

Chapter 7

Page
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Alternating Current

Note → In this chapter page numbers
~~start from 33 (11 pages) to 742~~ are

33 to 742

33 to 742

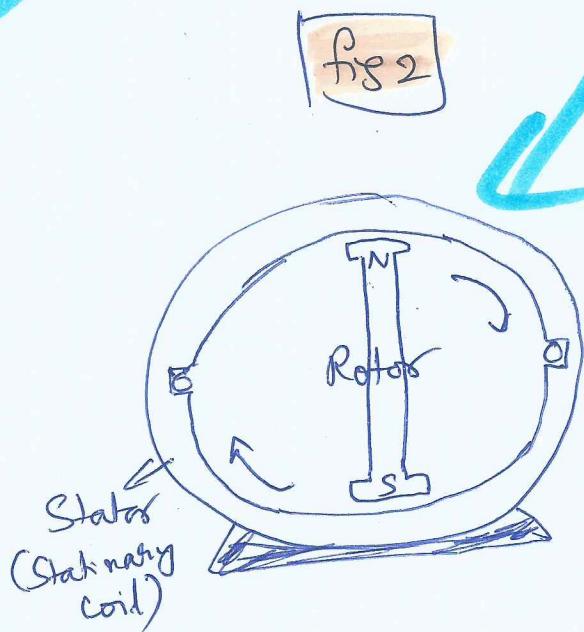
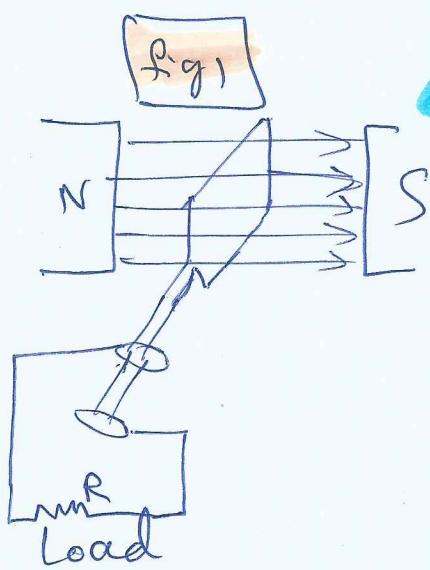
75a

- 7.1 → Induction
- 7.2 → AC Voltage applied to resistor
- 7.3 → Representation of AC current and voltage by Rotating Vectors
— called phasors
- 7.4 → AC Voltage applied to an Inductor
- 7.5 → _____ & _____ a Capacitor
- 7.6 → AC Vol. applied to a Series LCR circuit.
- 7.7 → power in Ac circuit : The power factor
- 7.8 → LC oscillations
- 7.9 → Transformers .

Alternating current Chapter (AC Chap.)

page 33 to 74

- We know that AC is generated by rotating a coil in a magnetic field (fig 1) OR by rotating a magnetic field within a stationary coil (fig 2)



It is seen that the induced emf varies as some function of the time angle ωt .

Consider a source that produces sinusoidally varying potential difference across its terminals. Let this pd also called as ac voltage be given by

$$V = V_m \sin \omega t \rightarrow 0$$

Where V_m is the peak value or the amplitude of the ac voltage and ω is its angular frequency.

Imp: $\omega = 2\pi f = \frac{2\pi}{T}$

To find the ~~ac~~ ac current, consider a loop rule $V_m \sin \omega t = i R$

$$i = \left(\frac{V_m}{R} \right) \sin \omega t$$

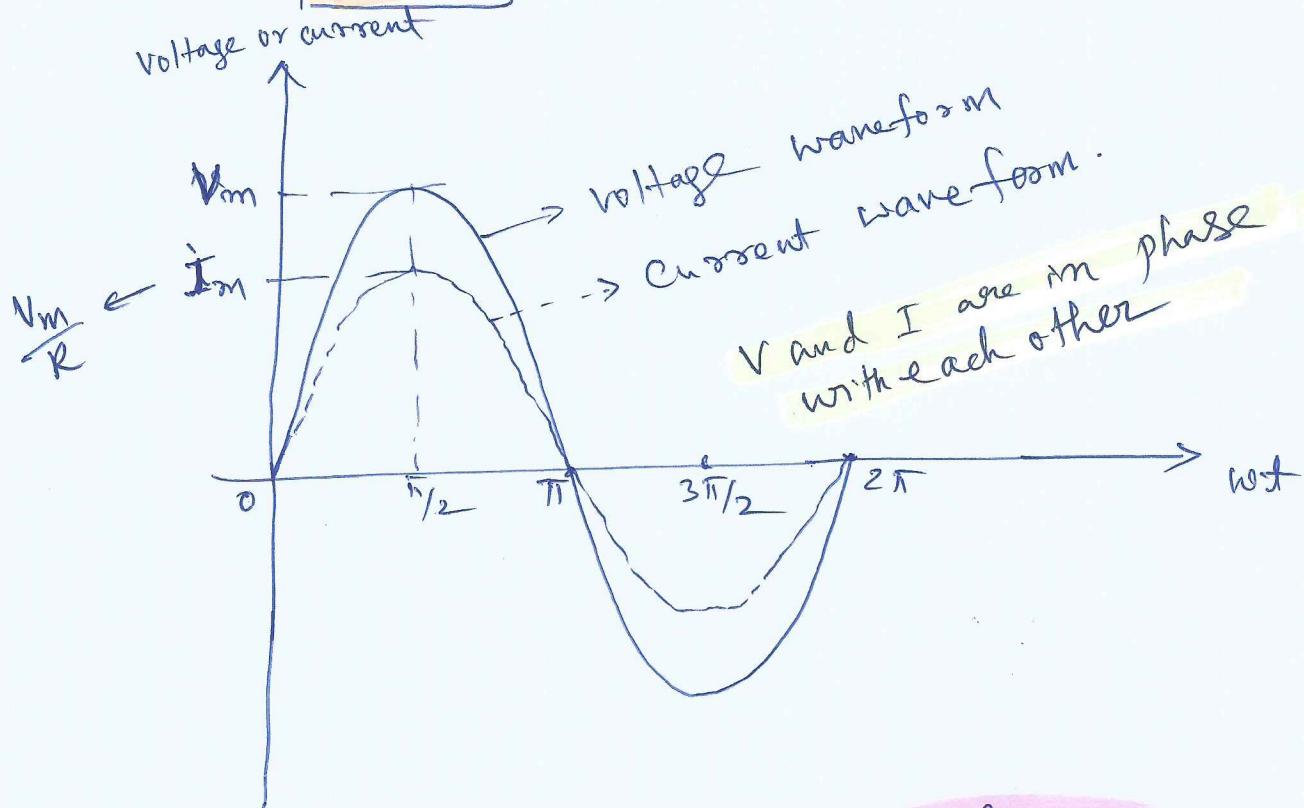
$$i = i_m \sin \omega t$$

$\rightarrow 2$

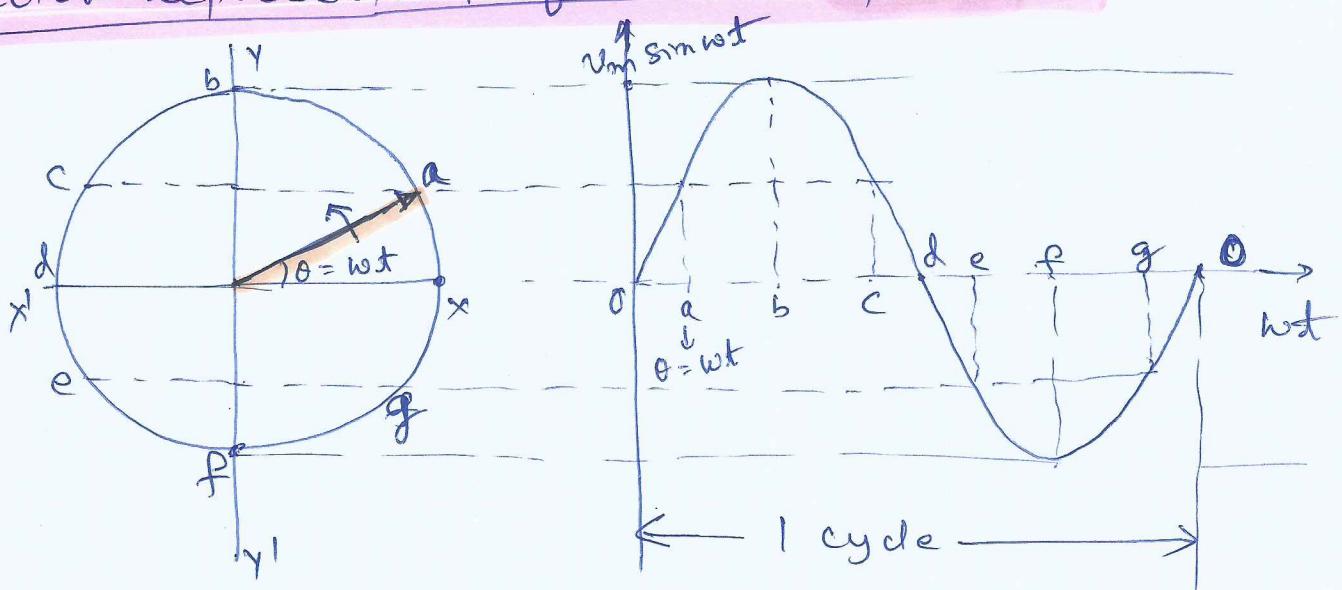
where $i_m = \frac{V_m}{R}$

P.T.O.

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Vector Representation of AC \rightarrow Phasors.



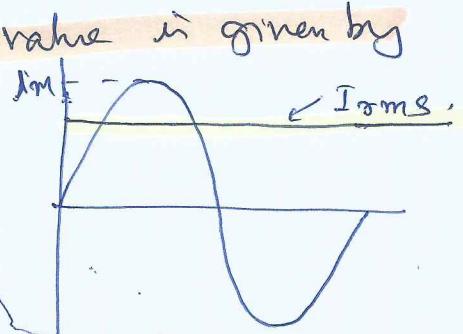
- A phasor is a vector which rotates about the origin with angular speed ω as shown above.
- To represent phase difference between AC voltage and AC current in a convenient way, we use notion of phasors. The analysis of AC circuit will be easier by the use of phasor diagram.
- From figure, the projection of voltage (or current) phasor on y-axis represents the value of voltage (or current) at that instant.

RMS value of an AC source

- Why RMS value?: ~~DC power~~ $P = I^2 R$, to express AC also in the same form as DC power, a special value of current (or voltage) is defined, which is called "root mean square" current (rms current or rms voltage)
- Relation b/w I (or V) with rms value is given by

$$V_{\text{rms}} = \frac{\cancel{0.707} V_m}{\sqrt{2}} = 0.707 V_m$$

$$I_{\text{rms}} = \frac{I_m}{\sqrt{2}} = 0.707 I_m$$



- To prove above relation:

Consider AC current (AC voltage is similar)

$$i = I_m \sin \omega t = I_m \sin \theta \rightarrow \text{the instantaneous value of current}$$

The mean of the squares of the instantaneous values of current over one complete cycle is

$$= \frac{1}{2\pi} \int_0^{2\pi} i^2 d\theta$$

Root of the above value is I_{rms}

$$\begin{aligned} I_{\text{rms}} &= \sqrt{\frac{1}{2\pi} \int_0^{2\pi} I_m^2 \sin^2 \theta d\theta} \\ &= \frac{I_m}{\sqrt{2} \cdot \sqrt{\pi}} \sqrt{\int_0^{2\pi} \left(\frac{1 - \cos 2\theta}{2} \right) d\theta} \\ &= \frac{I_m}{2\sqrt{\pi}} \sqrt{\left[\theta - \frac{\sin 2\theta}{2} \right]_0^{2\pi}} \\ &= \frac{I_m}{2\sqrt{\pi}} \sqrt{2\pi - 0 - 0 + 0} \end{aligned}$$

We know that

$$\sin^2 \theta = \frac{1 - \cos 2\theta}{2}$$

$$= \frac{I_m}{2\sqrt{\pi}} \times \sqrt{2\pi} = \frac{I_m \sqrt{2} \cdot \sqrt{\pi}}{\sqrt{2} \sqrt{2} \sqrt{\pi}}$$

$$\text{Hence } I_{\text{rms}} = \frac{I_m}{\sqrt{2}} = 0.707 I_m$$

$$V_{\text{rms}} = \frac{V_m}{\sqrt{2}} = 0.707 V_m$$

$$\therefore \text{peak value} = I_m = \sqrt{2} I_{\text{rms}}$$

$$\text{P-P Value} = 2I_m = 2\sqrt{2} I_{\text{rms}}$$

Defn: The rms value of AC is given by that steady current (dc) which when flowing through a given circuit for a given time produces the same heat as produced by the alternating current when flowing through the same circuit for the same time.

Average value of AC :

- For one complete cycle, the average value of AC = 0

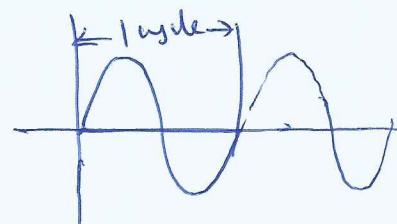
$$i = I_m \sin \theta$$

I_{average} over one complete cycle

$$I_{\text{av}} = \int_0^{2\pi} \frac{i d\theta}{2\pi}$$

$$= \frac{1}{2\pi} \int_0^{2\pi} I_m \sin \theta d\theta = \frac{I_m}{2\pi} [-\cos \theta]_0^{2\pi}$$

$$= \frac{I_m}{2\pi} [\cos 0]_0^{2\pi} = \frac{I_m}{2\pi} [1 - 1] \\ = 0$$



$$\begin{aligned} \therefore \cos 2\pi &= 1 \\ \cos 0 &= 1 \\ \cos \pi &= -1 \end{aligned}$$

$\therefore I_{\text{av}}$ over one complete cycle of AC = 0

- However, the Average Value over Half a cycle of AC is denoted as

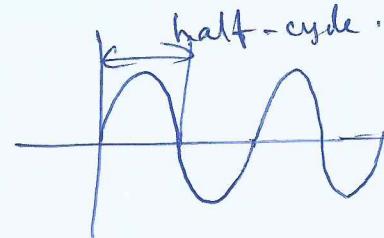
$$i = I_m \sin \theta$$

$$I_{\text{av}} \text{ over half a cycle} = \int_0^{\pi} \frac{i d\theta}{\pi}$$

$$= \frac{1}{\pi} \int_0^{\pi} I_m \sin \theta d\theta$$

$$= \frac{I_m}{\pi} [-\cos \theta]_0^{\pi} = \frac{I_m}{\pi} [\cos 0]_0^{\pi}$$

$$= \frac{I_m}{\pi} [1 - (-1)]$$



$$I_{\text{av}} \text{ over half a cycle} = \frac{2I_m}{\pi} = 0.637 I_m \quad (\pi = \frac{22}{7})$$

$\therefore 63.7\%$ of peak value.

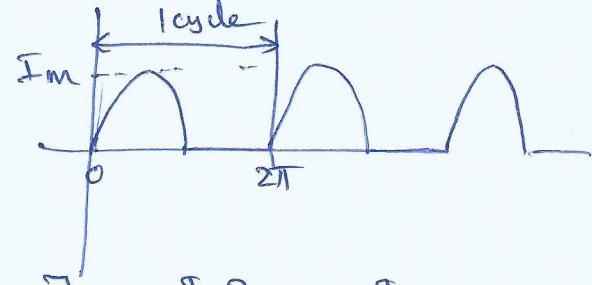
$$\begin{aligned} \int \sin x dx &= -\cos x \\ \int \cos x dx &= \sin x \\ \int e^x dx &= e^x \end{aligned}$$

Average value of Half-wave Rectified AC

$$i = I_m \sin \theta$$

$$I_{av} = \frac{1}{2\pi} \int_0^{\pi} I_m \sin \theta \, d\theta$$

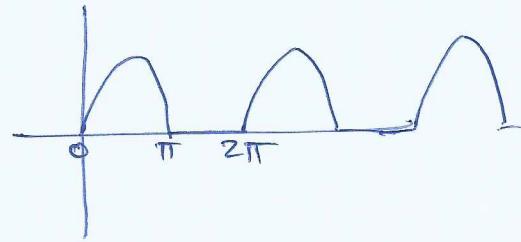
$$= \frac{I_m}{2\pi} \left[\cos \theta \right]_0^\pi = \frac{I_m}{2\pi} [1 - (-1)] = \frac{I_m 2}{2\pi} = \frac{I_m}{\pi}$$



$$\boxed{I_{av} = \frac{I_m}{\pi}} = 0.318 I_m \quad (31.8\% \text{ of peak value})$$

RMS value of Half-wave Rectified AC

$$\begin{aligned} I_{RMS} &= \sqrt{\frac{1}{2\pi} \int_0^{\pi} i^2 \, d\theta} \\ &= \sqrt{\frac{I_m^2}{2\pi} \int_0^{\pi} \sin^2 \theta \, d\theta} \\ &= \sqrt{\frac{I_m^2}{2\pi} \int_0^{\pi} \frac{(1 - \cos 2\theta)}{2} \, d\theta} \\ &= \sqrt{\frac{I_m^2}{4\pi} \left[\theta - \frac{\sin 2\theta}{2} \right]_0^\pi} \\ &= \sqrt{\frac{I_m^2}{4\pi} [(\pi - 0) - (0 - 0)]} \\ &= \sqrt{\frac{I_m^2 \pi}{4\pi}} = \frac{I_m}{2} \end{aligned}$$



$$\therefore \sin^2 \theta = \frac{(1 - \cos 2\theta)}{2}$$

$$\begin{array}{l} \therefore \sin 2\pi = 0 \\ \therefore \sin 0 = 0 \end{array}$$

$$\boxed{I_{av} = \text{RMS value of Half-wave rectified AC current} = \frac{I_m}{2} = 0.5 I_m}$$

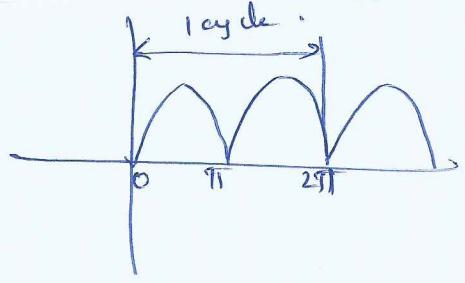
Average value of Full-wave Rectified current

$$i = I_m \sin \theta$$

$$i_{av} = \frac{2 \times \int_0^{\pi} I_m \sin \theta \, d\theta}{2\pi}$$

$$= \frac{I_m}{\pi} \int_0^{\pi} \sin \theta \, d\theta = \frac{I_m}{\pi} (\cos \theta) \Big|_0^{\pi} \\ = \frac{I_m}{\pi} [1 - (-1)] = \frac{2 I_m}{\pi}$$

i_{av} of full-wave rectified AC current = $\frac{2}{\pi} I_m = 0.637 I_m$
 (63.7% of peak current)



Avg RMS value of Full-wave Rectified current

$$(i = I_m \sin \theta)$$

$$I_{RMS} = \sqrt{2 \times \int_0^{\pi} \frac{i^2 \, d\theta}{2\pi}}$$

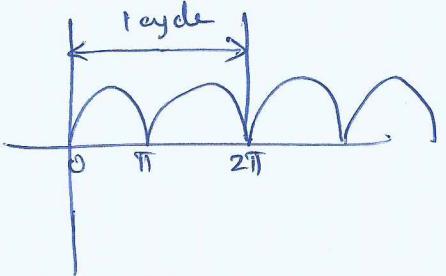
$$= \sqrt{\frac{I_m^2}{\pi} \int_0^{\pi} \sin^2 \theta \, d\theta}$$

$$= \sqrt{\frac{I_m^2}{\pi} \int_0^{\pi} \frac{(1 - \cos 2\theta)}{2} \, d\theta}$$

$$= \sqrt{\frac{I_m^2}{2\pi} \left[\theta - \frac{\sin 2\theta}{2} \right]_0^{\pi}}$$

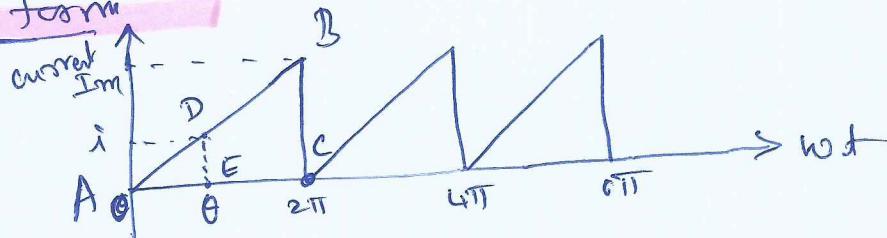
$$= \sqrt{\frac{I_m^2}{2\pi} [\pi - 0] - 0(0 - 0)}$$

$$= \sqrt{\frac{I_m^2 \pi}{2\pi}} = \frac{I_m}{\sqrt{2}}$$



$$I_{RMS} = \frac{I_m}{\sqrt{2}} = 0.707 I_m = \boxed{\text{Same as normal sinusoidal current.}}$$

Consider Waveform



① Average value of current ?

By symmetry, we can conclude it is $\frac{I_m}{2}$

Proof: From figure

$$\frac{DE}{AE} = \frac{BC}{AC} = \frac{I_m}{2\pi}$$

Consider current i at an angle θ , then

$$\frac{DE}{AE} = \frac{i}{\theta} = \frac{I_m}{2\pi} \therefore i = \frac{I_m \theta}{2\pi}$$

This is equation of current for one cycle.

$$I_{av} = \frac{I_m}{2\pi} \int_0^{2\pi} \frac{\theta}{2\pi} d\theta = \frac{I_m}{4\pi^2} \left[\frac{\theta^2}{2} \right]_0^{2\pi} = \frac{4\pi^2}{4\pi^2} \frac{I_m}{2}$$

$I_{av} = I_m/2$

② RMS value of above waveform :

Current equation of the waveform :

It is a straight line with slope m $\therefore y = mx + c$

$$i = \frac{I_m}{2\pi} (\omega t)$$

let $\omega t = \theta$ radians

$$i = \frac{I_m}{2\pi} \theta$$

$$i^2 = \frac{I_m^2}{4\pi^2} \theta^2$$

$$\text{Mean Square Value} = \frac{1}{2\pi} \int_0^{2\pi} i^2 d\theta = \frac{1}{2\pi} \int_0^{2\pi} \frac{I_m^2}{4\pi^2} \theta^2 d\theta$$

$$= \frac{I_m^2}{8\pi^3} \left[\frac{\theta^3}{3} \right]_0^{2\pi}$$

$$= \frac{I_m^2 \times 8\pi^3}{8\pi^3 \times 3}$$

$$= \frac{I_m^2}{3}$$

$$\therefore \text{RMS} = \sqrt{\frac{I_m^2}{3}}$$

$$= \boxed{\frac{I_m}{\sqrt{3}}}$$

IMP

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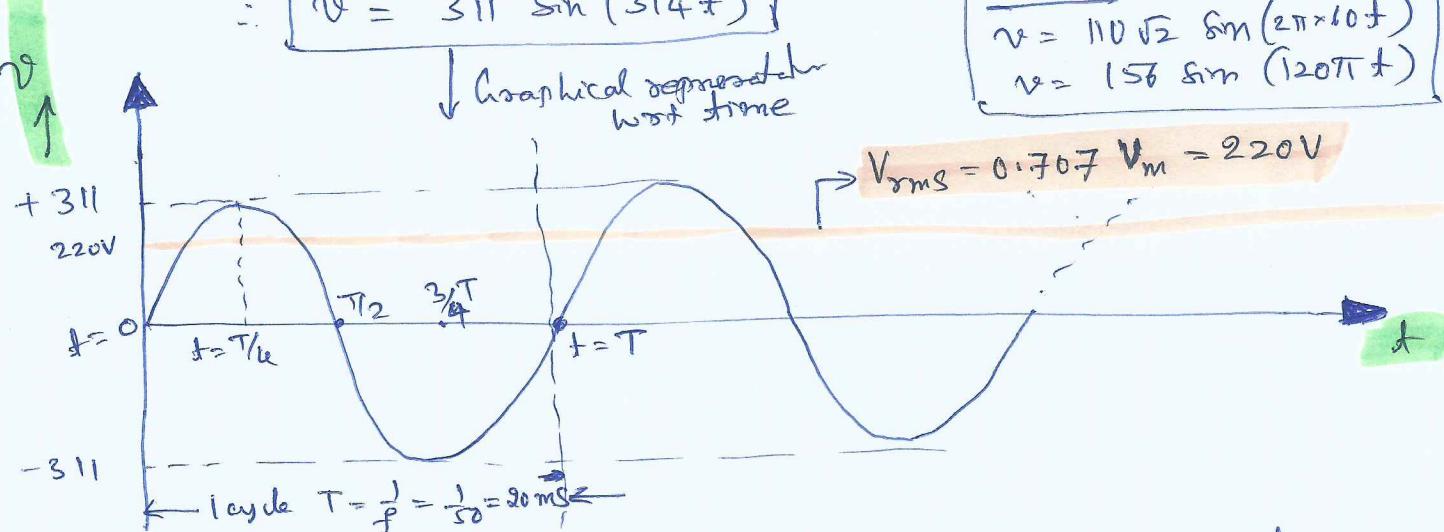
- ① The electric mains in a house in India is marked 220 V - 50 Hz. Write the eqn for instantaneous voltage.

EG. for Instantaneous voltage $v = V_m \sin(\omega t)$
 $= V_m \sin(2\pi f t)$

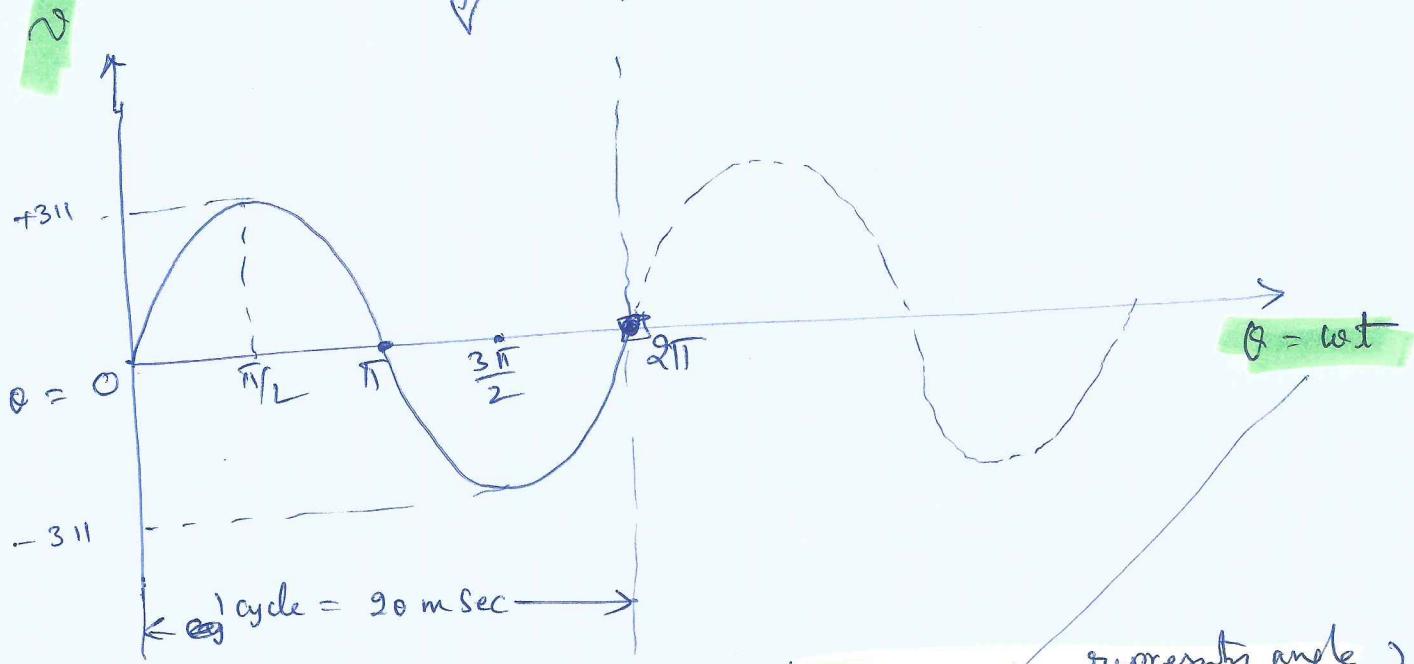
Given $V_{rms} = 220V \quad \therefore V_m = 220\sqrt{2} = 311V$
 $f = 50Hz \quad \therefore 2\pi f = 100\pi = 314$

$\therefore v = 311 \sin(314t)$

For VS
 $v = 110\sqrt{2} \sin(2\pi \times 50t)$
 $v = 156 \sin(314t)$



Graphical Representation not degrees angle



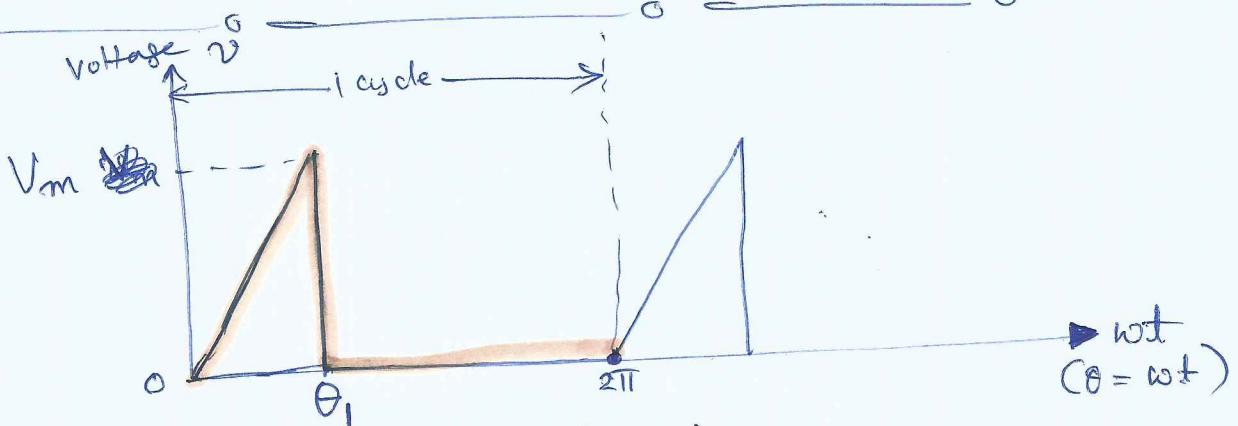
How (ωt) is represents angle?

$$\begin{aligned} \omega t &= 2\pi f t && \text{in Sec} \\ &= \frac{2\pi}{T} t && \text{in S} \end{aligned}$$

$\omega t = 2\pi$ rad.
 \rightarrow represents angle

RMS value of Δ^{1st} waveform

- What is the RMS value of a periodic signal?
→ When a periodic signal is generated by a source connected to a load (e.g. Resistor), the RMS value is the continuous ~~fixed~~ DC value which would deliver the same power to the load as the periodic signal.



Equation of the above waveform:

$$v = \frac{V_m}{\theta_1} (\omega t) \quad \text{or} \quad v = \frac{V_m}{\theta_1} \theta \quad \rightarrow \textcircled{1} \quad 0 \leq \theta \leq \theta_1$$

① is the linear function of signal and is repeated every 2π angle.

- Square of ① $v^2 = \frac{V_m^2}{\theta_1^2} \theta^2$
- Mean $\rightarrow \frac{1}{2\pi} \int_0^{\theta_1} v^2 d\theta = \frac{1}{2\pi} \int_0^{\theta_1} \frac{V_m^2}{\theta_1^2} \theta^2 d\theta$
- Square root of above $V_{\text{rms}} = \sqrt{\frac{1}{2\pi} \int_0^{\theta_1} \frac{V_m^2}{\theta_1^2} \theta^2 d\theta}$

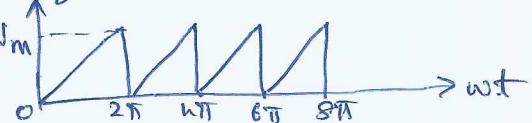
$$= \sqrt{\frac{V_m^2}{2\pi \theta_1^2} \left[\frac{\theta^3}{3} \right]_0^{\theta_1}} = \sqrt{\frac{V_m^2}{2\pi \theta_1^2} \frac{\theta_1^3}{3}}$$

$$= \frac{V_m}{\sqrt{2} \sqrt{\pi} \theta_1} \cdot \frac{\sqrt{\theta_1}}{\sqrt{3}}$$

$$V_{\text{rms}} = \frac{V_m}{\sqrt{3}} \sqrt{\frac{\theta_1}{2\pi}}$$

Hence $\frac{\theta_1}{2\pi}$ is the duty-cycle $\rightarrow \textcircled{2}$

In ② if duty-cycle is 100% $\Rightarrow \theta_1 = 2\pi$

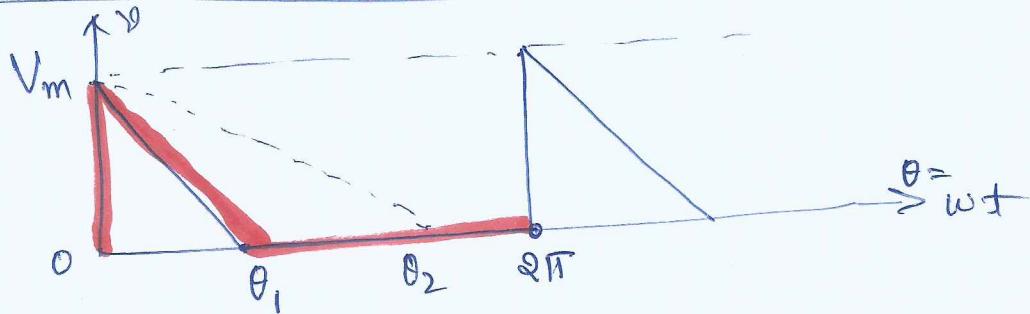


② becomes

$$V_{\text{rms}} = \frac{V_m}{\sqrt{3}} \rightarrow \textcircled{3}$$

P.T.O.

Consider following waveform



θ is a running value on x -axis.

Equation for waveform (slope @ $\theta_1 = -\frac{V_m}{\theta_1}$)

$$v = \left(-\frac{V_m}{\theta_1} \right) (\theta - \theta_1)$$

$$\boxed{v = \frac{V_m}{\theta_1} (\theta_1 - \theta)}$$

$$0 \leq \theta < \theta_1$$

$$\bullet \quad v^2 = \frac{1}{2\pi} \int_0^{\theta_1} \frac{V_m^2}{\theta_1^2} (\theta_1 - \theta)^2 d\theta$$

$$= \frac{V_m^2}{2\pi \theta_1^2} \int_0^{\theta_1} x^2 dx = \frac{V_m^2}{2\pi \theta_1^2} \left[\frac{x^3}{3} \right]_0^{\theta_1} = \frac{V_m^2 \theta_1^3}{2\pi \cdot 3}$$

$$\boxed{\text{Let } \theta_1 - \theta = x}$$

$$V_{rms} = \sqrt{\frac{V_m^2}{2\pi \cdot 3}} \theta_1$$

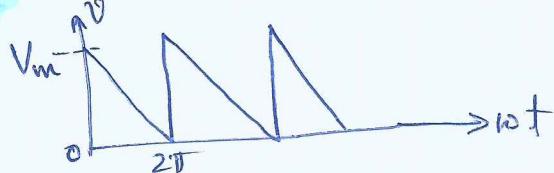
$$\boxed{V_{rms} = \frac{V_m}{\sqrt{3}} \cdot \sqrt{\frac{\theta_1}{2\pi}}} \rightarrow \textcircled{4}$$

$\frac{\theta_1}{2\pi}$ is duty cycle

for 100% duty cycle

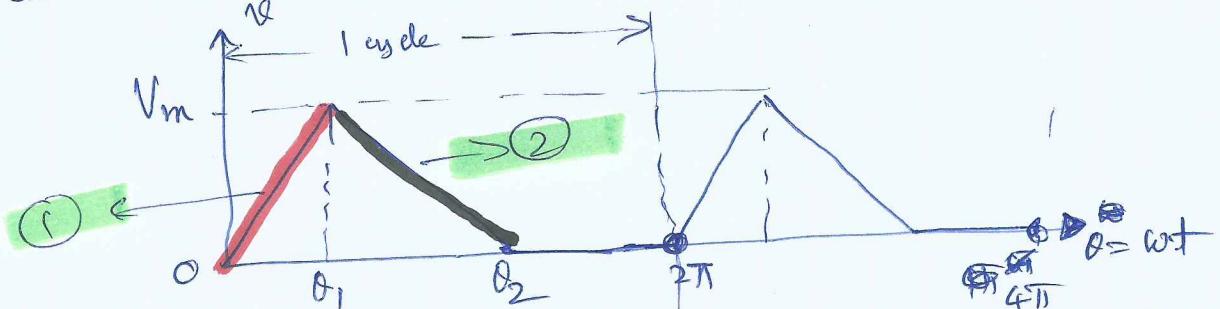
$$\theta_1 = 2\pi$$

$$\boxed{V_{rms} = \frac{V_m}{\sqrt{3}}}$$



PTO.

Consider wave form



Two parts in above wave form (as shown by red and black)

$$\text{ESM for } ① \quad v_1 = \frac{V_m}{\theta_1} \theta \quad \rightarrow 0 \leq \theta < \theta_1$$

$$\begin{aligned} \text{eq2 for } ② \quad v_2 &= \frac{V_m}{\theta_2 - \theta_1} (\theta_2 - \theta) \\ &= V_m \left(\frac{\theta_2 - \theta_1}{\theta_2 - \theta_1} \right) \quad \theta_1 \leq \theta < \theta_2 \end{aligned}$$

From eq ② in page ④, we get

$$v_1^2 (\text{rms}) = \frac{V_m^2 \theta_1}{6\pi} \rightarrow ⑤$$

From eq ④ in page 42

$$v_2^2 (\text{rms}) = \frac{V_m^2 (\theta_2 - \theta_1)}{6\pi} \rightarrow ⑥$$

Adding ⑤ + ⑥

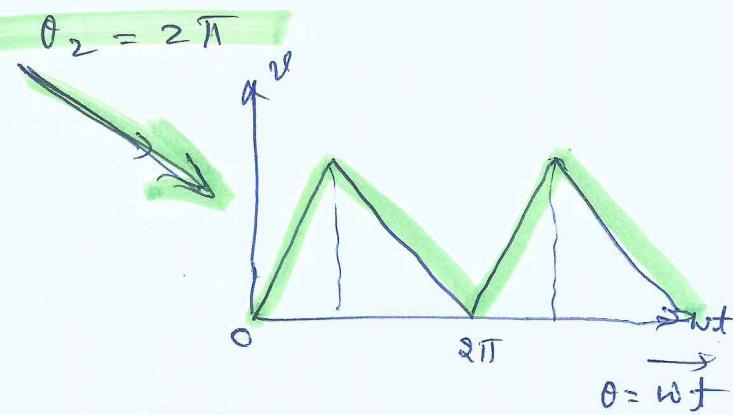
$$v_{\text{rms}}^2 = \frac{V_m^2 \theta_1}{6\pi} + \frac{V_m^2 (\theta_2 - \theta_1)}{6\pi}$$

$$v_{\text{rms}} = \sqrt{v_{\text{rms}}^2} = \sqrt{\frac{V_m^2 \theta_1}{6\pi} + \frac{V_m^2 (\theta_2 - \theta_1)}{6\pi}} = \sqrt{\frac{V_m^2 \theta_1}{6\pi} + \frac{V_m^2 \theta_2}{6\pi} - \frac{V_m^2 \theta_1}{6\pi}}$$

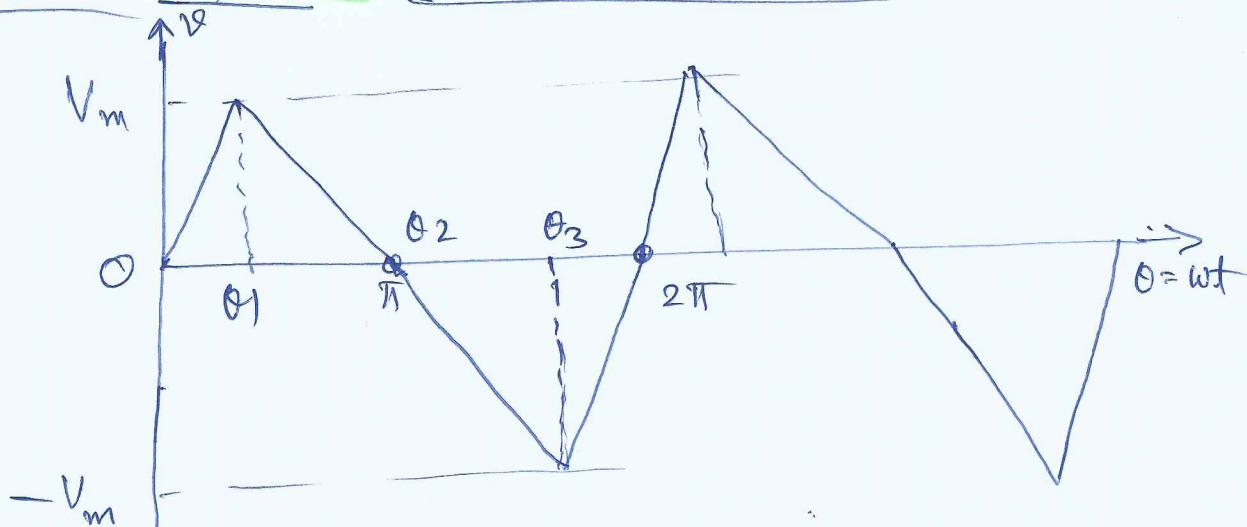
$$v_{\text{rms}} = \frac{V_m \sqrt{\theta_2}}{\sqrt{6\pi}} = \frac{V_m \sqrt{\theta_2}}{\sqrt{3} \sqrt{2\pi}} = \frac{V_m}{\sqrt{3}} \sqrt{\frac{\theta_2}{2\pi}} \rightarrow ⑦$$

If Duty cycle = 100%, $\theta_2 = 2\pi$

$$v_{\text{rms}} = \frac{V_m}{\sqrt{3}}$$



Consider bipolar Δ^{le} as shown below



- From 7 page 43, RMS value from 0 to θ_2 is known.

$$V_{\text{rms}}(0 \text{ to } \theta_2) = \frac{V_m}{\sqrt{3}} \sqrt{\frac{\theta_2}{2\pi}} \rightarrow 7$$

- RMS value from θ_2 to 2π is same as 0 to θ_2 , with the difference that we need to replace θ_2 with $(2\pi - \theta_2)$

$$V_{\text{rms}}(\theta_2 \text{ to } 2\pi) = \frac{V_m}{\sqrt{3}} \sqrt{\frac{2\pi - \theta_2}{2\pi}} \rightarrow 8$$

- The reason is it does not matter whether the signal is positive or negative, the power delivered to the load is the same.

$$\therefore V_{\text{rms_triple-bipolar}} = \sqrt{V_{\text{rms}}(0 \text{ to } \theta_2)^2 + V_{\text{rms}}(\theta_2 \text{ to } 2\pi)^2}$$

From 7 and 8

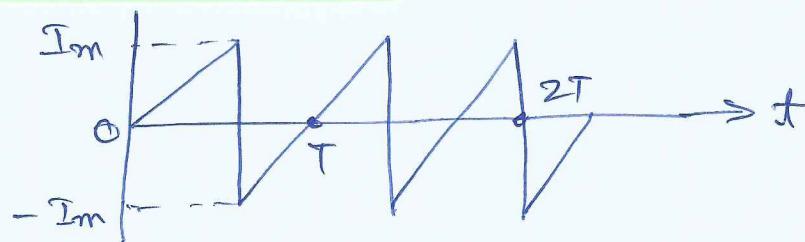
$$V_{\text{rms_triple-bipolar}} = \sqrt{\frac{V_m^2}{3} \frac{\theta_2}{2\pi} + \frac{V_m^2}{3} \frac{(2\pi - \theta_2)}{2\pi}}$$

$$= \sqrt{\frac{V_m^2}{3} \cdot \frac{\theta_2}{2\pi} + \frac{V_m^2 \cdot 2\pi}{3 \cdot 2\pi} - \frac{V_m^2 \cdot \theta_2}{3 \cdot 2\pi}}$$

$$V_{\text{rms_t-b}} = \frac{V_m}{\sqrt{3}} \rightarrow 9$$

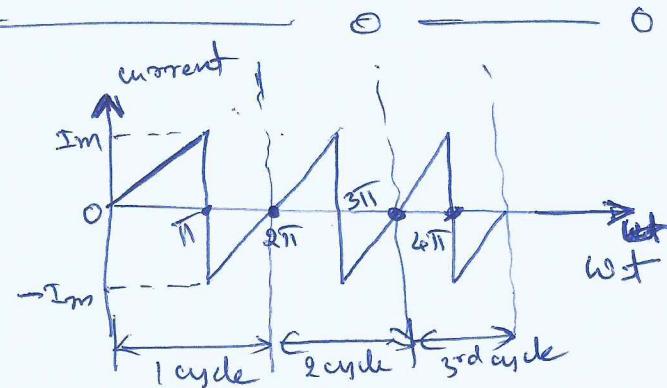
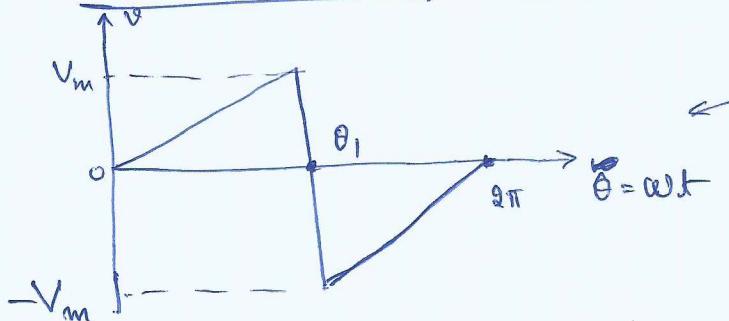
P.T.O.

Consider waveform



- ① Average value = Average current = 0 since half cycle is +ve and other half cycle current is negative

- ② RMS value of current



This is same as in page 44

$$V_{rms} (0 \text{ to } \theta_1) = \frac{V_m}{\sqrt{3}} \sqrt{\frac{\theta_1}{2\pi}} \rightarrow 10$$

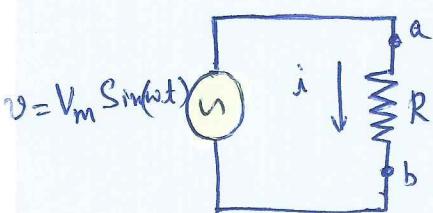
$$V_{rms} (\theta_1 \text{ to } 2\pi) = \frac{V_m}{\sqrt{3}} \sqrt{\frac{2\pi - \theta_1}{2\pi}} \rightarrow 11$$

$$\begin{aligned} V_{rms\text{-bipolar}} &= \sqrt{\frac{V_m^2}{3} \frac{\theta_1}{2\pi} + \frac{V_m^2}{3} \cdot \frac{(2\pi - \theta_1)}{2\pi}} \\ &= \sqrt{\frac{V_m^2 \cdot \theta_1}{3 \cdot 2\pi} + \frac{V_m^2 \cdot 2\pi}{3 \cdot 2\pi} - \frac{V_m^2 \cdot \theta_1}{3 \cdot 2\pi}} \end{aligned}$$

$$V_{rms\text{-bipolar}} = \frac{V_m}{\sqrt{3}}$$

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Resistor in an AC-Circuit



Using Kirchhoff's loop rule, $v = iR$

$$\therefore V_m \sin wt = iR$$

$$\therefore i = \left[\frac{V_m}{R} \right] \sin wt$$

$$\therefore i = I_m \sin wt$$

where $I_m = \frac{V_m}{R}$

and $v = V_m \sin wt$

Since V_m and R are constants for a given waveform, the shape of the voltage and the current will be sinusoidal.

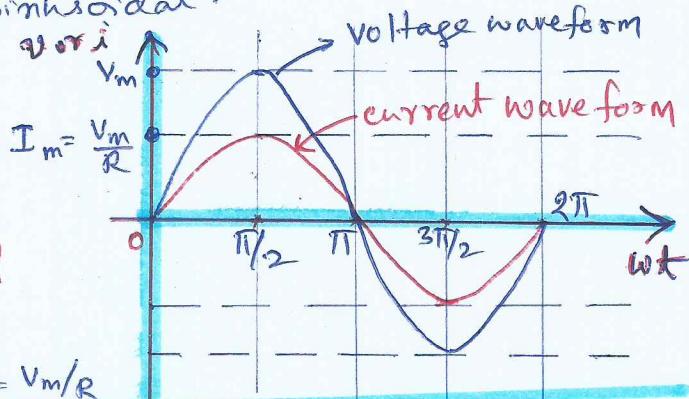
• **v and i are in phase**

⇒ When v goes to peak, i also peaks; When $v=0$, $i=0$.

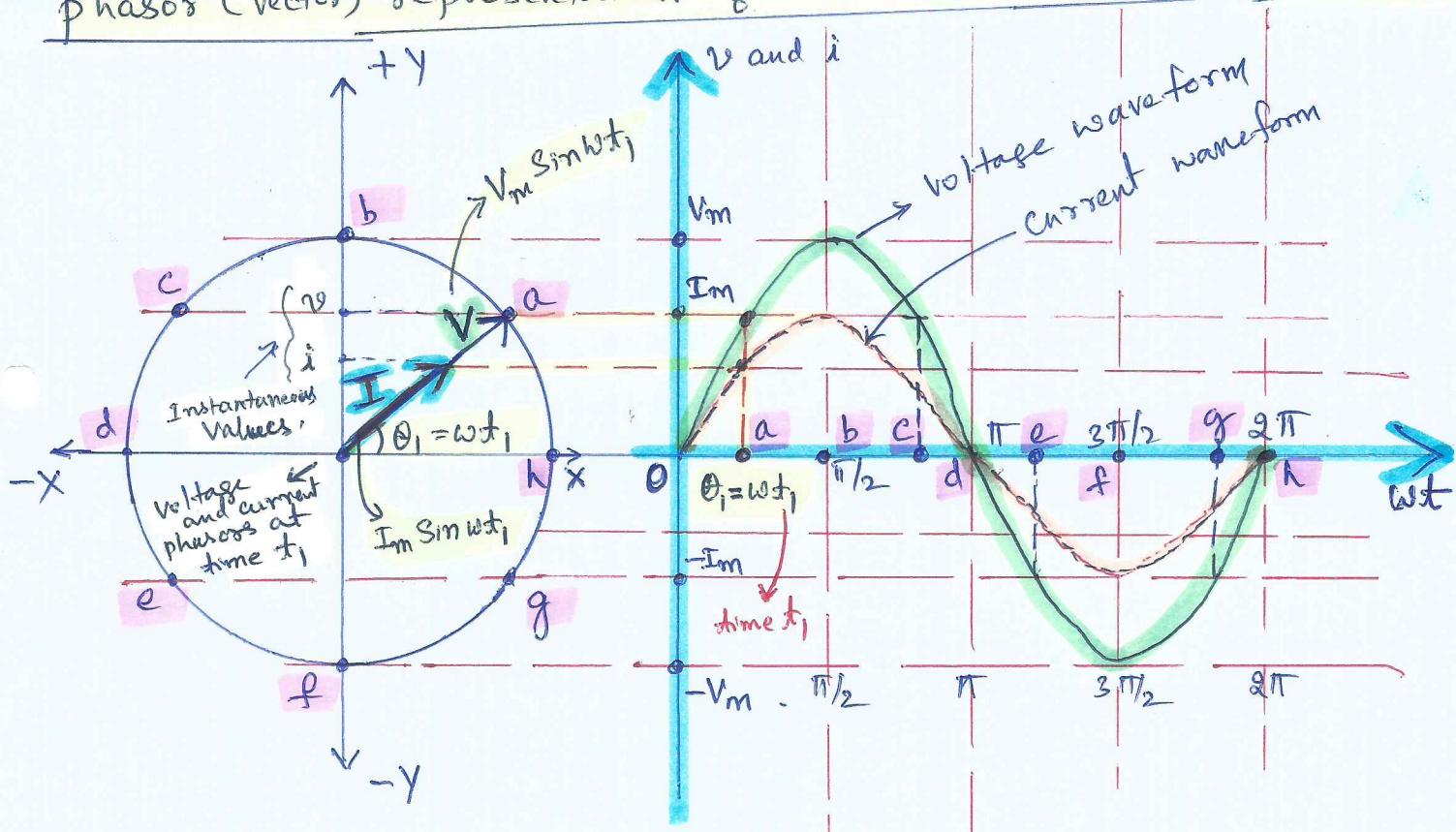
→ only difference in amplitude of current < amplitude of voltage (based on selected R)

→ V_m = peak value of voltage waveform

→ I_m = peak value of current waveform = V_m/R



phasor (vector) representation of Resistor in AC circuit (NEET book)



This diagram shows that the phase difference between I and V is zero.

Simplified diagram in the next page.