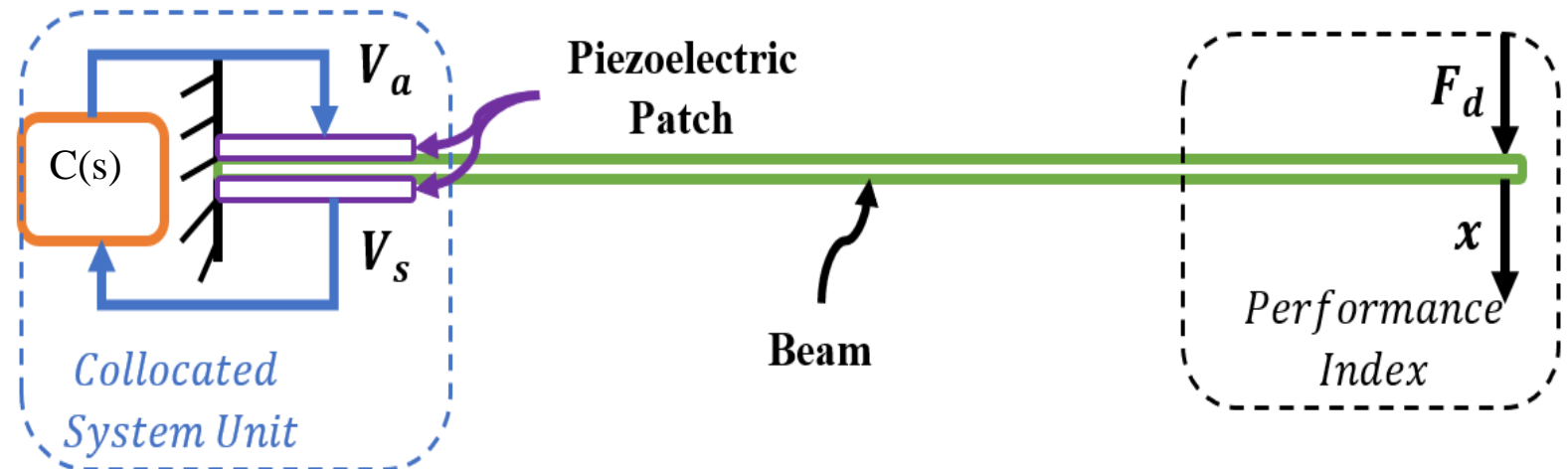
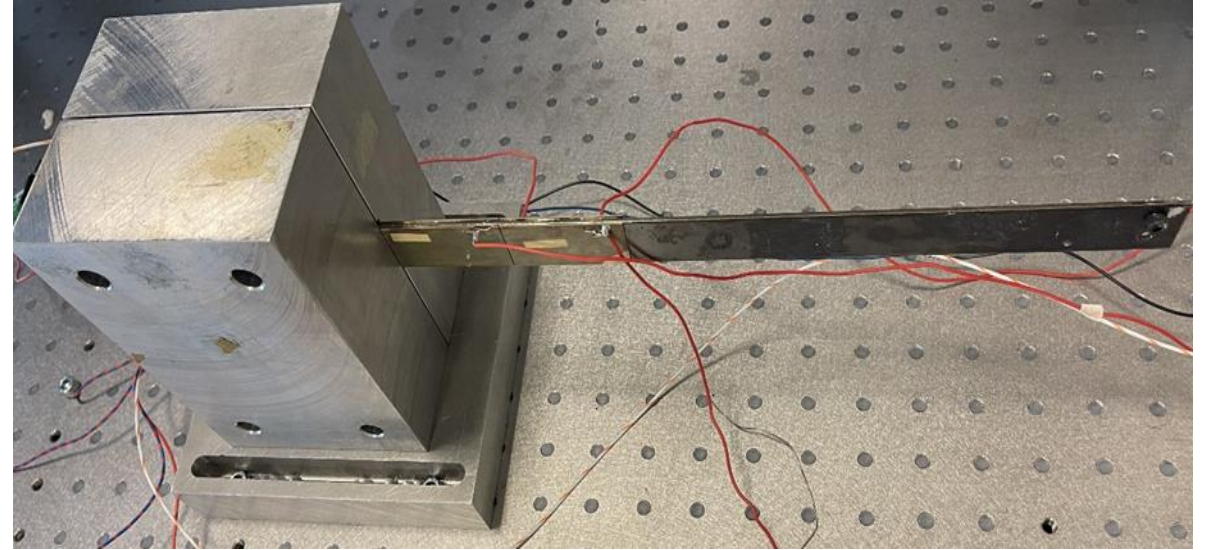


# Active Vibration Control strategies

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### Topics of Seminar:

- 1- Collocated systems
- 2- Positive position feedback (PPF) Controller
- 3- Negative Derivative Feedback Controller
- 4- Application on the beam
- 5- Application on the Bladed Rail



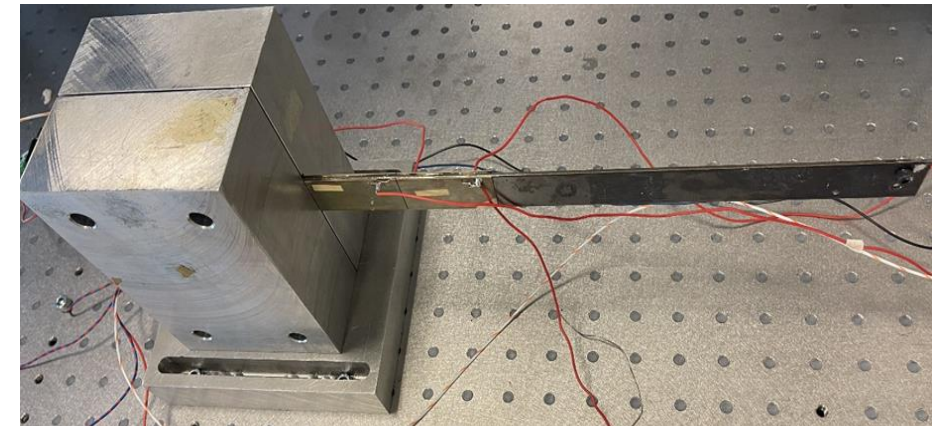
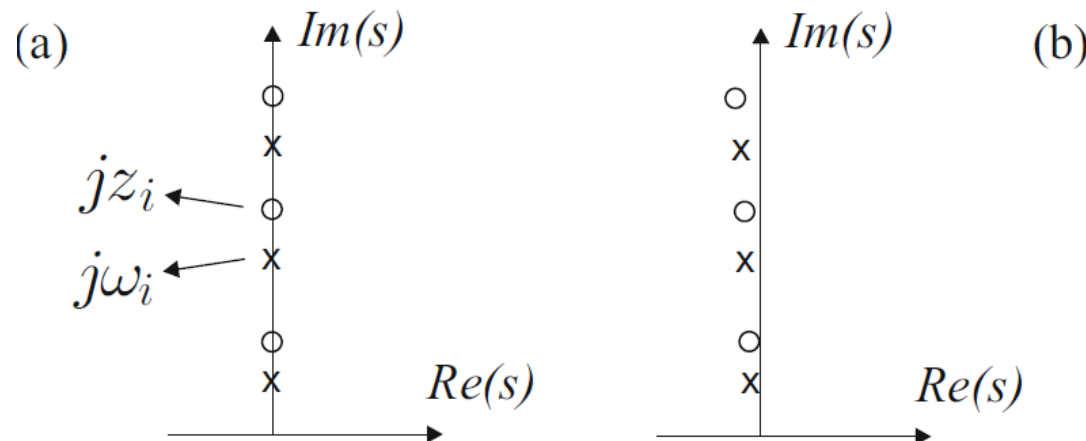
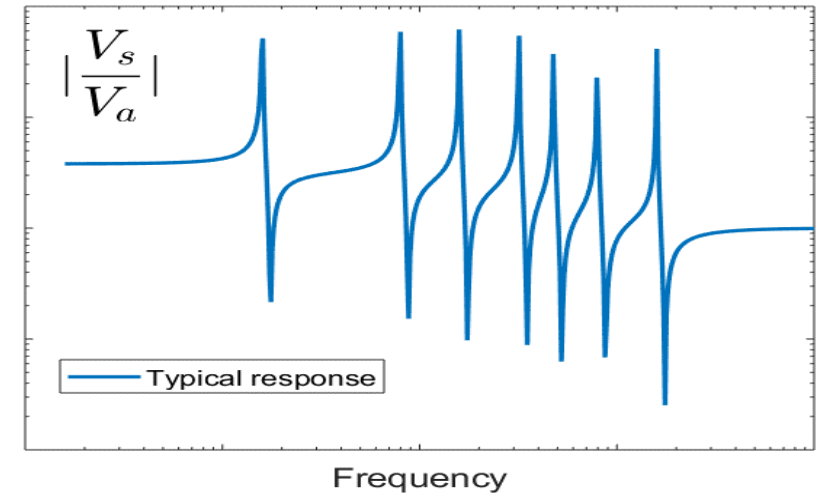
# Collocated systems

Location of Sensor and actuator are considered very close to each other or the same

This creates an alternating poles and zeros in the open loop  
 (Frequency Response from actuator to sensor)

The asymptotic stability of a wide class of single-input single-output (SISO) control systems, even if the system parameters are subject to large perturbations. This is because the root locus plot keeps the same general shape, and remains entirely within the left half plane when the system parameters are changed from their nominal values.

Such a control system is said to be *robust* with respect to stability.

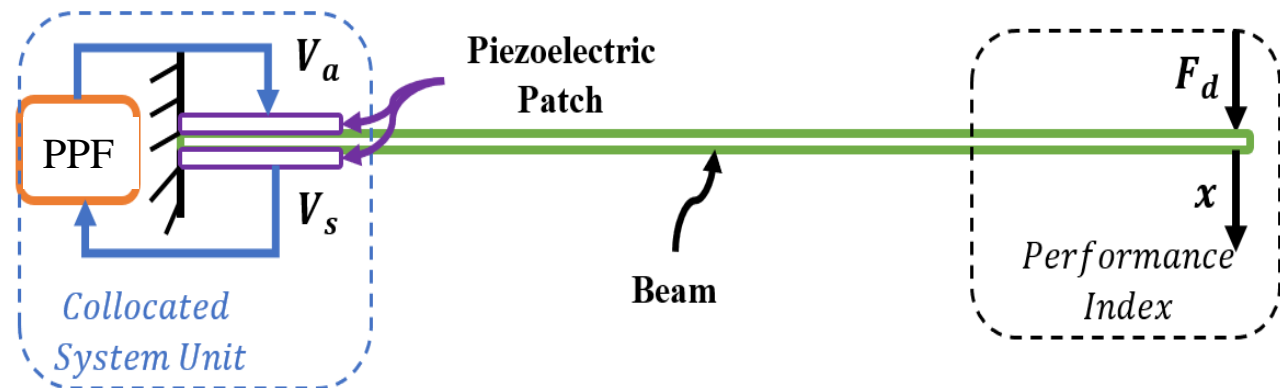
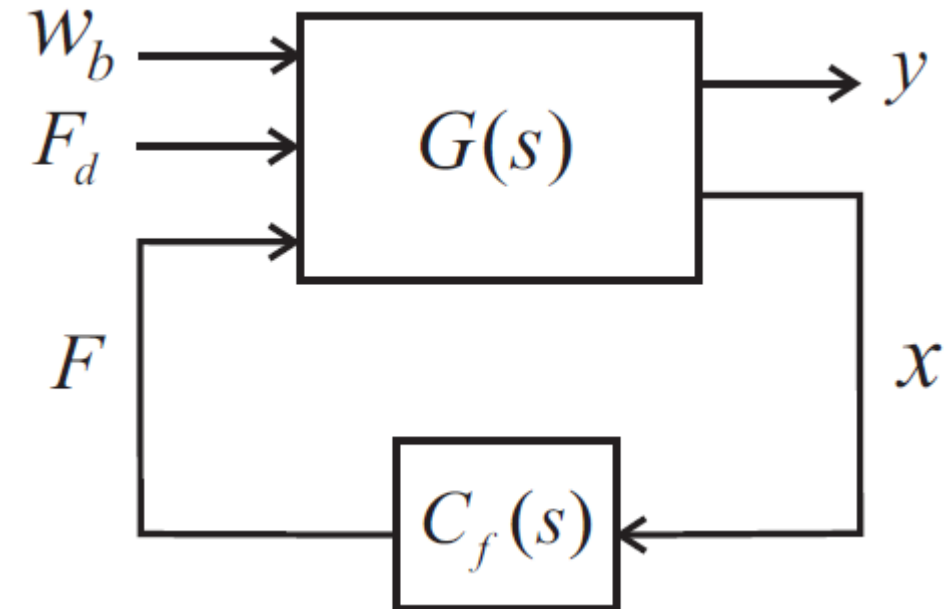
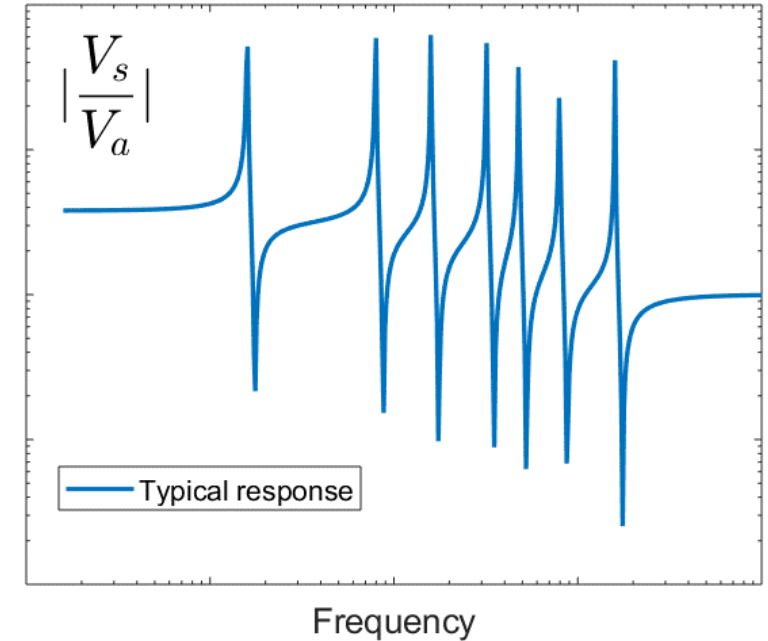
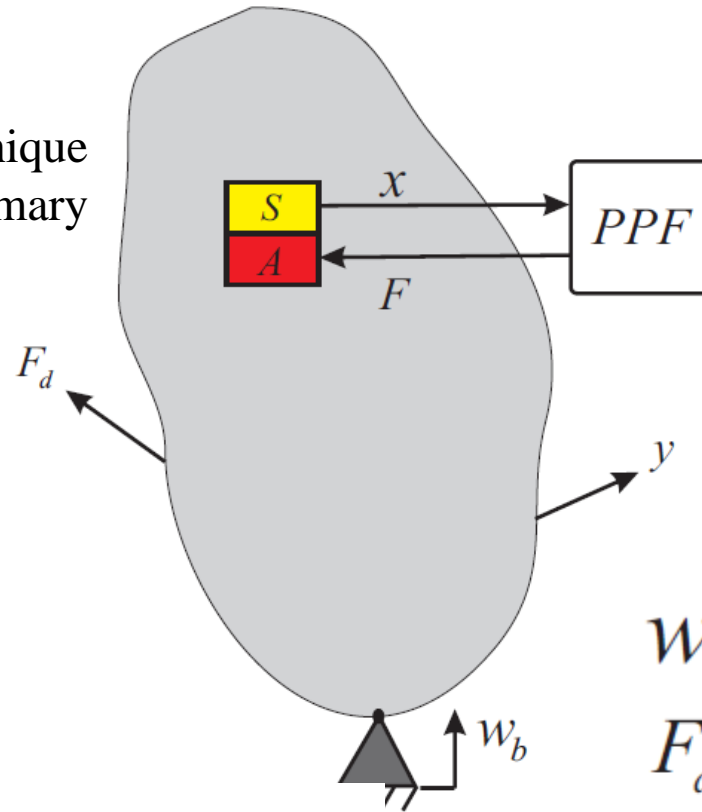


# PPF Controller

PPF is one of the most effective control technique to damp a certain resonance of collocated primary system with multiple modes.

$$G(s) = \frac{V_s}{V_a} = g_0 \frac{s^2 + 2\xi_z \omega_z s + \omega_z^2}{s^2 + 2\xi_p \omega_p s + \omega_p^2}$$

$$C_f(s) = + \frac{g_f}{s^2 + 2\xi_f \omega_f s + \omega_f^2}$$



- *Characteristic Eq. from general case*
- $1 + GC = (s^2 + 2\xi_f\omega_f s + \omega_f^2)(s^2 + \omega_p^2) - g_0g_f(s^2 + \omega_z^2) = s^4 + 2\xi_f\omega_f s^3 + \omega_p^2 + \omega_f^2 - g_0g_f s^2 + 2\xi_f\omega_f\omega_p^2 s + \omega_f^2\omega_p^2 - g_0g_f\omega_z$

### Maximum Damping Method

$$(s^2 + 2\xi_c\omega_c s + \omega_c^2)^2 = s^4 + 4\xi_c\omega_c s^3 + (4\xi_c^2\omega_c^2 + 2\omega_c^2)s^2 + 4\xi_c\omega_c^3 s + \omega_c^4$$

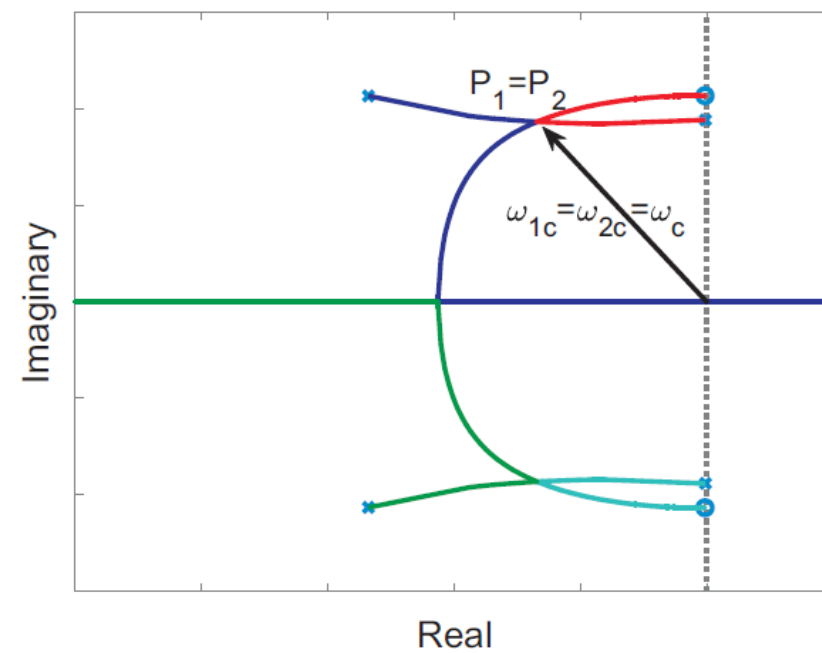
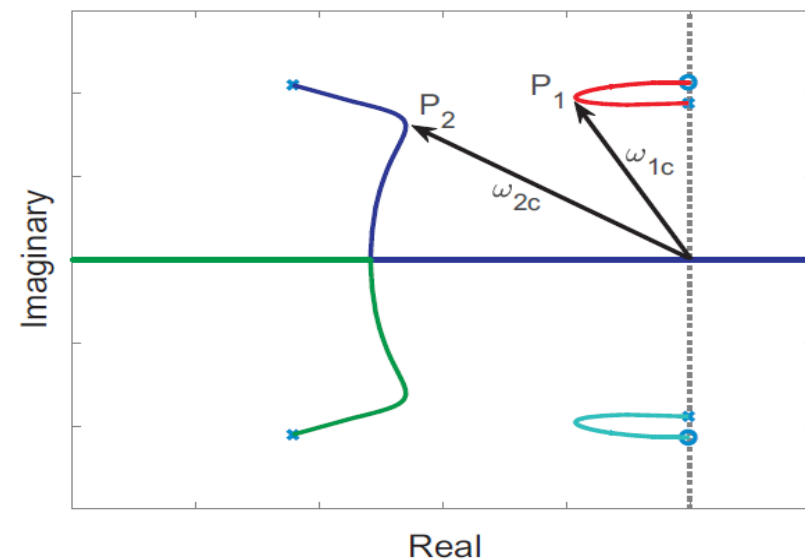
$$\omega_c = \omega_p$$

$$\begin{cases} 4\xi_c\omega_c = 2\xi_f\alpha\omega_p \\ (4\xi_c^2 + 2)\omega_c^2 = (\alpha^2 + 1)\omega_p^2 - \frac{\beta\alpha^2}{\gamma^2}g_0^2\omega_p^2 \\ 4\xi_c\omega_c^3 = 2\xi_f\alpha\omega_p^3 \\ \omega_c^4 = \alpha^2\omega_p^4 - \beta\alpha^2g_0^2\omega_p^4 \end{cases}$$

$$\alpha = \frac{\omega_f}{\omega_p}$$

$$\beta = \frac{g_f}{g_0}$$

$$\gamma = \frac{\omega_z}{\omega_p}$$



## Maximum Damping Method



- $\alpha = \sqrt{\frac{4\gamma^2\xi_c^2}{\gamma^2-1} + 1}$
- $\xi_f = 2\xi_c \sqrt{\frac{\gamma^2-1}{4\gamma^2\xi_c^2+\gamma^2-1}}$
- $\beta = \frac{1}{g_0^2} \frac{4\gamma^2\xi_c^2}{4\gamma^2\xi_c^2+\gamma^2}$

The norm wanted to minimized:  $\frac{G}{1+GC} =$

$$\frac{g_0(s^2 + 2\xi_z\omega_z s + \omega_z^2)}{s^4 + 2\xi_f\omega_f s^3 + \omega_p^2 + \omega_f^2 - g_0g_f s^2 + 2\xi_f\omega_f\omega_p^2 s + \omega_f^2\omega_p^2 - g_0g_f\omega_z}$$

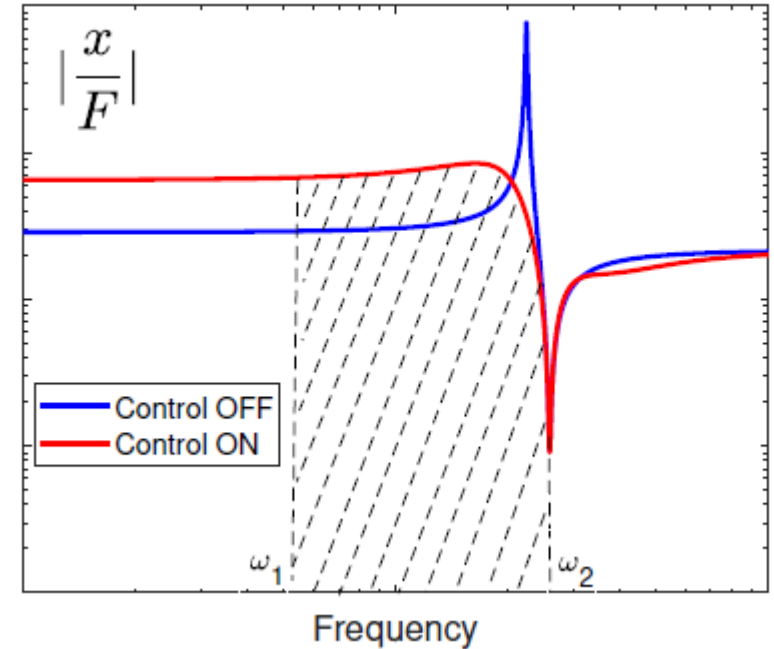
Considering  $s = j\omega$  the value of  $\left| \frac{G}{1+GC} \right|$  can be extracted

Using  $H_2$  and  $H_\infty$  optimization method For Finding an optimal value of closed loop damping

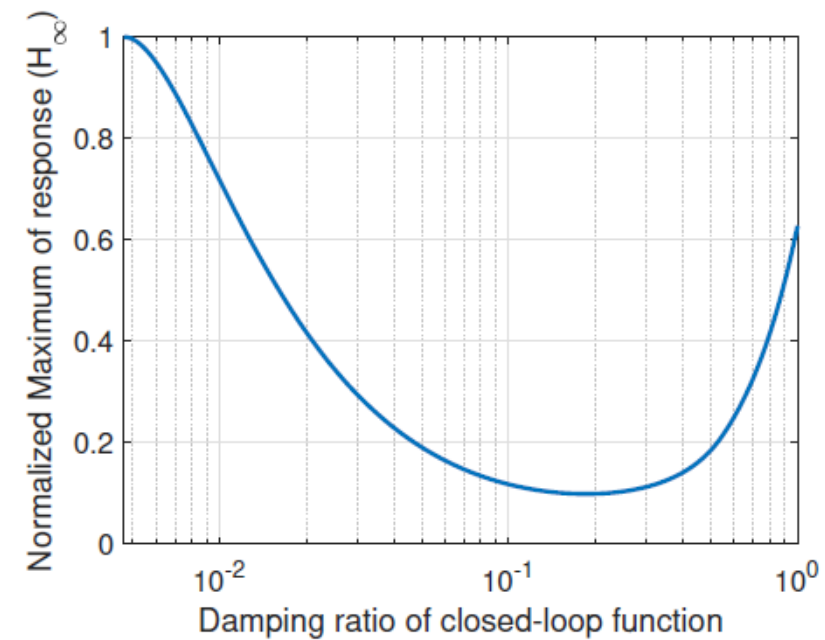
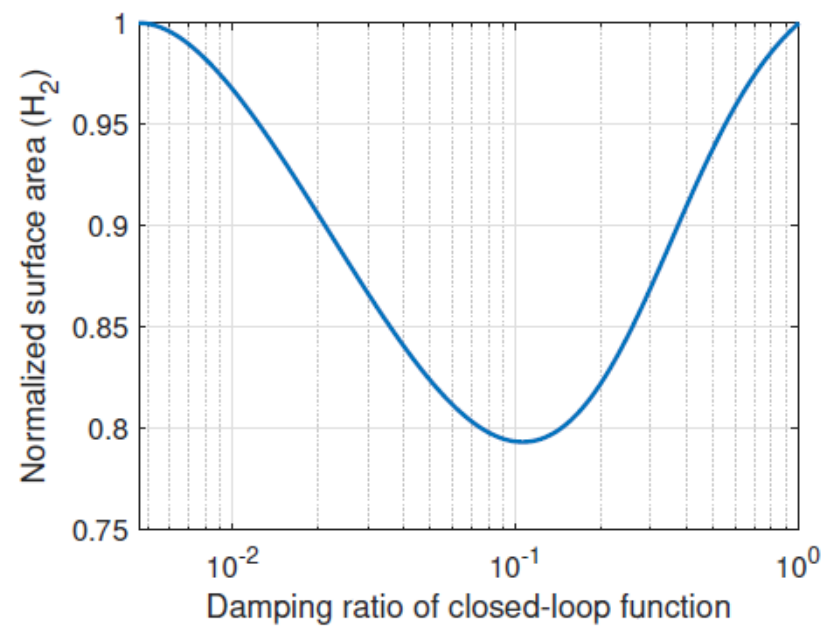
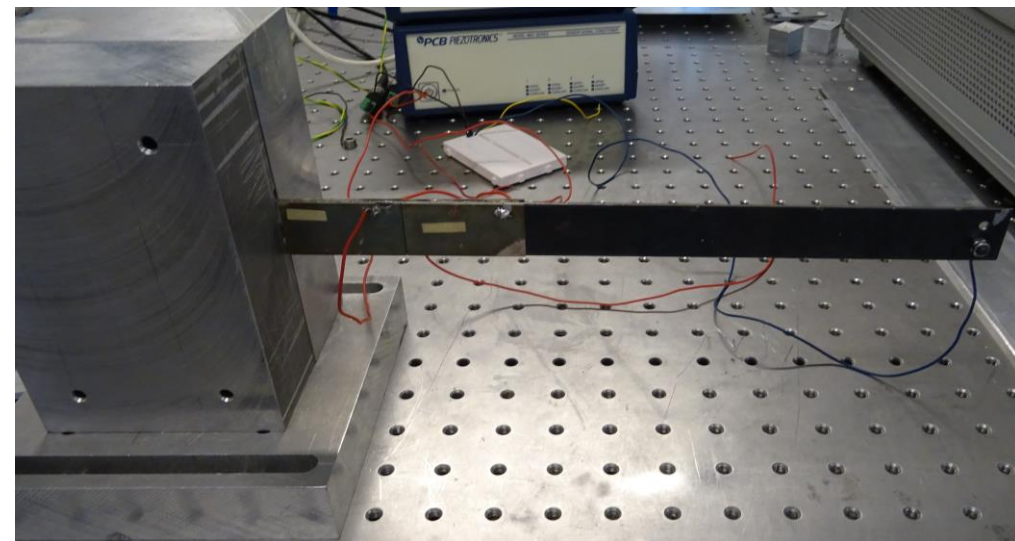
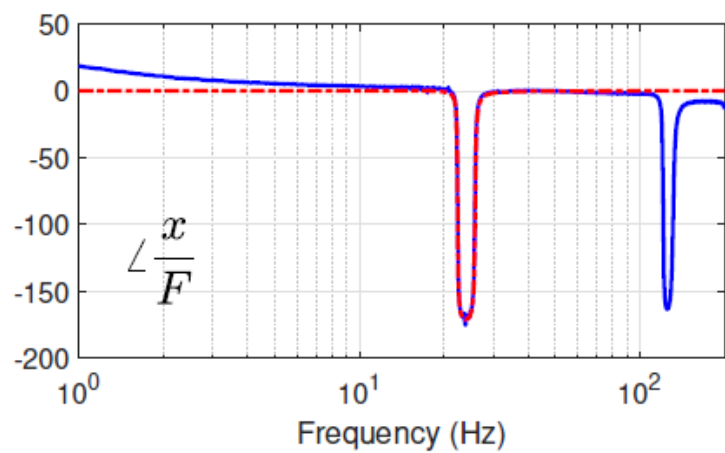
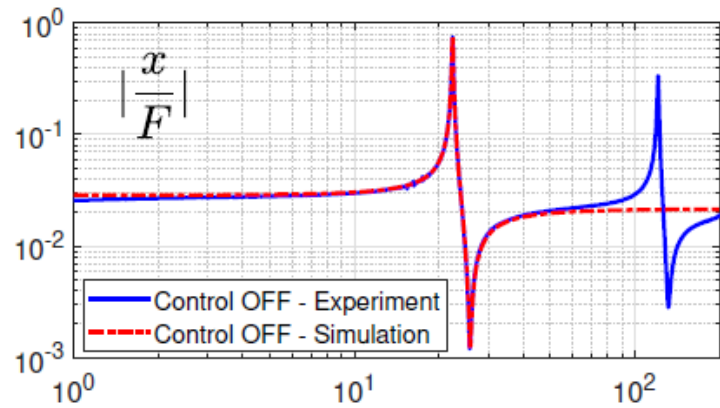
$$H_2 = \int_{\omega_1}^{\omega_2} \left| \frac{G}{1+GC} \right| d\omega$$

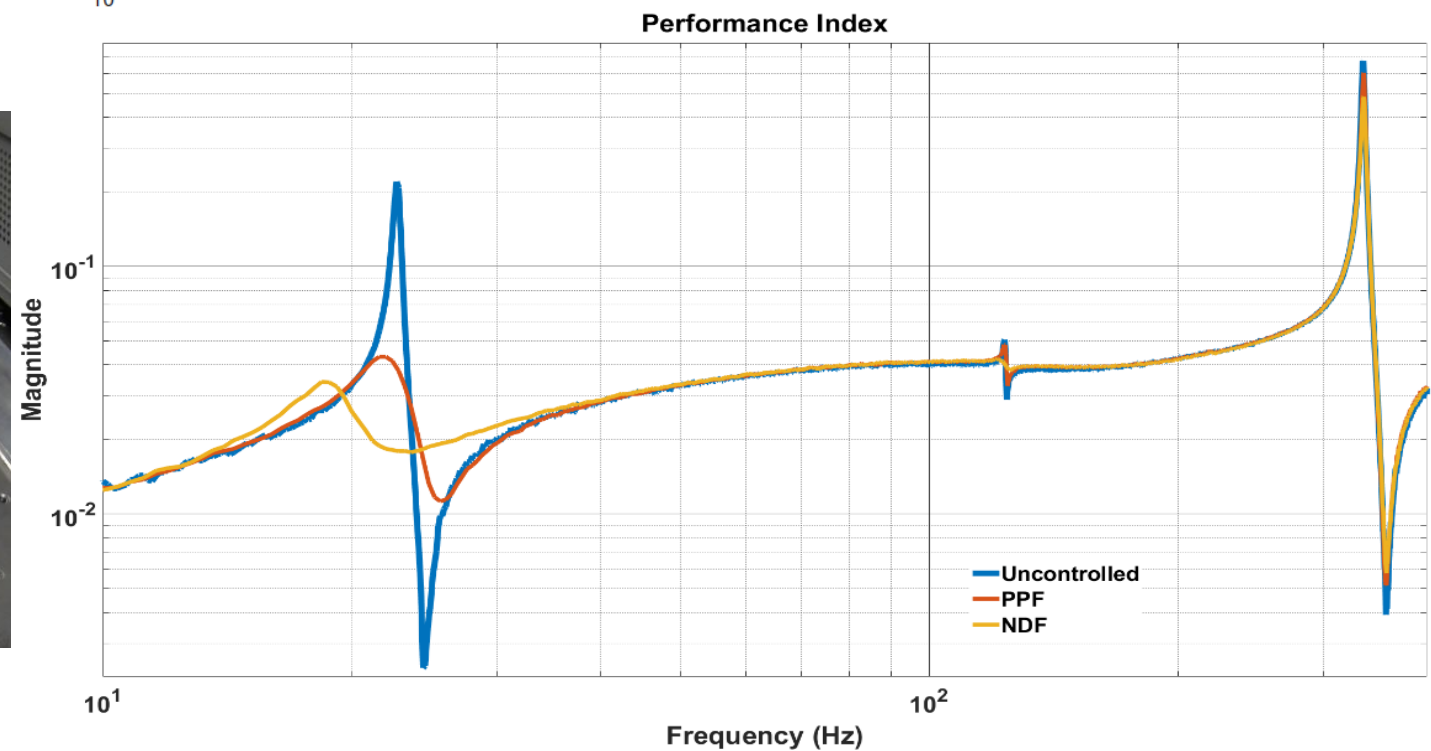
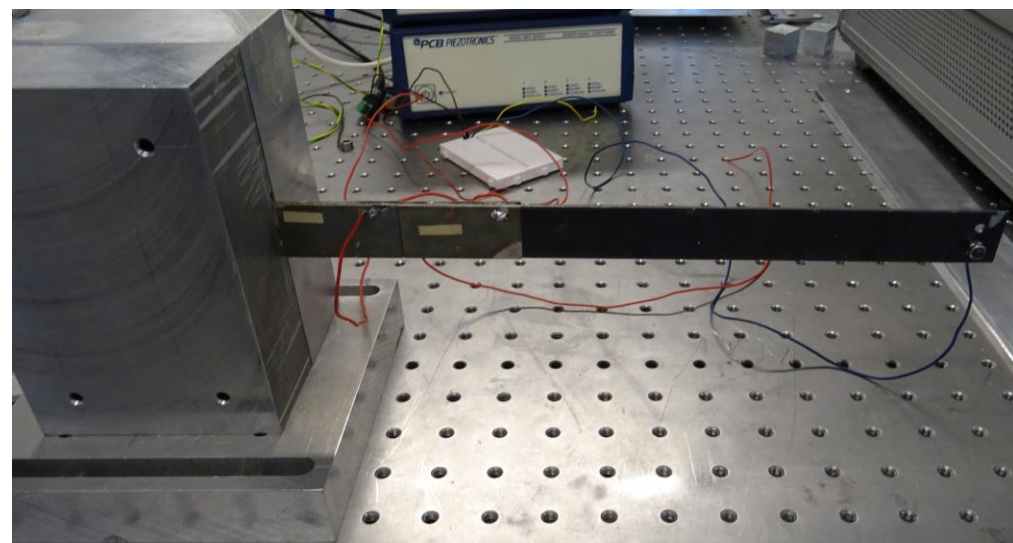
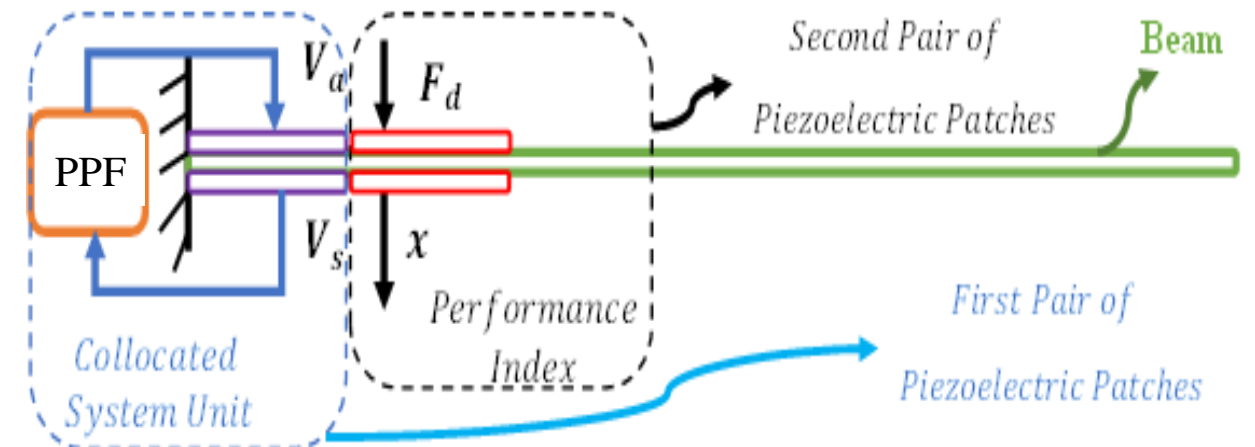
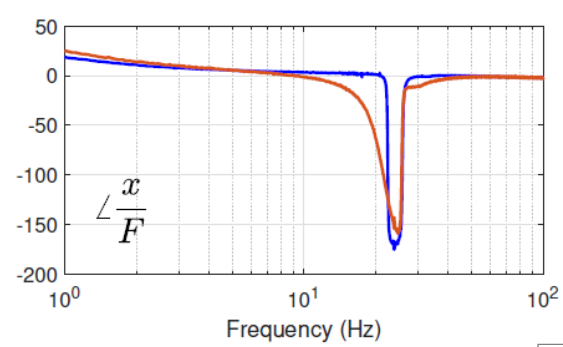
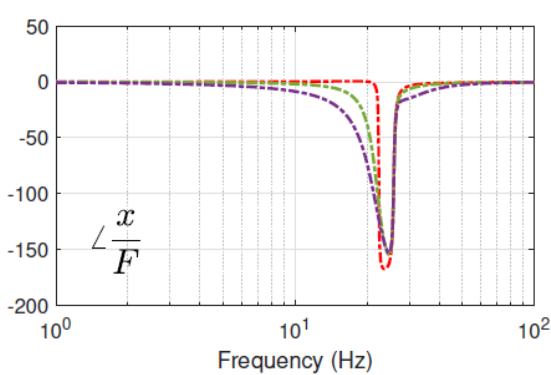
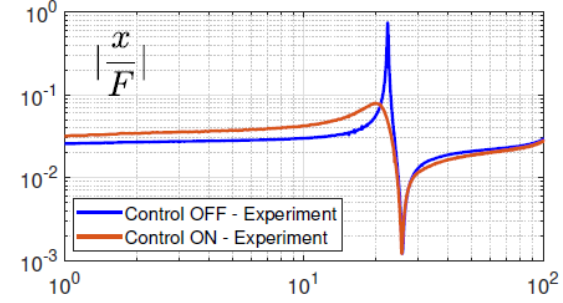
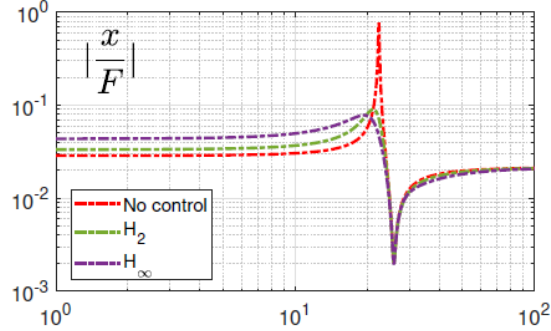
$$H_\infty = \min \left( \left| \frac{G}{1+GC} \right| \right) \Big|_{@w_p}$$

Minimizing the cost function with respect to values of closed loop damping (from zero to 1)

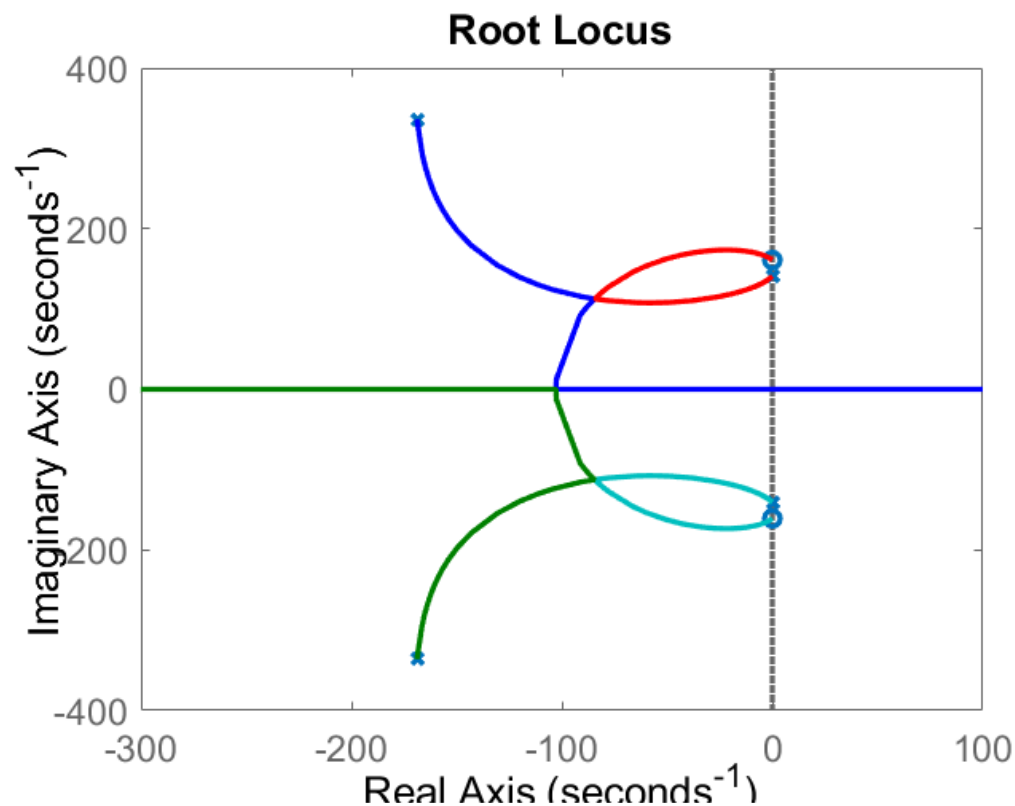




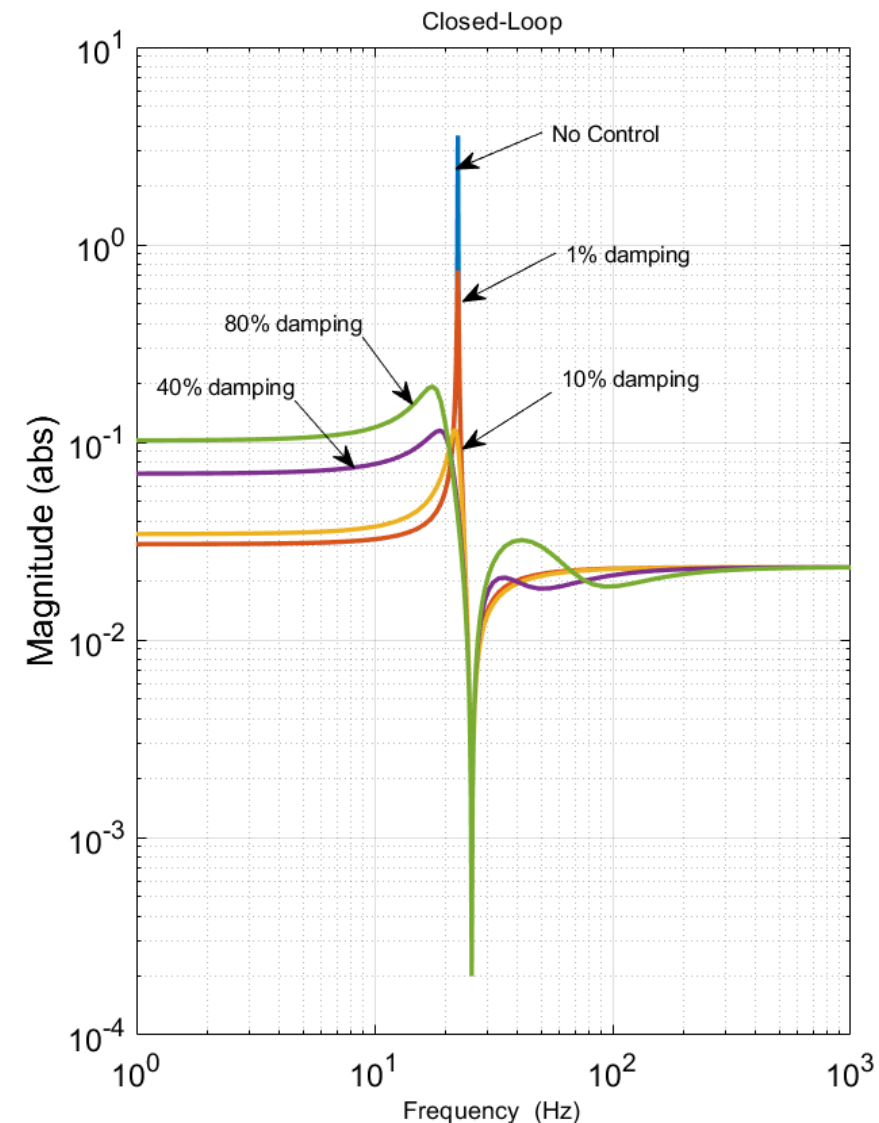








By increasing the value of the feedback gain, one pole of the closed-loop system crosses the imaginary axis and makes the system unstable.



The other trade-off in the system is that the more damping is added to the system, the more amplification of the static response the closed-loop system has.

## PPF Controller

- Maximum damping happens when the two loops in Root-Locus are intersecting
- For each value of closed-loop damping, there is one controller which causes merged poles of closed-loop and subsequently maximum damping
- More damping requires higher control frequency! This results in some problems
- The amplification at low frequency
- Coupling with the next modes (for continuous structure) The control frequency should be tuned far enough from the next resonance frequency
- Although the formula has been developed for undamped primary system, they can be used for lightly damped structure as well
- The biggest issue with this new method is that the presented method is only applicable for pole before zero pattern!

## Why NDF?

1- The NDF compensator is designed to work as a band-pass filter, cutting off the control action far from the natural frequencies associated with the controlled modes and reducing the so-called spillover effect.

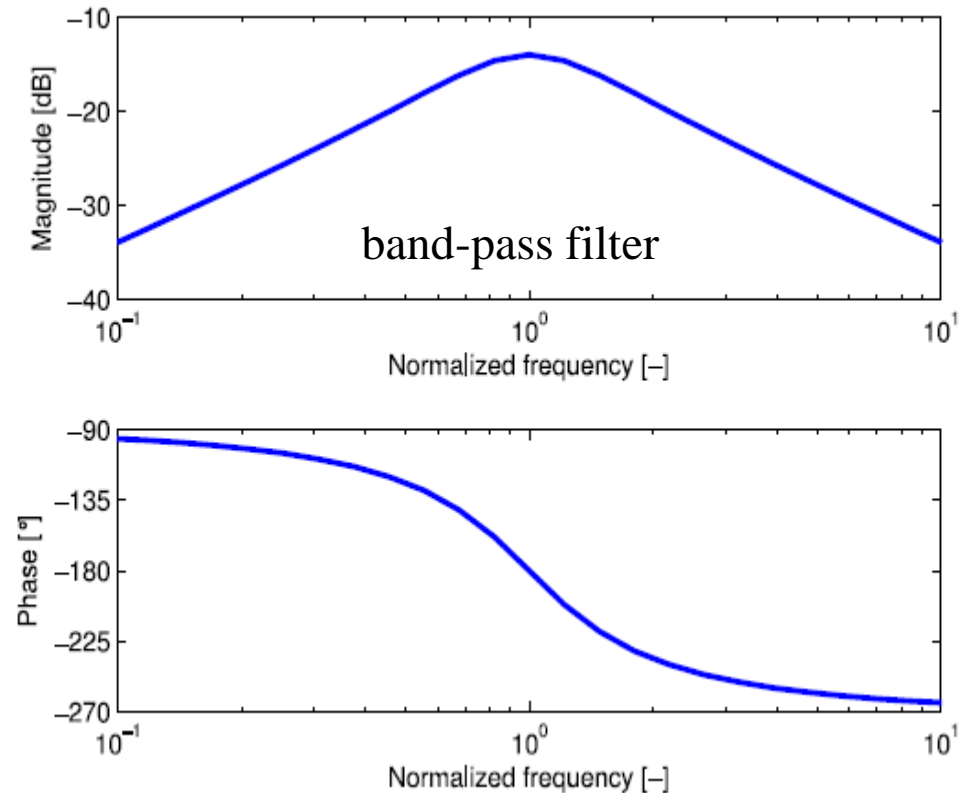
### 2- Comparison with PPF:

As a low-pass filter, PPF is very sensitive to low-frequency disturbances.

To overcome this shortcoming of PPF controller, NDF controller, which acts as a bandpass filter and can effectively control the lower and higher frequency disturbances, has been developed recently.

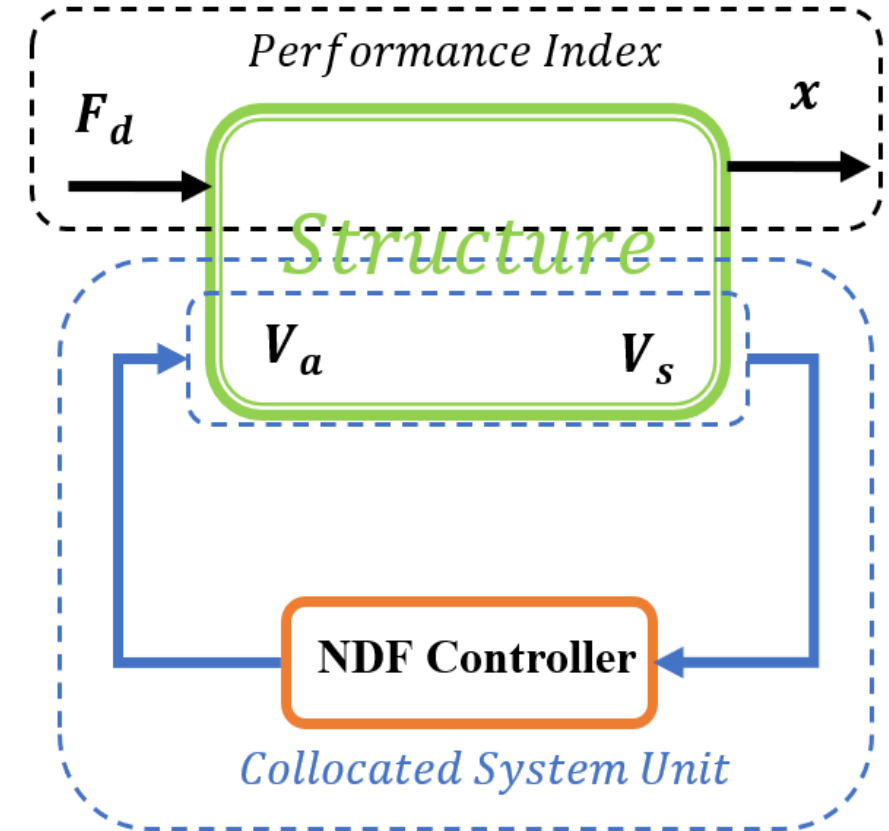
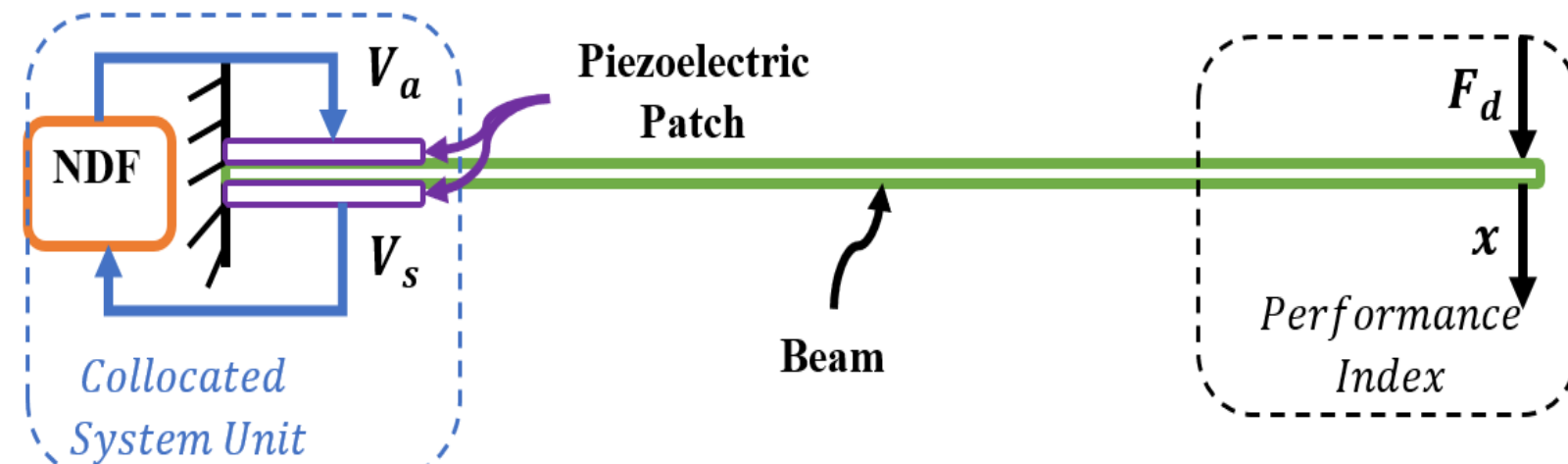
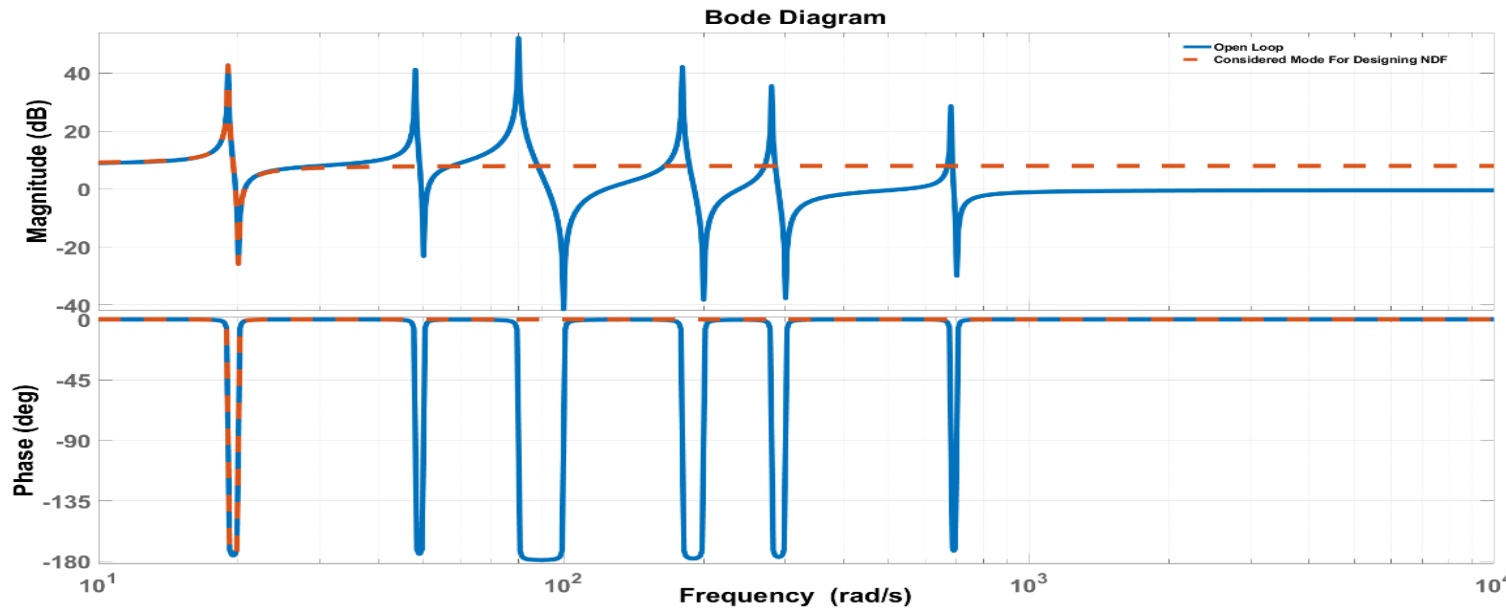
3- Negative derivative feedback (NDF) proves particularly robust against spillover since modal velocity is fed back through a band-pass filter so that undesired effects can be limited both at high and low frequencies.

4- This type of Controller is a ideal controller for low damped structure vibration reduction



$$NDF = - \frac{\alpha \beta s}{s^2 + 2\xi \alpha s + \alpha^2}$$

# Negative derivative feedback controller (NDF)



$$G(s) = \frac{V_s}{V_a} = g_0 \frac{s^2 + 2\xi_z \omega_z s + \omega_z^2}{s^2 + 2\xi_p \omega_p s + \omega_p^2}$$

$$C(s) = -\frac{K_c \omega_c s}{s^2 + 2\xi_c \omega_c s + \omega_c^2}$$

The characteristic equation of the system will be:

$$s^4 + (2\xi_p\omega_p + 2\xi_c\omega_c + k_c\omega_c g_0)s^3 + (\omega_p^2 + \omega_c^2 + 4\xi_p\xi_c\omega_p\omega_c + 2\xi_z\omega_z K_c\omega_c g_0)s^2 + (2\xi_p\omega_p\omega_c^2 + 2\xi_c\omega_c\omega_p^2 + g_0\omega_z^2 K_c\omega_c)s + \omega_p^2\omega_c^2 = 0$$

Considering Maximum damping Method:

$$(s^2 + 2\xi_f\omega_f s + \omega_f^2)^2 =$$

$$s^4 + 4\xi_f\omega_f s^3 + (4\xi_f^2\omega_f^2 + 2\omega_f^2)s^2 + 4\xi_f\omega_f^3 s + \omega_f^4 = 0$$

Considering:

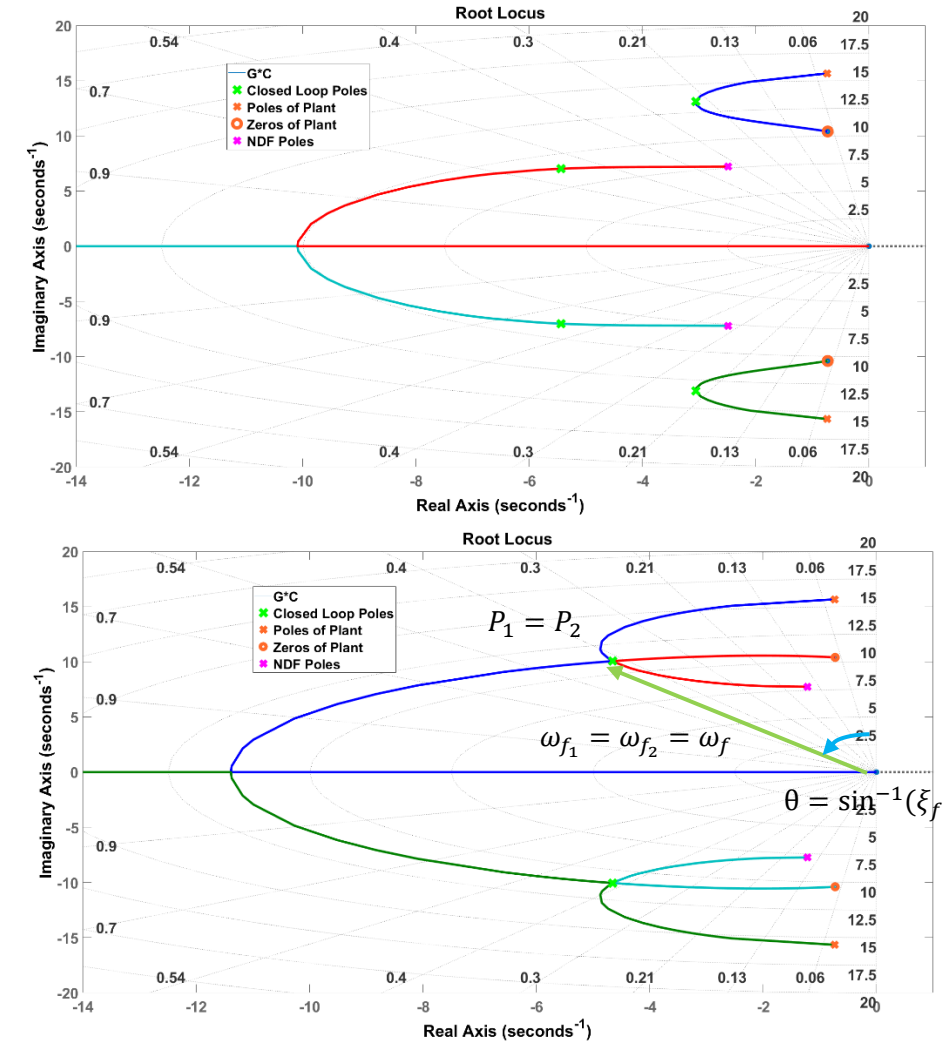
Negative derivative feedback controller works for both of zero before pole and zero after pole!

The pattern dose not matter!

$$\alpha = \frac{\omega_c}{\omega_p}$$

$$\gamma = \frac{\omega_z}{\omega_p}$$

$$\begin{cases} \gamma < 1 & \text{zero before pole} \\ \gamma > 1 & \text{zero after pole} \end{cases}$$





$$\begin{cases} 2\xi_p\omega_p + 2\xi_c\omega_c + k_c\omega_c g_0 = 4\xi_f\omega_f & (1) \\ \omega_p^2 + \omega_c^2 + 4\xi_p\xi_c\omega_p\omega_c + 2\xi_z\omega_z K_c\omega_c g_0 = 4\xi_f^2\omega_f^2 + 2\omega_f^2 & (2) \\ 2\xi_p\omega_p\omega_c^2 + 2\xi_c\omega_c\omega_p^2 + g_0\omega_z^2 K_c\omega_c = 4\xi_f\omega_f^3 & (3) \\ \omega_p^2\omega_c^2 = \omega_f^4 & (4) \end{cases}$$

$$(2) \begin{cases} \alpha < 1 & \text{then } \alpha = (\xi_{eq} + 1) - \sqrt{\xi_{eq}^2 + 2\xi_{eq}} \\ \alpha > 1 & \text{then } \alpha = (\xi_{eq} + 1) + \sqrt{\xi_{eq}^2 + 2\xi_{eq}} \end{cases}$$

### Based on Maximum Damping Method

$\alpha = \frac{\omega_c}{\omega_0}$	$\alpha$	$(\xi_{eq} + 1) \pm \sqrt{\xi_{eq}^2 + 2\xi_{eq}}$ ( $\xi_{eq} = 2\xi_f^2 - 2\xi_p\xi_c - \xi_z\gamma K_c g_0$ )
(Cutoff frequency)		
$\xi = \xi_c$	$\xi_c$	$\frac{(2\xi_f\sqrt{\alpha} - \xi_p)(1 - \gamma^2) - (1 - \alpha)[2\xi_f\sqrt{\alpha} - \xi_p(1 + \alpha)]}{\alpha(1 - \gamma^2)}$
(Damping Ratio)		
$K_c$	$K_c$	$\frac{2}{g_0} \frac{(1 - \alpha)[2\xi_f\sqrt{\alpha} - \xi_p(1 + \alpha)]}{\alpha(1 - \gamma^2)}$
(Gain)		

The norm wanted to minimized:  $\frac{G}{1+GC} =$

$$\frac{g_0(s^2 + 2\xi_z \omega_z s + \omega_z^2)}{s^4 + (2\xi_p \omega_p + 2\xi_c \omega_c + k_c \omega_c g_0)s^3 + (\omega_p^2 + \omega_c^2 + 4\xi_p \xi_c \omega_p \omega_c + 2\xi_z \omega_z K_c \omega_c g_0)s^2 + (2\xi_p \omega_p \omega_c^2 + 2\xi_c \omega_c \omega_p^2 + g_0 \omega_z^2 K_c \omega_c)s + \omega_p^2 \omega_c^2}$$

Considering  $s = j\omega$  we have

$$\left| \frac{G}{1+GC} \right| = \sqrt{\frac{(\omega_z^2 - \omega^2)^2 + (2\xi_z \omega_z \omega)^2}{(\omega^4 - (\omega_p^2 + \omega_c^2 + 2\xi_z \omega_c K_c \omega_c g_0)\omega^2 + \omega_p^2 \omega_c^2)^2 + (-(2\xi_p \omega_p + 2\xi_c \omega_c + g_0 K_c \omega_c)\omega^3 + (2\xi_p \omega_p \omega_c^2 + 2\xi_c \omega_c \omega_p^2 + g_0 \omega_z^2 K_c' \omega_c)\omega)^2}}$$

**Considering  $H_2$  and  $H_\infty$  optimization method**

$$H_2 = \int_0^\infty \sqrt{\frac{(\omega_z^2 - \omega^2)^2 + (2\xi_z \omega_z \omega)^2}{(\omega^4 - (\omega_p^2 + \omega_c^2 + 2\xi_z \omega_c K_c \omega_c g_0)\omega^2 + \omega_p^2 \omega_c^2)^2 + (-(2\xi_p \omega_p + 2\xi_c \omega_c + g_0 K_c \omega_c)\omega^3 + (2\xi_p \omega_p \omega_c^2 + 2\xi_c \omega_c \omega_p^2 + g_0 \omega_z^2 K_c \omega_c)w)^2}} d\omega$$

$$H_\infty = \min \left( \sqrt{\frac{(\omega_z^2 - \omega^2)^2 + (2\xi_z \omega_z \omega)^2}{(\omega^4 - (\omega_p^2 + \omega_c^2 + 2\xi_z \omega_c K_c \omega_c g_0)\omega^2 + \omega_p^2 \omega_c^2)^2 + (-(2\xi_p \omega_p + 2\xi_c \omega_c + g_0 K_c \omega_c)\omega^3 + (2\xi_p \omega_p \omega_c^2 + 2\xi_c \omega_c \omega_p^2 + g_0 \omega_z^2 K_c \omega_c)w)^2}} \right) \Big|_{@w_p}$$

**Minimizing the cost function with respect to values of closed loop damping (from zero to 1)**

## Stability of the System for various patterns

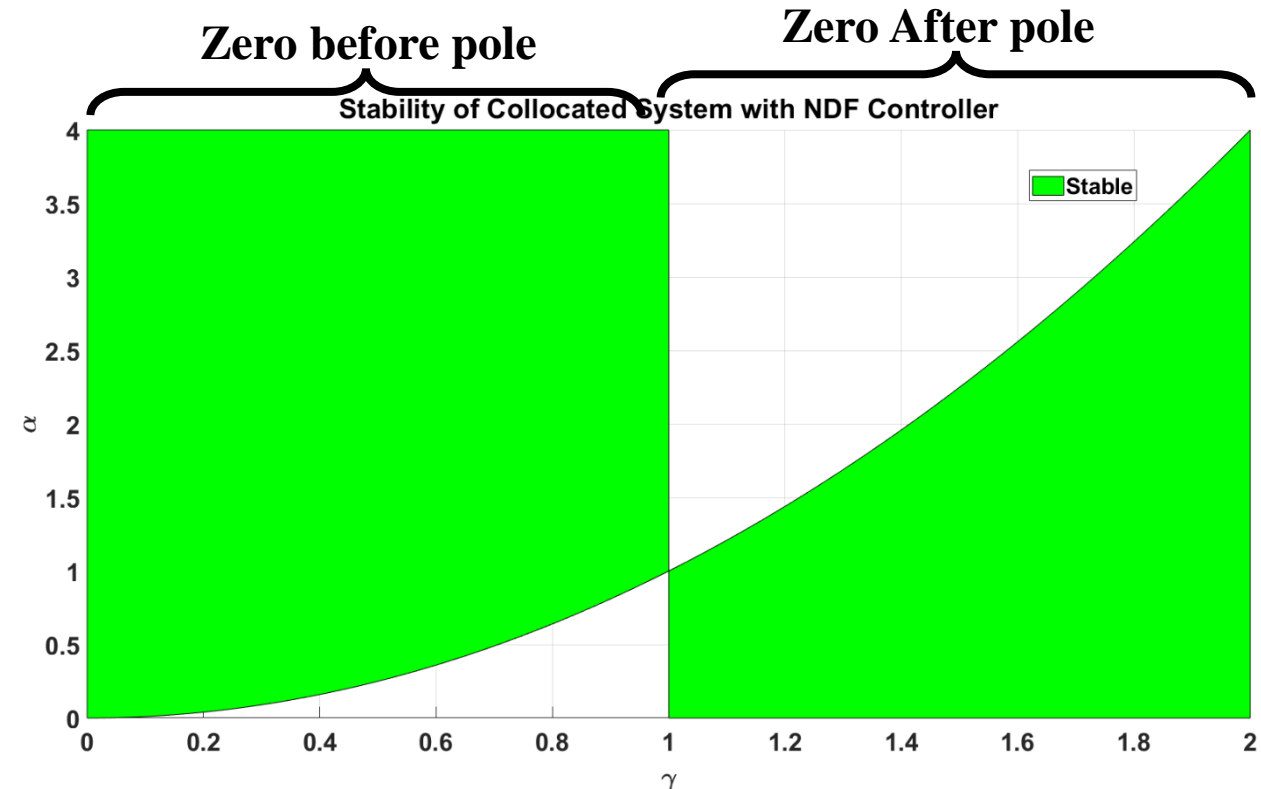
$$s^4 + (2\xi_p\omega_p + 2\xi_c\omega_c + k_c\omega_cg_0)s^3 + (\omega_p^2 + \omega_c^2 + 4\xi_p\xi_c\omega_p\omega_c + 2\xi_z\omega_zK_c\omega_cg_0)s^2 + (2\xi_p\omega_p\omega_c^2 + 2\xi_c\omega_c\omega_p^2 + g_0\omega_z^2K_c\omega_c)s + \omega_p^2\omega_c^2 = 0$$

Since the equations become very long in this case, a more simple method is considered.

It is considered that the damping of pole and zero to become zero.  $\xi_p = 0$  ,  $\xi_z = 0$

$\alpha$	$(2\xi_f^2 + 1) \pm 2\xi_f\sqrt{\xi_f^2 + 1}$
$\xi_c$	$\frac{2\xi_f(\alpha - \gamma^2)}{\sqrt{\alpha}(1 - \gamma^2)}$
$K_c$	$\frac{4\xi_f(1 - \alpha)}{g_0\sqrt{\alpha}(1 - \gamma^2)}$

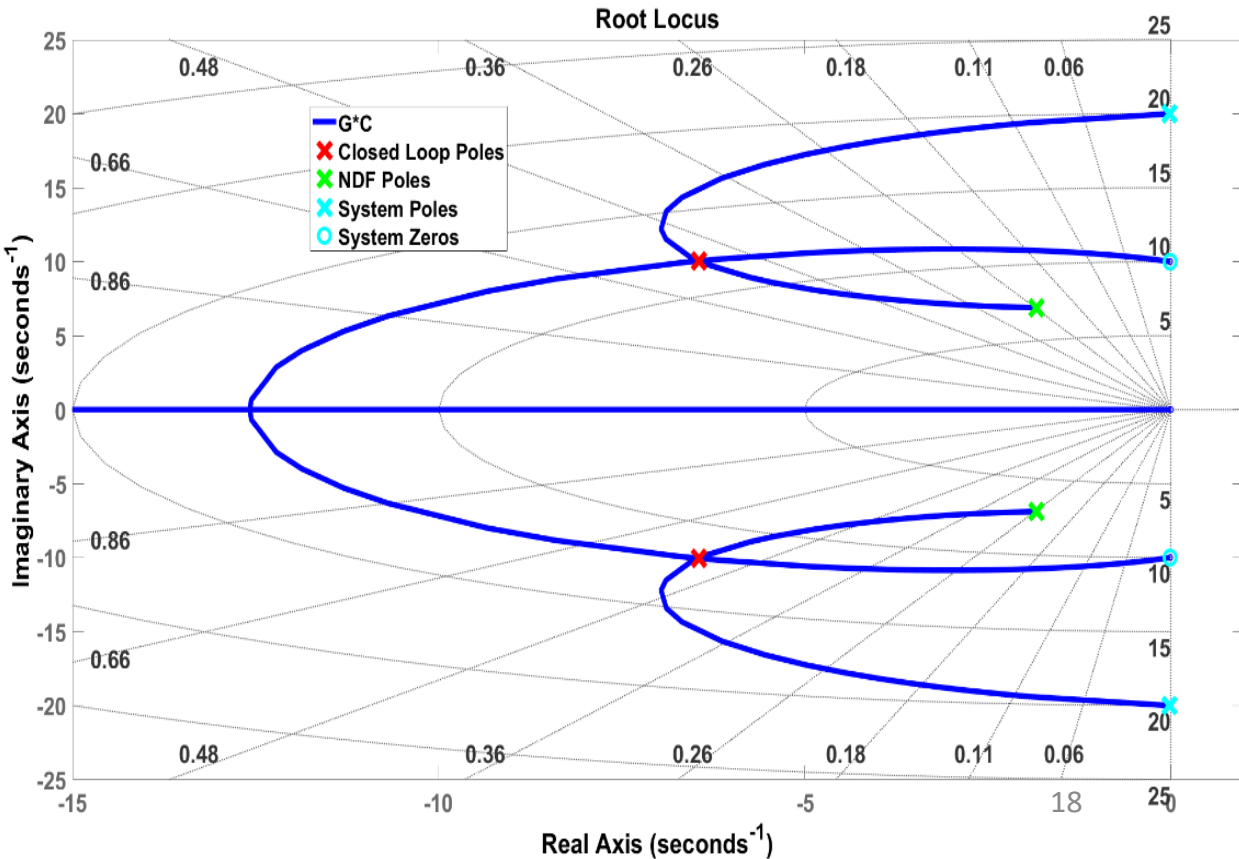
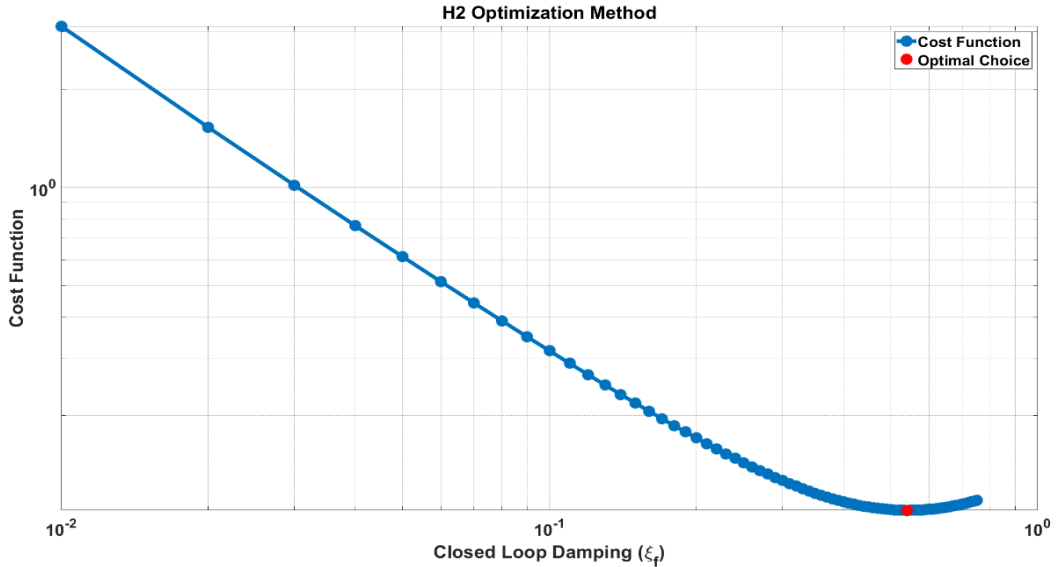
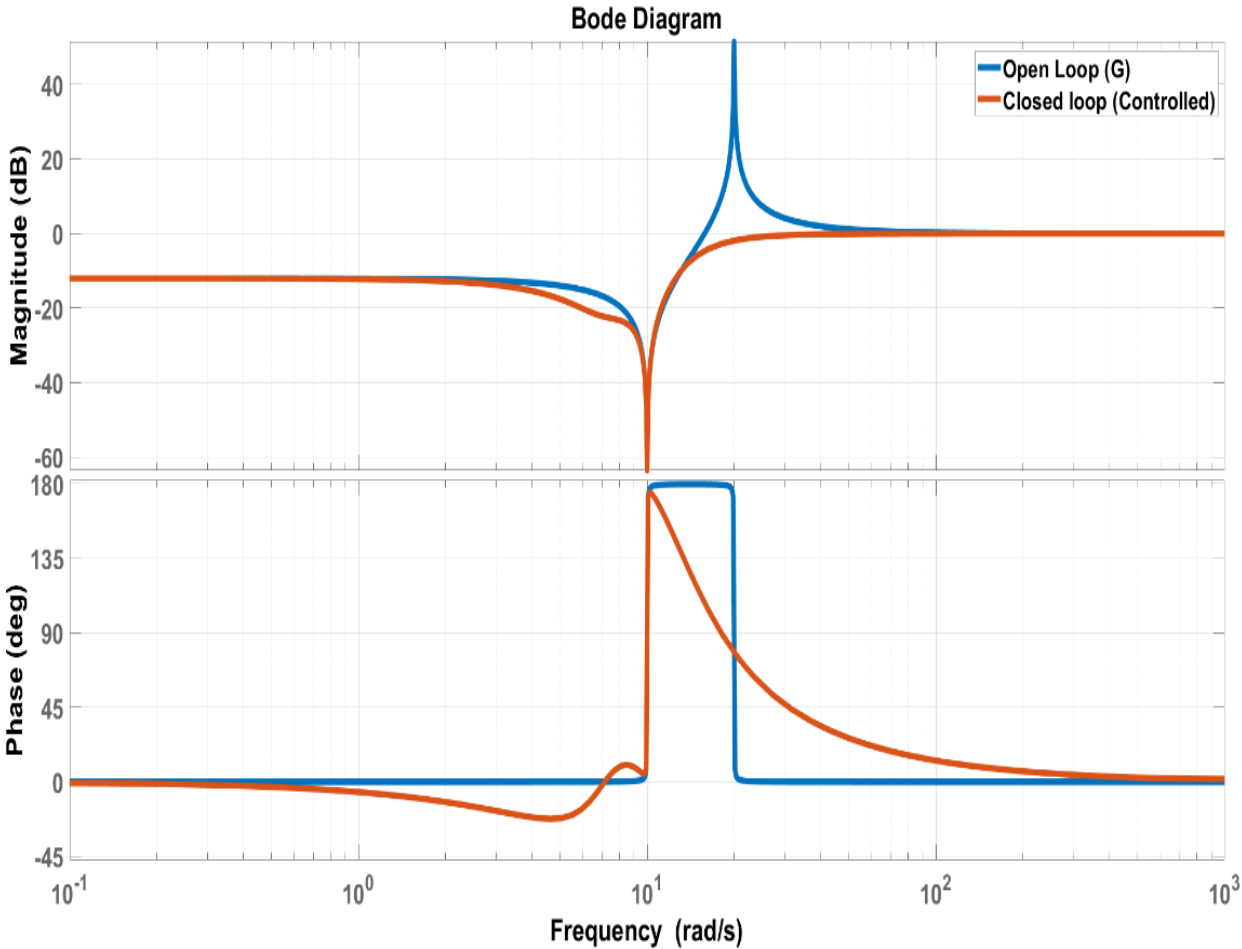
$$\begin{cases} \gamma < 1 & \text{then } \alpha > \gamma^2 \\ \gamma > 1 & \text{then } \alpha < \gamma^2 \end{cases}$$



# Zero Before Pole ( $\lambda > 1$ )

Example 1:

$$G(s) = \frac{s^2 + 0.02 * s + 100}{s^2 + 0.04 * s + 400}$$

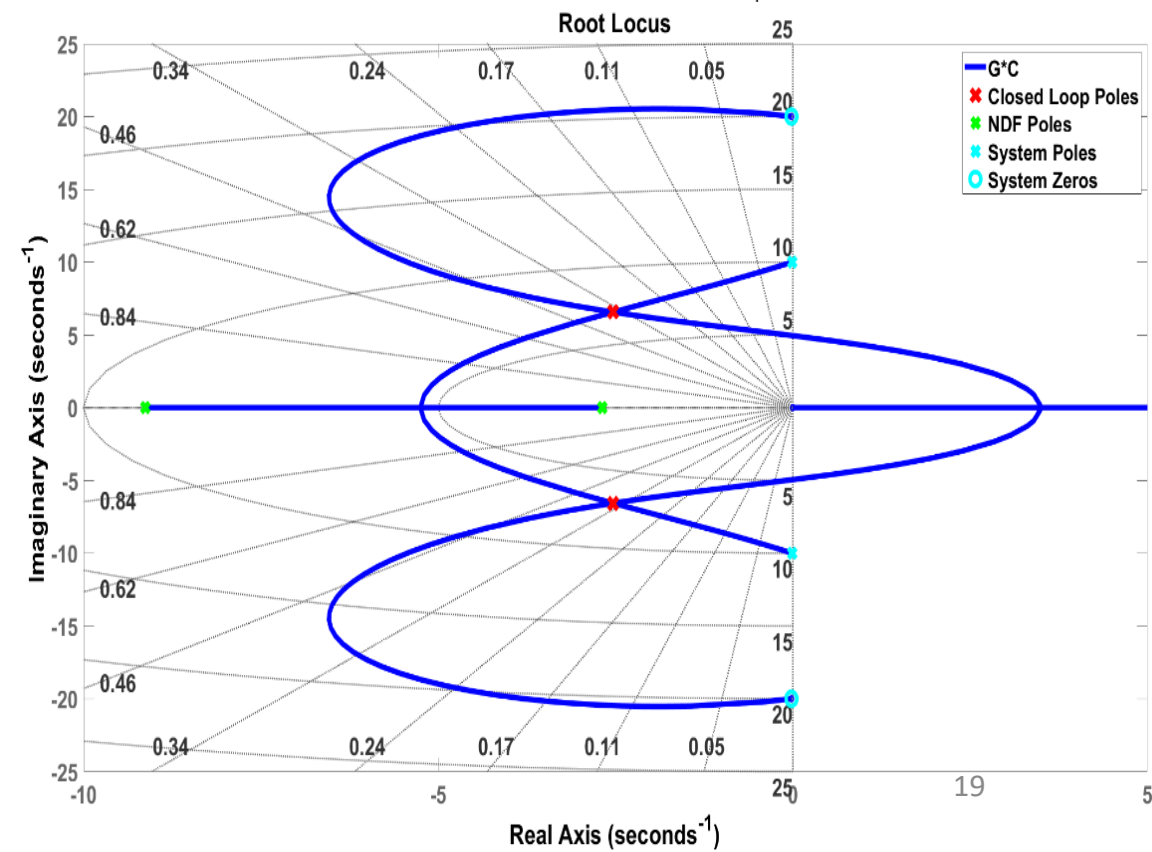
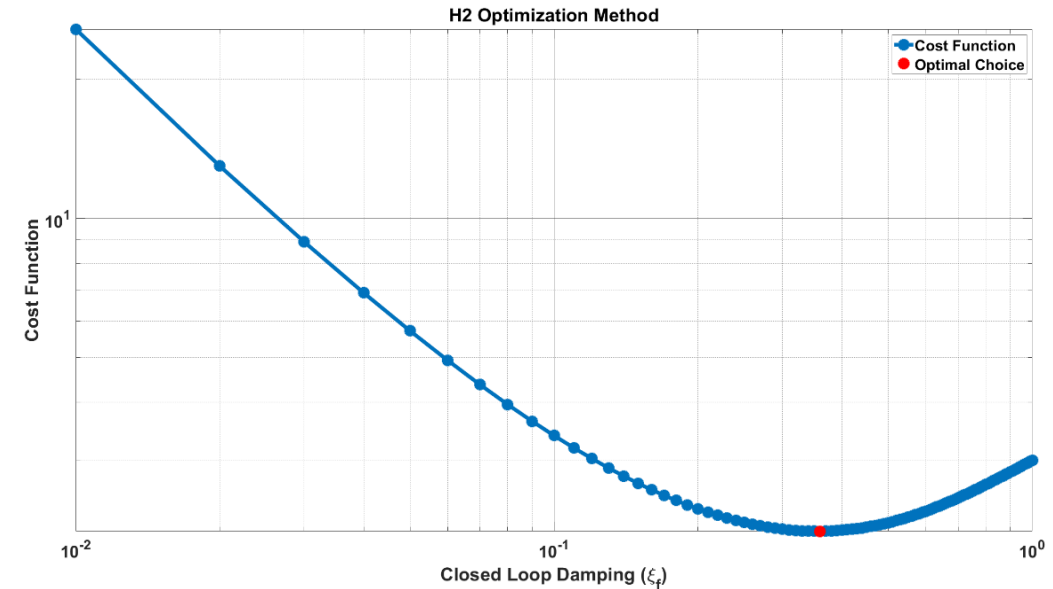
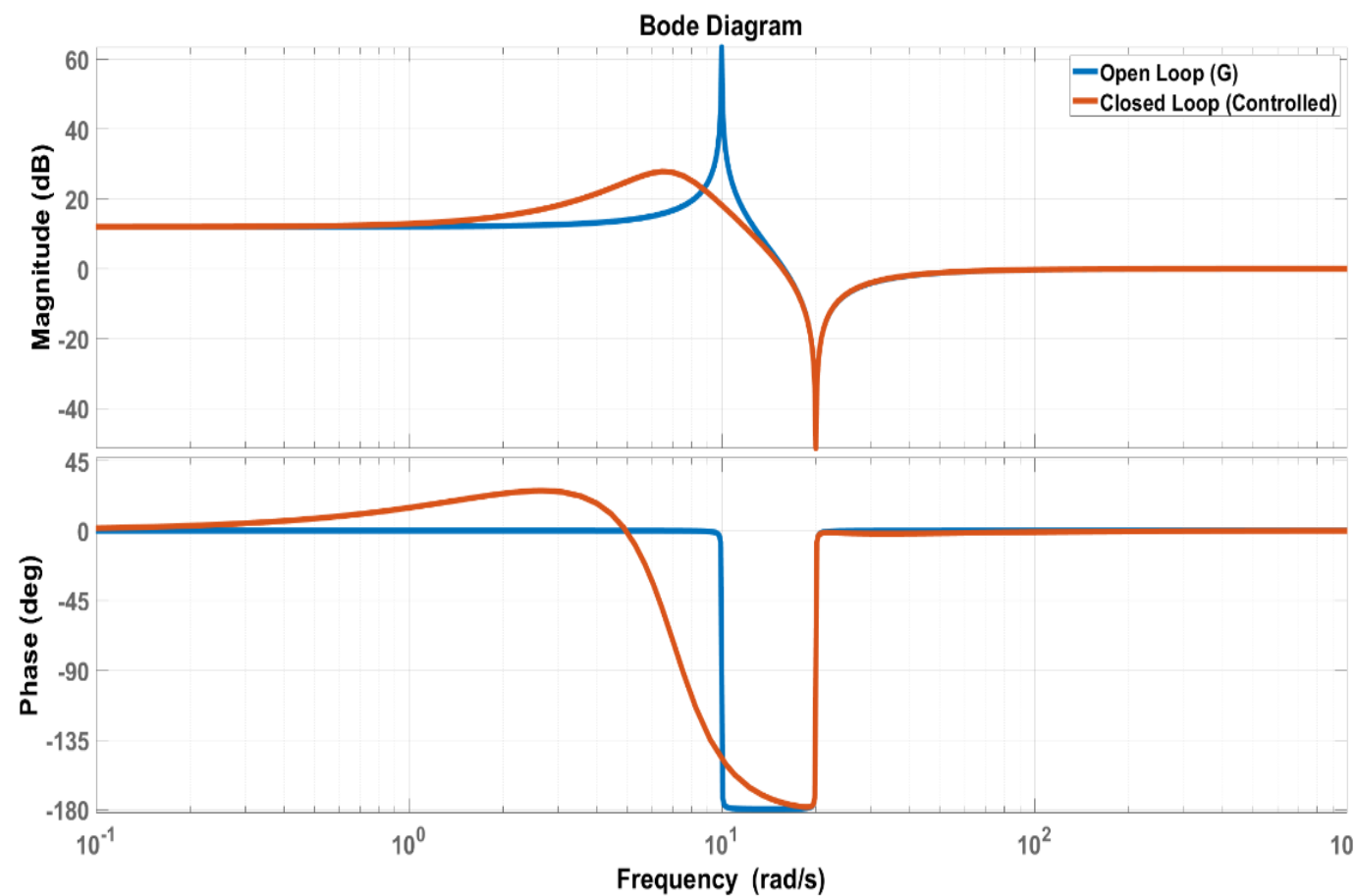




# Zero After Pole ( $\lambda < 1$ )

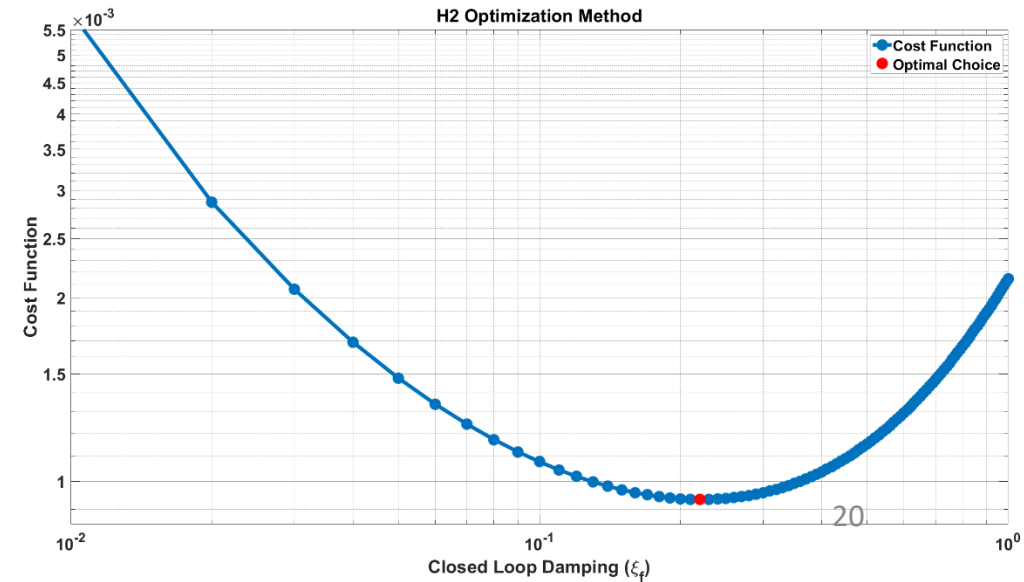
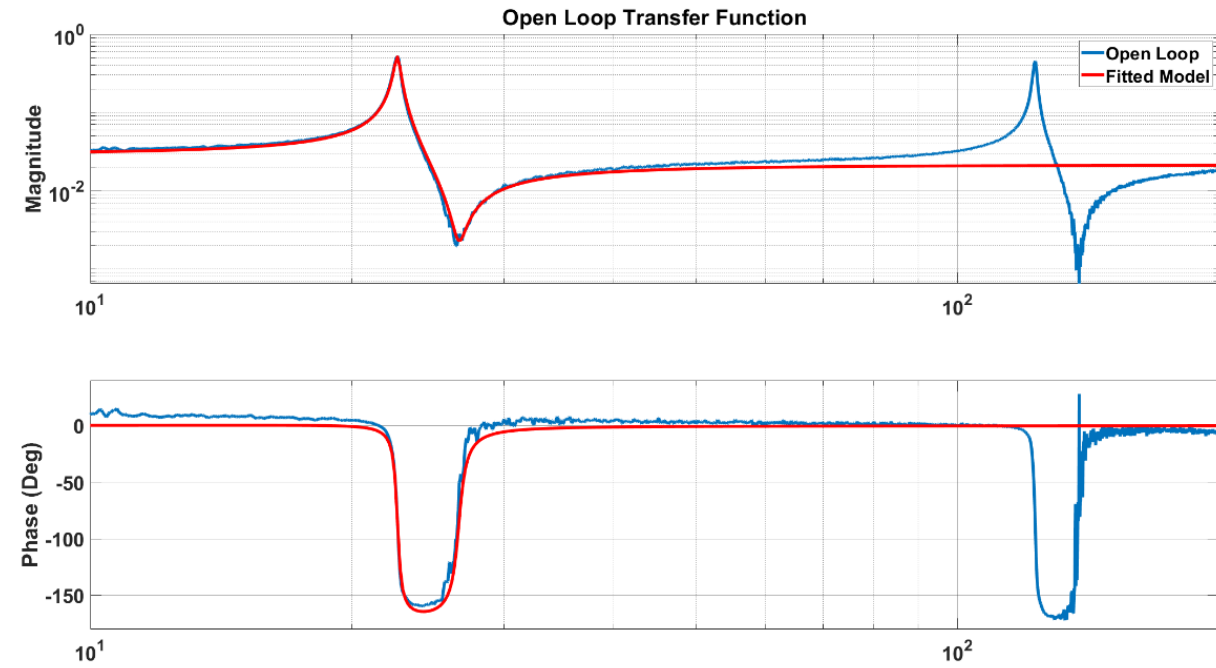
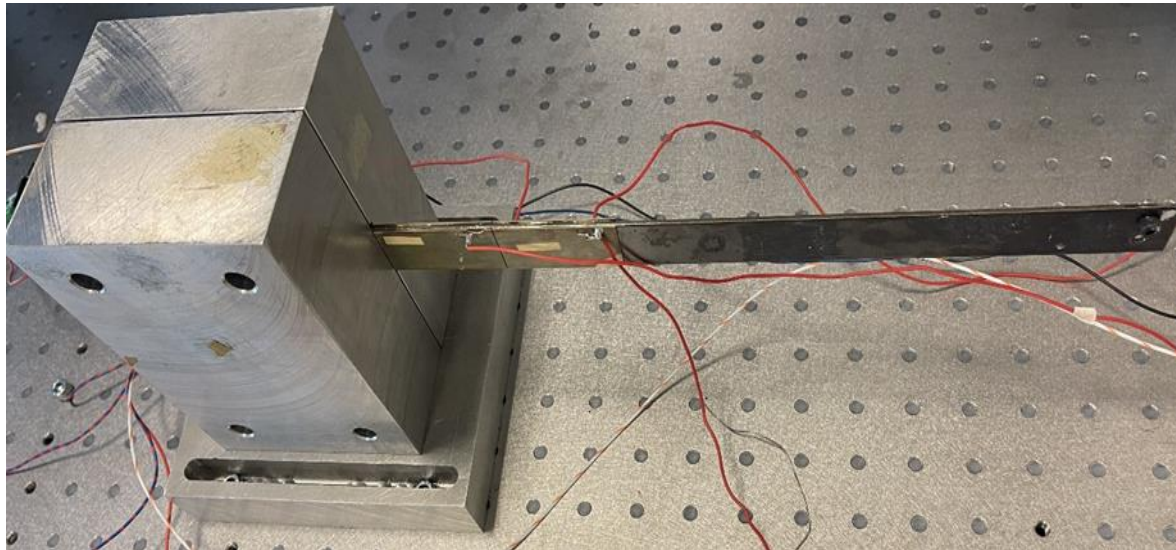
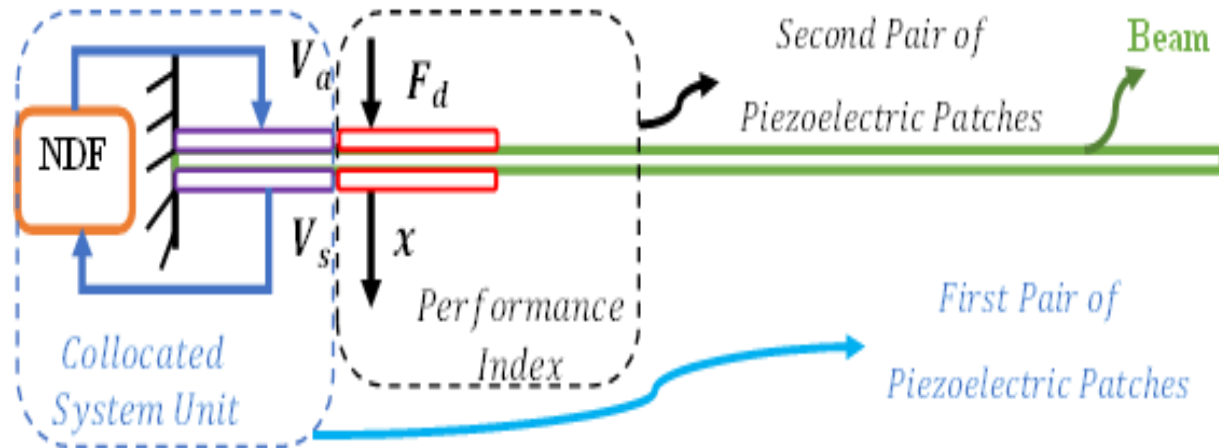
## Example 2:

$$G(s) = \frac{s^2 + 0.04 * s + 400}{s^2 + 0.02 * s + 100}$$

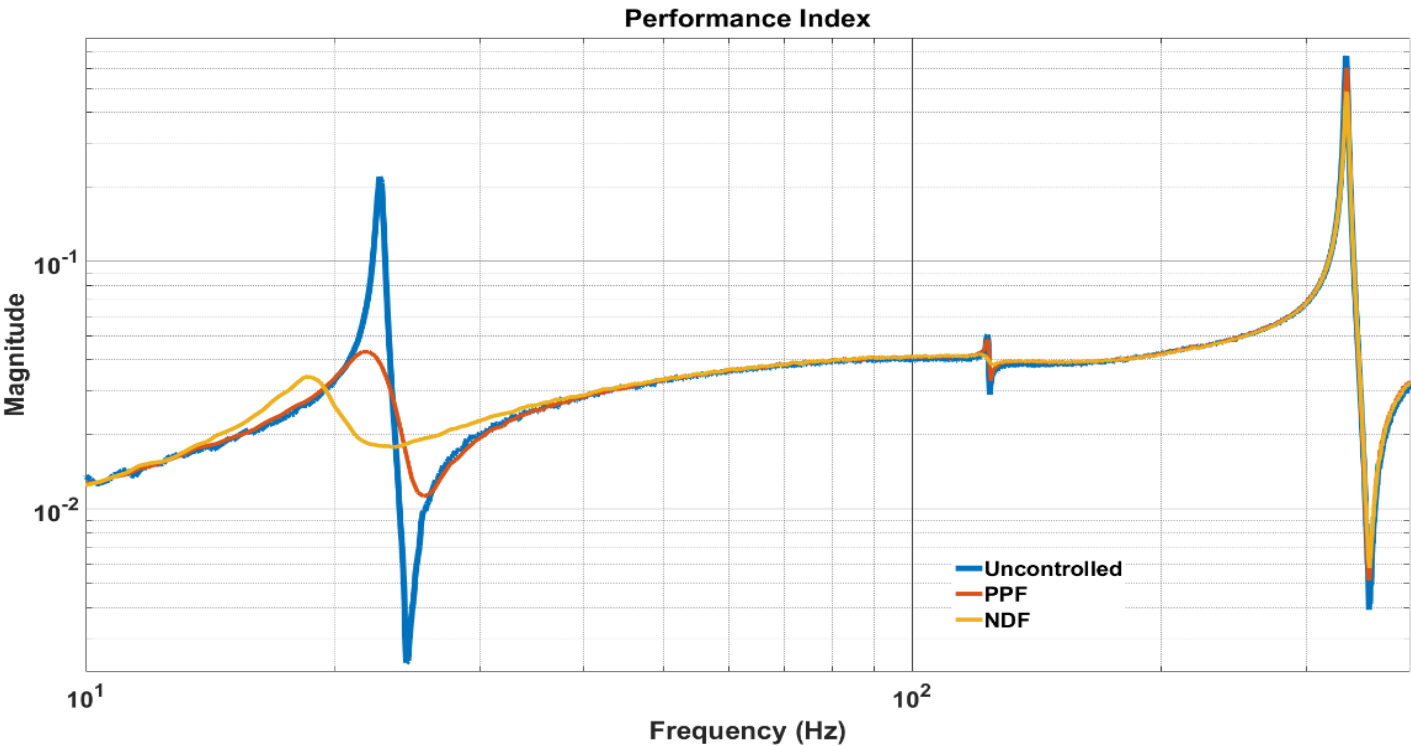
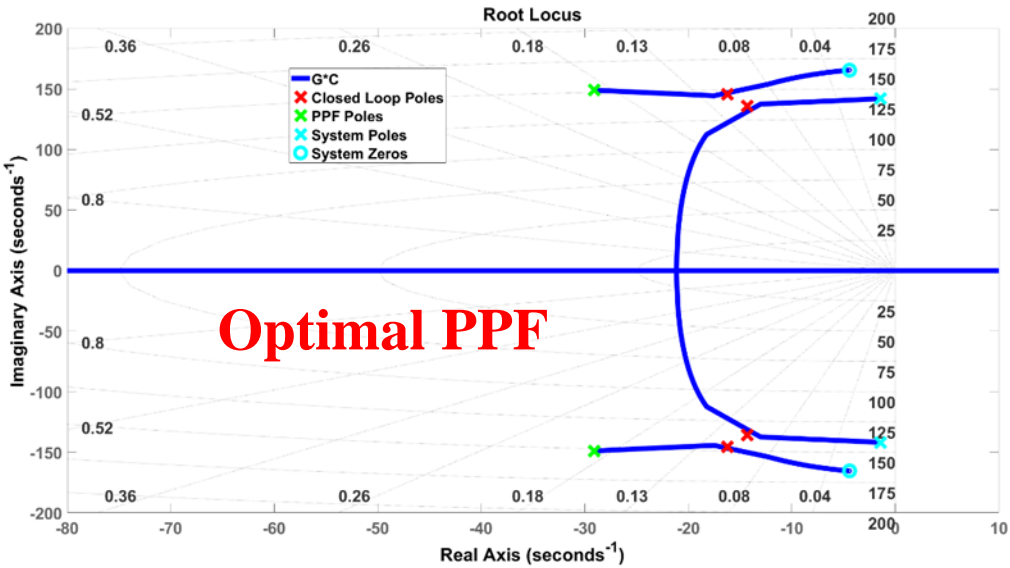
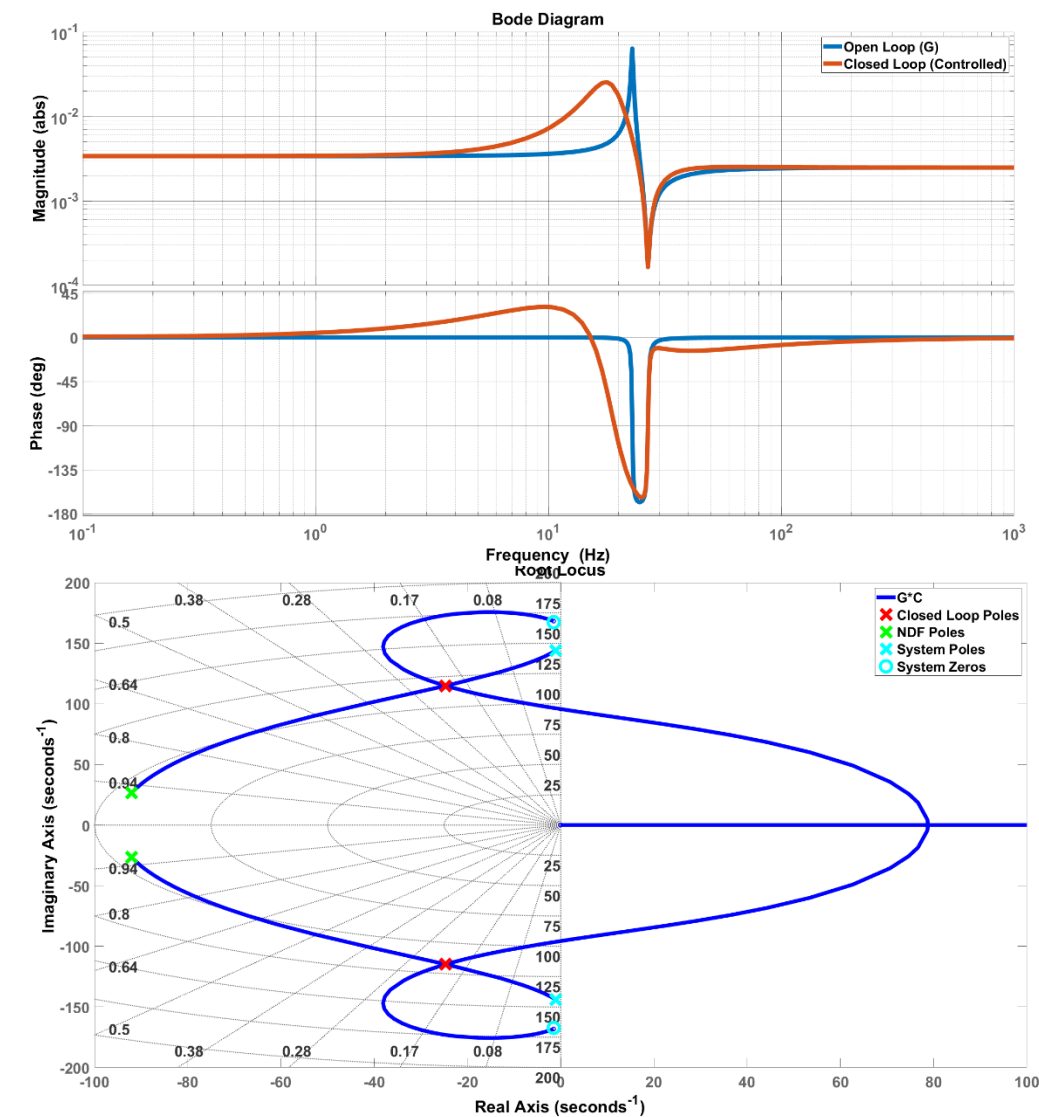


# Experimental Verifications

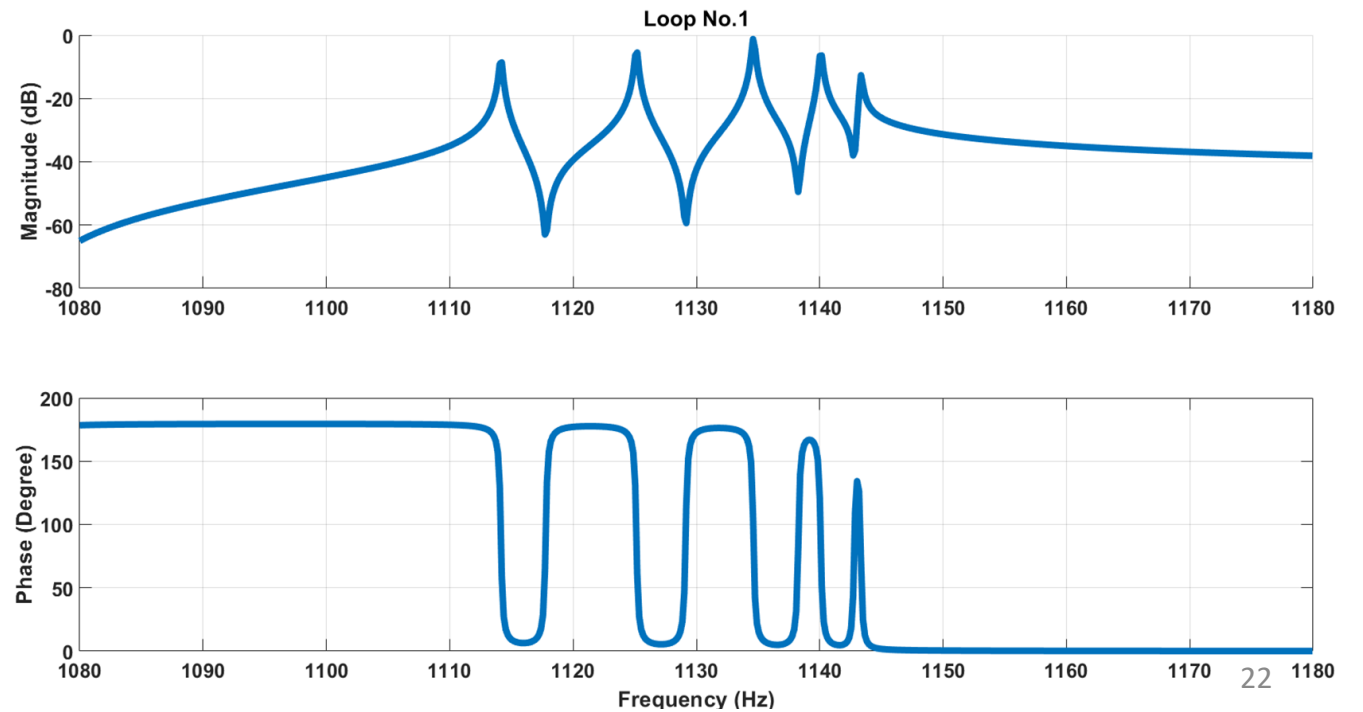
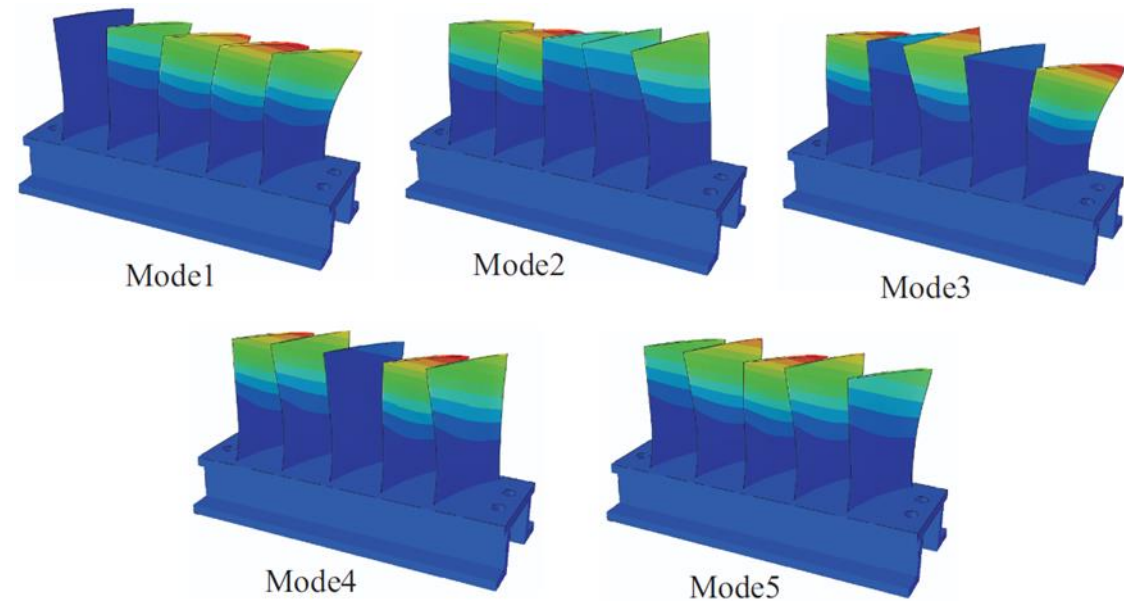
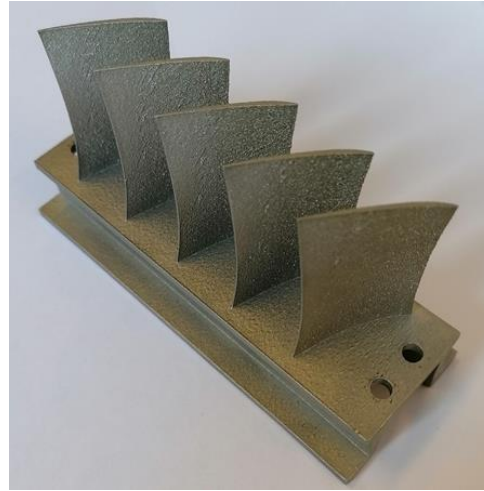
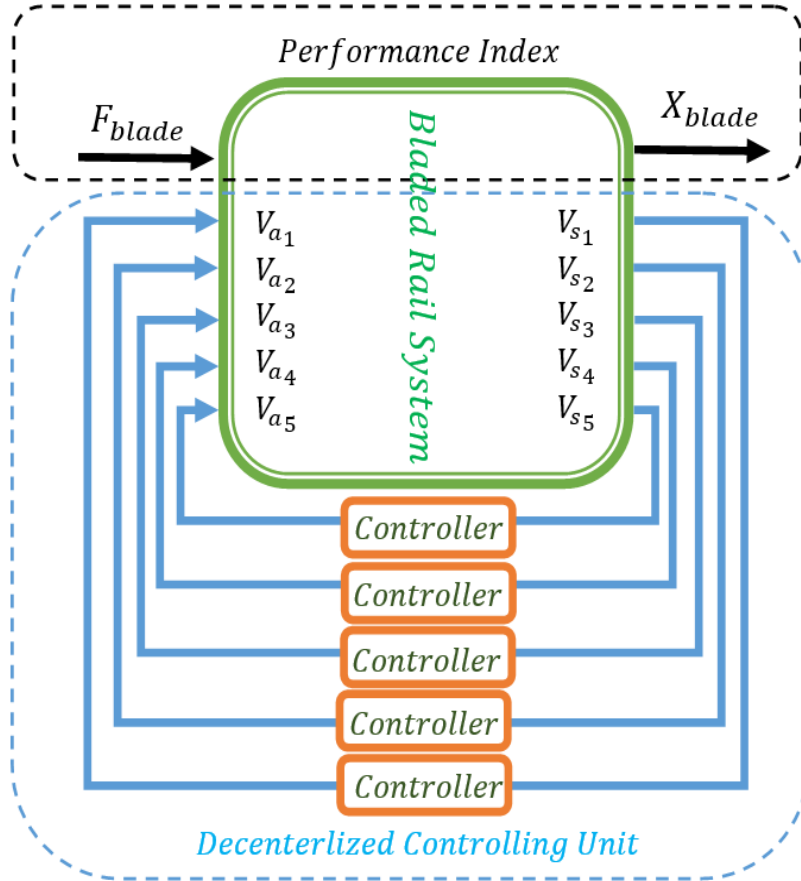
## Beam



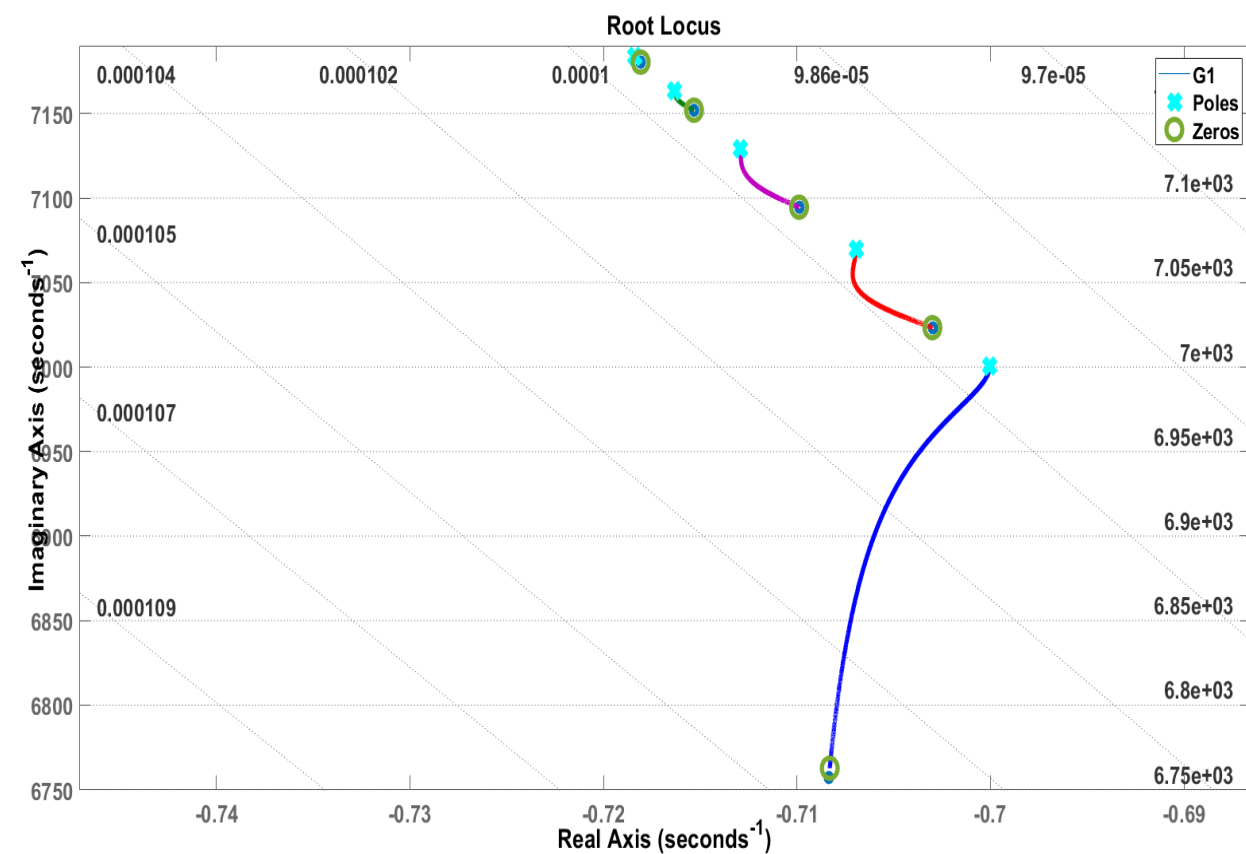
# Optimal NDF



# Bladed Rail

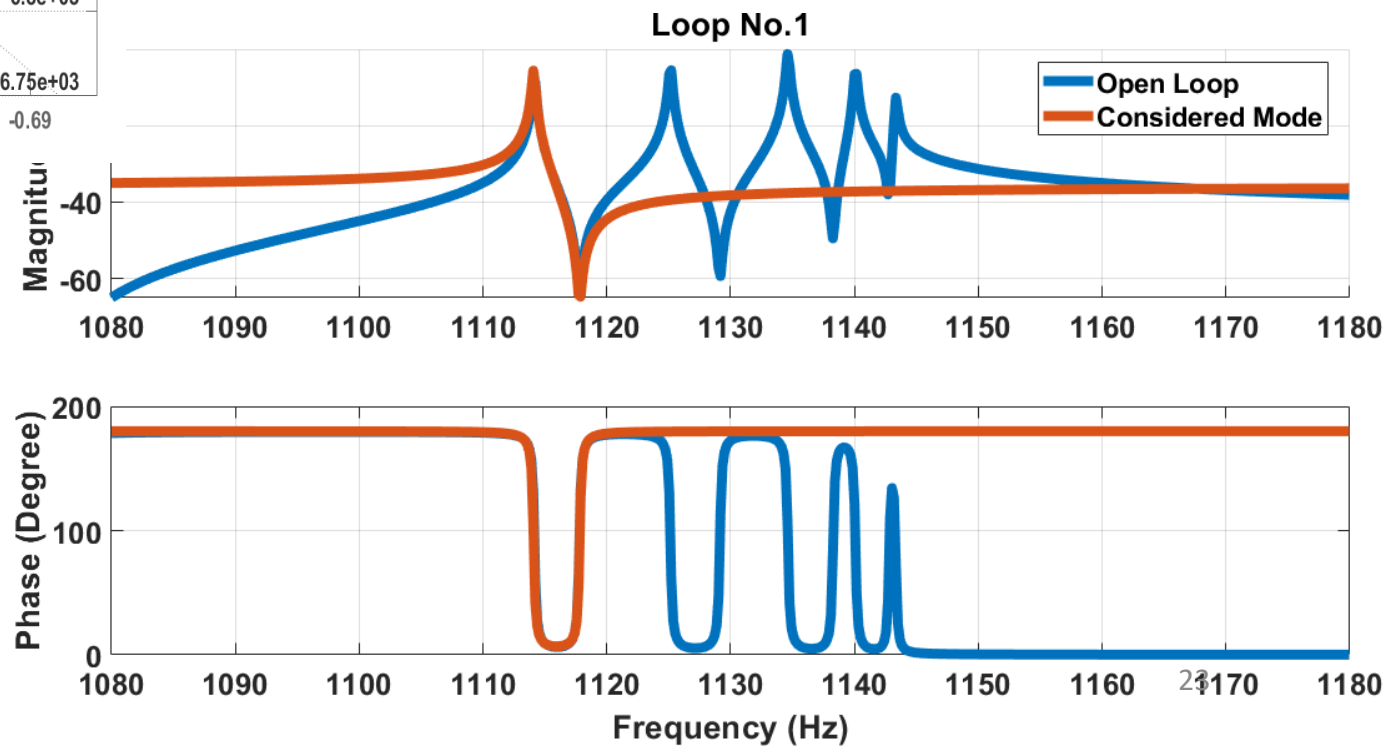




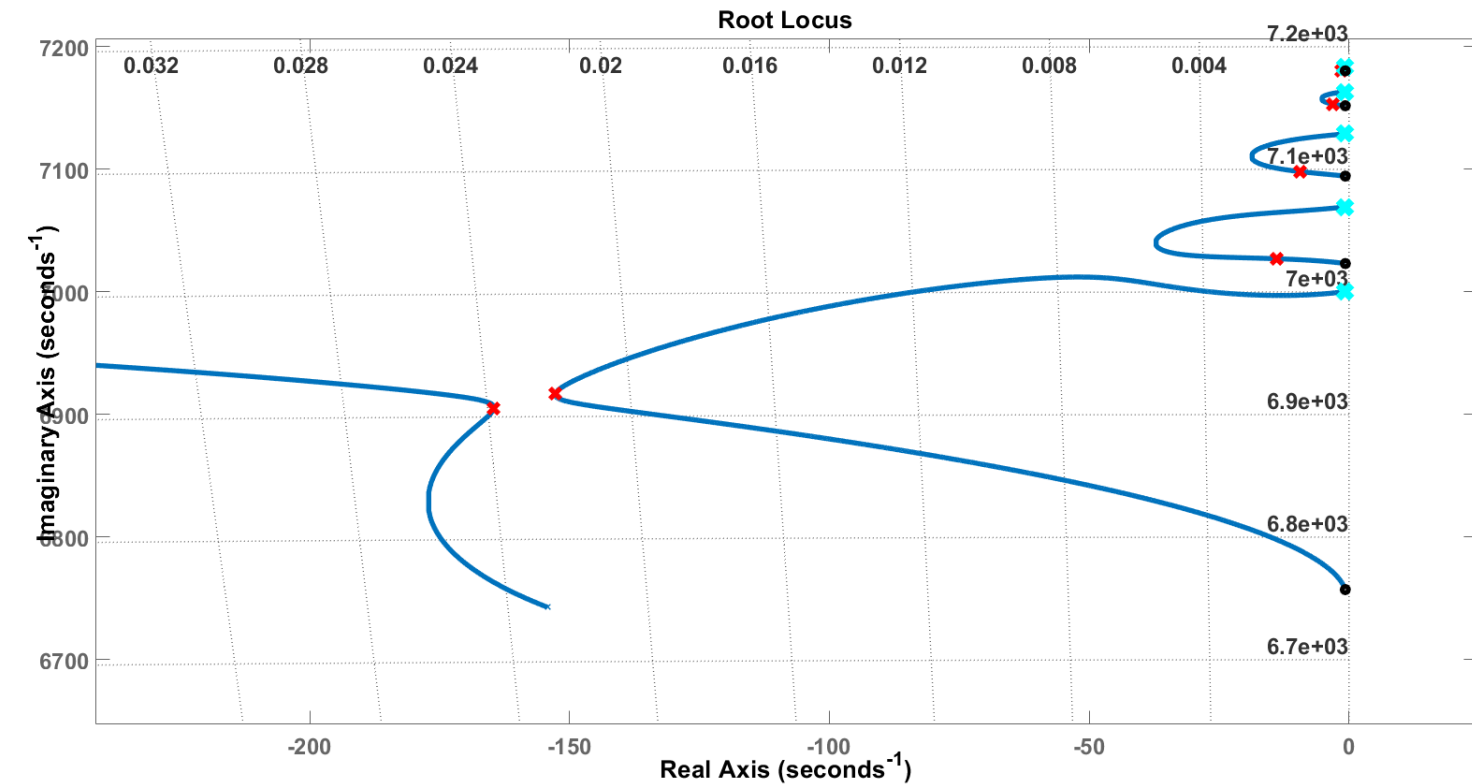


**Distances between zero and pole  
decreases in higher modes!**

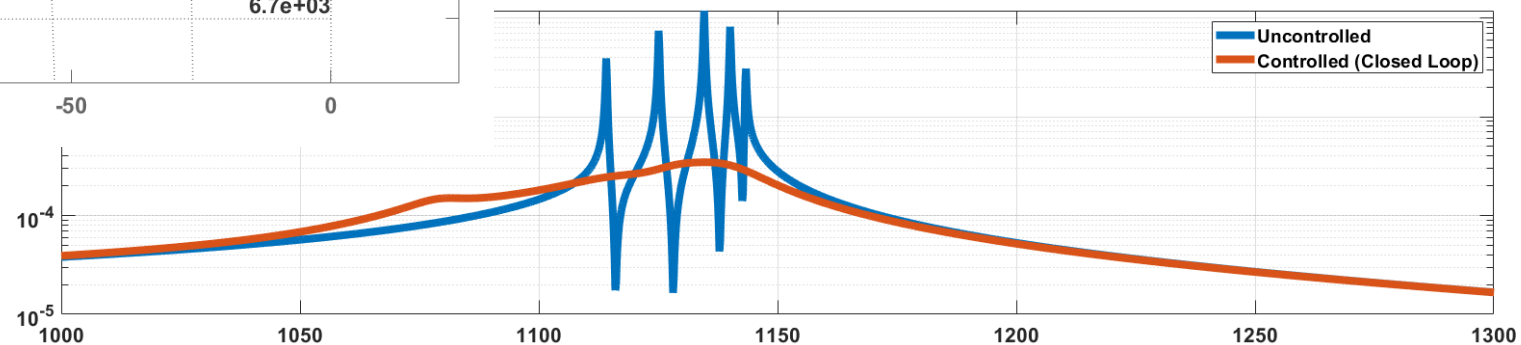
**So highest damping can be occurred  
only in the first Mode!**



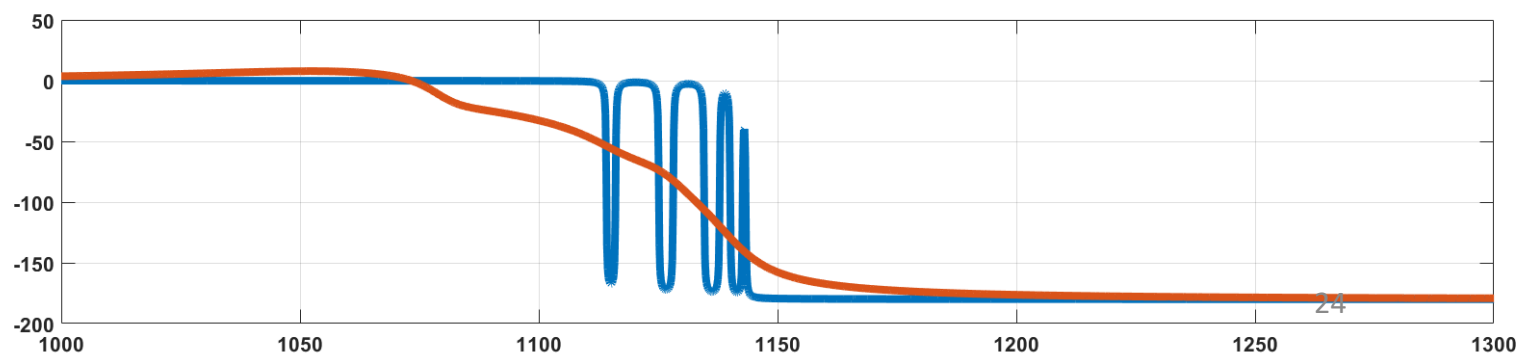




**First Mode damping Increased From  $10^{-4}$  to 0.024!**

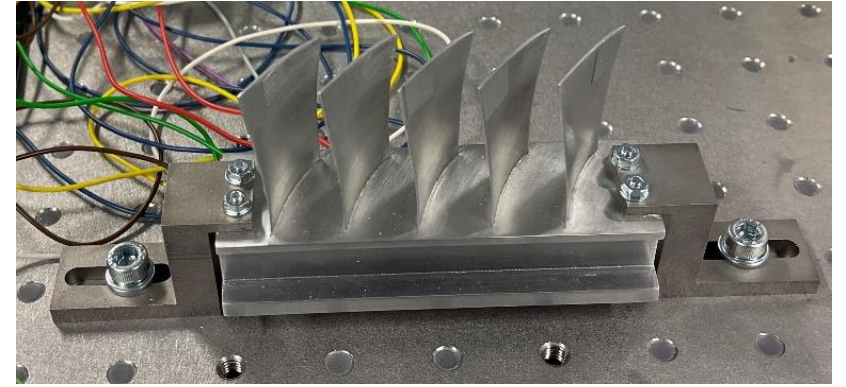


**Performance Index shows all the Family mode is damped!**

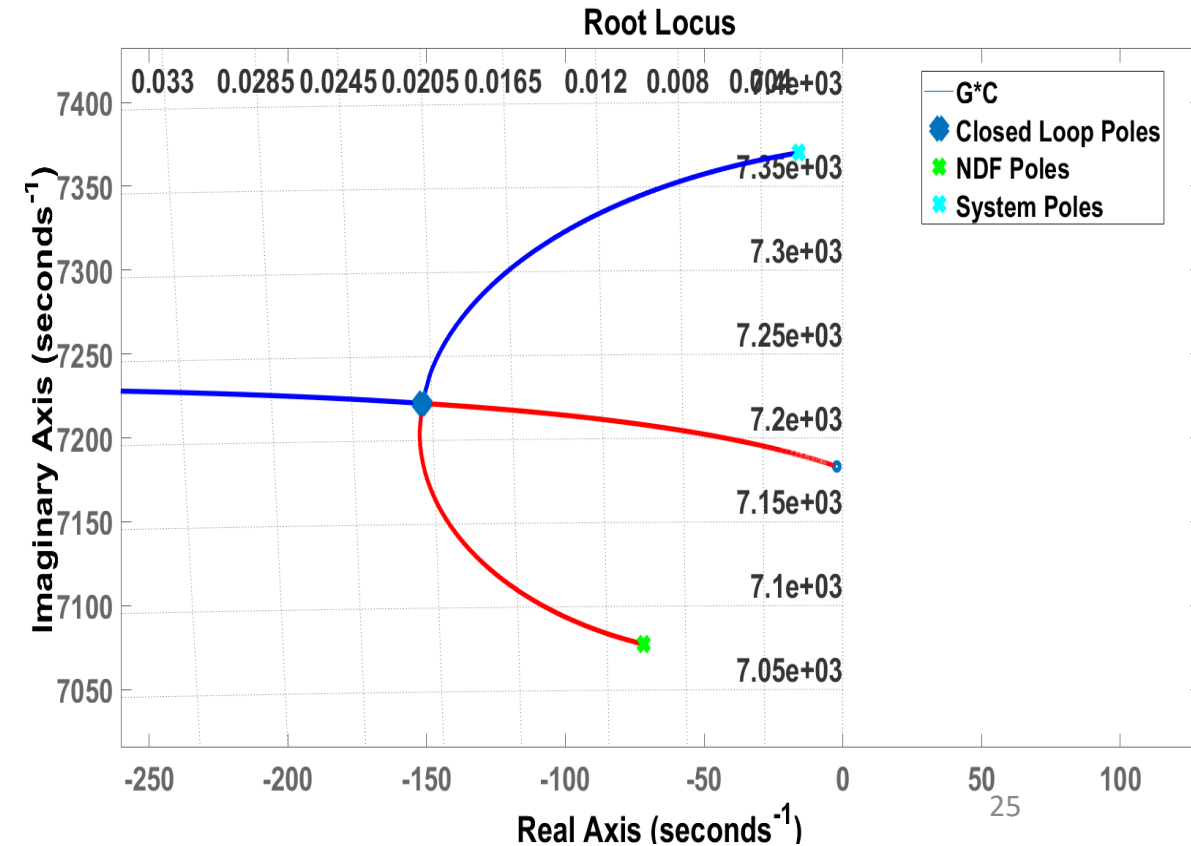
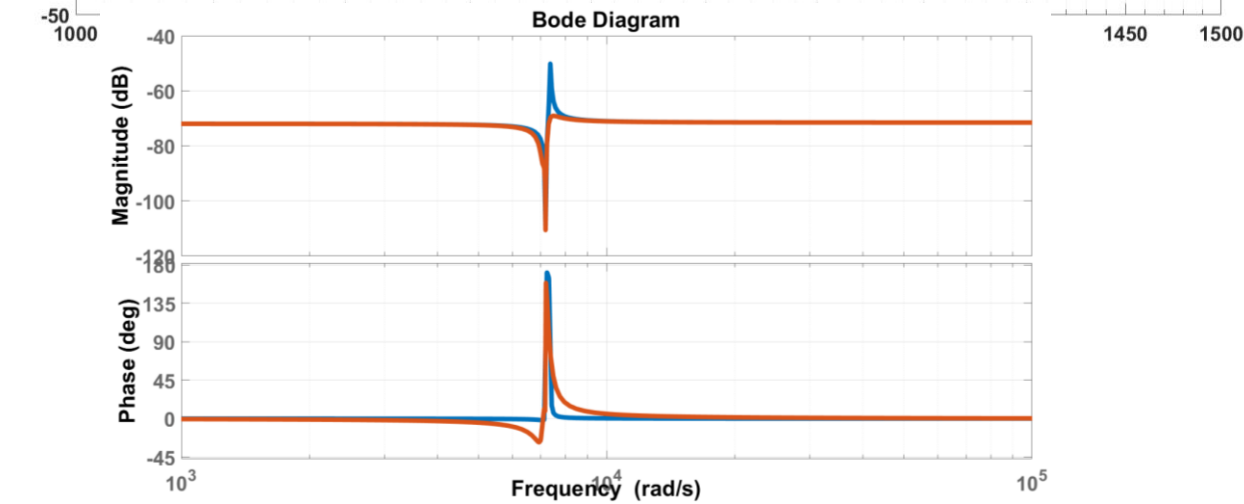
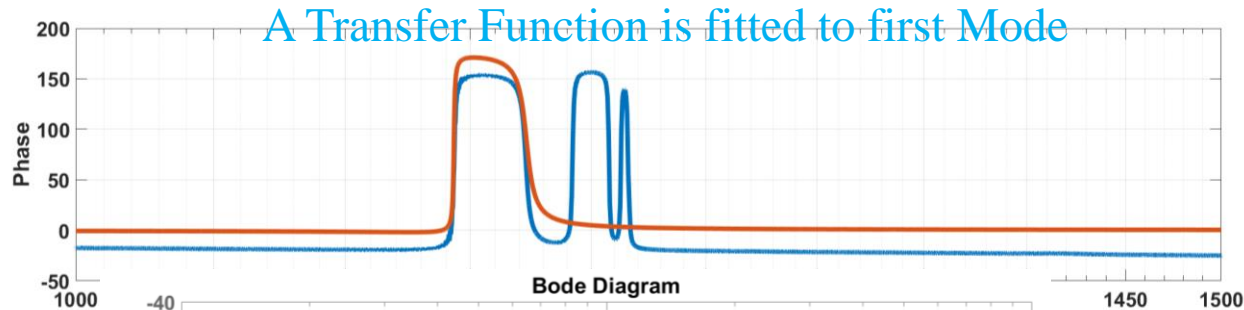
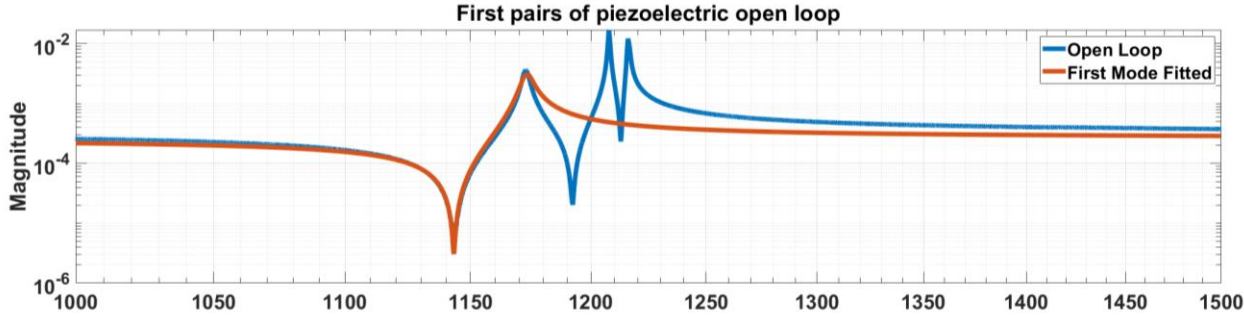


# Experimental Test: First Bladed Rail

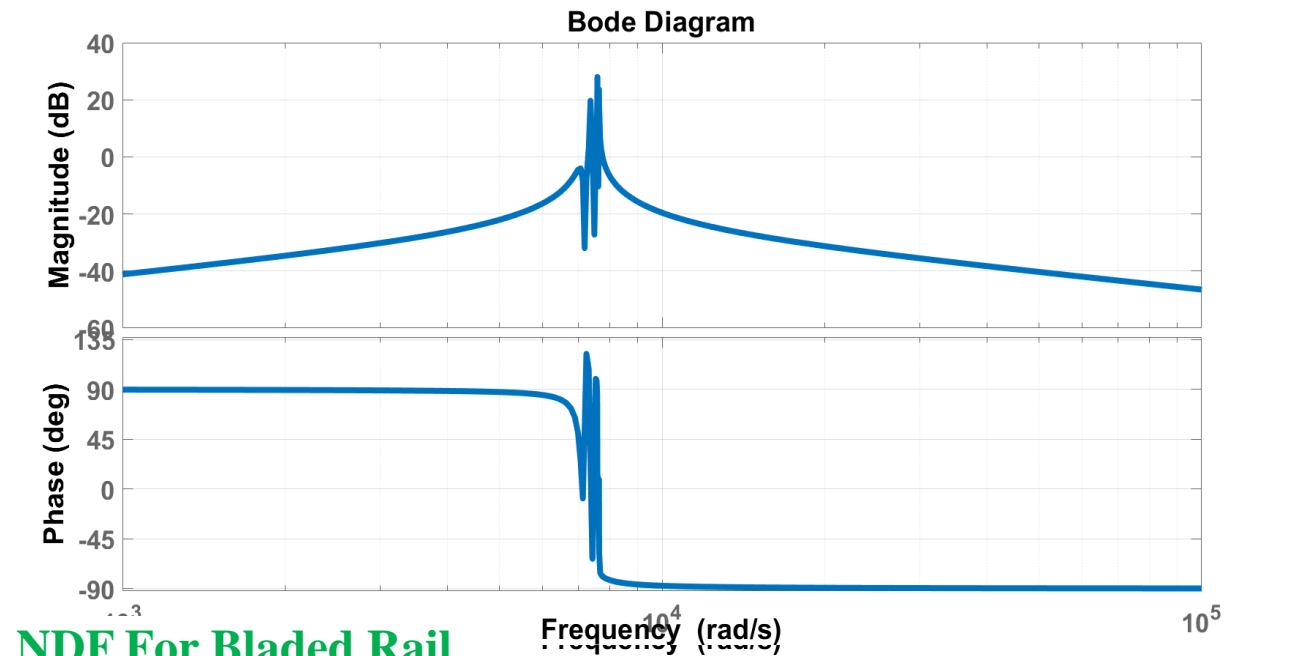
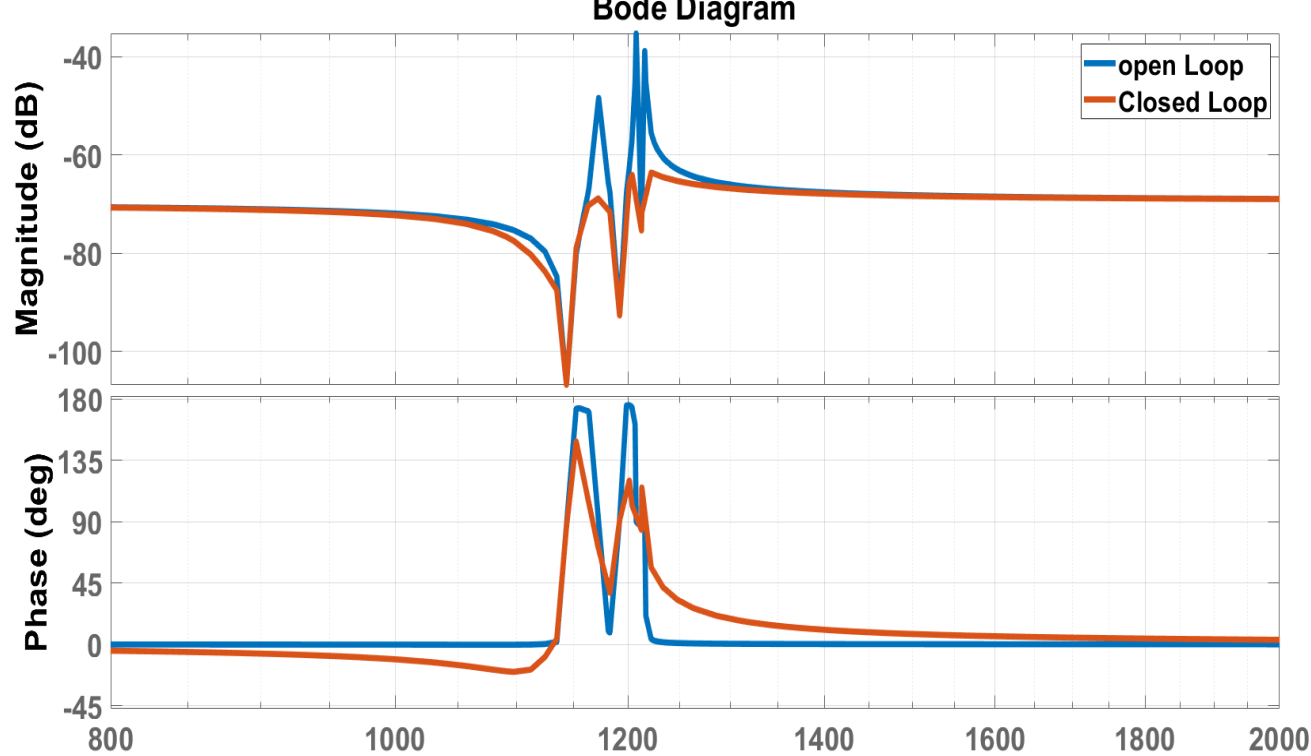
## Zero Before Pole Pattern



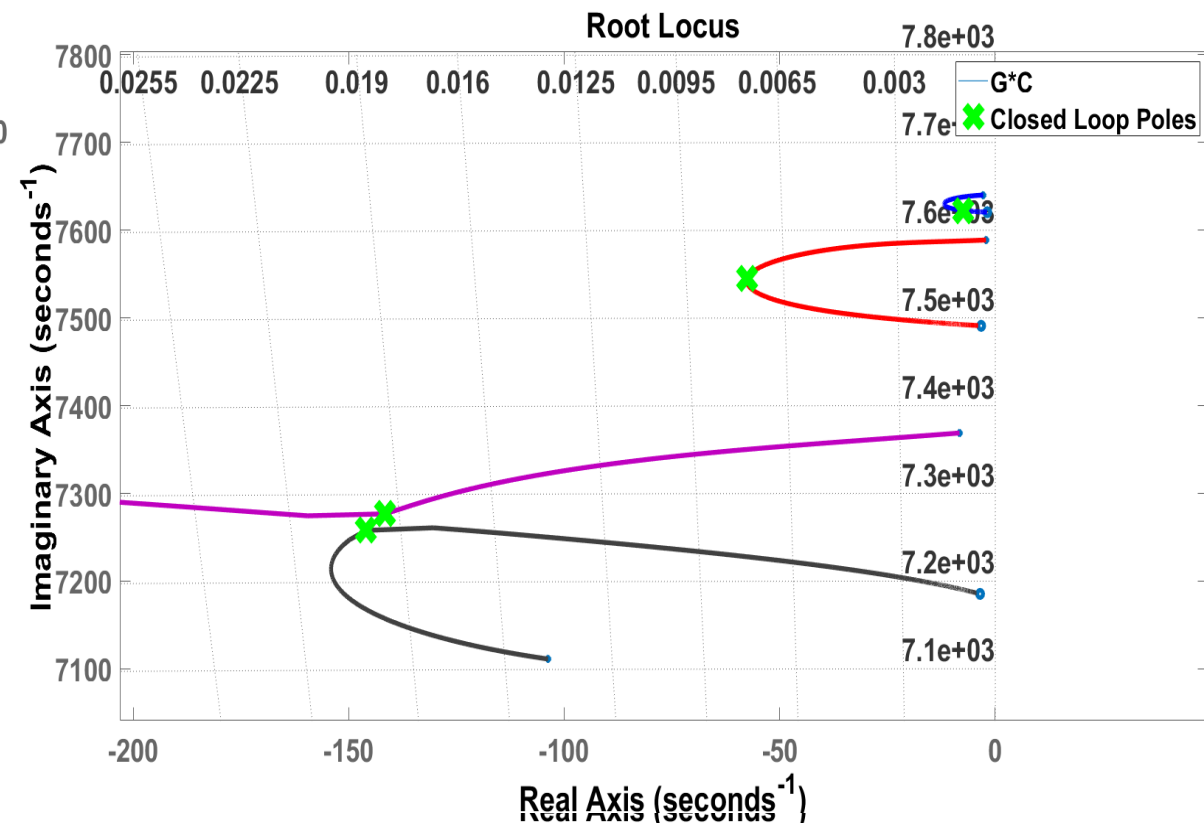
## NDF Design



## NDF For Bladed Rail



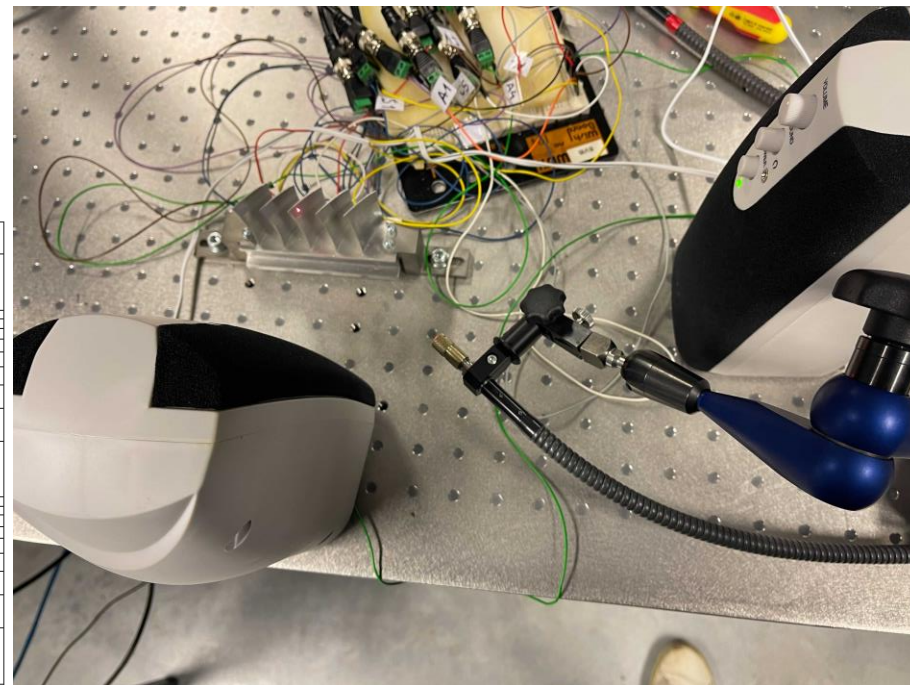
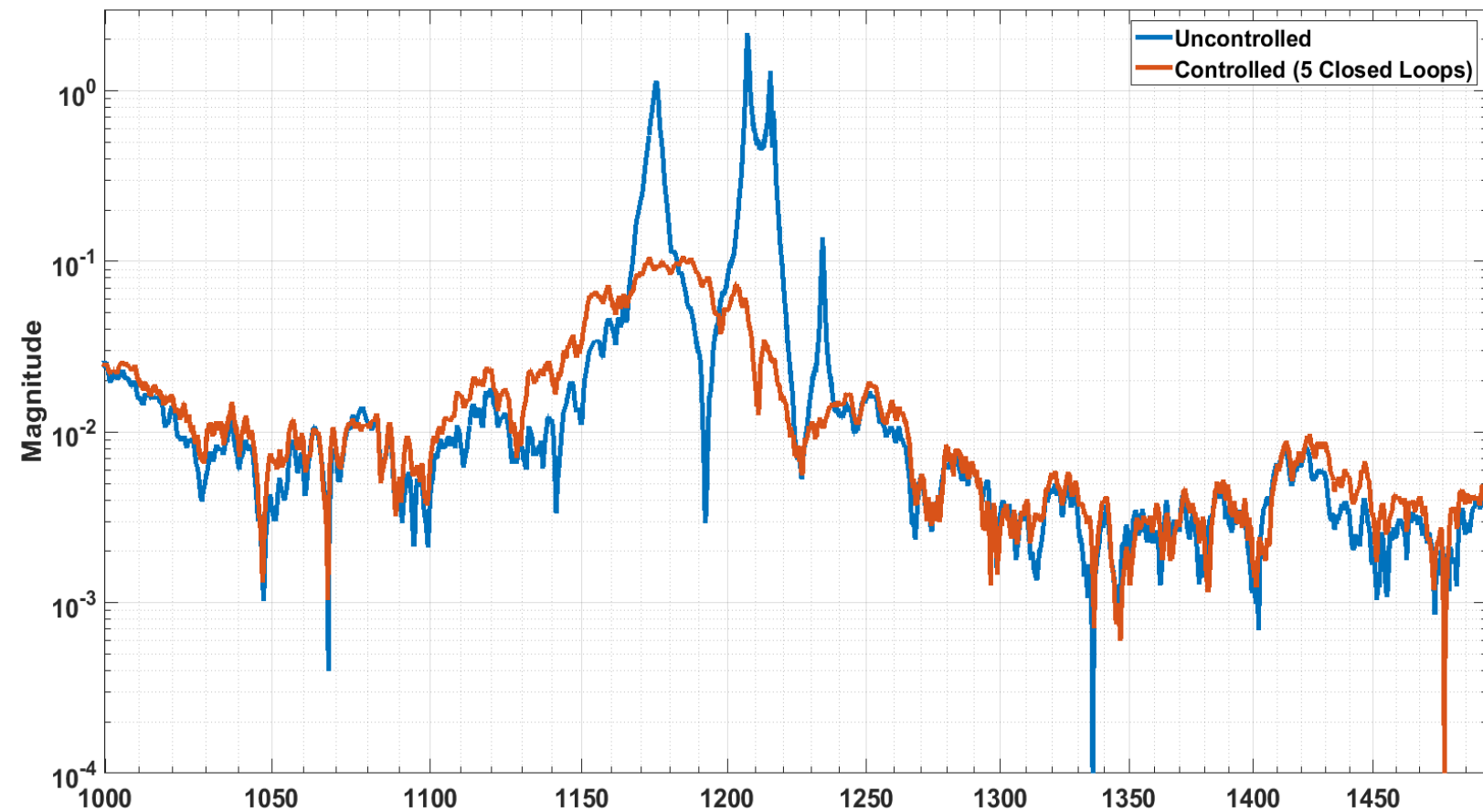
**NDF For Bladed Rail**



**NDF Design**



## *Test Result*

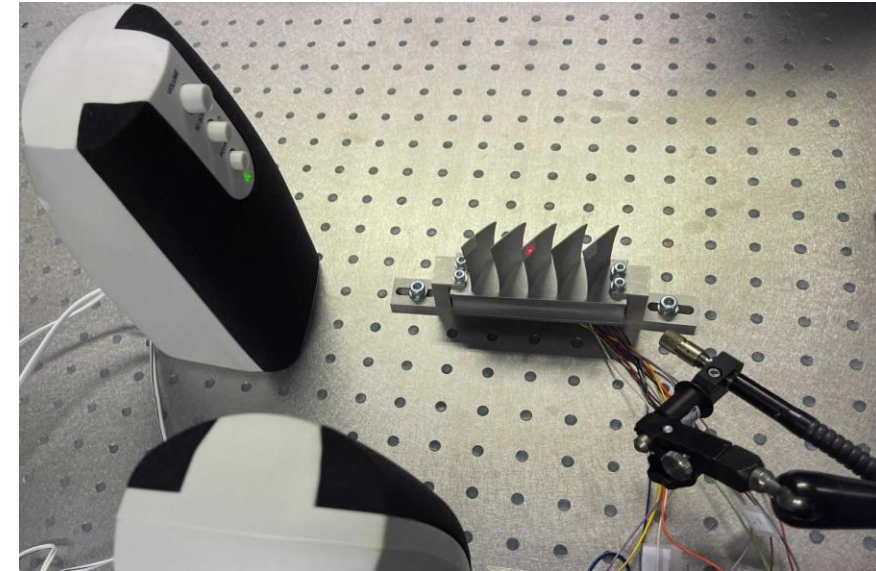
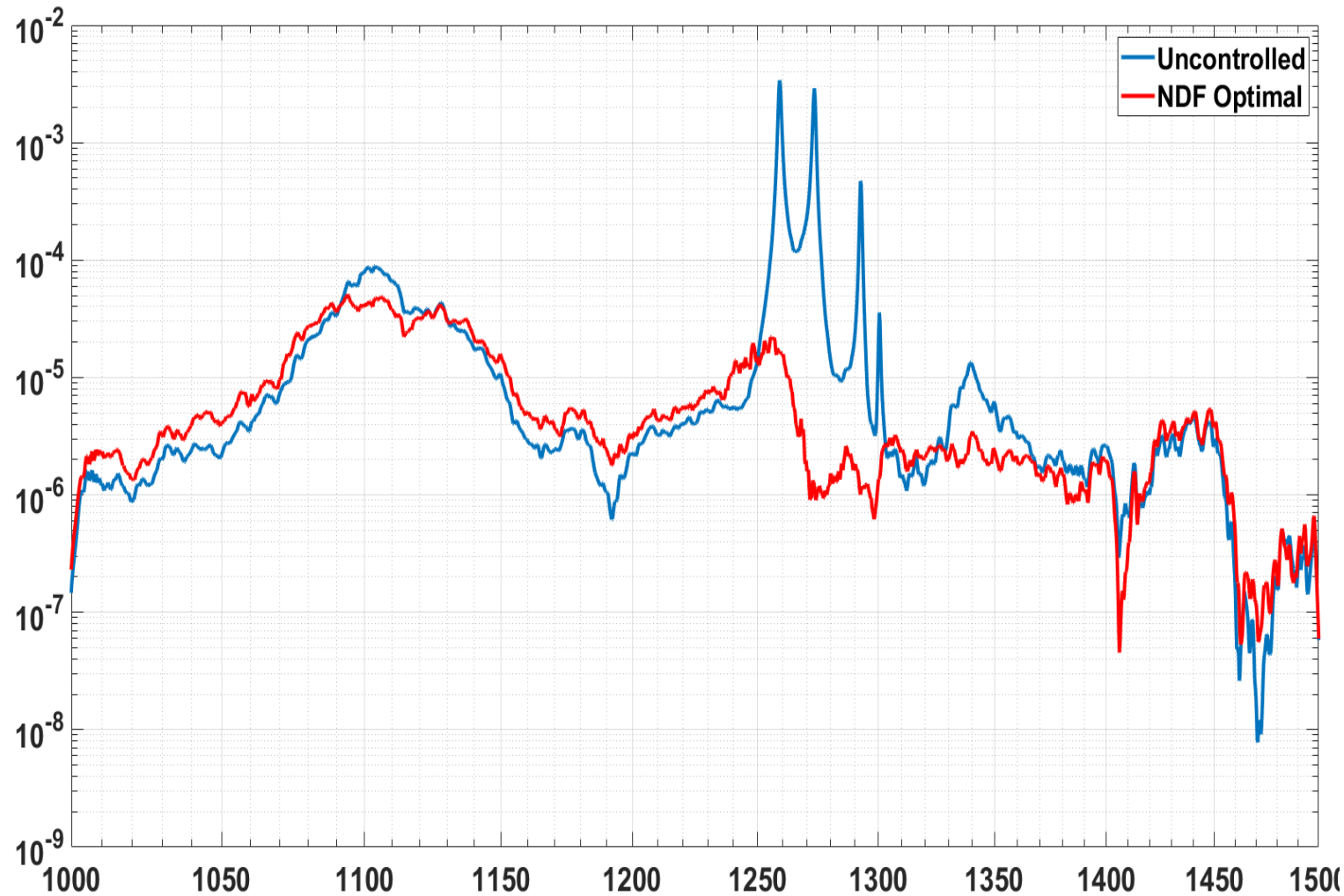




# Experimental Test: Second Bladed Rail

## Zero After Pole Pattern

The Same Procedure has been done.

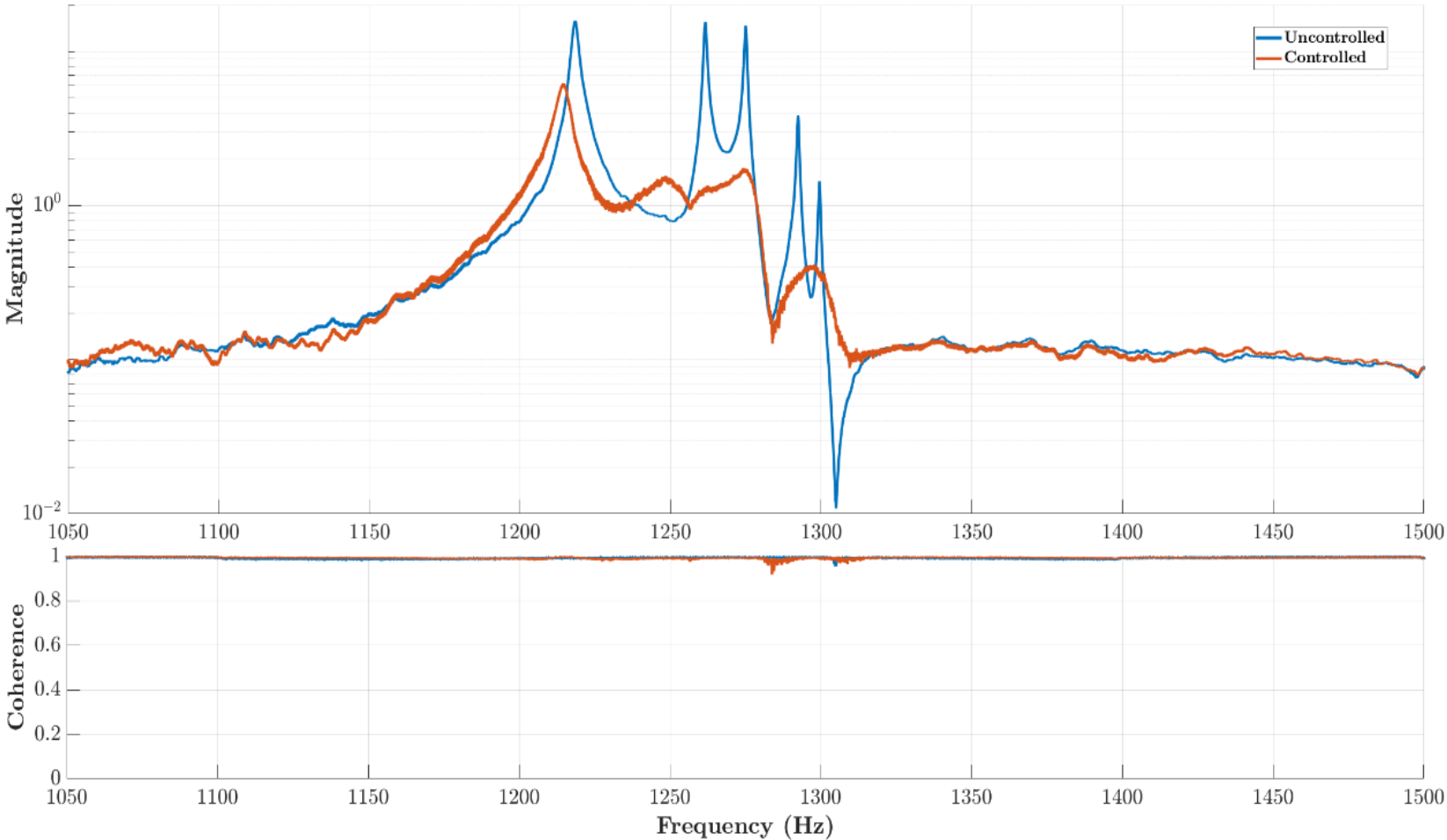




# PPF Controller

Controllers are not the same!!!!

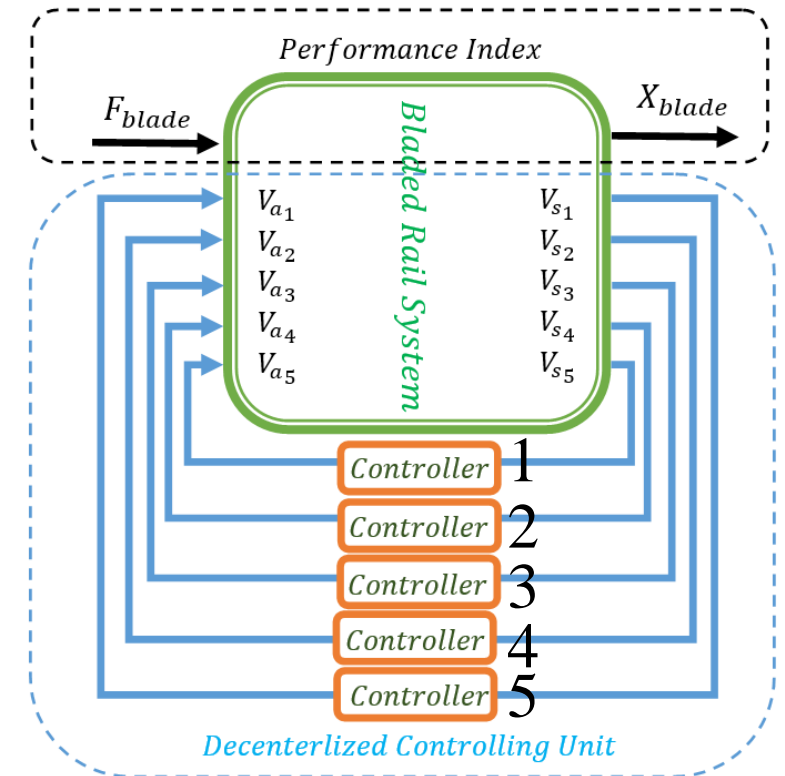
$$C(s) = + \frac{g_f \omega_f^2}{s^2 + 2\xi_f \omega_f s + \omega_f^2}$$



$$\alpha = \sqrt{4\eta^2 + 1}$$

$$\beta = \frac{4\eta^2}{4\eta^2 + 1}$$

$$\xi_f = \frac{2\eta}{\sqrt{4\eta^2 + 1}}$$



# Thank You