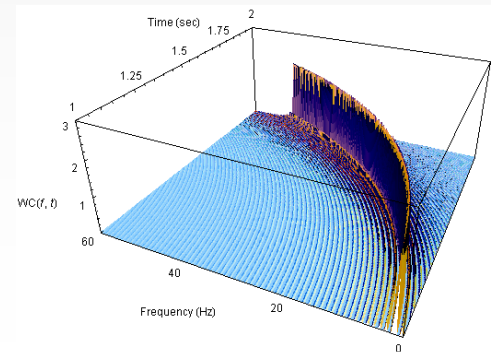


The Chirp-Wigner Transform

Prof. L. Gelman

New Time-Frequency Transform for Non-Stationary Signals with Any Nonlinear Polynomial Variation of the Instantaneous Frequency



Background

- The Wigner-Ville distribution is an effective time-frequency transform for signals with **only linear** change of the instantaneous frequency (i.e. the chirps)
- It is not suitable for non-stationary signals with **nonlinear** change of the instantaneous frequency (i.e. the higher order chirps).
- An effective time-frequency transform is required to handle the higher order chirps.

Problem Statement

- Normally, for transient vibro-acoustical signals from complex mechanical systems (e.g. turbines, gearboxes, etc.) time variations of the instantaneous frequency are *known* from independent synchronous measurements (e.g. from tachometer signals).
- Sometimes, these frequency variations even are known *a priori*.
- Therefore, the main task for these mechanical systems is **amplitude estimation** for transient signals (on the background of an interference) with *known* nonlinear time variation of the instantaneous frequency.

Signals with the Polynomial Variation of the Instantaneous Frequency

Signal

$$\begin{aligned}x(t) &= A \exp(j2\pi\theta(t)) \\ &= A \exp(j2\pi(f_0 t + V(t)))\end{aligned}$$

$$V(t) = (c_{20}t^2 / 2) + (c_{30}t^3 / 3) + (c_{40}t^4 / 4) + \dots + (c_{n0}t^n / n)$$

The instantaneous frequency

$$\begin{aligned}f_i(t) &= \frac{d}{dt}(f_0 t + V(t)) \\ &= f_0 + c_{20}t + c_{30}t^2 + c_{40}t^3 + \dots + c_{n0}t^{n-1}\end{aligned}$$

The initial frequency

The chirp rate

The frequency acceleration

The Chirp-Wigner Transform

The new transform based on the Wigner distribution is proposed by the author for signals with the polynomial variation of the instantaneous frequency

$$W_c(f, t) = \int_{-\infty}^{\infty} x^*\left(t - \frac{\tau}{2}\right) x\left(t + \frac{\tau}{2}\right) e^{-j2\pi\left[f\tau + \frac{c_2(t)}{2}\tau^2 + \frac{c_3(t)}{3}\tau^3 + \frac{c_4(t)}{4}\tau^4 + \dots + \frac{c_n(t)}{n}\tau^n\right]} d\tau$$

$x(t)$ – signal

$\frac{c_2(t)}{2}$ – frequency speed

$\frac{c_3(t)}{3}$ – frequency acceleration

$\frac{c_N(t)}{n}$, for $n \geq 4$ – higher order parameters

The Chirp-Wigner Transform

The transform is real valued

**Kernel coefficients:
matching conditions**

$$c_{2k+1}(t) = \frac{V^{(2k+1)}(t)}{2^{2k}(2k)!}, \quad k = 1, \dots, r \quad \text{and} \quad r = \lfloor (n-1)/2 \rfloor$$

$$c_3(t) = \frac{3c_{40}t + c_{30}}{4}, \quad c_2(t) = c_4(t) = \dots = c_n(t) = 0 \quad n = 4$$

The Wigner Distribution and the Chirp-Wigner Transform

Let's assume that the higher order chirp is of order 4 and time variation of the instantaneous frequency is *known* from a frequency sensor

The Wigner distribution

$$W(t, f) = A^2 \int_{-\infty}^{\infty} \exp\{-j2\pi\tau[f - f_i(t)]\} \exp\left\{-j2\pi\left[\left(-\frac{c_{30}}{12} - \frac{c_{40}}{4}t\right)\tau^3\right]\right\} d\tau$$

The chirp-Wigner transform

$$W_c(t, f) = A^2 \int_{-\infty}^{\infty} \exp\{-j2\pi\tau[f - f_i(t)]\} \exp\left\{-j2\pi\left[\frac{c_2(t)}{2}\tau^2 + \left(\frac{c_3(t)}{3} - \frac{c_{30}}{12} - \frac{c_{40}}{4}t\right)\tau^3 + \frac{c_4(t)}{4}\tau^4 + \dots + \frac{c_N(t)}{N}\tau^N\right]\right\} d\tau$$

The Wigner Transform

- It can be seen that the traditional Wigner distribution of the signal under consideration has the *non-unity time-varying* multiplier defined by the second exponential function in the integral
- This multiplier depends on time, the frequency acceleration and the higher order parameter of the signal.
- One can find that

$$W(t, f) \neq A^2 \int_{-\infty}^{\infty} \exp\{-j2\pi\tau[f - f_i(t)]\} d\tau = A^2 \delta[f - f_i(t)]$$

The Chirp-Wigner Transform (CWT)

- It can be seen from equation of the CWT that the proposed transform also has the time-varying multiplier defined by the second exponential function in the integral
- However, one can find from expression of the CWT that if the following adaptation conditions

$$c_3(t) = \frac{3c_{40}t + c_{30}}{4}, \quad c_2(t) = c_4(t) = \dots = c_N(t) = 0$$

apply for the chirp-Wigner transform, then this multiplier is equal to unity at any time and parameters and, hence, the new transform with adaptation conditions can be written as follows:

$$W_c(t, f) = A^2 \delta[f - f_i(t)]$$

Property of the New Transform

It can be seen that the chirp-Wigner transform with the *adaptation conditions* ideally follows the nonlinear polynomial frequency variation of the signal without amplitude errors.

Properties of the New Transform

- The transform kernel is matched to the instantaneous phase of the signal
- The transform is adaptive
- The transform is quadratic in the signal, and, hence non-linear.
- **Kernel parameters are time dependent**

Adaptation of the New Transform

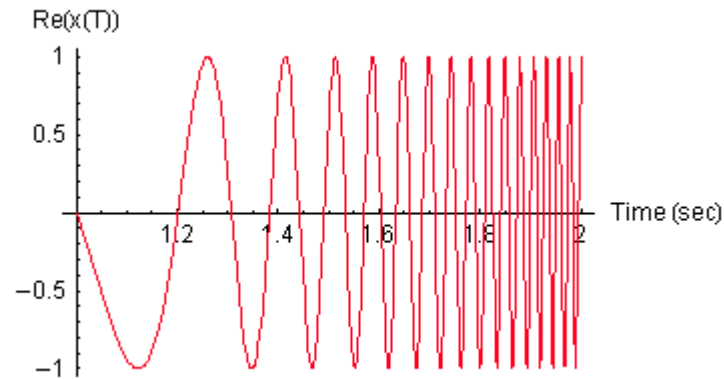
- The new transform is adaptive; this is the major advantage over the Wigner-Ville distribution.
- Adaptive amplitude estimation of the higher-order chirp could be carried out using a two-step procedure:
 - independent estimation of the time dependency of the instantaneous frequency
 - use of the chirp-Wigner transform with the adapted kernel coefficients

Case Study

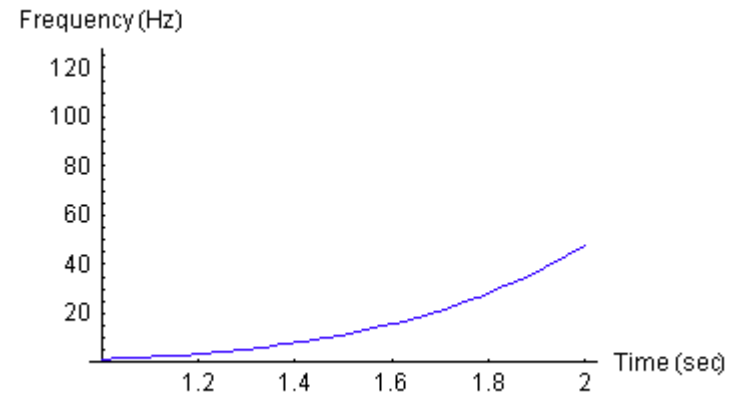
Signal $x(t) = \exp(j2\pi t^6/4)$

**The
instantaneous
frequency**

$$f_i(t) = 6t^5/4$$

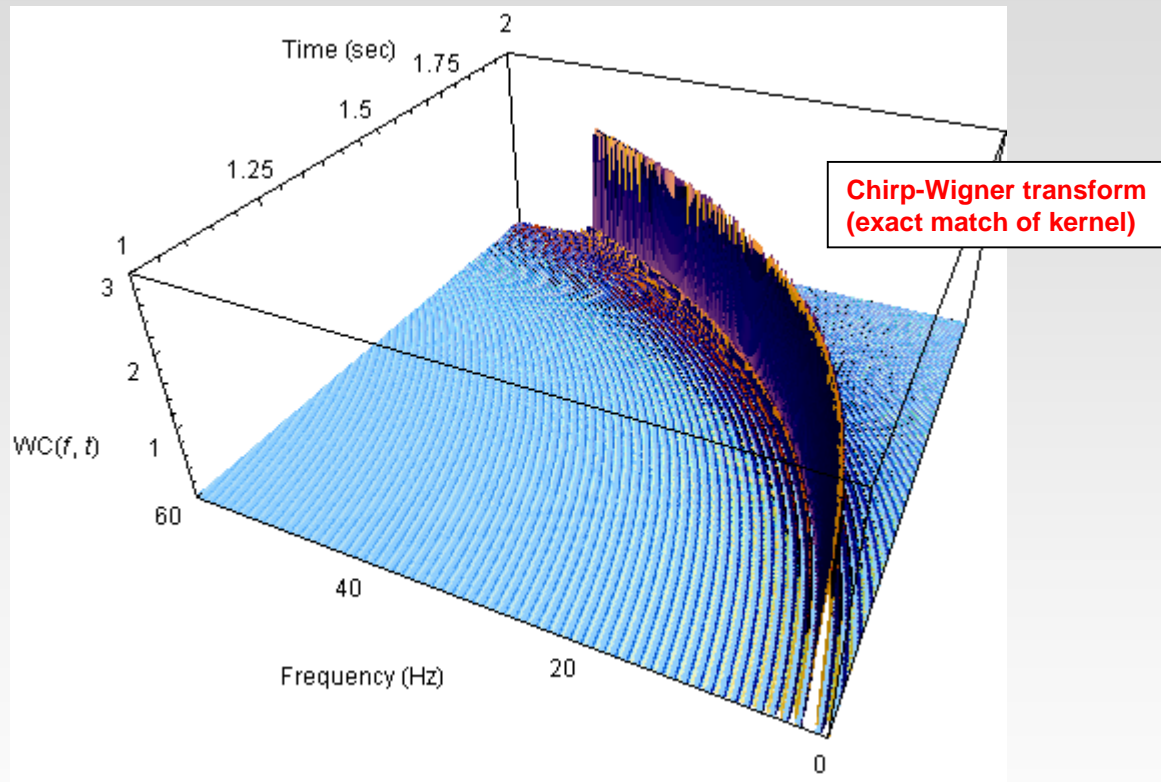


Signal



**The instantaneous
frequency**

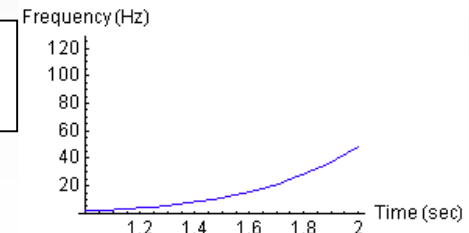
The Chirp-Wigner Transform



Signal

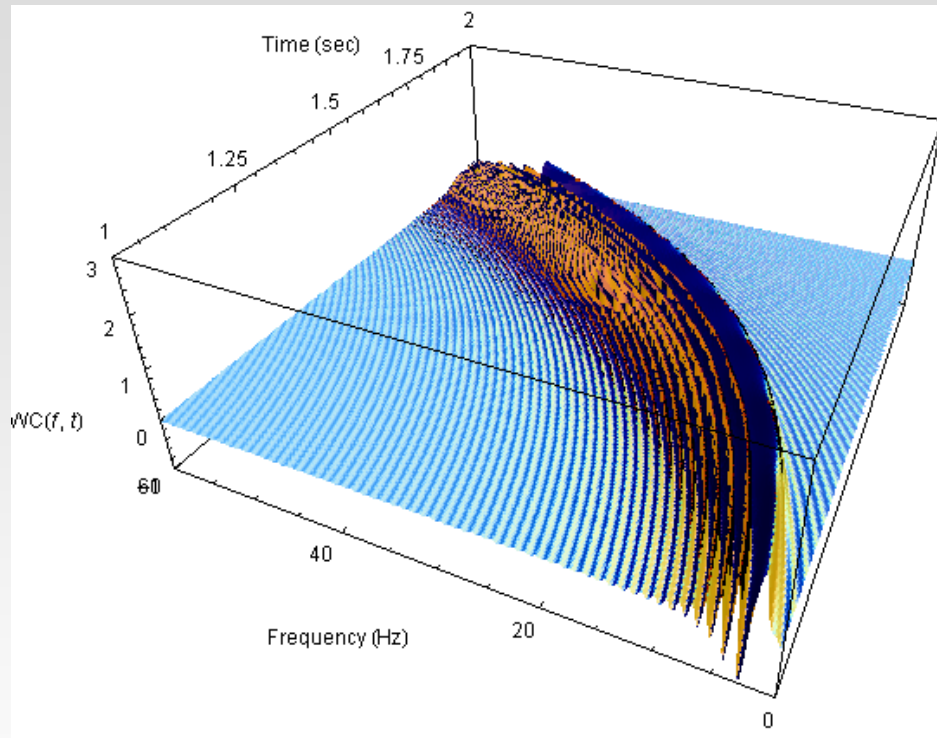
$$x(t) = \exp(j2\pi t^6/4)$$

The instantaneous
frequency of the signal



The chirp-Wigner transform ideally follows the instantaneous frequency without amplitude errors

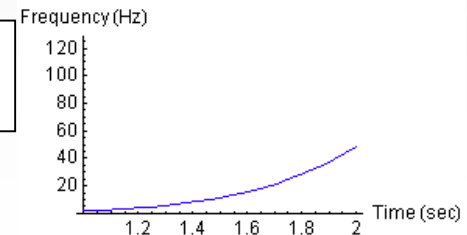
The Wigner Transform



Signal

$$x(t) = \exp(j2\pi t^6/4)$$

The instantaneous
frequency of the signal



The Wigner transform follows the instantaneous
frequency **with essential amplitude errors**

The Discrete Chirp-Wigner Transform: Finite Signal

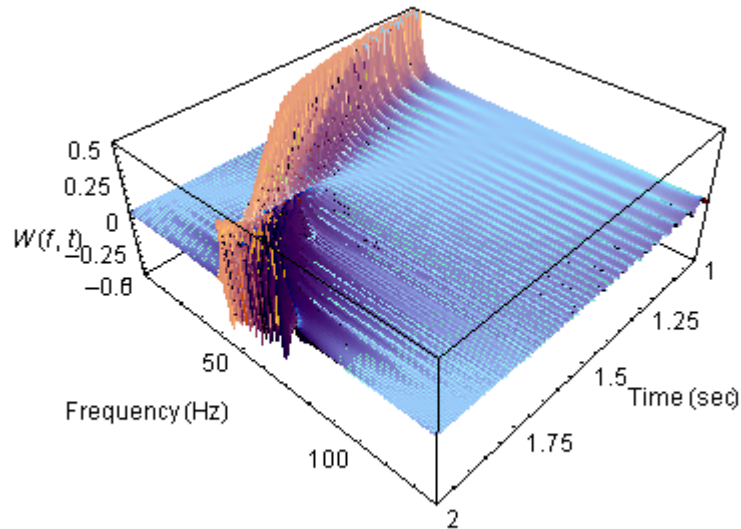
$$W_{cd}(n, r) = \frac{2}{N} \sum_{k=0}^{N-1} x^*(n-k)x(n+k) \exp \left(-j2\pi \left[\frac{2rk}{N} + \frac{c_3(n)}{3}(2k)^3 + \dots + \frac{c_{2r+1}(n)}{2r+1}(2k)^{2r+1} \right] \right)$$

$$W_{cd}(n, r/2) = DFT_N \left(2x^*(n-k)x(n+k) \exp \left(-j2\pi \left[\frac{c_3(n)}{3}(2k)^3 + \dots + \frac{c_{2r+1}(n)}{2r+1}(2k)^{2r+1} \right] \right) \right)$$

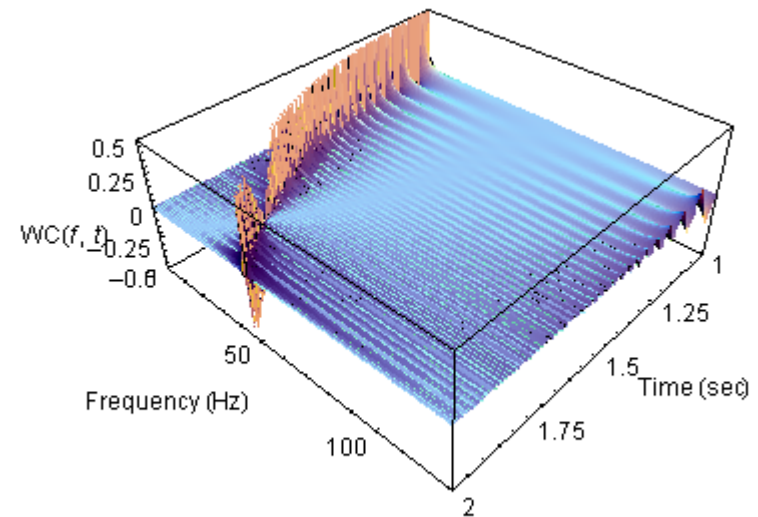
The DFT/FFT could be employed for estimation of the discrete chirp-Wigner transform.

The Chirp-Wigner Transform vs. the Wigner Transform

Discrete Wigner-Ville

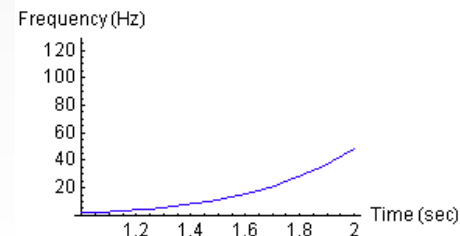


Discrete chirp-Wigner



Signal

$$x(t) = \exp(j2\pi t^6/4)$$



The Chirp-Wigner Transform vs. the Wigner Transform

- For signals with nonlinear variation of instantaneous frequency, the chirp-Wigner transform tracks the instantaneous **amplitude** better than the Wigner distribution.
- The new transform is a radical improvement over the traditional Wigner distribution.