

第11章 电路的频率响应

本章重点

*网络函数的概念,幅频特性和相频特性的意义;

*RLC串联的谐振分析;

RLC串联电路的频率响应及Q值、带宽、通频带;

*RLC并联电路的谐振分析;

RC低通滤波器、RC高通滤波器。

11.1 网络函数



单一正弦激励下,电路稳态响应的都为同频率的正弦量。

1. 定义

电路的工作状态随频率而变化的现象, 率特性 (频率响应)

网络函数 响应与激励之间的函数关系称为

$$H(j\omega)$$

$$H(j\omega) = \frac{R_k(j\omega)}{\dot{E}_{Sj}(j\omega)}$$

$$R_k(j\omega)$$

端口k的正弦稳态响应相量 $\dot{I}_k(j\omega)$ 或 $\dot{U}_k(j\omega)$

$$\dot{E}_{Sj}(j\omega)$$

端口j处的输入变量(正弦激励) $\dot{I}_{\mathrm{S}j}(\mathrm{j}\omega)$ 或 $\dot{U}_{\mathrm{S}j}(\mathrm{j}\omega)$

网络函数与激励无关,是系统参数和结构决定的。

网络函数是一个复数。
$$H(j\omega) = |H(j\omega)| \angle \varphi(j\omega)$$

2. 网络函数是一个复数



$$H(j\omega) = \frac{\dot{R}_k(j\omega)}{\dot{E}_{Sj}(j\omega)} = |H(j\omega)| \angle \varphi(j\omega)$$

幅频特性:正弦量有效值或振幅值之比

相频特性:正弦量初相位之差

己知网络函数,已知激励,可以求响应。

3. 网络函数的物理意义

网络函数有多种类型

当激励和响应属于同一对端子时,称为驱动点函数。

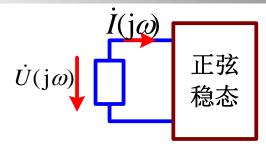
当激励和响应不属于同一对端子时,称为转移函数。

驱动点函数



驱动点阻抗
$$Z = H(j\omega) = \frac{\dot{U}(j\omega)}{\dot{I}(j\omega)}$$

驱动点导纳
$$Y = H(j\omega) = \frac{\dot{I}(j\omega)}{\dot{U}(j\omega)}$$



转移函数

转移阻抗
$$H(j\omega) = \frac{U_k(j\omega)}{\dot{I}_{Sj}(j\omega)}$$

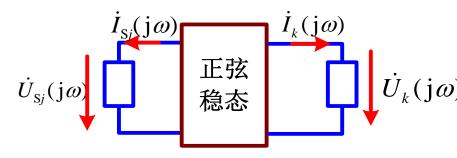
转移阻抗
$$H(j\omega) = \frac{\dot{U}_k(j\omega)}{\dot{I}_{Sj}(j\omega)}$$

转移导纳 $H(j\omega) = \frac{\dot{I}_k(j\omega)}{\dot{U}_{Sj}(j\omega)}$

转移电压比
$$H(j\omega) = \frac{\dot{U}_k(j\omega)}{\dot{U}_{Sj}(j\omega)}$$

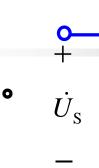
转移电流比 $H(j\omega) = \frac{\dot{I}_k(j\omega)}{\dot{I}_{Si}(j\omega)}$

转移电流比
$$H(j\omega) = \frac{I_k(j\omega)}{\dot{I}_{S_i}(j\omega)}$$





求电路的网络函数 $rac{\dot{I}_2}{\dot{U}_{ m S}}$, $rac{\dot{U}_{ m L}}{\dot{U}_{ m S}}$.



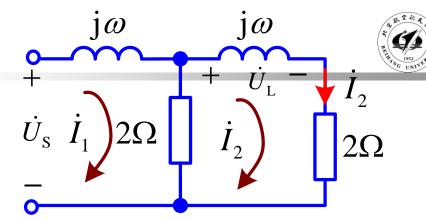


$$\begin{cases} (2 + j\omega)\dot{I}_{1} - 2\dot{I}_{2} = \dot{U}_{S} \\ -2\dot{I}_{1} + (4 + j\omega)\dot{I}_{2} = 0 \end{cases}$$

$$\dot{I}_2 = \frac{2\dot{U}_S}{4 + (j\omega)^2 + j6\omega}$$

$$\frac{\dot{I}_2}{\dot{U}_S} = \frac{2}{\left(4 - \omega^2\right) + j6\omega}$$

转移导纳

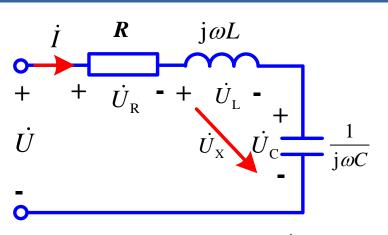


$$\frac{\dot{U}_{L}}{\dot{U}_{S}} = \frac{j2\omega}{\left(4 - \omega^{2}\right) + j6\omega}$$

转移电压比

11.2 RLC串联电路的谐振

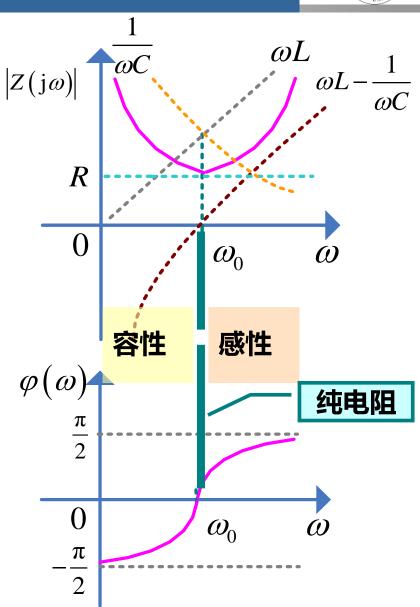




驱动点阻抗 $Z = R + j \left(\omega L - \frac{1}{\omega C}\right)$

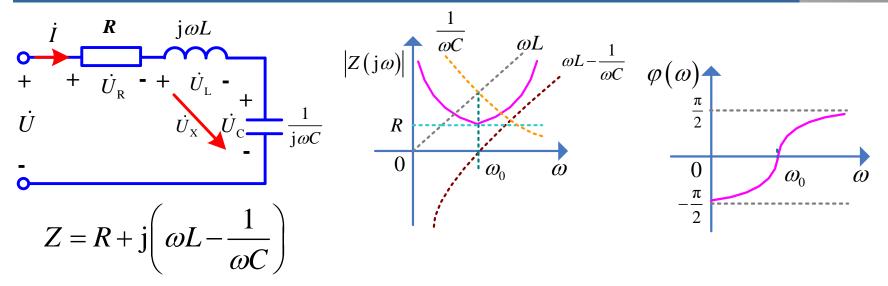
$$|Z(j\omega)| = \sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}$$

$$\varphi(j\omega) = \arctan\left(\frac{\omega L - \frac{1}{\omega C}}{R}\right)$$



11.2 RLC串联电路的谐振





1. 定义

端口电压、电流出现同相位的现象时,称 电路发生了谐振;对于RLC串联电路,则 称为串联谐振。

$$\operatorname{Im}[Z(j\omega)] = 0$$

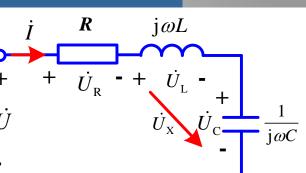
11.2 RLC串联电路的谐振



$$\operatorname{Im}[Z(j\omega)] = 0$$

*由谐振条件得串联电路实现谐振的方式为:

- (1) L 、C 不变,改变 ω 达到谐振。
- (2) 电源频率不变,改变 L 或 C (常改变 C) 达到谐振。



谐振角频率

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

谐振频率

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3. 特点 (阻抗、电流、电压、功率)



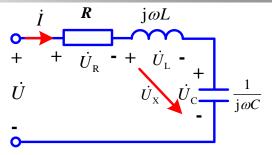
$$Z(j\omega_0) = R + j\left(\omega_0 L - \frac{1}{\omega_0 C}\right) = R$$

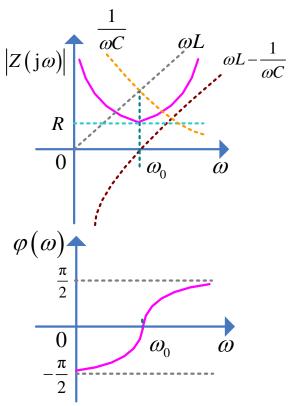
$$(1)$$
 \dot{U} 、 \dot{I} 同相 $\varphi = 0^{\circ}$

$$Z = R$$
 $|Z| = |Z|_{\min} = R$

(2) 若U一定

$$I = \frac{U}{|Z|} = \frac{U}{R} = I_{\text{max}}$$





(3)
$$\dot{U}_{R} = R\dot{I} = R\frac{\dot{U}}{R} = \dot{U}$$

$$\dot{U}_{X} = \dot{U}_{L} + \dot{U}_{C} = 0$$



电压谐振 L串C部分视作短路

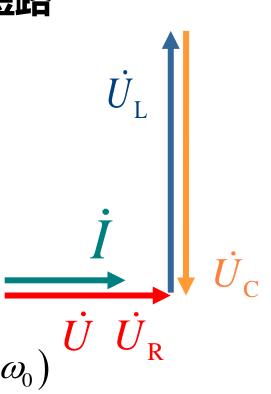
$$U_{\rm L} = U_{\rm C}$$

(4)
$$P = UI = RI^2 = \frac{U^2}{R}$$

$$Q = Q_{\rm L} + Q_{\rm C} = 0$$

$$W_{\text{grif}}(j\omega_0) = W_L(j\omega_0) + W_C(j\omega_0)$$

$$W_{\mathrm{存储}}(\mathrm{j}\omega_{0}) = \frac{1}{2}LI_{\mathrm{m}}^{2}(\mathrm{j}\omega_{0}) = \frac{1}{2}CU_{\mathrm{Cm}}^{2}(\mathrm{j}\omega_{0})$$



 $j\omega L$

4. 品质因数Q



谐振时
$$U_{\rm C} = U_{\rm L} = \omega_0 L I = \omega_0 L \frac{U}{R} = \frac{\omega_0 L}{R} U$$

定义
$$Q = \frac{U_L(\omega_0)}{U}$$

$$Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 CR} = \frac{1}{R} \sqrt{\frac{L}{C}}$$

$$Q = \frac{Q_{L}(j\omega_{0})}{P(j\omega_{0})} = \frac{|Q_{C}(j\omega_{0})|}{P(j\omega_{0})}$$

$$Q = \frac{\omega_{0}LI^{2}(j\omega_{0})}{RI^{2}(j\omega_{0})}$$

$$W(j\omega_0) = \frac{1}{2} LI_m^2(j\omega_0) = \frac{1}{2} CU_{Cm}^2(j\omega_0) = CQ^2U_S^2(j\omega_0)$$

! Q综合反映了电路三个参数对谐振状态的影响!

11.3 RLC串联电路的频率特性



驱动点阻抗
$$Z = H(j\omega) = \frac{\dot{U}(j\omega)}{\dot{I}(j\omega)}$$

转移电压比
$$H(j\omega) = \frac{\dot{U}_{R}(j\omega)}{\dot{U}_{S}(j\omega)}$$
 $H(j\omega) = \frac{\dot{U}_{L}(j\omega)}{\dot{U}_{S}(j\omega)}$ $H(j\omega) = \frac{\dot{U}_{C}(j\omega)}{\dot{U}_{S}(j\omega)}$

(1) 电阻电压频率特性
$$H_{R}(j\omega) = \frac{\dot{U}_{R}(j\omega)}{\dot{U}_{S}(j\omega)} = \frac{R}{R + j\left(\omega L - \frac{1}{\omega C}\right)}$$

$$H_{R}(j\omega) = \frac{1}{1 + j\left(\frac{\omega L}{R} - \frac{1}{R\omega C}\right)} = \frac{1}{1 + j\left(\frac{\omega}{\omega_{0}} | \frac{\omega_{0}L}{R} - \frac{1}{R\omega_{0}C} | \frac{\omega_{0}}{\omega}\right)}$$

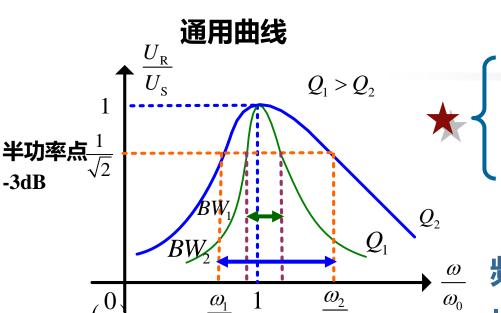
$$Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 CR} \qquad \eta = \frac{\omega}{\omega_0}$$



$$H(j\eta) = \frac{1}{1 + jQ\left(\eta - \frac{1}{\eta}\right)}$$



$$\begin{cases} |H(j\eta)| = \frac{1}{\sqrt{1 + Q^2 \left(\eta - \frac{1}{\eta}\right)^2}} \\ \varphi(j\eta) = -\arg \tan \left[Q\left(\eta - \frac{1}{\eta}\right)\right] \end{cases}$$



$$|H(j\eta)| = \frac{1}{\sqrt{1 + Q^2 \left(\eta - \frac{1}{\eta}\right)^2}}$$

$$\varphi(j\eta) = -\arg \tan \left[Q \left(\eta - \frac{1}{\eta} \right) \right]$$

频率选择性: Q 越大越好

抑制能力: Q 越大抑制能力越强

★ 通频带 (通带)

$$\frac{\pi}{2}$$

 ω_0

 ω_0

品质因数试验方法

$$Q = \frac{\omega_0}{\omega_2 - \omega_1} = \frac{f_0}{f_2 - f_1} = \frac{1}{\eta_2 - \eta_1}$$

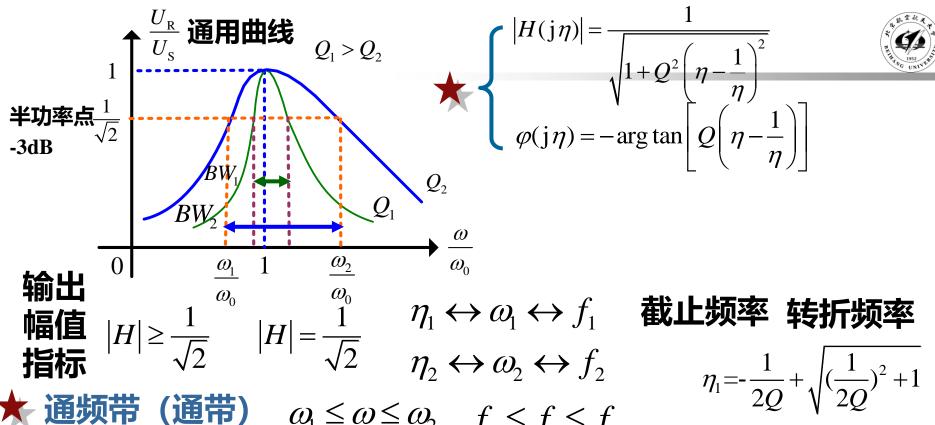
! Q反映电路选择性的好坏!

电路 自动化科学与电气工程学院

通频带带宽

 $\mathbf{0}$

 π



$$\eta_1 = -\frac{1}{2Q} + \sqrt{(\frac{1}{2Q})^2 + 1}$$

$$\eta_2 = \frac{1}{2Q} + \sqrt{(\frac{1}{2Q})^2 + 1}$$

$$|H| \ge \frac{1}{\sqrt{2}}$$

$$|H| = \frac{1}{\sqrt{2}}$$

$$n_2 \leftrightarrow \omega_2 \leftrightarrow f_2$$

$$\eta_2 \leftrightarrow \omega_2 \leftrightarrow f_2$$

通频带(通带)
$$\omega_1 \le \omega \le \omega_2$$
 $f_1 \le f \le f_2$

帶竞
$$BW = \begin{cases} \omega_2 - \omega_1 \\ f_2 - f_1 \\ \eta_2 - \eta_1 \end{cases}$$

$$Q = \frac{\omega_0}{BW} \sim \frac{f_0}{BW} \sim \frac{1}{BW}$$

波特图 $\begin{cases} H_{dB} = 20 \lg |H(j\omega)| \\ \log \omega \end{cases}$ $\begin{cases} \varphi(j\omega) \\ \log \omega \end{cases}$

$$\begin{cases} \varphi(j\omega) \\ \log \omega \end{cases}$$

(2) 电感电压和电容电压频率特性



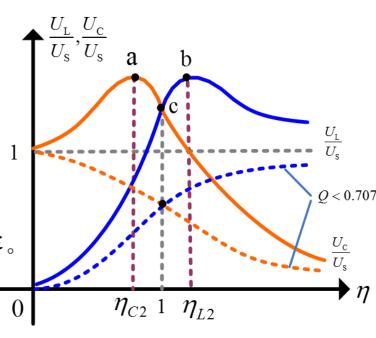
$$H_{C}(j\eta) = \frac{\dot{U}_{C}(j\eta)}{\dot{U}_{S}(jl)} = \frac{-jQ}{\eta + jQ(\eta^{2} - 1)}$$

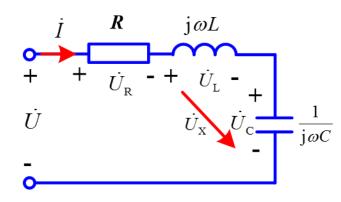
低通频率特性

若Q>>1,则 $U_c>>U_s$,可应用于无线电系统; 而电力系统应避免谐振,以免 U_c 过大,击毁电容。

$$H_{L}(j\eta) = \frac{\dot{U}_{L}(j\eta)}{\dot{U}_{S}(jl)} = \frac{jQ}{\frac{1}{\eta} + jQ\left(1 - \frac{1}{\eta^{2}}\right)}$$

高通频率特性





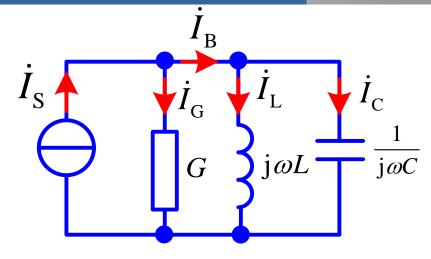
11.4 RLC并联谐振电路



$$Y = G + j(\omega C - \frac{1}{\omega L})$$

1. 定义

端口电压、电流出现同相位的现象时,称电路发生了谐振;对于RLC并联电路,则称为并联谐振。



2. 条件

$$I_{m}[Y(j\omega)] = 0$$

谐振角频率

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

谐振频率

3. 特点 (导纳、电压、电流、功率)



(1) \dot{U} 、 \dot{I} 同相, $\varphi = 0^{\circ}$

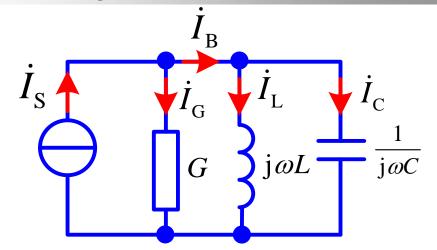
$$Y = G$$

$$|Y| = |Y| \min = G$$

$$|Z| = \frac{1}{|Y|} = |Z|_{\text{max}}$$

(2) 若I一定

$$U = I |Z| = IR = U_{\text{max}}$$

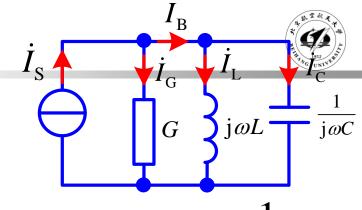


$$Y = G + j(\omega C - \frac{1}{\omega L})$$

(3)
$$\dot{I}_{\rm B} = \dot{I}_{\rm L} + \dot{I}_{\rm C} = 0$$

$$I_{\rm L} = I_{\rm C}$$

$$I_{\rm S} = I_{\rm G}$$



$$Y = G + j(\omega C - \frac{1}{\omega L})$$



电流谐振 L并C视作开路

(4)
$$P = UI = RI^2 = \frac{U^2}{R}$$

$$Q = Q_{\rm L} + Q_{\rm C} = 0$$

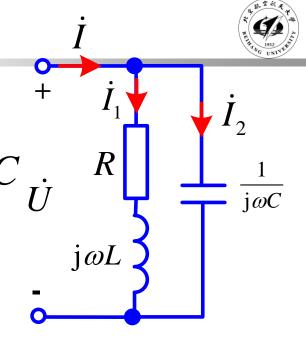
$$Y = \frac{1}{R + j\omega L} + j\omega C$$

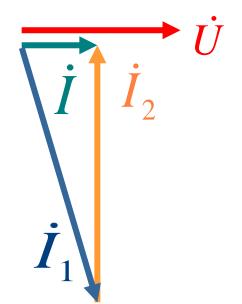
$$= \frac{R}{R^2 + (\omega L)^2} - j\frac{\omega L}{R^2 + (\omega L)^2} + j\omega C_{\dot{U}}$$

$$I_{m}[Y] = 0$$

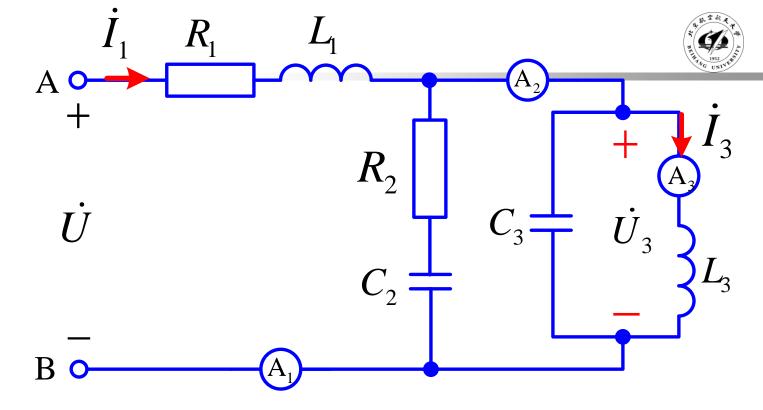
$$\omega C - \frac{\omega L}{R^2 + (\omega L)^2} = 0$$

$$\omega_0 = \frac{1}{\sqrt{LC}} \sqrt{1 - \frac{CR^2}{L}}$$





【例】

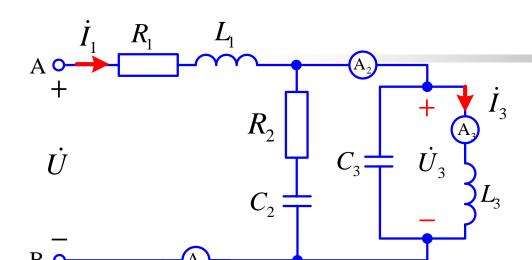


三知:
$$R_1 = 50\Omega$$
 $R_2 = 50\Omega$ $L_1 = 200$ mH $C_2 = 5$ μF $L_3 = 100$ mH $C_3 = 10$ μF $U = 200$ V

电流表A2指示为零,所有电流表内阻忽略不计

- 求: (1) A1, A3的读数
 - (2) 输入AB端的功率和功率因数。





解

(1)
$$\omega C_3 = \frac{1}{\omega L_3}$$

$$\omega = \frac{1}{\sqrt{L_3 C_3}} = \frac{1}{\sqrt{100 \times 10^{-3} \times 10 \times 10^{-6}}} = 10^3 \text{ rad/s}$$



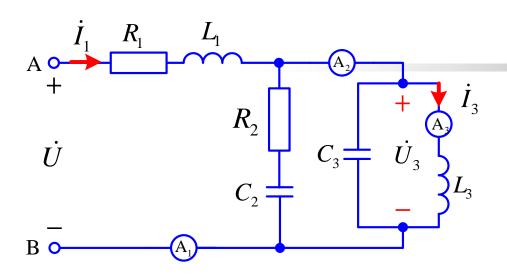
$$Z = (R_1 + R_2) + j \left(\omega L_1 - \frac{1}{\omega C_2}\right)$$

$$= 50 + 50 + j \left(10^3 \times 200 \times 10^{-3} - \frac{1}{10^3 \times 5 \times 10^{-6}}\right)$$

$$= 100 \angle 0^0 \Omega$$

$$I_1 = \frac{U}{|Z|} = \frac{200}{100} = 2A$$





$$U_3 = I_1 |Z_2| = I_1 \sqrt{R_2^2 + \left(\frac{1}{\omega C_2}\right)^2} = 412.3 \text{V}$$

$$I_3 = \frac{U_3}{\omega L_3} = \frac{412.3}{10^3 \times 100 \times 10^{-3}} = 4.12A$$



$$\varphi = 0^{\circ}$$

$$\cos \varphi = 1$$

$$P = UI_1 \cos \varphi = 200 \times 2 = 400 \text{W}$$

或
$$P = I_1^2 (R_1 + R_2) = 2^2 \times (50 + 50) = 400$$
W

【例】





既有串联谐振, 也有并联谐振。

先求
$$Z$$
, $I_m[Z] = 0$, 串联谐振

再求
$$Y$$
, $I_m[Y] = 0$,并联谐振

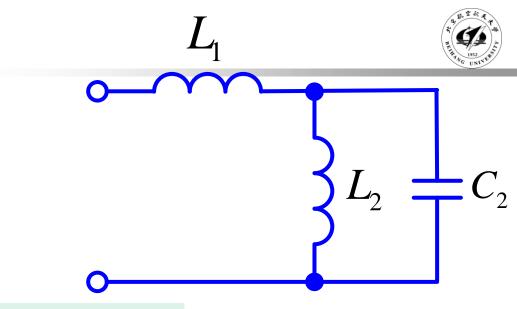
$$Z = j\omega L_{1} + \frac{j\omega L_{2}\left(-j\frac{1}{\omega C_{2}}\right)}{j\omega L_{2} - j\frac{1}{\omega C_{2}}} = j\frac{\omega^{3}L_{1}L_{2}C_{2} - \omega(L_{1} + L_{2})}{\omega^{2}L_{2}C_{2} - 1}$$

串联谐振
$$I_m[Z] = 0$$

$$\omega_{01} = \sqrt{\frac{L_1 + L_2}{L_1 L_2 C_2}}$$

并联谐振 $I_m[Y] = 0$

$$\omega_{02} = \sqrt{\frac{1}{L_2 C_2}}$$



谐振特点: 电流电压同相; LC间能量交换,与 外部没有能量交换

低频段,并联环节呈感性,整个电路呈感性;

 ω 上升,达到 ω_{02} ,发生并联谐振;

 ω 继续上升,并联环节呈容性,当 $\omega = \omega_{01}$ 时, 发生串联谐振。

【练习】

分析图示电路的串并联谐振。



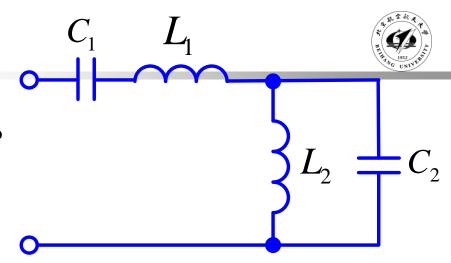
既有串联谐振, 也有并联谐振。

求
$$Z$$
, $I_m[Z] = 0$, 串联谐振

求
$$Y$$
, $I_m[Y] = 0$,并联谐振

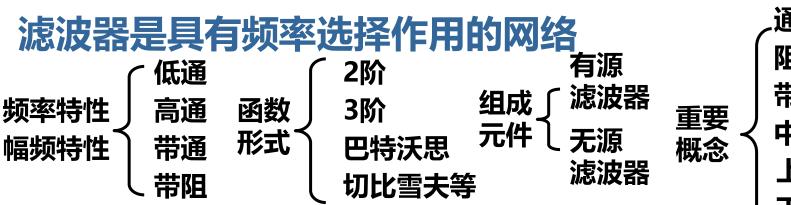
 C_2, L_2 并联,并联谐振

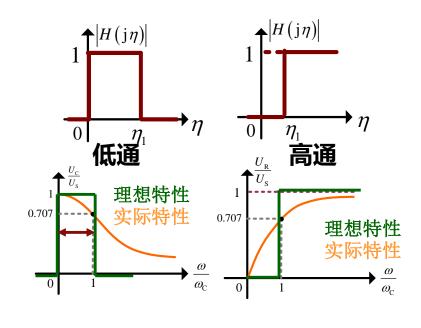
 C_1, L_1 串联, C_2, L_2 并联, 整个电路串联谐振

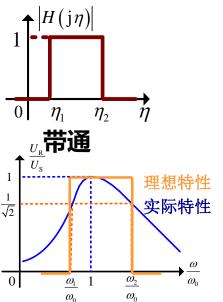


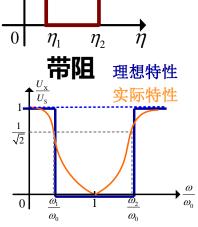
11.5 滤波器简介









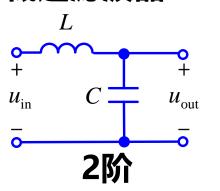


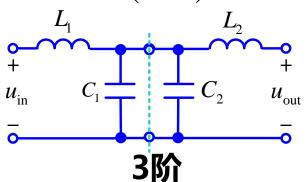
 $1 \uparrow^{|H(j\eta)|}$

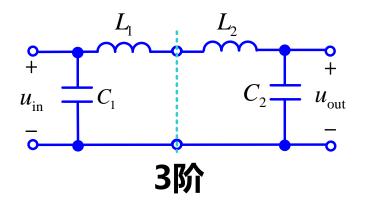
11.5 滤波器简介



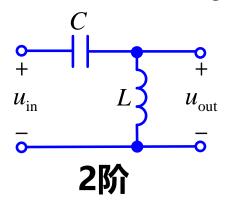
低通滤波器Low Pass Filter(LPF)

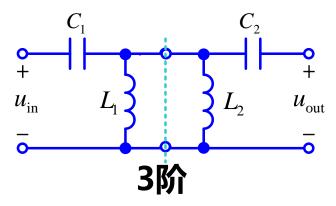


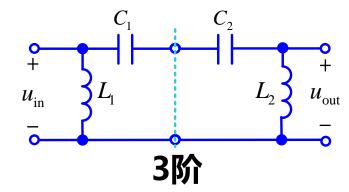




高通滤波器High Pass Filter(HPF)



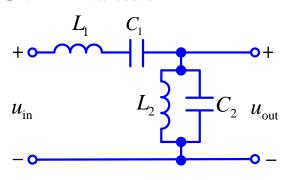


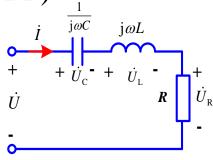


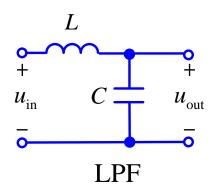
11.5 滤波器简介 有源滤波器 无源滤波器



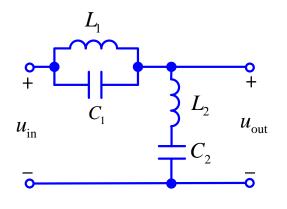
带通滤波器Band Pass Filter(BPF)

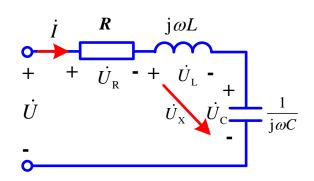


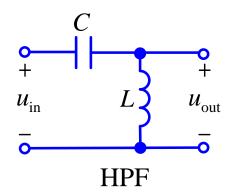




带阻滤波器Band Reject Filter(BRF)







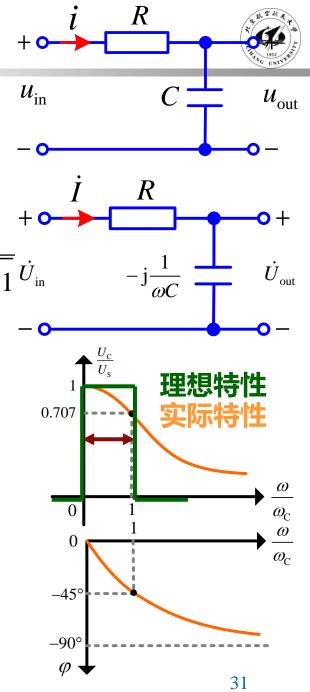
(1) RC低通滤波器

$$\dot{U}_{\text{out}} = \frac{-j\frac{1}{\omega C}}{R - j\frac{1}{\omega C}}\dot{U}_{\text{in}} = \frac{1 - jR\omega C}{\left(R\omega C\right)^2 + 1}\dot{U}_{\text{in}}$$

$$U_{\text{out}} = \frac{1}{\sqrt{(R\omega C)^2 + 1}} U_{\text{in}} \left| H(j\omega) \right| = \frac{U_{\text{out}}}{U_{\text{in}}} = \frac{1}{\sqrt{(R\omega C)^2 + 1}} \dot{U}_{\text{in}}$$

$$\omega_{\rm C} = \frac{1}{\tau} = \frac{1}{RC} \qquad \left| H(j\omega) \right| = \frac{1}{\sqrt{\left(\frac{\omega}{\omega_{\rm C}}\right)^2 + 1}}$$

是一个截止频率为 ω 低通滤波器。



(2) RC高通滤波器

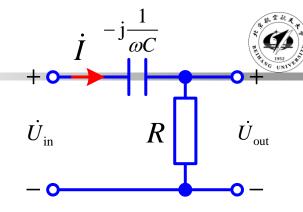
$$\dot{U}_{\text{out}} = \frac{R}{R - j \frac{1}{\omega C}} \dot{U}_{\text{in}} = \frac{R\omega C (R\omega C + j)}{(R\omega C)^2 + 1} \dot{U}_{\text{in}}$$

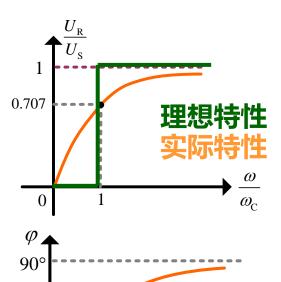
$$U_{\text{out}} = \frac{R\omega C}{\sqrt{\left(R\omega C\right)^2 + 1}} U_{\text{in}} = \frac{1}{\sqrt{1 + \left(\frac{1}{R\omega C}\right)^2}} U_{\text{in}}$$

$$|H(j\omega)| = \frac{U_{\text{out}}}{U_{\text{in}}} = \frac{1}{\sqrt{1 + \left(\frac{1}{R\omega C}\right)^2}}$$

$$\omega_{\rm C} = \frac{1}{\tau} = \frac{1}{RC} \qquad |H(j\omega)| = \frac{1}{\sqrt{\left(\frac{\omega_{\rm C}}{\omega}\right)^2 + 1}}$$

是一个截止频率为 $\omega_{\mathbb{C}}$ 高通滤波器。



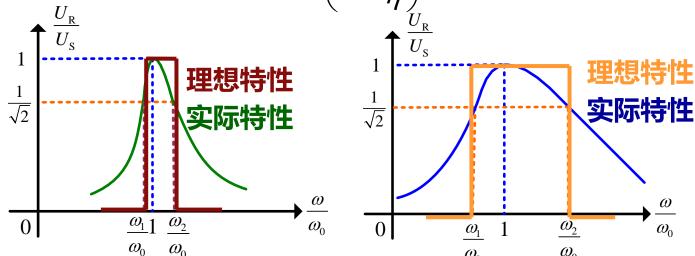


(3) *RLC*带通滤波器

(5)
$$RLC$$
 FIEURIX AS
$$H_{R}(j\omega) = \frac{\dot{U}_{R}(j\omega)}{\dot{U}_{S}(j\omega)} = \frac{R}{R + j\left(\omega L - \frac{1}{\omega C}\right)}$$

$$\eta = \frac{\omega}{\omega_0}$$

$$H(j\eta) = \frac{1}{1 + jQ\left(\eta - \frac{1}{\eta}\right)}$$



通频带为 ω_1,ω_2 之间,带通滤波器。

jωL

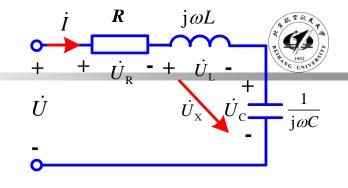
 ω_0

(4) RLC带阻滤波器

$$H_{X}(j\omega) = \frac{\dot{U}_{X}(j\omega)}{\dot{U}_{S}(j\omega)} = \frac{j\left(\omega L - \frac{1}{\omega C}\right)}{R + j\left(\omega L - \frac{1}{\omega C}\right)}$$

$$\eta = \frac{\omega}{\omega_0}$$

$$H(j\eta) = \frac{jQ\left(\eta - \frac{1}{\eta}\right)}{1 + jQ\left(\eta - \frac{1}{\eta}\right)}$$



理想特性 $\frac{U_x}{U_s}$ 实际特性 $\frac{1}{\sqrt{2}}$ $\frac{\omega_1}{\omega_0}$ $\frac{\omega_2}{\omega_0}$ $\frac{\omega}{\omega_0}$

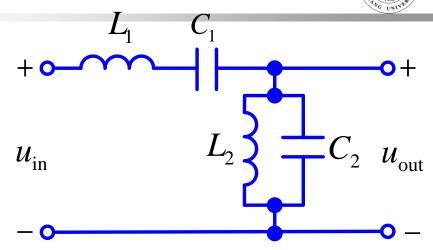
阻频带为 ω_1,ω_2 之间,带阻滤波器。



思考:

 $u_{\rm in}$ 中有多个频率谐波信号,

希望 U_{out} 中只有一种 频率信号无衰减,分 析电路工作状态。



$$L_1C_1$$
 串联谐振 L_2C_2 并联谐振 L_2

其谐振频率是输出信号频率。