

$$\sum_{i=0}^{M-1} |H_i(e^{j\omega})|^2 =$$

↓

$$\sum_{i=0}^{M-1} |H_i(e^{j\omega})|^2 =$$

$$\sum_{i=0}^{\infty} |H_i(e^{j\omega})|^2 = 1$$

$$H_o(z) H_o(\bar{z}') + H_i(z) H_i(\bar{z}')$$

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$$H_1(z) = \frac{1}{2} \left[A_0(z) H_0(z') + A_1(z) H_1(z') \right]$$

$$\begin{aligned}
 H_1(z) &= H_1(z^{-1}) + A_0(z) A_0(z^{-1}) + A_1(z) A_1(z^{-1}) \\
 &\quad - A_0(z) A_1(z^{-1}) - A_1(z) A_0(z^{-1}) \\
 H_0(z) &= H_0(z^{-1}) + H_1(z) H_1(z^{-1}) \\
 &= \frac{1}{q} \left[2 A_0(z) A_0(z^{-1}) + 2 A_1(z) \right] = 1
 \end{aligned}$$

$$= \frac{1}{4} [2 + 2] = 1.$$

$H_0(z)$ & $H_1(z)$ are also power complementary.

Ans: If $H_0(z)$ & $H_1(z)$ are APC, then they are also power complementary. (Not necessarily true for more than 2 TFs).

$$|H_0(e^{j\omega})|^2 + |H_1(e^{j\omega})|^2 = \gamma_2$$

$\xrightarrow{\omega = \omega_c}$

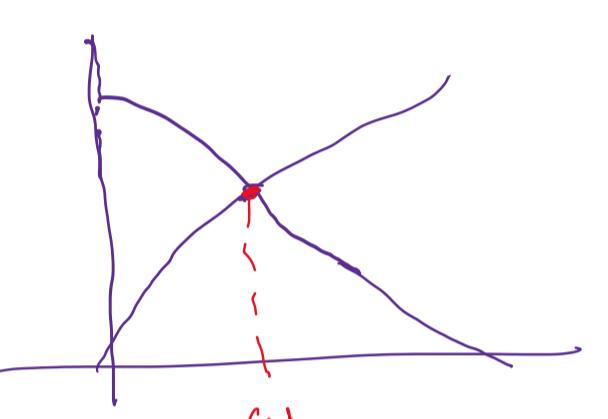
$$|H_0(e^{j\omega_c})|^2 = \gamma_2$$

$\xrightarrow{\therefore}$

$$|H_1(e^{j\omega_c})|^2 = \gamma_2$$

$\xrightarrow{\therefore}$

Since $|H(e^{j\omega_c})| = |A_1(e^{j\omega_c})|$,
 will be a cross-over at $\omega = \omega_c$.
 Therefore, ω_c is known as Cross-over frequency.



A P.C & power complementary filters are
 1. Known as cross-over filters.

$\{ H_i(z) \}$, $0 \leq i \leq M-1$, satisfying
 both allpass & power complementary.

$$\begin{aligned}
 &= \frac{1}{2} \left[A_0(z) + \frac{\pi_1(z)}{z} \right] \\
 H_1(z) &= \frac{1}{2} \left[1 - \frac{-\alpha + z^{-1}}{1 - \alpha z^{-1}} \right] \\
 &= \frac{1}{2} \left[\frac{1 - \alpha z^{-1} + \alpha - z^{-1}}{1 - \alpha z^{-1}} \right] \\
 &= \frac{1 + \alpha}{2} \frac{1 - z^{-1}}{1 - \alpha z^{-1}}
 \end{aligned}$$

1st order IIR
HPF

$w_c = \cos^{-1} \frac{2\alpha}{1 + \alpha^2}$

cross-over freq.

Magnitude Complementary Filter

$$\{G_i(z)\}_{0 \leq i \leq M-1}$$

$$\sum_{\bar{z}=0}^{M-1} |h_{\bar{z}}(e^{j\omega})| = 1.$$

$$H_0(z) = \frac{1}{2} \left[A_0(z) + A_1(z) \right]$$

$$H_1(z) = \frac{1}{2} \left[A_0(z) - A_1(z) \right].$$

$H_1(z) = H_0(z)$.

magnitude complementary pair.

$H_0(e^{j\omega})^2 = |H_0(e^{j\omega})|^2$ ✓

$H_1(e^{j\omega})^2 = |H_1(e^{j\omega})|^2$ ✓

$$\Rightarrow \left| h_0(e^{j\omega}) \right| + \left| h_1(e^{j\omega}) \right|$$

$$\begin{aligned}
 &= |H_0^2(e^{j\omega})| + |H_1^2(e^{j\omega})| \\
 &= |H_0(e^{j\omega})|^2 + |H_1(e^{j\omega})|^2 \\
 &= 1.
 \end{aligned}$$

Hence $h_0(z)$ & $h_1(z)$ are magnitude complementary filter pair.

A block diagram representing a discrete-time system $H_1(z)$. The input signal, shown as a blue line, enters a rectangular block labeled $H_1(z)$. The output signal, also a blue line, exits from the bottom of the block. A red curved arrow originates from the right side of the block and points downwards to the label $G_1(z)$, indicating that the output of $H_1(z)$ is $G_1(z)$.