

Final Exam

EE 250 (Control Systems Analysis) Spring 2010 *

DEPARTMENT OF ELECTRICAL ENGINEERING, IIT KANPUR.

Instructions

1. Before you turn in your answer books, please verify that your answer booklet contains one semilog graph paper.
2. You may use a pencil for trial and error, but your final drawing should use a pen.
3. Show all your assumptions.
4. You will need the following items: pen, pencil, ruler, eraser.

$$t_s = \frac{4}{\zeta\omega_n} \left(\begin{array}{c} 2\% \text{ tube,} \\ 0 < \zeta < 0.8 \end{array} \right), \quad M_p = e^{-\pi\zeta/\sqrt{1-\zeta^2}}.$$

$$PM^\circ \approx 100 \times \zeta,$$

PM	25°	40°	60°	70°
M_p	45%	25%	10%	5%

Useful information

Rule 1: The RL has n branches.

Rule 2: The RL starts at $K = 0$ at the poles and ends at $K = \infty$ on the m zeros, and $n - m$ of its branches end at ∞ .

Rule 3: On the real axis, the RL exists only on those segments to the right of which the sum of the number of poles and the number of zeros is odd.

Rule 4: The RL is symmetric about the real axis.

Rule 5: As $K \rightarrow \infty$, $n - m$ branches of the RL go asymptotically along rays that emanate from the centroid

$$s = -\sigma = -\frac{\sum_{i=1}^n p_i - \sum_{i=1}^m z_i}{n - m}$$

Rule 6: The angles that these rays form with the real axis are

$$\frac{\pm 180^\circ (2q + 1)}{n - m}, q = 0, 1, 2, \dots$$

Rule 7: The roots of multiplicity greater than one occur on the RL where s satisfies $dG/ds = 0$, or, equivalently, $dK/ds = 0$.

Rule 8: In order to determine the angle of departure from a pole and the angle of arrival at a zero, use the phase criterion (PC) along with the idea of a point $s = \hat{s}$ that is infinitesimally close to the pole (zero) the angle of departure (arrival) from (to) which needs to be calculated.

Rule 9: In order to determine the $j\omega$ -axis intercepts of the RL, use the Routh-Hurwitz method.

Rule 10: If necessary, calculate a few additional suitably located points using the PC.

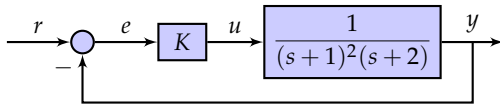
$\Phi_{\max} (\Phi_{\min})$ needed	$90^\circ (-90^\circ)$	$60^\circ (-60^\circ)$	$45^\circ (-45^\circ)$	$30^\circ (-30^\circ)$
DD between corner	∞	1.1439	0.76555	0.47712
freq-s of BP in decades		≈ 1.144	≈ 0.766	≈ 0.477

$$t_r = \frac{\pi - \arccos \zeta}{\omega_n \sqrt{1 - \zeta^2}}, \quad t_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}},$$

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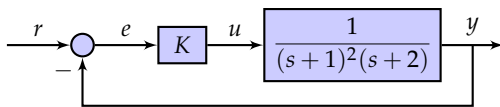
Problems

1. Suppose you wish to use the ultimate gain method to tune a PID controller for a plant that has the following exact transfer function in the open-loop: $\frac{1}{(s+1)^2(s+2)}$. Suppose you discover that this system displays sustained oscillations at $K = 18$ in the following feedback configuration when $r = 1(t)$:



- 1.1. [2 points] Determine the period of these oscillations.
- 1.2. [1 points] What is the critical gain of this system?
- 1.3. [2 points] About which value do these oscillations happen — about 1 or 0 or about some other value? Show your calculations.
- 1.4. [1 points] Using the second Ziegler Nichols method, determine the parameters of a PID controller that can be used in the place of K .
- 1.5. [1 points] Write the transfer function of this PID controller.

2. For the following control system



- 2.1. [2 points] Calculate the breakaway and breakin points if any.
- 2.2. [2 points] If breakaway points exist, then calculate the angle of departure from them.
- 2.3. [1 points] Calculate the centroid, if it exists.
- 2.4. [1 points] Calculate the angles of the asymptotes, if they exist.
- 2.5. [2 points] Sketch the root locus.
- 2.6. [1 points] What is the value of K for which this root locus intersects the $j\omega$ -axis?
- 2.7. [2 points] Is this system conditionally stable? Explain.

3. For the state space equation

$$\dot{x} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 2 & 0 \\ 0 & 0 & 3 \end{bmatrix} x + \begin{bmatrix} 2 \\ 1 \\ 3 \end{bmatrix} u, \quad y = [1 \quad 1 \quad 2] x$$

we wish to design a tracker of the form $u = -Kx + Ny_{\text{ref}}$ such that the unit step response of the closed-loop system has an approximately second-order behavior with an overshoot of approximately 25%, and a settling time of approximately 2 s.

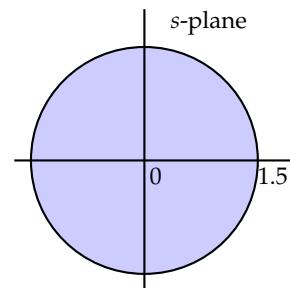
[5 points] For this purpose, determine the location of the desired closed-loop eigenvalues neglecting the effect of closed-loop zeros.

4. Given the open-loop transfer function, $G(s) = \frac{K}{s^2 + \omega_0^2}$, we wish to use “Bode plot to polar plot” method to sketch the Nyquist plot (NP) of $G(s)$ for $K \in [0, \infty)$.

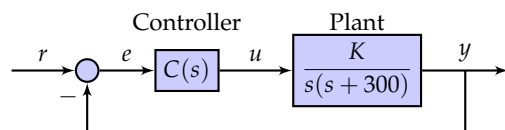
- 4.1. [2 points] Sketch the Bode plot (BP).
- 4.2. [1 points] Sketch the polar plot (PP) section of the NP. Show this section by thick solid lines.
- 4.3. [2 points] Work out a few points on the s -plane contour and the $G(s)$ -plane contour that will help complete the NP.
- 4.4. [2 points] Sketch the NP. Label the sections of the s -plane contour $C1, C2, \dots$ and the corresponding sections of the NP $C1', C2', \dots$. Label a few points on the s -plane contour $1, 2, \dots$, and the corresponding points on the NP $1', 2', \dots$.

5. [2 points] For one clockwise traversal of the following s -plane contour, determine the number of times the $G(s)$ -plane contour encircles the origin of the $G(s)$ -plane in the clockwise direction, given that

$$G(s) = \frac{(s+1)(s+2)}{(s^2+s+1)}.$$



6. Consider an aircraft attitude control system shown in the following figure.



We wish to determine $C(s)$ and K such that the following performance specifications are satisfied:

- Steady-state error due to unit-ramp input ≤ 0.000443 ,
- Maximum overshoot 10% – 45%,
- Settling time $t_s \leq 0.005$ s,
- The controller must be first order.

- 6.1. [3 × 2 points] Explain how you will satisfy the first 3 specifications.
- 6.2. [3 points] On the semilog graph paper provided design the loopgain using loop-shaping to satisfy these specifications.
- 6.3. [2 points] Explain why your design will satisfy the specification on overshoot and controller order.

[illegible]