# HW3 Solutions

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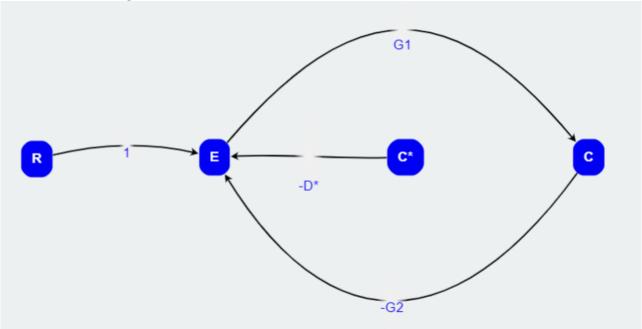
۱ بخش اجباری

#### Disclaimer:

This is the solution manual to the homework assigned to students of Digital Control - Dr. Talebi. We do not guarantee that this solution is precise and thorough so please contact your TA to propose your innovative solutions and/or any probable mistakes.

سوال اول

First draw the original SFG :



$$C = G_1 E$$

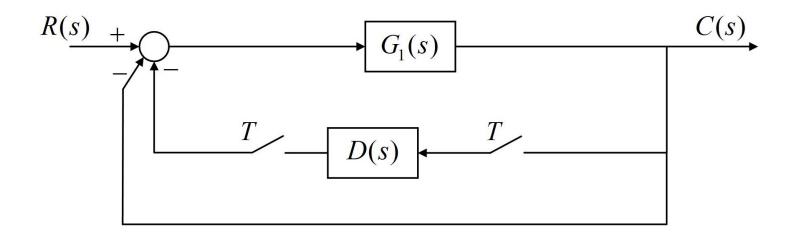
$$E = R - G_2 C - D^* C^*$$

$$C = \frac{G_1 R}{G_1 G_2 + 1} - \frac{D^* C^*}{1 + G_1 G_2}$$
Derive  $C^*$  as follows:
$$C^* = \left(\frac{G_1 R}{1 + G_1 G_2}\right)^* - \left(\frac{1}{1 + G_1 G_2}\right)^* D^* C^*$$

$$C^* = \frac{\left(\frac{G_1 R}{1 + G_1 G_2}\right)^*}{\left(1 + \left(\frac{1}{G_1 G_2}\right)^* D^*\right)}$$

$$C = \frac{G_1 R}{G_1 G_2 + 1} - \frac{D^* \left(\frac{G_1 R}{1 + G_1 G_2}\right)^*}{1 + G_2 G_2}$$

HW3 Solutions ٣



سوال دوم

Recall the equivalent for forward discretization:

$$s \to \frac{z-1}{T}$$

we must substitute it in the transfer function to get the following discrete transfer function:

$$H(z)_{forward} = 10 \frac{z - 0.75}{z + 1.5}$$

 $H(z)_{forward}=10\frac{z-0.75}{z+1.5}$  from the original transfer function we have:

$$\frac{s+1}{0.1s+1} \xrightarrow{s \to j\omega} \frac{j\omega + 1}{0.1j\omega + 1} \frac{j\omega + 1}{0.1j\omega + 1} = \frac{\sqrt{\omega^2 + 1}}{\sqrt{(0.1\omega)^2 + 1}} \angle \tan^{-1}\omega - \tan^{-1}0.1\omega$$

$$\omega = 3 \to \varphi = \tan^{-1}3 - \tan^{-1}0.3 \simeq 55^{\circ}$$

$$H(z)_{forward} = 10 \frac{z - 0.75}{z + 1.5} \xrightarrow{z = e^{j\omega T}} 10 \frac{e^{j\omega T} - 0.75}{e^{j\omega T} + 1.5}$$

$$10 \frac{\cos \omega T - 0.75 + j \sin \omega T}{\cos \omega T + 1.5 + j \sin \omega T}$$

$$\varphi = \tan^{-1}\frac{\sin \omega T}{\cos \omega T - 0.75} - \tan^{-1}\frac{\sin \omega T}{\cos \omega T + 1.5}$$

$$\varphi = \tan^{-1}\frac{\sin 0.75}{\cos 0.75 - 0.75} - \tan^{-1}\frac{\sin 0.75}{\cos 0.75 + 1.5}$$

$$\varphi = \tan^{-1}\frac{.68}{.0183} - \tan^{-1}\frac{0.68}{0.23168} = 88.46 - 16.94 = 71.52$$
Discretize using the Backward rule to get:

$$H(z)_{forward} = 10 \frac{z - 0.75}{z + 1.5} \xrightarrow{z = e^{j\omega T}} 10 \frac{e^{j\omega T} - 0.75}{e^{j\omega T} + 1.5}$$

$$10\frac{\cos\omega T + 1.5 + j\sin\omega T}{\cos\omega T + 1.5 + j\sin\omega T}$$

$$\varphi = \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.75} - \tan^{-1} \frac{\sin \omega T}{\cos \omega T + 1.5}$$

$$\varphi = \tan^{-1} \frac{\sin 0.75}{\cos 0.75 - 0.75} - \tan^{-1} \frac{\sin 0.75}{\cos 0.75 + 1.5}$$

$$\varphi = \tan^{-1} \frac{.68}{.0183} - \tan^{-1} \frac{0.68}{0.23168} = 88.46 - 16.94 = 71.52$$

Discretize using the Backward rule to get:

$$s \to \frac{z-1}{Tz}$$

$$H(z)_{Backward}^{1z} = 3.57 \frac{z - 0.8}{z - 0.286}$$

$$H(z) = 3.57 \frac{z - 0.8}{z - 0.286} \xrightarrow{z \to e^{j\omega T}} 3.57 \frac{e^{j\omega T} - 0.8}{e^{j\omega T} - 0.286}$$

$$3.57 \frac{\cos \omega T - 0.8 + j \sin \omega T}{\cos \omega T - 0.286 + j \sin \omega T}$$

$$3.57 \frac{\cos \omega T - 0.286}{\cos \omega T - 0.88 + j \sin \omega T}$$

$$\varphi = \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.286} - \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.286}$$

$$\varphi = \tan^{-1} \frac{\sin 0.75}{\cos 0.75 - 0.8} - \tan^{-1} \frac{\sin 0.75}{\cos 0.75 - 0.286}$$

$$\varphi = \tan^{-1} \frac{0.68}{0.0683} - \tan^{-1} \frac{0.68}{0.4457}$$

$$\varphi = \tan^{-1} \frac{0.68}{0.0683} - \tan^{-1} \frac{0.68}{0.4457}$$

$$\varphi \simeq 40$$

سوال سوم

Recall the formula for tustin with prewarping:

$$s \to \frac{1}{\tan(\frac{\omega_1 T}{2})} \frac{z-1}{z+1}$$

we must substitute it in the transfer function to get the following discrete transfer function:

$$H(z) = 4.89 \frac{z - 0.768}{z - 0.161}$$

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$$H(z) = 4.89 \frac{z - 0.768}{z + 0.161} \xrightarrow{z \to e^{j\omega T}} 4.89 \frac{e^{j\omega T} - 0.768}{e^{j\omega T} + 0.161}$$

$$4.89 \frac{\cos \omega T - 0.768 + j \sin \omega T}{\cos \omega T + 0.161 + j \sin \omega T}$$

$$\varphi = \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.768} - \tan^{-1} \frac{\sin \omega T}{\cos \omega T + 0.161}$$

$$\varphi = \tan^{-1} \frac{0.68}{0.363} - \tan^{-1} \frac{0.68}{0.8926}$$

$$\varphi = -86.944 - 37.3$$

$$\varphi = -55.756$$

$$4.89 \frac{\cos \omega T - 0.768 + j \sin \omega T}{\cos \omega T + 0.161 + i \sin \omega T}$$

$$\varphi = \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.768} - \tan^{-1} \frac{\sin \omega T}{\cos \omega T + 0.161}$$

$$\varphi = \tan^{-1} \frac{0.68}{0.363} - \tan^{-1} \frac{0.68}{0.8926}$$

$$\varphi = -86.944 - 37.3$$

$$\varphi = 55.756$$

Recall the equivalent formula for zero pole matching. Substitute it to get the discrete transfer function as bellow:

$$H_{PZ} = 4.150 \frac{z - 0.779}{z - 0.082}$$

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$$H_{PZ} = 4.150 \frac{z - 0.779}{z - 0.082} \xrightarrow{z \to e^{j\omega T}} 4.150 \frac{e^{j\omega T} - 0.779}{e^{j\omega T} - 0.082}$$

$$4.150 \frac{\cos \omega T - 0.779 + j \sin \omega T}{\cos \omega T - 0.082 + j \sin \omega T}$$

$$\varphi = \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.779} + \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.082}$$

$$\varphi = \tan^{-1} \frac{0.68}{-0.047} - \tan^{-1} \frac{0.68}{0.6496}$$

$$\varphi = -86.046 - 46.3097$$

$$4.150\frac{\cos\omega T - 0.779 + j\sin\omega T}{\cos\omega T - 0.082 + j\sin\omega T}$$

$$\varphi = \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.779} + \tan^{-1} \frac{\sin \omega T}{\cos \omega T - 0.082}$$

$$\varphi = \tan^{-1} \frac{0.68}{-0.047} - \tan^{-1} \frac{0.68}{0.6496}$$

$$\varphi = -86.046 - 46.3097$$

$$\varphi = 47.64$$

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Looking back at the calculations of section 4.3.3 of Digital Control of Dynamic Systems by G.Franklin:

Let's choose T=1

$$x(k+1) = \Phi x(k) + \Gamma u(k)$$
  

$$y(k) = Hx(k)$$
(1)

$$\begin{split} \Phi &= e^{AT} \\ \Gamma &= \int\limits_0^T e^{A\eta} d\eta B \\ \Phi &= L^{-1} \left\{ (sI - AT)^{-1} \right\} \\ AT &= \begin{pmatrix} 0 & 1 \\ -1 & -1 \end{pmatrix} \\ \Phi &= L^{-1} \left\{ \begin{pmatrix} \frac{s+1}{s^2+s+1} & \frac{-1}{s^2+s+1} \\ \frac{1}{s^2+s+1} & \frac{s+1.57 \times 10^{-16}}{s^2+s+1} \end{pmatrix} \right\} \end{split}$$

The rest is routine and left as an exercise to students.

سوال ششم

Follow section 6.1, equations 6.27 through 6.31 of Digital Control of dynamic systems by G.Franklin for the solution.

## بخش امیتازی

سوال هفتم

#### Solution:

To obtain  $V_o$ , we will replace the R-2R network with its Thevenin equivalent circuit, as shown in Fig. 2. The computation of the Thevenin resistance and Thevenin voltage is shown in Figs. 3 and 4, respectively, where repeated use of Thevenin's theorem has been made to simplify the circuit systematically.

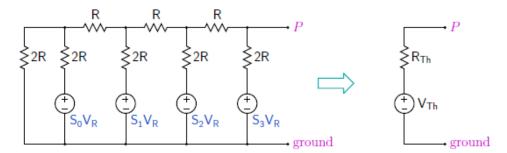
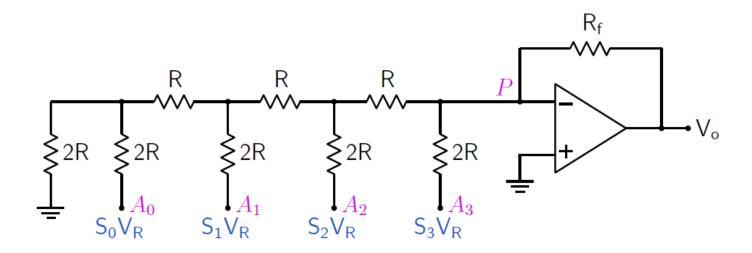


Figure 2: Representation of R-2R ladder with Thevenin equivalent circuit.

Finally, we replace the R-2R network in the original DAC circuit of Fig. 1 with its Thevenin equivalent circuit and obtain the circuit shown in Fig. 5. This circuit is simply an inverting amplifier, and the output voltage is given by

$$V_o = -\frac{R_f}{R}V_{\text{Th}} = -\frac{V_R}{8} = -0.625 \,\text{V},$$
 (1)

Y HW3 Solutions



شكل ٢: شكل سوال هفتم

### سوال هشتم

This task must be done in multiple steps:

Firstly, compare 10 which is half of 20 with the input voltage:

10 < 13.478

MSB Gets Set we have : 10000000

10 + 5 > 13.478

the next bit remains 0:10000000

10 + 2.5 < 13.478

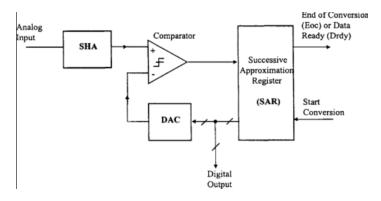
the next bit gets set: 1010000

10 + 2.5 + 1.25 > 13.478

the next bit remains 0:10100000

10 + 2.5 + 0.625 < 13.478

the next bit gets set: 10101000 10 + 2.5 + 0.625 + 0.3125 < 13.478the next bit gets set: 10101100 10 + 2.5 + 0.625 + 0.3125 + 0.15625the next bit remains 0: 10101100



شكل ٣: شكل سوال هشتم