

EE23BTECH11020 - Raghava Ganji

Audio Filtering

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Abstract—This manual provides a simple introduction to digital signal processing.

1 SOFTWARE INSTALLATION

Run the following commands

```
sudo apt-get update
sudo apt-get install libffi-dev libsndfile1 python3
  -scipy python3-numpy python3-matplotlib
sudo pip install cffi pysoundfile
```

2 DIGITAL FILTER

2.1 Download the sound file from

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/RAF.wav>

2.2 You will find a Spectrogram at <https://academo.org/demos/spectrum-analyzer>.

Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the Spectrogram. What do you find?

Solution: There are a lot of yellow lines in between 440 to 10KHz . These represent the synthesizer key tones. Also, the key strokes are audible along with background noise. It

clearly shows that tonal frequency is under 10KHz. And above 10KHz only noise is present.

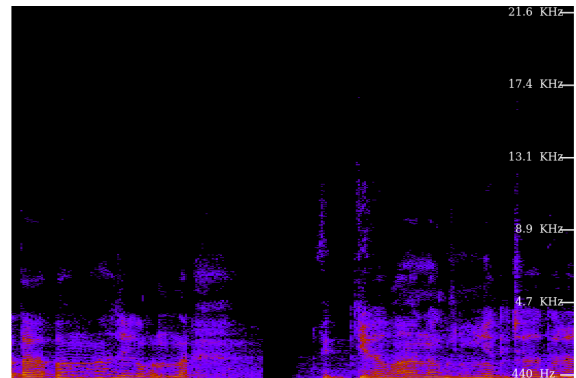


Fig. 1. Spectrogram of original sound

Write the python code for removal of out of band noise and execute the code.

Solution:

```
import soundfile as sf
from scipy import signal

input_signal, fs = sf.read('keyboard.wav')

sampler_freq = fs

order = 4

cutoff_freq = 10000.0

Wn = 2 * cutoff_freq / sampler_freq

b, a = signal.butter(order, Wn, 'low')

output_signal = signal.lfilter(b, a,
                               input_signal)

sf.write('Sound_With_ReducedNoise.wav',
        output_signal, fs)
```

2.3 The output of the python script in Problem ?? is the audio file Sound_With_ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 10KHz.

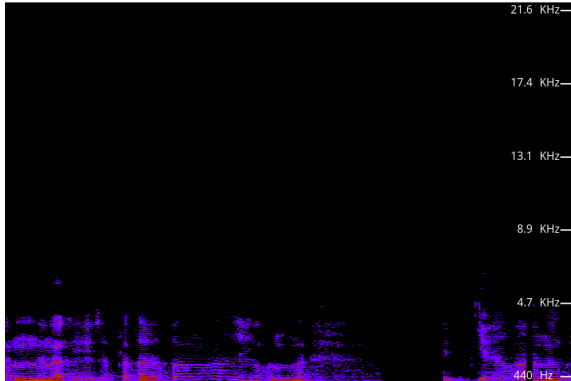


Fig. 2. Spectrogram with reduced noise

3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ \underset{\uparrow}{1}, 2, 3, 4, 2, 1 \right\} \quad (1)$$

Sketch $x(n)$.

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (2)$$

Sketch $y(n)$.

Solve

Solution: The C code calculates $y(n)$ and generates values in a text file.

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/3.2c>

The following code plots (1) and (2)

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/3.2py>

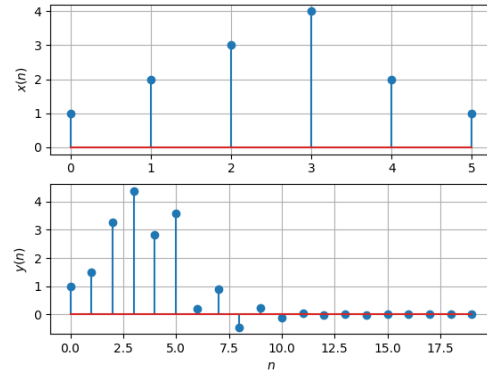


Fig. 3. plot of $x(n)$ and $y(n)$

4 Z-TRANSFORM

4.1 The Z-transform of $x(n]$ is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n} \quad (3)$$

Show that

$$\mathcal{Z}\{x(n-1)\} = z^{-1}X(z) \quad (4)$$

and find

$$\mathcal{Z}\{x(n-k)\} \quad (5)$$

Solution: From (3),

$$\begin{aligned} \mathcal{Z}\{x(n-k)\} &= \sum_{n=-\infty}^{\infty} x(n-k)z^{-n} \\ &= \sum_{n=-\infty}^{\infty} x(n)z^{-(n+k)} = z^{-k} \sum_{n=-\infty}^{\infty} x(n)z^{-n} \end{aligned} \quad (6)$$

$$= z^{-k} X(z) \quad (7)$$

resulting in (4). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \quad (8)$$

4.2 Find

$$H(z) = \frac{Y(z)}{X(z)} \quad (9)$$

from (2) assuming that the Z-transform is a linear operation.

Solution: Applying (8) in (2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z) \quad (10)$$

$$\Rightarrow \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \quad (11)$$

4.3 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases} \quad (12)$$

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \geq 0 \\ 0 & \text{otherwise} \end{cases} \quad (13)$$

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \quad (14)$$

Solution: It is easy to show that

$$\delta(n) \xleftrightarrow{Z} 1 \quad (15)$$

and from (13),

$$U(z) = \sum_{n=0}^{\infty} z^{-n} \quad (16)$$

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \quad (17)$$

using the formula for the sum of an infinite geometric progression.

4.4 Show that

$$a^n u(n) \xleftrightarrow{Z} \frac{1}{1 - az^{-1}} \quad |z| > |a| \quad (18)$$

Solution:

$$a^n u(n) \xleftrightarrow{Z} \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n} \quad (19)$$

$$= \sum_{n=0}^{\infty} a^n z^{-n} \quad (20)$$

$$= 1 + az^{-1} + a^2 z^{-2} + \dots \quad (21)$$

$$= \frac{1}{1 - az^{-1}} \quad |z| > |a| \quad (22)$$

4.5 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}). \quad (23)$$

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the *Discrete Time Fourier Transform* (DTFT) of $x(n)$.

Solution: The following code plots the magnitude of transfer function.

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/4.5.py>

Substituting $z = e^{j\omega}$ in (11), we get

$$|H(e^{j\omega})| = \left| \frac{1 + e^{-2j\omega}}{1 + \frac{1}{2}e^{-j\omega}} \right| \quad (24)$$

$$= \sqrt{\frac{(1 + \cos 2\omega)^2 + (\sin 2\omega)^2}{\left(1 + \frac{1}{2}\cos \omega\right)^2 + \left(\frac{1}{2}\sin \omega\right)^2}} \quad (25)$$

$$= \frac{4|\cos \omega|}{\sqrt{5 + 4\cos \omega}} \quad (26)$$

$$|H(e^{j(\omega+2\pi)})| = \frac{4|\cos(\omega + 2\pi)|}{\sqrt{5 + 4\cos(\omega + 2\pi)}} \quad (27)$$

$$= \frac{4|\cos \omega|}{\sqrt{5 + 4\cos \omega}} \quad (28)$$

$$= |H(e^{j\omega})| \quad (29)$$

Therefore its fundamental period is 2π , which verifies that DTFT of a signal is always periodic.

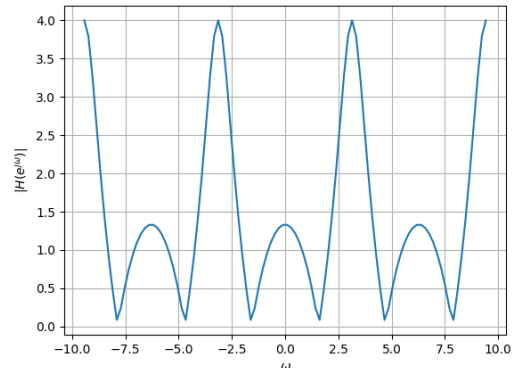


Fig. 4. $|H(e^{j\omega})|$

5 IMPULSE RESPONSE

5.1 Find an expression for $h(n)$ using $H(z)$, given that

$$h(n) \xleftrightarrow{Z} H(z) \quad (30)$$

and there is a one to one relationship between $h(n)$ and $H(z)$. $h(n)$ is known as the *impulse*

response of the system defined by (2).

Solution: From (11),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}} \quad (31)$$

$$\Rightarrow h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (32)$$

using (18) and (8).

5.2 Sketch $h(n)$. Is it bounded? Convergent?

Solution: The following code plots $h(n)$

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/5.2.py>

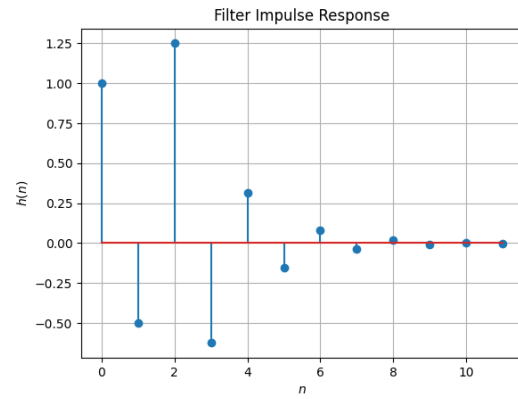


Fig. 5. $h(n)$ as the inverse of $H(z)$

5.3 The system with $h(n)$ is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \quad (33)$$

Is the system defined by (2) stable for the impulse response in (30)?

Solution: For stable system (33) should converge.

By using ratio test for convergence:

$$\lim_{n \rightarrow \infty} \left| \frac{h(n+1)}{h(n)} \right| < 1 \quad (34)$$

$$(35)$$

For large n

$$u(n) = u(n-2) = 1 \quad (36)$$

$$\lim_{n \rightarrow \infty} \left(\frac{h(n+1)}{h(n)} \right) = 1/2 < 1 \quad (37)$$

Hence it is stable.

5.4 Compute and sketch $h(n)$ using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (38)$$

This is the definition of $h(n)$.

Solution:

Definition of $h(n)$: The output of the system when $\delta(n)$ is given as input.

The following code plots Fig. 6. Note that this is the same as Fig. 5.

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/5.4.py>

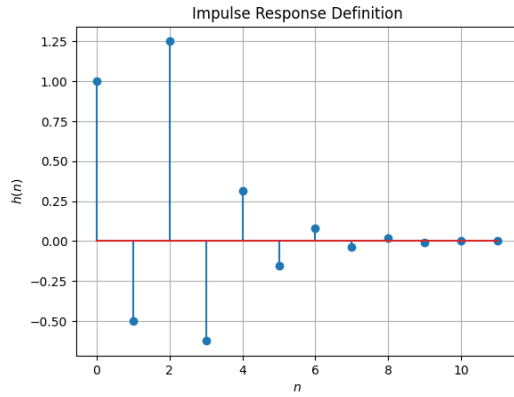


Fig. 6. $h(n)$ from the definition is same as Fig. 5

5.5 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (39)$$

Comment. The operation in (39) is known as *convolution*.

Solution: The following code plots Fig. 7. Note that this is the same as $y(n)$ in Fig. 3.

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/5.5.py>

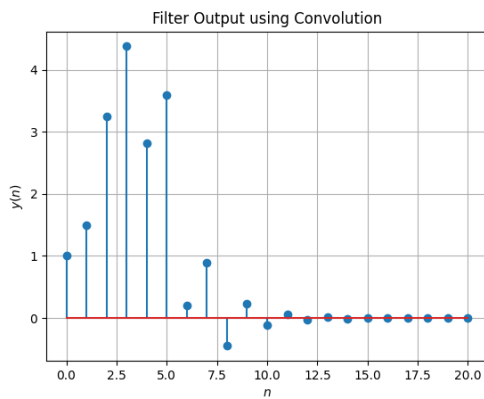


Fig. 7. $y(n)$ from the definition of convolution

5.6 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k) \quad (40)$$

Solution: In (39), we substitute $k = n - k$ to get

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (41)$$

$$= \sum_{n-k=-\infty}^{\infty} x(n-k)h(k) \quad (42)$$

$$= \sum_{k=-\infty}^{\infty} x(n-k)h(k) \quad (43)$$

6 DFT AND FFT

6.1 Compute

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (44)$$

and $H(k)$ using $h(n)$.

6.2 Compute

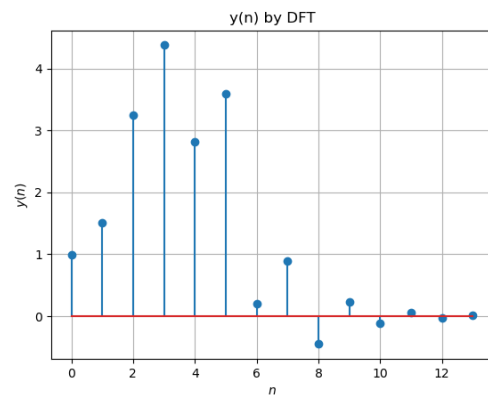
$$Y(k) = X(k)H(k) \quad (45)$$

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1 \quad (46)$$

Solution: The above three questions are solved using the code below.

https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/6_sol.py



6.4 Repeat the previous exercise by computing $X(k)$, $H(k)$ and $y(n)$ through FFT and IFFT.

Solution: The solution of this question can be found in the code below.

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/6.4.py>

This code verifies the result by plotting the obtained result with the result obtained by DFT.

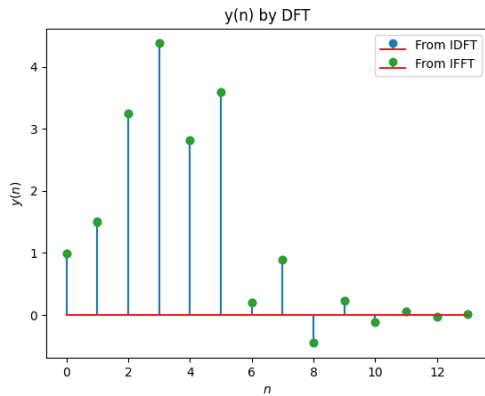


Fig. 8. $y(n)$ obtained from IDFT and IFFT is plotted and verified

6.5 Wherever possible, express all the above equations as matrix equations.

Solution: The DFT matrix is defined as :

$$\mathbf{W} = \begin{pmatrix} \omega^0 & \omega^0 & \dots & \omega^0 \\ \omega^0 & \omega^1 & \dots & \omega^{N-1} \\ \vdots & \vdots & \ddots & \vdots \\ \omega^0 & \omega^{N-1} & \dots & \omega^{(N-1)(N-1)} \end{pmatrix} \quad (47)$$

where $\omega = e^{-\frac{j2\pi}{N}}$. Now any DFT equation can be written as

$$\mathbf{X} = \mathbf{W}\mathbf{x} \quad (48)$$

where

$$\mathbf{x} = \begin{pmatrix} x(0) \\ x(1) \\ \vdots \\ x(n-1) \end{pmatrix} \quad (49)$$

$$\mathbf{X} = \begin{pmatrix} X(0) \\ X(1) \\ \vdots \\ X(n-1) \end{pmatrix} \quad (50)$$

Thus we can rewrite (45) as:

$$\mathbf{Y} = \mathbf{X} \cdot \mathbf{H} = (\mathbf{W}\mathbf{x}) \cdot (\mathbf{W}\mathbf{h}) \quad (51)$$

The below code computes $y(n)$ by DFT Matrix and then plots it.

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/6.5.py>

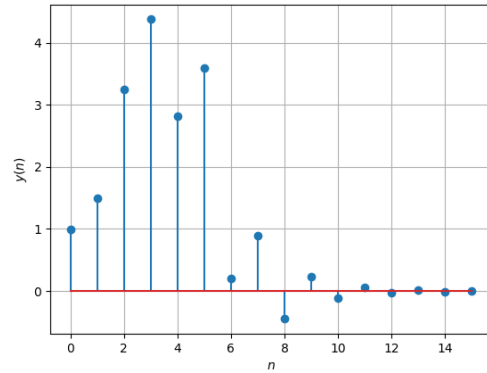


Fig. 9. $y(n)$ obtained from DFT Matrix

7 EXERCISES

Answer the following questions by looking at the python code in Problem ??.

7.1 The command

```
output_signal = signal.lfilter(b, a,
input_signal)
```

in Problem ?? is executed through the following difference equation

$$\sum_{m=0}^M a(m) y(n-m) = \sum_{k=0}^N b(k) x(n-k) \quad (52)$$

where the input signal is $x(n)$ and the output signal is $y(n)$ with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.

Solution: The below code gives the output of an Audio Filter with and without using the built in function `signal.lfilter`.

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/7.1.py>

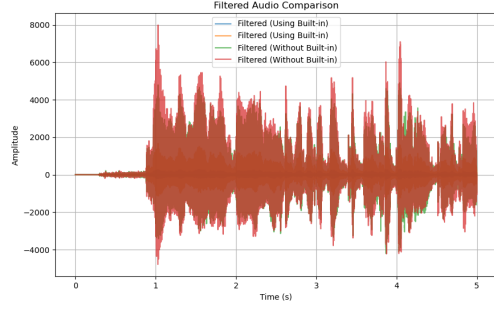


Fig. 10. Both the outputs using and without using function overlap

7.2 Repeat all the exercises in the previous sections for the above a and b .

Solution: The code in ?? generates the values of a and b which can be used to generate a difference equation.

And,

$$M = 5 \quad (53)$$

$$N = 5 \quad (54)$$

From 52

$$a(0)y(n) + a(1)y(n-1) + a(2)y(n-2) + a(3)y(n-3) + a(4)y(n-4) = b(0)x(n) + b(1)x(n-1) + b(2)x(n-2) + b(3)x(n-3) + b(4)x(n-4) \quad (55)$$

$$y(n-3) + a(4)y(n-4) = b(0)x(n) + b(1)x(n-1) + b(2)x(n-2) + b(3)x(n-3) + b(4)x(n-4)$$

Difference Equation is given by :

$$\begin{aligned} & y(n) - (3.66)y(n-1) + (5.05)y(n-2) \\ & - (3.099)y(n-3) + (0.715)y(n-4) \\ & = (1.45 \times 10^{-5})x(n) + (5.74 \times 10^{-5})x(n-1) \\ & + (8.62 \times 10^{-5})x(n-2) + (5.74 \times 10^{-5})x(n-3) \\ & + (1.43 \times 10^{-5})x(n-4) \end{aligned} \quad (56)$$

From (52)

$$H(z) = \frac{b_0 + b_1z^{-1} + b_2z^{-2} + \dots + b_Mz^{-M}}{a_0 + a_1z^{-1} + a_2z^{-2} + \dots + a_Nz^{-N}} \quad (57)$$

$$H(z) = \frac{\sum_{k=0}^N b(k)z^{-k}}{\sum_{k=0}^M a(k)z^{-k}} \quad (58)$$

Partial fraction on (58) can be generalised as:

$$H(z) = \sum_i \frac{r(i)}{1 - p(i)z^{-1}} + \sum_j k(j)z^{-j} \quad (59)$$

Now,

$$a^n u(n) \xleftrightarrow{Z} \frac{1}{1 - az^{-1}} \quad (60)$$

$$\delta(n-k) \xleftrightarrow{Z} z^{-k} \quad (61)$$

Taking inverse z transform of (59) by using (60) and (61)

$$h(n) = \sum_i r(i)[p(i)]^n u(n) + \sum_j k(j)\delta(n-j) \quad (62)$$

The below code computes the values of $r(i)$, $p(i)$, $k(i)$ and plots $h(n)$

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/7.2.py>

$r(i)$	$p(i)$	$k(i)$
$-0.58129538 - 2.51766121j$	$0.13676769 + 0.1953307j$	2.0248078
$-0.58129538 + 2.51766121j$	$0.13676769 - 0.1953307j$	$-$
$-0.40462158 + 0.3262658j$	$0.18896814 + 0.6515556j$	$-$
$-0.40462158 - 0.3262658j$	$0.18896814 - 0.6515556j$	$-$

TABLE I
VALUES OF $r(i)$, $p(i)$, $k(i)$

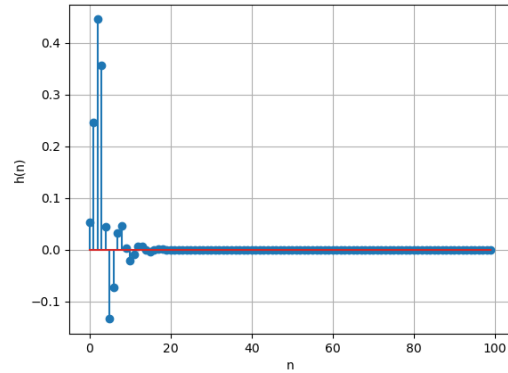


Fig. 11. $h(n)$ of Audio Filter

Stability of $h(n)$:

According to (33)

$$H(z) = \sum_{n=0}^{\infty} h(n)z^{-n} \quad (63)$$

$$H(1) = \sum_{n=0}^{\infty} h(n) = \frac{\sum_{k=0}^N b(k)}{\sum_{k=0}^M a(k)} < \infty \quad (64)$$

As both $a(k)$ and $b(k)$ are finite length sequences they converge.

The below code plots Filter frequency response

<https://github.com/Raghava11020/Signals-and-Systems/blob/main/4/codes/filterresponse.py>

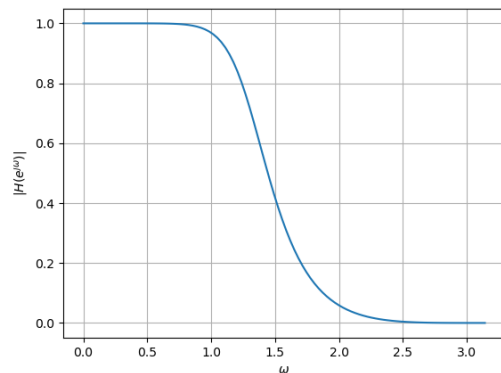


Fig. 12. Frequency Response of Audio Filter

7.3 What is the sampling frequency of the input signal?

Solution: The Sampling Frequency is 44.1KHz

7.4 What is type, order and cutoff-frequency of the above butterworth filter

Solution: The given butterworth filter is lowpass with order=4 and cutoff-frequency=1KHz.

7.5 Modify the code with different input parameters and get the best possible output. **Solution:**

A better filtering was found on setting the order of the filter to be 5.