0 + \epsilon])$  tends to 1 for  $N \to \infty$ .

## Unconstrainded Optimization

Theorem 1: (First Order Necessary Conditions)

If  $x^* \in D$  is local minimizer of  $f: D \to \mathbb{R}$  and  $f \in C^1$  then  $\nabla f(x^*) = 0$  Definition (Stationary Point) A point  $\bar{x}$  with  $\nabla f(\bar{x}) = 0$  is called a stationary point of f.

Theorem 2: (Second Order Necessary Conditions)

If  $x^* \in D$  is local minimizer of  $f:D \to R$  and  $f \in C^2$  then  $\nabla^2 f(x^*) \succ 0$ 

Theorem 3: (Second Order Sufficient Conditions and Stability under Perturbations)

Assume that  $f:D\to R$  is  $C^2$ . If  $x^*\in D$  is a stationary point and  $\nabla^2 f(x^*)\succ 0$  then  $x^*$  is a strict local minimizer of f. In addition, this minimizer is locally unique and is stable against small perturbations of f, i.e. there exists a constant C such that for sufficiently small  $p\in\mathbb{R}^n$  holds

$$||x^* - \arg\min_{x} (f(x) + p^T x)|| \le C||p||$$

## Linear Least Squares Estimation

Preliminaries: i.i.d. and Gaussian noise

Overall Model:  $y(k) = \phi(k)^T \theta + \varepsilon(k)$ 

LS cost function as sum:  $\sum_{k=1}^{N} (y(k) - \phi(k)^T \theta)^2$ 

LS cost function:  $f(\theta) = \|y_N - \Phi_N \theta\|_2^2$ 

Unique minimizers:  $\hat{\theta}_{LS} = \underset{\theta \in \mathbb{R}}{arg \min} f(\theta)\theta^* = \underbrace{(\Phi^T \Phi)^{-1} \Phi^T}_{*\perp} y$ 

Pseudo Inverse:  $\Phi^+ = (\Phi^T \Phi)^{-1} \Phi^T$ 

#### Weighted Least Squares (unitless)

For i.i.d noise: Unweight Least Squares is optimal: W = I

$$f_{WLS}(\theta) = \sum_{k=1}^{N} \frac{\left(y(k) - \phi(k)^T \theta\right)^2}{\sigma_{\epsilon}^2(k)} = \|y_N - \Phi_N \theta\|_W^2$$

$$= \|W^{\frac{1}{2}}y - W^{\frac{1}{2}}\Phi_N\theta\|_2^2 = (y_N - \Phi \cdot \theta)^T \cdot W \cdot (y_N - \Phi \cdot \theta)$$

Solution for WLS:

$$\begin{split} \hat{\theta}_{WLS} &= \tilde{\Phi}^+ \tilde{y} & \text{mit } \tilde{\Phi} = W^{\frac{1}{2}} \Phi \text{ und } \tilde{y} = W^{\frac{1}{2}} y \\ &= \underset{\theta \in \mathbb{R}}{\operatorname{argmin}} f_{WLS}(\theta) = \left(\Phi^T W \Phi\right)^{-1} \Phi^T W y \end{split}$$

## Ill-Posed Least Squares

Singular Value Decomposition:  $A = USV^T \in \mathbb{R}^{mxn}$  with  $U \in \mathbb{R}^{mxm}$ ,  $V \in \mathbb{R}^{nxn}$  and  $S \in \mathbb{R}^{mxn}$  where S is a diagonal Matrix with non-negative elements  $(\sigma_1, \ldots, \sigma_r, 0, \ldots, 0)$ 

Moore Penrose Pseudo Inverse:

$$\Phi^{+} = VS^{+}U^{T} = V(S^{T}S + \alpha I)^{-1}S^{T}U^{T}$$

 $\Phi^+$  therefore selects  $\theta^* \in S^*$  with minimal norm

Regularization for Least Squares:

$$\lim_{a \to 0} (\Phi^T \Phi + \alpha I)^{-1} \Phi^T = \Phi^+ \quad \text{with } \Phi^+ MPPI$$

$$\theta^* = (\Phi^T \Phi + \alpha \mathbb{I})^{-1} \Phi^T y$$

#### Statistical Analysis of WLS

Expectation of Least Squares Estimator:

$$E\{\hat{\theta}_{WLS}\} = E\{(\Phi_N^T W \Phi_N)^{-1} \Phi_N^T W y_N\} = \theta_0$$

Covariance of the least squares estimator:

$$\begin{split} & \operatorname{cov}(\hat{\theta}_{WLS}) = \left(\boldsymbol{\Phi}_N^T W \boldsymbol{\Phi}_N\right)^{-1} = \left(\boldsymbol{\Phi}_N^T \boldsymbol{\Sigma}_{\in N}^{-1} \boldsymbol{\Phi}_N\right)^{-1} \\ & \operatorname{cov}(\hat{\theta}_{WLS}) \succeq \left(\boldsymbol{\Phi}_N^T W \boldsymbol{\Phi}_N\right)^{-1} \end{split}$$

## Example LLS

Example of the Linear Least Square Estimator for: N=2

$$\varepsilon(1) \sim \mathcal{N}(0|\sigma_1^2)$$
  $\varepsilon(1)$ 

$$\varepsilon(2) \sim \mathcal{N}(0|\sigma_2^2)$$

$$N=2; \quad \Sigma_{\varepsilon_N} = \begin{bmatrix} \sigma_1^2 & 0 \\ 0 & \sigma_2^2 \end{bmatrix} \qquad W^{OPT} = \Sigma_{\varepsilon_N}^{-1} = \begin{bmatrix} \frac{1}{\sigma_1^2} & 0 \\ 0 & \frac{1}{\sigma_2^2} \end{bmatrix}$$

$$\begin{aligned} \cos(\hat{\theta}_{WLS}) &= (Y_N - \Phi_N \theta)^T \cdot W \cdot (Y_N - \Phi_N \theta) \\ &= \sum_{k=1}^2 (y(k) - \phi(k)^T \theta) \cdot \frac{1}{\sigma_k^2} \cdot (y(k) - \phi(k)^T \theta) \end{aligned}$$

Measuring the goodness of Fit using:  $R^2 - (0 \le R^2 \le 1)$ 

$$\begin{split} R^2 &= 1 - \frac{\|y_N - \Phi_N \hat{\theta}\|_2^2}{\|y_N\|_2^2} = 1 - \frac{\|\epsilon_N\|_2^2}{\|y_N\|_2^2} \\ &= \frac{\|y_N\|_2^2 - \|\epsilon_N\|_2^2}{\|y_N\|_2^2} = \frac{\|\hat{y}_N\|_2^2}{\|y_N\|_2^2} \end{split}$$

Residual:  $\epsilon_N \uparrow \rightarrow R^2 \rightarrow 0 \ (\Rightarrow bad)$ 

# Estimating the Covariance with the Single Experiment

$$\hat{\sigma}_{\varepsilon}^{2} := \frac{1}{N - d} \sum_{k=1}^{N} (y(k) - \phi(k)^{T} \hat{\theta}_{LS})^{2} = \frac{\|y_{N} - \phi_{N} \hat{\theta}_{LS}\|_{2}^{2}}{N - d}$$

$$\hat{\Sigma}_{\hat{\theta}} \coloneqq \hat{\sigma}_{\varepsilon}^2 (\phi_N^T \phi_N)^{-1} = \sigma_{\varepsilon}^2 (\Phi_N^+ \Phi_N^{+T}) = \frac{\|y_N - \phi_N \hat{\theta}_{LS}\|_2^2}{N - d} \cdot (\phi_N^T \phi_N)^{-1}$$

### Bayesian Estimation and the Maximum a Posteriori Estimate

## Assumptions:

- Measurement:  $y_N \in \mathbb{R}^N$  has i.i.d. noise Linear Model:  $M(\theta) = \phi_N \cdot \theta$  and  $\theta \in \mathbb{R}$

$$p(\theta|y_N) = \frac{p(y_N, \theta)}{p(y_N)} = \frac{p(y_N|\theta) \cdot p(\theta)}{p(y_N)}$$

$$\hat{\theta}_{MAP} = \arg\min_{\theta \in \mathbb{R}} \{ -\log(p(y_N | \theta)) - \log(p(\theta)) \}$$

Max. Likelihood prev. knowledge

MAP Example: Regularised Least Squares

$$\theta = \bar{\theta} \pm \sigma_{\theta}$$
 with  $\bar{\theta} = \theta_{a\text{-priori}}$ 

$$\hat{\theta}_{MAP} = \mathop{\rm argmin}_{\theta \in \mathbb{R}} \frac{1}{2} \cdot \frac{1}{\sigma_{r^2}} \cdot \left\| y_N - \Phi_N \cdot \theta \right\|_2^2 + \frac{1}{2} \cdot \frac{1}{\sigma_{\theta}^2} \cdot \left(\theta - \bar{\theta}\right)^2$$

#### Maximum Likelihood Estimation

## L<sub>2</sub> Estimation: Maximum Likelihood Estimation (ML):

- Measurement Errors assumed to be Normally distributed
- Model described by a non-linear function  $M(\theta)$
- Every unbiased estimator needs to satisfy the Cramer-Rao inequality, which gives a lower bound on the covariance matrix

Model:  $y = M(\theta) + \varepsilon$ 

$$p(y|\theta) = C \prod_{i=1}^{N} \exp\left(\frac{-(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}\right) \quad C = \prod_{i=1}^{N} \frac{1}{\sqrt{2 \cdot \pi \sigma_i^2}}$$

Positive log-Likelihood: (log changes prod

$$\log p(y|\theta) = \log(C) + \sum_{i=1}^{N} -\frac{(y_i - M_i(\theta))^2}{2 \cdot \sigma_i^2}$$

#### Negative log-Likelihood:

$$\hat{\theta}_{ML} = \arg\max_{\theta \in \mathbb{R}^d} \ p(y|\theta) = \arg\min_{\theta \in \mathbb{R}^d} \sum_{i=1}^N \frac{\left(y_i - M_i(\theta)\right)^2}{2\sigma_i^2}$$

$$= \arg\min_{\theta \in \mathbb{R}^d} \frac{1}{2} \sum_{i=1}^{N} \left( \frac{y_i - M_i(\theta)}{\sigma_i} \right)^2$$

$$= \arg\min_{\theta \in \mathbb{R}^d} \frac{1}{2} \|S^{-1}(y - M(\theta))\|_2^2 \quad \text{with: } S = \begin{bmatrix} \sigma_1 & & & \\ & \ddots & & \\ & & \ddots & \\ & & \sigma_N \end{bmatrix}$$

#### $L_1$ Estimation:

- Measurement Errors assumed to be Laplace distributed and more robust against outliers.

$$\min_{\theta} \|y - M(\theta)\|_1 = \min_{\theta} \sum_{i=1}^{N} |y_i - M_i(\theta)|$$

$$\Rightarrow \text{ median of } \{Y_1, \dots, Y_N\}$$

## Recursive Linear Least Squares

$$\begin{aligned} Q_{N+1} &= \alpha \cdot Q_N + \varphi(N+1) \cdot \varphi(N+1)^T \\ \hat{\theta}_{ML}(N+1) &= \hat{\theta}_{ML}(N) + Q_{N+1}^{-1} \cdot \varphi(N+1) \\ & \cdot [ \underbrace{y(N+1)}_{\text{new measurement}} - \underbrace{\varphi(N+1)^T \cdot \hat{\theta}_{ML}(N)}_{\text{old prediction}} ] \end{aligned}$$

 $Q_0$  and  $\hat{\theta}_0$  have to be chosen.

 $Q_0$  should be non-singular, small and positive definite. (e.g.  $10^{-3} \cdot I$ )  $Q_N \approx \Sigma_{\hat{\theta}_{\mathrm{ML}}(N)}^{-1}$ 

## Cramer-Rao-Inequality (Fisher information Matrix M)

$$\Sigma_{\hat{\theta}} \succeq M^{-1} = \underbrace{(\Phi_N^T \cdot \Sigma^{-1} \cdot \Phi_N)^{-1}}_{\text{for lin. model w. } \varepsilon \sim \mathcal{N}(\mu, \sigma)}$$
$$M = \int_{\mathcal{U}_T} \nabla_{\theta}^2 L(\theta_0, y_n) \cdot p(y_n | \theta_0) \mathrm{d}y_n$$

- Minimising a Linear Model
- Gaussian Noise:  $X \sim \mathcal{N}(0, \Sigma)$ )

$$L(\theta, y_N) = -\log(p(y_N|\theta))$$

(if lin. model, etc.) = 
$$\frac{1}{2} \cdot (\Phi_N \cdot \theta - y_N)^T \cdot \Sigma^{-1} \cdot (\Phi_N \cdot \theta - y_N)$$
$$M = \mathbb{E}\{\nabla_{\theta}^2 L(\theta, y_N)\} = \nabla_{\theta}^2 L(\theta, y_N) (= \Phi_N^T \cdot \Sigma^{-1} \cdot \Phi_N)$$

 $\Rightarrow W = \Sigma^{-1}$  is the optimal weighting Matrix for WLS.

## Continuous Time Systems

Ordinary Differential Equations (ODE):

$$\dot{x} = f(x(t), u(t), \epsilon(t), p)$$

Differential Algebraic Equations(DAE):

$$\dot{x} = f(x(t), u(t), \epsilon(t), p)$$

$$0 = g(x, z).$$

LTI Sytem (ODE):

$$\dot{x} = Ax + Bu$$
  $y = Cx + Du$ 

$$G(s) = C(sI - A)^{-1}B + D$$

## Numerical Integration Methods

### Euler Integration Step

$$\tilde{x}(t; x_0, u_{\text{const}}) = x_0 + t f(x_0, u_{\text{const}}), \quad t \in [0, \Delta t]$$

$$\tilde{x}_{j+1} = \tilde{x}_j + h f(\tilde{x}_j, u_{\text{const}}), \quad j = 0, ..., M - 1$$

- Approximation becomes better by decreasing the step size h
- Concistency Error: h<sup>2</sup>
- Total Number of steps:  $\Delta t/h$
- Error in the final step of order  $h\Delta t$
- Linear in step size → order one
- Taking more steps is more accurate but needs more computation

## Runge-Kutta Method of Order Four (RK4)

$$k_1 = f(\tilde{x}_j, u_{\text{const}})$$

$$k_2 = f(\tilde{x}_j, \frac{h}{2}k_1, u_{\text{const}})$$

$$k_3 = f(\tilde{x}_j, \frac{h}{2}k_2, u_{\text{const}})$$

$$k_4 = f(\tilde{x}_j, hk_3, u_{\text{const}})$$

$$\tilde{x}_{j+1} = \tilde{x}_j + \frac{h}{6}(k_1 + 2k_2 + 2k_3 + k_4)$$

One Step of RK4 is thus as expensive as four steps of euler accurrency of final approximation is of order  $h^4\Delta$  t

→ RK4 needs fewer functions to obtain the same accuracy level as euler

## Discrete Time Systems

Det. Model as State Space Stoch. Model as State Space

Det. Model as Input-Output Stoch. Model as Input-Output

#### State Space Model

 $x_{k+1} = f_k(x_k, u_k), k = 0, 1, \dots, N-1$  with input vector  $u_k$  and state

Input-Output Model

$$y(k) = h(u(k), \dots, u(k-n), y(k-1), \dots, y(k-n))$$

LTI system as State-Space Model:

$$x_{k+1} = Ax_k + Bu_k, \quad k = 0, 1, ..., N-1.$$

$$G(s) = \frac{b_0 + b_1 s + \dots + b_n s^n}{a_0 + a_1 s + \dots + a_{n-1} s^{n-1} + s^n} \quad | \cdot s = z^{-1}$$

$$G(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_n z^{-n}}{a_0 + a_1 z^{-1} + \dots + a_n z^{-n}}$$

$$= \frac{b_0 z^n + b_n z^{n-1} + \dots + b_n}{a_0 z^n + a_1 z^{n-1} + \dots + a_n} \Rightarrow \text{Also called "polynomial model"}.$$

#### Deterministic Model

The output of the system can be obtained with absolute certainty. The Output y or the state x, depend on the known inputs  $u(1), \ldots, u(N)$ , the previous Outputs  $y(1), \ldots, y(N)$  or state x(n-1) and initial conditions. State Space Model:

$$x(t+1) = f(x(k), u(k))$$
$$y(k) = g(x(k), u(k))$$

Initial conditions:  $x(1) = x_0$ 

Input-Output Model

$$y(k) = h(u(k), ..., u(k-n), y(k-1), ..., y(k-n))$$

Initial conditions:  $y(1) = y_1, \ldots, y(n) = y_n$   $u(1) = u_1, \ldots, u(n) = u_n$  Finite Impulse Response (FIR):

$$y(k) = b_0 u(k) + ... + b_{n_b} u(k - n_b)$$

$$\begin{split} G(z) &= b_0 + b_1 z^{-1} + \ldots + b_{n_b} z^{-n_b} \quad | \cdot \frac{z^{n_b}}{z^{n_b}} \\ &= \frac{b_0 z^{n_b} + b_1 z^{n_b - 1} + \ldots + b_{n_b}}{z^{n_b}} \end{split}$$

Auto Regressive Models with Exogenous Inputs (ARX):

$$a_0y(k) + \cdots + a_{n_a}y(k - n_a) = b_0u(k) + \cdots + b_{n_b}u(k - n_b)$$

$$G(z) = \frac{b_0 z^n + b_1 z^{n-1} + \dots + b_n}{a_0 z^n + a_1 z^{n-1} + \dots + a_n}$$

The next output depends on the previous output. Also called  $\mathbf{IIR}$  (infinite impulse response)

## Stochastic Model

Real systems are far from deterministic.

- there is stochastic noise  $\varepsilon(k)$
- $\bullet\,$  there are constant and unknown parameters p
- measured outputs y(k) depend in both,  $\varepsilon(k)$  and p

Assumptions: noise is  $\mathbf{i}.\mathbf{i}.\mathbf{d}$  and enters the model like a normal input, but as a random variable

#### State Space Model

$$x(t+1) = f(x(k), u(k), \varepsilon(k))$$
$$y(k) = q(x(k), u(k), \varepsilon(k))$$

#### Input-Output Model

Only interested in input and output, not the whole model state

$$y(k) = h(u(k), ..., u(k-n), y(k-1), ..., y(k-n), \varepsilon(k), ..., \varepsilon(k-n))$$
  
for  $k = n+1, n+2, ...$ 

Measurement Noise (Output Error Model)

$$y(k) = M(k; U, x_0, p) + \varepsilon(k)$$

## Stochastic Disturbance (Equation Errors)

$$y(k) = h(p, u(k), ..., u(k-n), y(k-1), ..., y(k-n)) + \varepsilon(k)$$
  
for  $k = n + 1, n + 2, ...$ 

Linear In the Parameters models (LIP):

$$y(k) = \sum_{i=1}^{d} \theta_i \phi_i(u(k), \dots, y(k-1), \dots) + \varepsilon(k)$$

$$y(k) = \varphi(k)^T \theta + \varepsilon(k)$$
 where  $\varphi = (\phi_1(\cdot), ..., \phi_d(\cdot))$ 

LIP-LTI Models with Equation Errors (ARX)
- Combining best of two worlds (LTI and LIP)

$$a_0 y(k) + \ldots + a_{n_0} y(k - n_0) = b_0 u(k) + \ldots + b_{n_b} u(k - n_b) + \varepsilon(k)$$

Auto-Regressive Moving Average with eXogeneous input (ARMAX):

$$a_0 y(k) + \ldots + a_{n_a} y(k-n_a) = b_0 u(k) + \ldots + b_{n_b} u(k-n_b) + \varepsilon(k) +$$

$$c_1 \varepsilon(k-1) + \dots + c_{n_T} \varepsilon(k-n_c)$$

Auto-Regressive Moving Average without inputs (ARMA):

$$a_0y(k)+\ldots+a_{n_a}y(k-n_a)=\varepsilon(k)+c_1\varepsilon(k-1)+\ldots+c_{n_x}\varepsilon(k-n_c)$$

Where  $c_i$  represent the noise coefficient, we have to use non-linear least squares with the unknown noise terms  $\varepsilon(k-i)$ 

## Difference between Deterministic and Stochastic Models

- stochastic noise  $\varepsilon(k)$
- $\bullet$  unknown but constant parameter p
- measured output y(k) depend on both,  $\varepsilon(k)$  and p

# Example for State Space Model

$$\ddot{a} = m \cdot \dot{a} + g \cdot a + c \cdot u$$

$$y = \dot{a}$$

$$x = \begin{bmatrix} a \\ \dot{a} \end{bmatrix} \dot{x} = \begin{bmatrix} \dot{a} \\ \ddot{a} \end{bmatrix} \dot{x} = Ax + Bu \quad y = Cx + Du$$

$$A = \begin{bmatrix} 0 & 1 \\ g & m \end{bmatrix} B = \begin{bmatrix} 0 \\ c \end{bmatrix} C = \begin{bmatrix} 0 & 1 \end{bmatrix} D = \begin{bmatrix} 0 \end{bmatrix}$$

## Pure Output Error (OE) Minimization

Assume: i.i.d. gaussian noise only affecting output using non-linear least squares

$$\theta_{ML} = \min_{\theta} \sum_{k=1}^{N} (y(k) - M(k; U, x_0 p))^2$$

Output Error Minimization for FIR Models: lead to convex problems, therefore global minimum can be found

$$y(k) = (u(k), u(k-1), ..., u(k-n_{n_h})) \cdot \theta + \varepsilon(k)$$

$$= \min_{\theta} \sum_{k=n_b+1}^{N} (y(k) - \underbrace{(u(k), u(k-1), ..., u(k-n_{n_b}))}_{\text{det. part is also } M(k: U.x_0, p)} \cdot \theta)^2$$

They often need a very high dimension  $n_b$  to obtain a reasonable fit. As a consequence ARX models are usually used instead.

Equation Error Minimization: Assume: i.i.d.  $\varepsilon(k)$  noise enters the input-output equation as additive disturbance

$$y(k) = h(p, u(k), ..., u(k-n), y(k-1), ..., y(k-n)) + \varepsilon(k)$$

for 
$$k = n + 1, n + 2$$

if the i.i.d noise is gaussian, a maximum likelihood formulation to estimate the unknown parameter vector  $\theta=p$  is given:

$$\theta_{ML} = \min_{\theta} \sum_{k=n+1}^{N} (y(k) - h(p, u(k), ..., y(k-1), ...)))^{2}$$

u and k are known input and output measurements, and the algorithm minimises the so called equation errors or prediction errors.

This problem is also known as **Prediction error minimisation(PEM)** Such a problem is convex if p enters linearly in f, i.e. if the model is linear-in-the-parameters (LIP)

#### PEM of LIP Models

$$y(k) = \varphi(k)^T \theta + \varepsilon(k)$$

where 
$$\varphi = (\phi_1(\cdot), ..., \phi_d(\cdot))^T$$
 are the regressor variables

considering this last expression, the prediction error minimisation (PEM) problem can be formulated as:

$$\min_{\theta} \underbrace{\sum_{k=1}^{N} (y(k) - \varphi(k)^{\mathrm{T}} \theta)^{2}}_{=\|y_{N} - \Phi_{N} \theta\|_{2}^{2}}$$

Which can be solved using LLS  $\theta^* = \Phi_N^+ y_N$ 

Special Case: PEM of LIP-LTI Models with Equation Errors(ARX) General ARX model equation

$$a_0 y(k) + \dots + a_{n_a} y(k - n_a) = b_0 u(k) + \dots + b_{n_b} u(k - n_b) + \varepsilon(k)$$

In order to have a determined estimation problem,  $a_0$  has to be fixed, otherwise the number of optimal solutions would be infinitive. Therefore we sually fix  $a_0=1$  and use  $\theta=(a_1,\dots,a_{n_a},b_0,\dots,b_{n_b})^{\rm T}$  as the parameter estimation vector. The regressor vector is given by:

$$\varphi = (-y(k-1), ..., -y(k-n_a), u(k), ..., u(k-n_b))^{\mathrm{T}}$$

leading to the optimal solution provided by LLS:

$$y(k) = \varphi(k)^{\mathrm{T}} \theta + \varepsilon(k)$$

## Pure Output Error (OE) Minimization

## Models with Input and Output Errors:

$$y(k) = M(k; U + \varepsilon_N^u, x_0, p) + \varepsilon^y(k)$$
 input noise  $\epsilon^u(t)$  output noise  $\epsilon^v(t)$  output noise  $\epsilon^v(t)$  measured input 
$$u(t)$$
 
$$\tilde{u}(t)$$
 System

parameters p

Assume: i.i.d. gaussian noise on both input and output with variance  $\sigma_u^2$  for the input and  $\sigma_u^2$  for the output

$$\begin{split} & \underset{\theta}{\operatorname{argmin}} \sum_{k=1}^{N} \frac{1}{\sigma_{y}^{2}} (y(k) - M(k; U + \varepsilon_{N}^{u}, x_{0}, p))^{2} + \frac{1}{\sigma_{u}^{2}} (\varepsilon_{u}(k))^{2} \\ & \underset{\theta}{\operatorname{argmin}} \sum_{k=1}^{N} \frac{1}{\sigma_{y}^{2}} (y(k) - M(k; \tilde{U}, x_{0}, p))^{2} + \frac{1}{\sigma_{u}^{2}} (u(k) - \tilde{u}(k))^{2} \\ & \tilde{U} := U + \varepsilon_{N}^{u} \end{split}$$

initial conditions

## Fourier Transformation

FT:

$$F\{F\}(\omega) = \int_{-\infty}^{\infty} f(t)e^{-j\omega t}dt$$

iFT

$$f(t) = F^{-1}{F}(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega)e^{-j\omega t} d\omega$$

DFT:

$$U(m) := \sum_{k=0}^{N-1} u(t)e^{-j\frac{2\pi mk}{N}}$$

iDFT:

$$u(n) := \sum_{k=0}^{N-1} U(k)e^{j\frac{2\pi kn}{N}}$$

## Useful frequency things

$$\omega = 2\pi f = \frac{2\pi}{T}$$
  $f_s > 2f_{max}$   $T = N\Delta t = \frac{N}{f_s}$ 

# Aliasing and Leakage Errors

Aliasing Error: Due to sampling of continous signal to discrete signal. Avoid with Nyquist Theoreme:

$$f_{Nyquist} = \frac{1}{2\Delta t} [\text{Hz}] \quad or \quad \omega_{Nyquist} = \frac{2\pi}{2\Delta t} [\text{rad/s}]$$

Leackage Error: Due to windowing.

$$\omega_{base} := \frac{2\pi}{N \cdot \Delta t} = \frac{2\pi}{T} \rightarrow \omega = m \frac{2\pi}{N \cdot \Delta t}$$

## Crest Factor = Scheitelfaktor

$$\begin{array}{ll} \text{Crest Factor } = \frac{u_{max}}{u_{rms}} & \text{with} : u_{rms} := \sqrt{\frac{1}{T} \int_0^T u(t)^2 \, dt} \\ & \text{and} \quad u_{max} := \max_{t \in [0,T]} \lvert u(t) \rvert \end{array}$$

### Optimising Multisine for optimal crest factor

Frequency: Choose frequencies in logarithmic manner as multiples of the base frequency.  $\omega_{k+1}/\omega_k \approx 1.05$ 

Phase: To prevent high peaks (Crest Factor) in the Signal, the phases of the different frequencies are modulated accordingly. (Positive interference)

# Multisine Identification Implementation procedure

Window Length: Integer multiple of sampling time:  $T = N \cdot \delta t$ 

Harmonics of base frequency: Are contained in multisine

 $\omega_{base} = \frac{2}{2}$ 

Highest contained Frequency: Is half of Nyquist frequency:  $\omega_{Nyquist} = 2\pi$ 

Experiment and Analysis: (Step 2): Insert Multisine periodically. Drop first Periods (till transients died out). Record M Periods, each with N samples, of input and output data. Average all the M periods and make the DFT (or vice versa). Finally build transfer function:  $\hat{G}_{j\omega_k} = \frac{\hat{Y}(k(p))}{\hat{U}(k(p))}$ 

## Nonparametric and Frequency Domain Identification Models

## Impulse response and transfer function:

$$y(t) = \int_{0}^{\infty} g(\tau)u(t-\tau)\,\delta t$$

$$Y(s) = G(s) \cdot U(s)$$

$$G(s) = \int_{0}^{\infty} e^{-st} g(t) dt$$

Bode diagram from frequency sweeps:

$$u(t) = A \cdot \sin(\omega \cdot t), \quad y(t) = ||G(j \cdot \omega)||A \cdot \sin(\omega \cdot t + \alpha)$$

# Bode Diagramm

 $Magnitude = Amplitude |G(j\omega)|$ 

Phase  $arg G(j\omega)$ 

## Recursive Least Squares

New Inverse Covariance:  $Q_K = Q_{k-1} + \phi_K \phi_K^T$ 

## Kalman Filter

## Valid for Discrete and Linear!

If recursive least squares:  $x_{k+1} = A_k \cdot x_k$ 

$$x_{k+1} = A_k \cdot x_k + \omega_k$$
 and  $y_k = C_k \cdot x_k + v_k$ 

Steps of Kalman Filter

1 Deadistic

$$\hat{x}_{[k|k-1]} = A_{k-1} \cdot \hat{x}_{[k-1|k-1]}$$

$$P_{[k|k-1]} = A_{k-1} \cdot P_{[k-1|k-1]} \cdot A_{k-1}^T \cdot W_{k-1}$$

If RLS, without:  $W_{k-1}$ 

#### 2 Innovation update

$$P_{[k|k]} = (P_{[k|k-1]}^{-1} + C_k^T \cdot V^{-1} \cdot C_k)^{-1}$$

$$\hat{x}_{[k|k]} = \hat{x}_{[k|k-1]} + P_{[k|k]} \cdot C_k^T \cdot V^{-1} \cdot (y_k - C_k \cdot \hat{x}_{[k|k-1]})$$