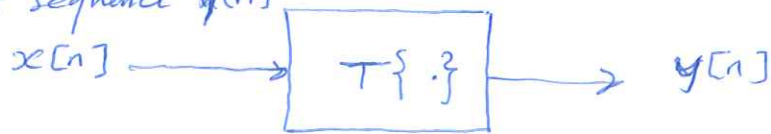


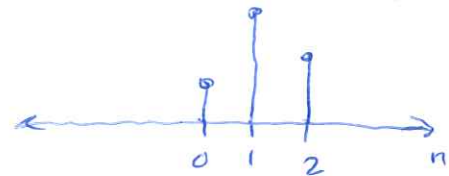
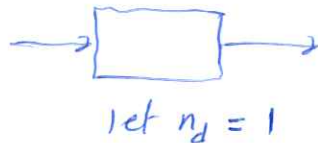
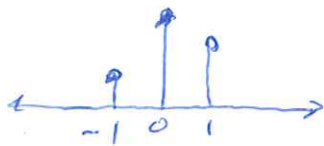
Discrete-Time Systems

* it is a transformation that maps an input sequence $x[n]$ into output sequence $y[n]$



$$y[n] = T\{x[n]\}$$

example: delay system $y[n] = x[n - n_d]$, $-\infty < n < \infty$



$$y[n] = x[n-1]$$

Properties:

1 Memoryless / Memory Systems:

Memoryless: if the output $y[n]$ at every value of n depends only on input $x[n]$ at the same value of n .

Examples:- a) $y[n] = (x[n])^2$

memoryless

b) $y[n] = x[n-1]$

not memoryless

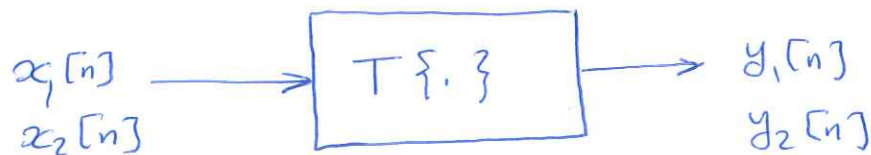
c) $y[n] = x[n+1]$

not memoryless.

d) $y[n] = x[n^2]$

not memoryless

2 Linear Systems:-



* The system is linear iff:-

$$T\{x_1[n] + x_2[n]\} = T\{x_1[n]\} + T\{x_2[n]\} \\ = y_1[n] + y_2[n]$$

$$\text{and } T\{ax[n]\} = aT\{x[n]\} = ay[n], \quad a: \text{constant}$$

* In general:

$$\text{let } x_3[n] = ax_1[n] + bx_2[n]$$

$$T\{x_3[n]\} = T\{ax_1 + bx_2\}$$

$$= aT\{x_1\} + bT\{x_2\}$$

} superposition principle

Example:- Ideal delay system: $y[n] = x[n - n_d]$

$$\text{let } x_3[n] = ax_1[n] + bx_2[n]$$

$$y_3[n] = T\{x_3[n]\} = x_3[n - n_d]$$

$$= ax_1[n - n_d] + bx_2[n - n_d]$$

$$= ay_1[n] + by_2[n]$$

\Rightarrow Linear

Example: $y[n] = (x[n])^2$

$$y_3[n] = T\{x_3[n]\} = (x_3[n])^2 = (ax_1 + bx_2)^2 \\ \neq a(x_1[n])^2 + b(x_2[n])^2 \\ \neq ay_1[n] + by_2[n]$$

It is not Linear

c) The accumulator system $y[n] = \sum_{k=-\infty}^n x[k]$

It is called so, since the output at time (n) is the accumulation (sum) of the present and all past input samples

$$\begin{aligned} \text{let } x_3[n] &= ax_1[n] + bx_2[n] \\ T\{x_3[n]\} &= T\{ax_1[n] + bx_2[n]\} \\ &= \sum_{k=-\infty}^n x_3[k] \\ &= \sum_{k=-\infty}^n ax_1[k] + bx_2[k] \\ &= a \sum_{k=-\infty}^n x_1[k] + b \sum_{k=-\infty}^n x_2[k] \\ &= ay_1[n] + by_2[n] \end{aligned}$$

The accumulator satisfies the superposition principle for all inputs \Rightarrow it is Linear system.

d) $w[n] = \log_{10}(|x[n]|)$

This system is nonlinear, to prove that, one counter example is enough.

$$\begin{aligned} x_1 &= 1 \longrightarrow w_1 = \log 1 = 0 \\ x_2 &= 10 \longrightarrow w_2 = \log 10 = 1 \\ x_3 = x_1 + x_2 &= 11 \longrightarrow w_3 = \log_{10} 11 \neq w_1 + w_2 \end{aligned}$$

[3] Time-Invariant (shift-Invariant) systems

It is a system for which a time shift (Delay) of the input sequence causes a corresponding (similar) shift in the output sequence.

$$\text{if } x[n] \longrightarrow \boxed{T\{\cdot\}} \longrightarrow y[n]$$

$$\text{then } x_1[n] = x[n-n_0] \longrightarrow \boxed{T\{\cdot\}} \longrightarrow y_1[n] = y[n-n_0] \\ \text{for all } n_0.$$

$$\left\{ \begin{array}{l} \text{Test: we solve for both } y[n-n_0] \\ y_1[n] = T\{x_1[n]\} \\ \text{where } x_1[n] = x[n-n_0]. \\ \text{if } y[n-n_0] = y_1[n] \implies \text{The system is time-Invariant} \end{array} \right.$$

Example: Accumulator $y[n] = \sum_{k=-\infty}^n x[k]$

we Evaluate $y[n-n_0] = \sum_{k=-\infty}^{n-n_0} x[k]$

we define $x_1[n] = x[n-n_0]$, then evaluate $y_1[n]$

$$\Rightarrow y_1[n] = \sum_{k=-\infty}^n x_1[k] = \sum_{k=-\infty}^n x[k-n_0]$$

change of variable $m = k - n_0$

$$\Rightarrow y_1[n] = \sum_{m=-\infty}^{n-n_0} x[m]$$

(m) and (k) are dummy indices

$$\Rightarrow y_1[n] = y[n-n_0] \Rightarrow \text{Accumulator is Time-Invariant}$$

$$\text{Examples: } \left. \begin{array}{l} \textcircled{1} y[n] = x[n-n_0] \\ \textcircled{2} y[n] = (x[n])^2 \\ \textcircled{3} y[n] = \log_{10}(|x[n]|) \end{array} \right\} \text{All are Time-Invariant}$$

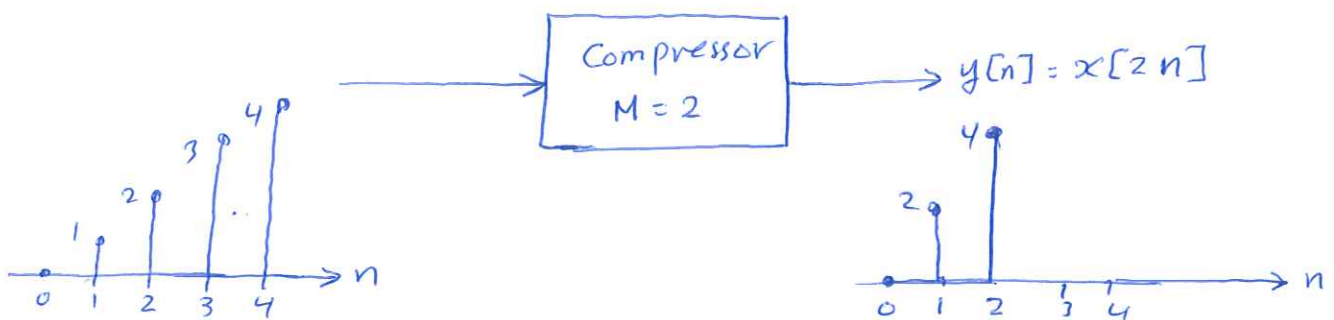
Example: Compressor System

$$y[n] = x[Mn]$$

$$-\infty < n < \infty$$

M : ~~p~~positive integer

The compressor discards $(M-1)$ samples out of M samples. It creates the output sequence by selecting every M^{th} samples.



Test: $x[2n] = y[n]$

$$x_1[n] = x[n-n_0] \Rightarrow y_1[n] = x_1[2n] = x[2n-n_0]$$

$$\text{also } y[n-n_0] = x[2(n-n_0)] = x[2n-2n_0] \neq y_1[n]$$

$\Rightarrow y[n] = x[Mn]$ is not time-invariant

4 Causality

A system is causal if, for every choice of n_0 , the output sequence value at $n = n_0$ depends only on the input sequence value for $n \leq n_0$.

Examples:- consider the following difference systems:-

① $y[n] = x[n+1] - x[n]$ (Forward difference system)

It is not causal, since the current value of the output depends on a future value of the input

② $y[n] = x[n] - x[n-1]$ (Backward difference system)

It is causal.

5 Stability

* A system is stable in the Bounded-Input, Bounded-Output (BIBO) sense if and only if every bounded input produces a bounded output.

* The input $x[n]$ is bounded if there exists a fixed positive finite value B_x such that $|x[n]| \leq B_x < \infty$ for all n .

* Stability requires that, for every bounded input, there exists a fixed +ve finite value B_y such that $|y[n]| \leq B_y < \infty$ for all n .

* If there is one bounded input for which the system property does not hold, then the system does not have the stability property.

Examples: consider the following systems:

① $y[n] = \log_{10}(|x[n]|)$

It is unstable : let $x[n] = 0 < B_x < \infty$
then $y[n] = \infty$

② $y[n] = (x[n])^2$

It is stable : for every $|x[n]| \leq B_x$

$$|y[n]| = |x[n]|^2 \leq B_x^2$$

$$\Rightarrow B_y = B_x^2$$

③ Accumulator $y[n] = \sum_{k=-\infty}^n x[k]$

It is unstable:

$$\text{let } x[n] = u[n] \leq B_x \quad \text{where } B_x = 1$$

$$\text{then } y[n] = \sum_{k=-\infty}^n u[k] = \begin{cases} 0 & n < 0 \\ n+1 & n \geq 0 \end{cases}$$

there is no finite choice for B_y such that

$$(n+1) \leq B_y < \infty \quad \text{for all } n.$$

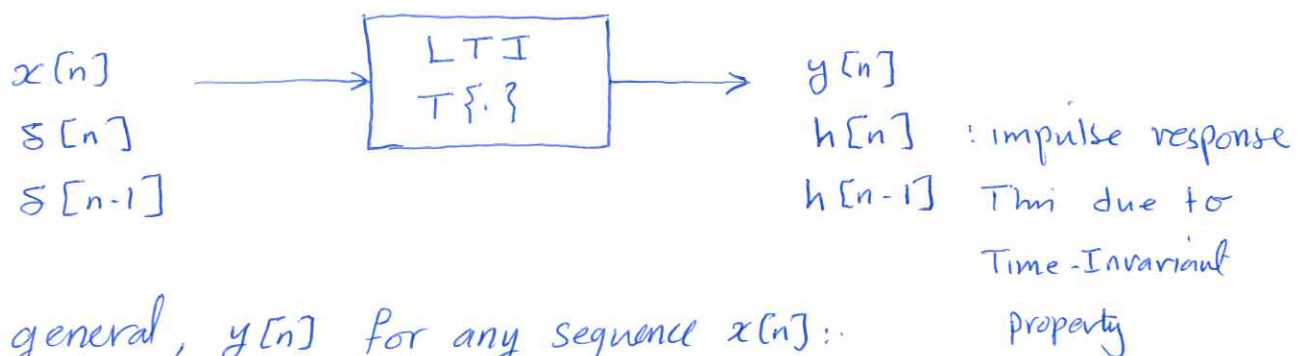
2.3 LTI Systems

* Linear-Time-Invariant (LTI) systems have significant Signal Processing applications

* Remember, any sequence can be expressed as

$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n-k]$$

* Consider the system $y[n] = T\{x[n]\}$



* In general, $y[n]$ for any sequence $x[n]$:

$$y[n] = T\{x[n]\} = T\left\{\sum_k x[k] \delta[n-k]\right\}$$

due to linearity:

$$\Rightarrow y[n] = \sum_k x[k] T\{\delta[n-k]\}$$

$$\boxed{y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]} \quad \text{for all } n$$
$$\boxed{y[n] = x[n] * h[n]} \quad \text{Convolution Sum}$$

* Conclusion: LTI system is completely characterized by its impulse response $h[n]$ in the sense that, given $x[n]$ and $h[n]$ for all n , it is possible to compute each sample of the output sequence $y[n]$.

* properties:

$$h[n-n_0] * x[n] = y[n-n_0]$$

$$h[n-n_1] * x[n-n_2] = y[n-n_1-n_2]$$

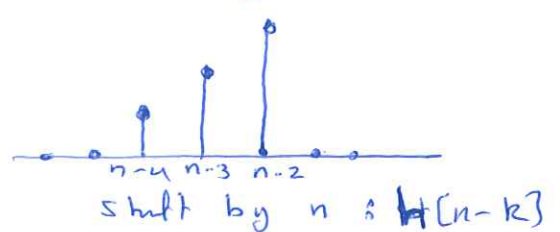
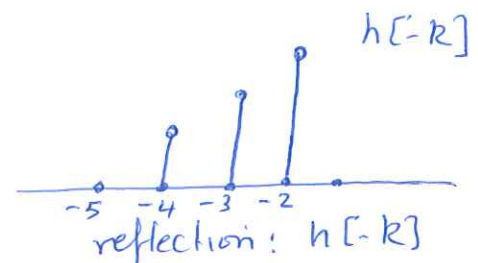
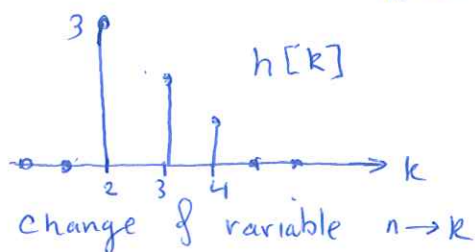
Example: Evaluate $y[n]$ for $h[n] = \{3, 2, 1\}$

$n=2$

$x[n] = \{2, 1, 3\}$

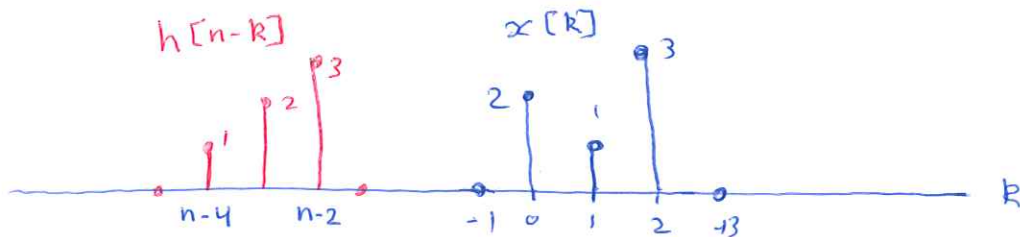
$n=0$

Solution: $y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$



Then,

Sweep \rightarrow Multiply \rightarrow Sum



For $n-2 < 0 \Rightarrow n < +2$, $y[n] = 0$

For $n-2 = 0 \Rightarrow n = 2$, $y[n] = 3 \times 2 = 6$

For $n-2 = 1 \Rightarrow n = 3$, $y[n] = 3 \times 1 + 2 \times 2 = 7$

$n-2 = 2 \Rightarrow n = 4$, $y[n] = 2 \times 1 + 2 \times 2 + 3 \times 3 = 13$

$n-2 = 3 \Rightarrow n = 5$, $y[n] = 1 \times 1 + 2 \times 3 = 7$

$n-2 = 4 \Rightarrow n = 6$, $y[n] = 1 \times 3 = 3$

$n-2 > 4 \Rightarrow n > 6$, $y[n] = 0$

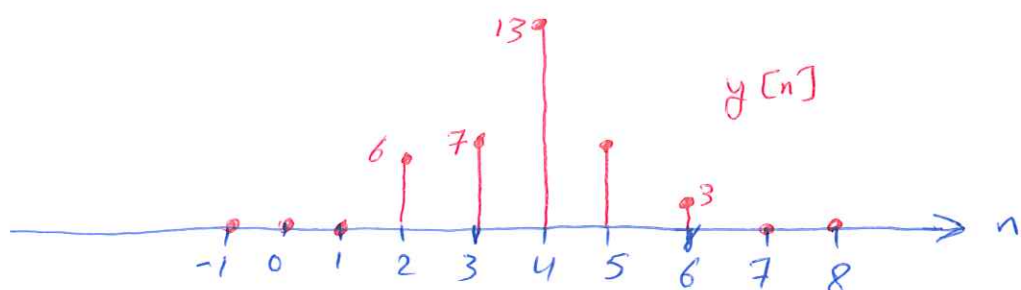


Table Method ① :-

$x[n] \backslash h[n]$	3	2	1
2	$2 \times 3 = 6$ → 6	$2 \times 2 = 4$ → 7	$2 \times 1 = 2$ → 13
1	$1 \times 3 = 3$	$1 \times 2 = 2$	$1 \times 1 = 1$
3	$3 \times 3 = 9$	$3 \times 2 = 6$	$3 \times 1 = 3$

$$\Rightarrow y[n] = \{6, 7, 13, 7, 3\}$$

\uparrow
 $n=2$

Note that

$$N_h : [2, 4]$$

$$N_x : [0, 2]$$

$$N_y : [2+0, 4+2] = [2, 6]$$

Table Method ② :-

n	2	3	4	5	6
$x[0] \times h[n]$	3 ⁶	2 ⁴	1 ²		
$x[1] \times h[n-1]$		3 ³	2 ²	1 ¹	
$x[2] \times h[n-2]$			3 ⁴	2 ⁶	1 ³
$y[n] = \Sigma$	6	7	13	7	3

OR

n	2	3	4	5	6
$\times 3$	2 ⁶	1 ³	3 ⁹		
$\times 2$		2 ⁴	1 ²	3 ⁶	
$\times 1$			2 ²	1 ¹	3 ³
$y[n] = \Sigma$	6	7	13	7	3

Example let $h[n] = \{1, 2, 1, -1\}$, Determine $y[n]$
 for $x[n] = \{1, 2, 3, 1\}$

$$N_y = [-1, 2] + [0, 3] = [-1, 5]$$

n	-1	0	1	2	3	4	5
$x[n-0]$	1	2	3	1			
$x[n-1]$		1	2	3	1		
$x[n-2]$			1	2	3	1	
$x[n-3]$				1	2	3	1
$h[0]x[n-0]$	1	2	3	1			
$h[1]x[n-1]$		2	4	6	2		
$h[2]x[n-2]$			1	2	3	1	
$h[3]x[n-3]$				-1	-2	-3	-1
$y[n] = \sum$	1	4	8	8	3	-2	-1

$$y[n] = \sum h[k] x[n-k]$$

$$y[-1] = \sum h[k] x[-1-k] = \dots h[-1]x[0] + h[0]x[-1] + h[1]x[-2] + \dots$$

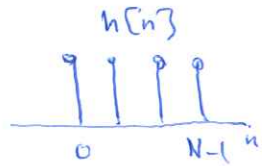
$$= 1$$

$$\Rightarrow y[n] = \{1, 4, 8, 8, 3, -2, -1\}$$

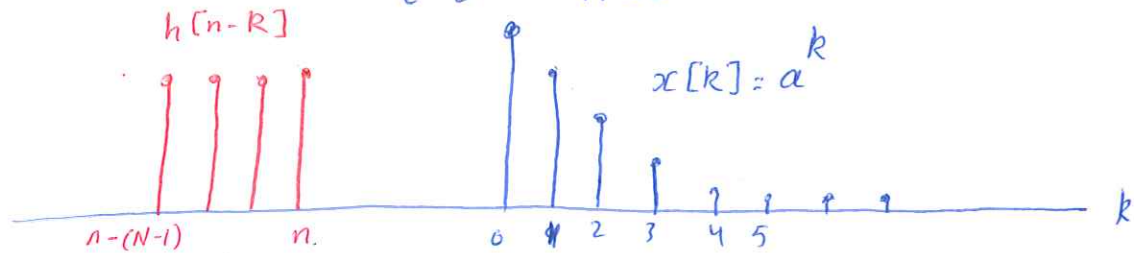
\uparrow
 $n=0$

Example 2.11 Analytical Evaluation of the convolution sum.

$$h[n] = u[n] - u[n-N] = \begin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$



$$x[n] = a^n u[n] = \begin{cases} a^n & n \geq 0 \\ 0 & n < 0 \end{cases}$$



we can find formulas for $y[n]$ for different sets of values of n .

Case I

for $n < 0$ no overlap $\Rightarrow y[n] = \sum_k x[k] h[n-k] = 0$

Case II

for $0 \leq n$ and $n-N+1 \leq 0 \Rightarrow 0 \leq n \leq N-1$
 $x[k] h[n-k] = a^k, 0 \leq k \leq n$, the mult. is non-zero

$$\Rightarrow y[n] = \sum_{k=0}^n a^k$$

using finite geometric series $\sum_{k=N_1}^{N_2} a^k = \frac{a^{N_1} - a^{N_2+1}}{1-a}, N_2 \geq N_1$

$$\Rightarrow y[n] = \frac{a^0 - a^{n+1}}{1-a} = \frac{1 - a^{n+1}}{1-a}$$

Case III

for $(n-N+1) > 0 \Rightarrow n > N-1$

$x[k] h[n-k] = a^k, n-N+1 \leq k \leq n$, the mult. is non-zero.

$$\Rightarrow y[n] = \sum_{k=n-N+1}^n a^k = \frac{a^{n-N+1} - a^{n+1}}{1-a} = a^{n-N+1} \left(\frac{1 - a^N}{1-a} \right)$$

$$\Rightarrow y[n] = \begin{cases} 0 & n < 0 \\ \frac{1-a^{n+1}}{1-a} & 0 \leq n \leq N-1 \\ a^{n-N+1} \left(\frac{1-a^N}{1-a} \right) & n > N-1 \end{cases}$$

note:-

since $h[n]$ or $x[n]$ is infinite sequence, the analytical evaluation is the appropriate method to evaluate the convolution sum.

if Both $h[n]$ and $x[n]$ are finite sequences, then the tabular method is easier.

Example : Find $y[n]$ for the LTI system

$$h[n] = a^n u[n], \quad |a| < 1$$

when $x[n] = u[n]$

Solution:

$$\begin{aligned} y[n] &= \sum_{k=-\infty}^{\infty} x[n-k] h[k] \\ &= \sum_{k=-\infty}^{\infty} u[n-k] \cdot a^k u[k] \end{aligned}$$

$$u[k] = 0 \quad \text{for } k < 0$$

$$u[n-k] = 0 \quad \text{for } n-k < 0 \Rightarrow n < k$$

$$\begin{aligned} \Rightarrow y[n] &= \sum_{k=0}^n a^k \\ &= \begin{cases} \frac{1 - a^{n+1}}{1 - a} & n \geq 0 \\ 0 & n < 0 \end{cases} \end{aligned}$$

$$\Rightarrow y[n] = \frac{1 - a^{n+1}}{1 - a} u[n]$$

$$\text{note : } \lim_{n \rightarrow \infty} y[n] = \frac{1}{1 - a}$$

OR The step response can be obtained directly from $h[n]$

$$u[n] \longrightarrow \boxed{h[n]} \longrightarrow y[n] = \sum_{k=-\infty}^n h[k]$$

$$\Rightarrow y[n] = \sum_{k=-\infty}^n a^k u[k] = \sum_{k=0}^n a^k$$

2.4 Properties of LTI Systems

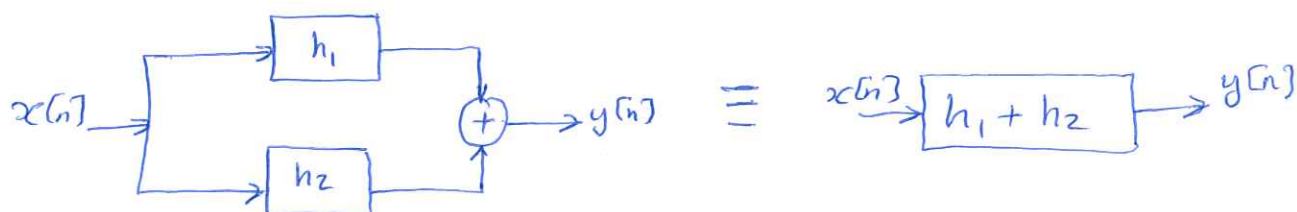
[1] Convolution operation is commutative

$$y[n] = x[n] * h[n] = h[n] * x[n]$$

$$y[n] = \sum_{m=-\infty}^{\infty} x[n-m] h[m] = \sum_{m=-\infty}^{\infty} h[n-m] x[m]$$

[2] Convolution is Distributive over addition

$$y[n] = x[n] * h_1[n] + x[n] * h_2[n] = x[n] * (h_1[n] + h_2[n])$$



[3] Convolution satisfies the Associative Property:

$$y[n] = (x[n] * h_1[n]) * h_2[n] = x[n] * (h_2[n] * h_1[n])$$



[4] LTI system is stable if and only if the impulse response ($h[n]$) is absolutely summable

$$B_h = \sum_{k=-\infty}^{\infty} |h[k]| < \infty$$

$$|y[n]| = \left| \sum_{k=-\infty}^{\infty} h[k] x[n-k] \right| \leq \sum_k |h[k]| |x[n-k]|$$

if $x[k]$ is bounded, so that $|x[n]| \leq B_x$

$\Rightarrow |y[n]| \leq B_x B_h = B_y$ iff $h[n]$ is absolutely summable

5 LTI system is causal if $h[n] = 0$ for $n < 0$

It is sometimes convenient to refer to a sequence that is zero for $n < 0$ as a causal sequence.

i.e., $f[n]$ is a causal sequence if $f[n] = 0$ for $n < 0$.

proof. $y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$

if $h[n-k]$ is causal then $h[n-k] = 0$ for $n-k < 0$
 $n < k$

$$\Rightarrow y[n] = \sum_{k=-\infty}^n x[k] h[n-k]$$

LTI system

Impulse Response

Stability
 $|\sum h[n]| < B_h$

① $y[n] = x[n-n_d]$

$h[n] = \delta[n-n_d]$

Yes

② $y[n] = \frac{1}{M_1+M_2+1} \sum_{k=-M_1}^{M_2} x[n-k]$

$$h[n] = \frac{1}{M_1+M_2+1} \sum_{k=-M_1}^{M_2} \delta[n-k]$$

$$= \begin{cases} \frac{1}{M_1+M_2+1} & -M_1 \leq n \leq M_2 \\ 0 & \text{else} \end{cases}$$

Yes

③ $y[n] = x[n+1] - x[n]$

$h[n] = \delta[n+1] - \delta[n]$

Yes

$h[n] = \{1, -1\}$
 \uparrow
 $n=0$

④ $y[n] = x[n] - x[n-1]$

$h[n] = \delta[n] - \delta[n-1]$

Yes

$h[n] = \{1, -1\}$
 \uparrow
 $n=0$

⑤ $y[n] = \sum_{k=-\infty}^n x[k]$

$$h[n] = \sum_{k=-\infty}^n \delta[k]$$

$$= \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases}$$

$$= u[n]$$

No, since

$\sum_{k=0}^{\infty} u[k] = \infty$

$\sum_{n=-\infty}^n |u[n]| = n+1$ as $n \rightarrow \infty$
 $\Sigma \rightarrow \infty$

Conclusions :-

- * In general, a system with a finite-duration impulse response (henceforth referred to as an FIR system) will always be stable, as long as the impulse response values are finite in magnitude. (see examples ①-④)
- * The impulse response of the accumulator (example ⑤) has infinite duration. This is an example of the class of systems referred to as infinite-duration impulse response (IIR) systems.
- * An example of an IIR system that is stable is

$$h[n] = a^n u[n] \quad \text{with } |a| < 1$$

$$\Rightarrow B_h = \sum_{n=0}^{\infty} |a|^n \quad (\text{infinite geometric series})$$

$$= \frac{1}{1 - |a|} < \infty \quad \text{for } |a| < 1$$

$$\Rightarrow \text{it is stable when } |a| < 1$$

however, if $|a| \geq 1$, then $B_h = \infty$

\Rightarrow the system is unstable.

Impulse Response

Causal if $h[n]=0, n < 0$

① $h[n] = \delta[n - n_d]$

Causal if $n_d \geq 0$

Non causal if $n_d < 0$

② $h[n] = \frac{1}{M_1 + M_2 + 1} \sum_{k=-M_1}^{M_2} \delta[n - k]$

Causal if $M_1 \leq 0$
and $M_2 \geq 0$

③ $h[n] = \sum_{k=-\infty}^n \delta[k] = u[n] = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases}$ Causal

④ $h[n] = \delta[n+1] - \delta[n]$

since $h[-1] = \delta[0] - \delta[-1]$

$$= 1 \neq 0$$

\Rightarrow Non causal.

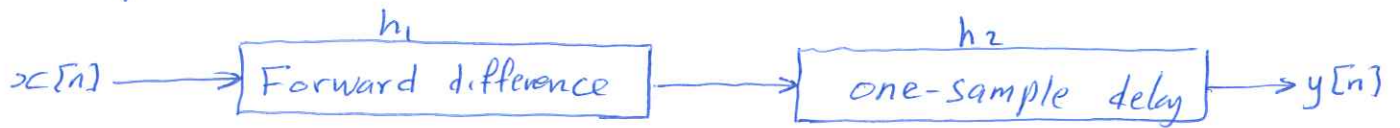
⑤ $h[n] = \delta[n] - \delta[n-1]$

causal.

Note: The convolution of a shifted impulse sequence $\delta[n - n_d]$ with any sequence $x[n]$ is simply evaluated by shifting $x[n]$ by the displacement of the impulse

$$x[n] * \delta[n - n_d] = x[n - n_d]$$

Example:



$$h_1[n] = \delta[n+1] - \delta[n]$$

$$h_2[n] = \delta[n-1]$$

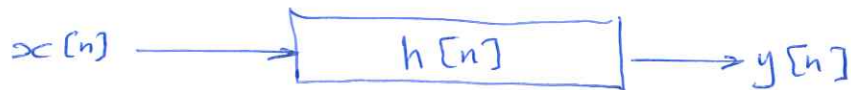
$$h[n] = h_1[n] * h_2[n]$$

$$= (\delta[n+1] - \delta[n]) * (\delta[n-1])$$

$$= \delta[n+1-1] - \delta[n-1]$$

$$= \delta[n] - \delta[n-1]$$

Backward difference system.



Note that the noncausal forward difference system is converted into causal system by cascading them with a delay.

⇒ In general, any noncausal FIR system can be made causal by cascading it with a sufficiently long delay.

6 Inverse System:

In general, if an LTI system has impulse response $h[n]$, then its inverse system, if it exists, has impulse response $h_i[n]$ defined by the relation

$$h[n] * h_i[n] = h_i[n] * h[n] = \delta[n]$$

In ch. 3, Z-transform provides a straight forward method of finding the inverse of an LTI system.

example:- Show that the accumulator is the inverse system of the backward difference system



$$\begin{aligned} h[n] &= u[n] * (\delta[n] - \delta[n-1]) \\ &= u[n] - u[n-1] \\ &= \delta[n] \end{aligned}$$

2.5 Linear Constant-Coefficient Difference equations

An important class of LTI systems consists of those systems for which the input $x[n]$ and the output $y[n]$ satisfy an N^{th} -order linear constant-coefficient difference equation of the form

$$\sum_{k=0}^N \alpha_k y[n-k] = \sum_{m=0}^M b_m x[n-m]$$

Example 2.12 Difference equation representation of the Accumulator.

$$y[n] = \sum_{k=-\infty}^n x[k] = x[n] + \sum_{k=-\infty}^{n-1} x[k]$$

$$= x[n] + y[n-1]$$

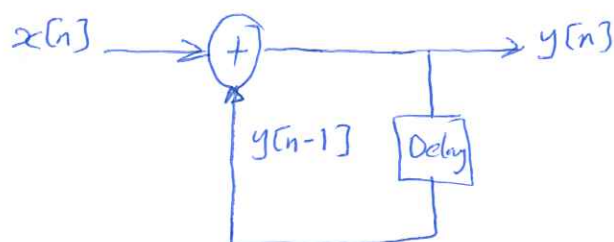
$$\text{or } y[n-0] - y[n-1] = x[n]$$

$$\alpha_0 y[n] + \alpha_1 y[n-1] = b_0 x[n]$$

$$\Rightarrow \alpha_0 = 1, \alpha_1 = -1, N = 1 \Rightarrow 1^{\text{st}} \text{ order Difference equation}$$

$$b_0 = 1, M = 0$$

This suggests a simple block diagram representation (Recursive representation)



- * In general, if we assume $\alpha_0 = 1, \alpha_k = 0$ for $k = 1, 2, \dots, N$, then
- $$y[n] = \sum_{m=0}^M b_m x[n-m]$$
- * the impulse response,
$$h[n] = \sum_{m=0}^M b_m \delta[n-m] = \begin{cases} b_m & n=0, \dots, M \\ 0 & \text{else} \end{cases}$$
- * it is causal

Example:- Determine the impulse response of the first order system given by :-

$$y[n] - a y[n-1] = x[n] \quad , \text{ assume } y[0] = 0, n < 0$$

(causal system)

Solution :-

$$x[n] = \delta[n] = \begin{cases} 1 & n = 0 \\ 0 & \text{else} \end{cases}$$

$$y[n] = x[n] + a y[n-1]$$

$$y[0] = x[0] + a y[-1] = 1 = h[0]$$

$$y[1] = x[1] + a y[0] = 0 + a = h[1]$$

$$y[2] = x[2] + a y[1] = 0 + a \cdot a = a^2 = h[2]$$

$$y[3] = x[3] + a y[2] = a^3 = h[3]$$

⋮

$$h[n] = a^n u[n] = \begin{cases} a^n & n \geq 0 \\ 0 & \text{else} \end{cases}$$

Let us assume that $y[n] = 0$ for $n > 0$ (noncausal system)

then $y[n-1] = \frac{y[n] - x[n]}{a}$

$$= a^{-1} (y[n] - x[n])$$

$$x[n] = \delta[n]$$

$$n = 2 \quad y[1] = a^{-1} (y[2] - x[2]) = 0$$

$$n = 1 \quad y[0] = a^{-1} (y[1] - x[1]) = 0$$

$$n = 0 \quad y[-1] = a^{-1} (y[0] - x[0]) = a^{-1} (0 - 1) = -a^{-1}$$

$$n = -1 \quad y[-2] = a^{-1} (y[-1] - x[-1]) = a^{-1} \cdot -a^{-1} = -a^{-2}$$

⋮

$$\Rightarrow h[n] = \begin{cases} -a^{-1} & n = 0 \\ -a^{-2} & n = -1 \\ -a^{-3} & n = -2 \end{cases} = -a^{n-1} u[-n] = -a^n u[-n-1]$$