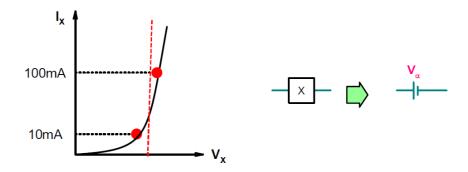
EE210: HW-2 Solution

Date: 17.01.19

Q.1 (a) A two terminal element X has I-V characteristics given by the relation $I_x = V_x^6$. A simple model for this device is a constant voltage of V_α . Determine this voltage, if the model is to be used in the current range 10mA-100mA. Determine the % error in the use of such a model over the specified current range. Will the error reduce if the non-linearity in the device became stronger (a higher order polynomial)?

Sol.



For 10mA current: $V_x = (10^{-2})^{1/6} = 0.464 \text{V}$

For 100mA current: $V_x = (10^{-1})^{1/6} = 0.681$ V

One estimate for V_{α} is the average of these two voltage values,

$$V_{\alpha} = \frac{0.464 + 0.681}{2} = 0.572V$$

Error at the two extreme ends of current range:

$$\varepsilon_1 = \frac{0.464 - 0.572}{0.464} \times 100 = -23.38\%$$

$$\varepsilon_2 = \frac{0.681 - 0.572}{0.681} \times 100 = 15.93\%$$
where order polynomial as slope will be store

Yes, error will reduce for higher order polynomial as slope will be steeper closely resembling to a constant voltage source.

(b) A battery of 5V is in series with a resistor and the element X. Determine an approximate value for the maximum % error in current, when the above model is used. (Note that error would increase as supply voltage gets closer to the voltage V_{α} .)

$$I = \frac{5 - V_x}{R} \Longrightarrow \Delta I = \frac{-\Delta V_x}{R} \Longrightarrow \frac{\Delta I}{I} = \frac{-\Delta V_x}{5 - V_x}$$

Maximum error:

$$\frac{\Delta I}{I} \simeq \frac{0.464 - 0.572}{5 - 0.572} = 2.4\%$$

(c) What would be a new suitable value of V_{α} , if the current range is changed to 1μ A- 10μ A? (Note that a model is a representation for a purpose).

For 1µA current:

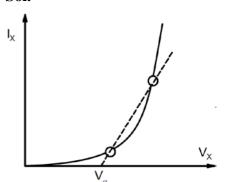
$$V_x = (10^{-6})^{1/6} = 0.1V$$

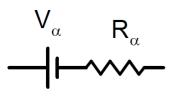
For $10\mu A$ current:

$$V_{\chi} = (10^{-5})^{1/6} = 0.147V$$

 $V_{\alpha} = \frac{0.1 + 0.147}{2} = 0.123V$

Q.2 An improved model for the element X is a constant voltage source V_{α} in series with a resistor R_{α} . Determine values of these elements for the current range 10mA-100mA. **Sol.**

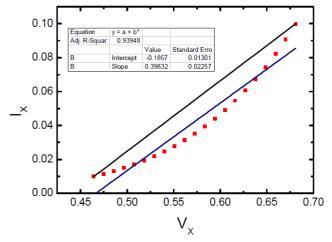


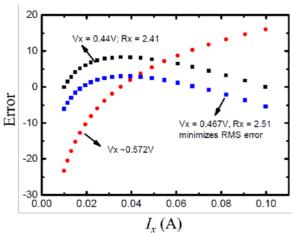


$$R_{\alpha} = \frac{V_{x2} - V_{x1}}{I_{x2} - I_{x1}} = \frac{0.681 - 0.464}{100mA - 10mA} = 2.41\Omega$$

$$V_{\alpha} = V_{x1} - I_{x1} * R_{\alpha} = 0.44V$$

This approach is accurate at the two extreme ends and error peaks somewhere in the middle. There are other ways of developing this model which may result in less overall error.





Minimum rms error:

$$I_x = \frac{(V - 0.467)}{2.51}$$

Q.3 (a) Determine the small signal model for element X, when it is biased at a voltage of 0.5V. Show that the small signal voltage across X has to be less than or equal to 21mV for error in small signal current to be less than or equal to 10%.

Sol.

Small signal model is a resistor of value:

$$\frac{1}{r_x} = \left(\frac{\partial I_x}{\partial V_x}\right)_{|V_x = 0.5V} = 6V_x^5_{|V_x = 0.5V} \Rightarrow r_x = 5.33\Omega$$

$$I_x + i_x = (V_x + v_x)^6 = V_x^6 * \left(1 + \frac{v_x}{V_x}\right)$$

$$i_x = I_x * \left\{ \left(1 + \frac{v_x}{V_x}\right)^6 - 1 \right\}$$

Small signal approximation:

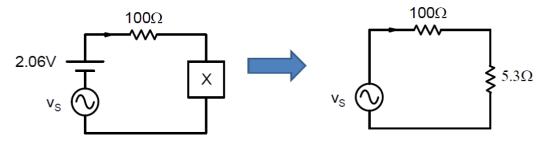
$$i'_x \cong I_x * \left\{ \left(1 + 6 \frac{v_x}{V_x} \right) - 1 \right\} = 6I_x \times \frac{v_x}{V_x}$$

Error:

$$\frac{i_{x} - i'_{x}}{i_{x}} = \frac{\left(1 + \frac{v_{x}}{V_{x}}\right)^{6} - 1 - \frac{6v_{x}}{V_{x}}}{\left(1 + \frac{v_{x}}{V_{x}}\right)^{6} - 1}$$

Substitution of $v_x = 21 \text{mV}$ and $V_x = 0.5 \text{V}$ gives error of 10%.

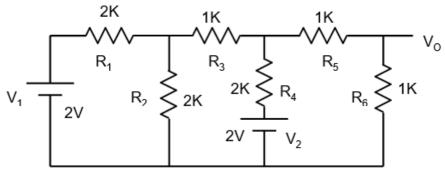
(b) How large can the ac voltage in the circuit be for the small signal model to be valid (with error < 10%) in the following circuit? (Note that X has a dc bias of 0.5V).



Sol.

$$v_x = v_s \times \frac{r_x}{100 + r_x} \le 21mV$$
$$v_s \le 0.415V$$

Q.4 For the circuit shown below, $V_o = 0.5$ V and $I_{R1} = 0.5$ mA for the given values of components. Using the small signal analysis technique, determine the approximate change in output voltage, if R_1 changes by 10%.



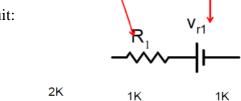
Sol.

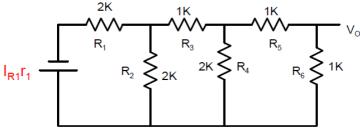
As a result of change in R1, current would change to $I_{R1}\!+\!i_{r1}.$

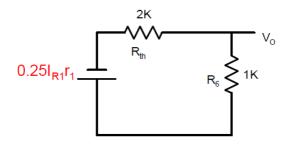
$$V_{R1} + v_{r1} = (I_R + i_{r1}) * (R_1 + r_1)$$

$$v_{r1} \cong i_{r1} * R_1 + I_R * r_1$$

Small signal equivalent circuit:

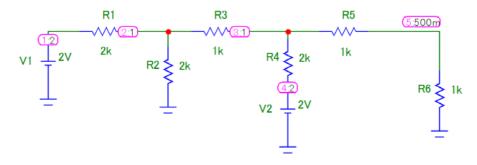


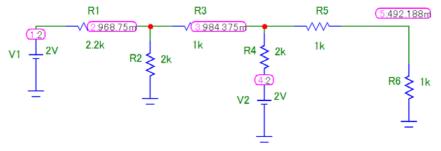




$$r_1 = 10\% \text{ of } 2K = 200\Omega$$

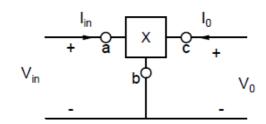
$$v_o = \frac{I_{r1}r_1}{12} = -\frac{0.5 \times 10^{-3} \times 200}{12} = -8.33 \text{mV}$$

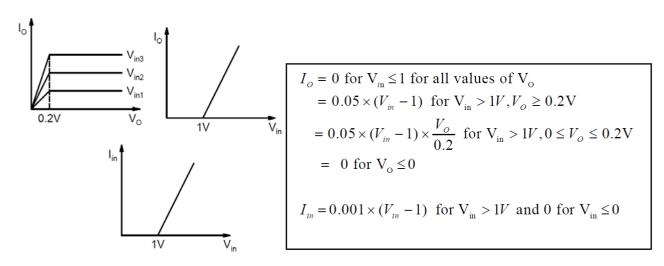




Difference in our hand calculation vs. simulation = $0.5 - 0.4921 = 7.9 \times 10^{-3}$

Q.5 A hypothetical 3-terminal (a, b & c) unilateral device (shown below) has the following characteristics.

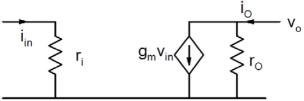




Determine the small signal model of the device around the bias point $V_{in} = 1.5 \text{V}$ and two values of output voltage $V_0 = 0.1 \text{V}$ and $V_0 = 2 \text{V}$.

Sol.

The general small signal model for a 3-terminal device is:



For
$$V_o = 2V$$
 and $V_{in} = 1.5V$,
$$\frac{1}{r_i} = \left(\frac{\partial I_{in}}{\partial V_{in}}\right)_{|V_{IN}=1.5V} = 10^{-3}\Omega^{-1} \Rightarrow r_i = 1K\Omega$$

$$\frac{1}{r_o} = \left(\frac{\partial I_o}{\partial V_o}\right)_{|V_o=1V} = 0 \Rightarrow r_o = \infty$$

$$g_m = \left(\frac{\partial I_o}{\partial V_{in}}\right)_{|V_{in}=1.5V, V_o=2V} = 0.05\Omega^{-1}$$

$$\downarrow i_{o} \qquad \downarrow i$$

For
$$V_o = 0.1$$
 V and $V_{in} = 1.5$ V
$$\frac{1}{r_i} = \left(\frac{\partial I_{in}}{\partial V_{in}}\right)_{|V_{IN} = 1.5V} = 10^{-3}\Omega^{-1} \Longrightarrow r_i = 1K\Omega$$

$$\frac{1}{r_o} = \left(\frac{\partial I_o}{\partial V_o}\right)_{|V_o = 0.1V} = 0.05 * (1.5 - 1) * \frac{1}{0.2} = 0.125\Omega^{-1}$$

$$g_m = \left(\frac{\partial I_o}{\partial V_{in}}\right)_{|V_{in} = 1.5V, V_o = 0.1V} = 0.05 * \frac{0.1}{0.2} = 0.025\Omega^{-1}$$

$$\downarrow i_{in} \qquad \downarrow i_{O} \qquad \downarrow V_o$$

$$1 \text{K}$$

$$0.025 \text{V}_{in} \qquad \downarrow \delta_{O} \qquad \downarrow \delta_{O$$