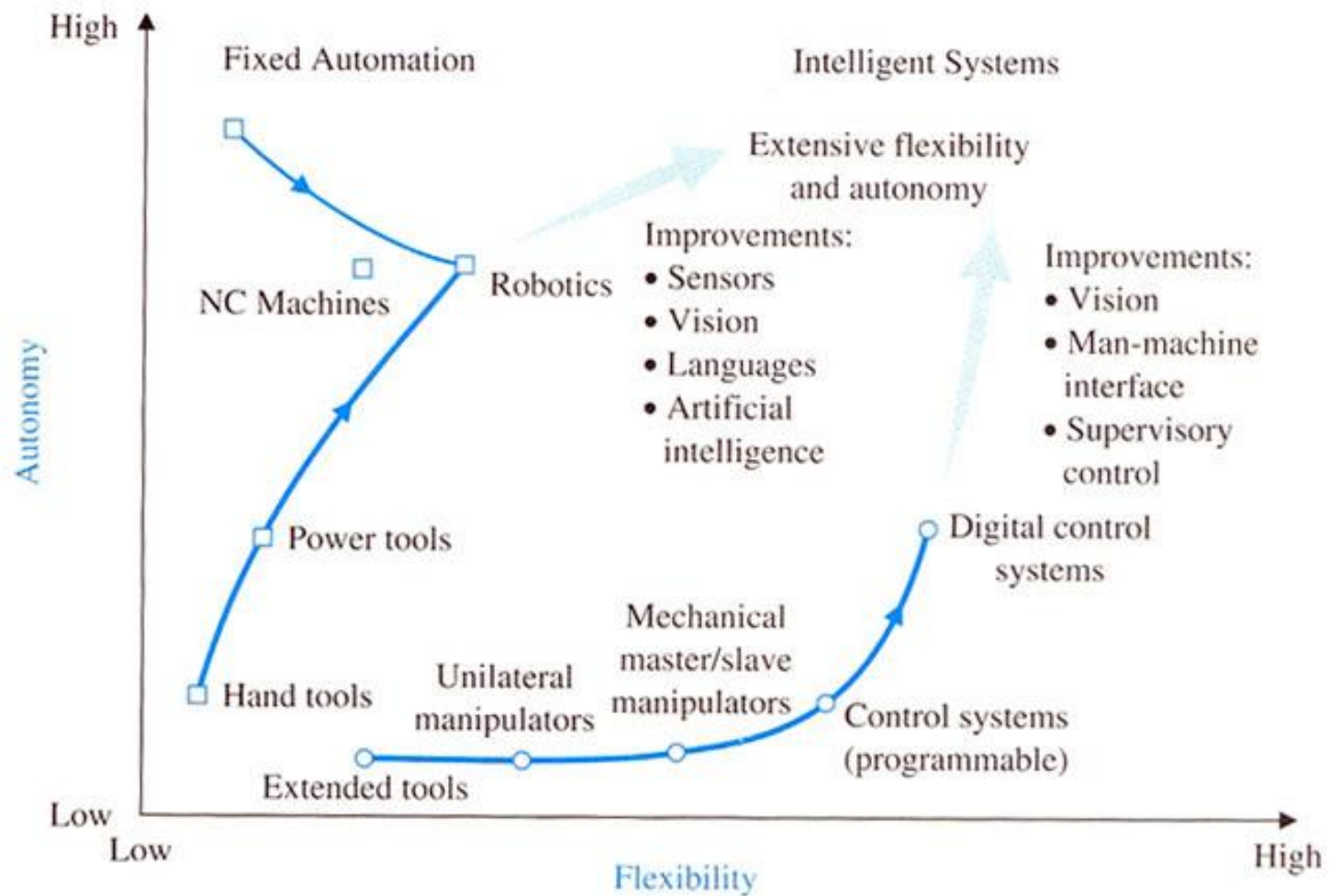
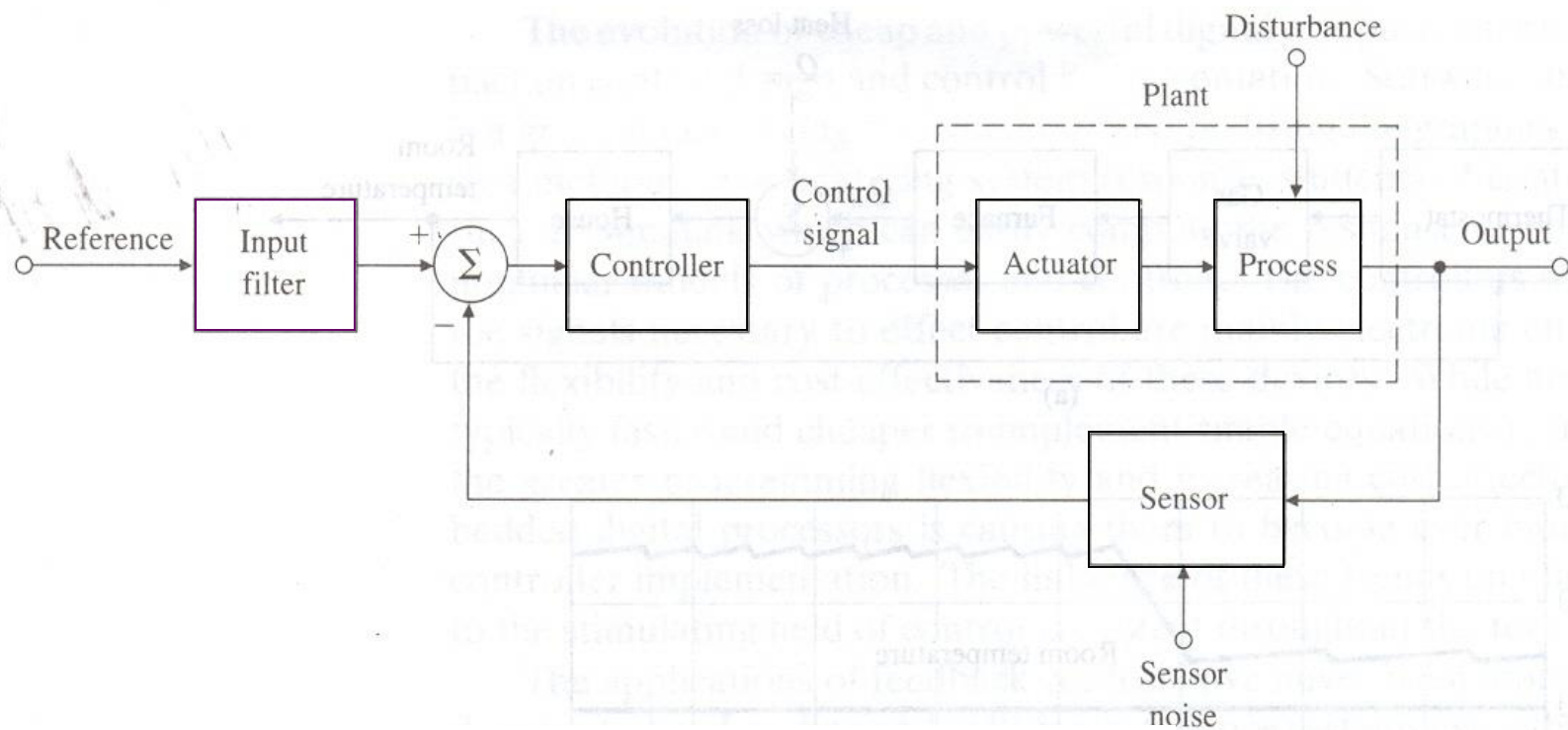


## ***Basic System Modelling***







*Closed Loop Control of a Plant*

# ***Modelling a Dynamical System***

*Models are a mathematical **representations** of system dynamics*

*Models allow the dynamics to be simulated and analyzed, without having to build the system*

*Models are never exact, but they can be predictive*

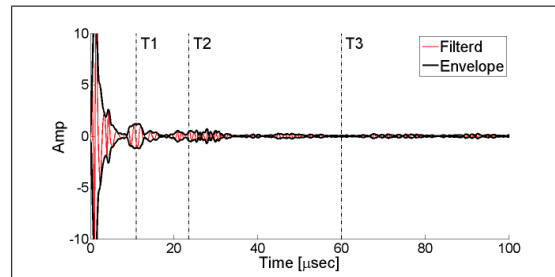
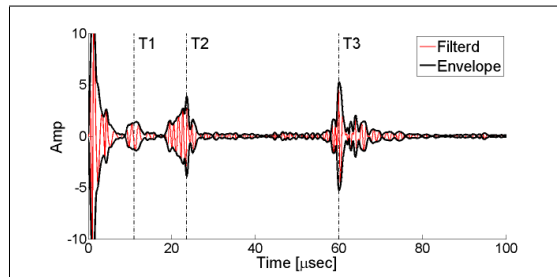
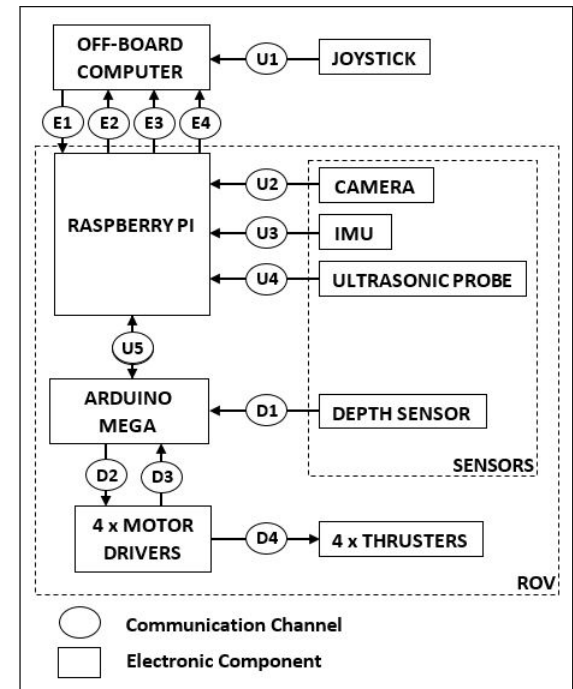
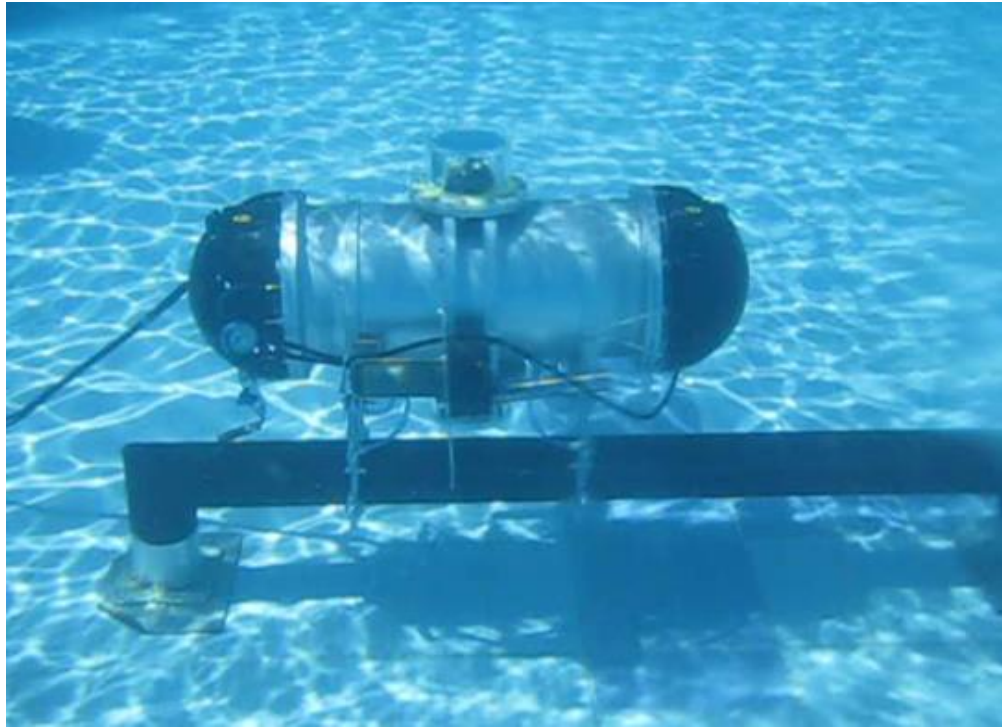
*Models can be used in ways that the system can't perform*

*Certain types of analysis (eg, parametric variations) can't easily be done on the actual system*

*Models can be run much more quickly*

*The model you use depends on the questions you want to answer*

*A single system may have many models*



***A system is modelled either***

- by defining the input-output characteristics or***
- by selecting a set of state variables***

*Inputs describe the external excitation of the dynamics. Inputs are extrinsic to the system dynamics (externally specified). Constant inputs are often considered to be parameters*

*Outputs are variables that are to be calculated or measured*

***Input-output relation is often governed by a generalised ODE of the form***

$$a_n y^{(n)} + \dots + a_2 y^{(2)} + a_1 y^{(1)} + a_0 y = b_m u^{(m)} + \dots + b_1 u^{(1)} + b_0 u(t)$$

***Where  $y^{(n)} = d^n y / dt^n$ ,  $u^{(m)} = d^{(m)} u / dt^m$***



*Choice of inputs and outputs depends on point of view*

*Inputs: what factors are external to the model that you are building*

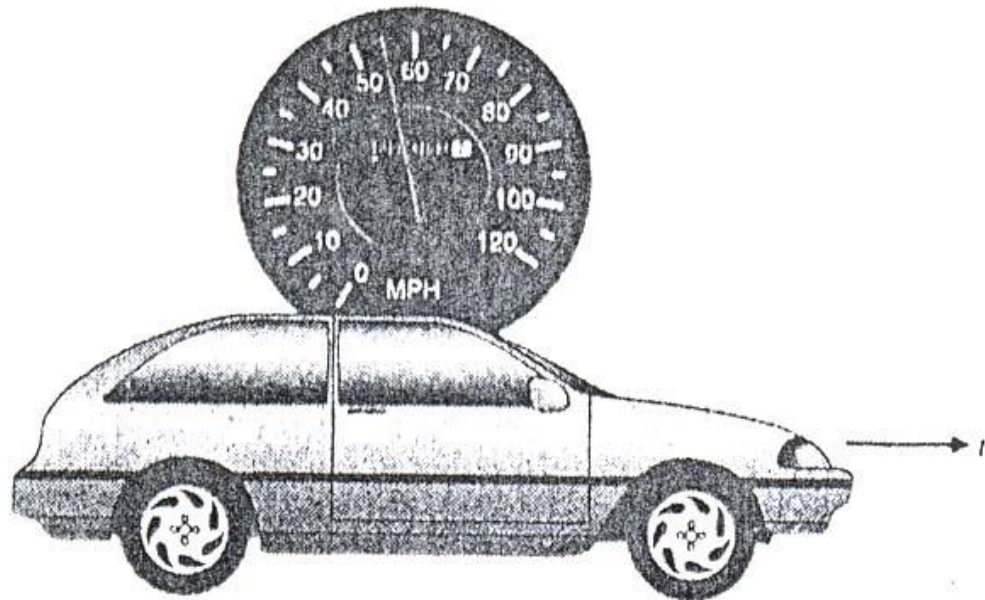
*Inputs in one model might be outputs of another model (eg, the output of a cruise controller provides the input to the vehicle model)*

*Outputs: what physical variables (often states) can you measure*

*Choice of outputs depends on what you can sense and what parts of the component model interact with other component models*

*States: Chosen such that their values at a reference time and the corresponding inputs are known. May or may not include Outputs.*

# A Cruise Control Model



Cruise-control model

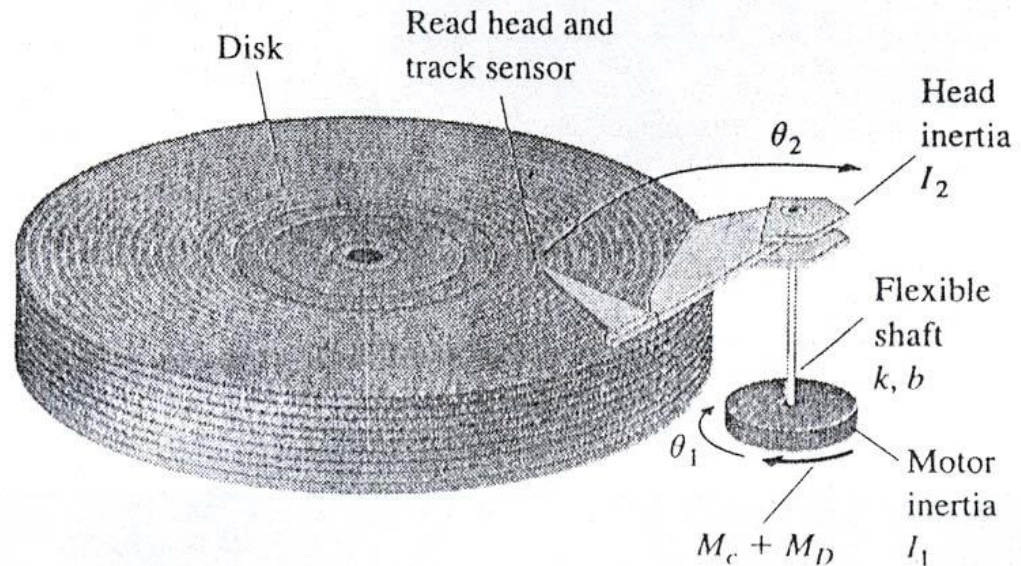
*EOM?*

*Transfer function?*



# Disk Reading

Disk read/write head  
schematic for modeling



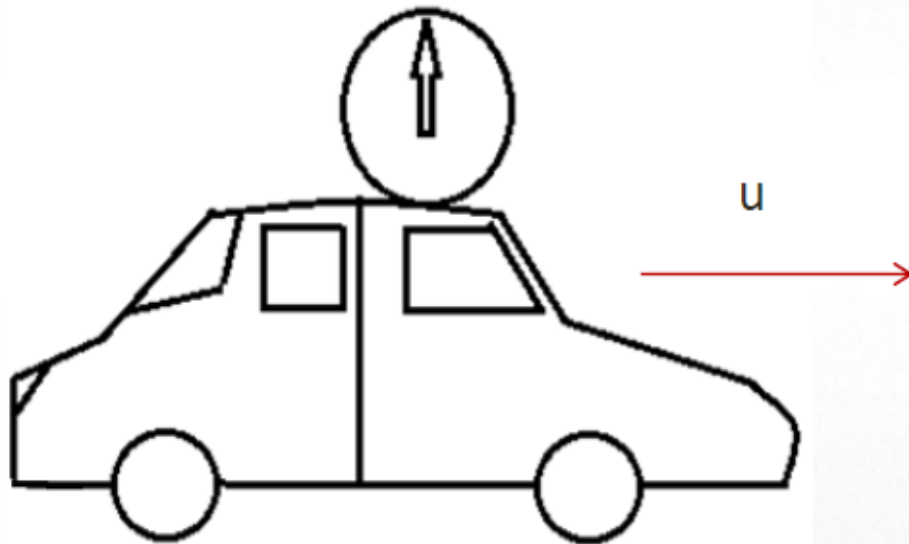
Non-collocated sensing

Input  $M_c$

Output  $\Theta_2$

Let us consider first the problems mooted in the last lecture

*Problem 1:*



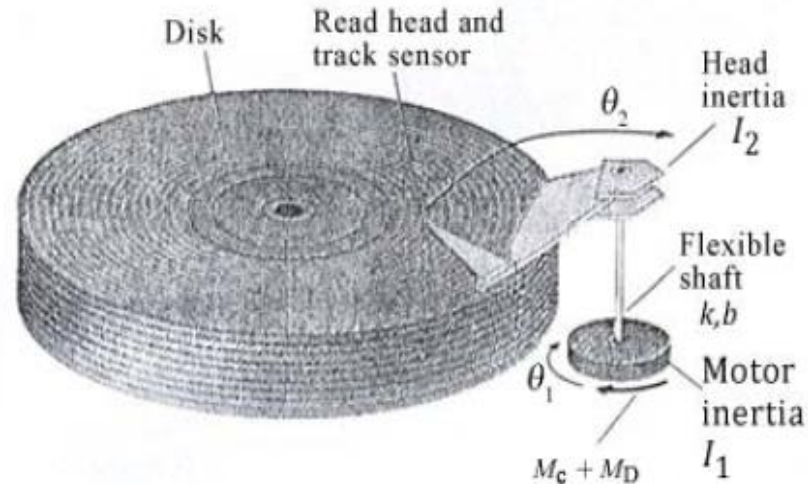
From the cruise control perspective: we can have a simplified model for the automobile system.

Assuming  $M$  to be the mass of the car and  $u$  to be the velocity, the governing EOM could be written as:

$$M \dot{u} + D u = F(t)$$

Where,  $D$  is the Drag /Resistance Coefficient and  $F(t)$  is the cruising force.

### Problem 2:



Input  $M_c$   
Output  $\theta_2$

Again, a simplified model could be developed by neglecting the disturbance, the EOM will be ( $C$  – shaft damping coefficient and  $k$  shaft stiffness)

$$I_2 \ddot{\theta}_2 + C (\dot{\theta}_2 - \dot{\theta}_1) + k (\theta_2 - \theta_1) = M_c(t)$$

$$I_1 \ddot{\theta}_1 + C (\dot{\theta}_1 - \dot{\theta}_2) + k (\theta_1 - \theta_2) = 0$$

# System Transfer Functions

- The last two examples relate to the description of the systems in Time Domain.
- By Laplace transformation one can convert the equations into frequency domain.
- Assuming zero initial condition, the first mathematical model could be expressed as :

$$sMU(s) + BU(s) = \bar{F}(s)$$

$$\frac{U(s)}{\bar{F}(s)} = \frac{1}{sM + B}$$

- Similarly, the second mathematical model could be expressed as:

$$\frac{\theta_2(s)}{\bar{M}_c} = \frac{1}{s^2 I_2 + sC + k}$$

The RHS of both the equations are known as Transfer Function that describes the relationship between output and input. Example 1, pertains to a first order transfer function, while Example 2 presents a second order transfer function. Such relationship could be graphically represented for better perception and this technique will be discussed in this lecture.



# State-Space Modelling

The *state* of a model of a dynamic system is a set of independent physical quantities, the specification of which (in the absence of excitation) completely determines the future evolution of the system

***Dynamics*** describes how the state evolves. The *dynamics* of a model is an update rule for the system state that describes how the state evolves, as a function on the current state and any external inputs

When we talk about electro-mechanical systems modeled by differential equations, such as masses and springs, electric circuits or satellites (rigid bodies) rotating in space, we can attach some additional intuition: the variables in the state should be adequate to specify the **energy** of the system.

For example, take a ball free-falling to earth: we can specify the position of the ball by specifying the height ( $h$ ) above the ground, but we also need to include the velocity of the ball ( $dh/dt$ ) to specify the total energy ( $E = 1/2 * m * (dh/dt)^2 + mgh$ ). Therefore, the state of the ball is  $(h, dh/dt)$ .



## State Space Modelling of a Single Degree of Freedom System

- Consider a SDOF system (with mass  $M$ , stiffness  $K$  and Damping constant  $C$ ) such that the equation of motion corresponding to force excitation is given by:

$$M \ddot{x} + C \dot{x} + K x = F(t)$$

- The following pair of states or their linear combinations could be considered for the modelling:

$$\begin{Bmatrix} x \\ \dot{x} \end{Bmatrix}, \begin{Bmatrix} x \\ \ddot{x} \end{Bmatrix}, \begin{Bmatrix} \dot{x} \\ \ddot{x} \end{Bmatrix}$$

## The EOM in State Space Form

- Consider for example, the position and velocity as the state coordinates.
- The state vector could be written as:

$$X = \begin{Bmatrix} x \\ \dot{x} \end{Bmatrix}$$

- Based on these states, the **EOM** could be rewritten as:

$$\frac{d}{dt} \begin{Bmatrix} x \\ \dot{x} \end{Bmatrix} = \begin{bmatrix} 0 & 1 \\ -k/m & -c/m \end{bmatrix} \begin{Bmatrix} x \\ \dot{x} \end{Bmatrix} + \begin{Bmatrix} 0 \\ 1 \end{Bmatrix} (f/m)$$

$$\dot{X} = AX + BU$$

$$A = \begin{bmatrix} 0 & 1 \\ -k/m & -c/m \end{bmatrix}, \quad B = \begin{Bmatrix} 0 \\ 1 \end{Bmatrix}, \quad U = f/m$$

## Output of a State-space System

- Many a times states of a system are not directly measurable and hence are not of direct interest. For example, if you consider, state-space representation of a finite element model pertaining to a Spacecraft. The number of states could be as high as three to four thousand! However, one cannot have so many sensors to measure all the states. In such cases, we fix a feasible number of outputs that are observable/measurable.
- Suppose for a system of  $n$ -states there are  $r$  outputs that are measurable. Then the output vector  $Y(t)$  of size  $r$  could be represented as a linear combination of input to the system and the states as follows:

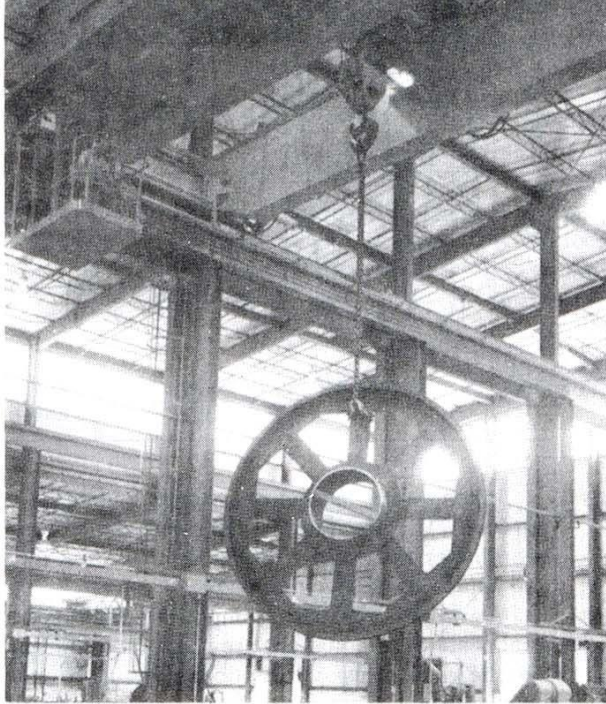
$$Y(t) = \begin{Bmatrix} y_1(t) \\ y_2(t) \\ . \\ . \\ y_r(t) \end{Bmatrix} = C X(t) + D U(t)$$

Where  $C$  &  $D$  are constants for an LTIV system. For majority of dynamic systems it is observed that  $D = 0$ , meaning outputs are not directly affected by the system inputs.

## Time Domain Solution for a Vector State Equation

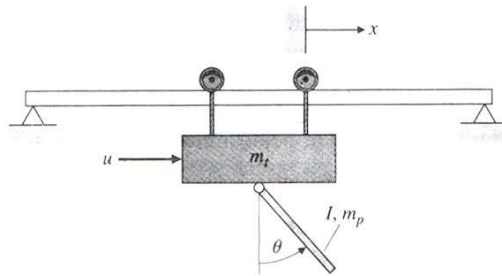
- $X(t) = e^{At} X_0 + \int_0^t e^{A(t-\tau)} B U(\tau) d\tau$
- $e^{At} = I + At + \frac{(At)^2}{2!} + \frac{(At)^3}{3!} + \dots$
- $X(s) = (sI - A)^{-1} B U(s)$
- Find out the eigen values and eigen vectors of  $sI - A$ , Obtain the transformation matrix and convert the state matrix into diagonal form
- Solve using a Discrete Time -Model





Can you model?

Schematic of the crane with  
hanging load

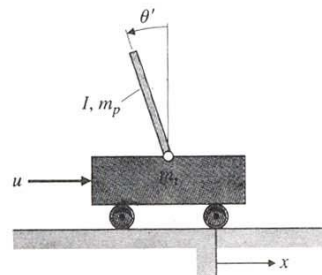


$$\Theta = 0$$

and

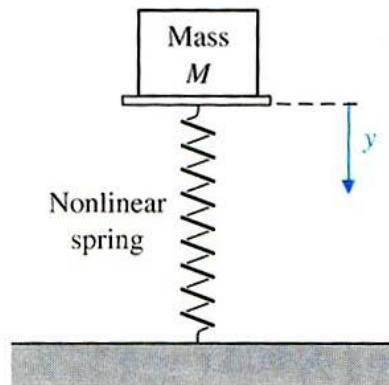
$$\Theta = \pi$$

Inverted pendulum

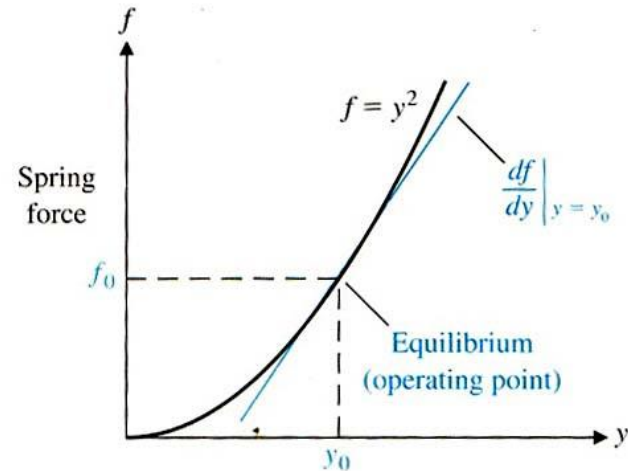


# Nonlinearity

(a) A mass sitting on a nonlinear spring. (b) The spring force versus  $y$ .



(a)



(b)

$$Y = g(x_0) + \left. \frac{dg}{dx} \right|_{x=x_0} (x - x_0)$$

When the principle of superposition and homogeneity gets violated

Linearizable for mechanical & electrical elements

Linearity: Zero at origin, law of addition and multiplication



Find the state-space and T.F. representations

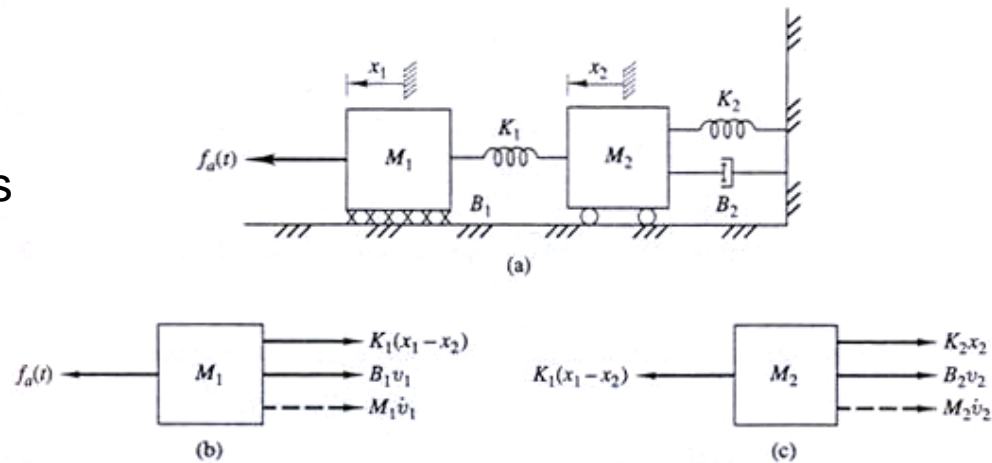
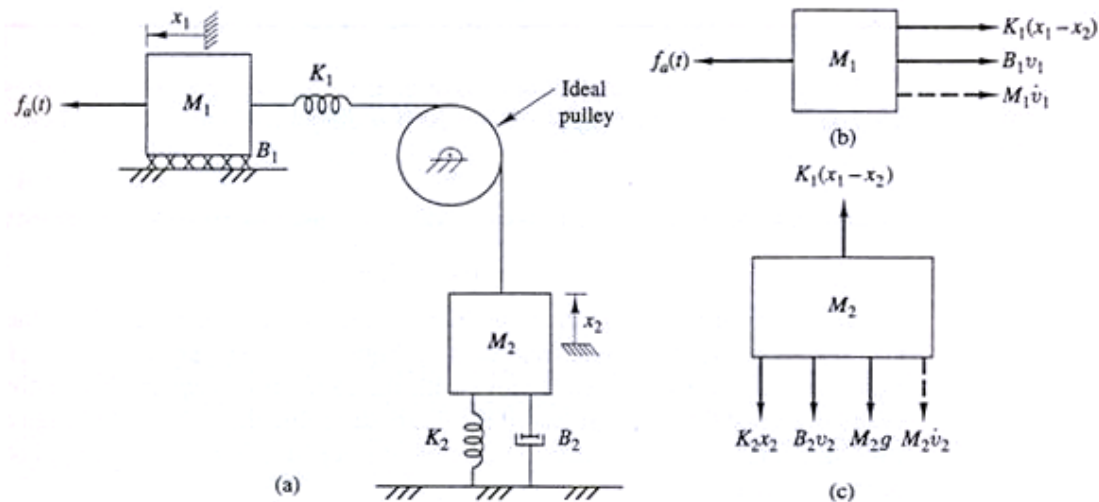
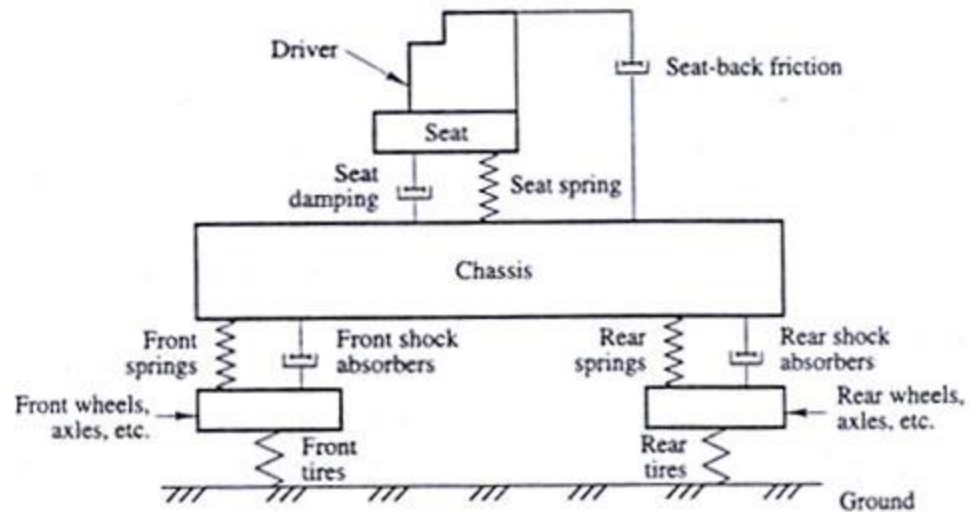
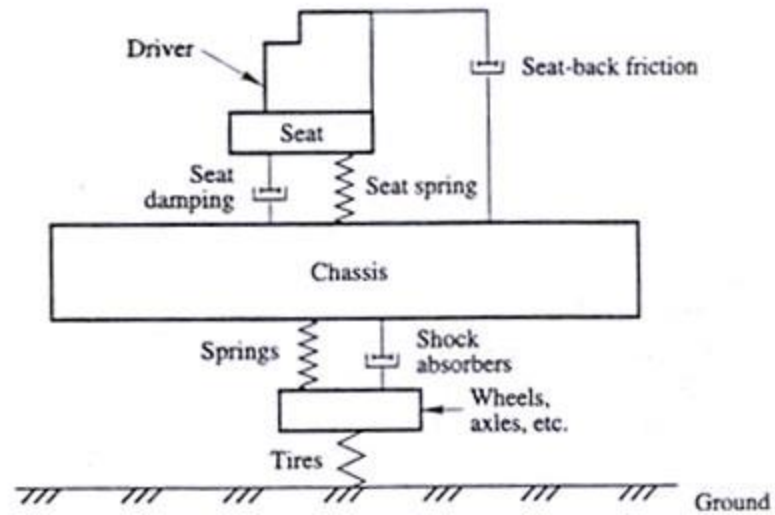


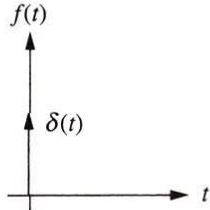
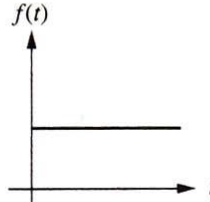
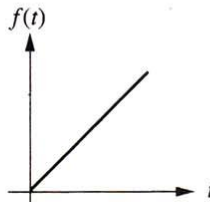
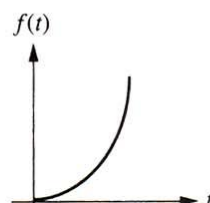
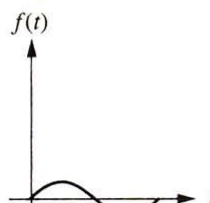
Figure 2.18 (a) Translational system for Example 2.7. (b), (c) Free-body diagrams.





(a)

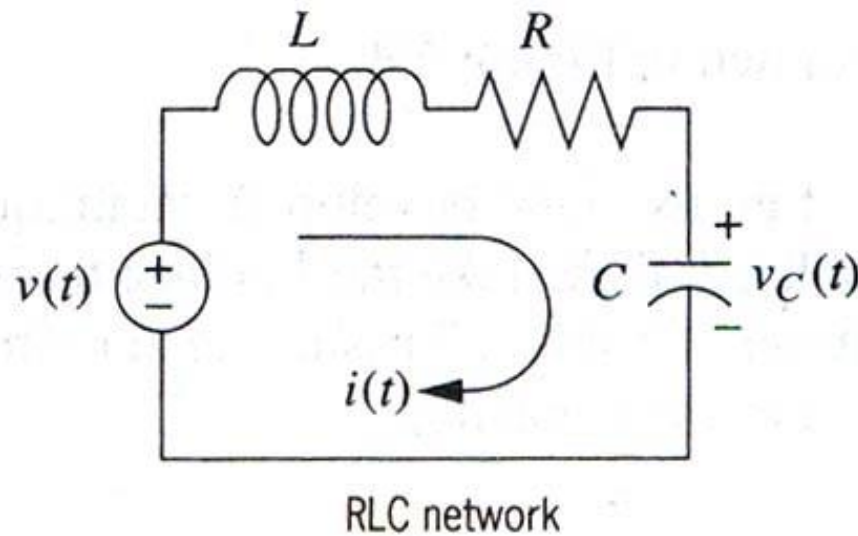


Input	Function	Description	Sketch	Use
Impulse	$\delta(t)$	$\delta(t) = \infty$ for $0- < t < 0+$ $= 0$ elsewhere $\int_{0-}^{0+} \delta(t) dt = 1$		Transient response Modeling
Step	$u(t)$	$u(t) = 1$ for $t > 0$ $= 0$ for $t < 0$		Transient response Steady-state error
Ramp	$tu(t)$	$tu(t) = t$ for $t \geq 0$ $= 0$ elsewhere		Steady-state error
Parabola	$\frac{1}{2}t^2u(t)$	$\frac{1}{2}t^2u(t) = \frac{1}{2}t^2$ for $t \geq 0$ $= 0$ elsewhere		Steady-state error
Sinusoid	$\sin \omega t$			Transient response Modeling Steady-state error

# Modelling of Electrical Elements




- **Kirchoff's Current Law:** The algebraic sum of currents leaving a junction or node equals the algebraic sum of currents entering the node.
- **Kirchoff's Voltage Law:** The algebraic sum of all voltages taken around a closed path in a circuit is zero.

# Modelling of Electrical System



# Impedance based representation

Voltage-current, voltage-charge, and impedance relationships for capacitors, resistors, and inductors

Component	Voltage-current	Current-voltage	Voltage-charge	Impedance $\hat{Z}(s) = V(s)/I(s)$	Admittance $\hat{Y}(s) = I(s)/V(s)$
 Capacitor	$v(t) = \frac{1}{C} \int_0^t i(\tau) d\tau$	$i(t) = C \frac{dv(t)}{dt}$	$v(t) = \frac{1}{C} q(t)$	$\frac{1}{Cs}$	$Cs$
 Resistor	$v(t) = Ri(t)$	$i(t) = \frac{1}{R} v(t)$	$v(t) = R \frac{dq(t)}{dt}$	$R$	$\frac{1}{R} = G$
 Inductor	$v(t) = L \frac{di(t)}{dt}$	$i(t) = \frac{1}{L} \int_0^t v(\tau) d\tau$	$v(t) = L \frac{d^2 q(t)}{dt^2}$	$Ls$	$\frac{1}{Ls}$

Note: The following set of symbols and units is used throughout this book:  $v(t) = V$  (volts),  $i(t) = A$  (amps),  $q(t) = Q$  (coulombs),  $C = F$  (farads),  $R = \Omega$  (ohms),  $G = \mathcal{U}$  (mhos),  $L = H$  (henries).



**THANK YOU**

