

Q1. Let $H(z) = 6 + z^{-1} - z^{-2}$

- Specify all possible causal systems that have the same magnitude response as $H(z)$. Although these systems may have different phase response.
- Plot the magnitude and phase response of all these equivalent systems using MATLAB. Use `freqz` command to get frequency response.
- Among all these systems which one is having minimum phase value at all frequencies.

AIM: To find the all-possible causal system transfer functions that have the same magnitude response of given $H(Z)$. And plot the magnitude and phase response of those transfer functions using `freqz` command.

Short Theory:

The system is described by a rational transfer function with real coefficients

$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_M z^{-M}}{a_0 + a_1 z^{-1} + \dots + a_N z^{-N}}.$$

- For a given transfer function $H(z)$, we can get the poles and zeros location. The magnitude response for the conjugate reciprocal locations of poles and zeros is also the same as the magnitude response of $H(z)$.
- a system is considered to be minimum phase if all of its zeros as well as all poles are inside the unit circle.

Key Commands:

`Filt % filt(num,den)` creates a discrete-time transfer function sys with numerator(s) num and denominator(s) den

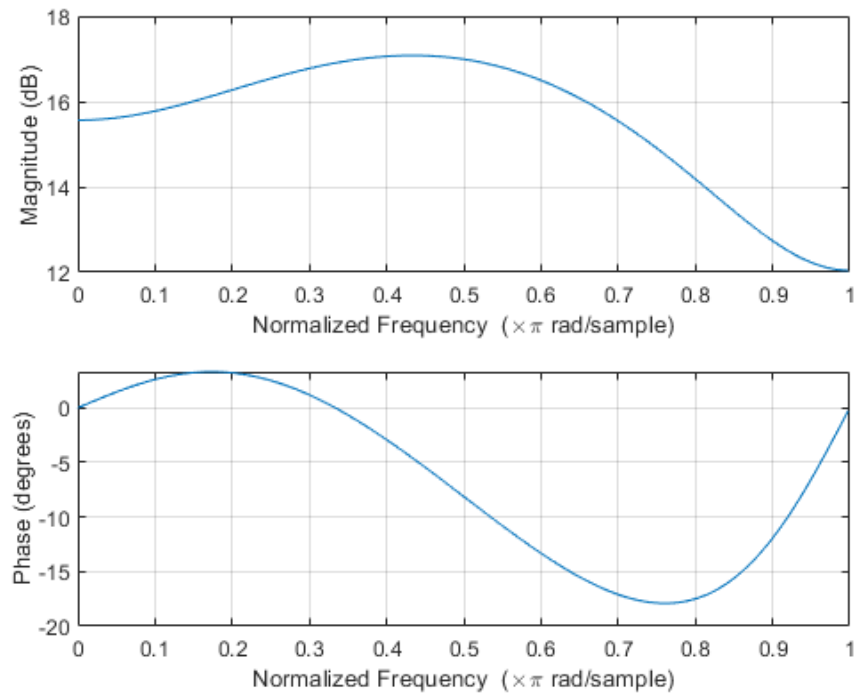
`Freqz % freqz(b,a,n)` returns the n-point frequency response vector h and the corresponding angular frequency vector w for the digital filter with transfer function coefficients stored in b and a.

`tf2zpk % tf2zpk(b,a)` finds the matrix of zeros z, the vector of poles p, and the associated vector of gains k from the transfer function parameters b and a

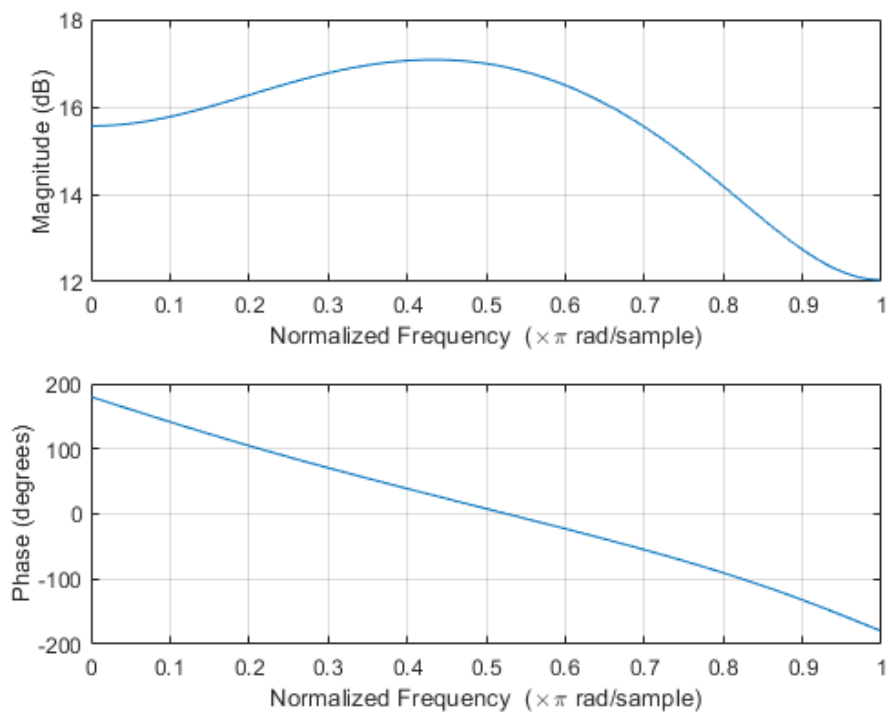
`zp2tf % zp2tf(z,p,k)` converts a factored transfer function representation

Plots:

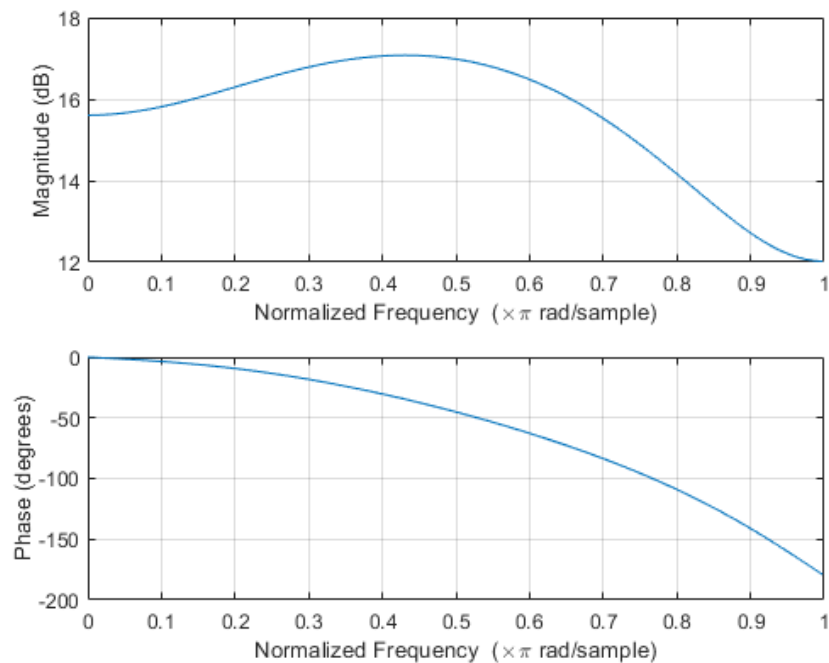
magnitude response of $H_1(z) = (6 + z^{-1} - z^{-2})$



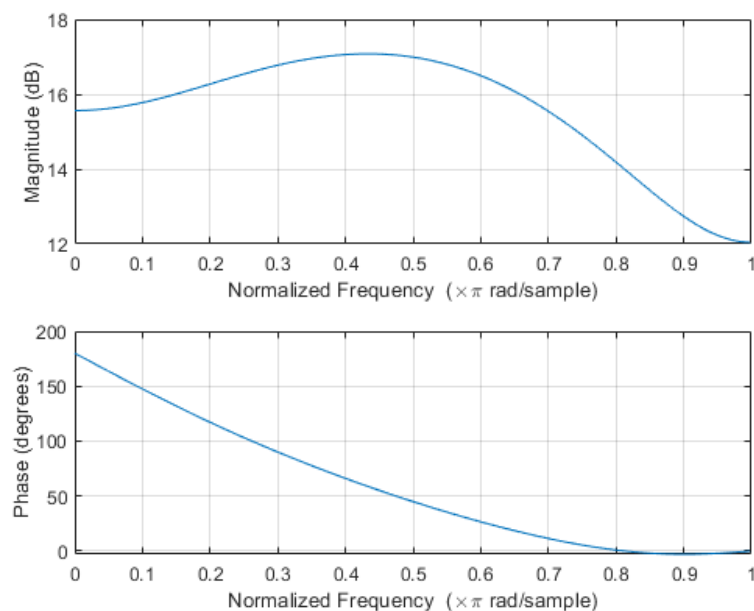
magnitude response of $H_2(z) = (1 - z^{-1} - 6 z^{-2})$



magnitude response of $H_3(z) = (3 + 5.01 z^{-1} - 1.98 z^{-2})$



magnitude response of $H_4(z) = (2 - 5 z^{-1} - 3 z^{-2})$



Inferences/comments:

- 1) The conjugate reciprocal locations of a poles and zeros of a $H(z)$ also same as the magnitude response of $H(z)$ even though the phase response is different.
- 2) A system function $H(z)$ is said to be a minimum phase system if all of its poles and zeros are within the unit circle. I.e. $H_1(z) = (6 + z^{-1} - z^{-2})$ has zeros at $z = -0.5$ and $1/3$.

Q2. A causal stable all-pass system $H_1(z)$ with real valued impulse response has only one pole at $z = -0.5$

(a) write $H_1(z)$ and draw its pole-zero plot.

(b) Plot the magnitude and phase response of $H_1(e^{j\omega})$ equivalent systems using MATLAB. Use freqz command.

AIM: To design the all pass system that is having one real pole at $z=-0.5$ and plot the magnitude and phase response using freqz command.

Short Theory:

The system which allow all the frequencies to pass through them is all pass system. The poles and zeros of a discrete time all pass system are present on the circles with radii r_k and $1/r_k$. If poles and zeros are complex, for every pole there will be conjugate reciprocal zero. All pass system is generally used to adjust the phase response of a n LTI system.

Key Commands:

Filt % filt(num,den) creates a discrete-time transfer function sys with numerator(s) num and denominator(s) den

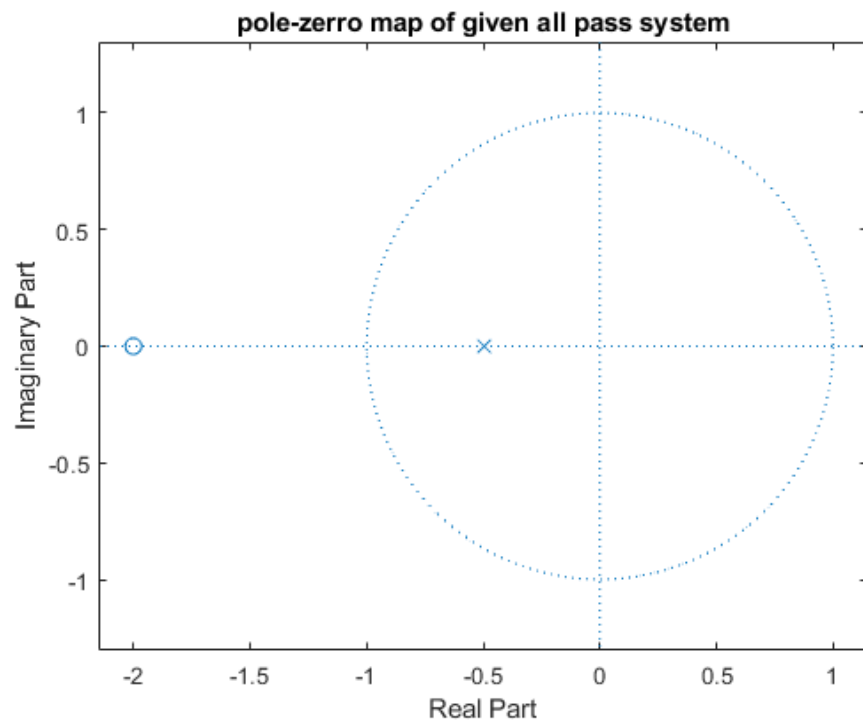
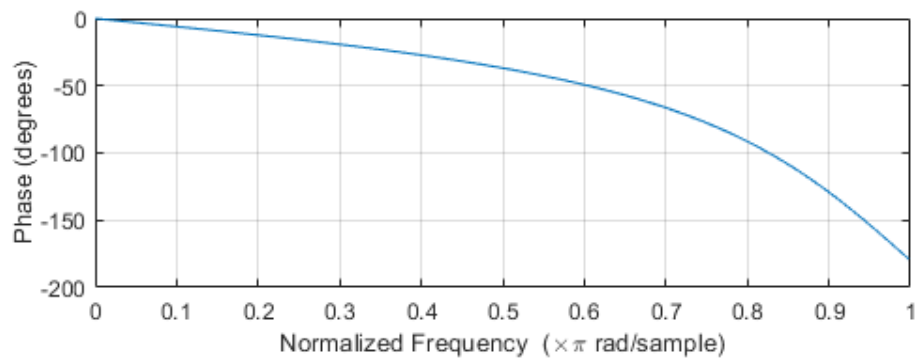
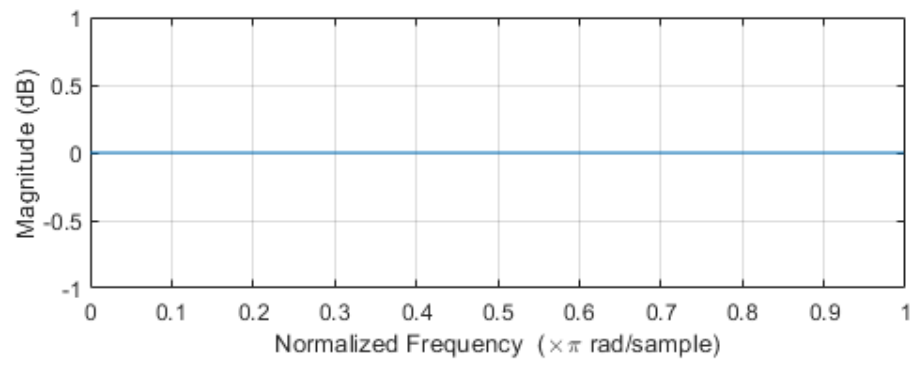
Freqz % freqz(b,a,n) returns the n-point frequency response vector h and the corresponding angular frequency vector w for the digital filter with transfer function coefficients stored in b and a.

tf2zpk % tf2zpk(b,a) finds the matrix of zeros z, the vector of poles p, and the associated vector of gains k from the transfer function parameters b and a

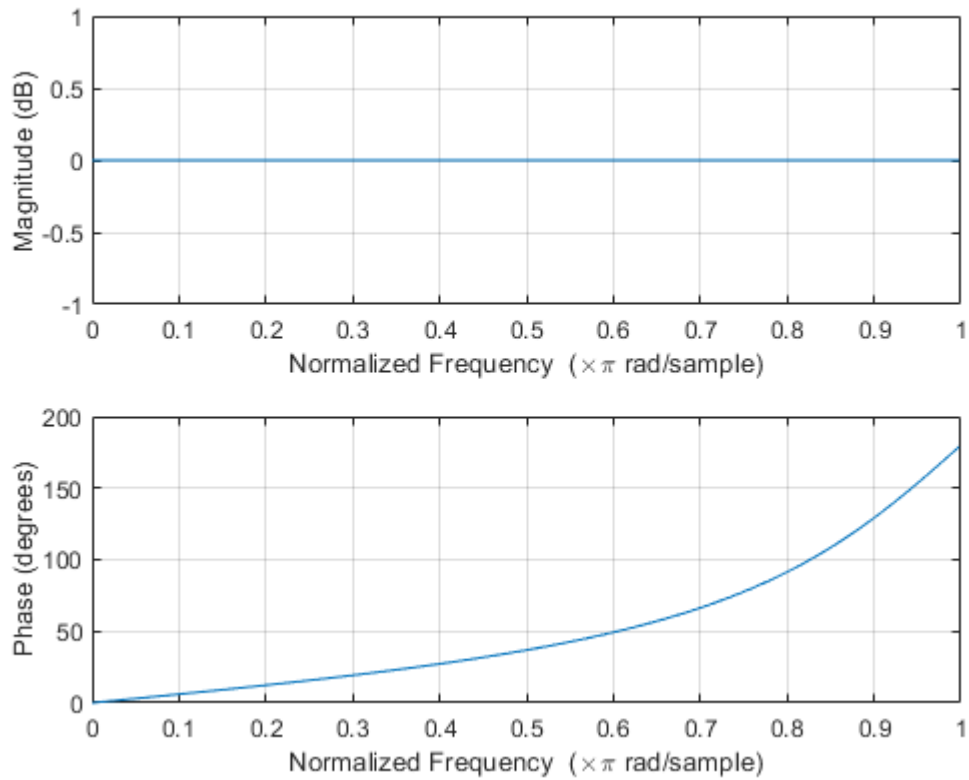
zp2tf % zp2tf(z,p,k) converts a factored transfer function representation

Plots:

$$\text{magnitude response of } H(z) = \frac{0.5 + z^{-1}}{1 + 0.5 z^{-1}}$$



frequency response of $H_2(z) = \frac{2 + z^{-1}}{1 + 2z^{-1}}$



Inferences/comments:

- 1) If the causal all pass system has a one pole at $z = -0.5$ then another zero at conjugate reciprocal location I,e, $z = -2$.
- 2) The magnitude response of all pass system is constant for all frequencies.
- 3) And the equivalent system for $H(z)$ has the same magnitude response as $H(z)$.

Q3. Repeat Q2 for a system $H_2(z)$ which has only one pole at $z = -2$

AIM: To design the all pass system that is having one real pole at $z=-2$ and plot the magnitude and phase response using freqz command.

Short Theory:

Consider an LSI system with transfer function

$$H(z) = (z^{-1} - a^*) / (1 - az^{-1})$$

Note that the system above has a pole at $z = a$ and a zero at $z = (1/a)^*$. This means that if the pole is located at $z = re^{j\theta}$, the zero would be located at $z = (1/re^{-j\theta})$. If $h[n]$ is real, the poles and zeros would each occur in complex-conjugate pairs. We note that even though the magnitude is constant, the phase does depend on frequency.

This property implies that you can invert the magnitude of a pole or zero location in an LSI system without changing the magnitude of the transfer function of that system.

Key Commands:

zp2tf % zp2tf(z,p,k) converts a factored transfer function representation

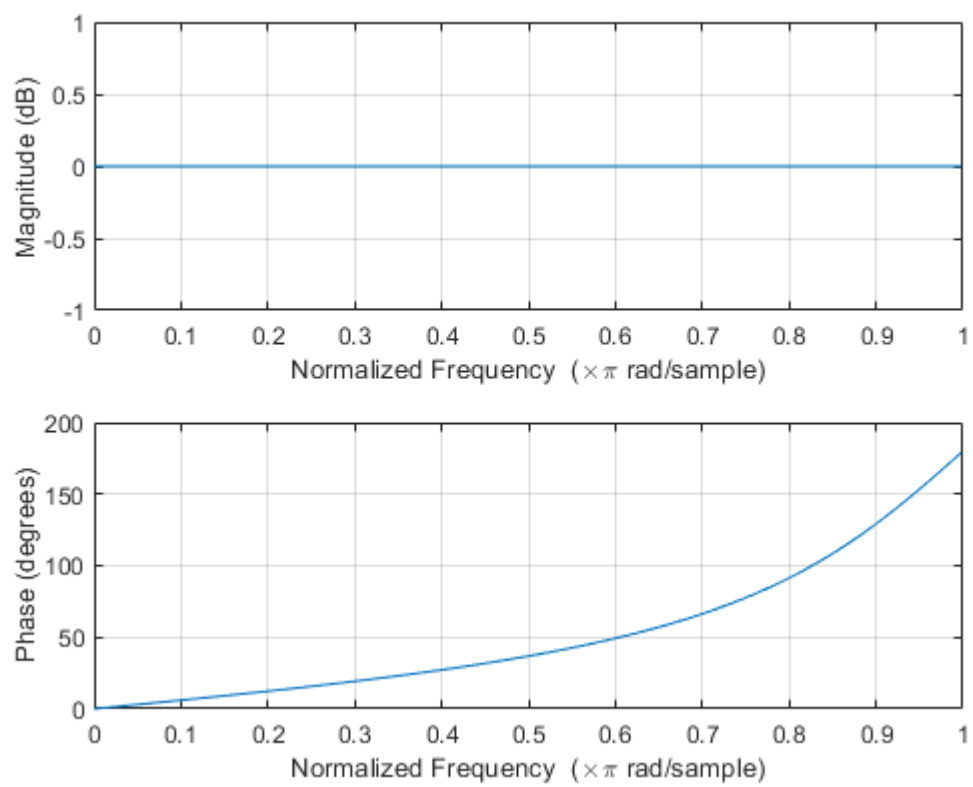
Freqz % freqz(b,a,n) returns the n-point frequency response vector h and the corresponding angular frequency vector w for the digital filter with transfer function coefficients stored in b and a.

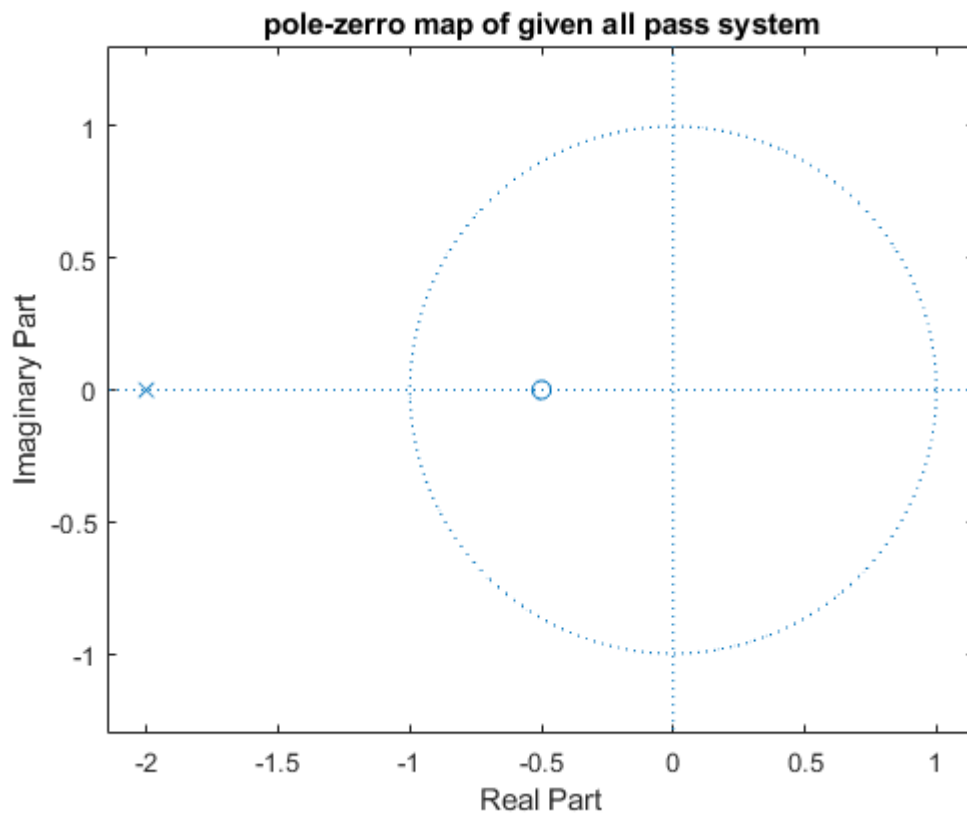
Filt % filt(num,den) creates a discrete-time transfer function sys with numerator(s) num and denominator(s) den

zplane % zplane(z,p) plots the zeros specified in column vector z and the poles specified in column vector p in the current figure window.

Plots:

frequency response of $H_2(z) = \frac{2 + z^{-1}}{1 + 2z^{-1}}$





Inferences/comments:

- 1) If the causal all pass system has a one pole at $z = -2$ then another zero at conjugate reciprocal location I.e, $z = -0.5$.
- 2) The magnitude response of all pass system is constant for all frequencies.
- 3) And the equivalent system for $H(z)$ has the same magnitude response as $H(z)$.

Q4. Repeat Q2 for a system $H_3(z)$ which is a cascade of $H_1(z)$ and $H_2(z)$ system defined earlier. comment on your observations.

AIM: To design the $H_3(z)$ transfer function from the above generated transfer functions $H_1(z)$ and $H_2(z)$ i.e, $H_3(z)=H_1(z) * H_2(z)$.

Short Theory:

Consider an LSI system with transfer function

$$H(z) = (z^{-1} - a^*) / (1 - az^{-1})$$

Note that the system above has a pole at $z = a$ and a zero at $z = (1/a)^*$. This means that if the pole is located at $z = re^{j\theta}$, the zero would be located at $z = (1/re^{-j\theta})$. If $h[n]$ is real, the poles and zeros would each occur in complex-conjugate pairs. We note that even though the magnitude is constant, the phase does depend on frequency.

This property implies that you can invert the magnitude of a pole or zero location in an LSI system without changing the magnitude of the transfer function of that system.

Key Commands:

`zp2tf` % `zp2tf(z,p,k)` converts a factored transfer function representation

`Freqz` % `freqz(b,a,n)` returns the n-point frequency response vector h and the corresponding angular frequency vector w for the digital filter with transfer function coefficients stored in b and a .

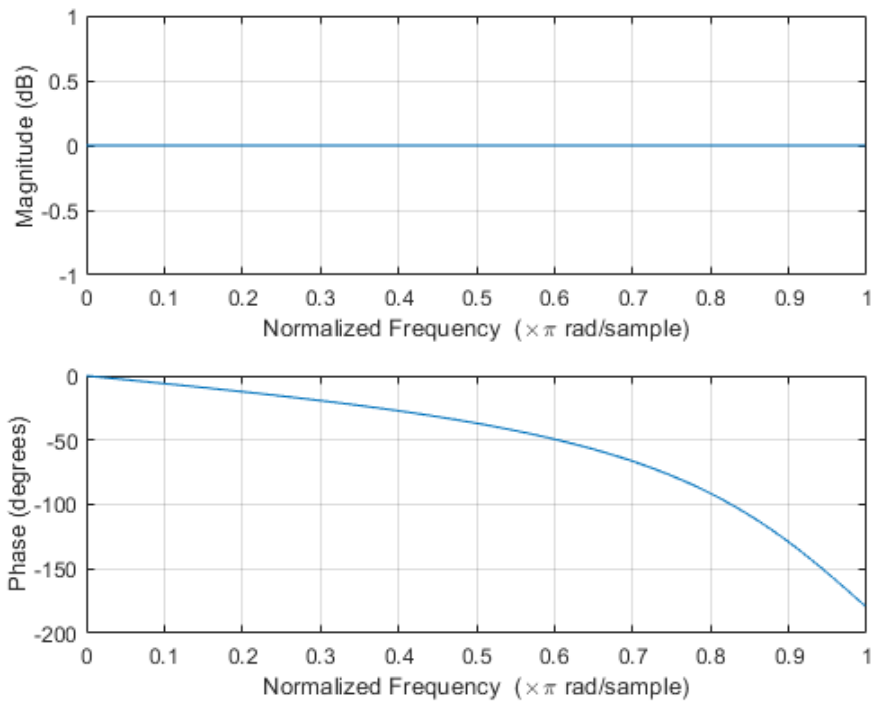
`Filt` % `filt(num,den)` creates a discrete-time transfer function sys with numerator(s) num and denominator(s) den

Plots:

$$0.5 + z^{-1}$$

frequency response of $H_1(z) = \frac{0.5 + z^{-1}}{1 + 0.5 z^{-1}}$

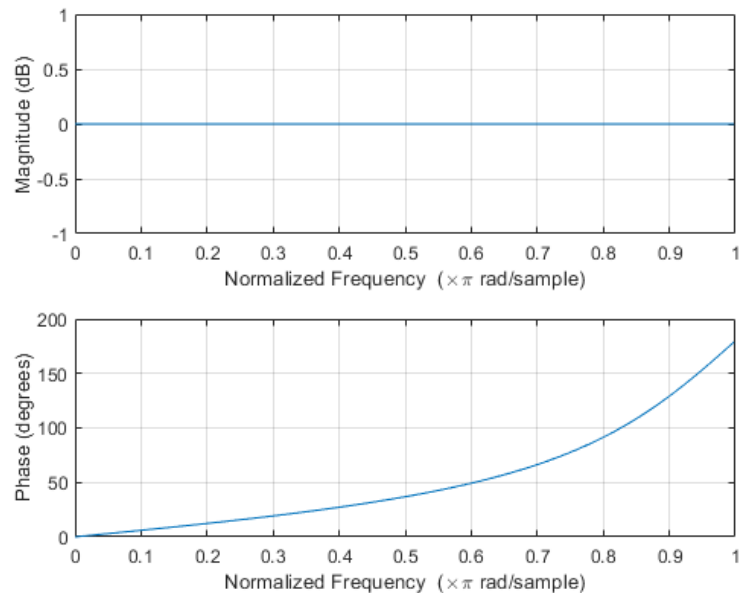
$$1 + 0.5 z^{-1}$$



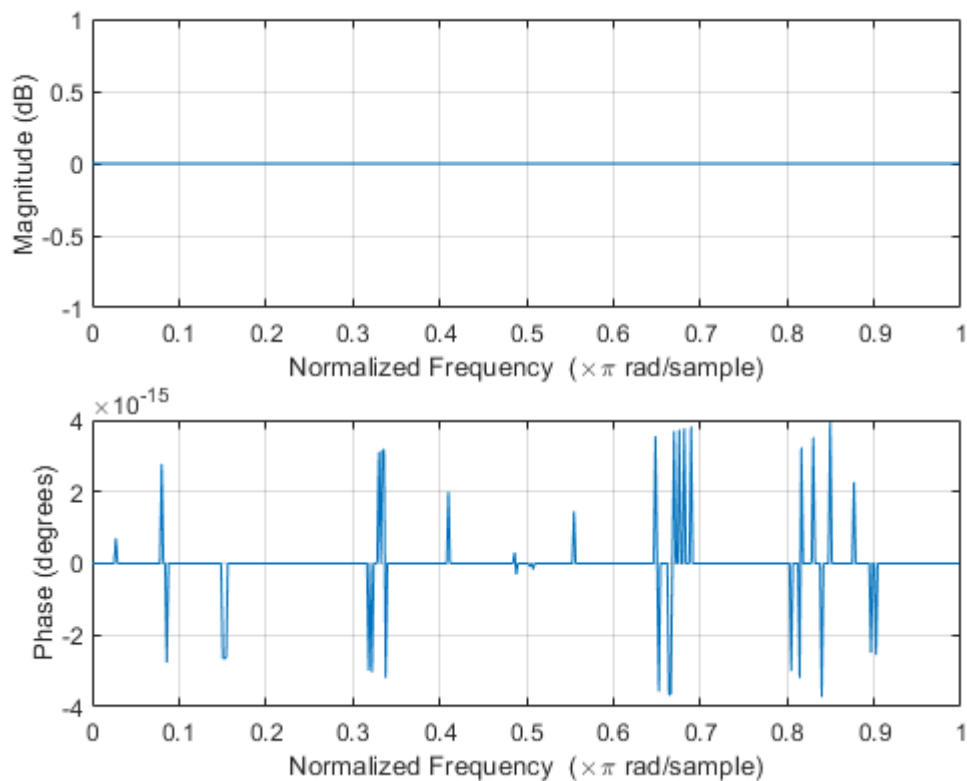
$$2 + z^{-1}$$

magnitude response of $h_2(z) = \frac{2 + z^{-1}}{1 + 2 z^{-1}}$

$$1 + 2 z^{-1}$$



$$\text{magnitude response of } H(z) = \frac{1 + 2.5 z^{-1} + z^{-2}}{1 + 2.5 z^{-1} + z^{-2}}$$



Inferences/comments:

- 1) If you cascade the 2 all pass systems, then the result is also a all pass system.
- 2) If you cascade the 2 all pass systems, the resulting all pass system order is increases.

Q5. Let $H(z) = (1-2z^{-1})/(1+1/3z^{-1})$. Express as a concatenation of Minimum phase & All-pass system.

AIM: To express the given $H(z)$ in concatenation of minimum phase and All-pass system.

Short Theory:

Minimum Phase Systems

A system function $H(z)$ is said to be a minimum phase system if all of its poles and zeros are within the unit circle.

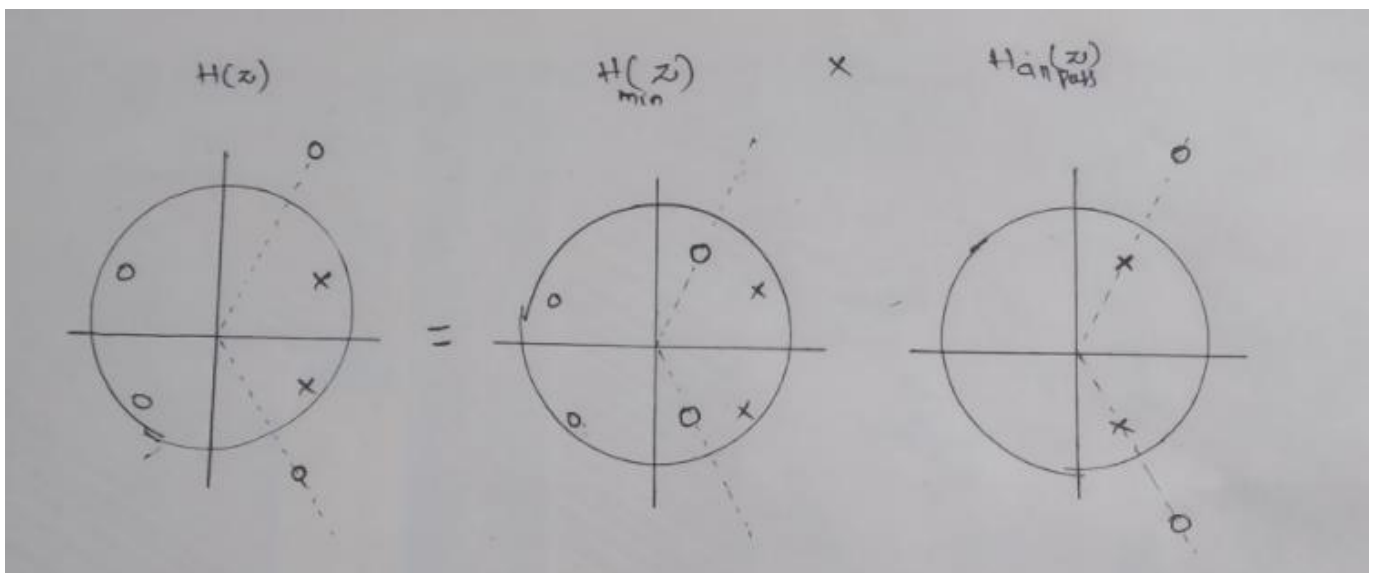
Allpass systems

An All-pass system is a system whose frequency response magnitude is constant for all frequencies, i.e.,

$$|H(e^{j\omega})| = \text{constant}, \omega \in [-\pi, \pi].$$

Consider a stable and causal LTI system with system function $H(z)$. Since the system is a stable system the poles of the system function are required to be inside the unit circle. The zeroes are however, free to wander outside. In case in our $H(z)$ any pole is outside the unit circle also we can represent the $H(z)$ into $H_{\min}(z) * H_{\text{ap}}(z)$ As shown bellow.

For example,



This implies that system function $H(z)$ can then be factorized into two parts as:

$$H(z) = H_{\min}(z) * H_{\text{ap}}(z)$$

Key Commands:

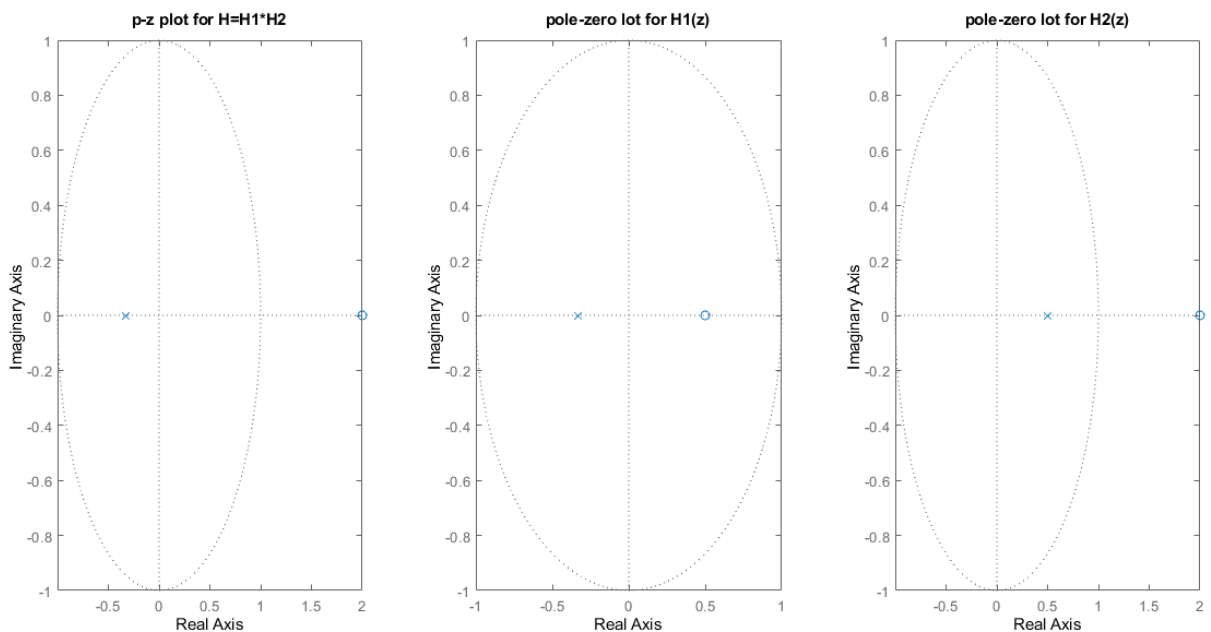
zp2tf % zp2tf(z,p,k) converts a factored transfer function representation

Freqz % freqz(b,a,n) returns the n-point frequency response vector h and the corresponding angular frequency vector w for the digital filter with transfer function coefficients stored in b and a.

Filt % filt(num,den) creates a discrete-time transfer function sys with numerator(s) num and denominator(s) den

minreal % minreal(sys) eliminates uncontrollable or unobservable state in state-space models, or cancels pole-zero pairs in transfer functions or zero-pole-gain models

Plots:



Inferences/comments:

- 1) Non minimum phase system are hard to realizable .If the given $H(z)$ is not a minimum phase system , to make it realizable , we convert given $H(z)$ into concatenation of minimum phase system and all pass system.
- 2) Given $H(z)$ is non minimum phase because it has a zero at $z=2$. So we can write $H(z)$ as

$$H(z) = \underbrace{\frac{z^{-1} - 2}{1 + \frac{1}{3}z^{-1}}}_{H_{\min}(z)} \cdot \underbrace{\frac{1 - 2z^{-1}}{z^{-1} - 2}}_{H_{\text{ap}}(z)}$$