

Lab 03 - Time-Varying Resonant Filters

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PROBLEM 1(a)

A resonant low-pass filter has the following transfer function:

$$H(s) = \frac{1}{(s/\omega_c)^2 + \frac{1}{Q}(s/\omega_c) + 1}.$$

The magnitude frequency response and phase response for various values of frequency and Q were plotted using the following MATLAB script (Figure 1).

```
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%
% HW 3 - Problem 1(a)

clear all;
close all;

% Constants:
wc = 2*pi*1000;           % Radians/second
Q = 2.^(-4:2);           % Unit-less
N = length(Q);           % Loop end
B = [0 0 1];             % Numerator Coeffs.

for i = 1 : N

    % Calculate denom. coeffs.
    A = [1 1/(wc*Q(i)) 1/(wc^2)];
    % Calculate freq. response
    [H,W] = freqs(B,A);
    % Add magnitude to plot:
    subplot(2,1,1);
    loglog(W, abs(H));
    hold all
    % Add phase to plot:
    subplot(2,1,2);
    semilogx(W, angle(H));
    hold all
end
```

Figure 1, MATLAB script for problem 1(a).

The following plots for phase and magnitude response were seen for values of Q ranging from $Q = 2^{-2}$ up to $Q = 2^4$ (Figure 2).

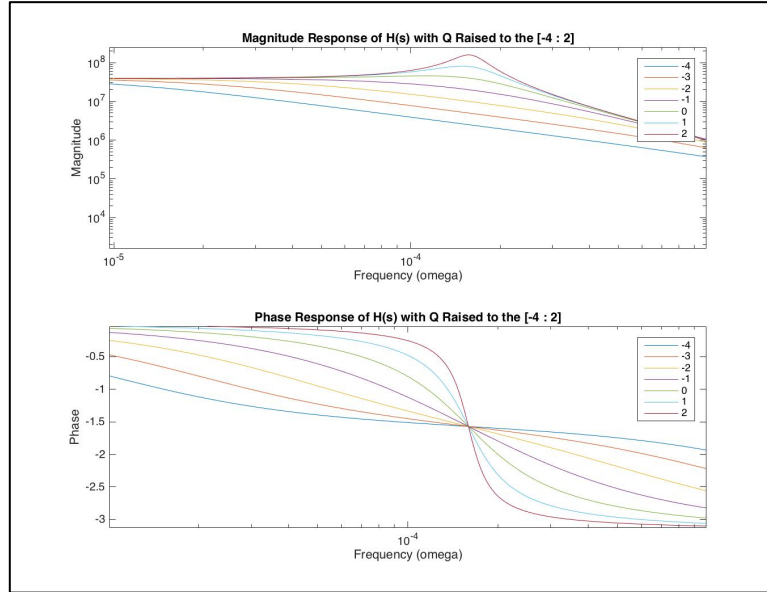


Figure 2, magnitude and phase for varying values of Q .

Next, frequency was varied from $2\pi 1000(2^{-2})$ up to $2\pi 1000(2^2)$ (Figure 3).

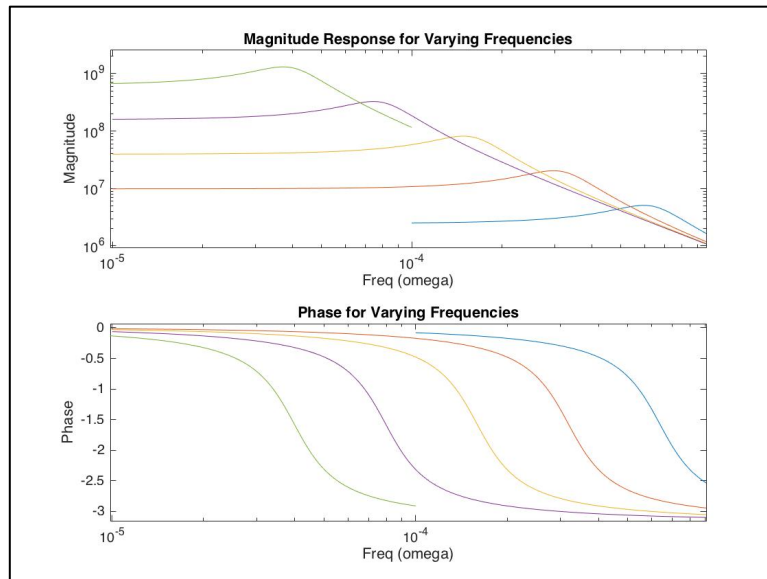


Figure 3, magnitude and phase for varying frequencies.

Finally, the poles and zeros for each case were plotted in Figures 4 and 5.

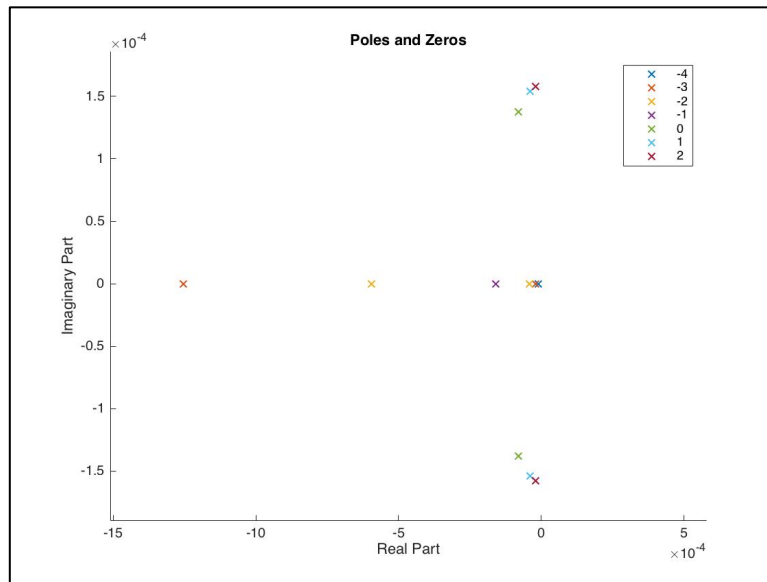


Figure 4, poles and zeros for varying Q .

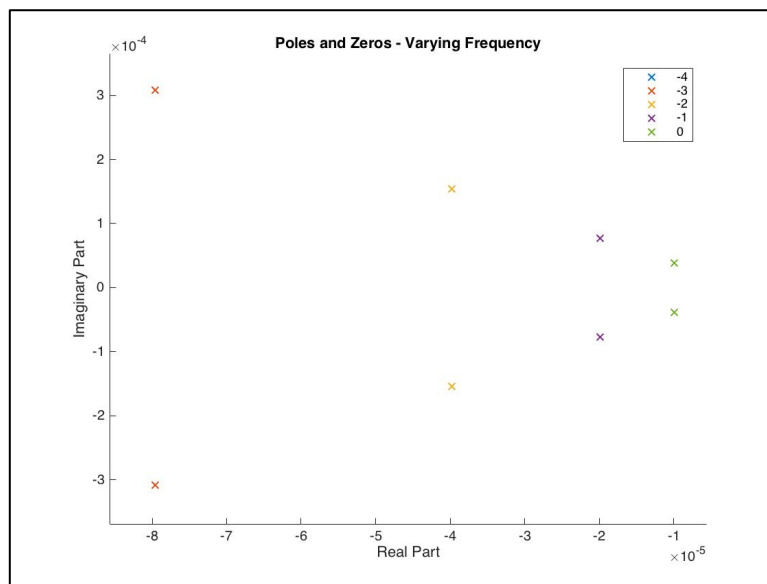


Figure 5, poles and zeros for varying frequency.

PROBLEM 1(b)

The bilinear transform has been used to map our analog filter into a digital representation (Figure 6).

$$H(s) = \frac{1}{(\frac{s}{w})^2 + (\frac{1}{Q})(\frac{s}{w}) + 1}$$

We let $s = (\frac{2}{T})(\frac{1-z^{-1}}{1+z^{-1}})$

$H(s)$ becomes $H(z)$...

$$H(z) = \frac{1}{(\frac{(\frac{2}{T})(\frac{1-z^{-1}}{1+z^{-1}})}{w})^2 + (\frac{1}{Q})(\frac{(\frac{2}{T})(\frac{1-z^{-1}}{1+z^{-1}})}{w}) + 1}$$

Figure 6, bilinear transform mapping from s to z.

Then, the bi-quad coefficients were implemented in C++ (Figure 7).

```
//////////BEGIN//////////
//TODO: design analog filter based on
//input gain, center frequency and Q
//////////START//////////

b0 = 1.0;
b1 = 0.0;
b2 = 0.0;
a0 = 1.0;
a1 = 1.0 / (center * qval * 2 * pi);
a2 = 1.0 / (center * center * 2 * pi);

//////////END//////////
// TODO: apply bilinear transform
//////////START//////////

double T = 1/fs;

az0 = ( a0*T*T + 2*a1*T + 4*a2 );
az1 = ( 2*a0*T*T - 8*a2 ) / az0;
az2 = ( a0*T*T - 2*a1*T + 4*a2 ) / az0;

bz0 = ( b0*t*T + 2*b1*T + 4*b2 ) / az0;
bz1 = ( 2*b0*T*T - 8*b2 ) / az0;
bz2 = ( b0*T*T - 2*b1*T + 4*b2 ) / az0;

az0 = 1;

//////////END//////////
```

Figure 7, bi-quad coefficients in C++.

PROBLEM 2(a)

Filter parameter controls have been fed into leaky integrators to help smooth changes between different states (Figure 8).

```
//////////////////////////////////////////
// Leaky integrator (same as Lab 01) //
//////////////////////////////////////////

void setTau(float tau, float fs) {
    ///////////////////////////////////////////////////START//////////////////////////////////////
    a1 = exp( -1.0 / ( tau * fs ) );
    b0 = 1 - a1;
    ///////////////////////////////////////////////////END//////////////////////////////////////
}

void reset() {
    // reset filter state
    z1=0;
}

void process (float input, float& output) {
    ///////////////////////////////////////////////////START//////////////////////////////////////
    z1 += b0 * (input - z1);
    output = z1;
    ///////////////////////////////////////////////////END//////////////////////////////////////
}
```

Figure 8, updated "SlewedParameter" class.

PROBLEM 2(b)

An LFO was implemented using the `sin()` of a phase counter (Figure 9).

```
float WahWah::LFO(float f0)
{
    // TODO: Implement
    ///////////////////////////////////////////////////START//////////////////////////////////////

    // Initialize:
    float phase = 0;
    // Increment by delta:
    float change = 2*pi*f0/fs;
    // Mod 2pi for phase:
    phase = fmodf(phase+change,2*pi);
    // Lookup next sample:
    float sample = sin(phase);
    return sample;

    ///////////////////////////////////////////////////END//////////////////////////////////////
}
```

Figure 9, LFO implementation in C++.

PROBLEM 2(c)

It seems like the frequency computer is already coded, and I'm not sure what to change. Still, some modest changes in the following sections were made (Figure 10).

```
void WahWah::processReplacing [... etc etc... ]

    // slew control signals
    float freqSlew, gainSlew, qSlew;
    fcSlewer.process(FcValue, freqSlew);
    gSlewer.process(GainValue, gainSlew);
    qSlewer.process(QValue, qSlew);

    // modulation signal
    float modulationSignal = frequencyComputer(freqSlew, RateValue, DepthValue);
}

// Not sure what to change in here:
float WahWah::frequencyComputer(float fc, float rate, float depth)
{
    float floSignal = pow(depth, .5*LF0(rate));

    // Slew the modulation signal as well
    modSlewer.process(floSignal, floSignal);

    return fc*floSignal;
}
```

PROBLEM 2(d)

I did not have time to finish this, but the peak detector portion of the code is presented below (Figure 11).

```
PeakDetector() {
[... etc ...]

    // Process one block (one sample):
    void process (float input, float threshold, float& output) {

        // Test (?) above or below threshold:
        if ( fabs( input ) > levelEstimate ) {

            // "Attack-state" update equation:
            levelEstimate += b0_a * ( fabs( input ) - levelEstimate );
        }

        else {

            // Release to threshold:
            levelEstimate += b0_r * ( threshold - levelEstimate );
        }

        // Update output:
        output = levelEstimate;
    }
};
```

Figure 11, peak detector scheme.