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The code used to generate all plots is given at the end of this document.

# P2.2

Below are the plots generated according to the instructions in P2.2 of the class textbook.

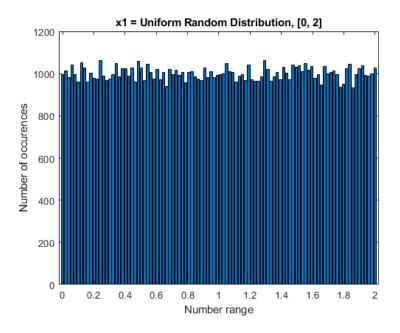


Figure 1: P2.2.1

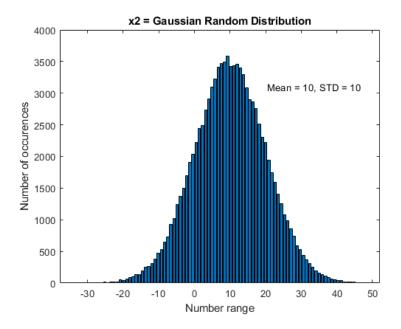


Figure 2: P2.2.2

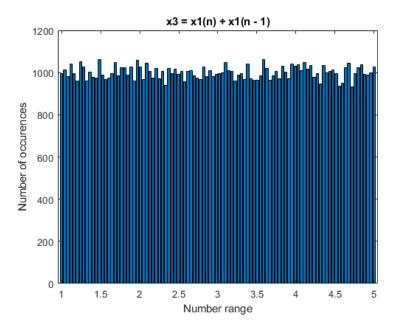


Figure 3: P2.2.3

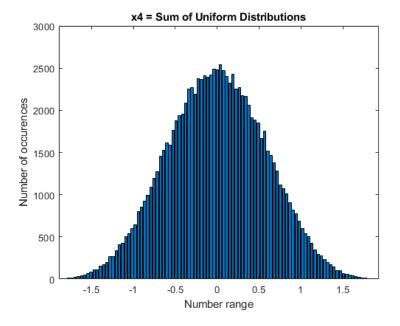


Figure 4: P2.2.4

Figure 3 and 4 were generated similarly: They are sums of multiple random variables uniformly distributed. Figure 3 is the sum of the random variable shown in Figure 1 added with an index-shifted version of itself, while Figure 4 is the sum of four individual random variables uniformly distributed over the range [-1/2, 1/2]. The difference between the two plots is self-evident: the former resembles a uniform distribution, while the latter resembles a normal distribution. This is due to the central limit theorem: informally, as one sums independent random variables with the same distribution, the sum of the random variables is itself normally distributed.

## P2.3

The following are plots generated for sections 2 and 3 of Problem 2.3.

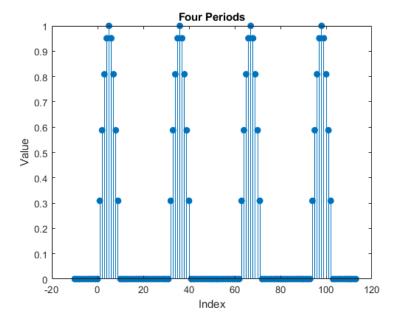


Figure 5: P2.3.3

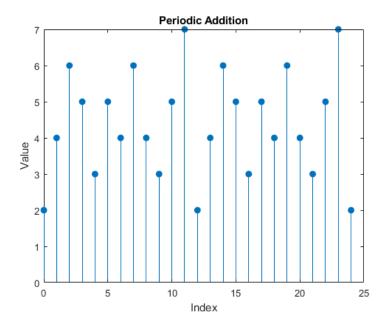


Figure 6: P2.3.4

The period of Figure 6 is the number of indexes between the lowest two points (in this case, when the value is 2); this means that the period is 12. The value 2 occurs at index 0, 12, and 24. This is because the periods of the two periodic signals used to create this plot are 3 and 4 (and  $3 \times 4 = 12$ ).

#### P2.5

The following theorem was given: "The complex exponential sequence  $\exp(j\omega_0 n)$  or the sinusoidal sequence  $\cos(\omega_0 n)$  are periodic if the *normalized* frequency  $f_0 \equiv \omega_0/2\pi$  is a rational number; that is,  $f_0 = K/N$  such that  $K, N \in \mathbb{Z}$ ."

Note that the following proof is for  $\cos(\omega_0 n)$ , but applies to both the real and imaginary parts of  $\exp(j\omega_0 n)$  as well, due to Euler's formula that states that  $\operatorname{Re}\left(e^{jx}\right) = \cos(x)$  and  $\operatorname{Im}\left(e^{jx}\right) = \sin(x)$ . In addition,  $\sin(\pi/2 - x) = \cos(x)$ . Therefore, this proof applies to both cases.

In the continuous case, it is always true that a sinusoid such as the cosine function is periodic. However, for the discrete case, the normalized frequency  $f_0 \equiv \omega_0/2\pi$  must be a rational number. Remember that a sequence x(n) is periodic if

$$x[n] = x[n+P] \text{ for all } n, \tag{1}$$

where P is the fundamental period (textbook, p. 25). In the case of  $f_0 \equiv K/N$ , N will be the fundamental period (i.e., x(n) = x(n+N) for all n). For example, if  $f_0 = 3/8$ , then 8 will be the fundamental period (x(n) = x(n+8)) for all n). The number 3 is the amount of continuous periods the sinusoid must go through before a period starts on an integer index. For example, for the signal  $x(n) = \cos(0.75\pi n)$ , x(0) = 1. x(8/3),  $x(2 \times 8/3)$ , and  $x(3 \times 8/3)$  all also equal 1, but  $3 \times 8/3 = 8$  is the only index that is an integer and thus valid for a discrete signal; in the discrete case, x[0] = 1, and the first sample that will also equal 1 is x[8]; thus, 8 will be the fundamental period. If  $f_0$  is not a rational number, N is not defined/does not exist. Therefore, for a discrete signal to be periodic,  $f_0$  must be rational.

The following are plots generated for Problem 2.5. Figure 7 was generated using  $z = \exp(j0.1\pi n)$ , and Figure 8 was generated using  $x = \cos(0.1n)$ .

<sup>&</sup>lt;sup>1</sup>This was discovered numerically by using the Desmos graphing calculator to graph sinusoids, available at https://www.desmos.com/calculator.

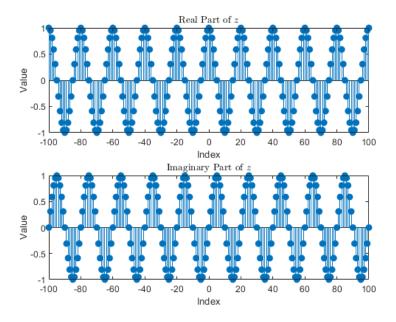


Figure 7: P2.5.2

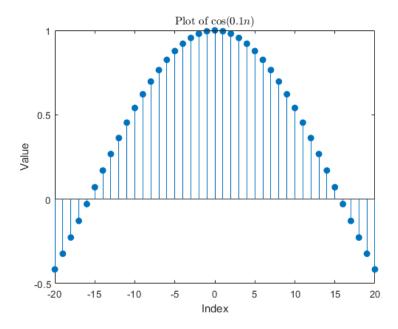


Figure 8: P2.5.3

Both the real and imaginary parts of Figure 7 are indeed periodic. The normalized frequency is

 $f_0 = \frac{0.1\pi}{2\pi} = \frac{1}{20}.$ 

Thus, K = 1 and N = 20, and its fundamental period is N = 20. This can be seen graphically by noticing that the value of zero is achieved on the indexes that are integer multiples of 20 ( $\{20k \mid k \in \mathbb{Z}\}$ ).

The sinusoid presented in Figure 8 is not periodic. It certainly looks like it is, but numerically, it does not satisfy the definition of periodicity (Equation (1)). Its normalized frequency is

$$f_0 = \frac{0.1}{2\pi} = \frac{1}{20\pi}.$$

K=1 and  $N=20\pi$ , so its fundamental frequency is  $20\pi$ . However, since this is not an integer, it cannot be used in the practice of discrete signals, where all indexes are integers. Thus, it is not periodic.

## P2.6

The following are plots generated for sections 3 and 4 of Problem 2.6. These graphs were obtained by using the evenodd function provided in the textbook (p. 34).

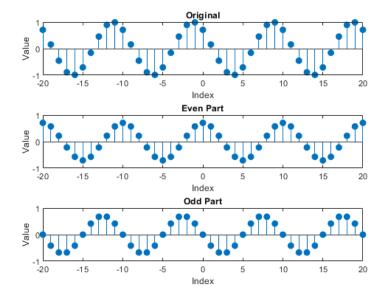


Figure 9: P2.6.3

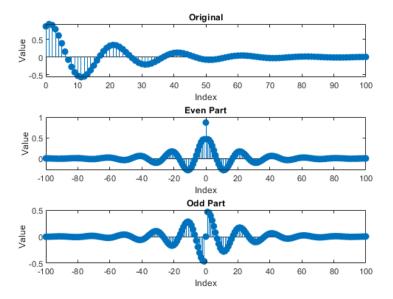


Figure 10: P2.6.4

## P2.10

Let  $y(n) = x(n) + \alpha x(n-k)$ , where x(n) is a signal, and  $\alpha x(n-k)$  is its echo (noise). The cross-correlation between y and x is given by

$$r_{yx}(\ell) = \sum_{n=-\infty}^{\infty} y(n)x(n-\ell)$$

by definition (the complex conjugate is not needed because both functions are real-valued). Substituting  $y(n) = x(n) + \alpha x(n-k)$ , we have

$$r_{yx}(\ell) = \sum_{n=-\infty}^{\infty} (x(n) + \alpha x(n-k))x(n-\ell)$$

$$= \sum_{n=-\infty}^{\infty} [x(n)x(n-\ell) + \alpha x(n-k)x(n-\ell)]$$

$$= \sum_{n=-\infty}^{\infty} x(n)x(n-\ell) + \sum_{n=-\infty}^{\infty} \alpha x(n-k)x(n-\ell)$$

$$= (x(n) \star x(n))(\ell) + \alpha(x(n-k) \star x(n))(\ell)$$

where  $(x(n) \star x(n))(\ell) = r_{xx}(\ell)$  is the autocorrelation of x(n).

To generate Figure 11, I used the following definitions:

$$x(n) = \cos(0.2\pi n) + 0.5\cos(0.6\pi n)$$
  

$$\alpha = 0.1, k = 50.$$

Note that the plot of x(n) ends at index 200, and the plot of  $\alpha x(n-k)$  starts at index 50. Theoretically, you can find the value of k by observing where  $r_{yx}(\ell)$  reaches its maximum. It is not possible to find the value of  $\alpha$ , however.

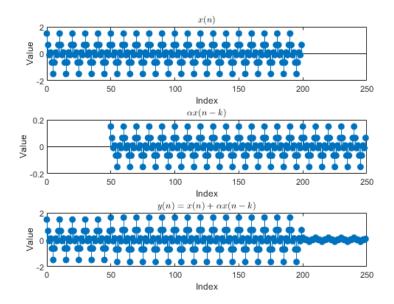


Figure 11: P2.10.2

# P2.11

#### System 1

$$\begin{split} T[x(n)] &= x(n)u(n) \\ aT[x(n)] + bT[y(n)] &= ax(n)u(n) + by(n)u(n) \\ T[ax(n) + by(n)] &= (ax(n) + by(n))u(n) = ax(n)u(n) + by(n)u(n) \\ ax(n)u(n) + by(n)u(n) &\stackrel{?}{=} ax(n)u(n) + by(n)u(n) \end{split}$$

The above equality is satisfied; thus, System 1 is indeed linear.

#### System 2

$$T[x(n)] = x(n) + nx(n+1)$$

$$aT[x(n)] + bT[y(n)] = ax(n) + anx(n+1) + by(n) + bny(n+1)$$

$$T[ax(n) + by(n)] = ax(n) + by(n) + n(ax(n+1) + by(n+1))$$

$$= ax(n) + by(n) + anx(n+1) + bny(n+1)$$

$$ax(n) + by(n) + anx(n+1) + bny(n+1)$$

$$\stackrel{?}{=} ax(n) + anx(n+1) + by(n) + bny(n+1)$$

The above equality is satisfied; thus, System 2 is indeed linear.

#### System 3

$$T[x(n)] = x(n) + \frac{1}{2}x(n-2) - \frac{1}{3}x(n-3)x(2n)$$

$$aT[x(n)] + bT[y(n)] = ax(n) + \frac{a}{2}x(n-2) - \frac{a}{3}x(n-3)x(2n)$$

$$+by(n) + \frac{b}{2}y(n-2) - \frac{b}{3}y(n-3)y(2n)$$

$$T[ax(n) + by(n)]$$

$$= ax(n) + by(n) + \frac{1}{2}(ax(n-2) + by(n-2))$$

$$-frac13(ax(n-3) + by(n-3))(ax(2n) + by(2n))$$

I can stop here, because as you can see on the last line, I am going to get  $a^2$ , which is not present in aT[x(n)] + bT[y(n)]. Therefore, System 3 is not linear.

#### System 4

$$T[x(n)] = \sum_{k=-\infty}^{n+5} 2x(k)$$

$$aT[x(n)] + bT[y(n)] = 2a \sum_{k=-\infty}^{n+5} x(k) + 2b \sum_{k=-\infty}^{n+5} y(k)$$

$$T[ax(n) + by(n)] = 2 \sum_{k=-\infty}^{n+5} (ax(k) + by(k))$$

$$= 2a \sum_{k=-\infty}^{n+5} x(k) + 2b \sum_{k=-\infty}^{n+5} y(k)$$

$$aT[x(n)] + bT[y(n)] \stackrel{?}{=} T[ax(n) + by(n)]$$

The above equality is satisfied; thus, System 4 is indeed linear.

#### System 5

$$T[x(n)] = x(2n)$$

$$aT[x(n)] + bT[y(n)] = ax(2n) + by(2n)$$

$$T[ax(n) + by(n)] = ax(2n) + by(2n)$$

$$aT[x(n)] + bT[y(n)] \stackrel{?}{=} T[ax(n) + by(n)]$$

The above equality is satisfied; thus, System 5 is indeed linear.

## System 6

$$T[x(n)] = \operatorname{round}(x(n))$$

$$aT[x(n)] + bT[y(n)] = a\operatorname{round}(x(n)) + b\operatorname{round}(y(n))$$

$$T[ax(n) + by(n)] = \operatorname{round}(ax(n) + by(n))$$

$$aT[x(n)] + bT[y(n)] \stackrel{?}{=} T[ax(n) + by(n)]$$

The above equality is not satisfied; thus, System 6 is not linear.

#### System 7

$$\begin{split} T[x(n)] &= x(-n) \\ aT[x(n)] + bT[y(n)] &= ax(-n) + by(-n) \\ T[ax(n) + by(n)] &= ax(-n) + by(-n) \\ aT[x(n)] + bT[y(n)] &\stackrel{?}{=} T[ax(n) + by(n)] \end{split}$$

Yes this is linear.

$$T[x(n-k)] = x(-n-k)$$
 
$$y(n-k) = x(-(n-k)) = x(-n+k)$$

Time variant.