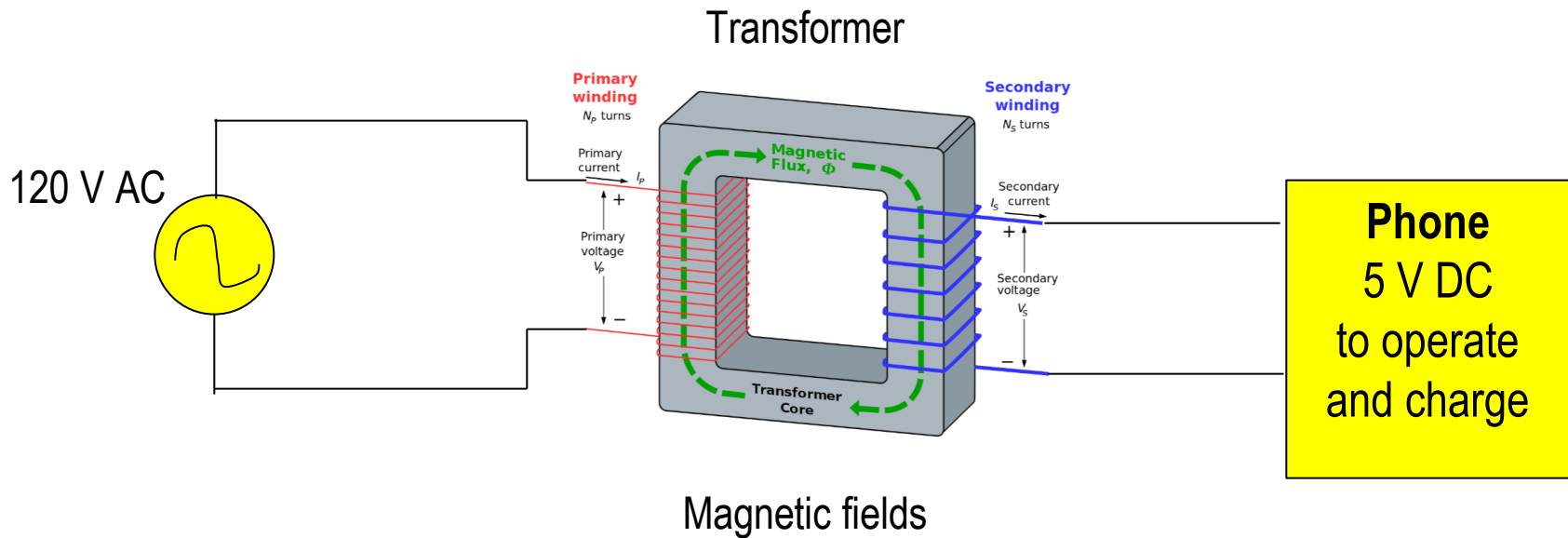
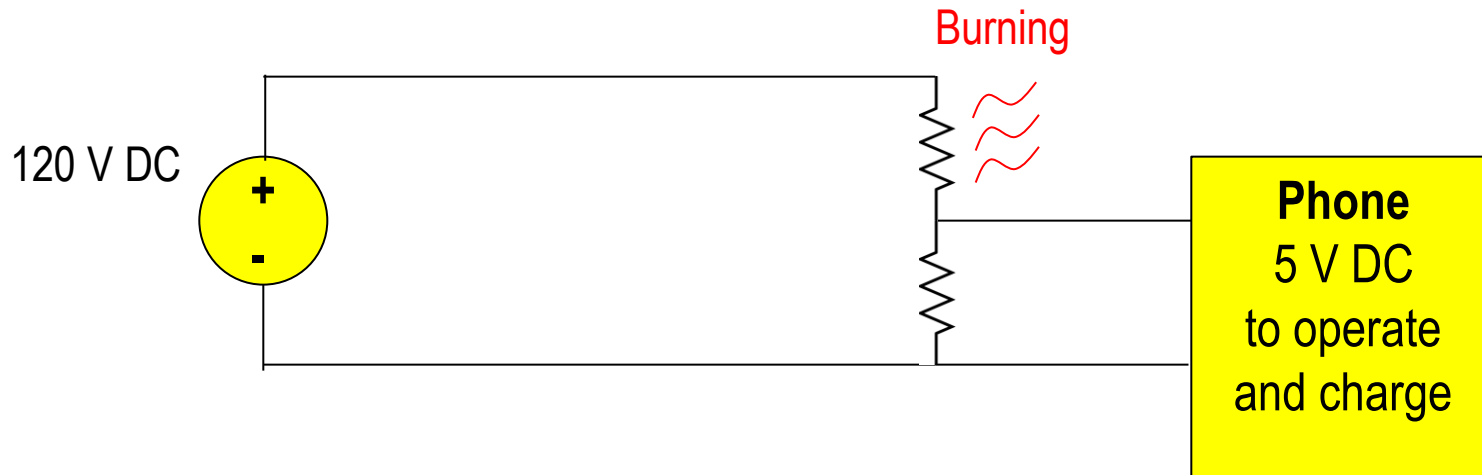


# Chapter 10

## ***Sinusoidal Steady-State Analysis***

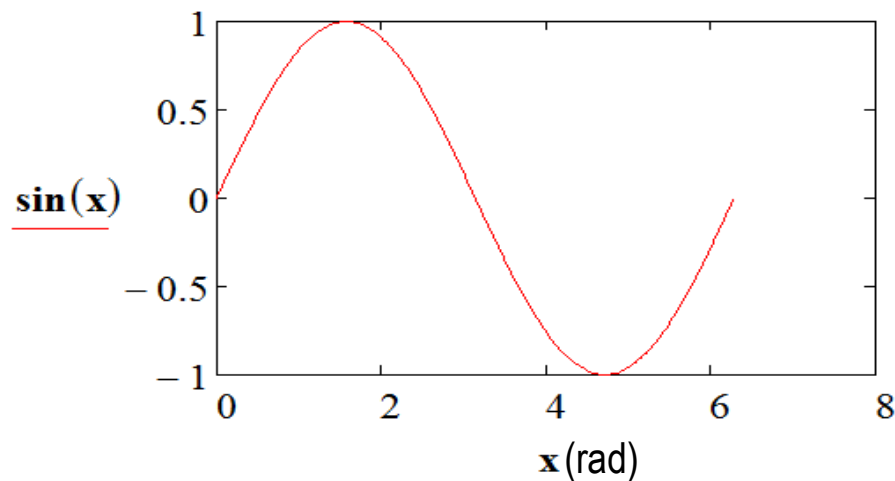
# Direct Current vs. Alternating Current (Loss)



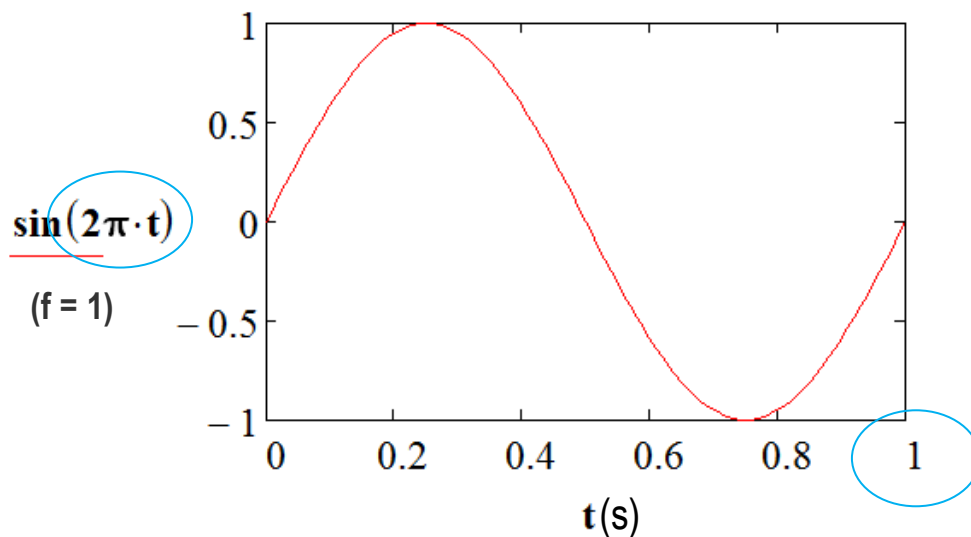
$$\sin(x)$$


$$v_s = V_m \sin(\omega t + \phi)$$

where  $\omega = 2\pi f$



Convert  $\sin(x)$  vs.  $x$  (rad) to  $\sin(t)$  vs. time (s)  
Plot 1 sinewave per second

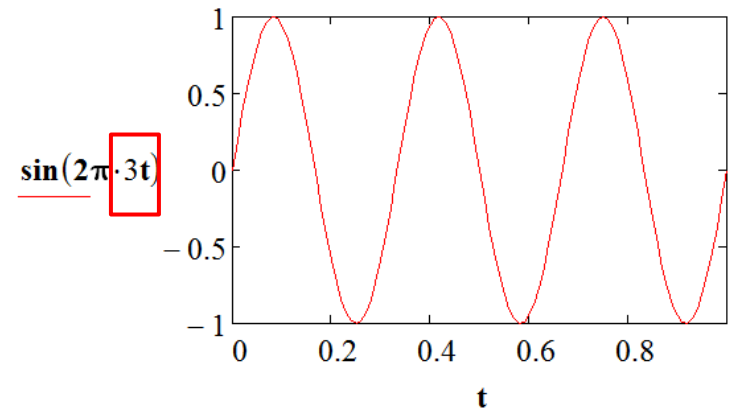
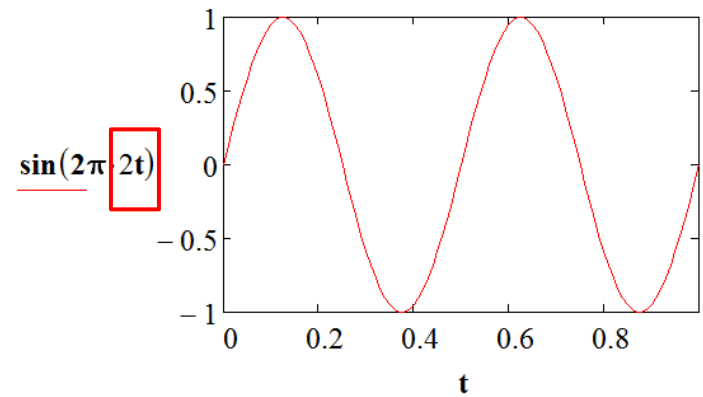
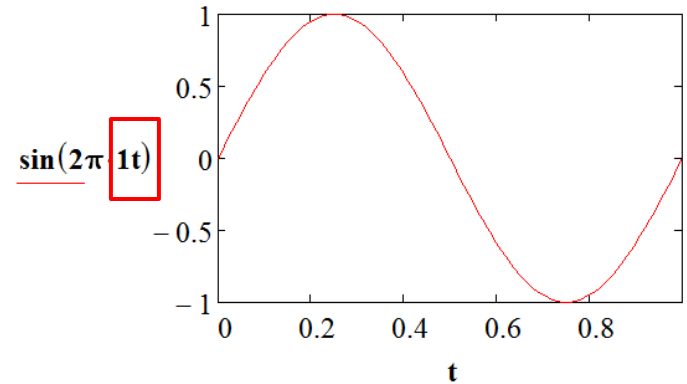


# Sinusoidal Sources – Frequency

$$\sin(2 \cdot \pi \cdot f \cdot t)$$



Frequency: Number of cycles per second  
Unit: Hz



$$\sin(2 \cdot \pi \cdot \boxed{1/T} \cdot t)$$

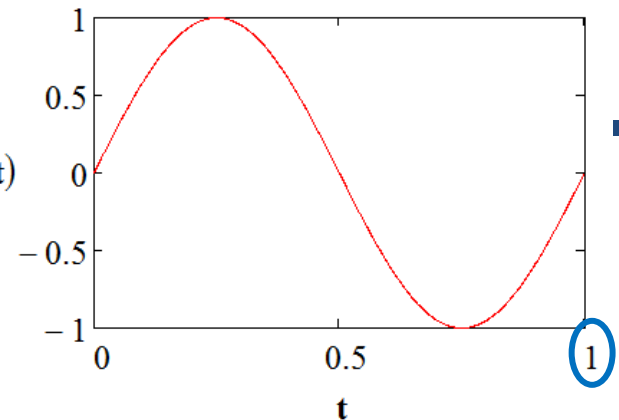


Period: Duration of one cycle in a repeating event

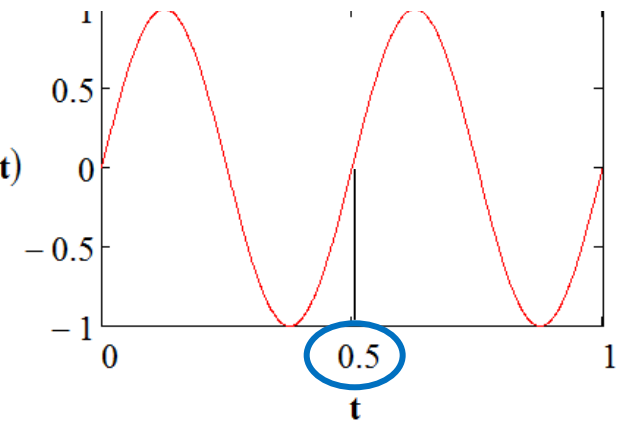
$$T = 1/f$$

Unit: s (time)

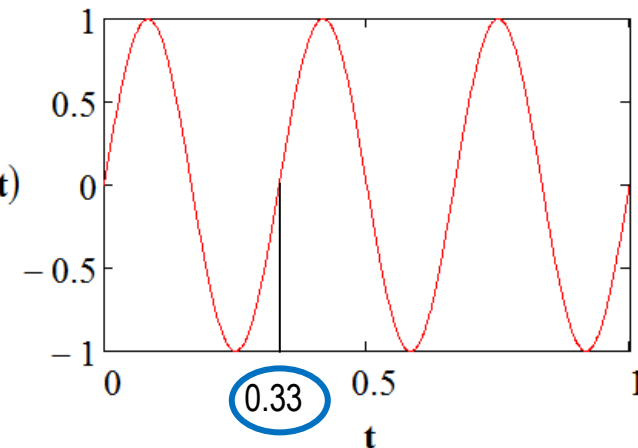
$$\sin(2\pi \cdot 1 \cdot t)$$



$$\sin(2\pi \cdot 2 \cdot t)$$



$$\sin(2\pi \cdot 3 \cdot t)$$



# Sinusoidal Sources – Summary

- A sinusoid is a periodic function defined by the property:

$$x(t+T) = x(t)$$

$T$  = period

$f = 1/T$  = frequency (number of cycles per second) [Hz]

$\omega$  = angular frequency (radian frequency) =  $2 \times \pi \times f$  [rad/s]

$$v_s = V_m \sin(\omega t)$$

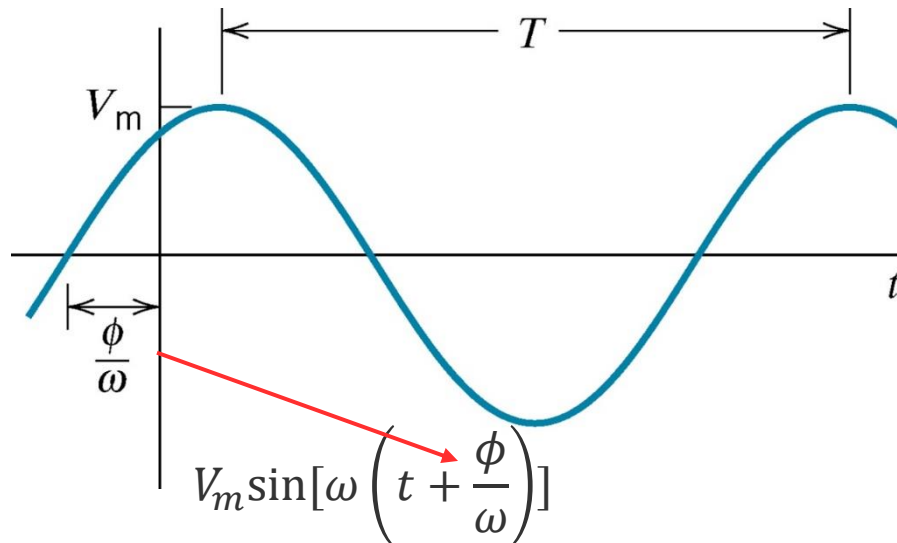
$$i_s = I_m \sin(\omega t)$$

# Sinusoidal Sources

$$v_s = V_m \sin(\omega t + \phi)$$

Maximum magnitude

Phase angle (phase shift)  
in rad or degrees



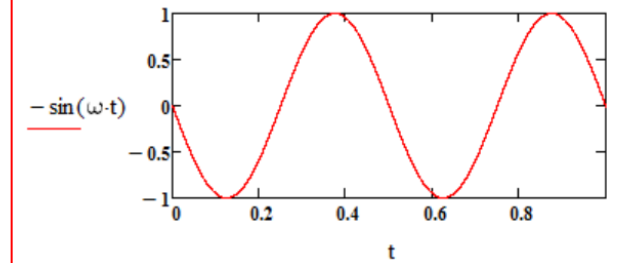
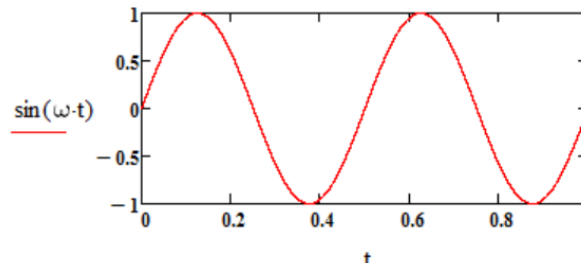
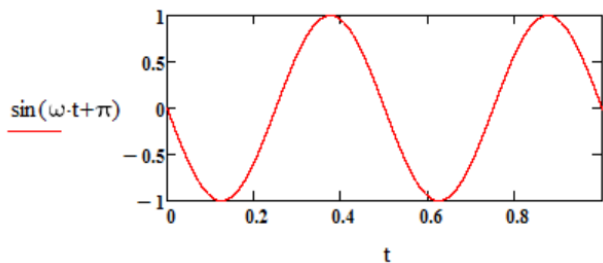
$$\omega = 2 \cdot \pi \cdot f$$
$$T = 1/f$$

# Review – Important Formulas

$$\sin(\omega t) = \cos(\omega t - 90^\circ) = \cos\left(\omega t - \frac{\pi}{2}\right)$$

$$\cos(\omega t) = \sin(\omega t + 90^\circ) = \sin\left(\frac{\pi}{2} + \omega t\right)$$

$$\sin(\omega t + 180^\circ) = -\sin(\omega t)$$



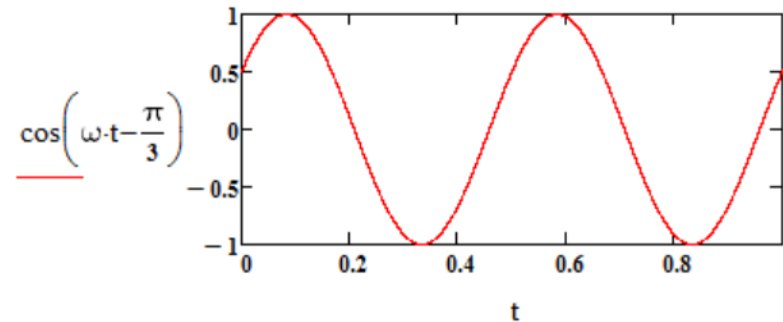
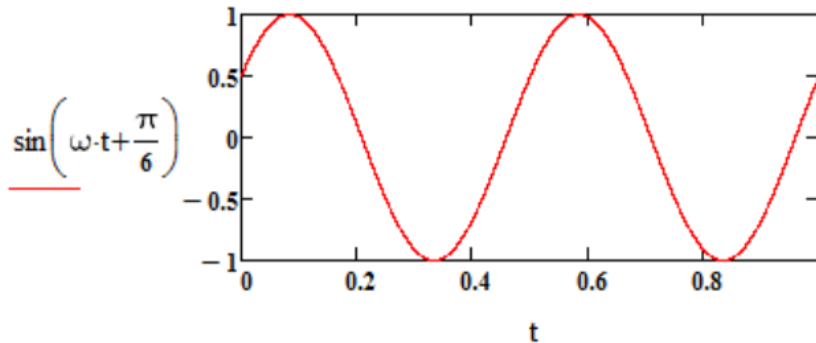


# Example 1

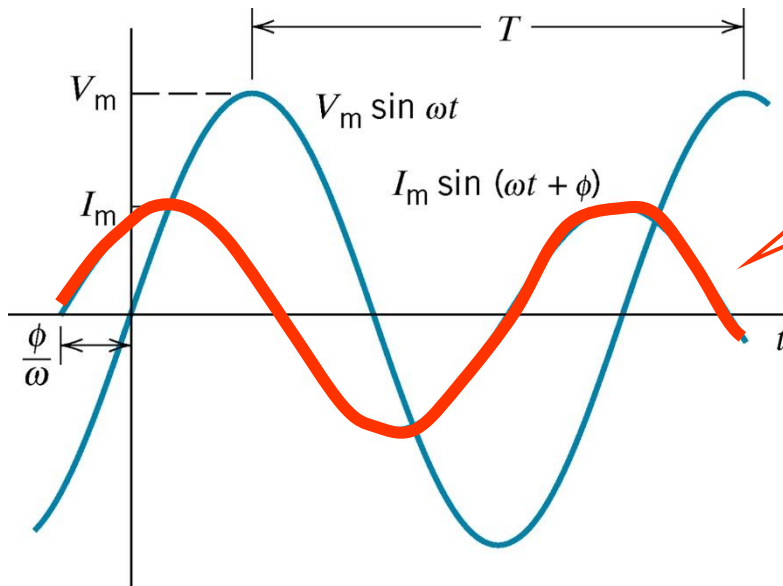
$$v_s = V_m \sin(\omega t + 30^\circ) = V_m \cos(?)$$

$$= V_m \cos(\omega t + 30^\circ - 90^\circ)$$

$$= V_m \cos(\omega t - 60^\circ)$$



# Sinusoidal Sources



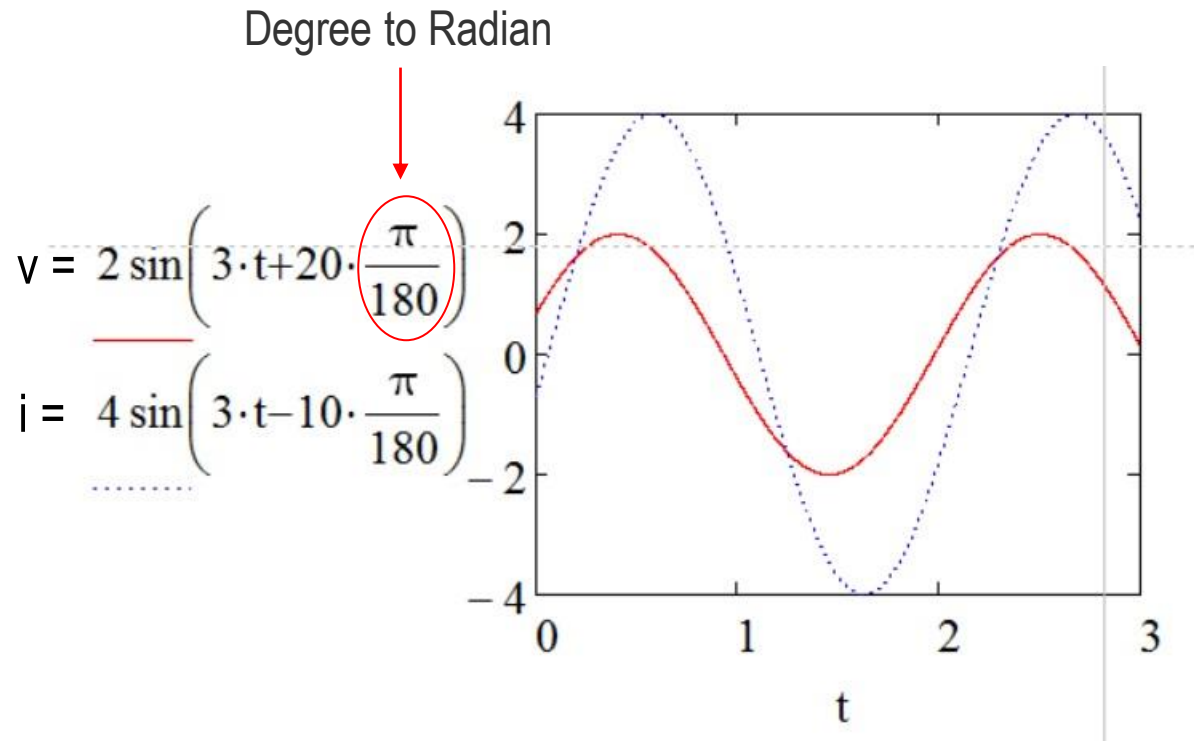
The current reaches its peak before the voltage: Current **leads** the voltage OR Voltage **lags** the current

## Example 2

$$v = 2 \sin(3t + 20^\circ)$$

$$i = 4 \sin(3t - 10^\circ)$$

Voltage,  $v$ , leads (or advances) the current,  $i$ , by  $20 - (-10) = +30^\circ$



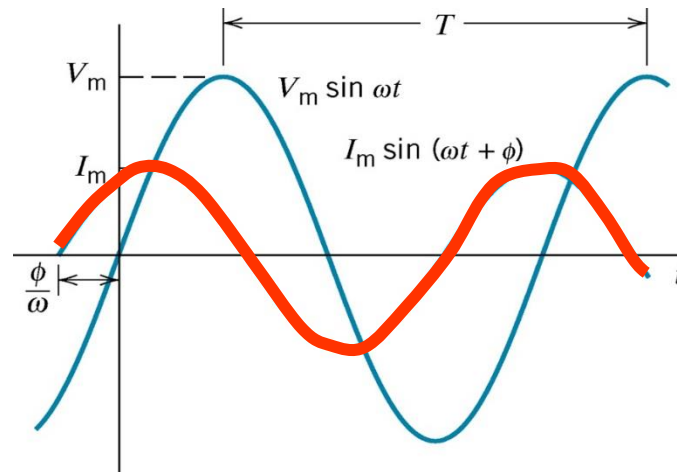
## Example 10.2-1

- Consider the voltages  $v_1 = 10\cos(200t+45^\circ)$  V and  $v_2 = 8\sin(200t+15^\circ)$  A. Determine the time by which  $v_2(t)$  is advanced or delayed with respect to  $v_1(t)$ .

$$\sin(\omega t) = \cos(\omega t - 90^\circ) = \cos\left(\omega t - \frac{\pi}{2}\right)$$

$$\cos(\omega t) = \sin(90^\circ - \omega t) = \sin\left(\frac{\pi}{2} - \omega t\right)$$

$$\sin(\omega t + 180^\circ) = -\sin(\omega t)$$



Current **leads** the voltage OR Voltage **lags** the current

## Example 10.2-1 Solution

$$v_2 = \sin(200t + 15^\circ)$$

$$v_2 = 8 \cos(200t + 15^\circ - 90^\circ) = 8 \cos(200t - 75^\circ) V$$

$$\theta_2 - \theta_1 = -75^\circ - 45^\circ = -120^\circ = -\frac{\pi}{3} \text{ rad}$$

$$\frac{\phi}{\omega} = -\frac{\frac{\pi}{3}}{200} = -5.2 \text{ ms}$$

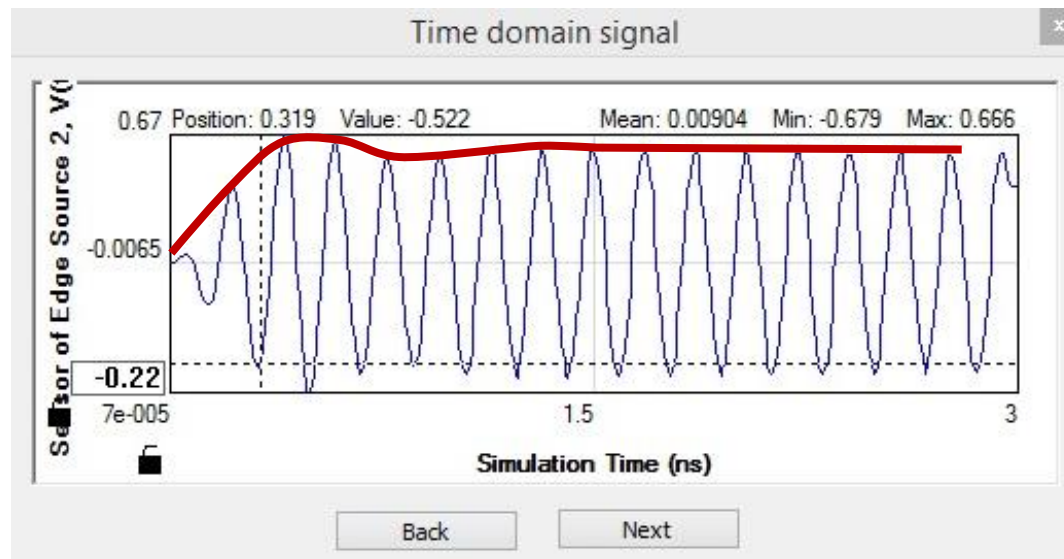
$$\theta_2 - \theta_1 < 0$$

**This indicates a delay (lag)**

# PHASORS

- Phasors may be used when:
  - the circuit is **linear**
  - the **steady-state response** is sought and
  - all independent sources are **sinusoidal** and have the **same frequency**

## Example of a Steady State Response



- Challenge in circuit analysis → sin or cos functions exist
  - Most of the realistic circuits include capacitors and/or inductors
  - The analytical equations will include:

$$C \frac{dv}{dt} \text{ \& \> } L \frac{di}{dt}$$

Can we make the analysis simpler?



# Exponential function

Widely used in physics, chemistry, engineering, mathematical biology, economics, mathematics, etc.

$$\ln e^x = x \quad \text{and} \quad e^{\ln x} = x.$$

$$\ln e = 1.$$

$$\frac{d}{dx} e^x = e^x$$

$$\int e^x dx = e^x + C$$

## Euler's Identity

$$e^{jx} = \cos(x) + j \sin(x)$$

$$e^{jx} = \cos(x) + j \sin(x)$$

$$v_s(t) = V_m \cos(\omega t) = \operatorname{Re}\{V_m e^{j\omega t}\}$$

$$v_s(t) = V_m \cos(\omega t + \phi) = \operatorname{Re}\{V_m e^{j(\omega t + \phi)}\} = \operatorname{Re}\{V_m e^{j\omega t} e^{j\phi}\}$$

$$i(t) = I_m \cos(\omega t) = \text{Re}\{I_m e^{j\omega t}\}$$

$$i(t) = I_m \cos(\omega t + \phi) = \text{Re}\{I_m e^{j(\omega t + \phi)}\} = \text{Re}\{I_m e^{j\omega t} e^{j\phi}\}$$

1. We can drop the  $\text{Re}\{\}$ , since we know that it is the real part
2. We can also drop the part with radian frequency, because we know that we are solving the circuit for *one specific frequency*
3. Then  $i(t)$  information can be represented by the **phasor** of  $i(t)$ ,  $\mathbf{I}$

$$\mathbf{I} = I_m e^{j\phi} = I_m \angle \phi$$

Time domain

$$i(t) = I_m \cos(\omega t + \phi)$$

$$= \operatorname{Re}\{I_m e^{j(\omega t + \phi)}\}$$

$$= \operatorname{Re}\{I_m e^{j\omega t} e^{j\phi}\}$$

Frequency domain

$$I = I_m e^{j\phi} = I_m \angle \phi$$

- Phasor notation:

$$i(t) = I_m \cos(\omega t + \phi)$$

$$\mathbf{I} = I_m e^{j\phi} = I_m \angle \phi$$

- Cosine** function is usually chosen as the **standard** for phasor notation
- Phasor quantities are complex (complex exponential function)
- Although we dropped the  $e^{j\omega t}$ , note that we are performing the calculations in the *frequency domain*, instead of in the *time domain*

- Phasor notation:

$$\mathbf{I} = I_m e^{j\phi} = I_m \angle \phi$$

- This way, we avoid the **complete solution of the circuits** (with energy storage elements) in time domain that requires differential equation solutions

$$C \frac{dv}{dt} \text{ \& \& } L \frac{di}{dt}$$

- Remember that **frequency domain analysis** only provides the **steady-state solution**
- If you need to know the **transient behavior** of the circuit, you **CANNOT** use frequency domain solution (phasors)

# Phasor Summary

**Table 10.5-1 Transformation from the Time Domain to the Frequency Domain**

1. Write the function in the time domain,  $y(t)$ , as a cosine waveform with a phase angle  $\phi$  as

$$y(t) = Y_m \cos(\omega t + \phi)$$

2. Express the cosine waveform as the real part of a complex quantity by using Euler's identity so that

$$y(t) = \operatorname{Re}\{Y_m e^{j(\omega t + \phi)}\}$$

3. Drop the real part notation.

4. Suppress the  $e^{j\omega t}$  while noting the value of  $\omega$  for later use, obtaining the phasor

$$\mathbf{Y} = Y_m e^{j\phi} = Y_m \angle \phi$$

**Table 10.5-2 Transformation from the Frequency Domain to the Time Domain**

1. Write the phasor in exponential form as

$$\mathbf{Y} = Y_m e^{j\beta}$$

2. Reinsert the factor  $e^{j\omega t}$  so that you have

$$Y_m e^{j\beta} e^{j\omega t}$$

3. Reinsert the real part operator  $\operatorname{Re}$  as

$$\operatorname{Re}\{Y_m e^{j\beta} e^{j\omega t}\}$$

4. Use Euler's identity to obtain the time function

$$y(t) = \operatorname{Re}\{Y_m e^{j(\omega t + \beta)}\} = Y_m \cos(\omega t + \beta)$$

## Example 3

- Transform the current expression in time domain  $i(t) = 5\sin(100t+120^\circ)$  to the frequency domain.

$$\sin(\omega t) = \cos(\omega t - 90^\circ) = \cos\left(\omega t - \frac{\pi}{2}\right)$$

$$\cos(\omega t) = \sin(90^\circ - \omega t) = \sin\left(\frac{\pi}{2} - \omega t\right)$$

$$\sin(\omega t + 180^\circ) = -\sin(\omega t)$$



## Example 4

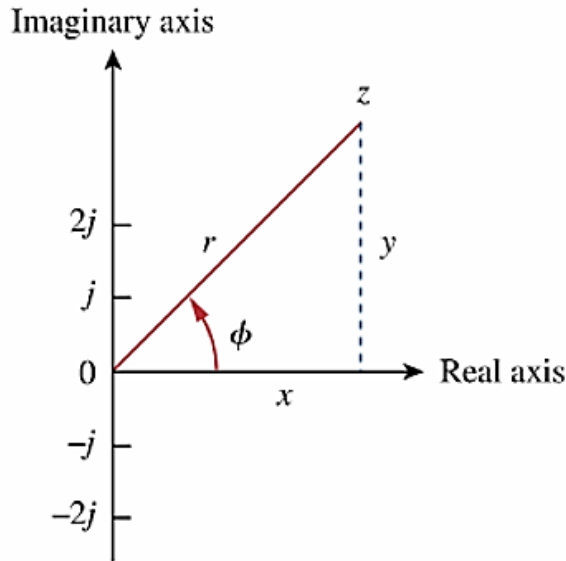
- Transform the voltage expression in frequency domain  $V = 24 \angle 125^\circ$  to time domain.

$$\sin(\omega t) = \cos(\omega t - 90^\circ) = \cos\left(\omega t - \frac{\pi}{2}\right)$$

$$\cos(\omega t) = \sin(90^\circ - \omega t) = \sin\left(\frac{\pi}{2} - \omega t\right)$$

$$\sin(\omega t + 180^\circ) = -\sin(\omega t)$$

# Review: Complex Numbers



**Figure 9.6**

Representation of a complex number  $z = x + jy = r \angle \phi$ .

- $z = x + jy$  (rectangular form)    + and –
- $z = r \angle \phi$  (polar form)     $\times$  and  $\div$
- $z = re^{j\phi}$  (exponential form)     $\times$  and  $\div$

Useful for the calculator

Rectangular to polar

$$r = \sqrt{x^2 + y^2} \quad \phi = \tan^{-1} \frac{y}{x} \text{ and}$$

$$\phi = 180^\circ - \tan^{-1} \frac{y}{x}, \text{ when } x < 0$$

$$z = \sqrt{x^2 + y^2} \angle \tan^{-1} \frac{y}{x}$$

Polar to rectangular

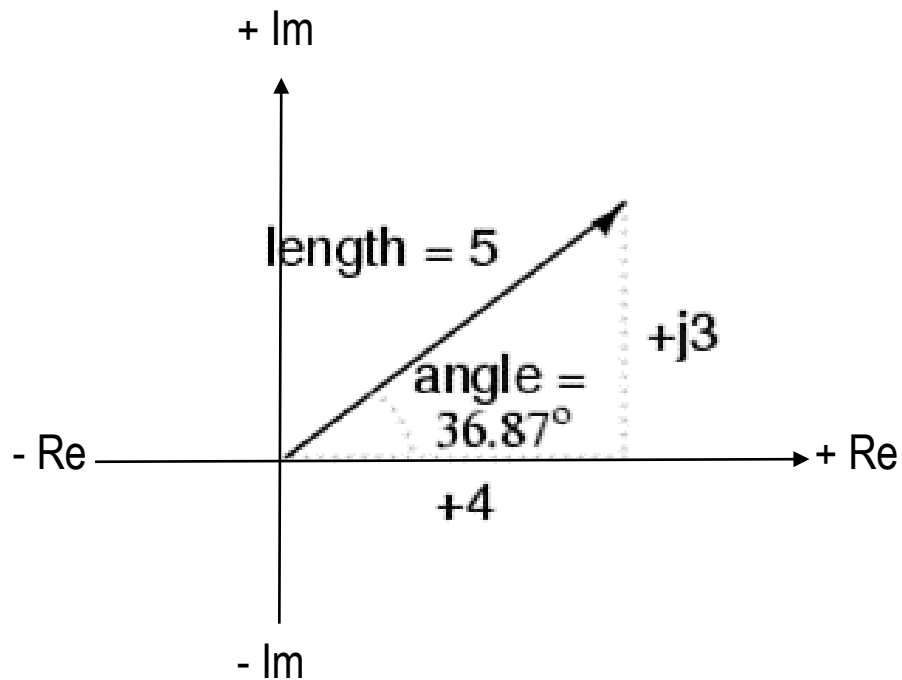
$$x = r \cos \phi$$

$$y = r \sin \phi$$

$$z = r(\cos \phi + j \sin \phi)$$

Inverse tan.

# Review: Complex Numbers – Example



Polar to rectangular

$5 \angle 36.87^\circ$  **(polar form)**

$(5)(\cos 36.87^\circ) = 4$  (real component)

$(5)(\sin 36.87^\circ) = 3$  (imaginary component)

$4 + j3$  **(rectangular form)**

Convert from rectangular to polar  
and vice versa

$$5 \angle 36.87^\circ \longleftrightarrow 4 + j3$$

Rectangular to polar

$4 + j3$  **(rectangular form)**

$$c = \sqrt{a^2 + b^2} \quad (\text{pythagorean theorem})$$

$$\text{polar magnitude} = \sqrt{4^2 + 3^2}$$

$$\text{polar magnitude} = 5$$

$$\text{polar angle} = \arctan \frac{3}{4}$$

$$\text{polar angle} = 36.87^\circ$$

$5 \angle 36.87^\circ$  **(polar form)**

# Review: Complex Numbers

Given:

$$Z_1 = x_1 + jy_1 = r_1 \angle \phi_1$$

$$Z_2 = x_2 + jy_2 = r_2 \angle \phi_2$$

Addition and Subtraction of Complex Number (rectangular)

$$Z_1 + Z_2 = (x_1 + x_2) + j(y_1 + y_2)$$

$$Z_1 - Z_2 = (x_1 - x_2) + j(y_1 - y_2)$$

Multiplication/Division of Complex Number (polar)

$$Z_1 \cdot Z_2 = r_1 \cdot r_2 \angle (\phi_1 + \phi_2)$$

$$r_1 e^{j\phi_1} \cdot r_2 e^{j\phi_2}$$

$$\frac{Z_1}{Z_2} = \frac{r_1}{r_2} \angle (\phi_1 - \phi_2)$$

$$= r_1 \cdot r_2 e^{j(\phi_1 + \phi_2)}$$

## Example 10.3-2

Consider the phasors

$$\mathbf{V}_1 = 4.25 \angle 115^\circ \text{ and } \mathbf{V}_2 = -4 + j3$$

Convert  $\mathbf{V}_1$  to rectangular form and  $\mathbf{V}_2$  to polar form.

Polar to rectangular

$$x = r \cos \phi$$

$$y = r \sin \phi$$

$$z = r(\cos \phi + j \sin \phi)$$

Rectangular to polar

$$r = \sqrt{x^2 + y^2} \quad \phi = \tan^{-1} \frac{y}{x}$$

$$z = \sqrt{x^2 + y^2} \angle \tan^{-1} \frac{y}{x}$$

## Example 10.3-3

Consider the phasors

$$\mathbf{V}_1 = -1.796 + j3.852 = 4.25 \angle 115^\circ \text{ and } \mathbf{V}_2 = -4 + j3 = 5 \angle 143^\circ$$

Determine  $\mathbf{V}_1 + \mathbf{V}_2$ ,  $\mathbf{V}_1 \cdot \mathbf{V}_2$  and  $\frac{\mathbf{V}_1}{\mathbf{V}_2}$ .

Addition and Subtraction of Complex Number (rectangular)

$$Z_1 + Z_2 = (x_1 + x_2) + j(y_1 + y_2)$$

$$Z_1 - Z_2 = (x_1 - x_2) + j(y_1 - y_2)$$

Multiplication/Division of Complex Number (polar)

$$Z_1 \cdot Z_2 = r_1 \cdot r_2 \angle (\phi_1 + \phi_2)$$

$$\frac{Z_1}{Z_2} = \frac{r_1}{r_2} \angle (\phi_1 - \phi_2)$$

# Example 10.3-4 – Kirchhoff's Law for AC Circuits

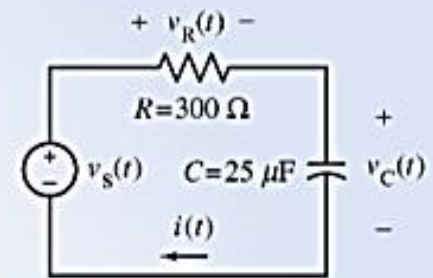
The input to the circuit shown in Figure 10.3-3 is the voltage source voltage,

$$v_s(t) = 25 \cos(100t + 15^\circ) \text{ V}$$

The output is the voltage across the capacitor,

$$v_C(t) = 20 \cos(100t - 22^\circ) \text{ V}$$

Determine the resistor voltage  $v_R(t)$ .



**FIGURE 10.3-3** The circuit in Example 10.3-4

Addition and Subtraction of Complex Number (rectangular)

$$Z_1 + Z_2 = (x_1 + x_2) + j(y_1 + y_2)$$

$$Z_1 - Z_2 = (x_1 - x_2) + j(y_1 - y_2)$$

Multiplication/Division of Complex Number (polar)

$$Z_1 \cdot Z_2 = r_1 \cdot r_2 \angle(\phi_1 + \phi_2)$$

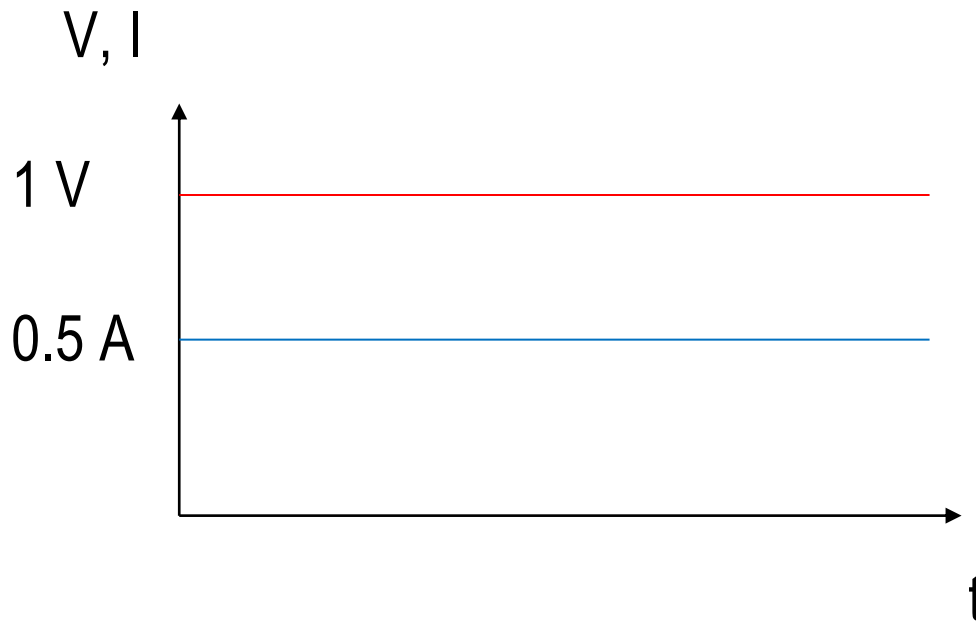
$$\frac{Z_1}{Z_2} = \frac{r_1}{r_2} \angle(\phi_1 - \phi_2)$$

# IMPEDANCE

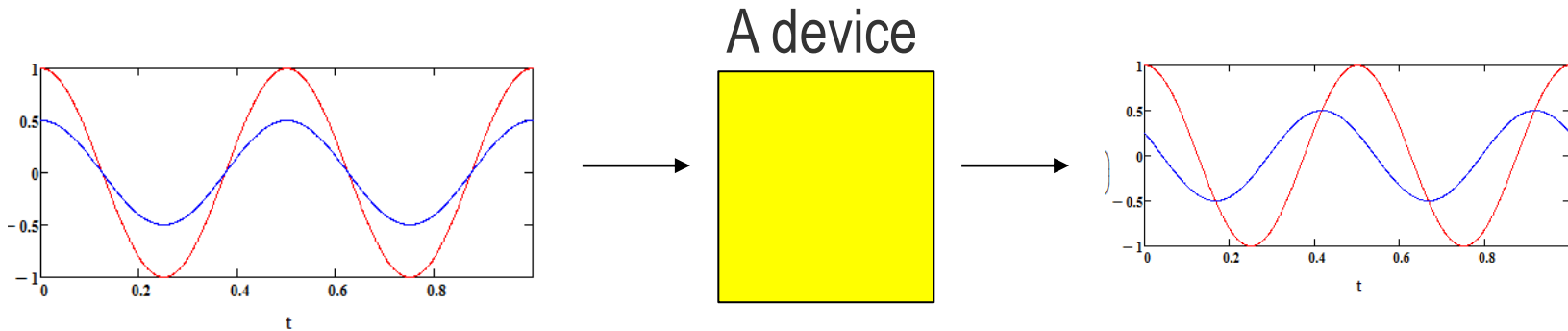


$$\text{if } \begin{cases} V = 1 \text{ V} \\ I = 0.5 \text{ A} \end{cases}$$

$$R = \frac{V}{I} = \frac{1}{0.5} = 2 \Omega$$



# Concept of Impedance in AC Circuits



**Does “R” describe this situation?**

Magnitude can be calculated by R

How about the phase difference?

The phase information is missing!

# Impedance

Freq.

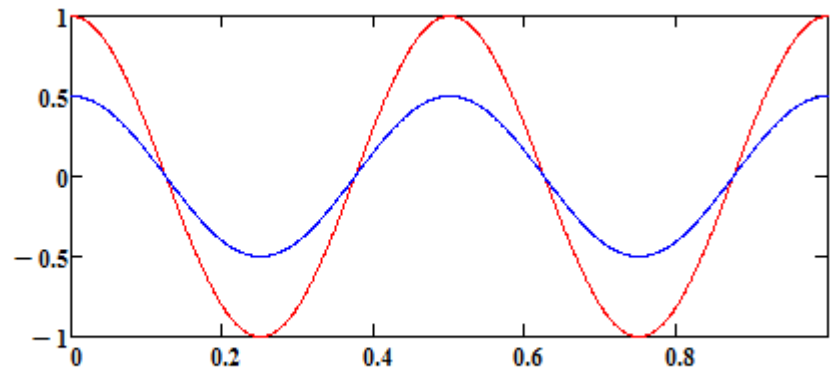
$$V = 1\angle 0^\circ$$

$$I = 0.5\angle 0^\circ$$

Time

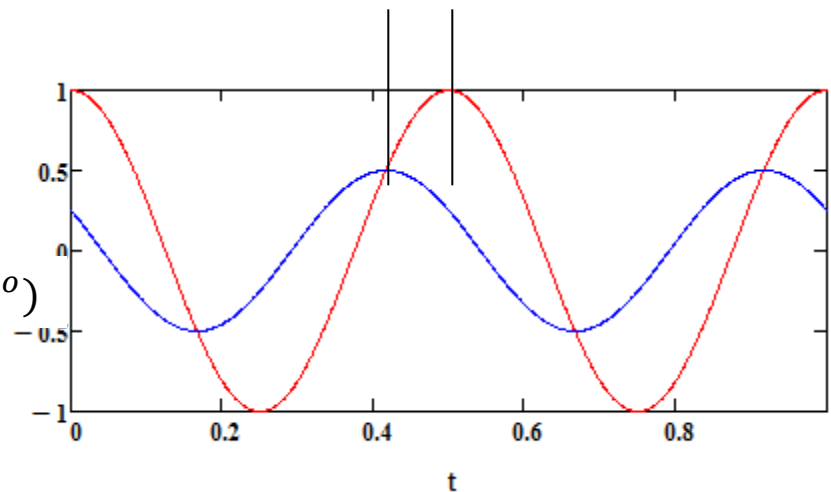
$$v(t) = 1 \cdot \cos(\omega t)$$

$$i(t) = 0.5 \cdot \cos(\omega t)$$



$$R = \frac{V}{I} = \frac{1\angle 0^\circ}{0.5\angle 0^\circ} = 2\angle 0^\circ \Omega \quad (2 \Omega \text{ resistor})$$

Phase



$$V = 1\angle 0^\circ$$

$$v(t) = 1 \cdot \cos(\omega t)$$

$$I = 0.5\angle 60^\circ$$

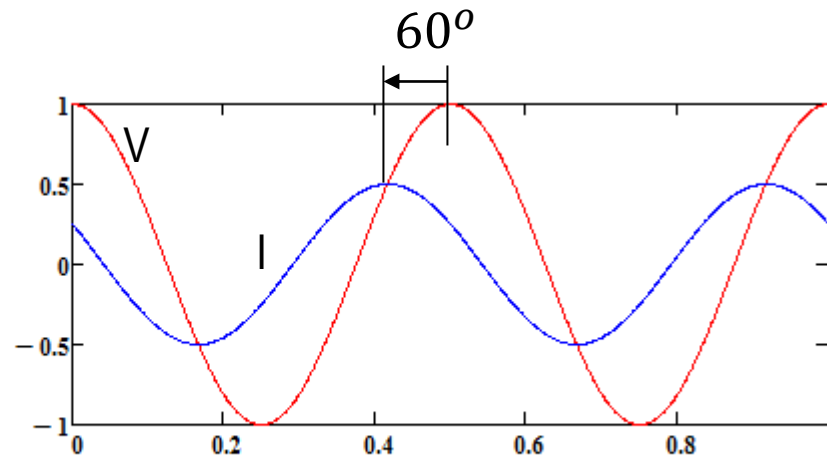
$$i(t) = 0.5 \cdot \cos(\omega t + 60^\circ)$$

$$\cancel{Z} R = \frac{V}{I} = \frac{1\angle 0^\circ}{0.5\angle 60^\circ} = 2\angle (-60^\circ) \Omega$$

*I leads V*

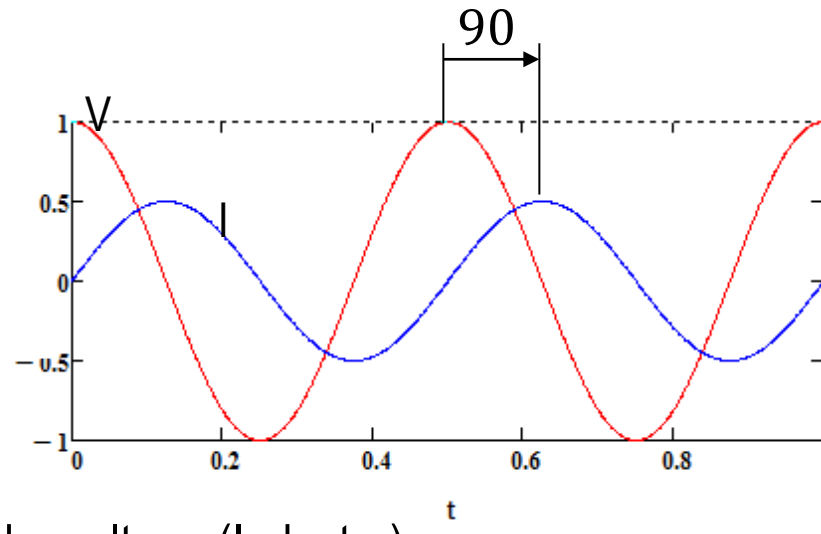
# Example of Impedance

$$Z = \frac{V}{I} = 2 \angle (-60^\circ) \Omega$$



Current *leads* the voltage OR Voltage *lags* the current (Capacitor)

$$Z = \frac{V}{I} = 2 \angle (90^\circ) \Omega$$



Voltage *leads* the current OR Current *lags* the voltage (Inductor)

# Impedance

- The ratio of the phasor voltage to the phasor current is defined as **impedance** and denoted by **Z**
- Impedance in AC circuits has a similar role to the role of resistance in DC circuits

$$V = V_m e^{j\phi}, I = I_m e^{j\beta}$$

$$\Rightarrow Z = \frac{V}{I} = \frac{V_m e^{j\phi}}{I_m e^{j\beta}} = \frac{V_m}{I_m} e^{j(\phi - \beta)} = \frac{V_m}{I_m} \angle (\phi - \beta)$$

Magnitude, **|Z|**

Phase angle

$$Z = |Z| \angle \theta \rightarrow \text{polar form}$$

$$= |Z| e^{j\theta} \rightarrow \text{exponential form}$$

$$= R + jX \rightarrow \text{rectangular form}$$

$$|Z| = \sqrt{R^2 + X^2}, \theta = \tan^{-1} \left( \frac{X}{R} \right)$$

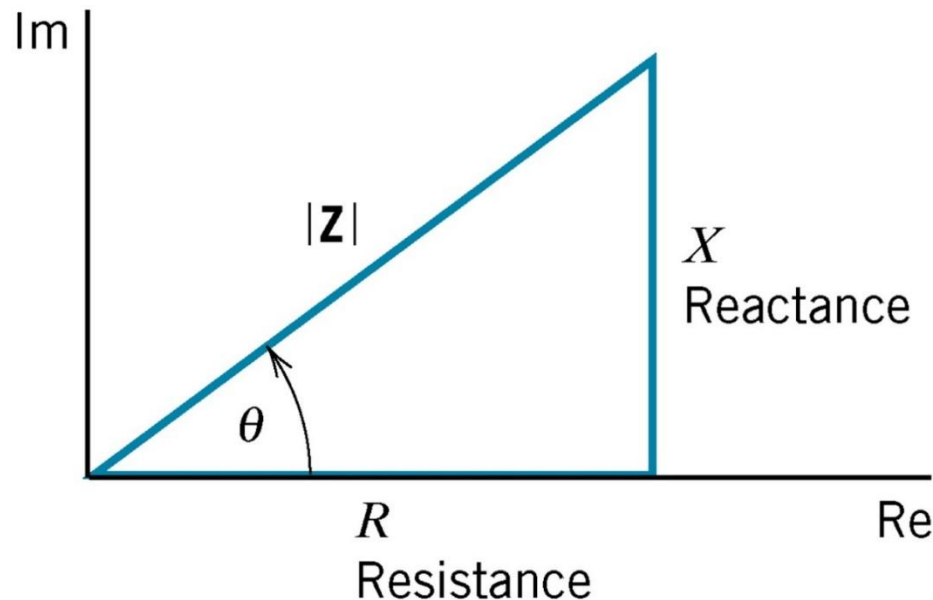
Real part:  
**Resistance** [ $\Omega$ ]

Imaginary part: **Reactance** [ $\Omega$ ]  
+  $jX$  (Positive)  $\rightarrow$  Inductor  
-  $jX$  (Negative)  $\rightarrow$  Capacitor

# Impedance

$$Z = |Z| \angle \theta = |Z| e^{j\theta} = R + jX$$

$$|Z| = \sqrt{R^2 + X^2}, \quad \theta = \tan^{-1}\left(\frac{X}{R}\right)$$



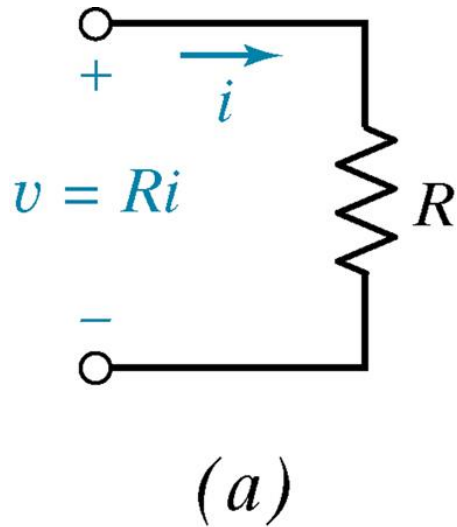
$$Y = \frac{1}{Z} = \frac{1}{|Z| \angle \theta} = \frac{1}{|Z|} \angle -\theta$$

$$Y = \frac{1}{R + jX} = \frac{R - jX}{R^2 + X^2} = \frac{R}{R^2 + X^2} - j \frac{X}{R^2 + X^2}$$
$$= \textcircled{G} + j \textcircled{B}$$

Real part: **Conductance**  
[Siemens]

Imaginary part: **Susceptance**  
[Siemens]  
+  $jB$  (Positive) → Capacitor  
-  $jB$  (Negative) → Inductor

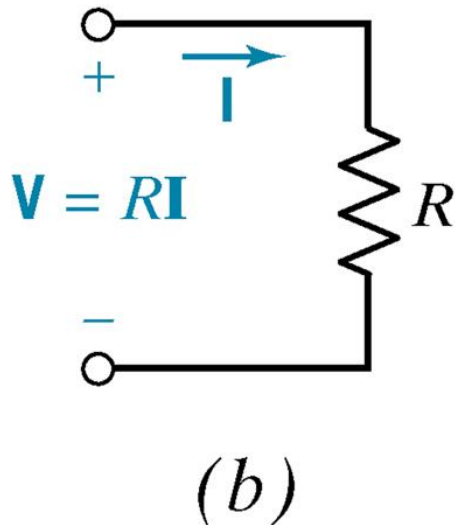


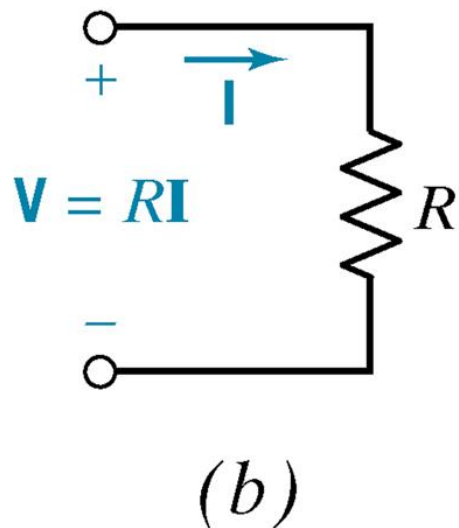
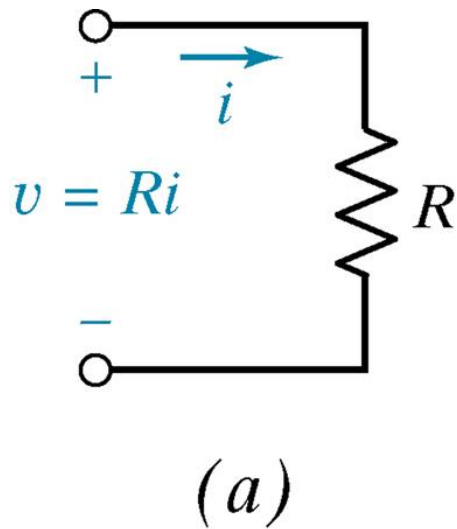


$$v(t) = Ri(t) \quad \underline{\text{Ohm's law}}$$

$$v = V_m \cos(\omega t + \phi) = \text{Re}\{V_m e^{j(\omega t + \phi)}\}$$

$$i = I_m \cos(\omega t + \beta) = \text{Re}\{I_m e^{j(\omega t + \beta)}\}$$





$$v(t) = Ri(t) \quad \text{Ohm's law}$$

$$v = V_m \cos(\omega t + \phi) = \text{Re}\{V_m e^{j(\omega t + \phi)}\}$$

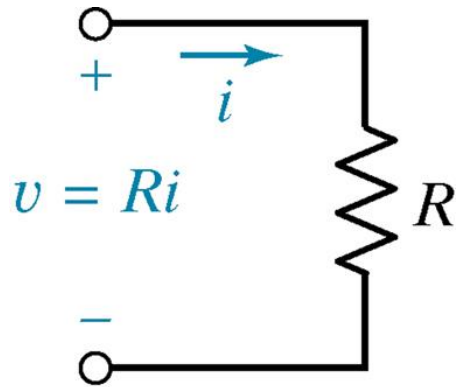
$$i = I_m \cos(\omega t + \beta) = \text{Re}\{I_m e^{j(\omega t + \beta)}\}$$

$$V_m e^{j(\omega t + \phi)} = R \times I_m e^{j(\omega t + \beta)}$$

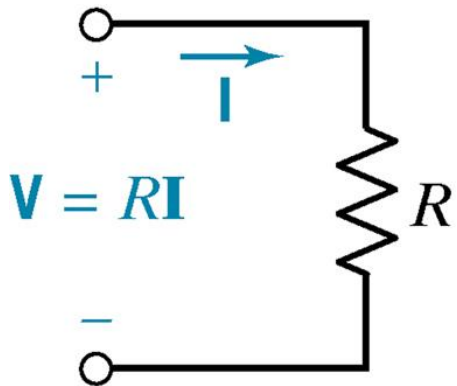
$$V_m e^{j\omega t} e^{j\phi} = R \times I_m e^{j\omega t} e^{j\beta}$$

$$V_m e^{j\phi} = R \times I_m e^{j\beta} \Rightarrow \phi = \beta$$

$$V = R \times I$$



(a)

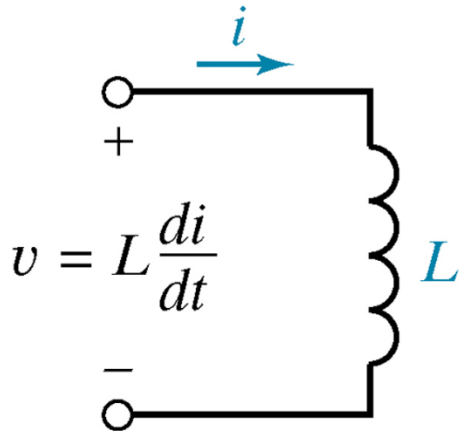


(b)

$$v(t) = 10 \cos(10t); i(t) = ?$$

$$V = R \times I \Rightarrow I = \frac{V}{R} = \frac{10 \angle 0^\circ}{R}$$

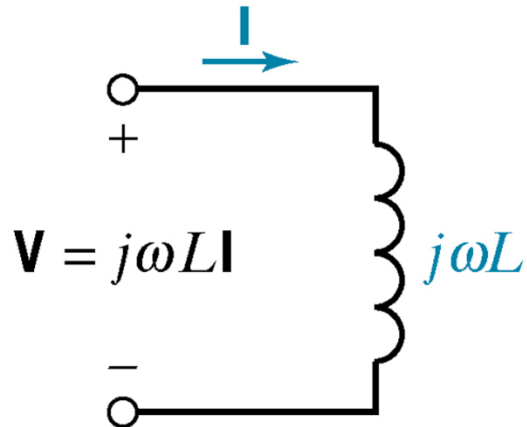
$$\Rightarrow i(t) = \frac{10 \cos(10t)}{R}$$

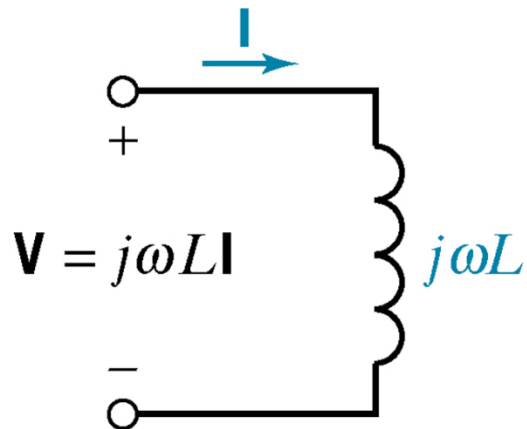
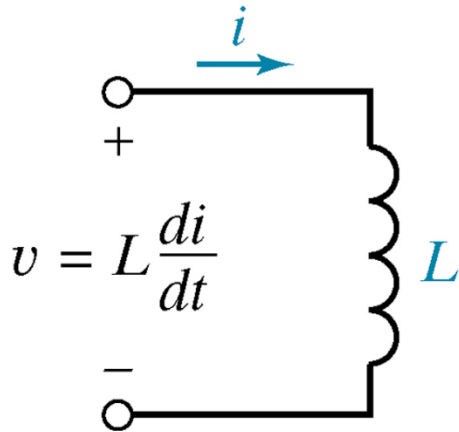


$$v(t) = L \frac{di(t)}{dt}$$

$$v = V_m \cos(\omega t + \phi) = \operatorname{Re}\{V_m e^{j(\omega t + \phi)}\}$$

$$i = I_m \cos(\omega t + \beta) = \operatorname{Re}\{I_m e^{j(\omega t + \beta)}\}$$





$$v(t) = L \frac{di(t)}{dt}$$

$$v = V_m \cos(\omega t + \phi) = \text{Re}\{V_m e^{j(\omega t + \phi)}\}$$

$$i = I_m \cos(\omega t + \beta) = \text{Re}\{I_m e^{j(\omega t + \beta)}\}$$

$$V_m e^{j(\omega t + \phi)} = L \times \frac{dI_m e^{j(\omega t + \beta)}}{dt}$$

$$V_m e^{j\omega t} e^{j\phi} = L \times \frac{dI_m e^{j(\omega t + \beta)}}{dt}$$

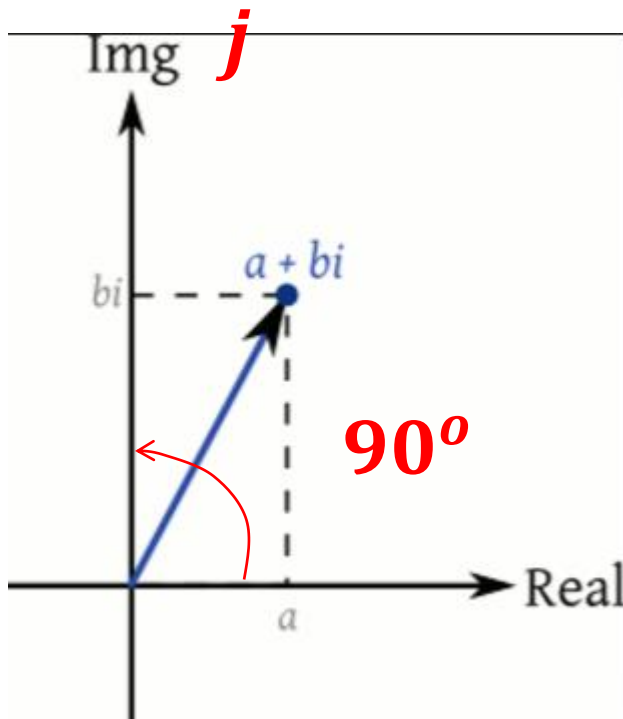
$$V_m e^{j\omega t} e^{j\phi} = L \times j\omega \times I_m e^{j(\omega t + \beta)}$$

$$V_m e^{j\omega t} e^{j\phi} = L \times j\omega \times I_m e^{j\omega t} e^{j\beta}$$

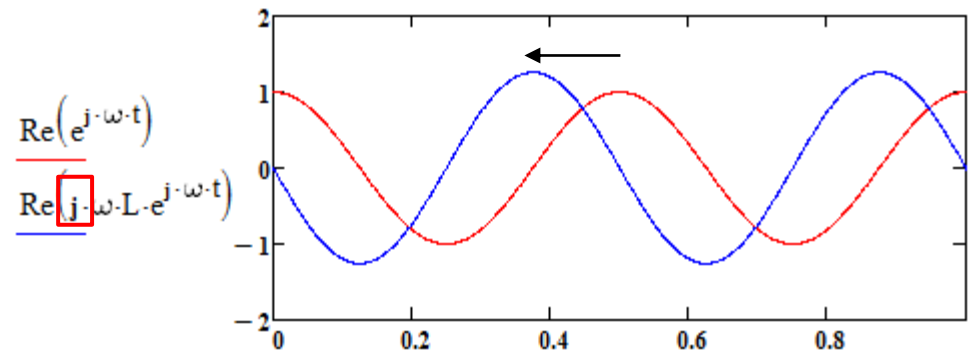
$$V_m e^{j\phi} = L \times j\omega \times I_m e^{j\beta}$$

$$V = j\omega L \times I$$

# Complex Number $j$

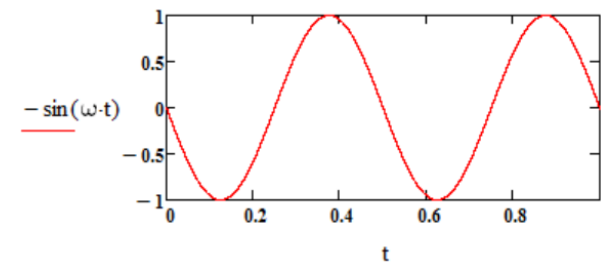
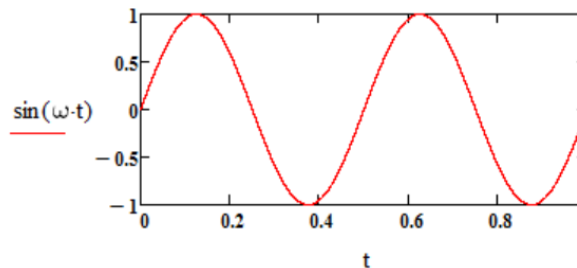


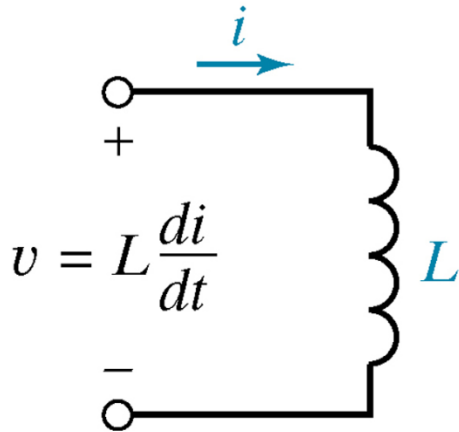
$j$  causes  $90^\circ$  phase shift: **leading**



$$e^{j90^\circ} = j$$

$$1 \cdot e^{j180^\circ} = -1$$

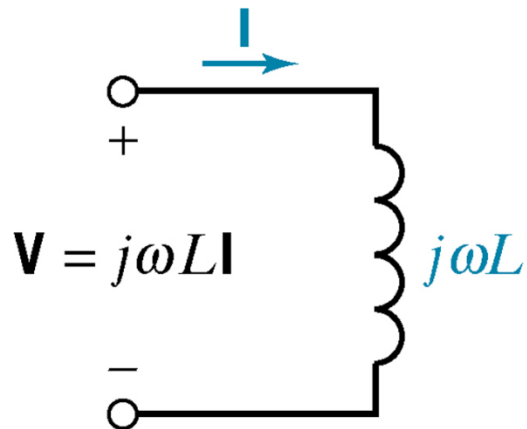




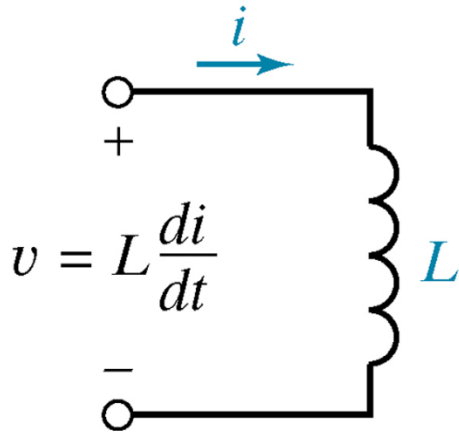
$$V = j\omega L \times I$$

$$j = e^{j90^\circ} \Rightarrow V_m e^{j\phi} = L \times \omega \times I_m e^{j90^\circ} e^{j\beta}$$

Inductor voltage leads the inductor current by exactly  $90^\circ$



## Example 5

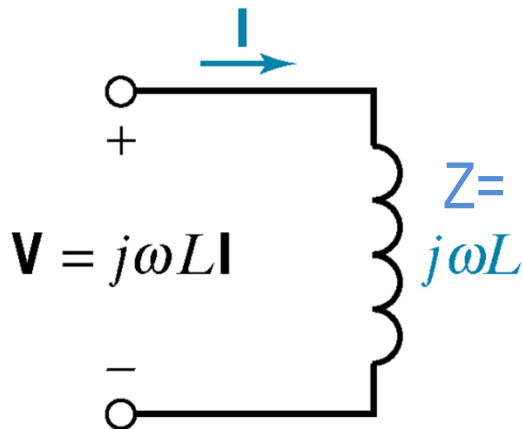


$$L = 2 \text{ H}, \omega = 100 \text{ rad/s}, v(t) = 10 \cos(\omega t + 50^\circ)$$

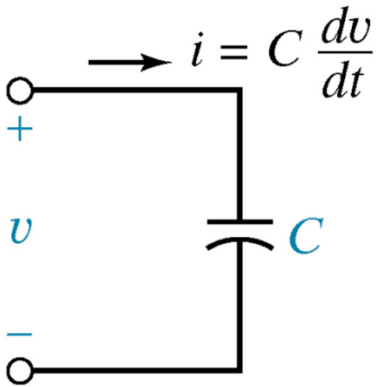
$$V = j\omega L \times I \Rightarrow I = \frac{V}{j\omega L} = \frac{10 \angle 50^\circ}{j \times 100 \times 2} = \frac{10 \angle 50^\circ}{j200}$$

$$= \frac{10 \angle 50^\circ}{200 \angle 90^\circ} = 0.05 \angle -40^\circ$$

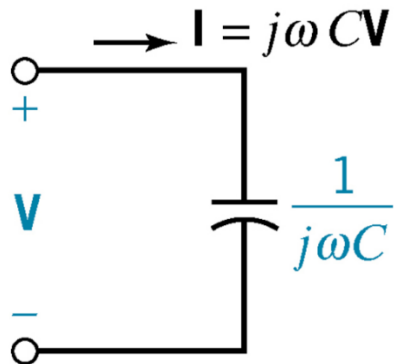
$$\Rightarrow i(t) = 0.05 \cos(100t - 40^\circ)$$







(a)



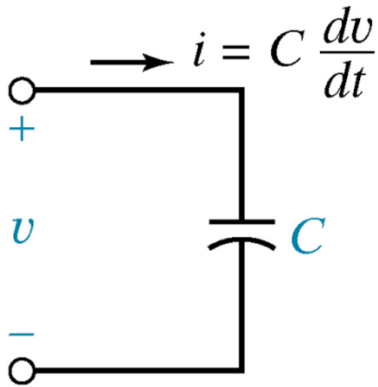
(b)

$$i(t) = C \frac{dv(t)}{dt}$$

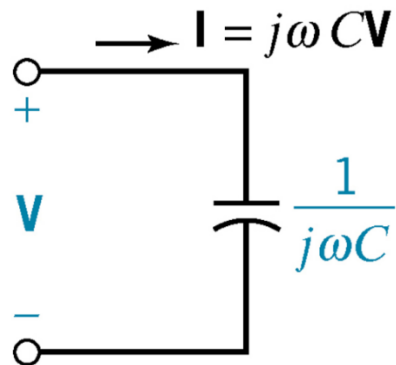
$$v = V_m \cos(\omega t + \phi) = \text{Re}\{V_m e^{j(\omega t + \phi)}\}$$

$$i = I_m \cos(\omega t + \beta) = \text{Re}\{I_m e^{j(\omega t + \beta)}\}$$

# Phasor of C



(a)



(b)

$$i(t) = C \frac{dv(t)}{dt}$$

$$v = V_m \cos(\omega t + \phi) = \text{Re}\{V_m e^{j(\omega t + \phi)}\}$$

$$i = I_m \cos(\omega t + \beta) = \text{Re}\{I_m e^{j(\omega t + \beta)}\}$$

$$I_m e^{j(\omega t + \beta)} = C \times \frac{dV_m e^{j(\omega t + \phi)}}{dt}$$

$$I_m e^{j\omega t} e^{j\beta} = C \times \frac{dV_m e^{j(\omega t + \phi)}}{dt}$$

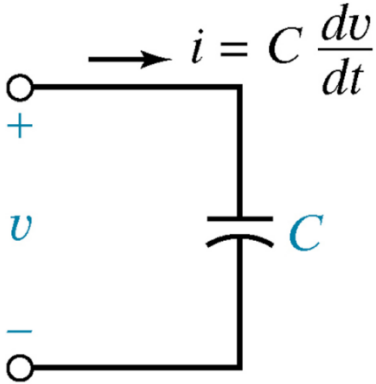
$$I_m e^{j\omega t} e^{j\beta} = C \times j\omega \times V_m e^{j(\omega t + \phi)}$$

$$I_m e^{j\omega t} e^{j\beta} = C \times j\omega \times V_m e^{j\omega t} e^{j\phi}$$

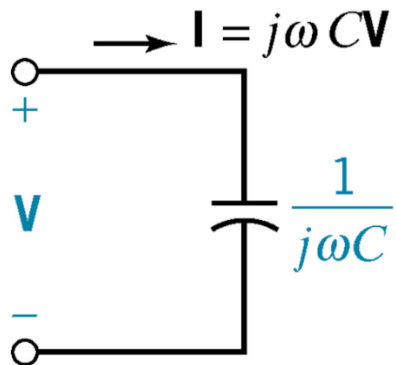
$$I_m e^{j\beta} = C \times j\omega \times V_m e^{j\phi}$$

$$I = j\omega C \times V$$

$$V = \frac{1}{j\omega C} \cdot I$$



(a)



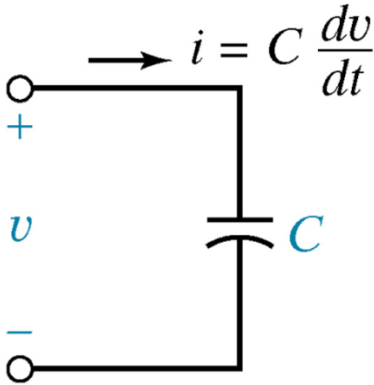
(b)

$$I = j\omega C \times V$$

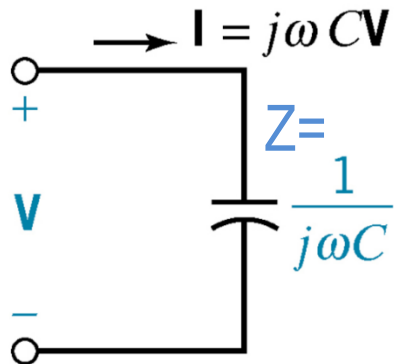
$$j = e^{j90^\circ} \Rightarrow I_m e^{j\beta} = L \times \omega \times V_m e^{j\phi} e^{j90^\circ}$$

Capacitor current leads the capacitor voltage by exactly  $90^\circ$

## Example 6



(a)



(b)

$$C = 1 \text{ mF}, v(t) = 100 \cos(1000t) \text{ V}$$

$$I = j\omega C \times V = j \times 1000 \times 1\text{m} \times 100 \angle 0^\circ$$

$$= j100 \angle 0^\circ = 100 \angle 90^\circ$$

$$\Rightarrow i(t) = 100 \cos(1000t + 90^\circ)$$

# Impedance Summary

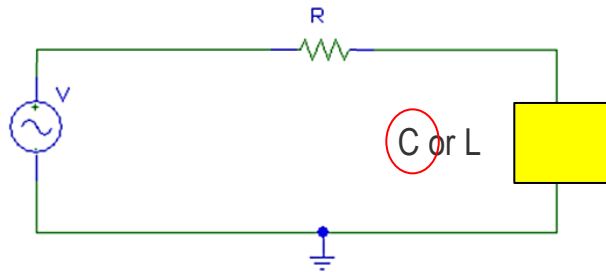
For a resistor  $R$ ;  $Z = R$

For an inductor  $L$ ;  $Z = j\omega L$

For a capacitor  $C$ ;  $Z = \frac{1}{j\omega C} = \frac{-j}{\omega C}$

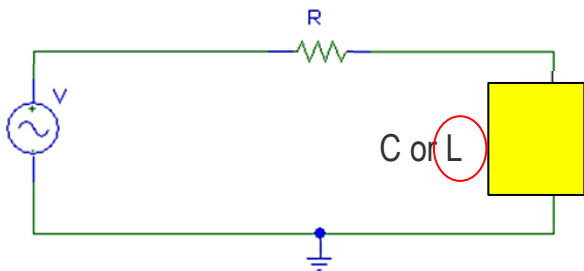
$$1 = 1\angle 0^\circ, j = 1\angle 90^\circ, -1 = 1\angle \pm 180^\circ, \\ -j = 1\angle -90^\circ \text{ or } 1\angle 270^\circ$$

# Example of Impedance



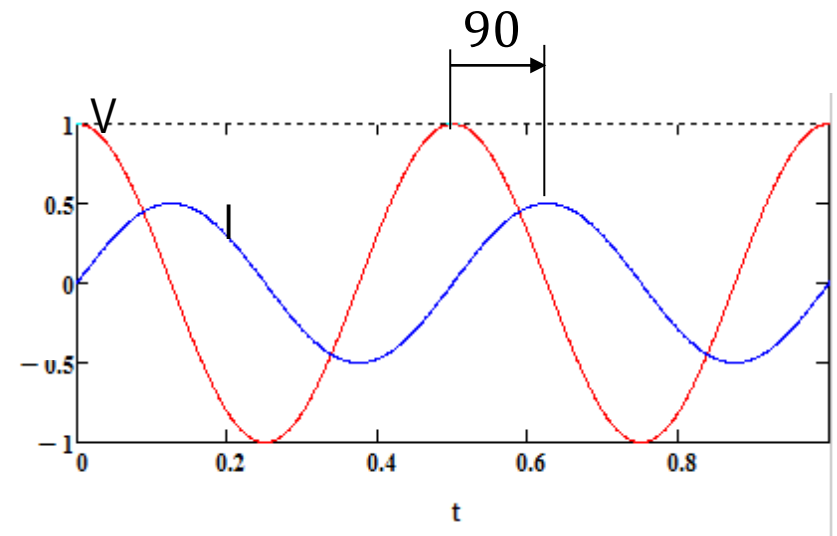
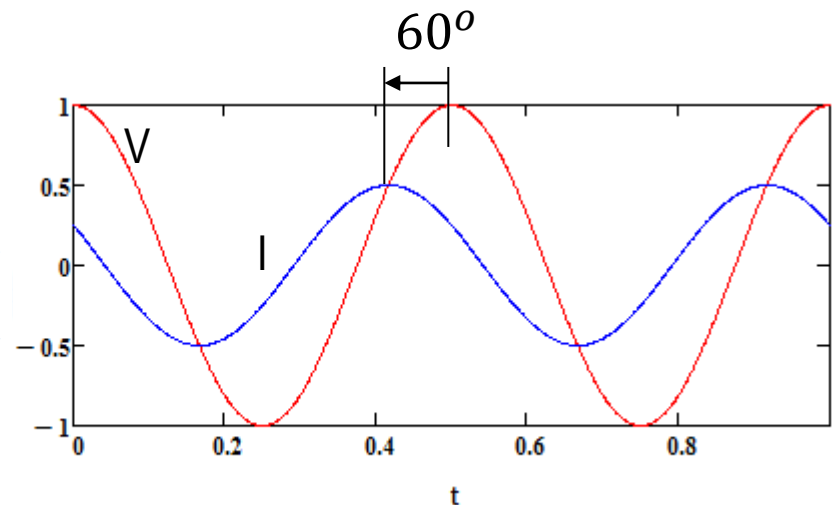
$$Z = \frac{V}{I} = 2 \angle (-60^\circ) \, \Omega = 1 - 1.732j \, \Omega$$

Polar form                  Rectangular form



$$Z = \frac{V}{I} = 2 \angle (90^\circ) \, \Omega = 0 + 2j \, \Omega$$

Polar form                  Rectangular form



- DC circuit analysis techniques compute only magnitudes of  $V$  and  $I$   
(Parallel/Series, Volt/Current, division, KVL, KCL, Mesh, Nodal, Thévenin/Norton Equivalent, Dependent sources, OpAmps, etc.)
- AC analysis is a lot more complex than DC analysis
  - In addition to magnitudes, phase should also be considered
  - Essentially, sine and cosine function(s) are to be used, but then
    - Analytical computations are too complicated to handle by human (or computers)
  - It is necessary to adapt simpler ways to compute AC circuits
  - It is also desired to utilize the known DC analysis techniques (listed above) for AC circuits

## Solution for AC circuit analysis (Advantages of phasors)

- Phasor is adapted to remove cosine functions from analytical equations
- By using phasors, the same DC circuit analysis techniques (that we have studied so far) can be used to analyze AC circuits

## Disadvantages of Phasor

- Phasors are complex numbers (still better than having COS and SIN functions)
- Students should be familiar with the complex mathematics (still better than using COS and SIN functions)
- Steady-state only (transient response cannot be obtained by phasor)

COS and SIN: time domain

Phasors: frequency domain

Element	Impedance	Admittance
$R$	$Z = R$	$Y = \frac{1}{R}$
$L$	$Z = j\omega L$	$Y = \frac{1}{j\omega L}$
$C$	$Z = \frac{1}{j\omega C}$	$Y = j\omega C$



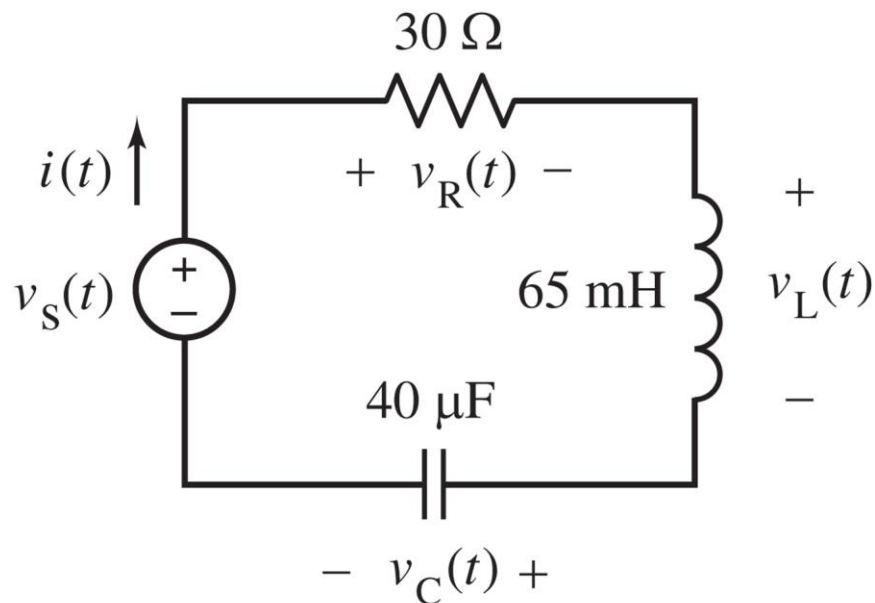
- **Kirchhoff's Laws**
  - Hold true in the frequency domain
  - The sum of sinusoidal currents in/out of a node equals zero
    - $I_1 + I_2 + I_3 + \dots + I_N = 0$
  - The sum of sinusoidal voltages around a loop equals zero
    - $V_1 + V_2 + \dots + V_N = 0$
- **Nodal Analysis** is valid in the frequency domain
  - First, transform a time-domain circuit to the frequency domain
  - Use nodal analysis, as you have done before
  - Finally, transform the answer to the time domain
- **Mesh Analysis** is also valid in the frequency domain
  - Follow the same steps as for Nodal analysis

## Example 10.4-1

The input to the AC circuit shown below is the source voltage

$$v_s(t) = 12 \cos(1000t + 15^\circ) \text{ V}$$

Determine (a) the impedances of the capacitor, inductor, and resistance and (b) the current  $i(t)$



Polar to rectangular

$$x = r \cos \phi$$

$$y = r \sin \phi$$

$$z = r(\cos \phi + j \sin \phi)$$

Rectangular to polar

$$r = \sqrt{x^2 + y^2} \quad \phi = \tan^{-1} \frac{y}{x}$$

$$z = \sqrt{x^2 + y^2} \angle \tan^{-1} \frac{y}{x}$$

## Addition and Subtraction of Complex Number (rectangular)

$$Z_1 + Z_2 = (x_1 + x_2) + j(y_1 + y_2)$$

$$Z_1 - Z_2 = (x_1 - x_2) + j(y_1 - y_2)$$

## Multiplication/Division of Complex Number (polar)

$$Z_1 \cdot Z_2 = r_1 \cdot r_2 \angle (\phi_1 + \phi_2)$$

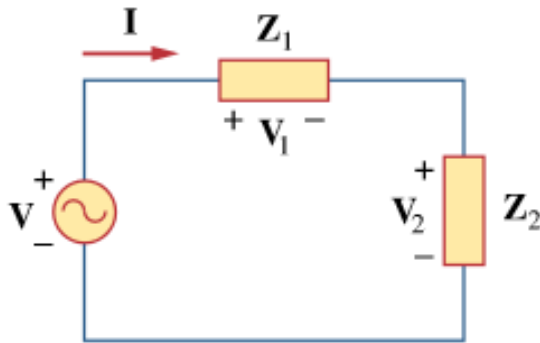
$$\frac{Z_1}{Z_2} = \frac{r_1}{r_2} \angle (\phi_1 - \phi_2)$$

Element	Impedance	Admittance
$R$	$Z = R$	$Y = \frac{1}{R}$
$L$	$Z = j\omega L$	$Y = \frac{1}{j\omega L}$
$C$	$Z = \frac{1}{j\omega C}$	$Y = j\omega C$

# Impedance Combination

## Series elements

$$Z_{eq} = Z_1 + Z_2 + \dots + Z_N$$



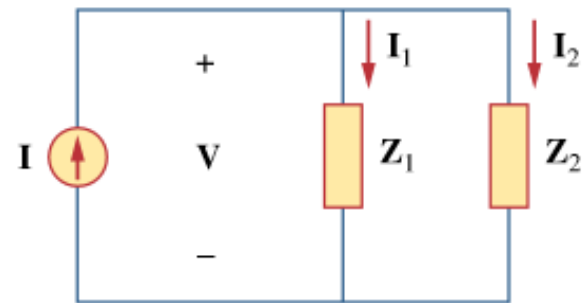
## Voltage Division

$$V_1 = V \frac{Z_1}{(Z_1 + Z_2)}$$

$$V_2 = V \frac{Z_2}{(Z_1 + Z_2)}$$

## Parallel elements

$$\frac{1}{Z_{eq}} = \frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_N}$$

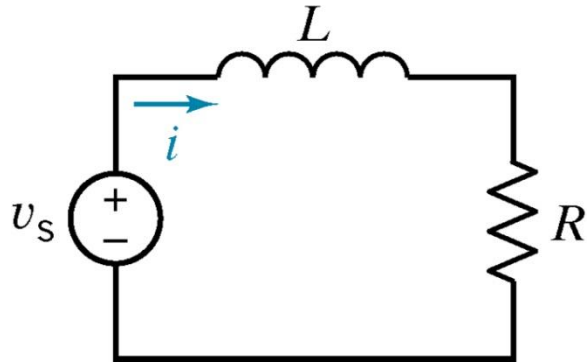


## Current Division

$$I_1 = I \frac{Z_2}{Z_1 + Z_2}$$

$$I_2 = I \frac{Z_1}{Z_1 + Z_2}$$

# Example 7



$$\omega = 100 \text{ rad/s}, R = 200 \Omega, L = 2 \text{ H}, v(t) = V_m \cos \omega t \text{ V}$$

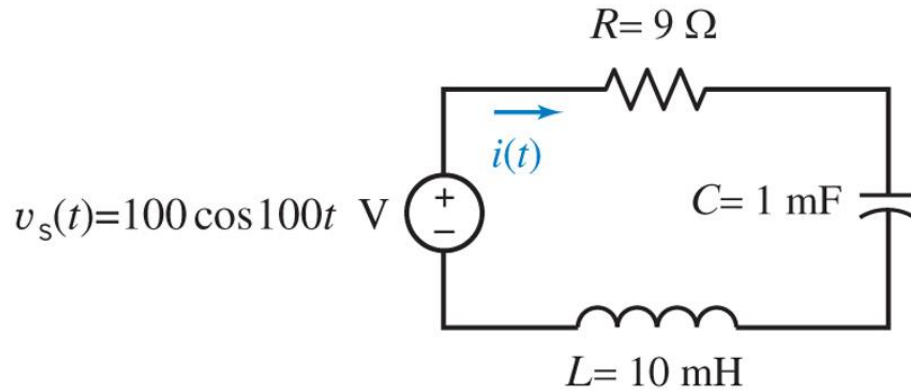
$$i = ?$$

$$I = \frac{V}{j\omega L + R} = \frac{V_m \angle 0^\circ}{j \times 100 \times 2 + 200} = \frac{V_m \angle 0^\circ}{j \times 200 + 200} = \frac{V_m \angle 0^\circ}{283 \angle 45^\circ} = \frac{V_m \angle -45^\circ}{283}$$

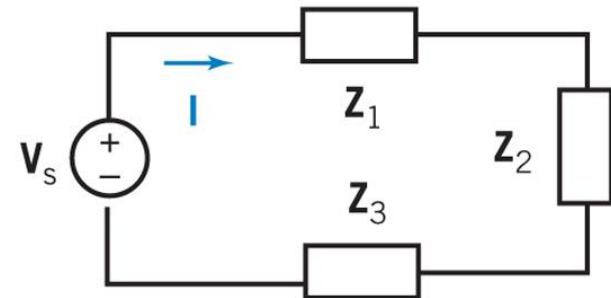
$$\Rightarrow i(t) = \frac{V_m}{283} \cos(100t - 45^\circ)$$

## Example 10.5-1 – KVL

- Determine the steady state current  $i(t)$  using phasor and impedances



(a)

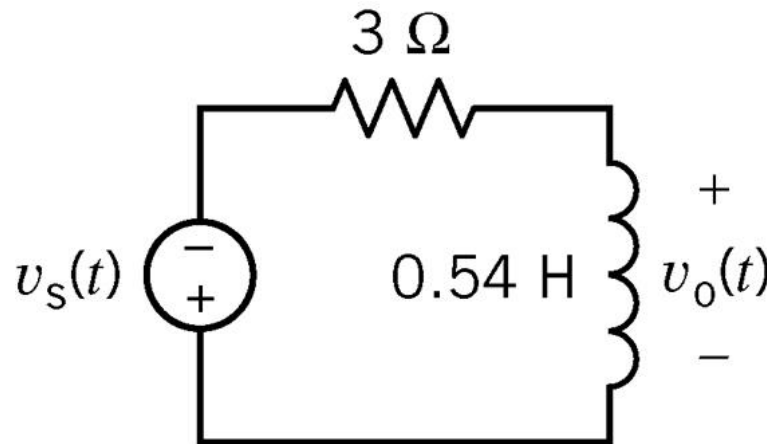


(b)

## Example 10.5-2 – Voltage Division

- Determine the steady-state output voltage,  $v_o(t)$  if the source voltage is:

$$v_s(t) = 7.28 \cos(4t + 77^\circ) \text{ V}$$



(a)

## Example 10.5-3 – KVL

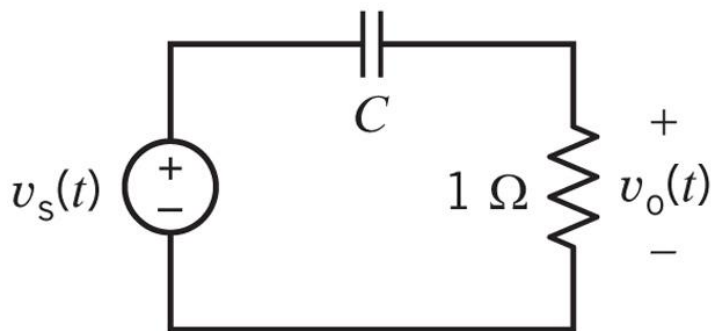
The input to the circuit is the voltage of the voltage source:

$$v_s(t) = 7.68\cos(2t + 47^\circ) \text{ V}$$

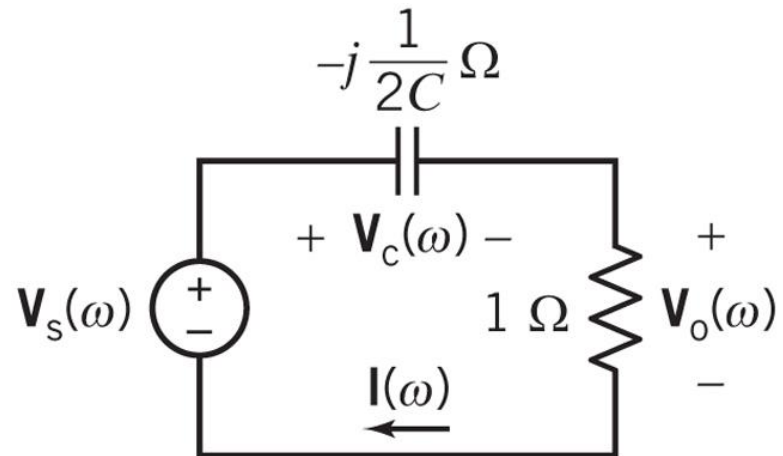
The output is the voltage across the resistor:

$$v_o(t) = 1.59 \cos(2t + 125^\circ)$$

Determine capacitance  $C$  of the capacitor.



(a)

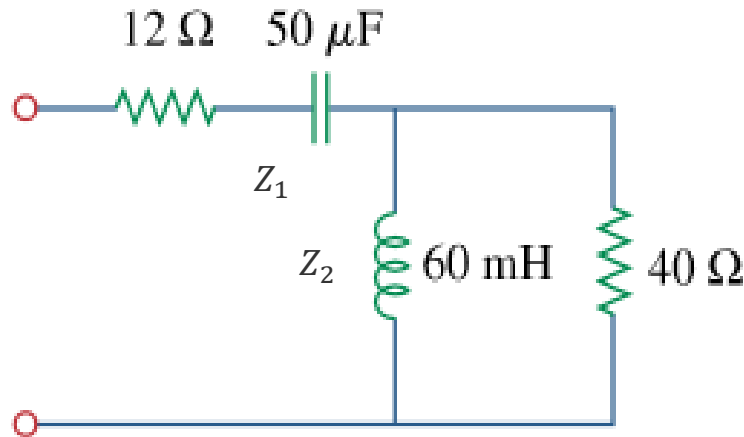


(b)



## Example 8 – $Z_{eq}$

- At  $\omega = 377 \text{ rad/s}$ , find the input impedance of the circuit.



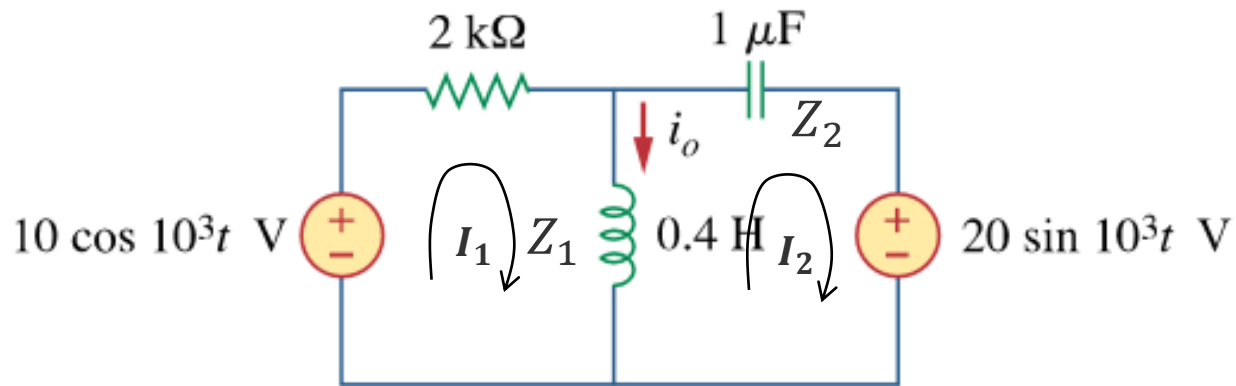
$$\text{R: } Z = R$$

$$\text{L: } Z = j\omega L$$

$$\text{C: } Z = -j \frac{1}{\omega C}$$

## Example 9 – Mesh Analysis

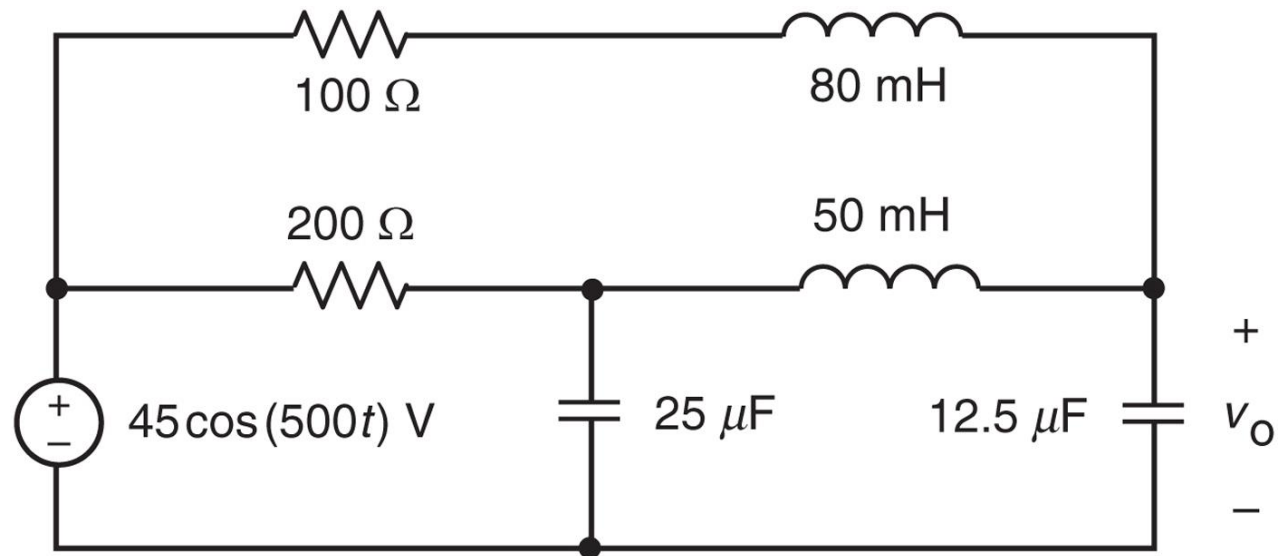
- Calculate  $i_o$



$$\sin(\omega t) = \cos(\omega t - 90^\circ) = 1 \angle -90^\circ = -1j$$

## Example 10.6-2

- Determine the mesh currents for the circuit shown below.

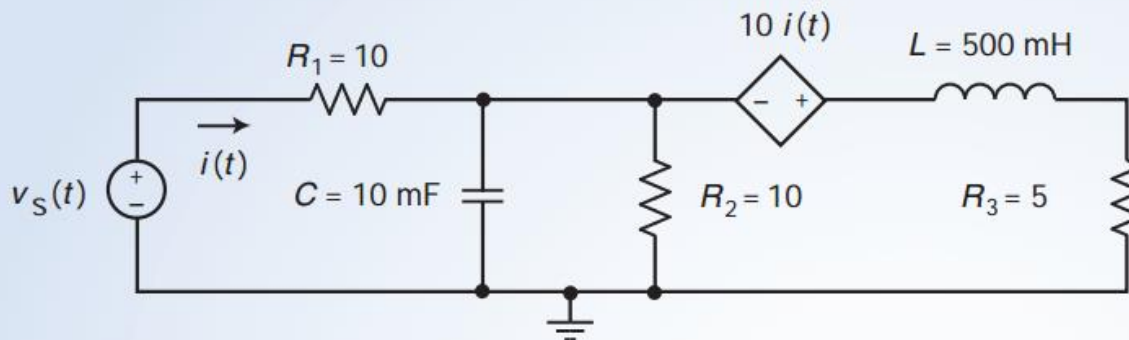


## Exercise 10.6-3 – Node Analysis

The input to the circuit shown in Figure 10.6-10 is the voltage source voltage

$$v_s(t) = 10 \cos(10t) \text{ V}$$

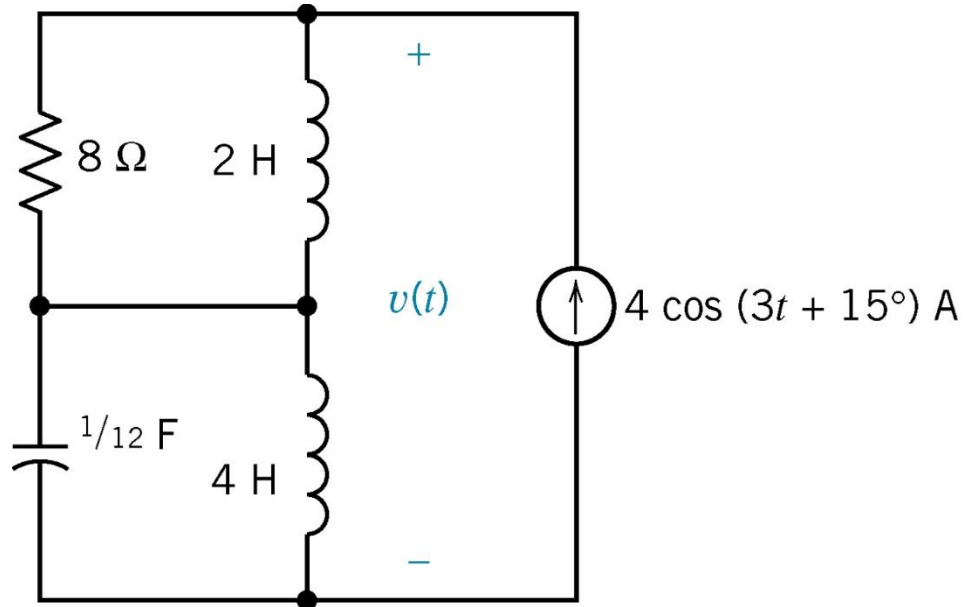
The output is the current  $i(t)$  in resistor  $R_1$ . Determine  $i(t)$ .



**FIGURE**  
Example

## Exercise 10.7-2

- Determine the phasor representation of each circuit element.



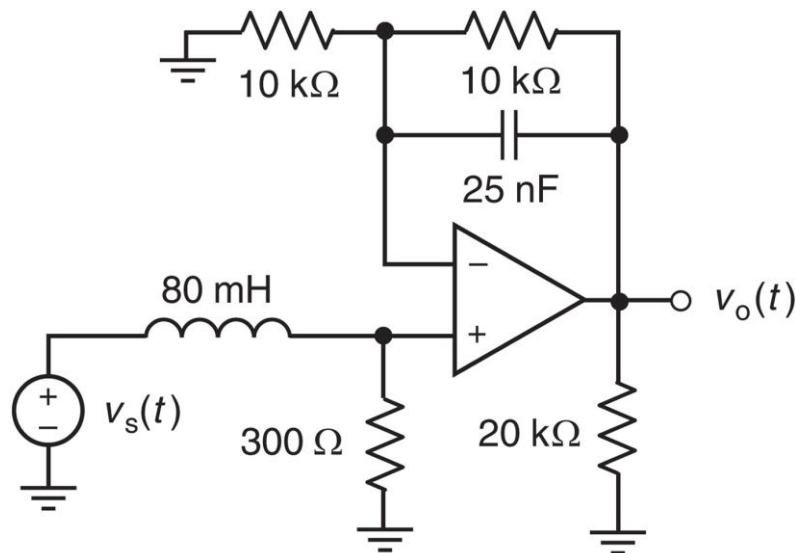
- **OpAmp**
- **Thévenin/Norton Equivalent**
  - No dependent sources: Just like in DC
  - Turn off independent sources
  - Determine the equivalent impedance
  - *Include impedance from capacitors and inductors*
  - Turn on independent sources
  - Determine phasor voltage (Thévenin) or current (Norton) at the terminals
- **Use source transformation to go from Norton to Thévenin**
- **With dependent sources**
  - Add a phasor source across the terminals
  - $V_m = 1$ , *angle* = 0
  - Determine the phasor current it provides to the circuit
  - Use Ohm's law to calculate the equivalent impedance

## Example 10.6-4 – OpAmp

The input to the ac circuit shown in Figure 10.6-13 is the voltage source voltage

$$v_s(t) = 125 \cos(500t + 15^\circ) \text{ mV}$$

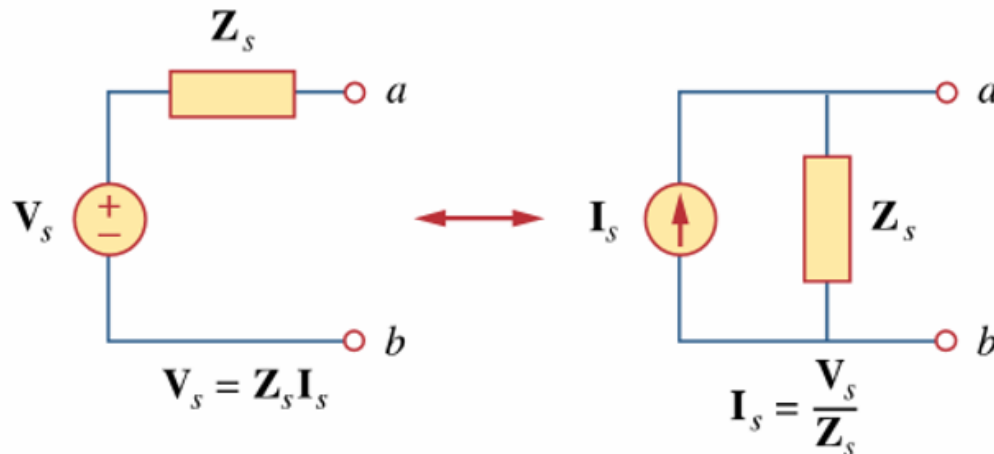
Determine the output voltage  $v_o(t)$ .



# Source Transformation

- Source transformation in the frequency domain is just like in the time domain, except that we use impedance instead of resistance

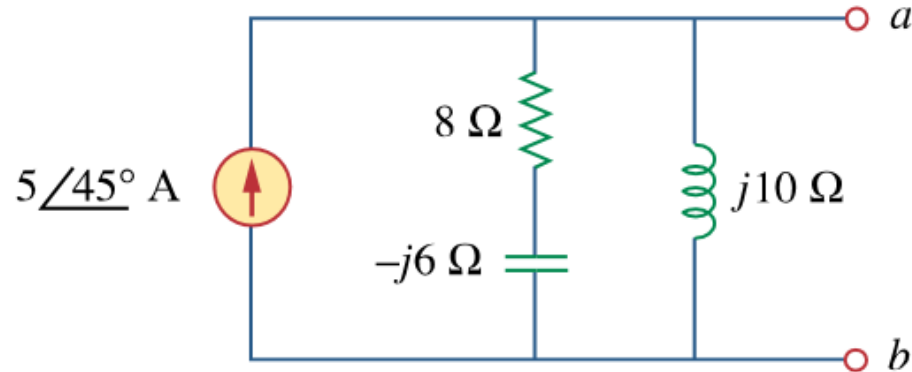
$$V_s = \frac{Z_s}{I_s}$$
$$I_s = \frac{V_s}{Z_s}$$





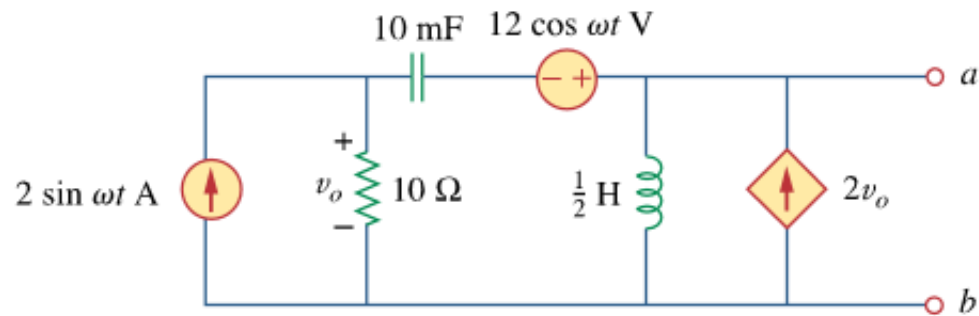
## Example 10

For the circuit depicted in Fig. 10.101, find the Thevenin equivalent circuit at terminals  $a$ - $b$ .



# Example 11

**10.66** At terminals  $a$ - $b$ , obtain Thevenin and Norton equivalent circuits for the network depicted in Fig. 10.109. Take  $\omega = 10$  rad/s.



# Example 12

