

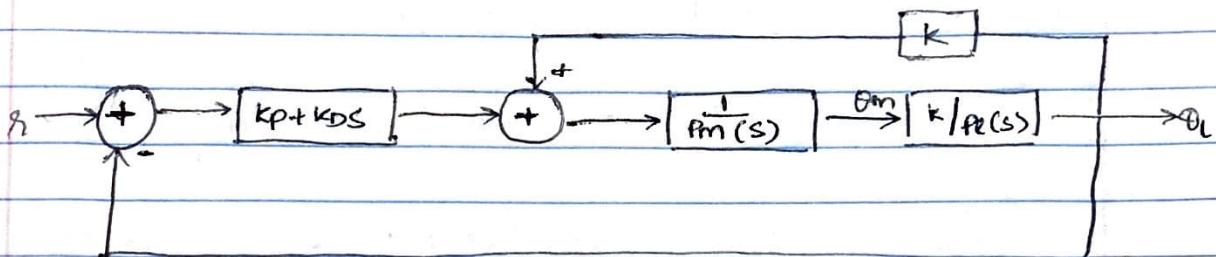
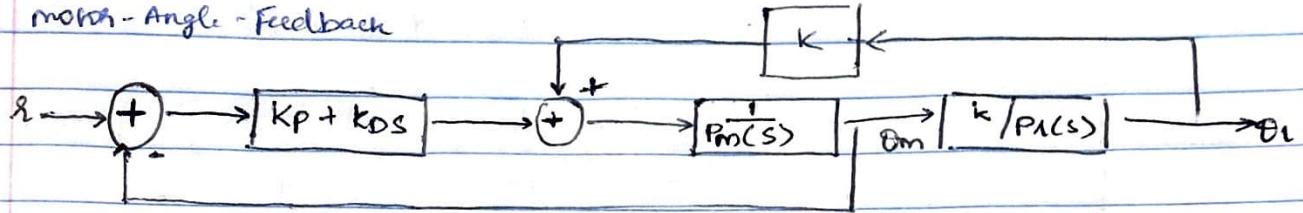
The eqⁿ of motion of the motor side is

$$J_m \ddot{\theta}_m + B_m \dot{\theta}_m - k(\theta_t - \theta_m) = u$$

The eqⁿ of motion in the load side is

$$J_L \ddot{\theta}_L + B_L \dot{\theta}_L + k(\theta_t - \theta_m) = 0$$

motor - Angle - Feedback



Load - angle - Feedback

(a) Motor Angle - Feedback $\Rightarrow u = k_p(\theta_{cmd} - \theta_m) - k_d \dot{\theta}_m$
 $\quad \quad \quad u = k_p(\theta_{cmd} - \theta_l) - k_d \dot{\theta}_l$

$$P_m(s) = J_m s^2 + B_m s + k ; \quad P_1(s) = J_L s^2 + B_L s + k$$

(b) Load Angle - Feed back

$$u = k_p(\theta_{cmd} - \theta_l) - k_d \dot{\theta}_l$$

The eqⁿ of motions are :-

$$Jm\ddot{\theta}_m + Bm\dot{\theta}_m - k(\theta_l - \theta_m) = u \quad \dots \quad (1)$$

$$J_1 \ddot{\theta}_1 + B_1 \dot{\theta}_1 + k(\theta_1 - \theta_m) = 0 \quad \dots \quad (2)$$

$$x_1 = \theta_m ; \quad x_2 = \dot{\theta}_m ; \quad x_3 = \theta_e ; \quad x_4 = \dot{\theta}_e \quad \dots \quad (3)$$

$$\dot{x}_1 = \theta_m = x_2 ; \quad \dot{x}_2 = \ddot{\theta}_m ; \quad \dot{x}_3 = \theta_L = x_4 ; \quad \dot{x}_4 = \ddot{\theta}_L \quad \dots \quad (4)$$

Sub" ③ + ④ in ① & ②

$$Jm x_2 + Bm x_2 - k(x_3 - x_1) = u \quad \dots \quad (5)$$

$$Jl \dot{x}_4 + Bl x_4 + K(x_3 - x_1) = 0 \quad \dots \quad (6)$$

Case - I (Motor Angle Feed back)

$$u = kp(\theta_{cmd} - \theta_m) - kd\dot{\theta}_m$$

$$u = k_p (\theta_{cmd} - x_1) - k_d (x_2) \quad \dots \quad (7)$$

Sub (7) in (5) we get.

$$3m\ddot{x}_2 + Bm\dot{x}_2 - k(x_3 - x_1) = kp(\theta_{cmd} - \dot{x}_1) - kd(x_2)$$

$$x_2 = kp(\text{Ocmd} - x_1) - kd(x_2) - Bm x_2 + k(x_3 - x_1)$$

Rearranging the terms we get

$$\dot{x}_2 = \underbrace{K_p \theta_{md} - (K_P + k)x_1 - (kd + Bm)x_2}_{Jm} + kx_3 \quad \text{--- (8)}$$

from (6) we get is

$$\dot{x}_4 = -\underline{B} \underline{l} x_4 - k (\underline{x}_3 - \underline{x}_1) \quad \dots \quad (9)$$

$$\begin{array}{|c|c|} \hline \dot{x}_1 & x_2 \\ \dot{x}_2 & = \frac{K_P \theta_{cmd} - (K_P + K) x_1 - (K_d + B_m) x_2 + K x_3}{J_m} \\ \dot{x}_3 & x_4 \\ \dot{x}_4 & -B_m x_4 - K(x_3 - x_1) \\ \hline \end{array}$$

$$A = \frac{\partial F}{\partial x} = \begin{bmatrix} \frac{\partial F_1}{\partial x_1} & \frac{\partial F_1}{\partial x_2} & \frac{\partial F_1}{\partial x_3} & \frac{\partial F_1}{\partial x_4} \\ \frac{\partial F_2}{\partial x_1} & \frac{\partial F_2}{\partial x_2} & \frac{\partial F_2}{\partial x_3} & \frac{\partial F_2}{\partial x_4} \\ \frac{\partial F_3}{\partial x_1} & \frac{\partial F_3}{\partial x_2} & \frac{\partial F_3}{\partial x_3} & \frac{\partial F_3}{\partial x_4} \\ \frac{\partial F_4}{\partial x_1} & \frac{\partial F_4}{\partial x_2} & \frac{\partial F_4}{\partial x_3} & \frac{\partial F_4}{\partial x_4} \end{bmatrix}$$

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{(K_P + K)}{J_m} & \frac{(-K_d + B_m)}{J_m} & \frac{K}{J_m} & 0 \\ 0 & 0 & 0 & 1 \\ \frac{K}{J_L} & 0 & -\frac{K}{J_L} & \frac{-B_m}{J_L} \end{bmatrix} \quad \text{--- (10)}$$

$$B = \frac{\partial F}{\partial \theta_{cmd}} = \begin{bmatrix} \frac{\partial F_1}{\partial \theta_{cmd}} \\ \frac{\partial F_2}{\partial \theta_{cmd}} \\ \frac{\partial F_3}{\partial \theta_{cmd}} \\ \frac{\partial F_4}{\partial \theta_{cmd}} \end{bmatrix} = \begin{bmatrix} 0 \\ \frac{K_P}{J_m} \\ 0 \\ 0 \end{bmatrix} \quad \text{--- (11)}$$

So for Case - I, The State Space Representation is given as:-
Sub $\begin{pmatrix} 10 \\ 11 \end{pmatrix}$

$$\begin{array}{c|c|c|c|c|c|c} \dot{x}_1 & 0 & 1 & 0 & 0 & x_1 & 0 \\ \dot{x}_2 & = & -\frac{(kp+k)}{Jm} & -\frac{(Bn+kd)}{Jm} & \frac{k}{Jm} & 0 & + & \frac{kp}{Jm} & \theta_{cmd} \\ \dot{x}_3 & & & & & x_2 & & 0 & \\ \dot{x}_4 & & & & & x_3 & & 0 & \\ \hline & 0 & 0 & 0 & 1 & x_4 & & 0 & \\ & \frac{k}{J_L} & 0 & -\frac{B_L}{J_L} & -\frac{B_L}{J_L} & & & 0 & \\ \hline \end{array}$$

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Case 2 :- load angle feed back.

$$u = k_p(\theta_{md} - \theta_e) - k_d \dot{\theta}_e$$

SWB (3) ue geb.

$$h = kp(\theta_{cmd} - x_3) - kd x_4 \quad \dots \quad (13)$$

Sub (13) in (5) we get

$$J_m x_2 + B_m x_2 - k(x_3 - x_1) = k_p (\Theta_{cmd} - x_3) - k_d x_4$$

$$\dot{x}_2 = \underbrace{k_p(\theta_{cmd} - x_3)}_{\text{Jm}} - kd x_4 - Bm x_2 + K(x_3 - x_1)$$

Re arranging terms we get.

$$\dot{x}_2 = k_p \theta_{cmd} + (k_r - k_p) x_3 - k_d x_4 - B_m x_2 - k_n x_1 \quad \text{--- (14)}$$

$$\dot{x}_4 = -Bl \underline{x_4} - k(x_5 - x_4)$$

$$\begin{array}{|c|c|c|c|} \hline
 & x_1 & & x_2 \\ \hline
 x_2 & = & \frac{k_p \theta_{cmd} + (k_p + k_d)x_3 - k_d x_4 - B_m x_2 - k x_1}{Jm} \\ \hline
 & x_3 & & x_4 \\ \hline
 & & -\frac{B_l x_4 - k(x_3 - x_1)}{JL} & \\ \hline
 \end{array}$$

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{k}{Jm} & -\frac{B_m}{Jm} & \frac{k - k_p}{Jm} & -\frac{k_d}{Jm} \\ 0 & 0 & 0 & 1 \\ \frac{k}{JL} & +\textcircled{B_L} & -\frac{k}{JL} & -\frac{B_L}{JL} \end{bmatrix}$$

$$B = \begin{bmatrix} 0 \\ \frac{k_p}{Jm} \\ 0 \\ 0 \end{bmatrix}$$

So the State-Space eqⁿ for Case-2 is :-

$$\begin{array}{|c|c|c|c|c|c|c|} \hline & x_1 & 0 & 1 & 0 & 0 & x_1 & 0 \\ \hline x_2 & = & -\frac{k}{2m} & -\frac{Bm}{2m} & \frac{k+kp}{2m} & -\frac{kq}{2m} & x_2 & + \frac{kp}{2m} \text{ cmd} \\ \hline x_3 & & 0 & 0 & 0 & 1 & x_3 & 0 \\ \hline x_4 & & \frac{k}{2L} & 0 & -\frac{k}{2L} & -\frac{Bk}{2L} & x_4 & 0 \\ \hline \end{array}$$

Act Bus

Problem -2

Case - 1. $u = kp(\theta_{cmd} - \theta_m) + kd\dot{\theta}_m$ (motor Angle Feedback)

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{(kp+k)}{Jm} & -\frac{(kd+Bm)}{Jm} & \frac{k}{Jm} & 0 \\ 0 & 0 & 0 & 1 \\ \frac{k}{JL} & 0 & -\frac{k}{JL} & -\frac{Bd}{JL} \end{bmatrix}$$

The characteristic polynomial is given as :-

$$\text{Inverses} \quad |S\bar{I} - A| = 0$$

$$|S\bar{I} - A| = \begin{vmatrix} + & - & + & - \\ S & -1 & 0 & 0 \\ \frac{k_p+k}{Jm} & \frac{S+(kd+Bl)}{Jm} & \frac{-k}{Jm} & 0 \\ 0 & 0 & S & -1 \\ \frac{-k}{JL} & 0 & \frac{k}{JL} & \frac{S+Bl}{JL} \end{vmatrix}$$

$$(S\bar{I} - A) = S \begin{vmatrix} + & - & + & - \\ S+\frac{(kd+Bl)}{Jm} & -\frac{k}{Jm} & 0 & 0 \\ 0 & S & -1 & 0 \\ 0 & \frac{k}{JL} & \frac{S+Bl}{JL} & \end{vmatrix}$$

$$(-1) \begin{vmatrix} \frac{k_p+k}{Jm} & -\frac{k}{Jm} & 0 & 0 \\ 0 & S & -1 & 0 \\ \frac{-k}{JL} & \frac{k}{JL} & \frac{S+Bl}{JL} & 0 \end{vmatrix} + 0 = 0$$

$$= S \left[\left(S + \frac{(kd+Bl)}{Jm} \right) \left(S + \frac{Bl}{JL} \right) - (-1) \left(\frac{k}{JL} \right) \right]$$

$$+ 1 \left[\frac{k_p+k}{Jm} \left((S) \left(S + \frac{Bl}{JL} \right) + 1 \left(\frac{k}{JL} \right) \right) \right]$$

$$+ \left[\frac{k}{Jm} \left(-\frac{k}{JL} \right) \right] = 0$$

$$S \left[\left(S + \frac{kd + Bm}{Jm} \right) f(S) \left(S + \frac{BL}{JL} \right) + \frac{k}{JL} \right]$$

$$+ \frac{k_p + k}{Jm} \left(S \left(S + \frac{BL}{JL} \right) + \frac{k}{JL} \right) + \frac{k}{Jm} \left(-\frac{k}{JL} \right) = 0$$

$$S \left(S' \right) + \left(\frac{kd + Bm}{Jm} \right) \left[S^2 + \frac{BLs}{JL} + \frac{k}{JL} \right]$$

$$+ \frac{k_p + k}{Jm} \left(S^2 + \frac{BLs}{JL} + \frac{k}{JL} \right) + \frac{k}{Jm} \left(-\frac{k}{JL} \right) = 0$$

$$S \left[S + \left(\frac{kd + Bm}{Jm} \right) S^2 + \left(\frac{BL}{JL} \right) \left(\frac{kd + Bm}{Jm} \right) S + \frac{k(kd + Bm)}{Jm JL} \right]$$

$$S \left[S^3 + \frac{BL}{JL} S^2 + \frac{k}{JL} S + \left(\frac{kd + Bm}{Jm} \right) S^2 + \frac{BL}{JL} \left(\frac{kd + Bm}{Jm} \right) S \right. \\ \left. + \frac{k}{JL} \left(\frac{kd + Bm}{Jm} \right) \right] = 0$$

$$+ \frac{k_p + k}{Jm} \left[S^2 + \frac{BLs}{JL} + \frac{k}{JL} \right] +$$

$$+ \frac{k_p + k}{Jm} S^2 + \left(\frac{BL}{JL} \right) \left(\frac{k_p + k}{Jm} \right) S + \frac{k}{JL} \left(\frac{k_p + k}{Jm} \right)$$

$$\frac{-k^2}{Jm JL} = 0$$

$$|S_I - A| = S^4 + \frac{BL}{JL} S^3 + \frac{k}{JL} S^2 + \left(\frac{kd + Bm}{Jm} \right) S^3 + \frac{BL}{JL} \left(\frac{kd + Bm}{Jm} \right) S^2$$

$$+ \frac{k}{JL} \left(\frac{kd + Bm}{Jm} \right) S + \left(\frac{k_p + k}{Jm} \right) S^2 + \left(\frac{BL}{JL} \right) \left(\frac{k_p + k}{Jm} \right) S$$

$$+ \left(\frac{k}{JL} \right) \left(\frac{k_p + k}{Jm} \right) - \frac{k^2}{Jm JL} = 0$$

Isolate the BIBns + BImS³ + Blks + Blkps + Bmks + Bmr

$$|SI - A| = BlBms^2 + B_1Jms^3 + Blkds^2 + Blkps + Bmks^3 + Bmks \\ + 2lJms^4 + JlkdS^3 + 3ekps^2 + Jmks^2 + kdkks + kpks = 0$$

JlJm

given. $Jm = Jl = 0.0097 \text{ kg.m}^2$, $bm = bl = 0.04169 \text{ Ns.m}^{-1}$
 $k = 100 \text{ Nm rad}^{-1}$. $kp = 10$

$$|SI - A| = s^4 + (103.093kd + 8.596)s^3 + (uu3.086kd + 21657.957)s^2 \\ + (1062812.201kd + 93048.145)s + 10628122.011 = 0$$

(a₀) (a₁) (a₂) (a₃) (a₄)

From this polynomial lets form the Hessian matrix.

$$H = \begin{vmatrix} a_1 & a_3 & 0 & 0 \\ a_0 & a_2 & a_4 & 0 \\ 0 & a_1 & a_3 & 0 \\ 0 & a_0 & a_2 & a_4 \end{vmatrix}$$

$$H = \begin{vmatrix} (103.093kd + 8.596) & (1062812.201kd) & 0 & 0 \\ 0 & (uu3.086kd + 21657.957) & 10628122.011 & 0 \\ 0 & (103.093kd + 8.596) & (1062812.201kd) & 0 \\ 0 & 0 & (uu3.086kd + 21657.957) & 10628122.011 \end{vmatrix}$$

$$\Delta_1 > 0 \Rightarrow a_1 > 0$$

$$\Rightarrow 103.093kd + 8.596 > 0$$

$$\Rightarrow kd > -0.083381 \quad \text{--- (1) Given } kd > 0 \}$$

$$\Delta_2 > 0 \Rightarrow a_1 a_2 - a_3 a_0 > 0$$

$$\Rightarrow a_1 a_2 > a_3 a_0$$

$$(103.093kd + 8.596)(443.086kd + 21657.957)$$

$$> (1062812.201kd + 93048.145)$$

$$43,679.065kd^2 + 2,232,783.8kd + 3,808.76kd \\ + 186,171.798 > 1062812.201kd + 93048.145$$

$$43,679.065kd^2 + 1,173,780.4kd + 93,123.653 > 0$$

The roots of the eqn are $kd > -0.0796$; $kd > -25.6167$

And given $kd > 0 \therefore \Delta_2 > 0$

$$\Delta_3 > 0$$

$$\sqrt{\Delta_4 = a_3}, \Delta_3 > 0 \Rightarrow a_4 > 0 \quad \{ \text{And } a_4 > 0 \}$$

$$\Delta_3 > 0$$

$$\Rightarrow a_1(a_2 a_3 - a_1 a_0) - a_3(a_0 a_3) > 0$$

$$a_1 a_2 a_3 - (a_1)^2 a_0 - (a_3)^2 a_0 > 0$$

$$= 4854826.2610.1464kd^3 + 1138800353467.1kd^2 + \\ 189354020875.417kd + 7879658342.69843 > 0$$

The roots of this eqⁿ are

$$kd > -23.289918, \quad kd > +0.088, \quad kd > -0.079$$

given $kd > 0$; $\therefore \Delta g > 0$.

The lowest value of kd at which the system can take is -0.079

$\therefore kd > -0.079$ for the system to be stable.

\Rightarrow The system is stable for $kd > -0.079$

\therefore Case 2:-

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{k}{J_m} & -\frac{B}{J_m} & \frac{k-k_p}{J_m} & \frac{-kd}{J_m} \\ 0 & 0 & 0 & 1 \\ \frac{k}{J_L} & 0 & -\frac{k}{J_L} & -\frac{B_L}{J_L} \end{bmatrix}$$

$$[SI - A] = \begin{bmatrix} s & -1 & 0 & 0 \\ \frac{+k}{J_m} & \frac{s+B}{J_m} & \frac{-(k-k_p)}{J_m} & \frac{+kd}{J_m} \\ 0 & 0 & s & -1 \\ -\frac{k}{J_L} & 0 & \frac{+k}{J_L} & \frac{s+B_L}{J_L} \end{bmatrix}$$

given $J_m = J_L = 0.0097$; $B_m = B_L = 0.04169$; $k = 100 \text{ Nm rad}^{-1}$

$$k_p = 10$$

Simplifying the determinant of $[SI - A]$ we get

$$B_1 B_m s^2 + B_1 J_m s^3 + B_m k s + B_m J_m s^3 + B_m k s + J_1 J_m s^4 + J_m k s^2 \\ + J_m k s^2 + k d k s + k p k$$

$J_1 J_m$

$$J_m = \omega = 0.0097 ; B_m = 0.04163 ; B_{-l} = 0.04163$$

$$k = 100 ; k_p = 10$$

$$(0.04163)(0.04163)s^2 + (0.04163)(0.0097)s^3 \\ + (0.04163)(100)s + (0.04163)(0.0097)s^3 + (0.04163) \\ (100)s + (0.0097)(0.0097)s^4 + (0.0097)(100)s^2 \\ + (0.0097)(100)s^2 + (kd)(100)s + (10)(100)$$

$(0.0097)(0.0097)$

$$= 0$$

$$\Rightarrow s^4 + 8.5958762886598s^3 + 20637.0289733234s^2 + \\ (10628122.20108402 + kd + 88617.2813263896)s + \\ 10628122.0108402.$$

α_4

The fourth harmonic makes it is

$$H = \begin{bmatrix} 8.596 & 10628122.20108402 + 88617.281 & 0 & 0 \\ 1 & 20637.0289 & 10628122.201084 & 0 \\ 0 & 8.596 & (10628122.20108402 + 88617.281) & 0 \\ 0 & 1 & 20637.0289 & 10628122.201084 \end{bmatrix}$$

$$\Delta_1 > 0$$

$$8 \cdot 596 > 0$$

$$\Delta_2 > 0$$

$$(8 \cdot 596)(20637 \cdot 029) - (1062812 \cdot 201kd + 88617 \cdot 281) > 0$$

$$= 0 \cdot 0835318 > kd$$

$$\Rightarrow kd < 0.0835318 \quad \dots \dots \dots \textcircled{1}$$

$$\Delta_3 > 0$$

$$\begin{vmatrix} 8 \cdot 596 & 1062812 \cdot 201kd + 88617 \cdot 281 & 0 \\ 1 & 20637 \cdot 029 & 10628122 \\ 0 & 8 \cdot 596 & 1062812 \cdot 201kd + 88617 \cdot 281 \end{vmatrix} > 0$$

This simplifies to

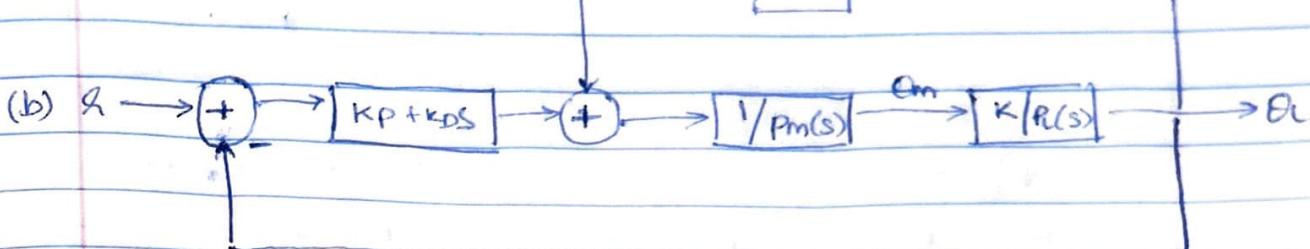
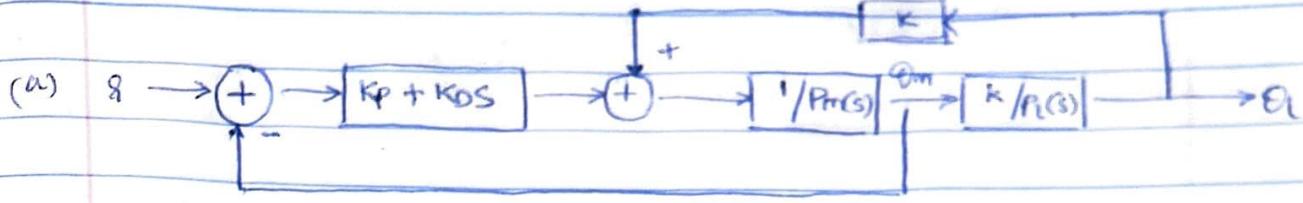
$$\begin{aligned} & -1129569774594 \cdot 46kd^2 + 171473359 \cdot 535792kd \\ & + 7081995081 \cdot 3222 > 0 \end{aligned}$$

The value of kd which satisfies this case.

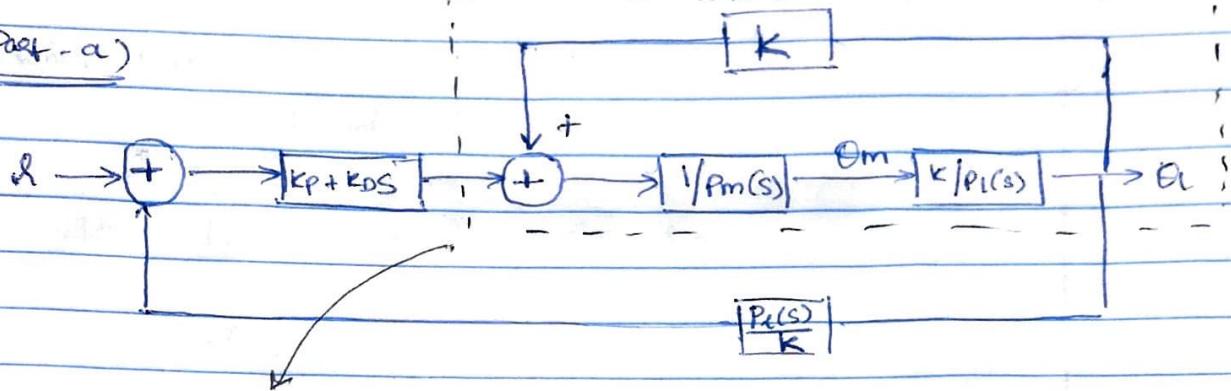
$$\boxed{-0.079105 < kd < 0.07925.}$$

→ This has to be the range of kd for the system to be stable.

Problem - 3



(Part - a)



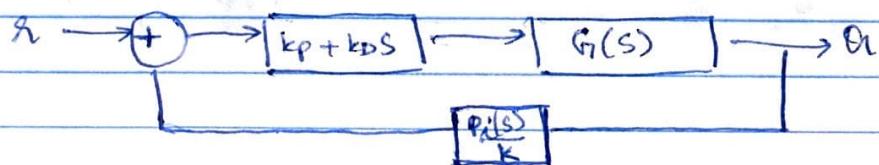
$$G_1(s) \text{ (open loop gain)} = \frac{K}{P_i(s) P_m(s)}$$

$$\text{Closed loop gain } G(s) = \frac{G_1(s)}{1 - K G_1(s)} = \frac{K}{P_i(s) P_m(s)}$$

$$= \frac{1}{1 - K \left[\frac{K}{P_i(s) P_m(s)} \right]}$$

$$G(s) = \frac{K}{P_i(s) P_m(s) - K^2}$$

now the loop simplifies to.



$$H(s) \text{ (Open-loop-gain)} = (k_p + k_d s)(G(s))$$

$$\begin{array}{l} H(s) \\ \left[\begin{array}{l} \text{closed loop gain} \\ \text{with -ve feed back} \end{array} \right] \end{array} = \frac{H'(s)}{1 + \frac{P_L(s)}{K} [H'(s)]} = \frac{(k_p + k_d s)(G(s))}{1 + \frac{P_L(s)}{K} [(k_p + k_d s)(G(s))]}$$

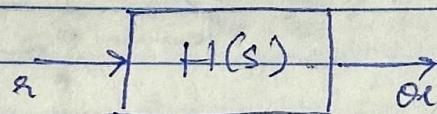
$$\begin{aligned} H(s) &= \frac{(k_p + k_d s)(G(s))}{k + P_L(s)(k_p + k_d s)(G(s))} = \frac{(k)(k_p + k_d s)(G(s))}{k + P_L(s)(k_p + k_d s)(G(s))} \\ &\quad \text{where } G(s) = k \end{aligned}$$

$$P_L(s) P_m(s) - k^2$$

$$\# P_L(s) = J_1 s^2 + B_1(s) + k$$

$$P_m(s) = J_m s^2 + B_m s + k$$

The block diagram simplifies to.



$$\text{Post: } H(s) = (k_p + k_d s)(k)(k)$$

$$\frac{(P_L(s) P_m(s) - k^2)}{k + P_L(s)(k_p + k_d s)(k)}$$

$$P_L(s) P_m(s) - k^2$$

$$H(s) = \frac{(k_p + k_d s)(k)(k)}{\cancel{(k)(P_L(s) P_m(s) - k^2)} + P_L(s)(k_p + k_d s)(k)}$$

$$H(s) = \frac{(k_p + k_d s)(k)}{(J_1 s^2 + B_1(s) + k)(J_m s^2 + B_m s + k)}$$

$$- k^2 + (J_1 s^2 + B_1(s) + k)(k_p + k_d s)$$

$$H(s) = k_{pk} + (k_{kd})s$$

$$\begin{aligned} & J_1 J_m s^4 + \cancel{J_1 B_m s^3} + \cancel{J_1 k s^2} + \cancel{J_m B_l s^3} + \cancel{B_m B_l s^2} \rightarrow B_{KL} s \\ & + \cancel{J_m k s^2} + \cancel{B_m k s} + \cancel{k} - \cancel{J_1 k p s^2} + \cancel{J_1 k D s^3} + \cancel{B_k k p s} \\ & + \cancel{B_k k D s^2} + k_{kp} + (k_{kd})s \end{aligned}$$

$$H(s) = k_{pk}s + k_{pd}$$

$$\begin{aligned} & (J_1 J_m)s^4 + (J_1 B_m + J_m B_l + J_1 k_D)s^3 + (J_1 k + J_m k \\ & + B_m B_l + J_1 k_p + B_k k_D)s^2 + (B_{KL} + B_m k + B_k k_p + k_{kd})s \\ & + k_{kp} \end{aligned}$$

$$W \cdot K \cdot T \quad J_m = J_1 = 0.0097; \quad B_m = B_l = 0.04169; \quad k = 100$$

(Subⁿ problem 2 data)

$$H(s) = \dots$$

$$H(s) = 100 k_D s + 100 k_p$$

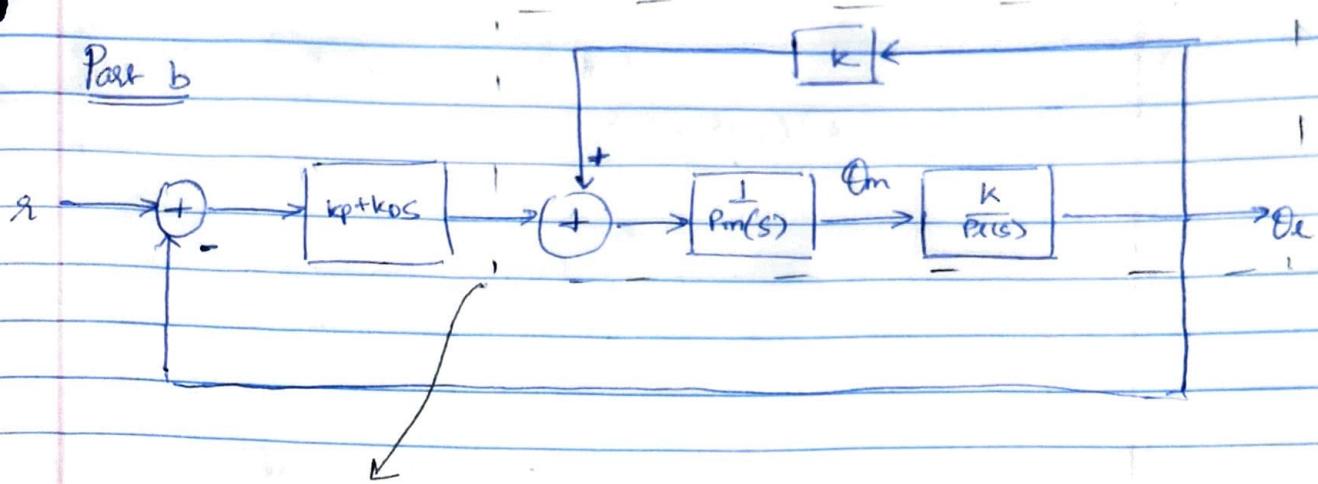
$$\begin{aligned} & (0.0009409)s^4 + (0.000808786 + 0.0097 k_D)s^3 \\ & + (1.94173806 + 0.0097 k_p + 0.04169 k_D)s^2 \\ & + (8.338 + 0.04169 k_p + 100 k_D)s + 100 k_p. \end{aligned}$$

$$\text{Sub } k_p = 10 \text{ ne getr.}$$

$$H(s) = H(s) = 100 k_D s + 1000$$

$$\begin{aligned} & (0.0009409)s^4 + (0.000808786 + 0.0097 k_D)s^3 + (+\cancel{0.04173806} \\ & + 0.0097 + 0.04169 k_D)s^2 + (-\cancel{8.338} + 0.04169 + 100 k_D)s + 1000 \\ & 2.038 \qquad \qquad \qquad 8.7569 \end{aligned}$$

Part b

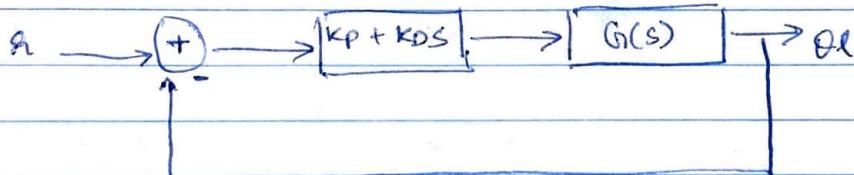


$$(\text{Open loop gain}) G_l(s) = \frac{1}{P_m(s)} \cdot \frac{K}{P_c(s)}$$

Close loop gain. (+ve Feedback)

$$G_l(s) = \frac{G_l(s)}{1 + K G_l(s)} = \frac{\frac{K}{P_m(s) P_c(s)}}{1 + \frac{-K^2}{P_m(s) P_c(s)}}$$

$$G_l(s) = \frac{K}{P_m(s) P_c(s) - K^2}$$

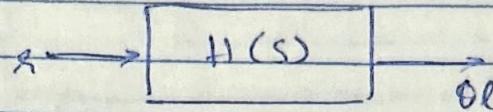


$$(\text{Open loop gain}) H'(s) = (K_p + K_d s) G_l(s)$$

$$\text{closed loop gain } H(s) = \frac{H'(s)}{1 + H'(s)} = \frac{(K_p + K_d s)(G_l(s))}{1 + (K_p + K_d s)(G_l(s))}$$

$$H(s) = \frac{(K_p + K_d s) \left(\frac{K}{P_m(s) P_c(s) - K^2} \right)}{1 + (K_p + K_d s) \left(\frac{K}{P_m(s) P_c(s) - K^2} \right)}$$

$$H(s) = \frac{(kp + kds)(k)}{P_m(s)P_L(s) - k^2 + (kp + kds)(k)}$$



where

$$P_m(s) = J_m s^2 + B_m s + k ; \quad P_L(s) = J_L s^2 + B_L s + k .$$

$$H(s) = \frac{(kp + kds)(k)}{(J_m s^2 + B_m s + k)(J_L s^2 + B_L s + k) - k^2 + (kp + kds)k}$$

$$H(s) = K_{KDS} s + K_{KP}$$

$$\begin{aligned} & J_m J_L s^4 + J_m B_L s^3 + J_m k s^2 + B_m J_L s^3 + B_m B_L s^2 + K B_m s \\ & + J_L k s^2 + B_L k s + k^2 - k^2 + K_{KDS} + K_{KP} \end{aligned}$$

$$H(s) = K_{KDS} s + K_{KP}$$

$$\begin{aligned} & J_m J_L s^4 + (J_m B_L + B_m J_L) s^3 + (J_m k + B_m B_L + J_L k) s^2 \\ & + (K B_m + B_L k + K_{KDS}) s + K_{KP} \end{aligned}$$

$$H(s) = \text{given } k = 100 ; \quad J_m = J_L = 0.00094097 ; \\ B_L = B_m = 0.04169$$

$$H(s) = 100 K_{KDS} s + 100 K_{KP}$$

$$\begin{aligned} & (0.00094097) s^4 + (0.000808786) s^3 + (1.94173806) s^2 \\ & + (8.338 + 100 K_{KDS}) s + 100 K_{KP} \end{aligned}$$

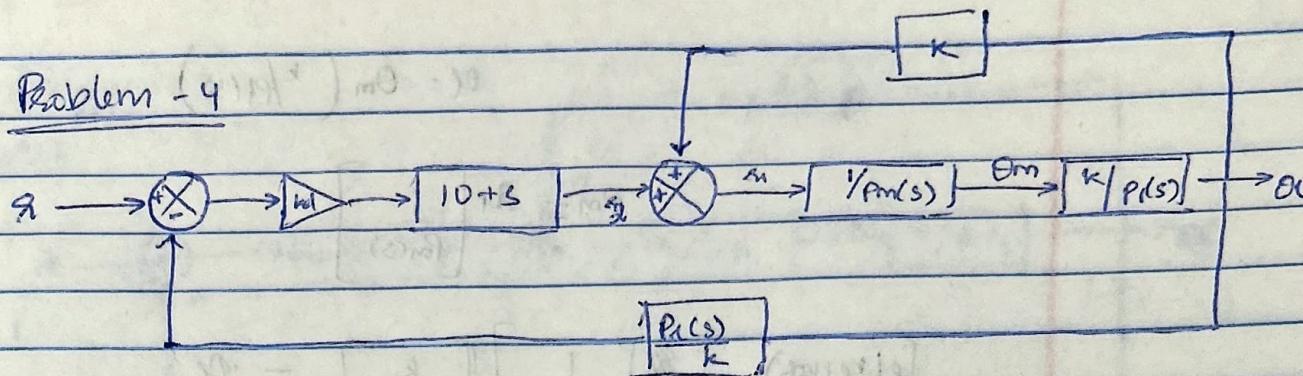
Sub $K_{KDS} = 10$ we get

$$H(s) = 100 K_{KDS} s + 1000$$

$$\begin{aligned} & (0.00094097) s^4 + (0.000808786) s^3 + (1.94173806) s^2 + (8.338 + 1000) s \\ & + 1000 \end{aligned}$$

Problem - 4

(Part a)



$$\Theta_l = \frac{k \theta_m}{P_l(s)}; \quad \theta_m = \frac{1}{P_m(s)} r_1 \Rightarrow \Theta_l = \frac{k r_1}{P_l(s) P_m(s)}$$

$$r_1 = r_2 + k \theta_l.$$

$$\Rightarrow \Theta_l = \frac{k}{P_l(s) P_m(s)} (k \theta_l + r_2) \Rightarrow \Theta_l = \frac{k}{P_l(s) P_m(s)} (k \theta_l + (k_p + k_d s) r)$$

Note → Taken $k_D = 1$; $k_P = 10$ (Maintaining the ratio $k_P/k_D = 10$)

$$Q(s) = \frac{Q_l(s)}{P_l(s)} = \frac{ks + 10k}{s^4 + (B_e J_m + B_m J_e) s^3 + (K_{J_l} + K_{J_m} + B_e B_m) s^2 + (k_B J_l + k_B J_m) s}$$

$$H(s) = \frac{Q(s) \cdot P_l(s)}{K} = \frac{(s + 10)(s^2 + B(s + k))}{s^4 + (B_e J_m + B_m J_e) s^3 + (k_{J_l} + k_{J_m} + B_e B_m) s^2 + (k_B J_l + k_B J_m) s}$$

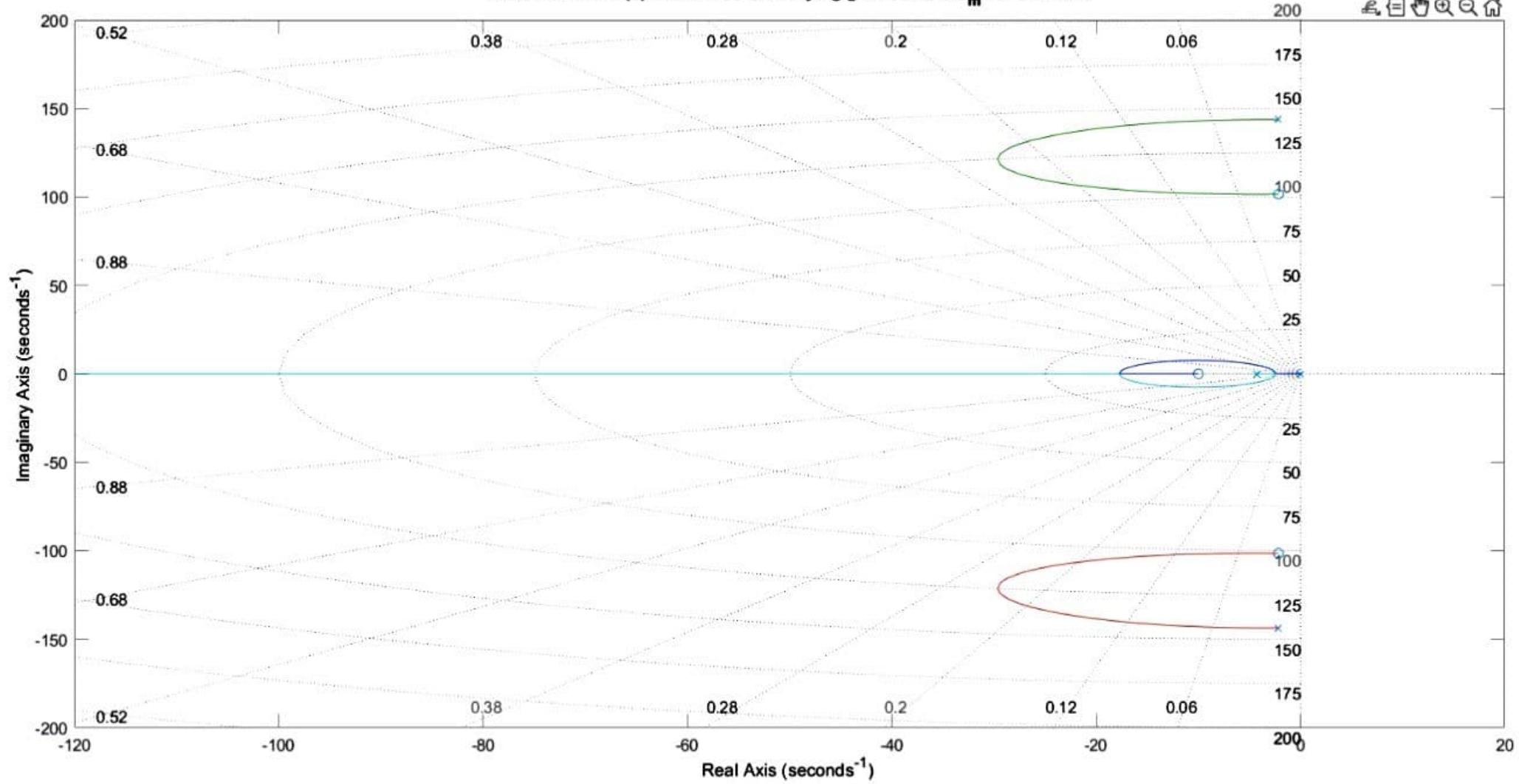
Sub Values $J_m = J_e = 0.0097$, $B_m = B_e = 0.00169$, $K = 100$, $k_p = 10$

$H(s) =$

$$H(s) = \frac{0.0097 s^3 + 0.1386 s^2 + 100.4169 s + 1000}{9.409 \times 10^5 s^4 + 0.00080878 s^3 + 0.921738056 s^2 + 8.3477 s}$$

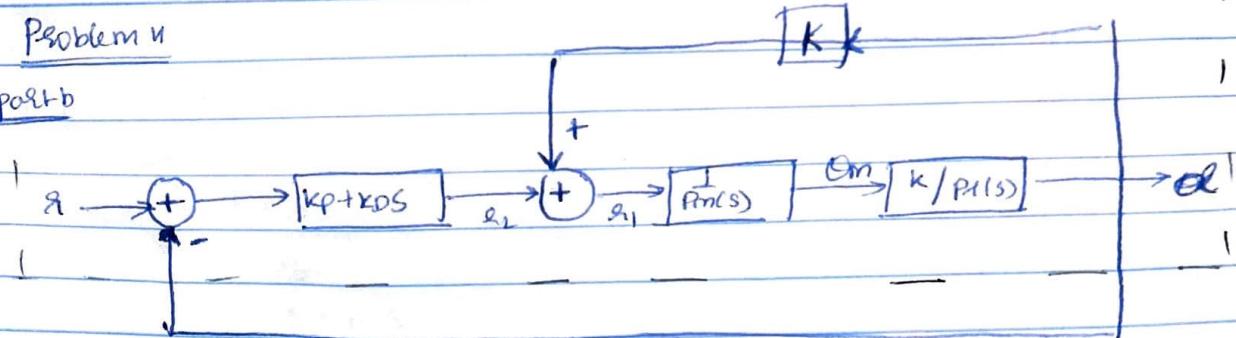
Root Locus of $H(s)$ with KD as the varying gain and θ_m as feedback

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Problem n

part b



$$\Theta_l = \Theta_m \frac{k}{P_L(s)} \quad \left\{ \Theta_m = \frac{\theta_1}{P_m(s)} \right\}$$

$$\Rightarrow \Theta_l = \frac{\theta_1 k}{P_m(s) P_L(s)} \quad \left\{ \theta_1 = k \Theta_l + \theta_2 \right\}.$$

$$\Rightarrow \Theta_l = \frac{(k \Theta_l + \theta_2) k}{P_m(s) P_L(s)} \quad \left\{ \theta_2 = (k_p + k_D s) \theta_1 \right\}$$

$$\Rightarrow \Theta_l = \frac{[k \Theta_l + (k_p + k_D s) \theta_1] k}{P_m(s) P_L(s)}$$

The open loop transfer function [Derived from Matlab]

$$Y(s) =$$

$$\Rightarrow \Theta_l = \frac{k^2 \Theta_l}{P_m(s) P_L(s)} + \frac{k \theta_1 (k_p + k_D s)}{P_m(s) P_L(s)}$$

$$\Theta_l \left[1 - \frac{k^2}{P_m(s) P_L(s)} \right] = \left[\frac{k (k_p + k_D s)}{P_m(s) P_L(s)} \right] \theta_1$$

$$Q(s) = \frac{\Theta_l}{\theta_1} = \frac{k (k_p + k_D s)}{P_m(s) P_L(s)} = \frac{k (k_p + k_D s)}{P_m(s) P_L(s) - k^2}$$

$$\frac{k(k_p + k_D s)}{(J_1 s^2 + B_1 s + k) (J_m s^2 + B_m s + k)} = \Theta(s)$$

$$\Theta(s) = \frac{K k_p + K k_D s}{J_1 J_m s^4 + J_1 (B_m s^3 + J_1 k s^2 + B_1 J_m s^3 + B_m B_1 s^2 + B_1 k s + k J_m s^2 + K B_m s + k^2 - k^2)}$$

$\Theta(s) \approx$ Subⁿ values i.e. $k_p = 10$; $k_D = 1 \Rightarrow \frac{k_p}{k_D} = 10$

$$+ J_1 = J_m = 0.0097; b_m = b_l = 0.04169 \quad k = 100$$

$$\Theta(s) = \frac{1000 + 100s}{0.00009409 s^4 + 0.000808786 s^3 + 0.9717380561 s^2 + 941 s^2 + 8.338 s + 100}$$

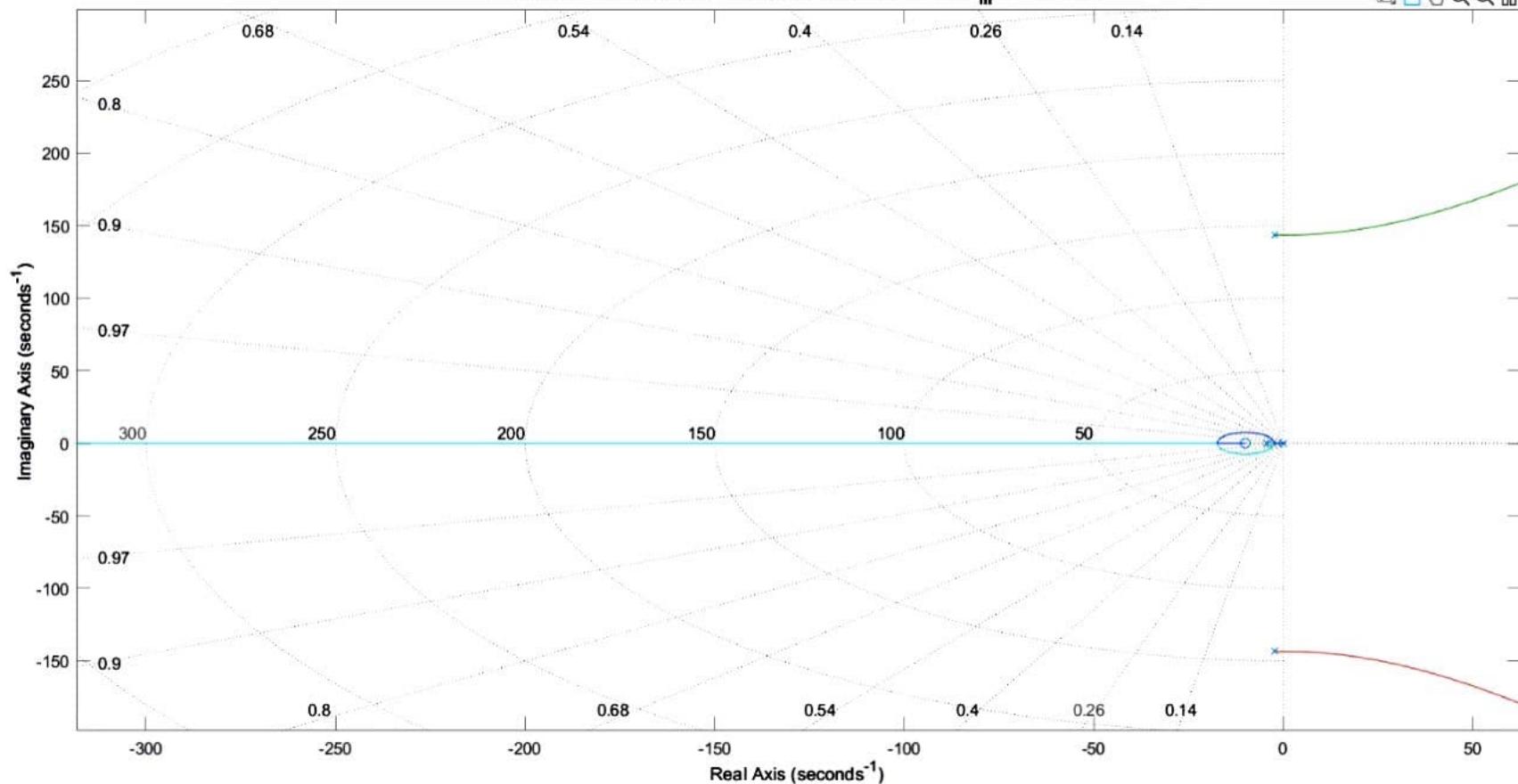
$$A(s) = \Theta(s) \times 1 \quad \leftarrow H(s) = \Theta(s)$$

$$\therefore H(s) = \Theta(s) \times 1$$

$$H(s) = 1000 + 100s$$

$$0.00009409 s^4 + 0.000808786 s^3 + 1.941 s^2 + 8.338 s + 0.$$

Root Locus of $H(s)$ with K_D as the varying gain and θ_m as feedback



Problem 4

(a) When the motor angle is fed back to the system & the root locus is obtained it can be observed that the gain k_d does not cross the left hand side of the open plane (or) OHP, that means for any value of $k_d \geq 0$ the system is stable & there is no upper bound.

→ To give a better estimate of the lower bound $k_d > -0.07$

2 The Routh-Hurwitz Criteria from which we get $k_d > -0.079$.
 { In graph below shows the min value of k_d on the root locus }
 { Gure could point it to -0.0327 but on closer inspection }
 { it can be seen that k_d is indeed greater than -0.079 }

(b) When the load angle is fed back to the system & the root locus is obtained it can be observed that the poles cross the open left hand plane & hence there will be an upper bound & lower bound for k_d when we use the root-locus to find these bounds it can be observed that there is a slight discrepancy with the Routh-Hurwitz Criteria.

The Routh-Hurwitz Criteria tells that the range of k_d for the system to be stable should be

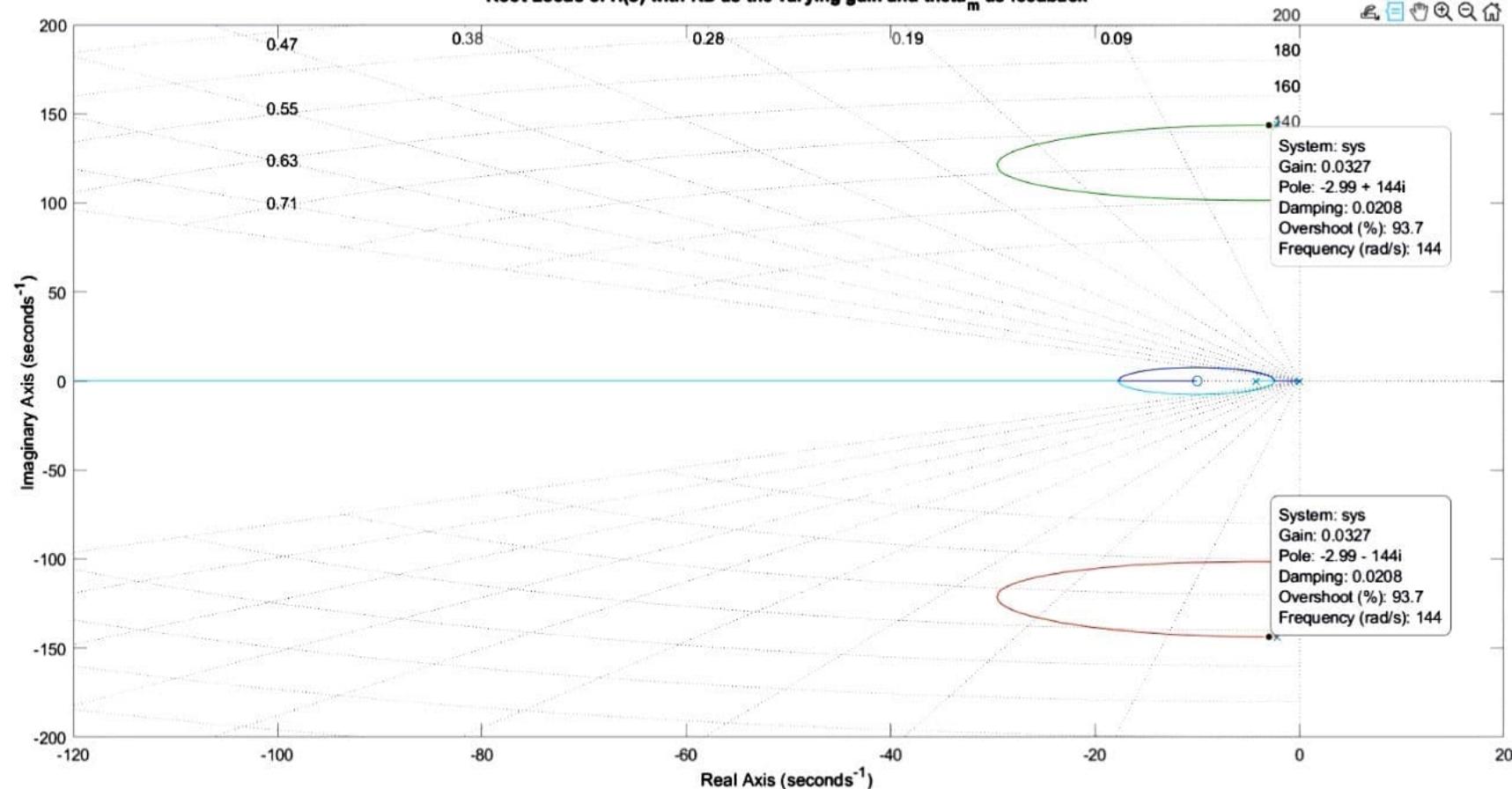
$$-0.0791052 \leq k_d \leq 0.07926$$

but that's not the case with the root-locus, the range of k_d from the root-locus is

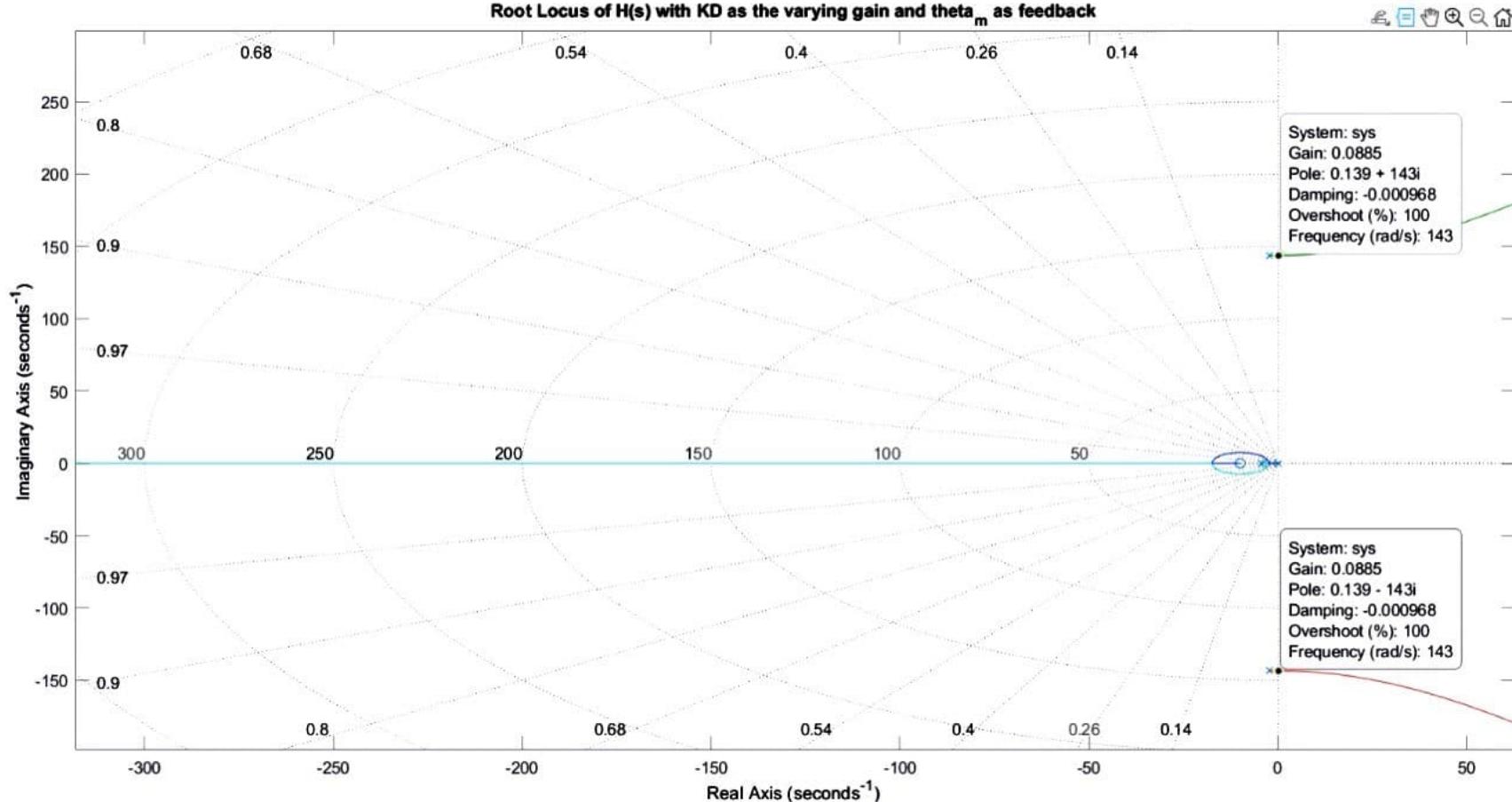
$$-0.083 \leq k_d \leq +0.083$$

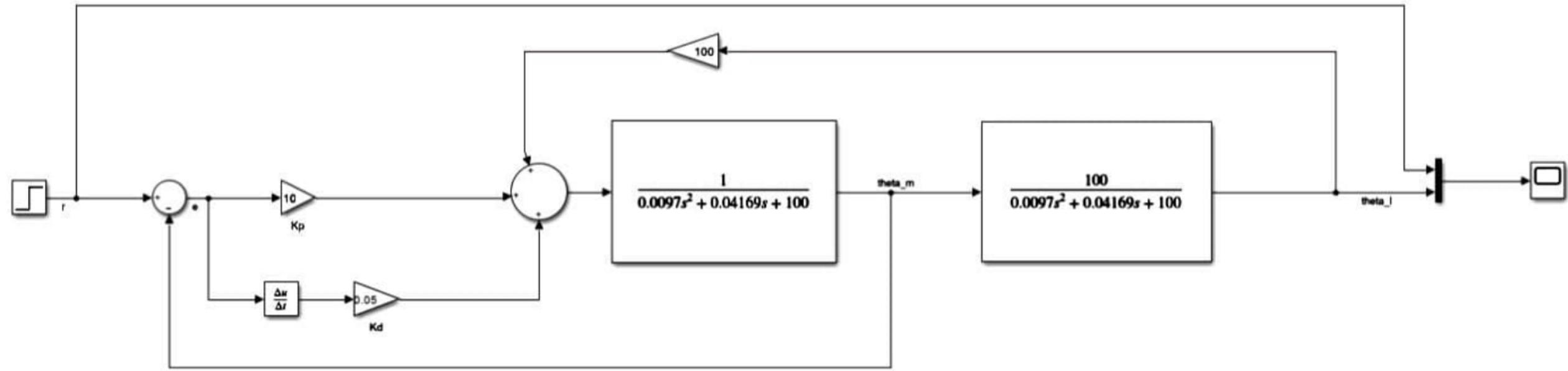
This discrepancy is because in the Routh-Hurwitz Criteria we fix k_p & then then find the value of k_d but in the root-locus we fix the ratio of k_p/k_d & then represent k_p in terms of k_d & then plot the root locus. So due to this the discrepancy occurs between the time domain (Routh-Hurwitz's) & frequency domain (Root-locus).

Root Locus of $H(s)$ with KD as the varying gain and θ_m as feedback



Root Locus of $H(s)$ with KD as the varying gain and θ_m as feedback

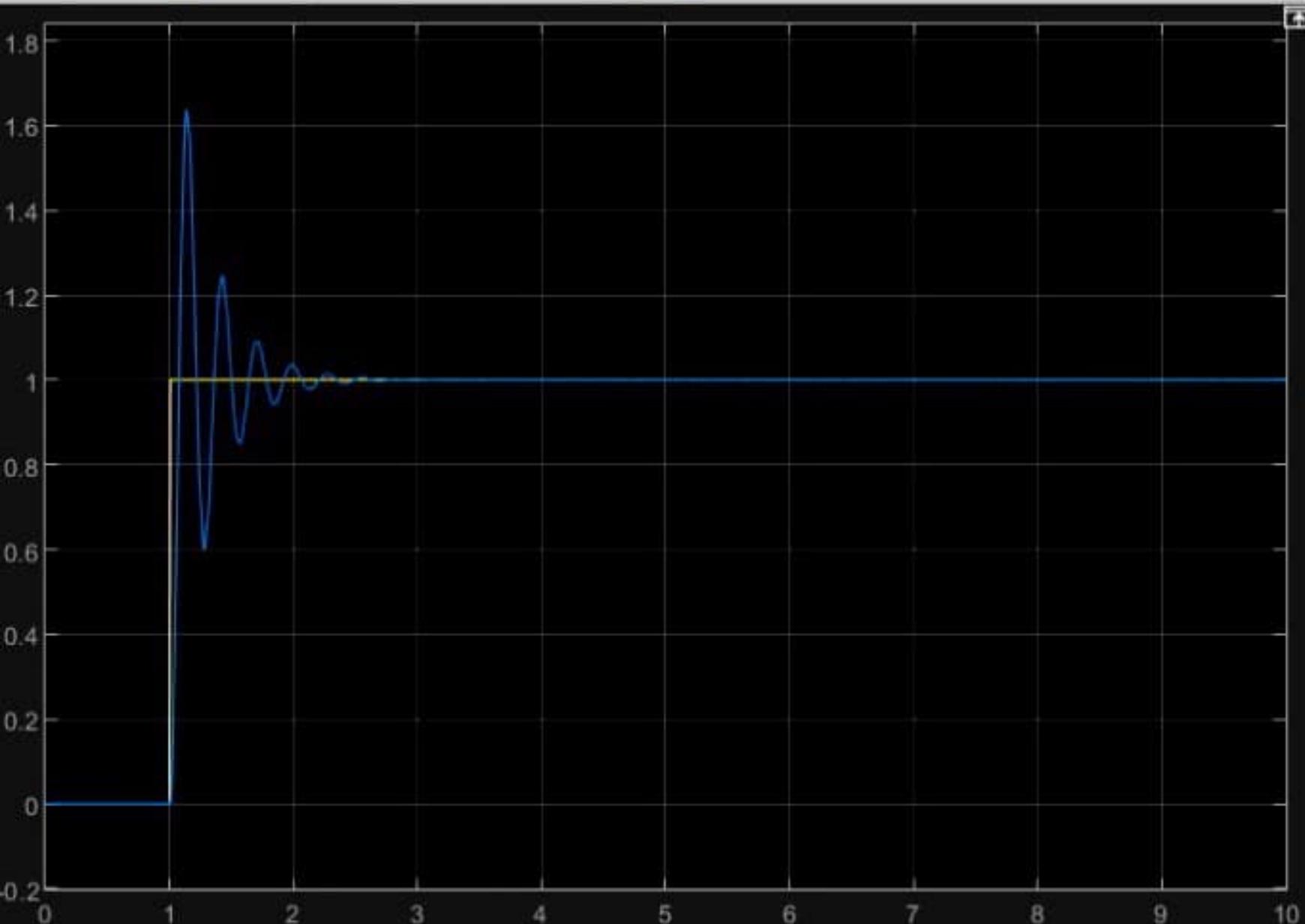


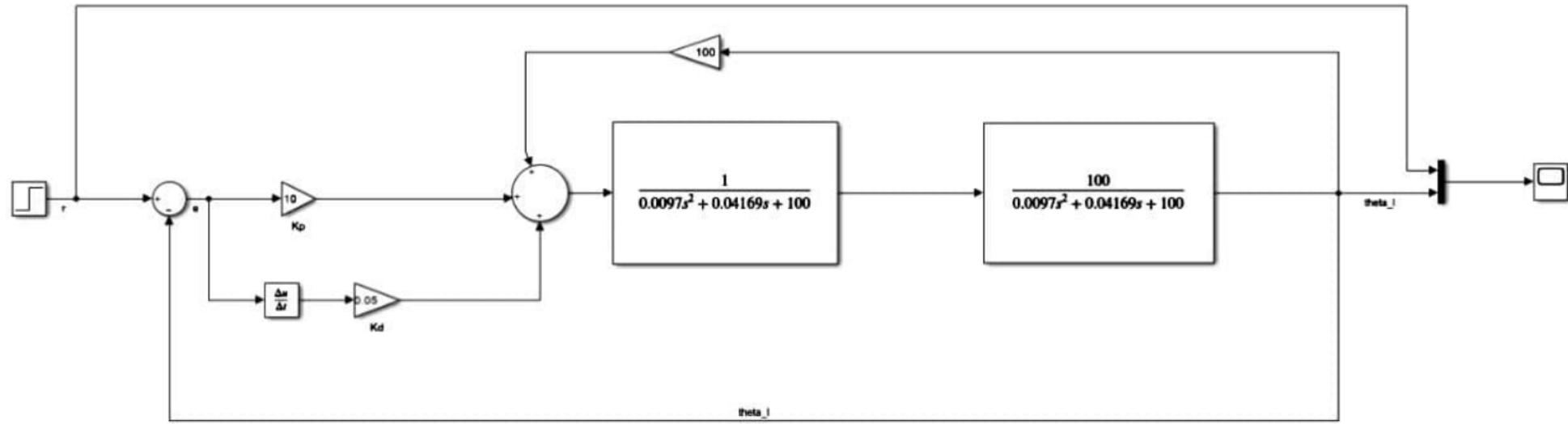


 Scope

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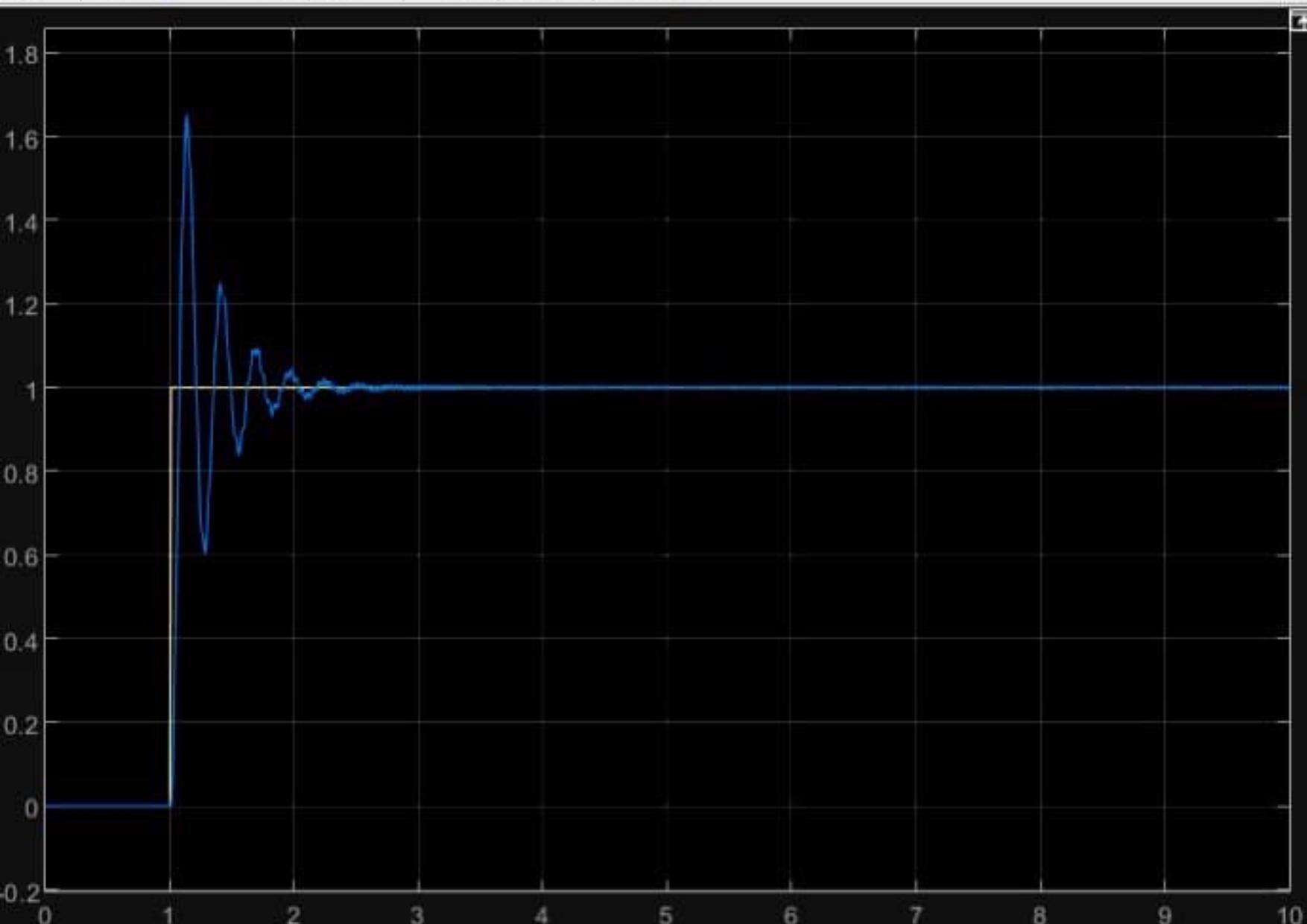
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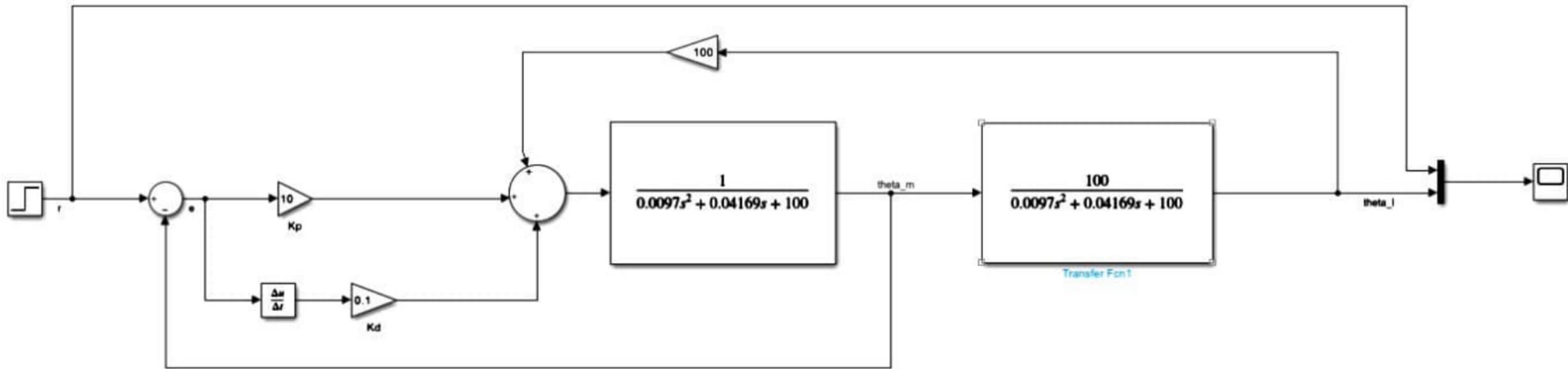




Scope

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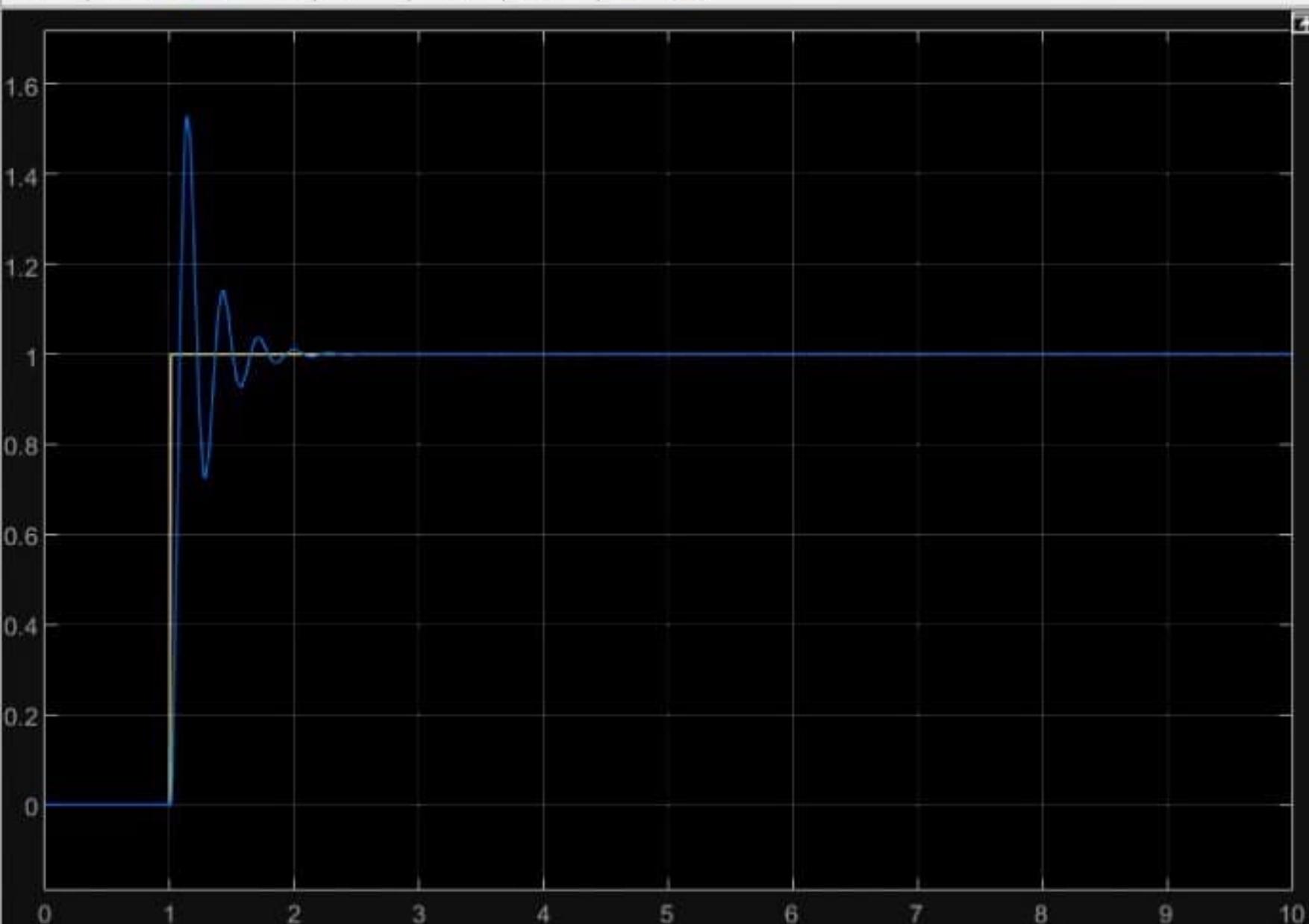
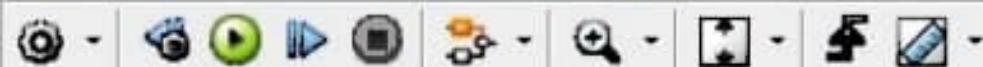


P5_b_theta_m

Scope

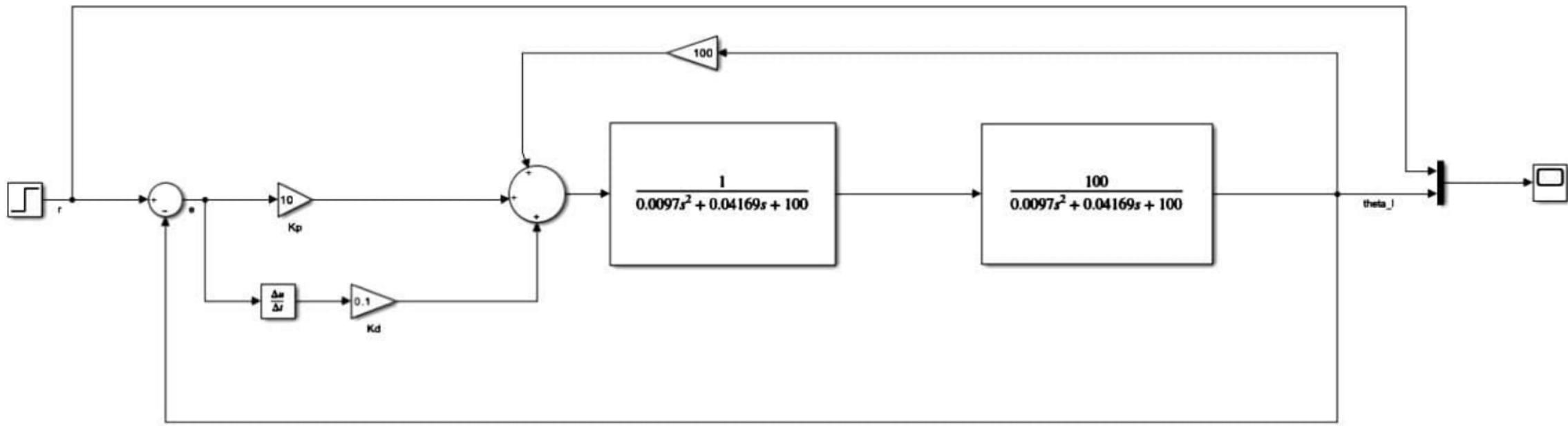


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Ready

Sample based T=10.000



P5_b_theta_1

