Speech Recognition Speech signal descriptors

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Course information



Schedule

- Meeting time: Wed 2pm-4pm, E110
- Lectures/labs interleave actual schedule tba

Course slides (supplementary material)

 Accessible from kslot.iis.p.lodz.pl under menu: Teaching - Courses in English – Speech Recognition

Assessment

- Test
- Lab projects

Contact information

- kslot@p.lodz.pl
- Office hours: Wednesdays, 3pm-4pm, building C7, room 6

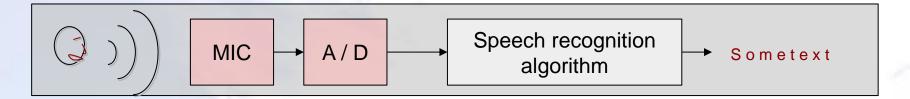


Course outline



Subject

- Speech recognition by computers
 - Input: a sequence of samples representing sounds (electroacoustic conversion, A/D conversion will not be covered)
 - Output: phonetic transcription / words



Topics

- Speech signal: production and characterization (phone classes)
- Speech signal representation (LPC, cepstrum, MFCC)
- Sequence matching (Dynamic Time Warping)
- Hidden Markov Models
- Language modeling



ASR preliminaries



Automatic Speech Recognition (ASR)

- Difficult task: patterns, variability, noise and distortions
- Huge demand: natural human-computer interface, data analysis for retreivial, identification, understanding
- Noticeable progress: operational continuous speech recognition (CSR) systems (for English), widespread isolated word recognition (IWR) applications
- Data-driven problem solution (training)

ASR contexts

- Vocabulary size: limited-large (modeling and recognition strategies)
- Universality: speaker-dependent vs.speaker independent systems
- Target domain: domain-specific or unconstrained
- Task definition: isolated word recognition or continuous speech
- Input quality: noise and reverberations vs. clean speech
- Modifying factors: emotions, ilness, age, gender
- Language dependence



ASR preliminaries

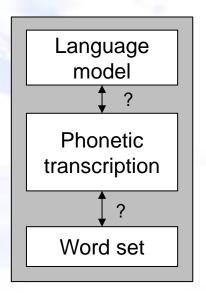


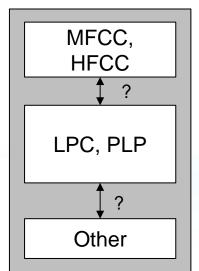
General framework

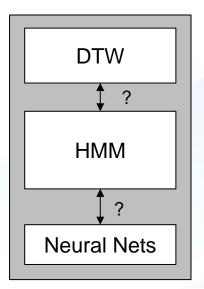
- Define the context and objectives, get training data
- Determine target categories and decide on class-modeling strategy
- Derive appropriate quantitative representation of speech signal
- Decide on recognition strategy
- Build models for considered categories and train their parameters
- Execute some adopted recognition scheme

IWR - CSR

Clean – noisy
Dep. – indep.
Small – large
voc.
...









ASR preliminaries



- Basic concepts
 - Sounds and their representation
 - Speech signal

- Acknowledgements: examples were created using the following software
 - Praat
 - Htk
 - Matlab



Sounds

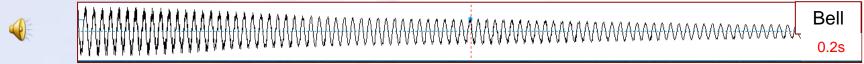


Sound perception and recognition

- Attributes: loudness (intensity), timbre (composition), variability, ...
- Recognition: requires quantitative description

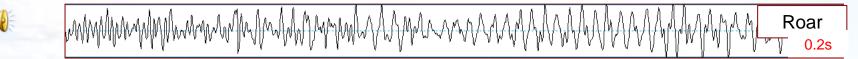
 Temporal waveforms

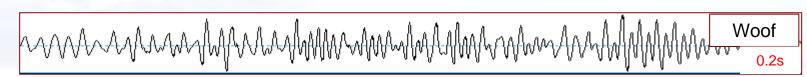










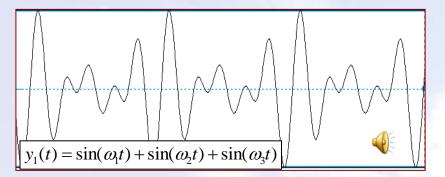


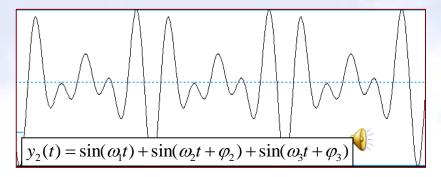


Sound perception and representation



- Representation as a function of time
 - Phase-insenitivity: waveforms cannot be the right representation

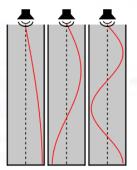


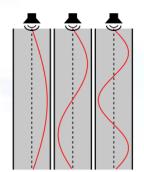


Different waveforms – the same perception

- Representation in terms of harmonic components
 - Physically-motivated representation of sounds: acoustic resonance

Some source is producing a collection of sine waves – only selected are emphasized and make a timbre of a sound





http://www.quora.com

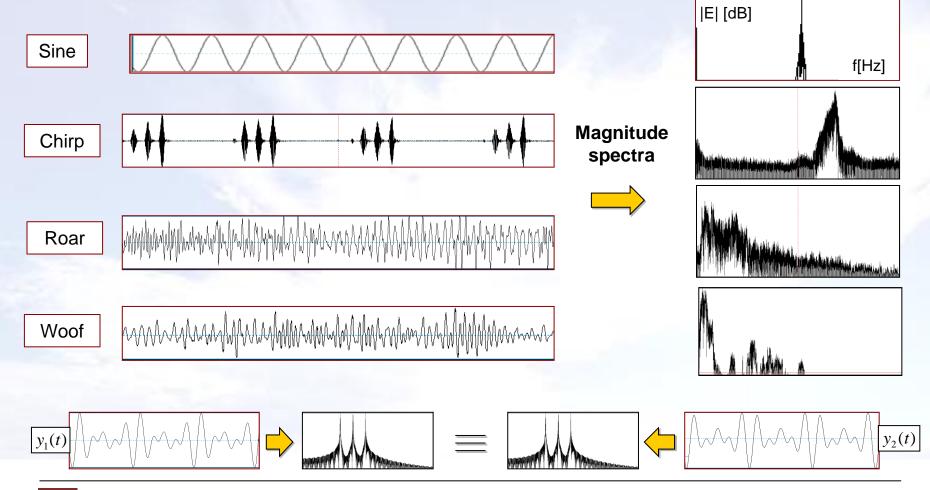


Sound perception and representation



Representation: a collection of harmonic components

 Fourier Transform and spectra (details – later): decomposition of a signal into harmonic components

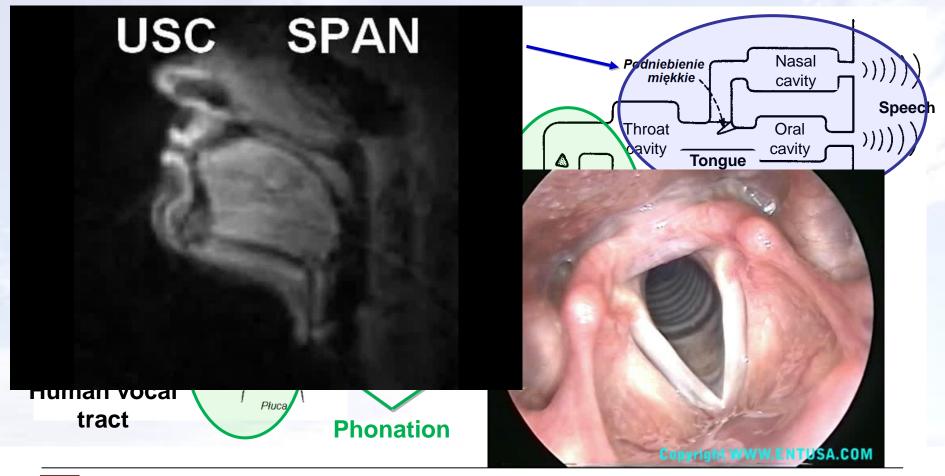




Speech production



- By humans (and many other species)
 - Generation of acosutic stimuli (phonation)
 - Forming the stimuli into differentiable sounds (articulation)





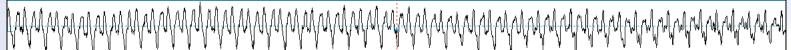
Speech production



Phonation

Vocal folds: vibrate (voiced) do not vibrate (voiceless)



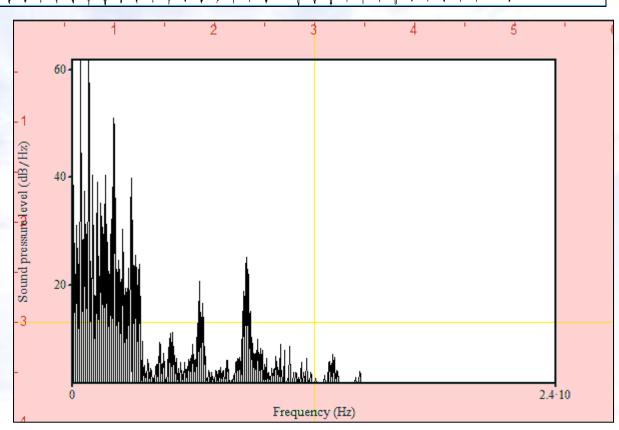


Voiced phonation

- Multi-harmonic
- Fundamental frequency (pitch)
- Voiced speech

Unvoiced phonation

- Vocal folds do not vibrate
- Turbulent flow
- Unvoiced speech





Speech production: source-filter model

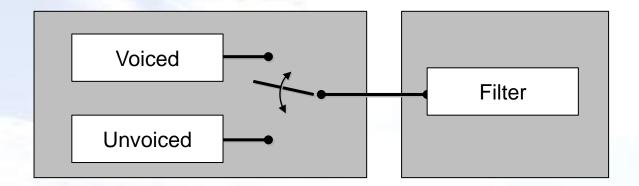


Source

- Voiced excitation: vocal folds vibrate
- Unvoiced excitation

Filter

Resonant cavities of dynamically changing structure





Speech signal: acoustics



Phones and phonemes

- Phoneme: the smallest structural unit that distinguishes meaning in a language
- Phone: a physical instance of phoneme

Phone types

- Vowels flow of air is not impeded: a e o i u
- Consonants impeded or stopped
 - Nasals formed by opening nasal cavity (m,n)
 - Fricatives formed by impeding air-flow: may be voiced (v,z, δ as in **th**is) or voiceless (f,s,sh, θ as in **th**ought, h)
 - Stops complete stopping then releasing air: voiced (b,d,g), voiceless (p,t,k)
 - Liquids r,l
- Semi-vowels: w as in wet, y as in yard



Phonetic transcription



Highlights

- Textual representation of sounds – of what we hear
- Several standards, the most common: IPA (International Phonetic Alphabet)
- There are more symbols than letters
- This is to be recognized in CSR

ðis iz ðə fərst lektsər

This is the first lecture

http://www.antimoon.com/how/pronunctransdemo.htm

| vowels | | | | |
|-----------------|---|--|--|--|
| IPA | examples | | | |
| ٨ | c <u>u</u> p, l <u>u</u> ck | | | |
| a: | <u>a</u> rm, f <u>a</u> ther | | | |
| æ | c <u>a</u> t, bl <u>a</u> ck | | | |
| e | m <u>e</u> t, b <u>e</u> d | | | |
| ə | <u>a</u> way, cin <u>e</u> m <u>a</u> | | | |
| 3: ^r | t <u>ur</u> n, l <u>ear</u> n | | | |
| I | h <u>i</u> t, s <u>i</u> tt <u>i</u> ng | | | |
| i: | s <u>ee,</u> h <u>ea</u> t | | | |
| b | h <u>o</u> t, r <u>o</u> ck | | | |
| ɔ : | c <u>a</u> ll, f <u>ou</u> r | | | |
| υ | p <u>u</u> t, c <u>oul</u> d | | | |
| u: | bl <u>ue</u> , f <u>oo</u> d | | | |
| aı | f <u>i</u> ve, <u>eye</u> | | | |
| aυ | n <u>ow</u> , <u>ou</u> t | | | |
| еі | s <u>ay, eigh</u> t | | | |
| oυ | g <u>o,</u> h <u>o</u> me | | | |
| IC | b <u>oy, joi</u> n | | | |

| Co | wii <u>ere, air</u> | | | |
|--------------|--|---|---|--|
| 19, | n <u>ear</u> , h <u>ere</u> | | | |
| υə | p <u>ure,</u> t <u>our</u> ist | | IPA | |
| | |] | | |
| consonants | | | | |
| IPA examples | | | | |
| b | <u>b</u> ad, la <u>b</u> | 5 | <u>s</u> un, mi <u>ss</u> | |
| d | <u>d</u> i <u>d</u> , la <u>d</u> y | ſ | <u>sh</u> e, cra <u>sh</u> | |
| f | <u>f</u> ind, i <u>f</u> | t | <u>t</u> ea, ge <u>tt</u> ing | |
| g | give, flag | t∫ | $\underline{\operatorname{ch}}$ eck, $\underline{\operatorname{ch}}$ ur $\underline{\operatorname{ch}}$ | |
| h | $\underline{\mathbf{h}}$ ow, $\underline{\mathbf{h}}$ ello | θ | <u>th</u> ink, bo <u>th</u> | |
| j | yes, yellow | ð | <u>th</u> is, mo <u>th</u> er | |
| k | <u>c</u> at, ba <u>ck</u> | v <u>v</u> oice, fi <u>ve</u> | | |
| I | <u>l</u> eg, <u>l</u> itt <u>l</u> e | w | <u>w</u> et, <u>w</u> indo <u>w</u> | |
| m | <u>m</u> an, le <u>m</u> on | Z | <u>z</u> oo, la <u>z</u> y | |
| n | <u>n</u> o, te <u>n</u> | 3 | plea <u>s</u> ure, vi <u>si</u> on | |
| ŋ | si <u>ng,</u> fi <u>n</u> ger | d3 | just, large | |
| р | pet, map | http://www.antimoon.com/ how/pronunc-trans.htm | | |
| r | red, try | | | |
| | | | | |

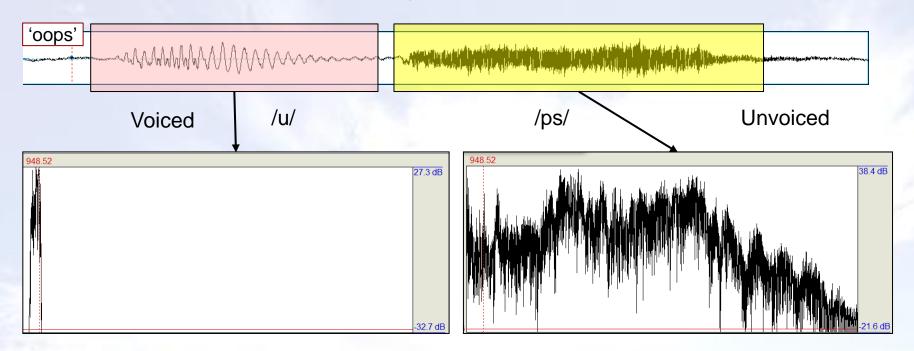
eər where, air

Voiced / unvoiced speech



Phone discrimination: phonation

Voiced speech has substantially different spectrum than unvoiced one



Clear differences

- Voiced: Few dominant frequencies
- Unvoiced: wide-band spectrum

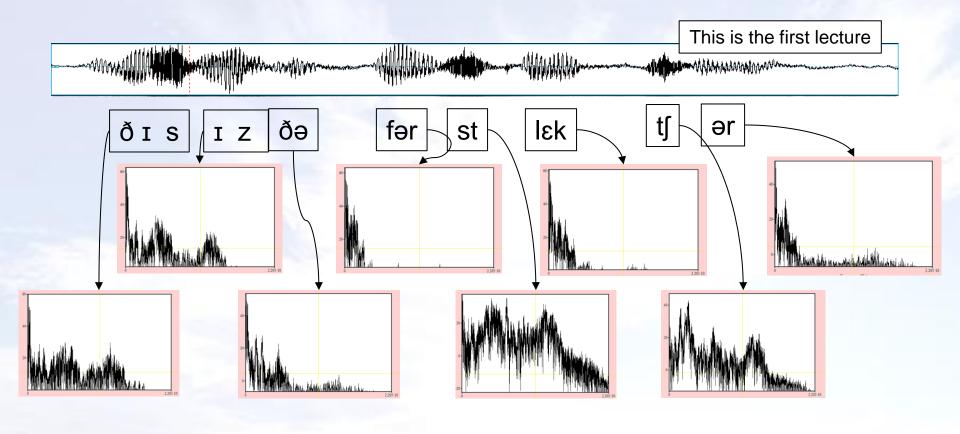


Speech signal properties



Phone discrimination: articulation

 Vowels, nasals, fricatives, stops, liquids etc: articulation differences are reflected in signal properties







Corollary

- Spectral representation seems adequate
- Information contents: very rich
- Is spectrum an appropriate basis for recognition?

Spectrum

- Decomposition of a function w.r.t. some set of bases
- Many possible basis functions
- Periodic basis functions: also several candidates
- Discrete functions discrete bases

Discrete Fourier Transform - DFT

- Harmonic components
- Fast computational algorithms



Discrete Fourier Transform: a review



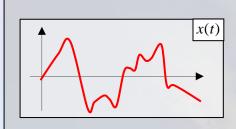
Highlights

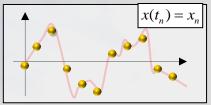
Decomposition of a (periodic) signal into harmonic components

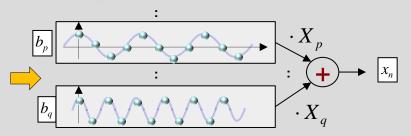
 $x(t_n)=x_n$ expressed as a composition of bases

$$x_n = c\sum_{k=0}^{N-1} X_k \overline{b}_{k,n}$$
 X_k : k-th weight $b_{k,n}$ k-th basis function

C: a constant







- DFT bases: complex exponentials (sampled at discrete points)

For a given signal x: determine the weights of the mixture X_k

 I_n : a discrete time instant

DFT
$$X_k = \sum_{n=0}^{N-1} x(t_n) e^{-j\omega_k t_n} = \sum_{n=0}^{N-1} x(t_n) (\cos \omega_k t_n - j \sin \omega_k t_n)$$

$$\omega_k = 2\pi f_k = 2\pi f_s \frac{k}{N} = \frac{2\pi k}{N}$$

 $f_{\rm s}$ Sampling frequency N Number of samples

Discrete Fourier Transform: a review



DFT
$$X_k = \sum_{n=0}^{N-1} x_n e^{-j\frac{2\pi}{N}kn}$$

Signal decomposition

Inverse DFT
$$x_n = \frac{1}{N} \sum_{k=0}^{N-1} X_k e^{j\frac{2\pi}{N}kn}$$

Signal reconstruction

Comments

- Weights regularly spaced in frequency domain
- Sampling frequency matters: Nyquist condition
- Number of samples in t and f is the same

$$f_k = \frac{k}{N} f_s$$

The more points, the higher resolution

A meaning of the adopted decomposition

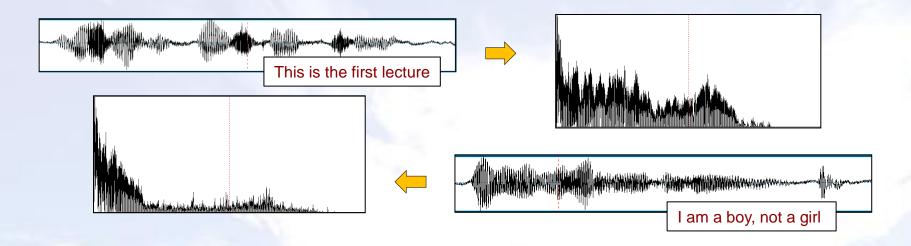
- Complex exponential evolving in time: a circle in Im-Re space
- Projections on a complex exponential: match to cos, match to sin rotated by -90 deg
- This matches arbitrary phase shift

$$x(t)\sin(\omega t + \varphi) = x(t)A_1\cos(\omega t) + x(t)A_2\sin(\omega t)$$

Signal pectrum



- Spectrum: a set of decomposition coefficients
 - Is magnitude spectrum an appropriate basis for speech recognition?



The answer

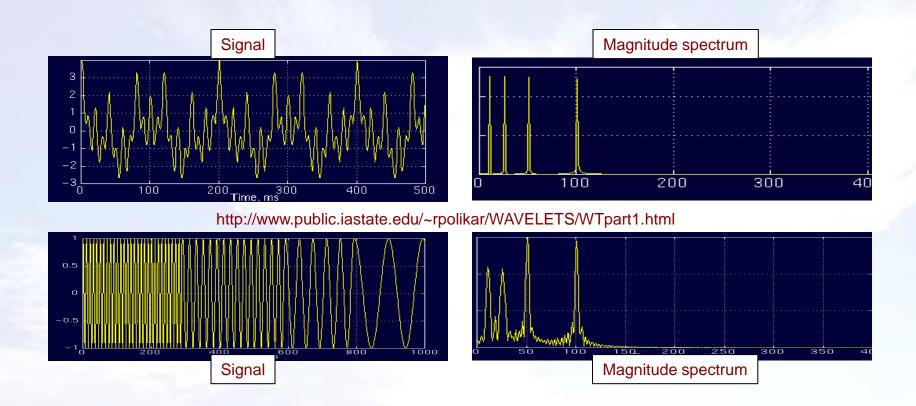
- Of course, not
- I am a boy, not a girl = I am a girl, not a boy
- Spectrum lacks a temporal resolution



Spectrum and signal recognition



- Spectrum lacks temporal clues
 - Unrelated time signals can produce same / similar spectra





Short-time Fourier Transform: STFT

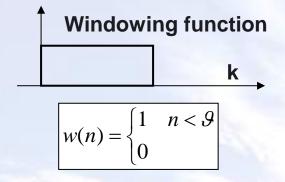


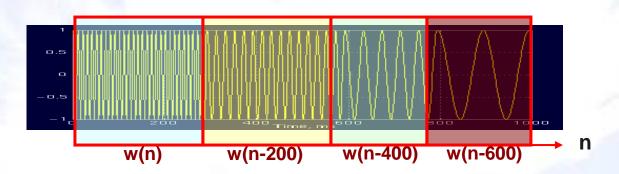
An objective

To preserve information on temporal evolution (provide time resolution)

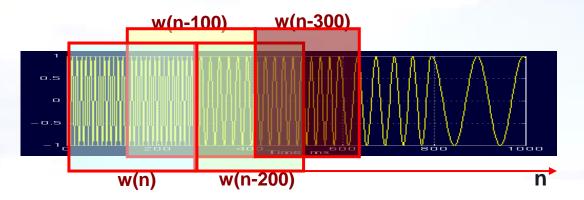
Temporal windows

Extract parts of a signal – define domains of subsequent DFT analyses





Overlapping windows

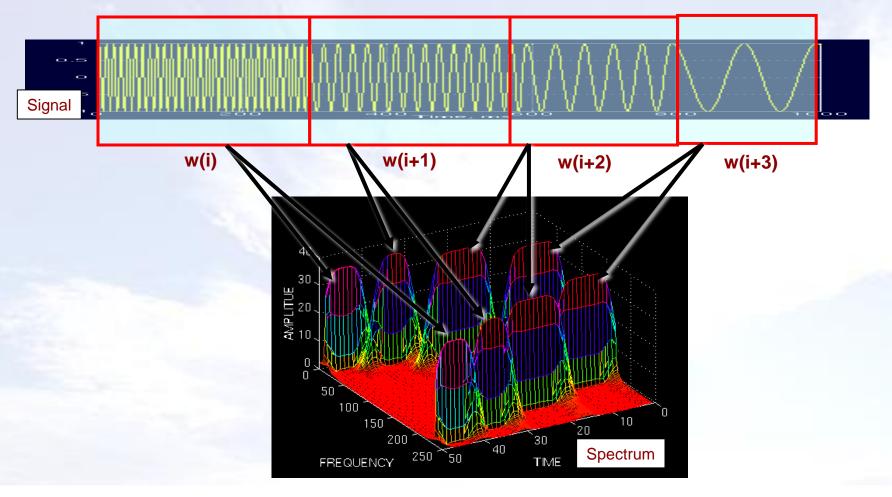




Short-time Fourier Transform: STFT



- Signal representation
 - Spectra computed for consecutive time 'frames'

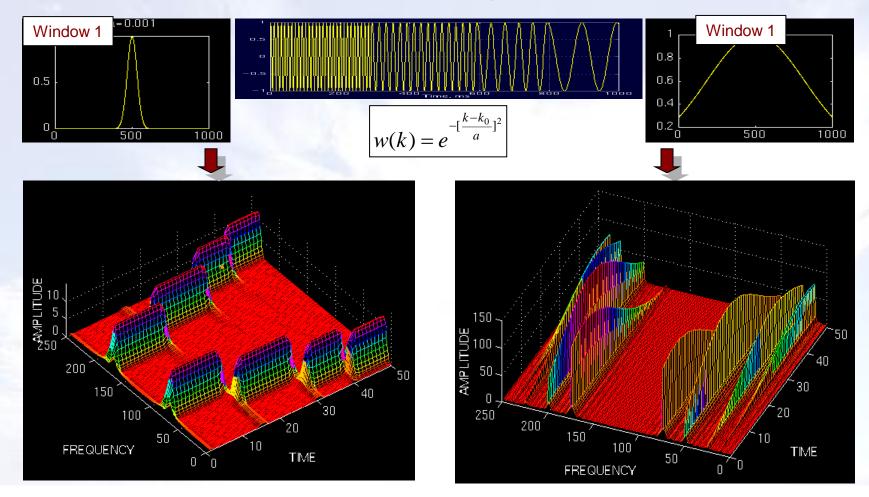




Short-time Fourier Transform: STFT



- Trade-off between temporal and frequency resolution (note: overlap)
 - Good time localization limited frequency

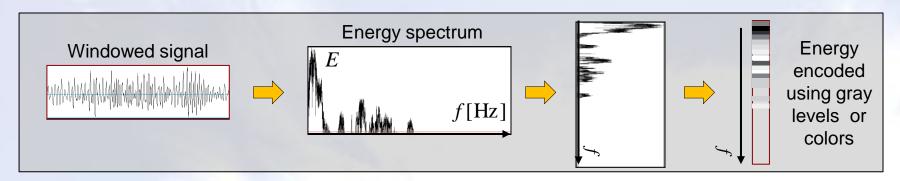


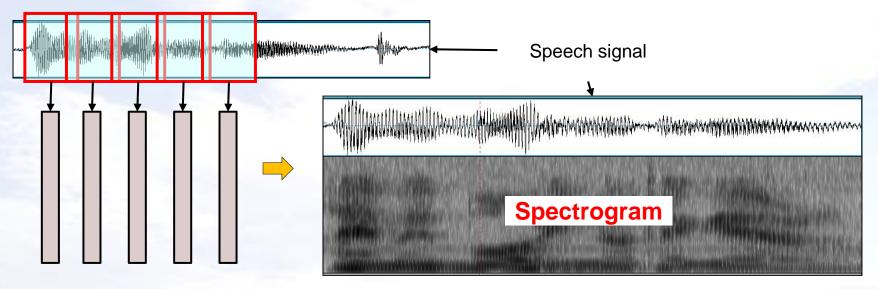


Spectrograms



- Time-frequency representation: spectrogram
 - Spectra computed for windows extracted at regular time intervals





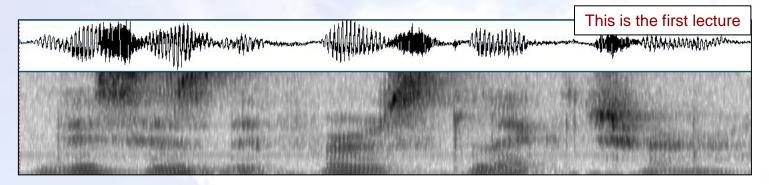


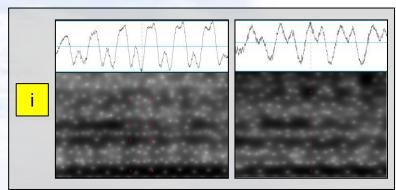
Spectrograms

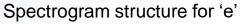


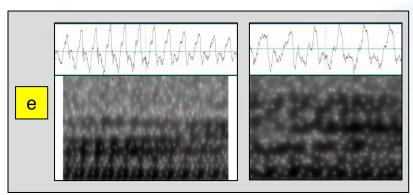
Spectrogram contents

- Information on temporal dynamics of speech (sound) production
- Resonant cavities characteristic for phones are reflected by spectral contents of spectrogram









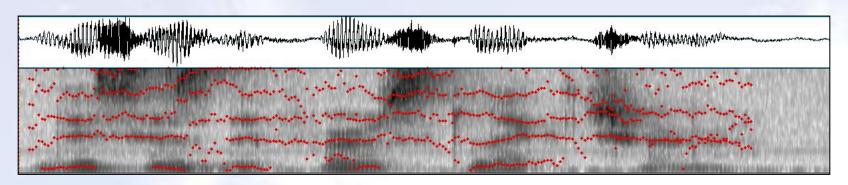
Spectrogram structure for 'i'





Spectrogram features

Bands of high energies: formants



Formants

- Produced in resonant cavities: local energy maxima in various frequency bands
- Vary in time as different phones get articulated
- Form patterns characteristic to phones
- Exist for voiced speech only , typically, up to four (can be hard to extract)
- Can be used e.g. for vowelclassification, are insufficient for speech representation – richer representation is required



Spectrogram computation considerations

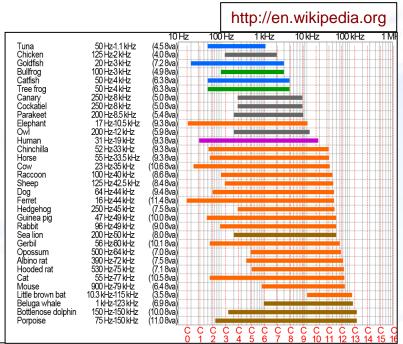


Spectrogram

- Provides information on temporal evolution and frequency composition of signal in windows
- Should carry all relevant information
 - Q1 what frequencies should be covered?
 - Q2 what should be duration of a window?

Frequency range for speech representation

- Tip: knowledge on auditory perception -20 Hz to 20 kHz (sampling frequency for digital music is 44 kHz)
- Telephony: channel bandwitdh: 300 Hz –
 3.4 kHz (sampled at 8 kHz)





Window length



Window length for speech representation

The window cannot be neither too long nor too short

Upper bound for window length

 Stationarity of the underlying process: a filter (articulation cavities) should not change its parameters (otherwise, frequency information becomes useless for phone identification)

Lower bound for window length

 Too short window – unable to represent low frequencies (to fit full period of 100 Hz wave, 10 ms window needs to be used)

Typical window size

- Within a range 10-30 ms
- Exact duration often adjusted to satisfy DFT requirements (to have a number of samples equal to some integer power of 2):
 - 8kHz sampling (T_s =0.125 ms) 256 samples give T=32 ms, which means that max freq < 4kHz, min freq approx. 30 Hz





A role of the window

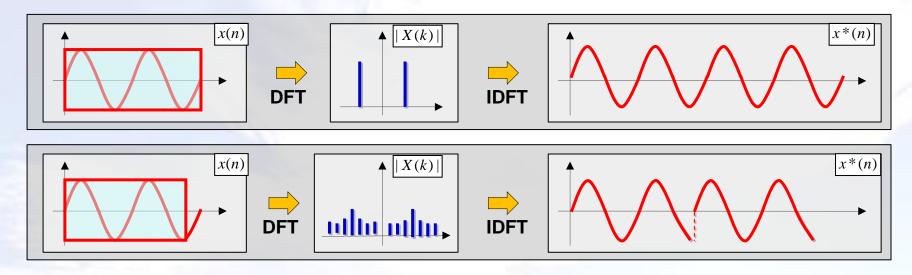
$$y(n) = w(n)x(n)$$

To preserve information on temporal order of events

$$w(n) = \begin{cases} 1 \\ 0 \end{cases}$$

Rectangular window

- The rectangular window ensures extraction of consecutive parts of a signal
- Is there any problem with a rectangular window?



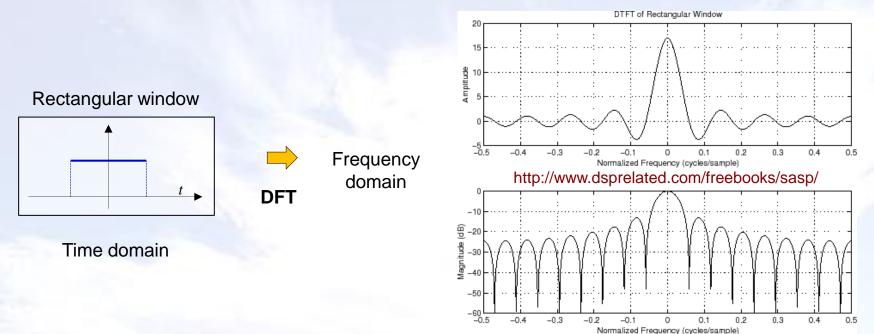
The problem

- DFT attempts to represent a discontinuous function (periodic)
- For rectangular window inevitable! distortions Spectrum 'leaks' over neigboring frequencies





- Problems caused by spectral leakage
 - Introduction of 'false' signal components
 - Potential hiding of lower, actual components



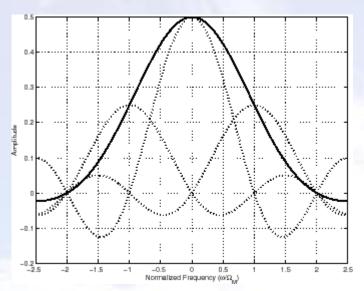
- A window should reduce the disconinuity effect
 - Time domain: reduce magnitudes at both signal ends;
 - Frequency domain; reduce side lobes





A gallery of potential candidates for windows

Hamming, Hanning (cosine combined with a rectangular window)



http://www.dsprelated.com/freebooks/sasp/

- Three sinc functions in frequency domain, scaled in proportions a, and two b and appropriately shifted
- Results in eliminating a side lobe (at the expense of widenning the central one)

$$W(n) = W_R(n) \left[a + 2b \cos \left(\frac{2\pi}{N} n \right) \right]$$



$$W_{HN}(n) = W_R(n)\cos^2\left(\frac{\pi}{N}n\right)$$

- Hamming window
 - a=0.54, b=0.23 (chosen to completely cancel the largest side lobe)

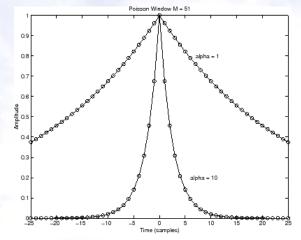




Other candidates

- Blackman-Harris window family: similar concept as for Hanning/Hamming, but more sinc functions are used
- Triangular Bartlett window (convolution of two rectangular windows)

Poisson window



http://www.dsprelated.com/freebooks/sasp/

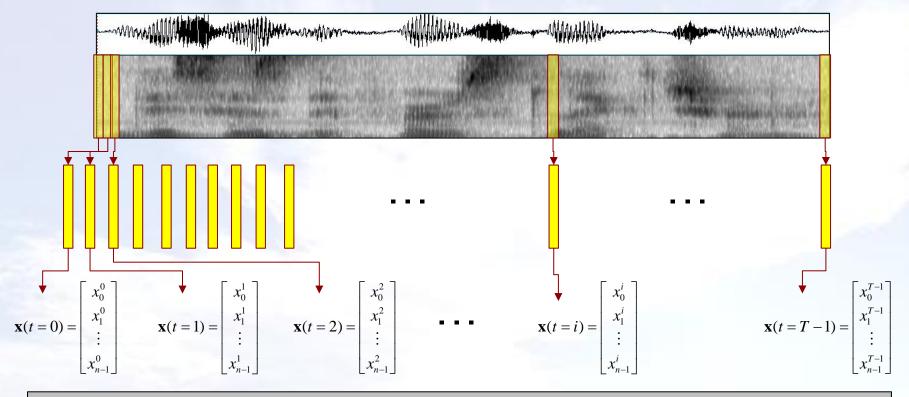
$$W_{HN}(n) = W_R(n)\cos^2\left(\frac{\pi}{N}n\right)$$

Gaussian, Kaiser ...



Spectrogram

- A sequence of slices. A slice: a vector of DFT magnitudes for a window
- A sequence of vectors



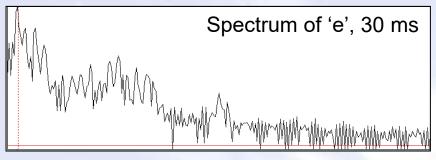
Spectrogram: a sequence of vectors $\{\mathbf x_0, \mathbf x_1 ... \mathbf x_{T-1}\}$ \Longrightarrow Speech signal representation

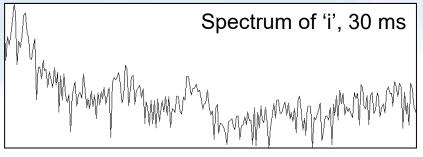


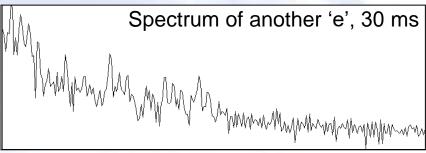


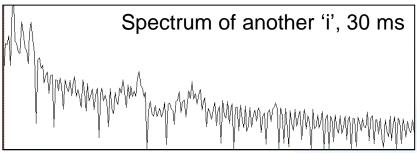
Feature vectors

– All DFT components?









Observation 1

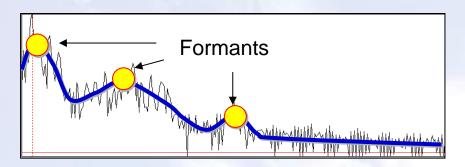
- Huge within-class variability, unclear between-class differences: probably useless representation
- Better representation needs to be found

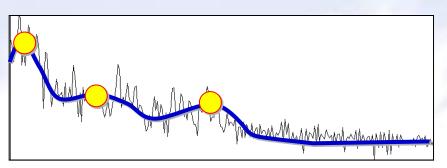




Observation 2

Spectrum looks like a mixture of slow-varying and fast-varying components





- 'Slow' (frequency axis) varying components: combination of filters corresponding to resonant cavities (articulation)
- Formant frequencies filter maxima
- Informative representation of speech
 - Slow-varying frequency-domain components
 - Analogy to performing low-pass filtering of temporal signals
- Filtering of temporal signals
 - Based on assumption of linearity (additive 'noise)





Temporal signal filtering

- Get Fourier spectrum
- Retain selected frequencies (low for elimination of noise)
- Underlying model: linear composition of a signal

$$y(t) = x(t) + n(t)$$
 \Rightarrow $Y(f) = X(f) + N(f)$ \Rightarrow $\widetilde{Y}(f) = X(f)$ \Rightarrow $\widetilde{y}(t) = x(t)$

Speech signal

- Convolution of source (vocal fold excitation) and filter (articulation cavities)
- Spectrum is a product of the two
- We want to get rid of excitation (for us this is a noise)
- We are in frequency domain: how to proceed?

$$s(t) = a(t) * p(t)$$

s(t) speech signal p(t) phonation

a(t) articulation (impulse response)



$$S(f) = A(f)P(f)$$



- The problem
 - Linear separation of components given in a product form

$$S(f) = A(f)P(f)$$

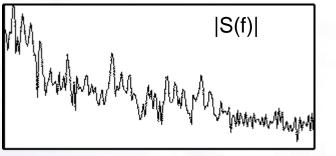
- Solution
 - Homomorphic filtering (simple trick: use logarithms)

$$S(f) = A(f)P(f)$$

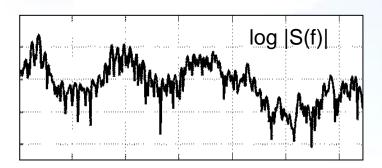


$$\log S(f) = \log A(f) + \log P(f)$$

Spectrum becomes a sum of components: can be approached using the presented framework

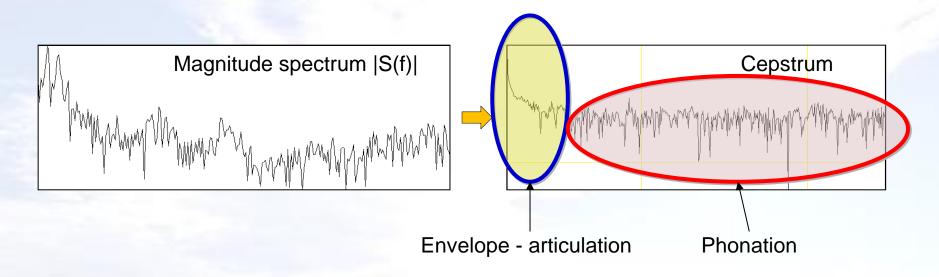








- 'Spectral decomposition' of log(abs(spektrum))
 - An objective: to separate slow-varying and fast-varying components
 - Apply inverse Fourier transform on log-abs spectrum
 - We go back to time axis, but this is not a reconstructed signal
 - To emphasize a difference it is called CEPSTRUM (reordering of SPECTRUM)
 - Cepstral analysis terminology: quefrency, liftering, ...

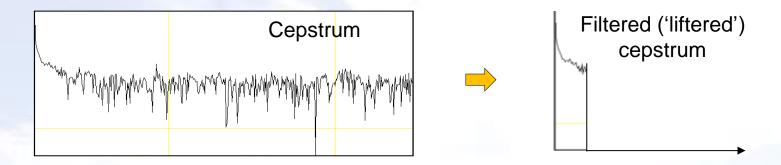






Low-order cepstral coefficients

- Low-order cepstral coefficients correspond to spectral envelope represent acoustic filters formed by resonant cavities of vocal tract (speech articulation)
- Possible methodology: cepstrum truncation (10-20 leading coefficients are left)



- The first representation of speech signal
 - A sequence of vectors (for each window) containing low-order cepstral coefficients



Cepstral mean subtraction



Channel impact on speech signal properties

- A channel: all components that exist between speech production point and a point where electrical speech waveform is produced
- Path variability: different laptops where ASR software is executed have different microphones so prototypes are hardware-specific, room-specific reverberations ...
- Channel effects: convolution in time (product in frequency)
- Real channels— non-ideal frequency response: spectral distortions

Cepstral mean

Cepstral vector: combination of speech and channel components

$$\mathbf{C}_i = \mathbf{C}_i^s + \mathbf{C}^{ch}$$

- Channel component does not change
- Mean of cepstral vectors
- Mean cepstrum of speech is zero (as mean of original signal is zero)

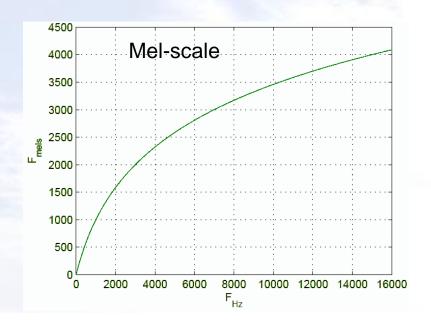
$$\frac{1}{n}\sum_{i=0}^{n-1}(\mathbf{C}_i^s + \mathbf{C}^{ch}) = \frac{1}{n}\sum_{i=0}^{n-1}\mathbf{C}_i^s + \mathbf{C}^{ch}$$

$$\widetilde{\mathbf{C}}_{k} = \left(\mathbf{C}_{k}^{s} + \mathbf{C}^{ch}\right) - \mathbf{\mu} = \mathbf{C}_{k}^{s}$$

 $\mu = \mathbf{C}^{ch}$



- Possible speech representation
 - A sequence of cepstrum-mean subtracted low-order cepstral coefficients
 - Why it is not commonly used?
- Psychology of hearing
 - Perception of frequencies is nonlinear: linear up to 1 kHz, then logarithmic
 - Masking phenomenon and critical bands



For F>1 kHz

$$F^{Mel} = 2595 \log_{10} \left(1 + \frac{F^{Hz}}{700} \right)$$



Masking

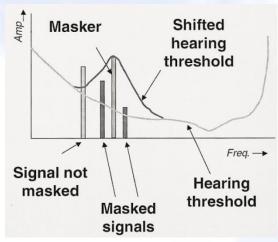
Hiding neighboring tones by a stronger component

Can be explained only if we assume that we perceive sounds within bands

(critical bands)

Ear acts as a set of filters

Cannot ignore it in modeling speech



Integration of frequency stimulus using filters

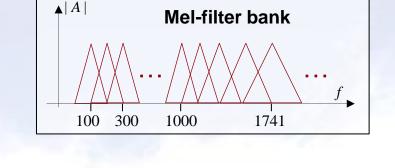
- Filters distributed evenly in subjective frequency domain (Mel-scale)
- Filters have same width in Mel-scale
- i.e. they are nonlinearly distributed along objective frequency axis





Mel-filter bank

- 24 triangular overlaping filters
- Centers linear in Mel-scale
- Bandwidth: masking range



| lack A A f | |
|------------------|--|
| f_0 B | |

| • | Mel-spectrum |
|---|--------------|

- Speech spectrum is integrated within critical bands
- Can be seen as spectrum downsampling

| f_0 | B | f_0 | B |
|-------|-----|-------|-----|
| 100 | 100 | 1516 | 211 |
| 200 | 100 | 1741 | 242 |
| 300 | 100 | 2000 | 278 |
| 400 | 100 | 2297 | 320 |
| 500 | 100 | 2639 | 367 |
| 600 | 100 | 3031 | 422 |
| 700 | 100 | 3482 | 484 |
| 800 | 100 | 4000 | 556 |
| 900 | 100 | 4595 | 639 |
| 1000 | 124 | 5278 | 734 |
| 1149 | 160 | 6063 | 843 |
| 1320 | 184 | 6964 | 969 |





MFCC

- Spectrum is accumulated in bins defined by Mel-filters
- 24 filters 24 intervals along frequency axis
- Mel-Cepstrum computed from 24-element Mel-spectrum
- Result: Mel-Frequency Cepstral Coefficients (MFCC)
- Low-order MFCC's use to represent articulation (typically up to 12)



Modeling articulation dynamics



Dynamics

- A rate of change: delta-MFCC: subtraction of MFCC's from consecutive frames
- Second derivative of MFCC changes: delta-delta-MFCC

Common speech representation

 A sequence of vectors (for each window) containing low-order MFCC, delta-MFCC and delta-delta coefficients,

