

A primer on Generalized, Normalized, Power and Active S-Parameters

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Generalized S-Parameters

- They may have different reference impedances at each port.
- The travelling waves at the two ports are defined as

$$\begin{aligned}a_1 &= \frac{V_1 + Z_{01}I_1}{2\sqrt{Z_{01}}} & a_2 &= \frac{V_2 + Z_{02}I_2}{2\sqrt{Z_{02}}} \\b_1 &= \frac{V_1 - Z_{01}I_1}{2\sqrt{Z_{01}}} & b_2 &= \frac{V_2 - Z_{02}I_2}{2\sqrt{Z_{02}}}\end{aligned}$$

- Simulation solvers internally give out generalized S-parameters with the port impedances as the reference impedance at each port. They are then converted to Normalized S-parameters.

- Suppose a solver gives the generalized 2-port S-parameters at a particular frequency as

$$\begin{aligned}\mathbf{S}_G &= \begin{bmatrix} \left. \frac{b_1}{a_1} \right|_{a_2=0} & \left. \frac{b_1}{a_2} \right|_{a_1=0} \\ \left. \frac{b_2}{a_1} \right|_{a_2=0} & \left. \frac{b_2}{a_2} \right|_{a_1=0} \end{bmatrix} = \begin{bmatrix} -0.234 + 0.2i & -0.628 - 0.672i \\ -0.628 - 0.672i & -0.328 + 0.213i \end{bmatrix} \\ |\mathbf{S}_G| &= \begin{bmatrix} 0.308 & 0.92 \\ 0.92 & 0.391 \end{bmatrix} = \begin{bmatrix} -17.0 & -0.291 \\ -0.291 & -11.9 \end{bmatrix} dB\end{aligned}$$

- The two device has port impedance of

$$\mathbf{Z}_P = \begin{bmatrix} Z_{01} & 0 \\ 0 & Z_{02} \end{bmatrix} = \begin{bmatrix} 50.93 & 0 \\ 0 & 88.07 \end{bmatrix}$$

i.e. The S-parameters S_G are calculated with the ports terminated with matched impedance according to Z_P

- A unique impedance matrix \mathbf{Z} can be calculated as

$$\begin{aligned}\mathbf{Z} &= \sqrt{\mathbf{Z}_P}(\mathbf{I} - \mathbf{S}_G)^{-1}(\mathbf{I} + \mathbf{S}_G)\sqrt{\mathbf{Z}_P} \\ &= \begin{bmatrix} 3.94 + 32.4i & -3.39 - 57.2i \\ -3.39 - 57.2i & 0.213 + 52.0i \end{bmatrix} \\ |\mathbf{Z}_P| &= \begin{bmatrix} 32.7 & 57.3 \\ 57.3 & 52.0 \end{bmatrix}\end{aligned}$$

- From this impedance matrix, the S-parameters can be normalized to a common reference impedance, usually to 50Ω as

$$\begin{aligned}\mathbf{Z}_N &= \begin{bmatrix} 50 & 0 \\ 0 & 50 \end{bmatrix} \\ \mathbf{S}_N &= \sqrt{\mathbf{Y}_N}(\mathbf{Z} - \mathbf{Z}_N)(\mathbf{Z} + \mathbf{Z}_N)^{-1}\sqrt{\mathbf{Z}_N} \\ &= \begin{bmatrix} -0.225 - 0.0569i & -0.708 - 0.667i \\ -0.708 - 0.667i & -0.0433 + 0.228i \end{bmatrix} \\ |\mathbf{S}_N| &= \begin{bmatrix} 0.232 & 0.973 \\ 0.973 & 0.232 \end{bmatrix} = \begin{bmatrix} -12.7 & -0.24 \\ -0.24 & -12.7 \end{bmatrix} dB\end{aligned}$$

Interpretation from \mathbf{S}_G

S_{11} Insertion Loss at Port 1 with Port 2 terminated in matched load ($a_2 = 0$)

S_{12} Reverse Voltage Gain (From port 2 to port 1)

S_{21} Forward Voltage Gain (From port 1 to port 2)

S_{22} Insertion Loss at Port 2 with Port 1 terminated in matched load ($a_1 = 0$)

Normalized S-Parameters

- The travelling waves at the two ports are defined with a common reference impedance of Z_0

$$\begin{aligned}a_1 &= \frac{V_1 + Z_0I_1}{2\sqrt{Z_0}} & a_2 &= \frac{V_2 + Z_0I_2}{2\sqrt{Z_0}} \\b_1 &= \frac{V_1 - Z_0I_1}{2\sqrt{Z_0}} & b_2 &= \frac{V_2 - Z_0I_2}{2\sqrt{Z_0}}\end{aligned}$$

- They usually refer to the measured S-parameters

- The nominal common reference impedance is $Z_0 = 50\Omega$.

- The Normalized S-parameters obtained from measurement or simulation with reference impedance of 50Ω be

$$\mathbf{S}_N = \begin{bmatrix} -0.255 - 0.196i & 0.739 + 0.592i \\ 0.739 + 0.592i & 0.247 + 0.205i \end{bmatrix}$$

- These Normalized S-Parameters can be converted into Generalized S-parameters with any reference impedance at the ports. For that, the unique impedance matrix can be calculated as

$$\begin{aligned}\mathbf{Z} &= \sqrt{\mathbf{Z}_N}(\mathbf{I} - \mathbf{S}_N)^{-1}(\mathbf{I} + \mathbf{S}_N)\sqrt{\mathbf{Z}_N} \\ \mathbf{Z} &= \begin{bmatrix} 0.000122 + 36.4i & 0.000173 + 75.0i \\ 0.000173 + 75.0i & 0.000756 + 87.3i \end{bmatrix}\end{aligned}$$

- Then the Generalized S-parameters can be obtained as

$$\begin{aligned}\mathbf{Z}_G &= \begin{bmatrix} 50.93 & 0 \\ 0 & 88.10 \end{bmatrix} \\ \mathbf{S}_G &= \sqrt{\mathbf{Y}_G}(\mathbf{Z} - \mathbf{Z}_G)(\mathbf{Z} + \mathbf{Z}_G)^{-1}\sqrt{\mathbf{Z}_G} \\ &= \begin{bmatrix} -0.221 + 0.0653i & 0.721 + 0.653i \\ 0.721 + 0.653i & -0.0433 + 0.226i \end{bmatrix}\end{aligned}$$

- This exercise simply says that the reference impedance of the S-parameters can be converted to the required reference impedances

Interpretation from \mathbf{S}_N

S_{11} Insertion Loss at Port 1 with Port 2 terminated in common reference load Z_0

S_{12} Reverse Gain (From port 2 to port 1)

S_{21} Forward Gain (From port 1 to port 2)

S_{22} Insertion Loss at Port 2 with Port 1 terminated in common reference load Z_0

The following definitions are valid w.r.t \mathbf{S}_N when the source and load impedances are matched to the reference impedance Z_0

$$\text{Transducer Gain } G_T = \frac{\text{Power transferred to Load}}{\text{Available power from Generator}} = |S_{21}|^2$$

$$\text{Available Power Gain } G_a = \frac{\text{Available power from Network}}{\text{Available Power from Generator}} = \frac{|S_{21}|^2}{1 - |S_{22}|^2}$$

$$\text{Operating Power Gain } G_P = \frac{\text{Power Transferred to Load}}{\text{Power fed to the DUT}} = \frac{|S_{21}|^2}{1 - |S_{11}|^2}$$

Power S-Parameters

- The **power waves** at the two ports can be defined as

$$\begin{aligned}a_1 &= \frac{V_1 + Z_G I_1}{2\sqrt{R_G}} & a_2 &= \frac{V_2 + Z_L I_2}{2\sqrt{R_L}} \\b_1 &= \frac{V_1 - Z_G^* I_1}{2\sqrt{R_G}} & b_2 &= \frac{V_2 - Z_L^* I_2}{2\sqrt{R_L}}\end{aligned}$$

- In the above definitions, it can be observed that the denominators have only real part of the impedance under the square root.

- These are defined by anticipating complex source and load of impedances of Z_G and Z_L respectively.

- The power waves can be interpreted in terms of power transferred to and from the DUT.

- Assume that the source and load impedance are same as the port impedance.i.e. $Z_G = Z_{01}$ and $Z_L = Z_{02}$. Then the Power S-parameters would be same as the Generalized S-Parameters

$$\mathbf{S}_P = \mathbf{S}_G = \begin{bmatrix} -0.234 + 0.2i & -0.628 - 0.672i \\ -0.628 - 0.672i & -0.328 + 0.213i \end{bmatrix}$$

- Consider a scenario where the reference impedance be complex. Say $Z_{01} = 45 + 90i\Omega$ and $Z_{02} = 60 - 80i\Omega$. The Generalized impedance in this case are

$$\begin{aligned}S_{GC} &= \begin{bmatrix} -0.234 + 0.2i & -0.628 - 0.672i \\ -0.628 - 0.672i & -0.328 + 0.213i \end{bmatrix} \\ |S_{GC}| &= \begin{bmatrix} 0.308 & 0.92 \\ 0.92 & 0.391 \end{bmatrix} = \begin{bmatrix} -10.2 & -0.721 \\ -0.721 & -8.15 \end{bmatrix} dB\end{aligned}$$

The Power S-Parameters are

$$\begin{aligned}S_{PC} &= \begin{bmatrix} 0.694 + 0.44i & -0.295 - 0.445i \\ -0.295 - 0.445i & -0.117 - 0.873i \end{bmatrix} \\ |S_{PC}| &= \begin{bmatrix} 0.822 & 0.534 \\ 0.534 & 0.881 \end{bmatrix} = \begin{bmatrix} -1.71 & -5.45 \\ -5.45 & -1.1 \end{bmatrix} dB\end{aligned}$$

- Consider a case where the source and load impedance are almost matched to the network i.e. $Z_{01} = 32\Omega$ and $Z_{02} = 49\Omega$, the Power S-Parameters are

$$\begin{aligned}S_{P2} &= \begin{bmatrix} -0.00506 - 0.0699i & -0.718 - 0.694i \\ -0.718 - 0.694i & -0.0434 + 0.00954i \end{bmatrix} \\ |S_{P2}| &= \begin{bmatrix} 0.0701 & 0.999 \\ 0.999 & 0.0445 \end{bmatrix}\end{aligned}$$

- The corresponding gains are $G_T = 0.997$, $G_a = 0.999$ and $G_P = 0.999$

Interpretation from \mathbf{S}_P

$$\text{Transducer Gain } G_T = \frac{\text{Power transferred to Load}}{\text{Available power from Generator}} = |S_{21}|^2$$

$$\text{Available Power Gain } G_a = \frac{\text{Available power from Network}}{\text{Available Power from Generator}} = \frac{|S_{21}|^2}{1 - |S_{22}|^2}$$

$$\text{Operating Power Gain } G_P = \frac{\text{Power Transferred to Load}}{\text{Power fed to the DUT}} = \frac{|S_{21}|^2}{1 - |S_{11}|^2}$$

Reference : <https://www.ece.rutgers.edu/orfanidi/ewa/>

Active S-Parameters

- Active S-Parameters are used to represent only the reflection coefficients when other ports are excited.

- One of the (possible) of definition of the Active S-parameter at port 1 (for a two port network is)

$$\text{Active } S_1 = S_{11} + S_{12} \frac{a_2}{a_1}$$

where a_1 and a_2 are related to the source powers at the ports 1 and 2.

- Similarly, the Active S-parameter at the port 2 is

$$\text{Active } S_2 = S_{22} + S_{21} \frac{a_1}{a_2}$$

- Note: When power supplied to port 2 is zero, the *Active* S_1 is equivalent to S_{11} and vice-versa

- As evident from the definition, active S-parameters vary depend on the power with which the other ports are excited. For the same structure with the Normalized S-parameters S_N , the active S-parameters for the two ports as the power is varied in the other port is tabulated below:

P1 Power	P2 Power	<i>Active</i> S_1
1W	0	$ -0.225 + 0.0568i = 0.23 = -12.7dB$
1W	0.1W	$ -0.449 - 0.154i = 0.474 = -6.4dB$
1W	0.5W	$ -0.726 - 0.415i = 0.836 = -1.6dB$
1W	1W	$ -0.933 - 0.610i = 1.114 = 0.94dB$

P1 Power	P2 Power	<i>Active</i> S_2
0	1W	$ -0.043 + 0.228i = 0.23 = -12.7dB$
0.1W	1W	$ -0.267 + 0.017i = 0.267 = -11.4dB$
0.5W	1W	$ -0.544 - 0.243i = 0.59 = -4.49dB$
1W	1W	$ -0.751 - 0.439i = 0.870 = -1.21dB$

- It can be interpreted that the Active S-parameters vary with the power at the other port.

- These definitions can be extended for a network of higher number of ports.

Interpretation from \mathbf{S}_A

S_{11} Insertion Loss at Port 1 when the other port is excited

S_{22} Insertion Loss at Port 2 when the other port is excited