#### Introduction to exceptional points

- Proposed by Wiersig in 2014
- Degeneracy in non-hermitian Hamiltonian is called an exceptional point (EP)
- Typical systems like 2 coupled non-linear oscillators show EP's
- Hamiltonian (assume  $\hbar = 1$ )

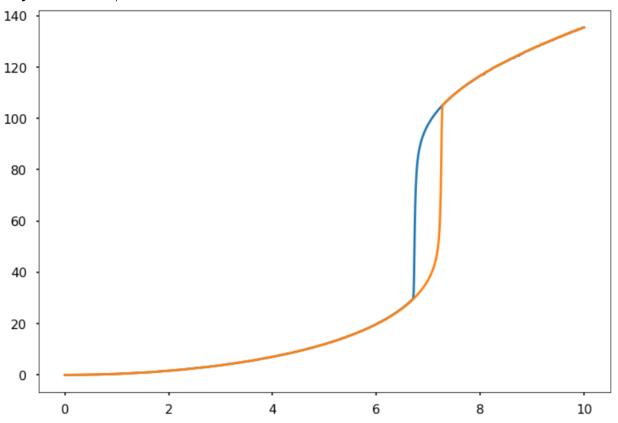
$$egin{pmatrix} \omega_1 & g \ g & \omega_2 \end{pmatrix}$$

- Eigenvalues:  $\omega_\pm= ilde{\omega}_{av}\pm ilde{\Delta}/2\sqrt{1+(rac{2g}{ ilde{\Delta}})^2}$  where  $w_{av}=( ilde{\omega}_1+ ilde{\omega}_2)/2$  and  $ilde{\Delta}= ilde{\omega}_1- ilde{\omega}_2$
- Here, exceptional point occurs at  $rac{2g}{\Delta}=i$  (square root singularity)
- Around exceptional point perturbation of the form  $\text{Re}(\tilde{\omega}_i) + \epsilon$  results in a splitting  $\text{Re}(\omega_+ \omega_-) \propto \sqrt{\epsilon}$  for small  $\epsilon$  S
- Tradeoff: Highly sensitive to noise due to enhanced sensitivity

### Alternative: Non-linear optical sensor

- Single mode coherently driven non-linear Kerr oscillator
- Measured signal: Splitting in transmitted intensities at 2 endpoints of hystersis cycle

• N (y axis) vs  $F/\sqrt{\Gamma}$  (x axis)



# Non-Linear Optical sensor cont...

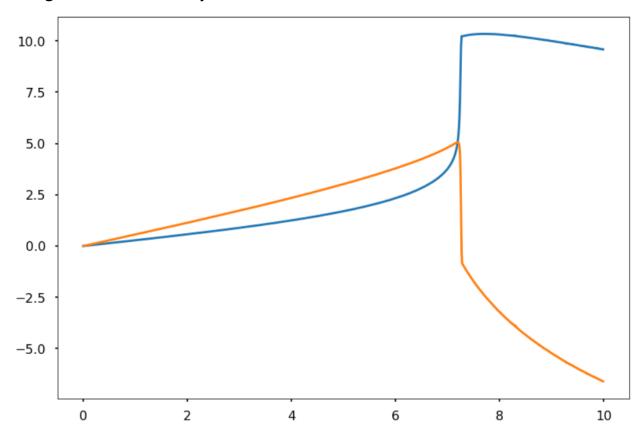
- Realised using a Fabry-Perot setup or any other architecture (whispering gallery, ring, photonic crystal etc.) having one mode spectrally distant from other modes and probes an intensity dependent refractive index
- In a frame rotating at driving frequency  $\omega$  intracavity field  $\alpha$  obeys following equation  $i\dot{\alpha}(t)=-\Delta-\frac{i\Gamma}{2}+U(|\alpha(t)|^2-1)\alpha(t)+i\sqrt{\kappa_L}F+D\xi(t)$ 
  - $\alpha$ : Intracavity Field
  - $\Delta$ : Detuning ( $\omega \omega_0$ ) where  $\omega_0$  is the resonance frequency
  - $\Gamma$ : Total Loss= $\gamma + \kappa_L + \kappa_R$  where  $\gamma$  is intrinsic loss,  $\kappa_L, \kappa_R$  is loss through left and right ports respectively
  - U: Kerr non-linearity strength
  - F: Amplitude of input laser
  - D $\xi(t)$ : Gaussian white noise with variance  $D^2$
  - Assume D=0 for now

# Solving for $\alpha$

Setting  $\alpha$  as a+ib where a and b are real, we transform the equation for  $\dot{\alpha}$  to the following coupled non-linear differential equations.

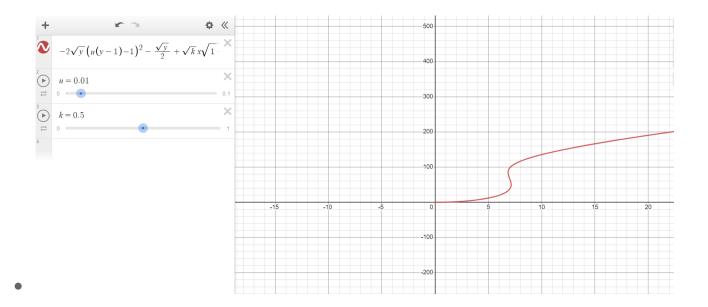
$$egin{aligned} \dot{a}&=(-\Delta+U(a^2+b^2-1))b-rac{a\Gamma}{2}+\sqrt{\kappa_L}F\ \dot{b}&=rac{-b\Gamma}{2}-a(-\Delta+U(a^2+b^2-1)) \end{aligned}$$

Hence we can see U represents the strength of non-linearity in this equation (blue-variation of a (y axis) with time (x axis)) (orange-variation of b (y axis) with time (x axis))



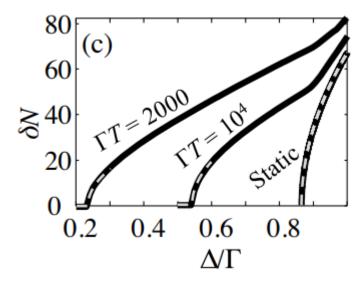
### Steady state solutions

- Set  $\dot{\alpha} = 0$  (steady state approximation)
- Also set  $\frac{d|F|^2}{dN} = 0$  (turning points)
- Solutions are  $ilde{N}_{\pm}=rac{2\Delta}{3U}\pmrac{2\Delta}{6U}\sqrt{1-(rac{\sqrt{3}\Gamma}{2\Delta})^2}$  (notice similarity with EP sensors)
- https://www.desmos.com/calculator/zhxszbbxrs



### **Properties**

- Notice that the transmitted signal ( $\delta N=N_+-N_-$ ) depends on  $\Delta$  and measurement time.
- The dependence (as plotted by the authors) has been shown below

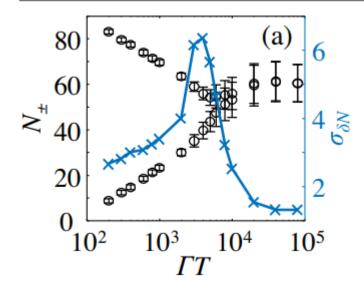


- Observe: For faster measurement time, less detuning needed for signal ⇒ better sensitivity for faster measurement times.
- Dashed Lines: Square root fit near each singularity

#### **Effects of Noise**

TIII now analysis was based on assumption D=0

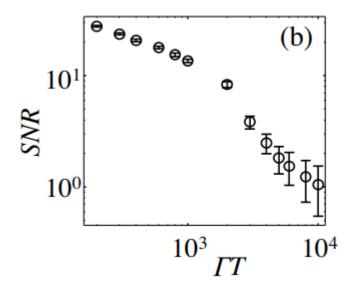
• With noise, solve numerically with different random seeds



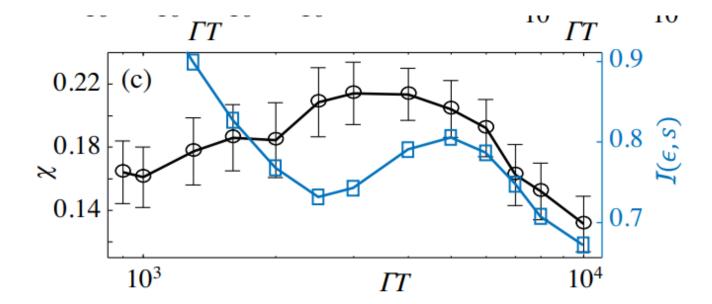
(plot from reference)

Parameters:  $\Delta=0.7\Gamma$ 

# Advantages cont...



(plots from reference)



SNR: Signal to noise ratio defined as  $\frac{\delta N}{\sigma_{\delta N}}$   $\chi$ : Precision figure of merit:  $\frac{\delta \bar{N}_{\epsilon} - \delta \bar{N}_{0}}{\sigma_{0} + \sigma_{\epsilon}}$ ; Quantifies the mean change in signal relative to measurement uncertainty

# Alternative viewpoint: Information theory

- Mutual information between 2 random variables: Measure of certainty of a random variable given knowledge of the other variable
- $I(\epsilon;S) = \sum_{s \in S} \sum_{\epsilon \in E} p_{\epsilon,S}(\epsilon,s) log \frac{p_{(\epsilon,S)}(\epsilon,s)}{p_{\epsilon}(\epsilon)p_{S}(s)}$  (from KL divergence between  $p_{(\epsilon,S)}(\epsilon,s)$  and  $p_{\epsilon}p_{S}(s)$ )
- KL divergence: Measure of similarity between to probability distributions P and Q
- Set  $p_{\epsilon}(\epsilon)$  gaussian with mean  $\Gamma/100$  and  $\sigma=\Gamma/1000$  and calculate  $p_s(s)$  numerically using detuning  $\Delta_0=0.7\Gamma$ . Then based on value of s we get the joint probability distribution
- Observe the enhancement of information content near  $T_{ss}$
- Correlation between square root singularity, mutual information and precision.
- Results are general for  $\epsilon << \Gamma$

#### **Effects of averaging**

- Smoothens effects of noise thus improving precision
- For same protocol F, it enhances sensitivity
- Tradeoff: Measurement time vs precision: averaging detrimental to fast sensing
- Non-trivial dependence of  $\chi$  on number of cycles n.
- Peak followed by fall (unexpected) followed by gradual rise (expected)
  Possibility of model parameters having best of both worlds (speed and precision)