

Elements of Hypothesis Testing - 1

Prof. C. Emre Koksal

ECE 7001: Stochastic Processes, Detection,
and Estimation



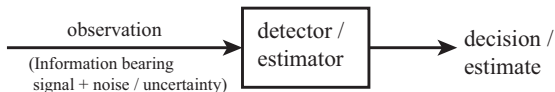
THE OHIO STATE UNIVERSITY

Outline

- 1 Introduction to Detection and Estimation, and Some Notations
- 2 Binary Hypothesis Testing
- 3 Bayesian Hypothesis Testing
- 4 Maximum A Posteriori (MAP) decision rule
- 5 Minimax Decision Rule

Introduction to Detection and Estimation, and Some Notations

Detection vs. Estimation



- **detection**: outcome is decision \in discrete set
- **estimator**: outcome is estimate \in continuous set

Detection Examples:

1 Digital Communication

$$\text{observe } y(t) = \begin{cases} x(t) + N(t), & '0' \text{ is sent} \\ -x(t) + N(t), & '1' \text{ is sent} \end{cases}, 0 \leq t \leq T$$

and decide whether "0" or "1" bit is transmitted.

2 Radar

$$\text{observe } y(t) = \begin{cases} \nu \sin(\omega_c(t - \tau) + \phi) + N(t), & \text{signal is present} \\ N(t), & \text{otherwise} \end{cases}$$

$\{\nu, \tau, \phi\}$ are known, i.e. decide whether an aircraft is present or not.

Estimation Examples:

1 Pulse Amplitude Modulation:

$$\text{observe } y(t) = Ab(t) \cos(\omega_C t) + N(t)$$

estimate $A \in \mathbb{R}$

2 Radar:

observe $y(t) = \nu \sin(\omega_c(t - \tau) + \phi) + N(t)$

estimate $\{\nu, \tau, \phi\}$

Notation:

- observation: y or \vec{y} or $y(t) \in \Gamma$
- outcome: $\theta \in \Lambda$: object of interest (decision or estimate), e.g., in binary hypothesis testing, $\Lambda = \{H_0, H_1\}$

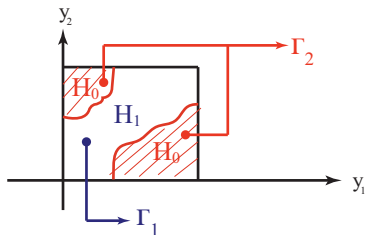
Binary Hypothesis Testing

- Two choices: $|\Lambda| = 2 \Rightarrow H_0$ (Null hypothesis) or H_1 (alternate hypothesis)
 - Observe: $\vec{Y} = \vec{y} \in \Gamma \subset \mathbb{R}^n$
-
- 1 **Bayes**: Apriori probabilities for H_0 , H_1 , costs for decision are known
 - 2 **Minimax**: Unknown priors - Minimum cost for the worst case priors
 - 3 **Neyman-Pearson**: Tradeoff between errors of different type
 - 4 **Composite**: Unknown parameters in the system

Bayesian Hypothesis Testing

Components of the system:

- Apriori probabilities: $\Pi_0 = P(H = H_0)$, $\Pi_1 = P(H = H_1) = 1 - \Pi_0$
- Measurement Model: $p(\vec{y}|H_0)$, $p(\vec{y}|H_1)$
- Decision Rule: $\delta(\vec{y}) : \Gamma \rightarrow \Lambda = \{H_0, H_1\}$

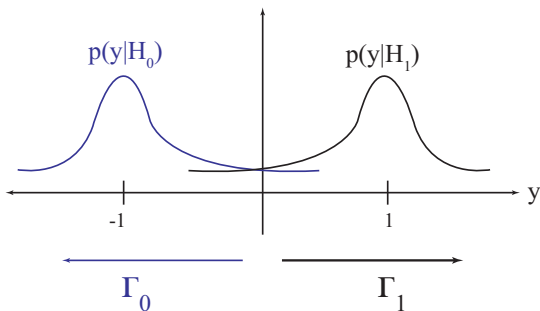


where $\Gamma_0 \cap \Gamma_1 = \emptyset$, $\Gamma_0 \cup \Gamma_1 = \Gamma$

Example:

Measurement model: $H_0 : Y \sim \mathcal{N}(-1, 1)$, $H_1 : Y \sim \mathcal{N}(1, 1)$

priors: $\Pi_0 = \Pi_1 = \frac{1}{2}$, $\Gamma = \mathbb{R}$, $\Lambda = \{H_0, H_1\}$



(may not be the best decision rule)

- Types of errors:

$\delta(y) = 0$ given H_1 (misdetection)

$\delta(y) = 1$ given H_0 (false alarm)

- Cost of decisions:

C_{ij} is cost of choosing H_i given H_j . In general, $C_{ii} < C_{ij}$, $i \neq j$. In radar

$C_{01} \gg C_{10}$.

- Conditional Risk:

The conditional expected cost of decision $\delta(\cdot)$ given H_j :

$$R_j(f) = \sum_i C_{ij} P_j(\Gamma_i)$$

where

$$P_j(\Gamma_i) = P(\vec{Y} \in \Gamma_i | H_j) = \int_{\Gamma_i} p(\vec{y} | H_j) d\vec{y} = \int_{\Gamma_i} \frac{P(H_j | \vec{Y} = \vec{y}) p(\vec{y})}{\Pi_j} d\vec{y}$$

- Bayes' risk:

$$r(\delta) = \sum_j \Pi_j R_j(\delta) = \sum_{i,j} C_{i,j} \int_{\Gamma_i} P(H_j | \vec{Y} = \vec{y}) p(\vec{y}) d\vec{y}$$

- Optimum decision rule: Minimize the Bayes' risk

$$\delta_{opt}(\cdot) = \arg \min_{\delta(\cdot)} r(\delta)$$

where

$$r(\delta) = \underbrace{\int_{\Gamma_0} \sum_j C_{0j} P(H_j | \vec{Y} = \vec{y}) p(\vec{y}) d\vec{y}}_* + \underbrace{\int_{\Gamma_1} \sum_j C_{1j} P(H_j | \vec{Y} = \vec{y}) p(\vec{y}) d\vec{y}}_{**}$$

objectives:

- For any given $\vec{y} \in \Gamma$, decide whether $\vec{y} \in \Gamma_0$ or $y \in \Gamma_1$ to minimize $r(\delta)$
- For any given \vec{y} , if $*$ $<$ $**$, choose $\vec{y} \in \Gamma_0$, if $**$ $<$ $*$, choose $\vec{y} \in \Gamma_1$

•

$$C_{00} P(H = H_0 | \vec{Y} = \vec{y}) + C_{01} P(H = H_1 | \vec{Y} = \vec{y}) \begin{matrix} \stackrel{\delta_{opt}(\vec{y})=1}{\geq} \\ \stackrel{\delta_{opt}(\vec{y})=0}{\leq} \end{matrix}$$

$$C_{10} P(H = H_0 | \vec{Y} = \vec{y}) + C_{11} P(H = H_1 | \vec{Y} = \vec{y})$$

•

$$\frac{P(H = H_1 | \vec{Y} = \vec{y})}{P(H = H_0 | \vec{Y} = \vec{y})} \begin{matrix} \stackrel{1}{\geq} \\ \stackrel{0}{\leq} \end{matrix} \frac{C_{10} - C_{00}}{C_{01} - C_{11}}$$

- Since $P(H = H_i | \vec{Y} = \vec{y}) = \frac{p(\vec{y}|H_i)\Pi_i}{p(\vec{y})}$
 \Rightarrow **Likelihood Ratio Test (LRT)**

$$\frac{p(\vec{y}|H_1)}{p(\vec{y}|H_0)} \underset{0}{\overset{1}{\geq}} \frac{\Pi_0(C_{10} - C_{00})}{\Pi_1(C_{01} - C_{11})}$$

where $\frac{p(\vec{y}|H_1)}{p(\vec{y}|H_0)}$ is the likelihood ratio, $L(\vec{y})$

Notes:

- $L(\vec{y})$ is a non-negative random variable
- $L(\vec{y}) : \mathbb{R}^k \rightarrow \mathbb{R}$ (reduces the dimension of the problem)

Maximum A Posteriori and Maximum Likelihood Detection

Objective: Minimize the decision error

- Consider $C_{00} = C_{11} = 0$, $C_{01} = C_{10} = 1 \rightarrow$ uniform costs
- $r(\delta) = P(\text{decide } H_1 | H = H_0) \Pi_0 + P(\text{decide } H_0 | H = H_1) \Pi_1 =$
probability of decision error $= P(E)$

Maximum A Posteriori (MAP) decision rule: (minimizes the prob. of error)

$$P(H = H_1 | \vec{Y} = \vec{y}) \underset{0}{\overset{1}{\geq}} P(H = H_0 | \vec{Y} = \vec{y})$$

where $P(H = H_1 | \vec{Y} = \vec{y})$ and $P(H = H_0 | \vec{Y} = \vec{y}) \rightarrow$ a posteriori probabilities

Maximum Likelihood (ML) decision rule:

- If $\Pi_0 = \Pi_1 = \frac{1}{2}$,

$$p(\vec{y}|H_1) \underset{0}{\overset{1}{\gtrless}} p(\vec{y}|H_0)$$

Note: If Π_0 and Π_1 are unknown, it may be reasonable to use $\Pi_0 = \Pi_1 = 1/2$ or other ideas (minimize decision rule).

Example: Digital communication system. Decide '0' or '1' based on Y . Under H_0 , $Y \sim \mathbf{N}(-1, 1)$ and H_1 , $Y \sim \mathbf{N}(1, 1)$.

Let $\Pi_0 = \Pi_1 = 1/2$. Find the detector, which minimizes the probability of error and find the associated $P(E)$.

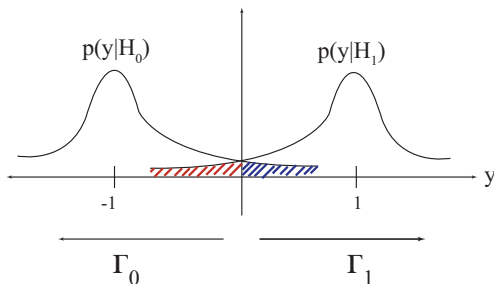
$$L(y) = \frac{p(y|H_1)}{p(y|H_0)} = \frac{\frac{1}{\sqrt{2\pi}} \exp(-\frac{(y-1)^2}{2})}{\frac{1}{\sqrt{2\pi}} \exp(-\frac{(y+1)^2}{2})} = \exp(2y)$$

ML decision:

$$L(y) \underset{0}{\overset{1}{\gtrless}} 1 \Rightarrow \underbrace{\log L(y)}_{\substack{\text{log likelihood} \\ \text{ratio (LLR)}}} \underset{0}{\overset{1}{\gtrless}} 0$$

$\Rightarrow y \underset{0}{\overset{1}{\gtrless}} 0$ where y : sufficient statistics

$$P(E) = r(\delta) = P(\delta(y) = H_1|H_0) \Pi_0 + P(\delta(y) = H_0|H_1) \Pi_1$$



$$\begin{aligned}
 P(E) &= P(Y > 0|H_0) \Pi_0 + P(Y < 0|H_1) \Pi_1 \\
 &= \frac{1}{2} \left(1 - \Phi \left(\frac{0 - (-1)}{\sqrt{1}} \right) \right) + \frac{1}{2} \Phi \left(\frac{0 - 1}{\sqrt{1}} \right) \\
 &\stackrel{a}{=} \Phi(-1)
 \end{aligned}$$

where (a) follows from $\Phi(-x) = 1 - \Phi(x)$

Minimax Decision Rule

Unknown Π_0, Π_1 , known C_{ij} .

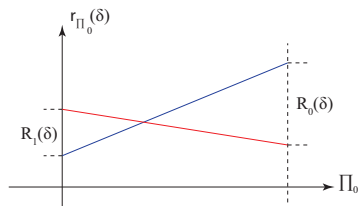
- Let $r_{\Pi_0}(\delta)$ be the expected risk for some (possibly suboptimal) $\delta(\cdot)$ and given the time prior Π_0 ;

Let $\delta_{\Pi_0}(\cdot) = \arg \min_{\delta(\cdot)} r_{\Pi_0}(\delta)$ (optimal decision rule for prior Π_0)

Objective: $\delta_M = \arg \min_{\delta(\cdot)} \left[\max_{\Pi_0} r_{\Pi_0}(\delta) \right]$

•

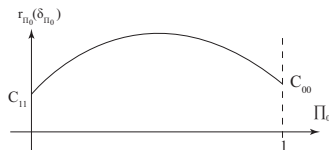
$$\begin{aligned} r_{\Pi_0}(\delta) &= \Pi_0 R_0(\delta) + \Pi_1 R_1(\delta) \\ &= (R_0(\delta) - R_1(\delta)) \Pi_0 + R_1(\delta) \geq r_{\Pi_0}(\delta_{\Pi_0}) \end{aligned}$$



$$\Pi_0 = 0 \Rightarrow \Gamma_0 = \emptyset \Rightarrow \delta_{\Pi_0}(\vec{y}) = 1$$

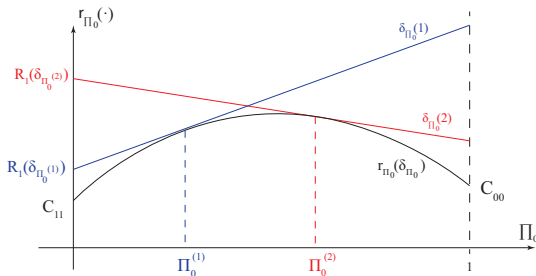
$$r_{\Pi_0}(\delta_{\Pi_0}) = C_{11}$$

Similarly, if $\Pi_1 = 0 \Rightarrow r_{\Pi_0}(\delta_{\Pi_0}) = C_{00}$

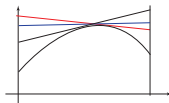


Combining, $r_{\Pi_0}(\delta_{\Pi_0})$ is concave, because

- 1 $r_{\Pi_0}(\delta) \geq r_{\Pi_0}(\delta_{\Pi_0})$ with equality if and only if δ is the decision rule optimized for Π_0 . In that case, line $r_{\Pi_0}(\delta)$ is tangent to $r_{\Pi_0}(\delta_{\Pi_0})$.
- 2 Item (1) is true for all $\Pi_0 \Rightarrow$ Every tangent line is always above $r_{\Pi_0}(\delta_{\Pi_0})$ is concave.

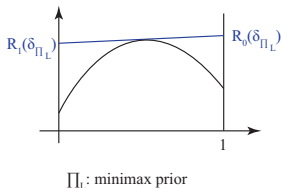


$$\bullet \delta_M = \arg \min_{\delta(\cdot)} \max_{\Pi_0} [r_{\Pi_0}(\delta)]$$

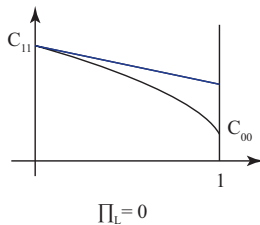


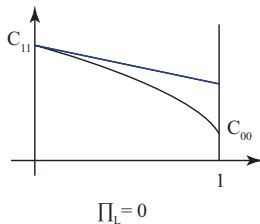
Three cases:

1 .



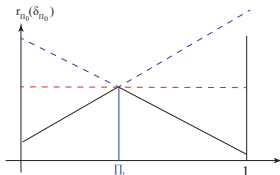
2 .





- δ_M minimizes the "sensitivity" of $r_{\Pi_0}(\delta_M)$ with respect to variations in Π_0 .

Note: Non-strictly concave $r_{\Pi_0}(\delta_{\Pi_0})$:



- $L(\vec{y})$ is a discrete r.v.
- One rule for $\Pi_0 < \Pi_L$ and another for $\Pi_L < \Pi_0$.
- **Minimax:** time share between two rules for the equalizer value.

Example: (In Poor referred to as Location Testing)

$$H_0 : Y \sim \mathcal{N}(\mu_0, \sigma^2), H_1 : Y \sim \mathcal{N}(\mu_1, \sigma^2), \mu_1 > \mu_0.$$

Find δ_M for unit costs ($C_{00} = C_{11} = 0$, $C_{10} = C_{01} = 1$).

The LRT:

$$L(y) = \exp \left(\frac{\mu_1 - \mu_0}{\sigma^2} \left(y - \frac{\mu_0 + \mu_1}{2} \right) \right) \stackrel{1}{\underset{0}{\gtrless}} \tau_L = \frac{\Pi_L}{1 - \Pi_L}$$

The log of both sides:

$$y \stackrel{1}{\underset{0}{\gtrless}} \underbrace{\frac{\mu_0 + \mu_1}{2} + \frac{\sigma^2}{\mu_0 - \mu_1} \log \tau_L}_{y_{\tau_L}}$$

- To find the equalizer rule, solve the equation: $R_1(\delta_L) = R_0(\delta_L)(*)$

$$\begin{aligned} R_j(\delta_L) &= \sum_i C_{ij} P_j(\Gamma_i^{(L)}) \Rightarrow R_1(\delta_L) = P_1(\Gamma_0^{(L)}) = P(Y < y_{\tau_L} | H_1) \\ &= \Phi\left(\frac{y_{\tau_L} - \mu_1}{\sigma}\right) \end{aligned}$$

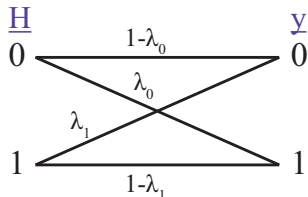
Similarly,

$$R_0(\delta_L) = P_0(\Gamma_1^{(L)}) = P(Y < y_{\tau_L} | H_0) = 1 - \Phi\left(\frac{y_{\tau_L} - \mu_0}{\sigma}\right) = \Phi\left(-\frac{y_{\tau_L} - \mu_0}{\sigma}\right)$$

- Solving (*), we obtain

$$\begin{aligned} \frac{y_{\tau_L} - \mu_1}{\sigma} &= \frac{y_{\tau_L} - \mu_0}{\sigma} \quad (\text{since } \Phi \text{ is one-to-one}) \\ \Rightarrow y_{\tau_L} &= \frac{\mu_1 + \mu_0}{2} \Rightarrow \tau_L = 1 \Rightarrow \Pi_L = \frac{1}{2} \end{aligned}$$

Example: Binary Channel: Simplified view of a communication channel



Assume $\Pi_0 = \Pi_1 = \frac{1}{2}$ and unit costs \Rightarrow ML decision rule

$$L(y) = \frac{P(Y = y|H_1)}{P(Y = y|H_0)} = \begin{cases} \frac{\lambda_1}{1-\lambda_0}, & y = 0 \\ \frac{1-\lambda_1}{\lambda_0}, & y = 1 \end{cases}$$

decision rule: $L(y) \underset{0}{\overset{1}{\gtrless}} 1$

$$\textcircled{i} \quad y = 0, \lambda_1 + \lambda_0 > 1 \Rightarrow \delta(y) = 1$$

$$\textcircled{ii} \quad y = 0, \lambda_1 + \lambda_0 < 1 \Rightarrow \delta(y) = 0$$

$$\textcircled{iii} \quad y = 1, \lambda_1 + \lambda_0 < 1 \Rightarrow \delta(y) = 1$$

$$\textcircled{iv} \quad y = 1, \lambda_1 + \lambda_0 > 1 \Rightarrow \delta(y) = 0$$

$$\delta(y) = \begin{cases} y, & \lambda_0 + \lambda_1 \leq 1 \\ 1 - y, & \lambda_0 + \lambda_1 > 1 \end{cases}$$

- What if $\lambda_0 + \lambda_1 = 1$

\textcircled{v}

$$\begin{aligned} \text{if } \delta(y) = y \Rightarrow r(\delta) &= \frac{1}{2}P_0(\Gamma_1) + \frac{1}{2}P_1(\Gamma_0) \\ &= \frac{1}{2}P(Y = 1|H_0) + \frac{1}{2}P(Y = 0|H_1) \\ &= \frac{1}{2}(\lambda_0 + \lambda_1) = \frac{1}{2} \end{aligned}$$



$$\begin{aligned}
 \text{if } \delta(y) = 1 - y \Rightarrow r(\delta) &= \frac{1}{2}P(Y = 0|H_0) + \frac{1}{2}P(Y = 1|H_1) \\
 &= \frac{1}{2}(1 - \lambda_0 + 1 - \lambda_1) \\
 &= 1 - \frac{1}{2}(\lambda_0 + \lambda_1) = \frac{1}{2}
 \end{aligned}$$

Both decisions are equally good if $\lambda_0 + \lambda_1 = 1$

- Binary symmetric channel: $\lambda_0 = \lambda_1 = \lambda$

$$\Rightarrow \delta(y) = \begin{cases} y, & \lambda \leq \frac{1}{2} \\ 1 - y, & \lambda > \frac{1}{2} \end{cases}$$

Signal detection under additive noise

Measurement model: $H_i : \vec{s}_i + \vec{N} = \vec{Y}$

Q: How do we find $p(\vec{y}|H_i)$, given $p_{\vec{N}}(\vec{n})$?

$$p(\vec{y}|H_i) = P_{\vec{N}}(\vec{y} - \vec{s}_i)$$

$$L(\vec{y}) = \frac{P_{\vec{N}}(\vec{y} - \vec{s}_1)}{P_{\vec{N}}(\vec{y} - \vec{s}_0)} = \frac{P_{N_1, N_2, \dots, N_n}(y_1 - s_{11}, y_2 - s_{12}, \dots, y_n - s_{1n})}{P_{N_1, N_2, \dots, N_n}(y_1 - s_{01}, y_2 - s_{02}, \dots, y_n - s_{0n})}$$

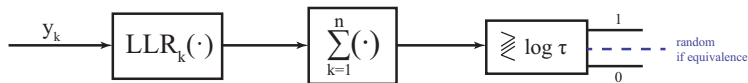
let \vec{N} be iid:

$$L(\vec{y}) = \frac{\prod_{k=1}^n P_{N_k}(y_k - s_{1k})}{\prod_{k=1}^n P_{N_k}(y_k - s_{0k})} = \prod_{k=1}^n L_k(y_k)$$

$$LLR(\vec{y}) \triangleq \log L(\vec{y}) = \underbrace{\sum_{k=1}^n LLR_k(y_k)}_{LRT} \underset{0}{\overset{1}{\gtrless}} \log \tau$$

signal is known

⇒ Coherent detector



Example:

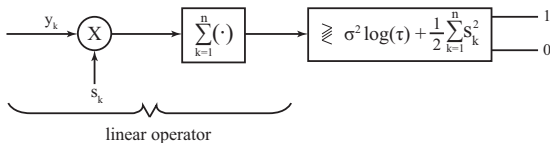
$H_1 : Y_k = s_k + N_k$, $H_0 : Y_k = N_k$, $N_k \sim \mathcal{N}(0, \sigma^2)$, iid

$$L_k(y_k) = \frac{\frac{1}{\sqrt{2\pi}\sigma^2} \exp\left(-\frac{1}{2} \frac{(y_k - s_k)^2}{\sigma^2}\right)}{\frac{1}{\sqrt{2\pi}\sigma^2} \exp\left(-\frac{1}{2} \frac{y_k^2}{\sigma^2}\right)} = \exp\left(-\frac{1}{2\sigma^2}(s_k^2 - 2s_k y_k)\right)$$

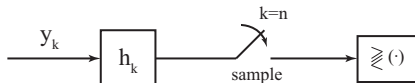
$$\sum_{k=1}^n LLR_k(y_k) = \frac{1}{\sigma^2} \sum_{k=1}^n \left(s_k y_k - \frac{1}{2} s_k^2\right) \stackrel{1}{\geq} \log \tau$$

$$\Rightarrow \underbrace{\sum_{k=1}^n s_k y_k}_{\substack{\text{projection of } \vec{y} \\ \text{on to } \vec{s}}} \underset{0}{\overset{1}{\gtrless}} \sigma^2 \log \tau + \frac{1}{2} \sum_{k=1}^n s_k^2$$

detector: linear correlator:



- The linear operator can be represented with a discrete-time filter, called the matched filter:



$$h_k = \begin{cases} s_{n-k}, & 0 \leq k \leq n-1 \\ 0, & \text{otherwise} \end{cases} \Rightarrow h_n \circledast y_n = \sum_{k=1}^n h_{n-k} y_k = \sum_{k=1}^n s_k y_k$$

read: detection in iid Laplacian noise in Poor.