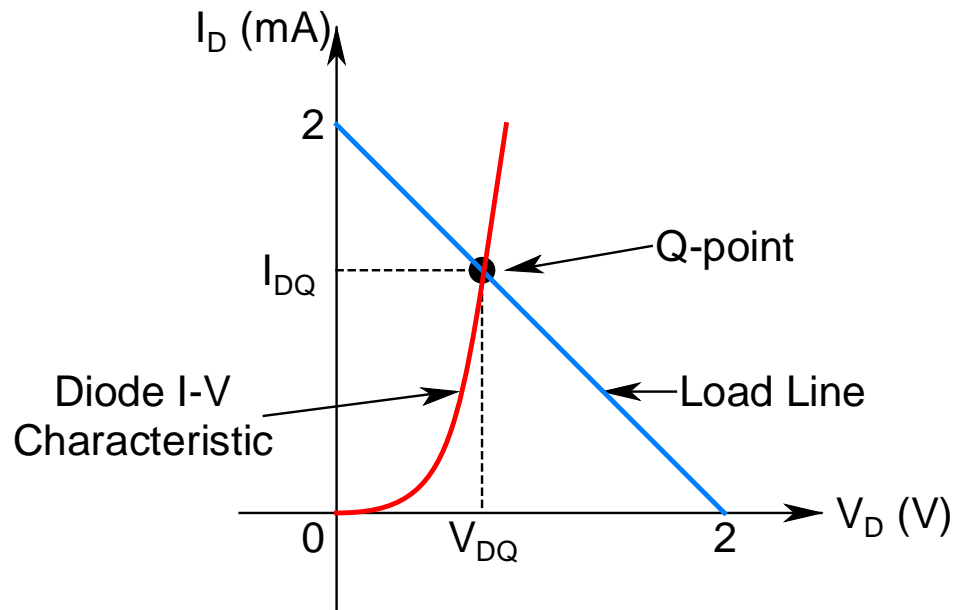


Finding the DC Operating Point

- Known as the ***Q-Point*** (*Quiescent Point*)
- *Defined by (I_D , V_D)*
- *Two solution techniques:*
 - *Graphical Method* (*Using Load Line*)
 - *Iterative Method* (*Self-Consistent*)
- *Graphical Method:*
 - *Use diode I-V relation:* $I_D = I_0 \exp(V_D/V_T)$
 - *Use Load Line equation:* $I_D = (2 - V_D)/(1 \text{ k}\Omega)$
 - *Intersection point is the operating point*



V_{DQ} , I_{DQ} : Quiescent values of V_D , I_D
 $= 0.7$ V, 1.3 mA
 (obtained from iterative method using $I_0 = 3$ fA)

DC Quiescent Power Dissipation
 $= V_{DQ} \times I_{DQ} = 0.9$ mW

Graphical Method

The two *end points* of the *load line*:

1. $V_D = 0$: $I_D = 2/(1 \text{ k}) = 2$ mA
2. $I_D = 0$: $V_D = 2$ V

- *Iterative Method:*

- Also known as *self-consistent analysis*
- *I-V relation* and *load line equation* form a set of *transcendental equations*
 - *Analytical solution not possible*
 - Have to resort to *numerical (iterative)* analysis
- *Procedure:*
 - *Choose $V_D = 0.7\text{ V}$ and find I_D from load line equation*
 - *Use this I_D to find V_D from I-V relation (convert to ln form first)*
 - *Repeat till convergence is achieved* (pretty quick!)

Series Resistance Effect

- The two *quasi-neutral regions* (p and n) have their own *bulk resistances*
- *Denoted as r_p and r_n*
- *For small I_D* , their *effects are negligible*
- However, *for large I_D* , the *potential dropped* across them *reduces* the *actual voltage* appearing across the *junction* of the diode
- Known as the *Series Resistance Effect*

- The *actual voltage* appearing across the *junction*:
 - $V_{D,\text{eff}} = V_D - I_D(r_p + r_n)$
 - V_D : *Diode terminal voltage*
- The *new diode I-V relation*:
 - $I_D = I_0 \exp(V_{D,\text{eff}}/V_T)$
- *Simultaneous solution* of the above *two equations* would yield the *Q-Point*
- *In presence of r_p and r_n , V_D may exceed V_0 , but $V_{D,\text{eff}}$ would always be less than V_0*

Diode Capacitances

- *Two components:*

- *Depletion Capacitance* C_{dep}

- *Diffusion Capacitance* C_{diff}

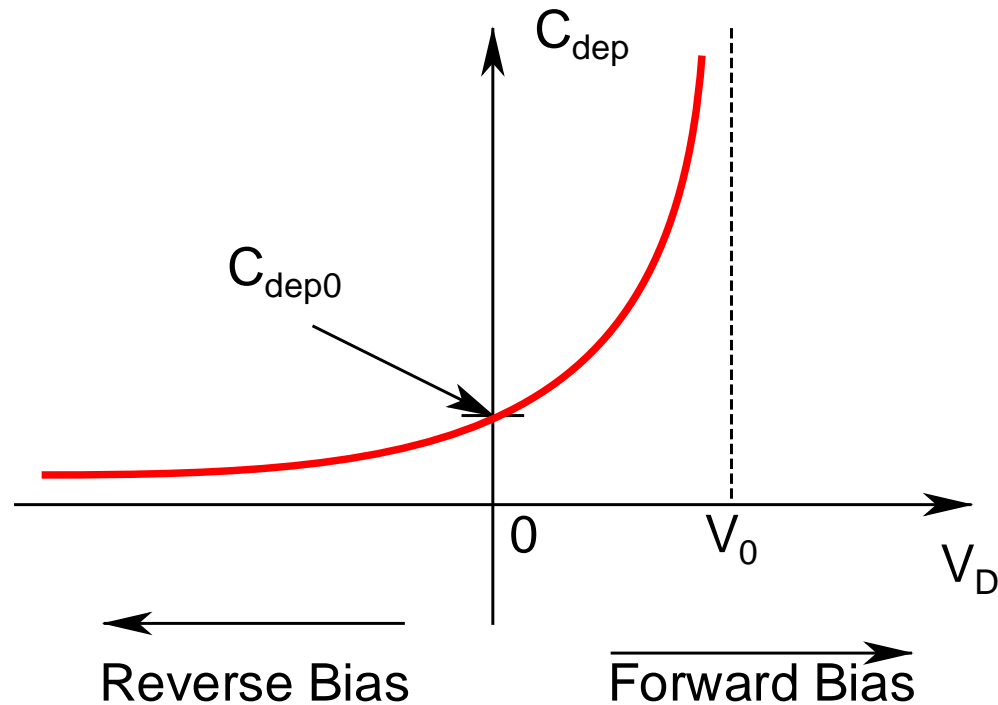
- *Depletion Capacitance:*

- Due to *depletion charge dipole* across the *junction*

- *Expressed by:* $C_{\text{dep}} = C_{\text{dep0}} / (1 - V_D/V_0)^{1/2}$

- $C_{\text{dep0}} = C_{\text{dep}}$ for $V_D = 0$

- *Present under both forward and reverse bias*



- * C_{dep} diverges as $V_D \rightarrow V_0$
- * C_{dep} in reverse bias < C_{dep} in forward bias
- * Rule of Thumb: C_{dep} in forward bias $\approx 2C_{\text{dep0}}$

- ***Diffusion Capacitance:***

- Due to *charge injection* in *both sides of the junction* (in the *quasi-neutral regions*)
- *Expressed by:* $C_{\text{diff}} = I_D \tau / V_T$
 - τ : *Injected carrier lifetime*
- *Present only in forward bias*, *negligible in reverse bias*
- *Much larger than C_{dep}*

- ***Total Diode Capacitance C_D :***

- $C_D = C_{\text{dep}} + C_{\text{diff}}$

Small-Signal Model

- *Needed for ac analysis*
- *Electrical equivalent* at the *DC bias point*
- Represented as a *network*, having various *circuit components* (*resistors*, *capacitors*, *current sources*, etc.)
- Also known as *incremental model*
- *Evaluated at (I_{DQ} , V_{DQ})*
- Appropriate for *small variations* of the *ac signal* around the *Q-point*

Small-Signal Model Parameters

- *Diode Resistance* (r_D)/*Diode Conductance* (g_D):

$$r_D = (g_D)^{-1} = \left(\frac{\partial I_D}{\partial V_D} \right)^{-1} \bigg|_{I_D = I_{DQ}} = \frac{V_T}{I_{DQ}}$$

- For $I_{DQ} = 1 \text{ mA}$, $r_D = 26 \Omega$
- *Under forward bias*, diode offers *very small resistance*

- **Diode Capacitance** (C_D):

- $C_D = C_{\text{dep}} + C_{\text{diff}}$

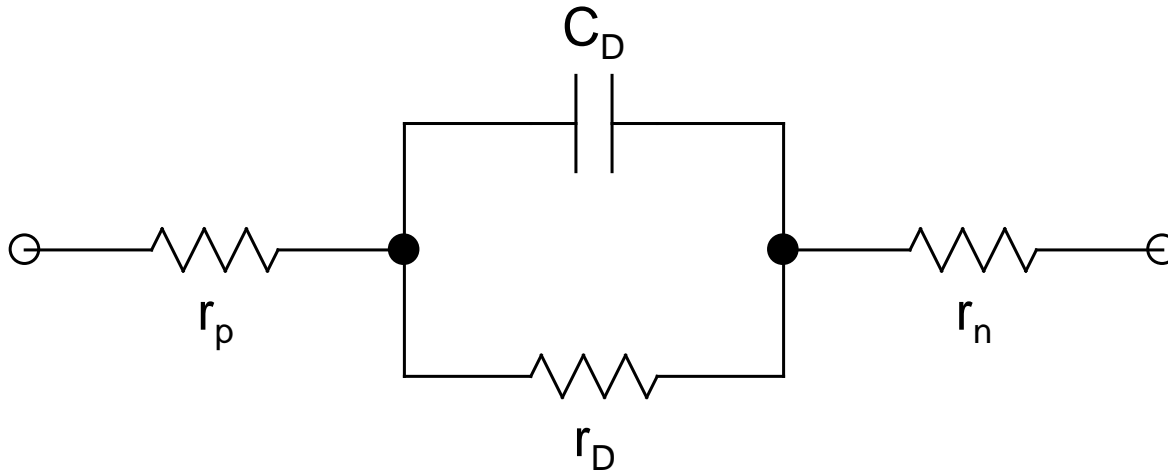
- $= C_{\text{dep0}} / (1 - V_D/V_0)^m + \tau/r_D$

- m: *Grading coefficient*

- (1/2 for *abrupt step junction*, 1/3 for *linearly graded junction*)

- *Both r_D and C_D appear in parallel across the junction*
- *The two quasi-neutral resistance r_p and r_n appear in series with this combination*

Small-Signal Equivalent



- *In absence of r_p and r_n , it's just a parallel RC circuit, and shorts out at high frequency*
 - *Diode time constant $\tau_D = r_D C_D = \tau$*